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LM9072 Dual Tracking Low-Dropout System Regulator

National Semiconductor

# LM9072 Dual Tracking Low-Dropout System Regulator

### **General Description**

The LM9072 is a high performance voltage regulator system with operational and protection features that address many requirements of automotive applications. Two regulated outputs are provided. The main regulator provides a precision 2% maximum tolerance 5V output at 350 mA with a low dropout characteristic. The second regulator provides a 5V output that tracks the main regulator output voltage within 1.5% with load currents up to 80 mA. The tracking output is ideal for use in powering remotely located sensors with outputs that are ratiometric to the main system supply. This output is fully protected from short circuits to ground or the unregulated input supply (ignition or battery potentials in automotive applications).

The LM9072 also contains a programmable delayed system reset output. Two control inputs are provided. An ON/OFF input intended for connection to an ignition switch, and a Keep Alive input to allow a system to remain powered after ignition has been switched OFF.

For EMC concerns the LM9072 remains fully operational and does not generate false rest signals while subjected to, 1 MHz to 400 MHz bulk current injection signals greater than 100 mA on the input supply and tracking output lines.

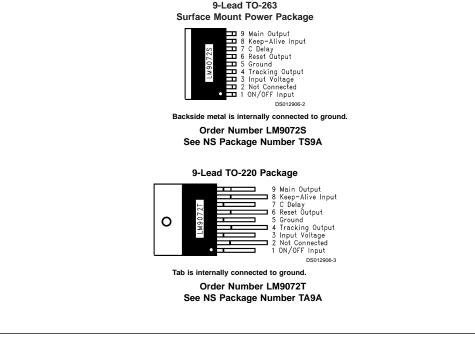
### Features

- Two 5V regulated outputs:
- 350 mA, 2% Main output
- 80 mA, 1.5% Tracking output
- Good EMI (1 MHz to 400 MHz, BCI) immunity
- Separate ON/OFF and Keep-Alive control inputs
- Less than 100 µA quiescent current in OFF state
- Programmable delayed reset output
- Input transient protection over 60V to -45V
- Tracking output protected from shorts to battery
- Less than 1V dropout at full load
- -40°C to +125°C operating temperature range
- Surface mount TO-263 Power Package and Standard TO-220 power package

# **Typical Applications**

- Automotive module supply power conditioning
- Remote sensor biasing
- Ratiometric to supply sensor detection
- Continuous operation for save routines and EPROM programming after power down command
- Safety related systems EMC operational

# **Connection Diagrams and Ordering Information**



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### Absolute Maximum Ratings (Note 1)

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/ Distributors for availability and specifications.

Input Voltage (Continuous)	-45V to 45V
Input Voltage (Transient, t ≤ 400 ms)	60V
Forced Output Voltages	
Main Output	-0.3V to 7V
Tracking Output	-0.3V to 27V
ON/OFF Input Voltage (Note 6)	-0.3V to 16V
ON/OFF Input Current	± 20 mA
Keep Alive In, Reset Out, C <sub>DELAY</sub> Voltage	-0.3V to 7V

Junction Temperature	150°C
Storage Temperature Range	-65°C to +150°C
ESD Susceptibility (Note 2)	2000V
Lead Temperature (Soldering, 10 seconds)	265°C

# **Operating Ratings** (Note 1)

Input Voltage Range	6V to 27V
Ambient Temperature Range	-40°C to +125°C
Thermal Resistance, $\theta_{J-C}$	4°C/W
Thermal Resistance, $\theta_{J-A}$	43°C/W

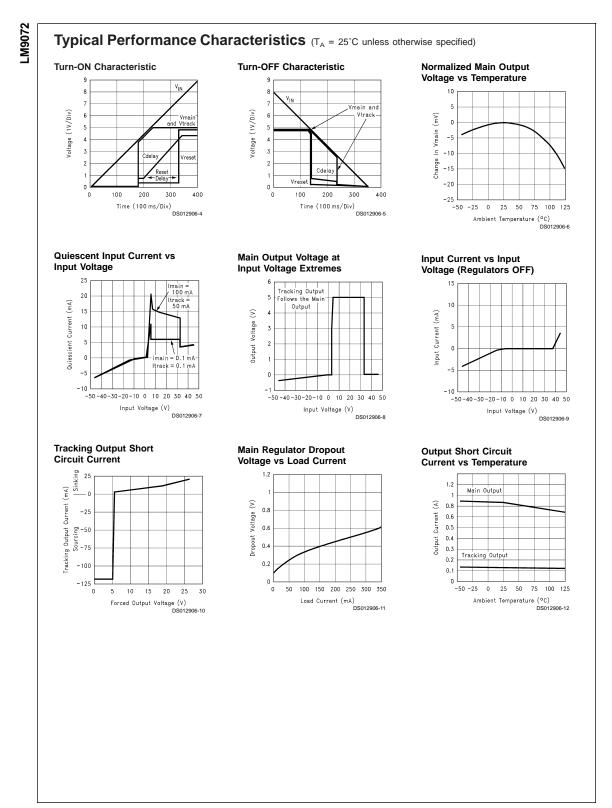
## **Electrical Characteristics**

 $6.0V \le V_{IN} \le 19V$ ,  $T_{CASE} = 25^{\circ}C$ , unless otherwise specified.  $C_{OUT} \ge 6 \ \mu F$  with  $0.3\Omega \le ESR \le 3\Omega$  on each regulator output.

