

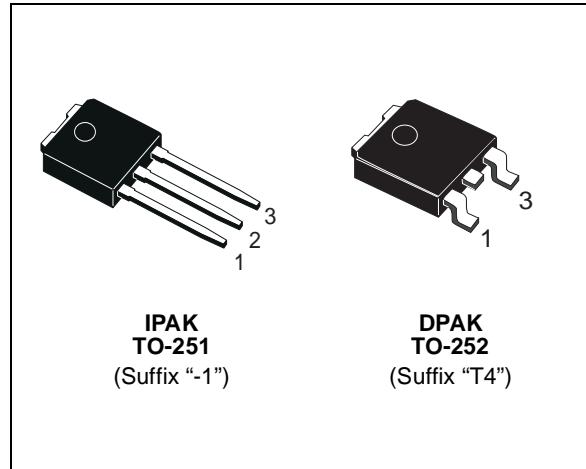


# STD60NH03L

## N-CHANNEL 30V - 0.0075 Ω - 60A DPAK/IPAK STripFET™ III POWER MOSFET

TYPE	V <sub>DSS</sub>	R <sub>DS(on)</sub>	I <sub>D</sub>
STD60NH03L	30 V	< 0.009 Ω	60 A

- TYPICAL R<sub>DS(on)</sub> = 0.0075 Ω @ 10 V
- TYPICAL R<sub>DS(on)</sub> = 0.009 Ω @ 5 V
- R<sub>DS(ON)</sub> \* Qg INDUSTRY's BENCHMARK
- CONDUCTION LOSSES REDUCED
- SWITCHING LOSSES REDUCED
- LOW THRESHOLD DEVICE
- THROUGH-HOLE IPAK (TO-251) POWER PACKAGE IN TUBE (SUFFIX "-1")
- SURFACE-MOUNTING DPAK (TO-252) POWER PACKAGE IN TAPE & REEL (SUFFIX "T4")



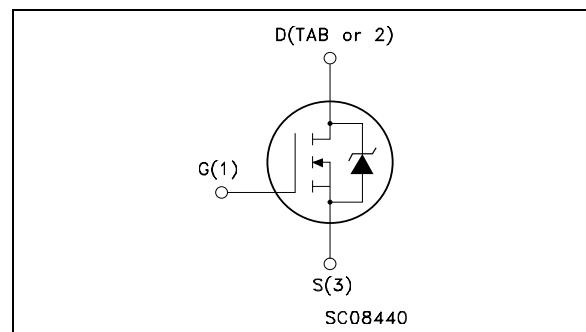
### DESCRIPTION

The STD60NH03L utilizes the latest advanced design rules of ST's proprietary STripFET™ technology. This is suitable for the most demanding DC-DC converter application where high efficiency is to be achieved.

### APPLICATIONS

- SPECIFICALLY DESIGNED AND OPTIMISED FOR HIGH EFFICIENCY DC/DC CONVERTERS

### INTERNAL SCHEMATIC DIAGRAM



### Ordering Information

SALES TYPE	MARKING	PACKAGE	PACKAGING
STD60NH03LT4	D60NH03L	TO-252	TAPE & REEL
STD60NH03L-1	D60NH03L	TO-251	TUBE

### ABSOLUTE MAXIMUM RATINGS

Symbol	Parameter	Value	Unit
V <sub>DS</sub>	Drain-source Voltage (V <sub>GS</sub> = 0)	30	V
V <sub>DGR</sub>	Drain-gate Voltage (R <sub>GS</sub> = 20 kΩ)	30	V
V <sub>GS</sub>	Gate-source Voltage	± 20	V
I <sub>D</sub>	Drain Current (continuous) at T <sub>C</sub> = 25°C	60	A
I <sub>D</sub>	Drain Current (continuous) at T <sub>C</sub> = 100°C	43	A
I <sub>DM</sub> <sup>(1)</sup>	Drain Current (pulsed)	240	A
P <sub>tot</sub>	Total Dissipation at T <sub>C</sub> = 25°C	70	W
	Derating Factor	0.47	W/°C
E <sub>AS</sub> <sup>(2)</sup>	Single Pulse Avalanche Energy	300	mJ
T <sub>stg</sub>	Storage Temperature	-55 to 175	°C
T <sub>j</sub>	Max. Operating Junction Temperature		

# STD60NH03L

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## THERMAL DATA

Rthj-case Rthj-amb Rthj-pcb T <sub>I</sub>	Thermal Resistance Junction-case Thermal Resistance Junction-ambient Thermal Resistance Junction-pcb(#) Maximum Lead Temperature For Soldering Purpose	Max Max Max	2.14 100 43 275	°C/W °C/W °C/W °C
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(#) When Mounted on 1 inch<sup>2</sup> FR-4 board, 2 oz of Cu.

