

FEATURES

- 1.2MHz Fixed Frequency Current-Mode PWM Operation.
- Efficiency Up to 84% at $V_{IN}=4.2V$, 3LEDs, $I_{LED}=20mA$
- Drives Up to 6LEDs in series
- Low Supply Current: 70 μA
- Matches LED Current
- Require Tiny Inductors and Capacitors
- Tiny SOT23-5 Package

APPLICATIONS

- Cellular Phones
- PDAs
- Digital Still Cameras
- Handheld Devices
- White LED Display Backlighting

DESCRIPTION

AIC1647 is a fixed frequency step-up DC/DC converter designed to drive white LEDs with a constant current to provide backlight in hand-held devices. Series connection of the LEDs provides identical LED currents resulting in uniform brightness. This configuration eliminates the need for ballast resistors. Low 95mV feedback voltage minimizes power loss in the current setting resistor for better efficiency.

AIC1647 is a step-up PWM converter, which includes an internal N-channel MOSFET switch for high efficiency. The high switching frequency, 1.2MHz, allows the use of tiny external components, saves the layout space and cost.

AIC1647 is available in a space-saving, 5-lead SOT-23-5 package.

TYPICAL APPLICATION CIRCUIT

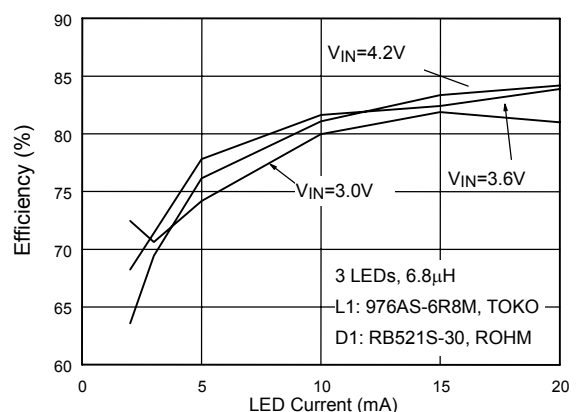
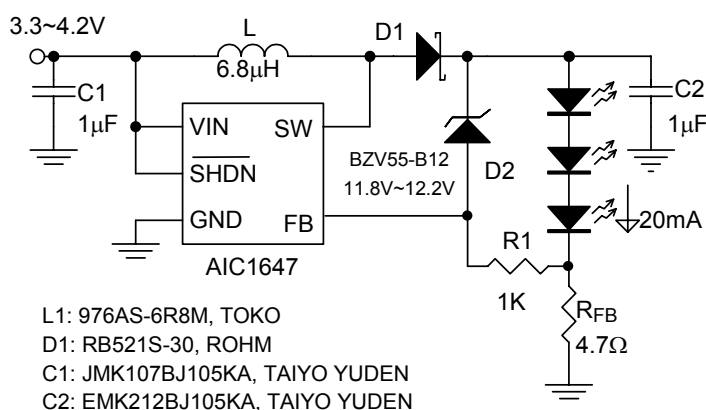
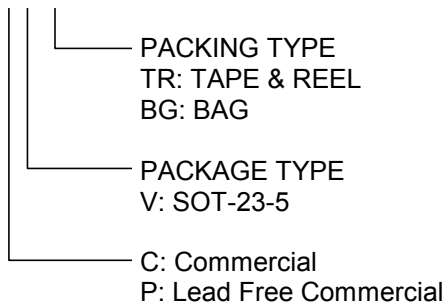
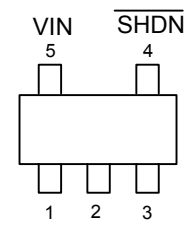