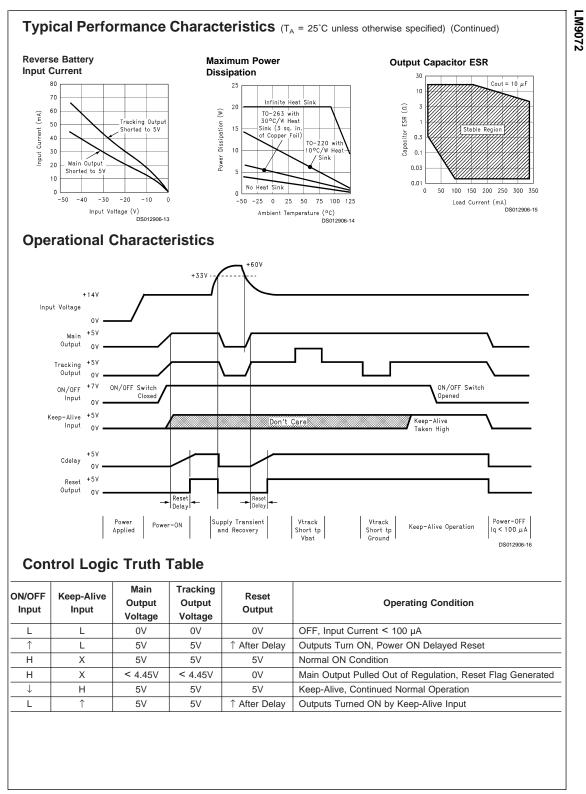
Symbol	Parameter	Conditions	Min	Max	Units
MAIN REG	GULATOR				
V <sub>MAIN</sub>	Output Voltage	$5 \text{ mA} \le I_{LOAD} \le 350 \text{ mA}$ -40°C $\le T_{CASE} \le 125°C$	4.9	5.1	V
		$19V \le V_{IN} \le V_{SD},$	4.8	5.2	V
		$5 \text{ mA} \leq I_{LOAD} \leq 350 \text{ mA}$			
		$-40^{\circ}C \le T_{CASE} \le 125^{\circ}C$			
R <sub>MLOAD</sub>	Load Regulation	$V_{\text{IN}}$ = 16V, 5 mA $\leq$ $I_{\text{LOAD}} \leq$ 350 mA		25	mV
R <sub>MLINE</sub>	Line Regulation	$I_{\text{LOAD}}$ = 350 mA, 8V $\leq V_{\text{IN}} \leq 16V$		25	mV
V <sub>MDO</sub>	Dropout Voltage, V <sub>IN</sub> -V <sub>MAIN</sub>	$V_{IN}$ > 5.5V, 5 mA $\leq I_{LOAD} \leq$ 350 mA (Note 5)		0.8	V
V <sub>SD</sub>	Overvoltage Shutdown Threshold		30	36	V
I <sub>MSC</sub>	Output Short Circuit Current	$R_{L} = 1\Omega$	450	1000	mA
R <sub>MRR</sub>	Ripple Rejection	$V_{IN}$ = 9V, 50 $\leq$ Freq $\leq$ 20 kHz, $V_{RIPPLE}$ = 4V <sub>P-P</sub>	40		dB
TRACKING	G REGULATOR				
V <sub>track</sub>	Output Voltage	$1 \text{ mA} \le I_{LOAD} \le 80 \text{ mA}$ -40°C $\le T_{CASE} \le 125$ °C	4.85	5.15	V
		$19V \le V_{IN} \le V_{SD}$	4.725	5.275	V
		$1 \text{ mA} \leq I_{LOAD} \leq 80 \text{ mA}$			
		$-40^{\circ}C \le T_{CASE} \le 125^{\circ}C$			
V <sub>error</sub>	Output Tracking Error (V <sub>MAIN</sub> -V <sub>TRACK</sub> )	$1 \text{ mA} \leq I_{LOAD} \leq 80 \text{ mA}$	-50	50	mV
V <sub>TDO</sub>	Dropout Voltage,	$V_{IN}$ > 5.5V, 1 mA $\leq I_{LOAD} \leq$ 80 mA		0.8	V
	V <sub>IN</sub> -V <sub>TRACK</sub>	(Note 5)		0.0	v
V <sub>SD</sub>	Overvoltage Shutdown Threshold		30	36	V
I <sub>TSC</sub>	Output Short Circuit Current	$R_{L} = 1\Omega$		200	mA
V <sub>TSC</sub>	Output Short Circuit Voltage	No Effect On Other Functions	-2	27	V
R <sub>TRR</sub>	Ripple Rejection	$V_{IN}$ = 9V, 50 $\leq$ Freq $\leq$ 20 kHz,	40		dB
		$V_{RIPPLE} = 4V_{P-P}$	40		u D
INPUT CU	RRENT				
I <sub>qOFF</sub>	Quiescent Input Current with	$8V \le V_{IN} \le 16V$		40	μA
	Both Regulators OFF	$16V \le V_{IN} \le 42V$		10	mA
l <sub>q</sub>	No Load Quiescent Current	$8V \le V_{IN} \le 19V, I_L = 0 \text{ mA}$		15	mA
In <sub>on</sub>	Additional Input Current with Both Regulators ON	$V_{IN} > 8V$ , $I_{Ltotal} = I_{Lmain} + I_{Ltrack}$ $I_{Ltotal} = 350 \text{ mA} + 80 \text{ mA} = 430 \text{ mA}$		1.2	x I <sub>Ltota</sub>
lin <sub>do</sub>	Additional Input Current in Dropout	$\begin{array}{l} 0 V < V_{IN} < 8 V, \mbox{ (Note 4)} \\ I_{Ltotal} = I_{Lmain} + I_{Ltrack} \\ I_{Ltotal} = 350 \mbox{ mA} + 80 \mbox{ mA} = 430 \mbox{ mA} \end{array}$		1.5	x I <sub>Ltota</sub>