## ELECTRICAL CHARACTERISTICS (T<sub>CASE</sub> = 25 °C UNLESS OTHERWISE SPECIFIED)

OFF

Symbol	Parameter	Test Conditions	Min.	Typ.	Max.	Unit
V <sub>(BR)DSS</sub>	Drain-source Breakdown Voltage	I <sub>D</sub> = 250 μA, V <sub>GS</sub> = 0	30			V
I <sub>DSS</sub>	Zero Gate Voltage Drain Current (V <sub>GS</sub> = 0)	V <sub>DS</sub> = Max Rating V <sub>DS</sub> = Max Rating T <sub>C</sub> = 125°C			1 10	μA μA
I <sub>GSS</sub>	Gate-body Leakage Current (V <sub>DS</sub> = 0)	V <sub>GS</sub> = ± 20V			±100	nA

## ON (4)

Symbol	Parameter	Test Conditions	Min.	Typ.	Max.	Unit
V <sub>GS(th)</sub>	Gate Threshold Voltage	V <sub>DS</sub> = V <sub>GS</sub> I <sub>D</sub> = 250 μA	1			V
R <sub>D(on)</sub>	Static Drain-source On Resistance	V <sub>GS</sub> = 10 V I <sub>D</sub> = 30 A V <sub>GS</sub> = 5 V I <sub>D</sub> = 30 A		0.0075 0.009	0.009 0.017	Ω Ω

## DYNAMIC

Symbol	Parameter	Test Conditions	Min.	Typ.	Max.	Unit
g <sub>fs</sub> (4)	Forward Transconductance	V <sub>DS</sub> = 15 V I <sub>D</sub> = 18 A		25		S
C <sub>iss</sub> C <sub>oss</sub> C <sub>rss</sub>	Input Capacitance Output Capacitance Reverse Transfer Capacitance	V <sub>DS</sub> = 10V f = 1 MHz V <sub>GS</sub> = 0		2200 380 49		pF pF pF
R <sub>G</sub>	Gate Input Resistance	f = 1 MHz Gate DC Bias = 0 Test Signal Level = 20 mV Open Drain		1.5		Ω

## ELECTRICAL CHARACTERISTICS (continued)

## SWITCHING ON

Symbol	Parameter	Test Conditions	Min.	Typ.	Max.	Unit
$t_{d(on)}$ $t_r$	Turn-on Delay Time Rise Time	$V_{DD} = 15 \text{ V}$ $I_D = 30 \text{ A}$ $R_G = 4.7 \Omega$ $V_{GS} = 5 \text{ V}$ (Resistive Load, Figure 3)		21 95		ns ns
$Q_g$ $Q_{gs}$ $Q_{gd}$	Total Gate Charge Gate-Source Charge Gate-Drain Charge	$V_{DD} = 15 \text{ V}$ $I_D = 60 \text{ A}$ $V_{GS} = 5 \text{ V}$		15.7 8.3 3.4	21	nC nC nC
$Q_{gls}^{(4)}$	Third-quadrant Gate Charge	$V_{DS} < 0 \text{ V}$ $V_{GS} = 10 \text{ V}$		15		nC

## SWITCHING OFF

Symbol	Parameter	Test Conditions	Min.	Typ.	Max.	Unit
$t_{d(off)}$ $t_f$	Turn-off Delay Time Fall Time	$V_{DD} = 15 \text{ V}$ $I_D = 30 \text{ A}$ $R_G = 4.7 \Omega$ , $V_{GS} = 5 \text{ V}$ (Resistive Load, Figure 3)		19 15		ns ns

## SOURCE DRAIN DIODE

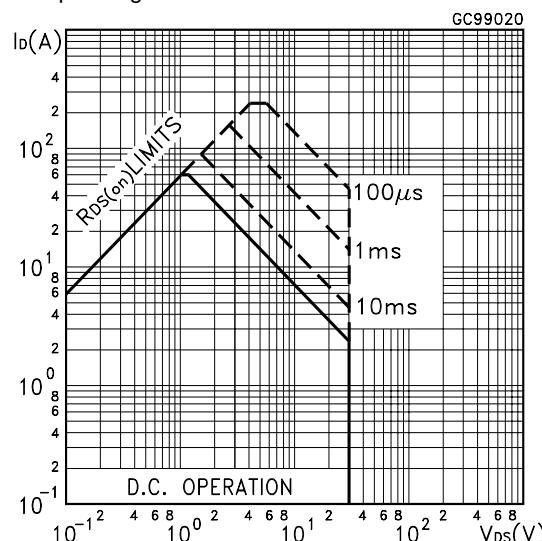
Symbol	Parameter	Test Conditions	Min.	Typ.	Max.	Unit
$I_{SD}$ $I_{SDM}$	Source-drain Current Source-drain Current (pulsed)				60 240	A A
$V_{SD}$	Forward On Voltage	$I_{SD} = 30 \text{ A}$ $V_{GS} = 0$			1.4	V
$t_{rr}$ $Q_{rr}$ $I_{RRM}$	Reverse Recovery Time Reverse Recovery Charge Reverse Recovery Current	$I_{SD} = 60 \text{ A}$ $di/dt = 100 \text{ A}/\mu\text{s}$ $V_{DD} = 20 \text{ V}$ $T_j = 150^\circ\text{C}$ (see test circuit, Figure 5)		32 51 3.2		ns nC A

(1) Pulse width limited by safe operating area  
 (2) Starting  $T_j = 25^\circ\text{C}$ ,  $I_D = 30\text{A}$ ,  $V_{DD} = 20\text{V}$