Fig. 1 Li-Ion Powered Driver with Over Voltage Protection for Three White LEDs

ORDERING INFORMATION

AIC1647XXXX



ORDER NUMBER	PIN CONFIGURATION
AIC1647CV&PV (SOT-23-5)	FRONT VIEW  VIN 5 SHDN 4 1 2 3 SW GND FB

Example: AIC1647CVTR
→ in SOT-23-5 Package & Tape & Reel Packing Type

MARKING

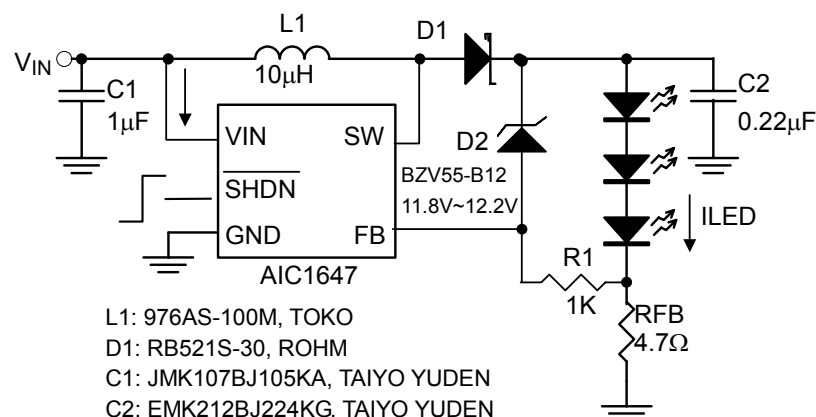
Part No.	CV	PV
AIC1647	1647	1647P

ABSOLUTE MAXIMUM RATINGS

Input Voltage (VIN)	6V
SW Voltage	33V
FB Voltage	6V
SHDN Voltage	6V
Operating Temperature Range	-40°C to 85°C
Maximum Junction Temperature	125°C
Storage Temperature Range	-65°C to 150°C
Lead Temperature (Soldering, 10 sec)	260°C

Absolute Maximum Ratings are those values beyond which the life of a device may be impaired.

TEST CIRCUIT



■ ELECTRICAL CHARACTERISTICS

($V_{SHDN}=3V$, $V_{IN}=3V$, $T_A=25^{\circ}C$, unless otherwise specified.) (Note 1)

PARAMETER	SYMBOL	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Minimum Operating Voltage	V _{IN}		2.5			V
Maximum Operating Voltage	V _{IN}				5.5	V
Supply Current	I _{IN}	Switching		1	5	mA
		Non switching		70	100	μA
		V _{SHDN} = 0V		0.1	1.0	
ERROR AMPLIFIER						
Feedback Voltage	V _{FB}		85	95	105	mV
FB Input Bias Current	I _{FB}	V _{FB} =95mV		1		nA
OSCILLATOR						
Switching Frequency	f _{OSC}		0.8	1.2	1.6	MHz
Maximum Duty Cycle	DC		85	90		%
POWER SWITCH						
SW ON Resistance	R _{DS(ON)}			1.4	5	Ω
Switch Leakage Current	I _{SW(OFF)}	V _{SW} =33V		0.1	1	μA
CONTROL INPUT						
SHDN Voltage High	V _{IH}	ON	1.5			V
SHDN Voltage Low	V _{IL}	OFF			0.3	V

Note 1: Specifications are production tested at $T_A=25^{\circ}C$. Specifications over the $-40^{\circ}C$ to $85^{\circ}C$ operating temperature range are assured by design, characterization and correlation with Statistical Quality Controls (SQC).