Symbol	Parameter	Conditions	Min	Max	Units
	JTPUT				
/ <sub>THRL</sub>	Low Switching Threshold	V <sub>MAIN</sub> Output Controls Reset	4.45	4.75	V
/ <sub>THRH</sub>	High Switching Threshold	V <sub>MAIN</sub> Output Controls Reset	5.40	5.75	V
/ <sub>LOW</sub>	Logic Low Output Voltage	$1V \le V_{MAIN} \le V_{THRL},$ $R_{RESET}$ = 50 k $\Omega$ to $V_{MAIN}$		0.4	V
/ <sub>HIGH</sub>	Logic High Output Voltage	Normal Operation, $V_{THRL} \le V_{MAIN} \le V_{THRH}$ , $I_{SOURCE} = 0$	V <sub>MAIN</sub> - 50 mV	V <sub>MAIN</sub>	V
₹ <sub>P-U</sub>	Internal Output Pull-Up Resistance		2.4	6.0	kΩ
DELAY	Reset Delay Interval	$C_{DELAY}$ = 0.1 µF (Low Leakage), I <sub>DELAY</sub> for Charging the Delay Capacitor is Typically 6 µA	35	70	ms
RISE	Output Rise Time	From 10% V <sub>MAIN</sub> to 90% V <sub>MAIN</sub> C <sub>LRESET</sub> = 50 pF		1.5	μs
- FALL	Output Fall Time	From 90% $V_{MAIN}$ to 10% $V_{MAIN}$ C <sub>LRESET</sub> = 50 pF		0.5	μs
ONTROL	INPUTS				
/ <sub>ON</sub>	ON Threshold for ON/OFF Input	R <sub>SERIES</sub> = 22 kΩ	3.5	4.5	V
/ <sub>OFF</sub>	OFF Threshold for ON/OFF Input	R <sub>SERIES</sub> = 22 kΩ (Note 3)	1.5	2.5	V
ON/OFF	ON/OFF Input Current	$1.4V \le V_{ON/OFF} \le 4.5V$ -0.3V $\le V_{ON/OFF} \le 7V$ (Note 6)	1 _1	12 5	μA mA
DN <sub>K-A</sub>	Turn ON Threshold for Keep Alive Input		2		V
)FF <sub>K-A</sub>	Turn OFF Threshold for Keep Alive Input	(Note 3)		0.8	V
₹ <sub>P-D</sub>	Pull-Down Resistance at Keep Alive Input	$0V \le V_{K-A} \le 5V$	5	40	kΩ
Note 3: If e Note 4: The load current Note 5: The temperature	on both outputs. The peak increase in current is lir	s will turn OFF. ators are in dropout conditions. The amount of addition mited to 50% of the total load current. saturation voltage of the PNP power transistors used in h an input voltage of only 5.5V.			