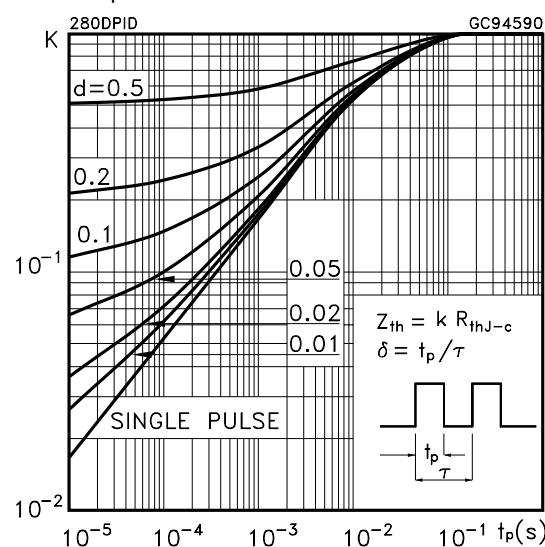
(3) Pulsed: Pulse duration = 300  $\mu\text{s}$ , duty cycle 1.5 %.

(4) Gate charge for synchronous operation. See Appendix A

## Safe Operating Area

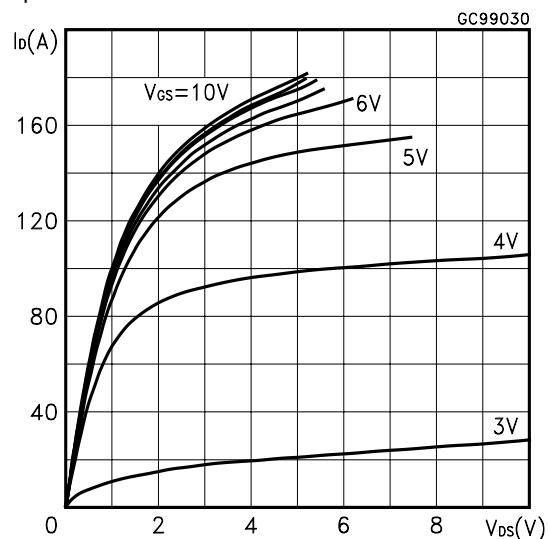


## Thermal Impedance

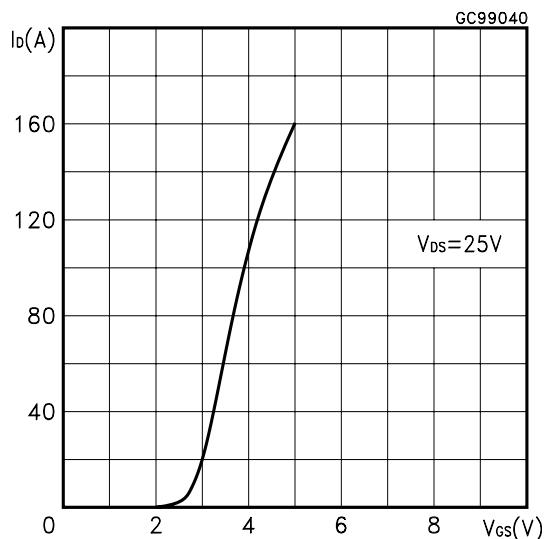


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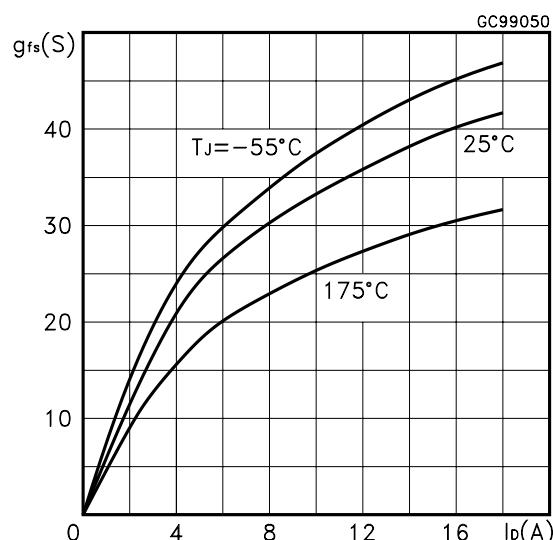
Output Characteristics