TYPICAL PERFORMANCE CHARACTERISTICS

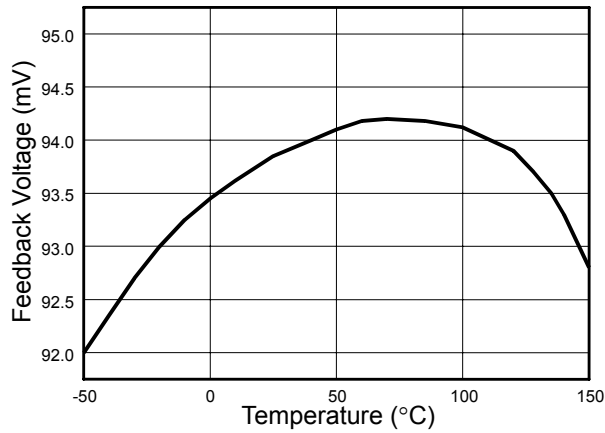


Fig. 2 Feedback Voltage vs. Temperature

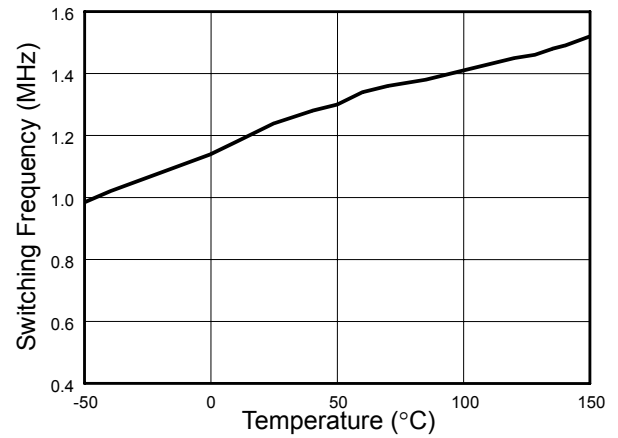


Fig. 3 Switching Frequency vs. Temperature

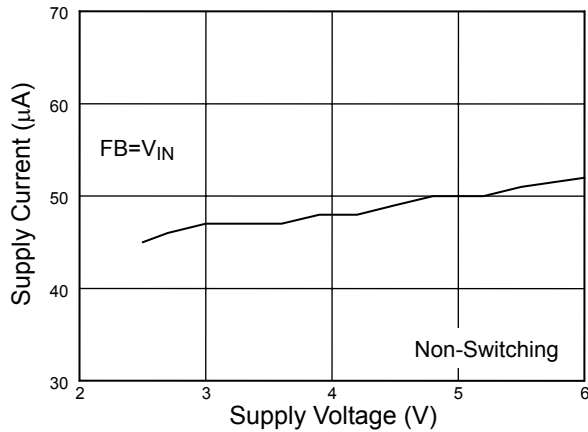


Fig. 4 Supply Current vs. Supply Voltage

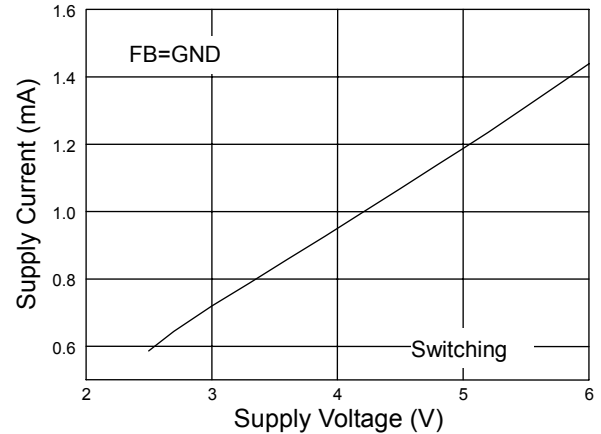


Fig. 5 Supply Current vs. Supply Voltage

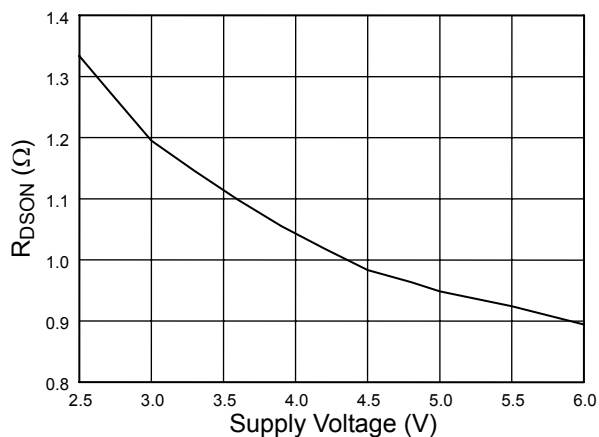
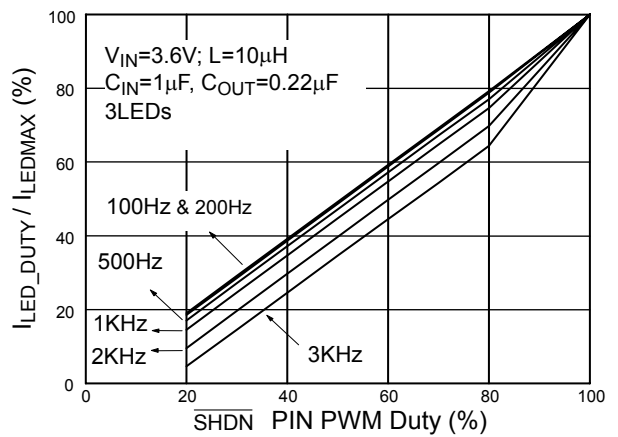