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### **Application Information**

#### PIN DESCRIPTION AND FUNCTIONALITY

The LM9072 is a precision dual tracking voltage regulator optimized for use in powertrain module applications but will also find use in a wide variety of automotive and industrial applications where precision supply regulation is required in harsh operating environments. The following will describe the functionality of each of the package pins.

#### **INPUT VOLTAGE (Pin 3)**

The LM9072 has been designed to connect directly to the ignition or battery supply in automotive applications. For this type of supply the regulator has been designed to withstand up to +60V and -45V supply transients such as load dump. An overvoltage shut down protection scheme turns OFF both of the regulator outputs should supply transients exceed typically +33V to fully protect all load circuitry. This higher threshold allows normal operation with 24  $V_{\rm DC}$  applied to the input as in the event when two batteries are used to start a vehicle. Protection of the system is also provided for inadvertent reverse polarity battery connections.

The current drain on the supply line is directly proportional to the load currents on the two voltage regulators. With no load current on either output the regulator requires 15 mA maximum quiescent current for biasing internal circuitry. During dropout conditions ( $V_{IN} < 5.5V$ ) the additional input current can rise to 50% of the total load current. With less than 4V applied to the input, internal biasing circuitry shuts OFF. When *switched* OFF the regulator can remain connected to the battery supply with a current drain of less than 100 µA.

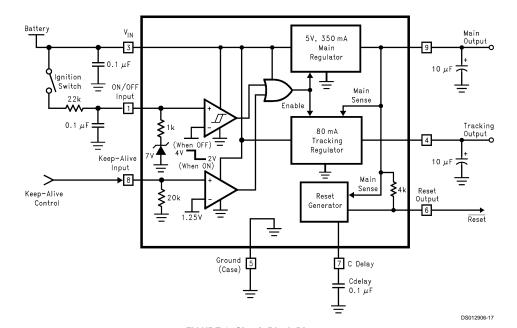


FIGURE 1. Circuit Block Diagram

### MAIN OUTPUT (Pin 9)

The Main Output regulator provides a well controlled (3% tolerance maximum) 5V supply line with a total load current ranging up to 350 mA. This relatively high level of output current is sufficient to provide power to a large number of load circuits in a variety of module applications.

This output has a short to ground current limit between 500 mA and 1A. The Main output can also withstand a short circuit to potentials up to 7V.

To maintain stability of this supply line an output bypass capacitor is required. This capacitor must be at least 6  $\mu$ F with an equivalent series resistance (ESR) between  $0.3\Omega$  and  $3\Omega$  over temperature.

The Main Output is sensed for the generation of the system reset output. Feedback from the Main Output is also used to control the output voltage of the tracking regulator.

#### TRACKING OUTPUT (Pin 4)

The Tracking Output regulator is a key feature of the LM9072. This output provides a voltage that directly tracks the main output voltage within 1.5%. This 80 mA output is provided to bias sensors and other devices located external to the main system module.

For providing remote power the Tracking Output is fully protected against short circuit connections to the battery or input supply (up to 27V) and to ground. These shorted fault conditions do not affect the operation of the main supply nor generate a reset of the system.

The tracking characteristic allows for ratiometric operation of sensors by providing power that is directly proportional to the system supply. Similar to the main output a bypass capacitor is required for stability. This capacitor should also be greater than 6  $\mu$ F with an ESR between 0.3 $\Omega$  and 3 $\Omega$ .

### Application Information (Continued)

#### RESET OUTPUT (Pin 6)

The Reset Output is an active low logic signal provided to reset a system microcontroller on power up and in the event that the Main Output supply falls out of regulation. This output is guaranteed to provide a logic low level (< 0.4V) whenever the Main Output supply is below 4.45V or is pulled above 5.75V. This general reset prevents erratic system operation which may occur with out-of-specification supply potentials.

The Reset Output has an active pull down to ground and a passive pull-up (through a 4 k $\Omega$  resistor) to the Main Output to ensure voltage compatibility with the system supply. Capacitive loading on this reset line will directly affect the rise time of the reset signal. The Reset Output will maintain a logic low level with a Main Output voltage of only 1V. Below 1V the active pull-down device switches OFF (sink current of only 500  $\mu$ A), but with such a low supply potential, system controllers are generally inoperative.

The Reset Output has a built-in delay time interval which is programmable by the selection of the delay capacitor.

#### **DELAY CAPACITOR (Pin 7)**

The Delay Capacitor ( $C_{DELAY}$ ) controls a time interval that the Reset Output remains low after the Main Output has established normal operating condition. This feature holds the system in reset for a time to allow all load circuitry to be properly biased before executing functions. This interval is applied at power-up and following any event that may trigger the system reset function.