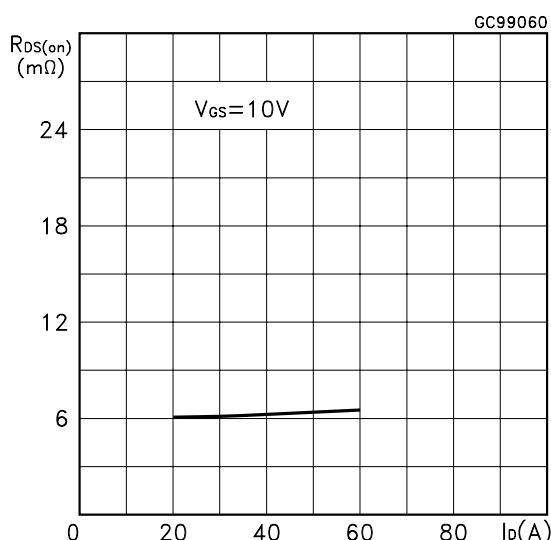
Transfer Characteristics



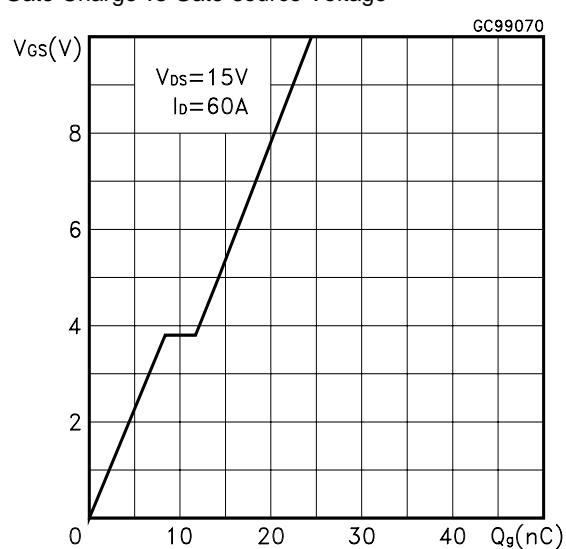
Transconductance



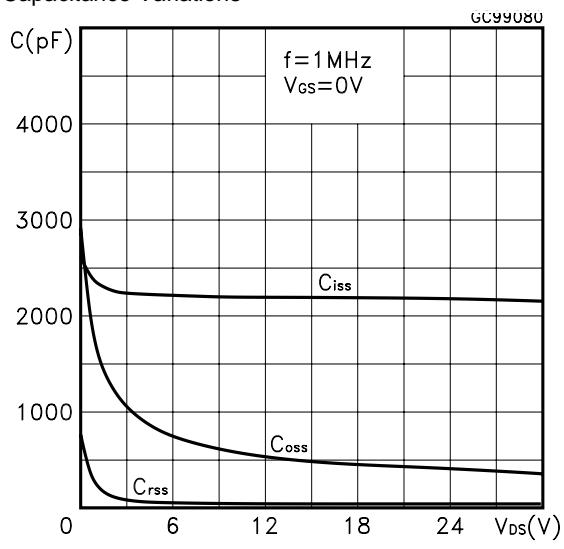
Static Drain-source On Resistance



Gate Charge vs Gate-source Voltage

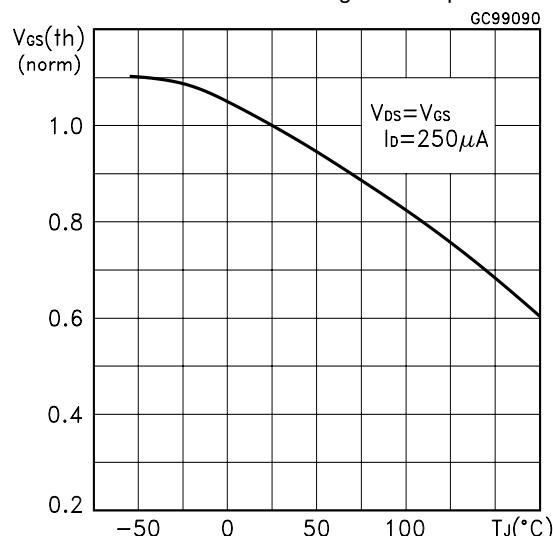


Capacitance Variations

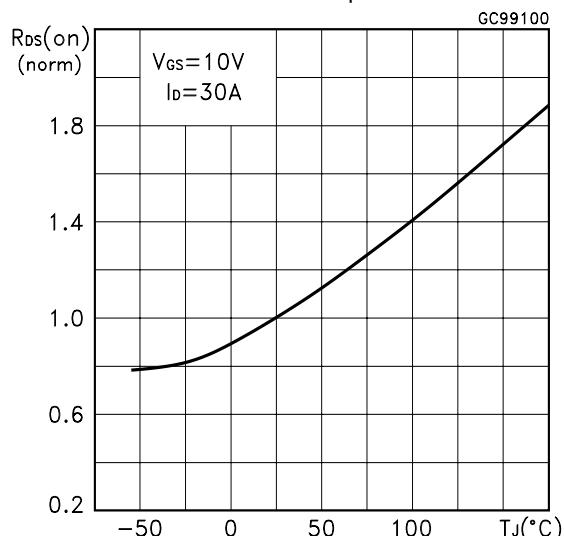


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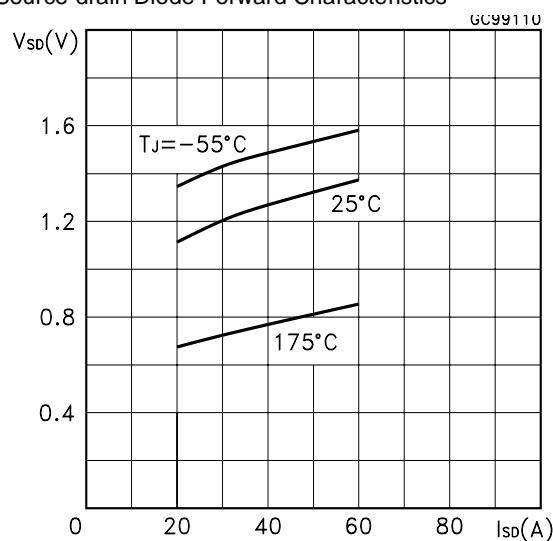
Normalized Gate Threshold Voltage vs Temperature