Fig. 6 R_{DS-ON} vs. Supply Voltage


Fig. 7 Dimming Control by Shutdown PIN

■ TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

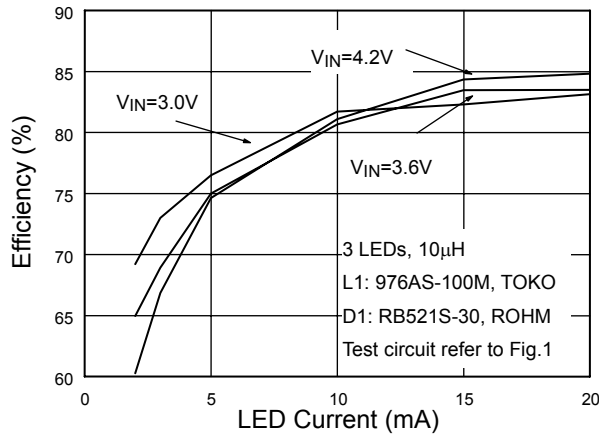


Fig. 8 3 LEDs Efficiency vs. LED Current

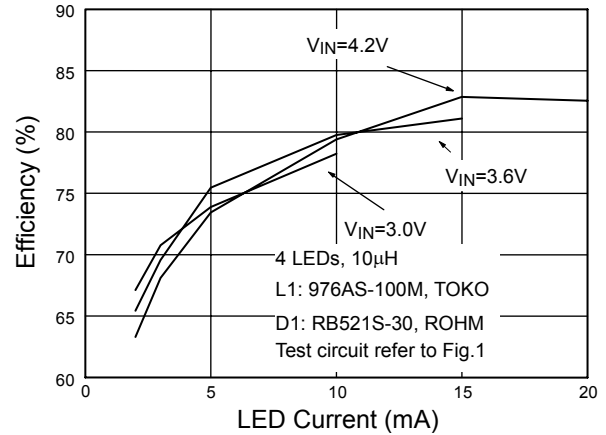


Fig. 9 4 LEDs Efficiency vs. LED Current

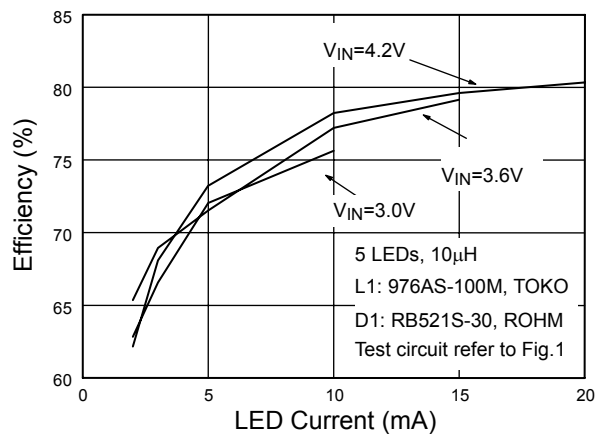


Fig. 10 5 LEDs Efficiency vs. LED Current

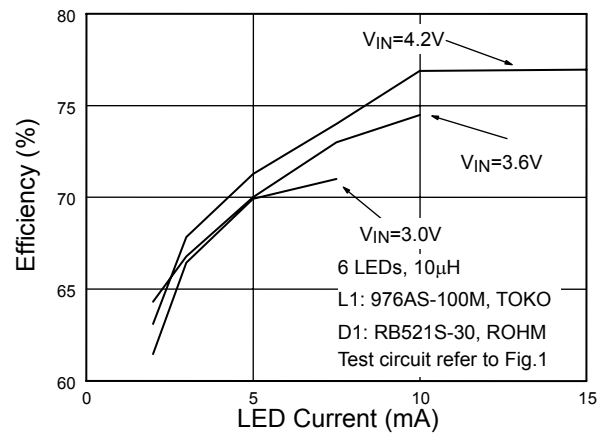


Fig. 11 6 LEDs Efficiency vs. LED Current

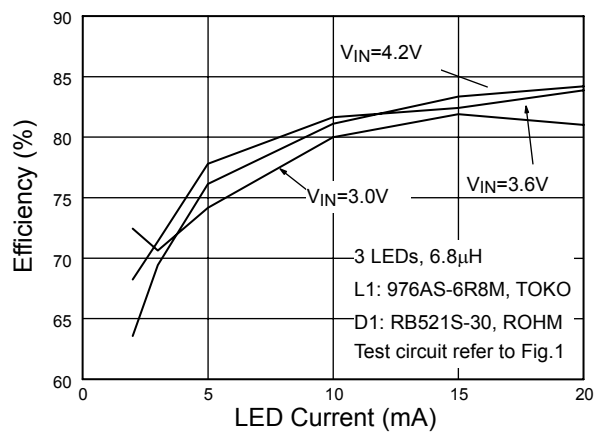


Fig. 12 3 LEDs Efficiency vs. LED Current

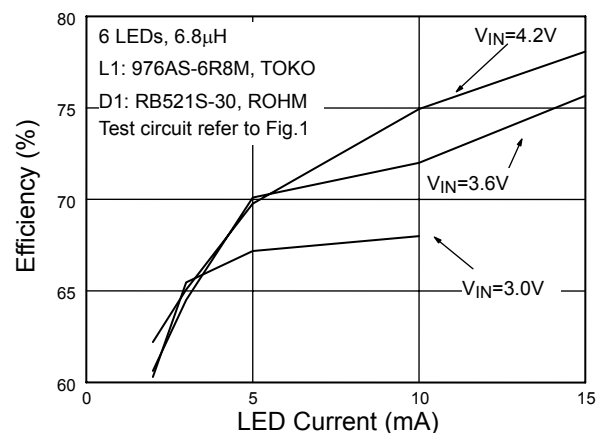
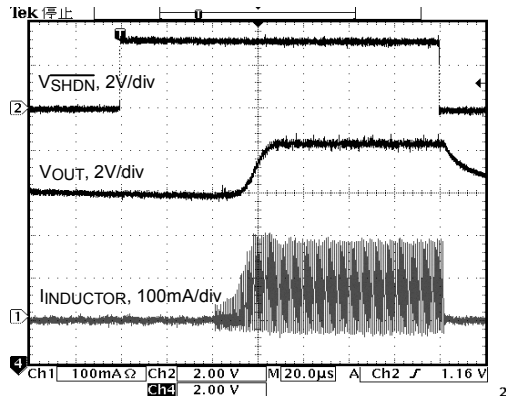