Figure 2 illustrates the delayed reset generator. Two comparators continually monitor the Main Output supply. Window comparators C1 and C2 detect if the Main Supply is below 4.6V or exceeds 5.5V typically. If this is true (at power-on, for example) the control logic turns ON the discharge transistor and holds C<sub>DELAY</sub> low (at 0.9V). Comparator C4 then outputs a logic low system Reset signal within 2 µs after detecting the out of regulation condition. The Delay Capacitor remains discharged until the window comparator senses that the Main Output is within normal operating range (C1 and C2 outputs are both low). When this condition is met, the discharge transistor is turned OFF and C<sub>DELAY</sub> is charged positively by an internal 6  $\mu$ A current source. The Reset Output will remain low un the delay capacitor has reached 4V, at which point it will go high and the system will begin normal operation. This delay time interval is controlled by the selection of C<sub>DELAY</sub> and can be determined from the following equation:

#### $T_{DELAY} = (0.5 \times 10^6) \cdot C_{DELAY}$

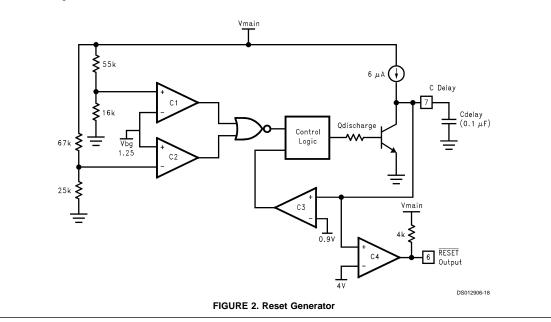
Typically a 0.1  $\mu\text{F}$  capacitor will produce a delay interval of 50 ms.

To ensure a consistent delay time interval, the discharge transistor is always latched ON by the window comparators, and can not be switched OFF to start a new delay interval until  $C_{\mathsf{DELAY}}$  has been discharged to less than or equal to 0.9V. This sets a fixed starting voltage (0.9V) and ending voltage (4V) for the charging of the Delay Capacitor.

#### ON/OFF INPUT (Pin 1)

The ON/OFF Input enables both the Main and Tracking outputs. In a typical application this input is connected to the input supply through a series resistor (nominally 22 kΩ) and a switch (Ignition, as an example). When the switch is closed this input is pulled high and switches ON both regulator outputs. This input is internally clamped to a 7V zener diode through a series 1 kΩ resistor. The external series resistor together with an optional 0.1 µF capacitor to ground are optional to provide filtering and current limiting to withstand any transients that may appear on the input supply in order to maintain normal operation of the system.

The switching threshold of the ON/OFF comparator has 2V of hysteresis to ensure noise free control of the system. To turn the regulators ON this input must be taken above 4V. To turn the system OFF the ON/OFF Input must be open circuited or taken below 2V.



### Application Information (Continued)

#### **KEEP-ALIVE INPUT (Pin 8)**

This CMOS logic level compatible input provides a system with the ability to control its own ON/OFF sequencing. The Keep-Alive Input is OR'ed with the ON/OFF Input so either one can independently control the regulators.

As shown in the Operational Characteristics a system controller can take the Keep-Alive input high at any time. If the ON/OFF switch is opened, this high level on Keep-Alive will keep the regulators ON and the entire system operational. This control is useful for providing as much time as neces-

sary for a system to perform "housekeeping" chores such as programming EEPROM with system information prior to turning itself OFF (by taking the Keep-Alive Input low) and reverting to the low guiescent current state.

A second use of the Keep-Alive Input can be from other modules which need information from the module powered by the LM9072, Figure 3. A CMOS logic high level (> 2V) on this input will power up the system as needed independent from the normal ON/OFF switch.

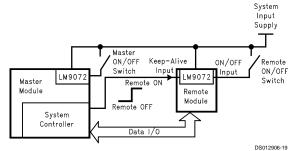


FIGURE 3. Remote ON/OFF Control

#### System Keep-Alive Operation

Figure 4 illustrates the basic concept of Keep-Alive operation. The LM9072 provides regulated supplies to an entire microcontroller based system or module including remote sensors. The system is switched ON or OFF by a switch connected to the unregulated input supply and the ON/OFF input, pin 1. When closed the regulators turn ON and the system is held in a reset state for the duration of the delayed reset interval controlled by C<sub>DELAY</sub>.

Once normal operation of the system begins, the controller needs to set an output line connected to the Keep-Alive input, pin 8, high. The system remains in normal operation until switched OFF by opening the ON/OFF switch. With Keep-Alive high the entire system remains normally biased and will remain operational until the Keep-Alive input is taken low

Transistor Q1 is shown as a means to inform the controller that the ON/OFF switch has been opened. This high level on an input line tells the controller that the system has been switched OFF. This indicates the start of the Keep-Alive interval. The system can perform whatever actions required to obtain the proper OFF state before actually powering down. These general housekeeping tasks can include putting external devices in the proper OFF condition and storing various system variables in EPROM for example. With the controller in command of the Keep-Alive interval these tasks can take whatever time necessary to complete.