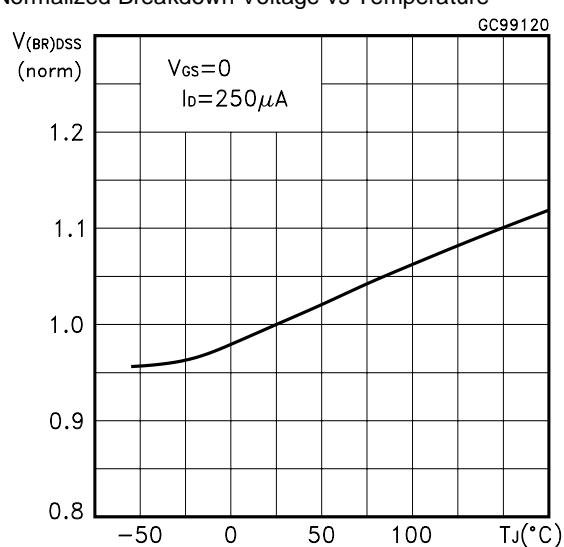
Normalized on Resistance vs Temperature



Source-drain Diode Forward Characteristics

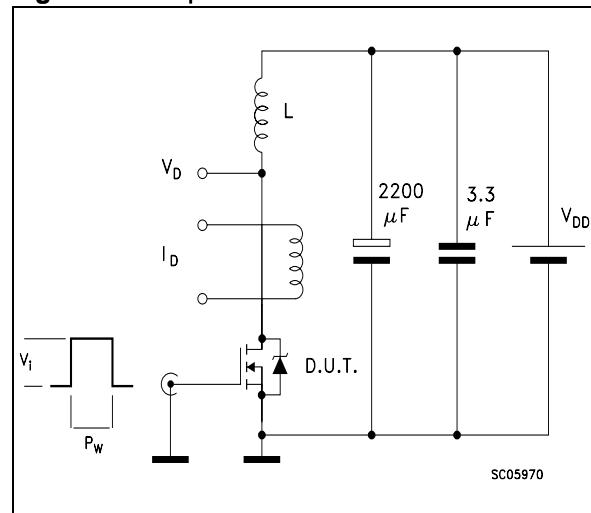


Normalized Breakdown Voltage vs Temperature

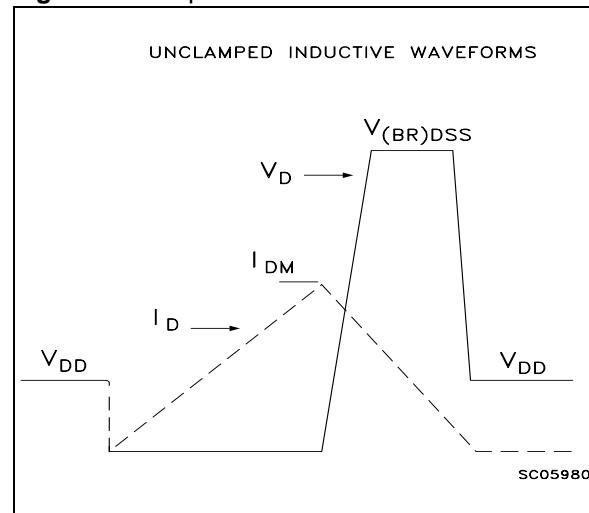


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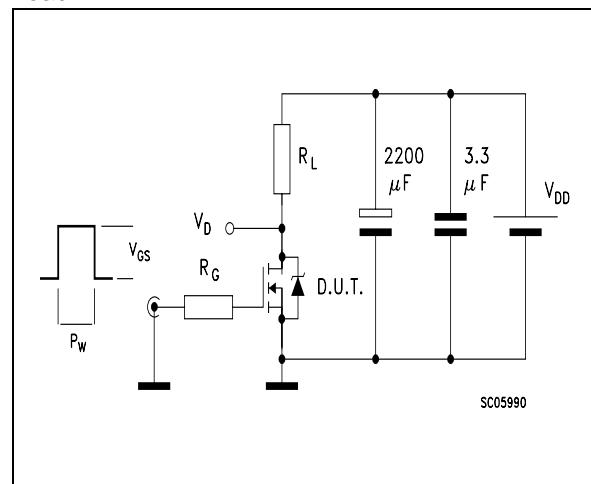
**Fig. 1: Unclamped Inductive Load Test Circuit**



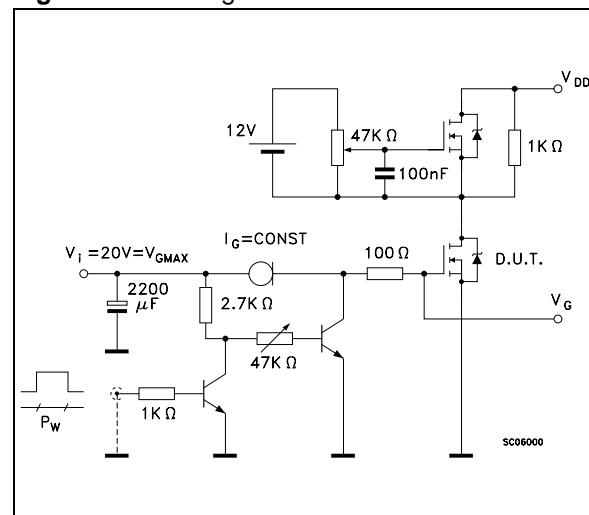
**Fig. 2: Unclamped Inductive Waveform**