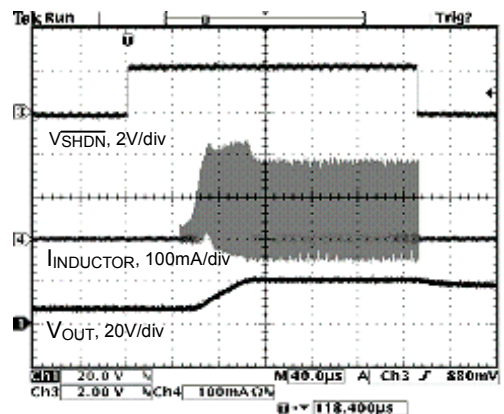
Fig. 13 6 LEDs Efficiency vs. LED Current

■ TYPICAL PERFORMANCE CHARACTERISTICS (Continued)



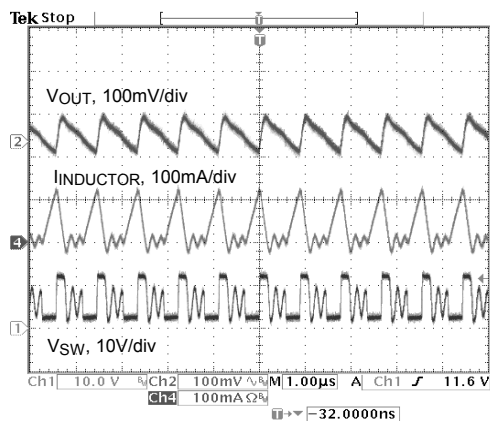
$V_{IN}=3.6V$; 3 LEDs; $L1=10\mu F$; $C_{OUT}=0.22\mu F$; $I_{LED}=20mA$