When completed the controller takes the Keep-Alive input to a low level and the entire system shuts down. The LM9072 powers down to a low quiescent current mode with less than 100 µA drawn from the input supply.

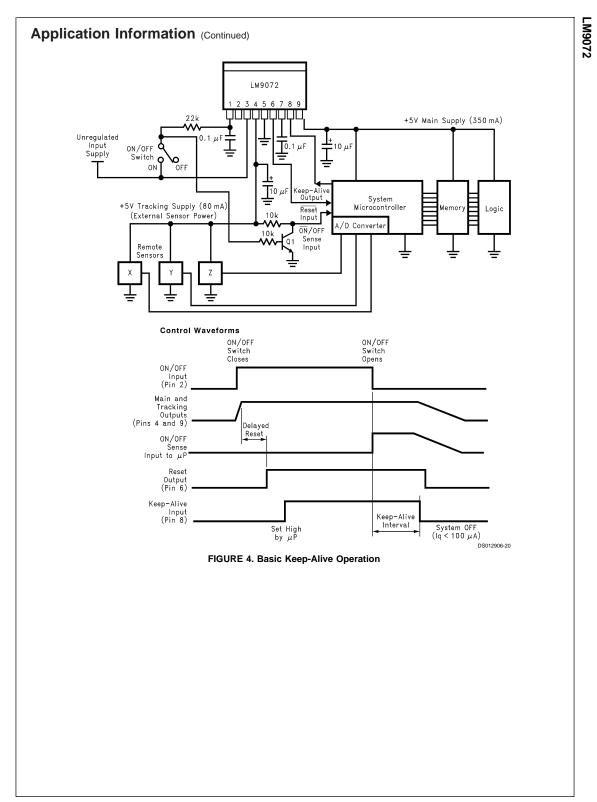
To initiate the Keep-Alive routine before actual power down, it is important for the system controller to know when the system has been switched OFF. To eliminate any interface between the controller and the ON/OFF switch and potentially noisy unregulated input supply, a simple logic scheme shown in Figure 5 can be used. With this circuitry the Reset output from the LM9072 provides the ON/OFF sensing input to the controller.

When switched OFF, the main regulator output will fall out of regulation and generate a low logic level on the Reset output. This input to the controller provides the switch OFF indication and initiates the Keep-Alive interval.

Control of the Keep-Alive duration is set by a logic 1 on the Keep-Alive output line from the system controller. This high level prevents the Reset output from resetting the entire system and also gates the Keep-Alive input signal to the LM9072. The inverted Reset signal provides a logic 1 to the Keep-Alive input of the LM9072.

The Main output will only drop out of regulation for a very short time before the Keep-Alive input turns it back on.

The Reset output remains low for the delay time interval. When it returns high the Main output switches OFF and back ON again very quickly. This continues until the system controller takes the Keep-Alive output line to a logic low level.



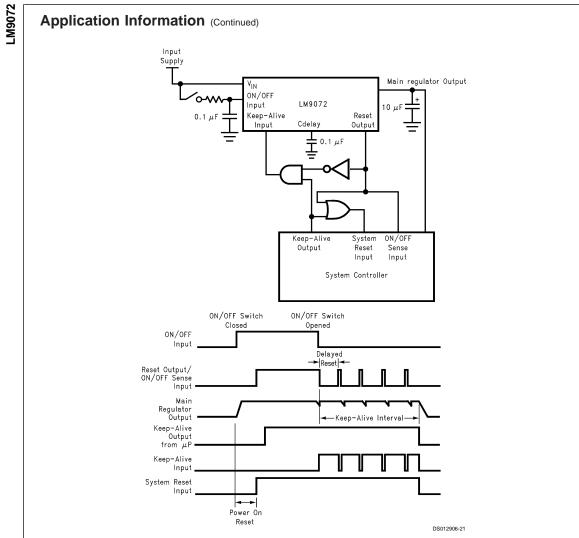


FIGURE 5. Simple Logic Configuration to Provide ON/OFF Sensing

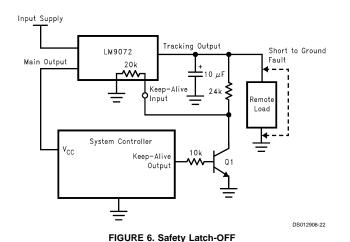
#### Safety Latch-OFF

To address a system FMEA (Failure Mode Effects Analysis) issue the Keep-Alive high level should be derived from the Tracking output regulator as shown in *Figure 6*. The issue stems from the "what-if" scenario whereby the system is in the Keep-Alive state and there is a short to ground on the Tracking Output regulator. If this output is powering remote sensors or systems, this becomes a more highly possible fault condition. Since a short on the Tracking output does not affect the Main output, which, if used to provide the

Keep-Alive input signal, would remain ON and draw 120 mA, the short circuit current of the Tracking regulator, from the input supply.