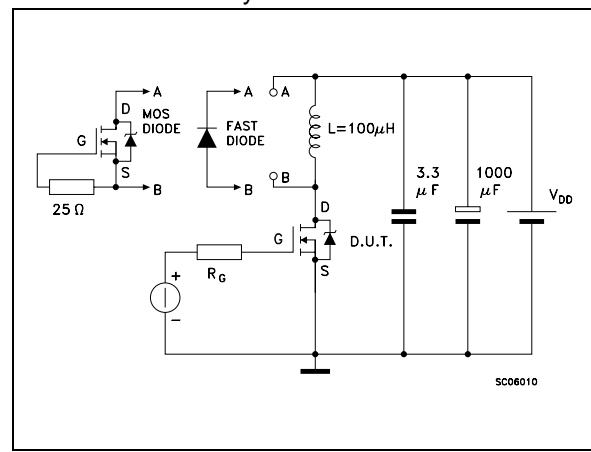
**Fig. 3: Switching Times Test Circuits For Resistive Load**



**Fig. 4: Gate Charge test Circuit**

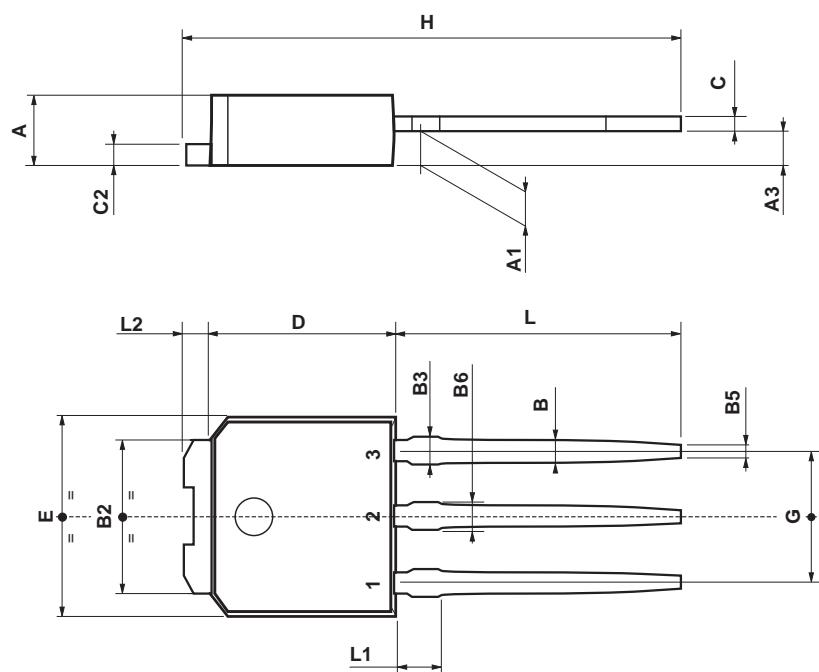


**Fig. 5: Test Circuit For Inductive Load Switching And Diode Recovery Times**



## TO-251 (IPAK) MECHANICAL DATA

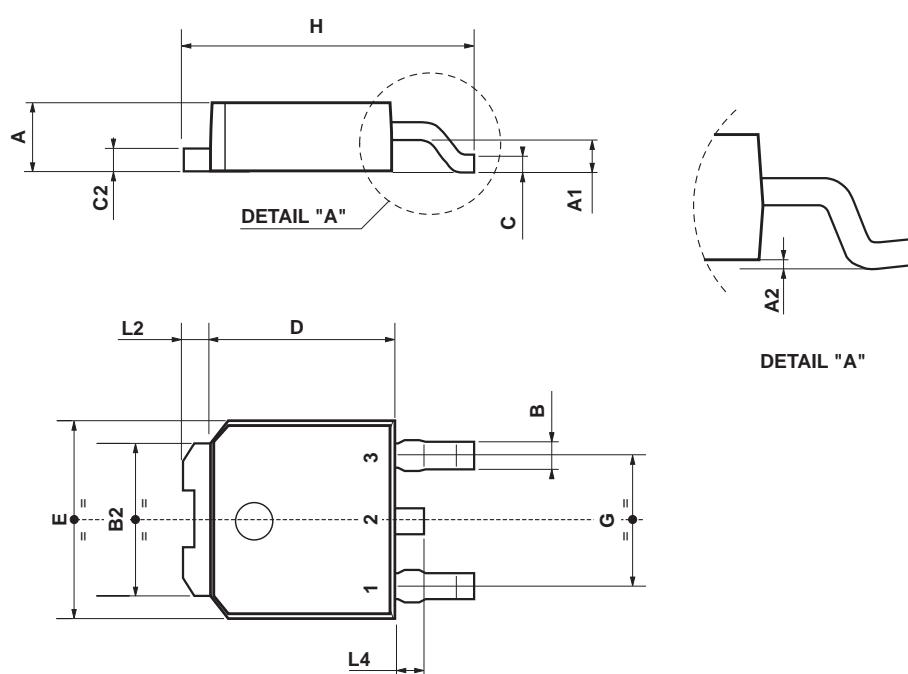
DIM.	mm			inch		
	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.
A	2.2		2.4	0.086		0.094
A1	0.9		1.1	0.035		0.043
A3	0.7		1.3	0.027		0.051
B	0.64		0.9	0.025		0.031
B2	5.2		5.4	0.204		0.212
B3			0.85			0.033
B5		0.3			0.012	
B6			0.95			0.037
C	0.45		0.6	0.017		0.023
C2	0.48		0.6	0.019		0.023
D	6		6.2	0.236		0.244
E	6.4		6.6	0.252		0.260
G	4.4		4.6	0.173		0.181
H	15.9		16.3	0.626		0.641
L	9		9.4	0.354		0.370
L1	0.8		1.2	0.031		0.047
L2		0.8	1		0.031	0.039