Fig. 14 Start-Up from Shutdown



$V_{IN}=3.6V$; 6 LEDs; $L1=10\mu F$; $C_{OUT}=0.22\mu F$; $I_{LED}=10mA$

Fig. 15 Start-Up from Shutdown



$V_{IN}=3.6V$; 3 LEDs; $L1=10\mu F$; $C_{OUT}=0.22\mu F$; $I_{LED}=10mA$

Fig. 16 Operation Wave Form

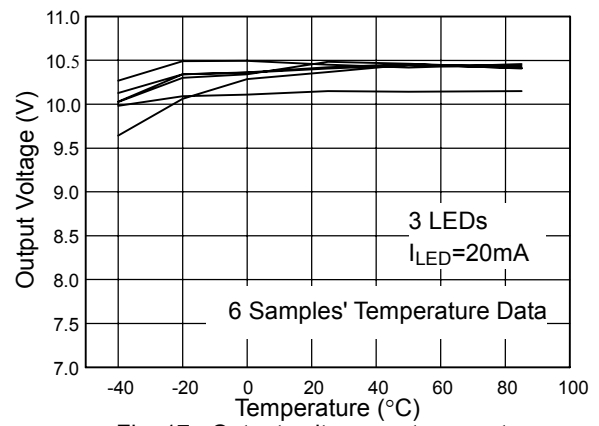


Fig. 17 Output voltage vs. temperature

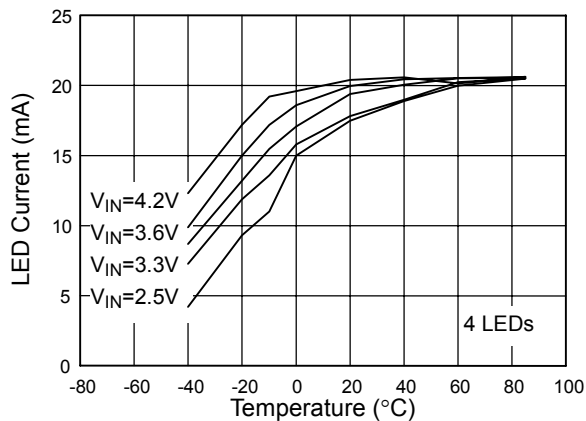
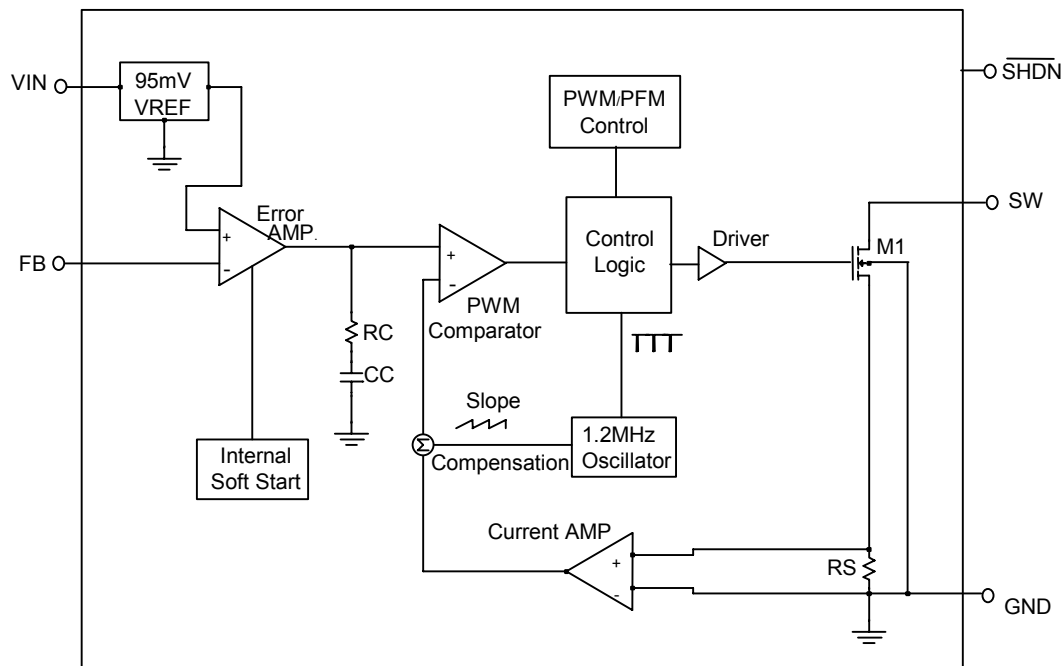


Fig. 18 LED Current vs. Temperature

■ BLOCK DIAGRAM



■ PIN DESCRIPTIONS

- | | | |
|------------|---|--|
| PIN 1: SW | - Switch Pin. Connect inductor/diode here. Minimize trace area at this pin to reduce EMI. | resistor here. Calculate resistor value according to the formula:
$R_{FB} = 95mV/I_{LED}$ |
| PIN 2: GND | - Ground Pin. Tie directly to local ground plane. | PIN 4: $\overline{SHDN}$ - Shutdown pin. Tie to higher than 1.5V to enable device, 0.3V or less to disable device. |
| PIN 3: FB | - Feedback Pin. Reference voltage is 95mV. Connect cathode of lowest LED and | PIN 5: VIN - Power input pin. Bypass VIN to GND with a capacitor sitting as close to VIN as possible. |

■ APPLICATION INFORMATION

Inductor Selection

A 10 μ H inductor is recommended for most AIC1647 applications. Although small size and high efficiency are major concerns, the inductor should have low core losses at 1.2MHz and low DCR (copper wire resistance).

Capacitor Selection

The small size of ceramic capacitors makes them ideal for AIC1647 applications. X5R and X7R types are recommended because they retain their capacitance over wider ranges of voltage and temperature than other types, such as Y5V or Z5U. 1 μ F input capacitor with 1 μ F output capacitor are sufficient for most AIC1647 applications.

Diode Selection

Schottky diodes, with their low forward voltage drop and fast reverse recovery, are the ideal choices for AIC1647 applications. The forward voltage drop of a Schottky diode represents the conduction losses in the diode, while the diode capacitance (CT or CD) represents the switching losses. For diode selection, both forward voltage drop and diode capacitance need to be considered. Schottky diodes with higher current ratings usually have lower forward voltage drop and larger diode capacitance, which can cause significant switching losses at the 1.2MHz switching frequency of AIC1647. An Schottky diode rated at 100mA to 200mA is sufficient for most AIC1647 applications.

LED Current Control

LED current is controlled by feedback resistor (R_{FB} in Fig. 1). The feedback reference voltage is 95mV. The LED current is $95\text{mV} / R_{FB}$. In order to have accurate LED current, precision resistors are preferred (1% recommended). The formula for R_{FB} selection is shown below.

$$R_{FB} = 95\text{mV} / I_{LED}$$

Open-Circuit Protection

In the cases of output open circuit, when the LEDs are disconnected from the circuit or the LEDs fail, the feedback voltage will be zero. AIC1647 will then switch to a high duty cycle resulting in a high output voltage, which may cause SW pin voltage to exceed its maximum 33V rating. A zener diode can be used at the output to limit the voltage on SW pin (Figure 1). The zener voltage should be larger than the maximum forward voltage of the LED string. The current rating of the zener should be larger than 0.1mA.