Using the Tracking output supply for Keep-Alive would prevent Keep-Alive operation during a short to ground fault and the excessive drain on the input supply. The inversion by the transistor will require a low level from the controller to allow Keep-Alive operation. The 24 k $\Omega$  pull-up resistor provides current limiting in the event of a Tracking output short to the unregulated/battery input supply.

### Application Information (Continued)



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#### Input Stability

Low dropout voltage regulators which utilize a PNP power transistor usually exhibit a large increase in current when in dropout (V<sub>IN</sub> < 5.5V). This increase is caused by the saturation characteristics ( $\beta$  reduction) of the PNP transistor. To significantly minimize this increase in current the LM9072 detects when the PNP enters saturation and reduces the operating current.

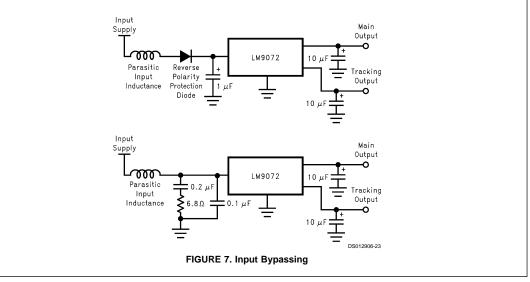
This reduction in input current can create a stability problem in applications with higher load current (> 200 mA) where the input voltage is applied through a long length of wire which in effect add a significant amount of inductance in series with the input. The drop in input current may create a positive input voltage transient which may take the PNP out of saturation. If the input voltage is held constant at the threshold where the PNP is going in and out of saturation, an oscillation may be created.

This is only observed where a large series inductance is present in the input supply line and when the rise and fall

time of the input supply is very slow. If the application and removal of the input voltage changes at a rate greater than 500 mV/ms it will move through the dropout region of the regulator (V<sub>IN</sub> of 3V to 5.5V) too quickly for an oscillation to be established.

In the event that an oscillation is present, input bypassing can also help de-tune the resonance. *Figure 7* illustrates two input bypassing approaches. The straightforward addition of a larger valued electrolytic capacitor could suffice. in this case however, if reverse battery connections are a possibility it is necessary to add a series protection diode as shown to prevent damaging the polarized input capacitor.

An alternative input bypassing scheme is also shown. This eliminates the use of polarized input capacitors and a series protection diode. The values shown were derived empirically in a representative typical application. Appropriate values for any given application require experimentation.



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# Application Information (Continued)

#### Thermal Management

The LM9072 is packaged in both a TO-263 surface mount power package and a narrow lead-pitch TO-220 package. To obtain operation over the highest possible load current and input voltage ranges, care must be taken to control the operating temperature of the device. Thermal shutdown protection is built in with a threshold above 150°C. Conventional heat sinking techniques can be used with the TO-220 package. When applying the TO-263 package on-board heat sinking is important to prevent shut down. More copper foil area under the tab of the device will directly reduce the operating junction temperature. Use of a double sided board with vias between two planes of copper as shown in *Figure 8* will improve performance and can optimize the PC board surface area required.

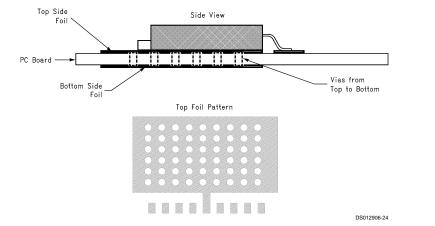


FIGURE 8. Typical TO-263 PC Board Heatsinking

#### Electro-Magnetic Compatibility (EMC)

The LM9072 offers good immunity to high frequency interference in a standard Bulk Current Injection (BCI) test (ISO11452 Part 4 test method). The following test conditions and configuration (*Figure 9*) can be used to observe this performance.

Frequency Range	1 MHz to 400 MHz
Modulation 1	CW (no modulation)
Modulation 2	1 kHz sine wave, 80% AM
Dwell Time	1 second
Frequency Steps	1 MHz (from 1 MHz to 10 MHz)
	2 MHz (from 10 MHz to 200 MHz)