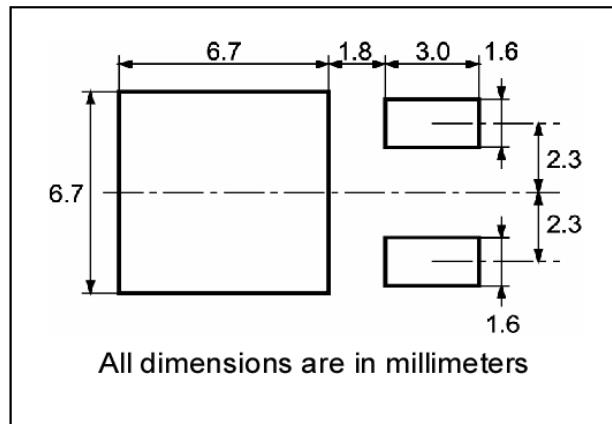
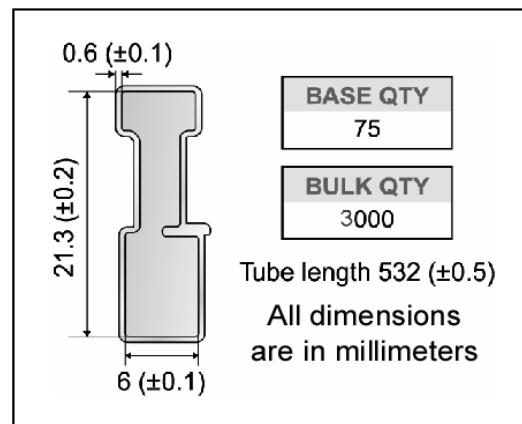
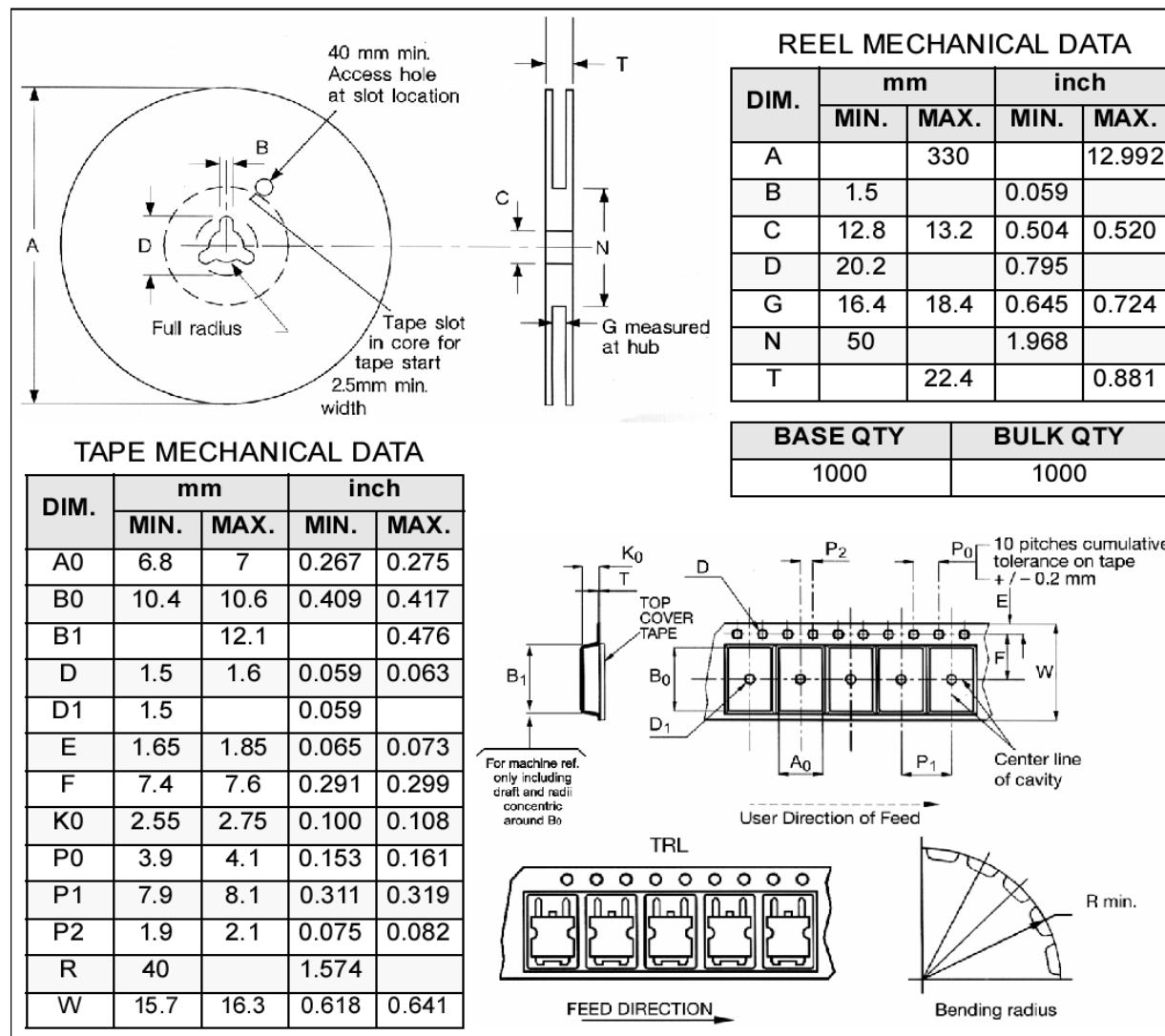
0068771-E