Dimming Control

There are three different ways of dimming control circuits as follows:

1. Using a PWM Signal

PWM brightness control provides the widest dimming range by pulsing the LEDs on and off at full and zero current, respectively. The change of average LED current depends on the duty cycle of the PWM signal. Typically, a 0.1kHz to 1kHz PWM signal is used. Two applications of PWM dimming with AIC1647 are shown in Figure 19 and Figure 20. One, as Figure 19, uses PWM signal to drive $\overline{\text{SHDN}}$ pin directly for dimming control. The other, as Figure 20, employs PWM signal going through a resistor to drive FB pin. If the $\overline{\text{SHDN}}$ pin is used, the increase of duty cycle results in LED brightness enhancement. If the FB pin is used, on the contrary, the increase of duty cycle will decrease its brightness. In this application, LEDs are dimmed by FB pin and turned off completely by $\overline{\text{SHDN}}$.

2. Using a DC Voltage

For some applications, the preferred method of a dimming control uses a variable DC voltage to adjust LED current. The dimming control using a DC voltage is shown in Figure 21. Cautiously selecting R1 and R2 is essential so that the current from the variable DC source is much smaller than the LED current and much larger

than the FB pin bias current. With a VDC ranging from 0V to 5V, the selection of resistors in Figure 21 results in dimming control of LED current from 20mA to 0mA, respectively.

3. Using a Filtered PWM Signal

Filtered PWM signal can be considered as an adjustable DC voltage. It can be used to replace the variable DC voltage source in dimming control. The circuit is shown in Figure 22.