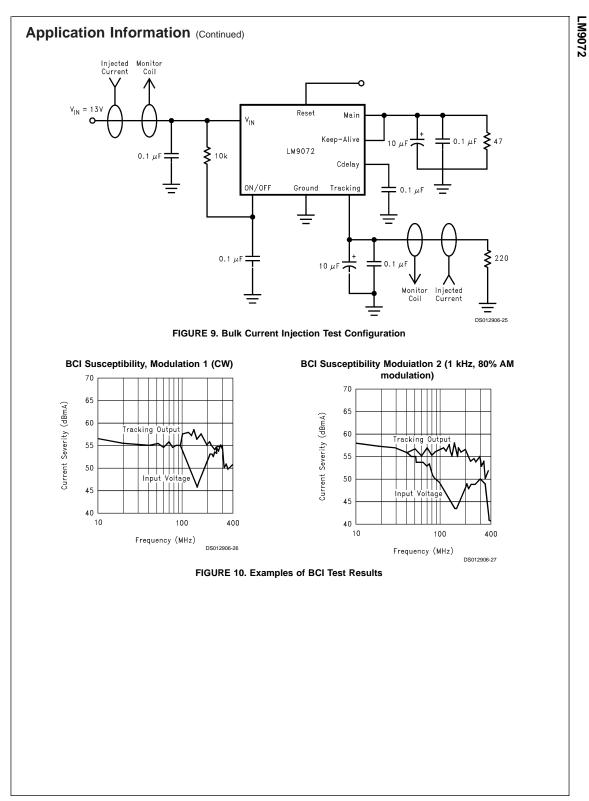
20 MHz (from 200 MHz to 400 MHz)

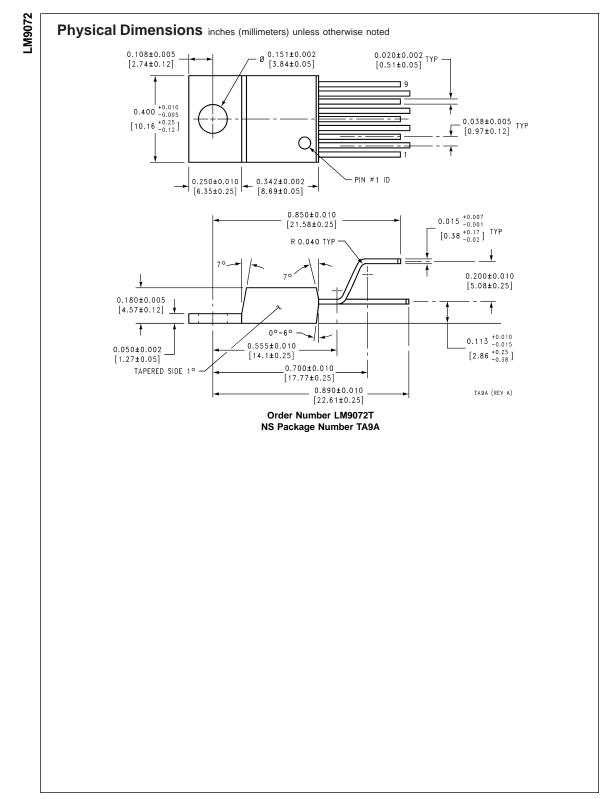
Test Method

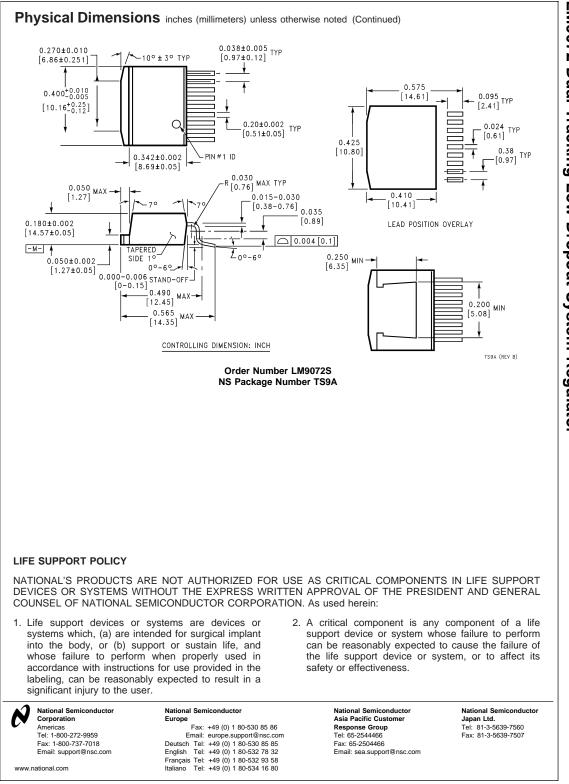
Closed loop current probe

In this test configuration the current injected into either the input pin or the tracking output pin is increased until a reset output is generated. These two pins are the most critical as they typically will connect to a module through long lengths of wire most likely to pick up high frequency energy. *Figure 10* illustrates examples of test results on the LM9072 with both types of modulation.

These results are just examples as actual results in any given application will depend on numerous external factors such as component selection, pc board layout, etc. The current power of the injected signal is expressed in dB relative to 1 mA (i.e., 40 dBmA = 100 mA).







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