## TO-252 (DPAK) MECHANICAL DATA

DIM.	mm			inch		
	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.
A	2.2		2.4	0.086		0.094
A1	0.9		1.1	0.035		0.043
A2	0.03		0.23	0.001		0.009
B	0.64		0.9	0.025		0.035
B2	5.2		5.4	0.204		0.212
C	0.45		0.6	0.017		0.023
C2	0.48		0.6	0.019		0.023
D	6		6.2	0.236		0.244
E	6.4		6.6	0.252		0.260
G	4.4		4.6	0.173		0.181
H	9.35		10.1	0.368		0.397
L2		0.8			0.031	
L4	0.6		1	0.023		0.039

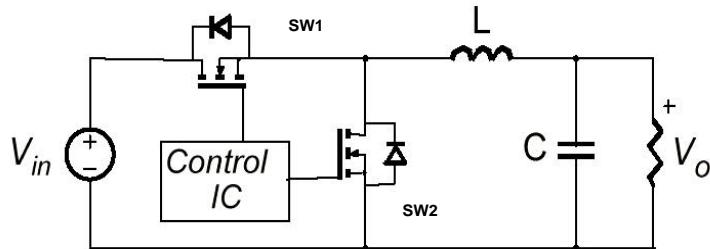


0068772-B

**DPAK FOOTPRINT****TUBE SHIPMENT (no suffix)\*****TAPE AND REEL SHIPMENT (suffix "T4")\***

## APPENDIX A

### Buck Converter: Power Losses Estimation



The power losses associated with the FETs in a Synchronous Buck converter can be estimated using the equations shown in the table below. The formulas give a good approximation, for the sake of performance comparison, of how different pairs of devices affect the converter efficiency. However a very important parameter, the working temperature, is not considered. The real device behavior is really dependent on how the heat generated inside the devices is removed to allow for a safer working junction temperature.

The low side (SW2) device requires:

- Very low  $R_{DS(on)}$  to reduce conduction losses
- Small  $Q_{gls}$  to reduce the gate charge losses
- Small  $C_{oss}$  to reduce losses due to output capacitance
- Small  $Q_{rr}$  to reduce losses on SW<sub>1</sub> during its turn-on
- The  $C_{gd}/C_{gs}$  ratio lower than  $V_{th}/V_{gg}$  ratio especially with low drain to source voltage to avoid the cross conduction phenomenon;

The high side (SW1) device requires:

- Small  $R_g$  and  $L_s$  to allow higher gate current peak and to limit the voltage feedback on the gate
- Small  $Q_g$  to have a faster commutation and to reduce gate charge losses
- Low  $R_{DS(on)}$  to reduce the conduction losses.

## STD60NH03L

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		High Side Switch (SW1)	Low Side Switch (SW2)
$P_{\text{conduction}}$		$R_{DS(\text{on})\text{SW1}} * I_L^2 * d$	$R_{DS(\text{on})\text{SW2}} * I_L^2 * (1-d)$
$P_{\text{switching}}$		$V_{in} * (Q_{gsth(\text{SW1})} + Q_{gd(\text{SW1})}) * f * \frac{I_L}{I_g}$	Zero Voltage Switching
$P_{\text{diode}}$	Recovery	Not Applicable	${}^1 V_{in} * Q_{rr(\text{SW2})} * f$
	Conduction	Not Applicable	$V_{f(\text{SW2})} * I_L * t_{\text{deadtime}} * f$
$P_{\text{gate}(Q_G)}$		$Q_{g(\text{SW1})} * V_{gg} * f$	$Q_{gls(\text{SW2})} * V_{gg} * f$
$P_{Qoss}$		$\frac{V_{in} * Q_{oss(\text{SW1})} * f}{2}$	$\frac{V_{in} * Q_{oss(\text{SW2})} * f}{2}$

Parameter	Meaning
$d$	Duty-cycle
$Q_{gsth}$	Post threshold gate charge
$Q_{gls}$	Third quadrant gate charge
$P_{\text{conduction}}$	On state losses
$P_{\text{switching}}$	On-off transition losses
$P_{\text{diode}}$	Conduction and reverse recovery diode losses
$P_{\text{gate}}$	Gate drive losses
$P_{Qoss}$	Output capacitance losses

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<sup>1</sup> Dissipated by SW1 during turn-on

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