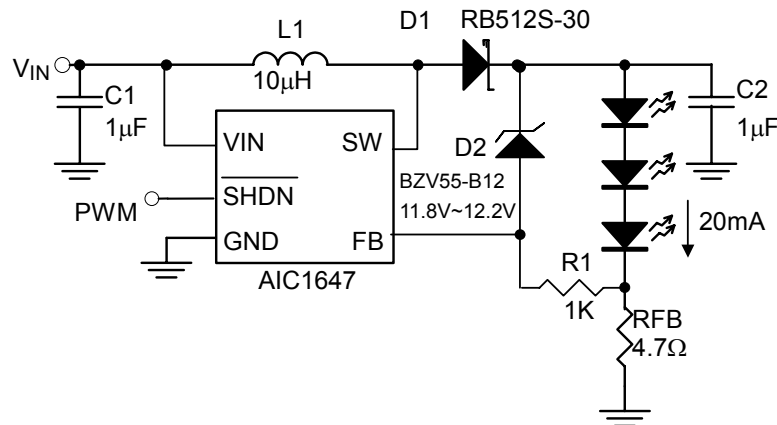


Fig. 19 Dimming Control Using a PWM Signal with Open-Circuit Protection

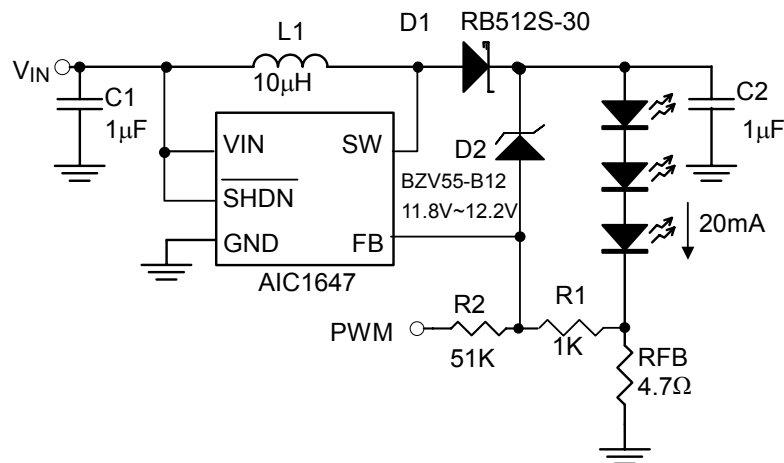


Fig. 20 Dimming Control Using a PWM Signal

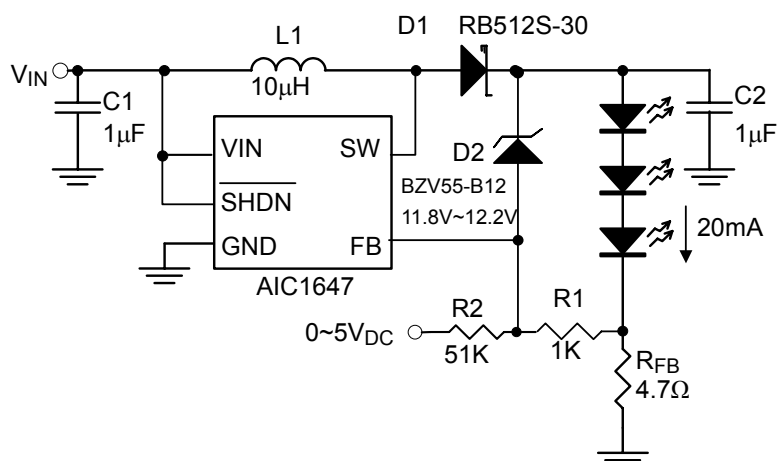


Fig. 21 Dimming Control Using a DC Voltage

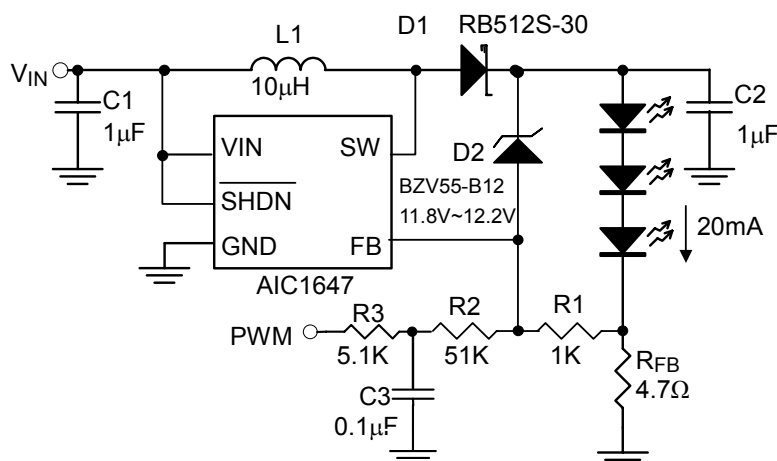


Fig. 22 Dimming Control Using a Filter PWM Signal

■ APPLICATION EXAMPLE

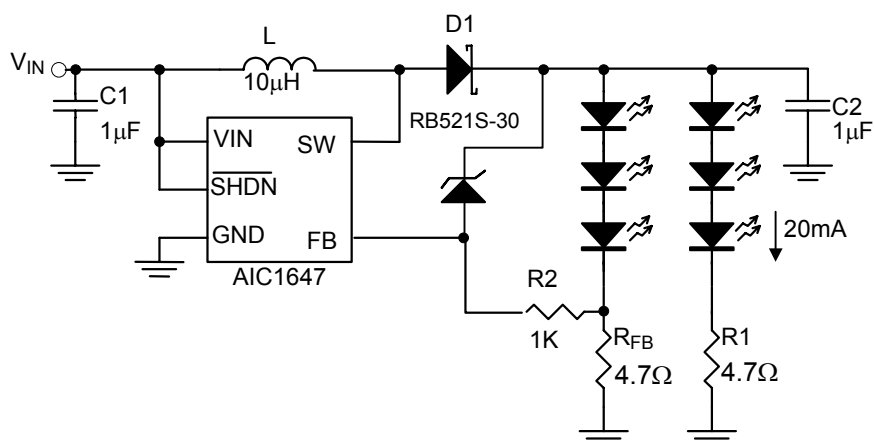
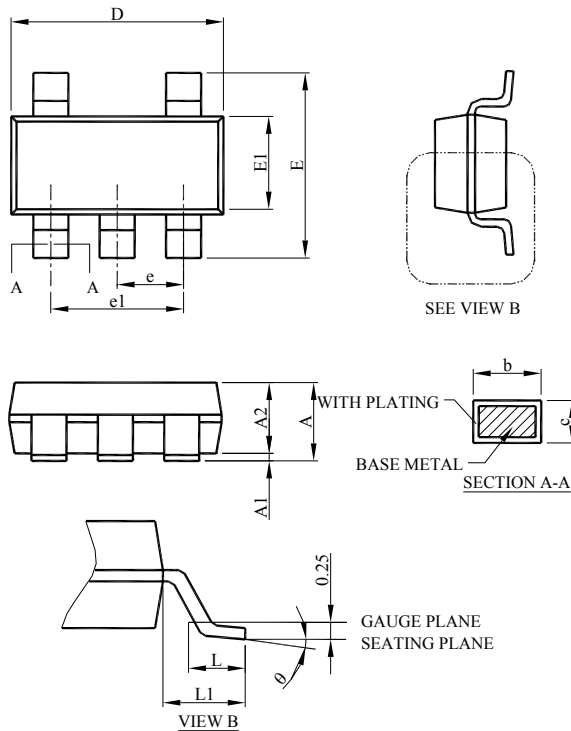


Fig. 23 Six white LEDs application in Li-Ion Battery

PHYSICAL DIMENSIONS (unit: mm)

SOT-23-5



S Y M B O L	SOT-25	
	MILLIMETERS	
	MIN.	MAX.
A	0.95	1.45
A1	0.05	0.15
A2	0.90	1.30
b	0.30	0.50
c	0.08	0.22
D	2.80	3.00
E	2.60	3.00
E1	1.50	1.70
e	0.95 BSC	
e1	1.90 BSC	
L	0.30	0.60
L1	0.60 REF	
θ	0°	8°

Note:

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