features

- Dual-Input, Single-Output MOSFET Switch With No Reverse Current Flow (No Parasitic Diodes)
- IN1 . . . 250-mΩ, 500-mA N-Channel; 16-μA Max Supply Current
- IN2 . . . 1.3-Ω, 10-mA P-Channel; 1.5-μA Max Supply Current (V_{AUX} Mode)
- Advanced Switch Control Logic
- CMOS- and TTL-Compatible Enable Input
- Controlled Rise, Fall, and Transition Times
- 2.7-V to 4 V Operating Range
- SOT-23-5 and SOIC-8 Package
- –40°C to 70°C Ambient Temperature Range
- 2-kV Human-Body-Model, 750-V CDM, 200-V Machine-Model Electrostatic-Discharge Protection

typical applications

- Notebook and Desktop PCs
- Palmtops and PDAs

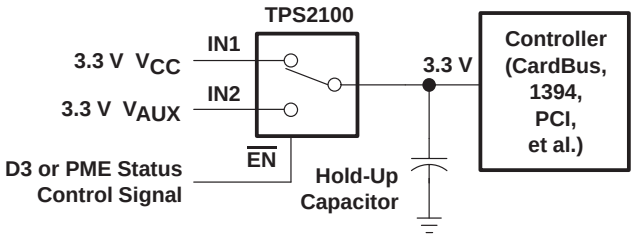


Figure 1. Typical Dual-Input Single-Output Application

description

The TPS2100 and TPS2101 are dual-input, single-output power switches designed to provide uninterrupted output voltage when transitioning between two independent power supplies. Both devices combine one n-channel (250 mΩ) and one p-channel (1.3 Ω) MOSFET with a single output. The p-channel MOSFET (IN2) is used with auxiliary power supplies that deliver lower current for standby modes. The n-channel MOSFET (IN1) is used with a main power supply that delivers higher current required for normal operation. Low on-resistance makes the n-channel the ideal path for higher main supply current when power-supply regulation and system voltage drops are critical. When using the p-channel MOSFET, quiescent current is reduced to 0.75 μA to decrease the demand on the standby power supply. The MOSFETs in the TPS2100 and TPS2101 do not have the parasitic diodes, found in discrete MOSFETs, which allow the devices to prevent back-flow current when the switch is off.

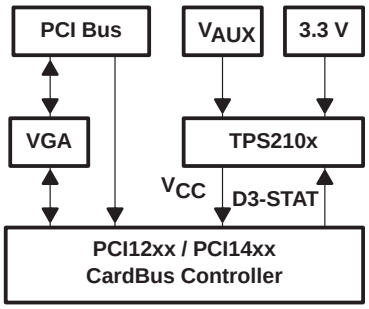
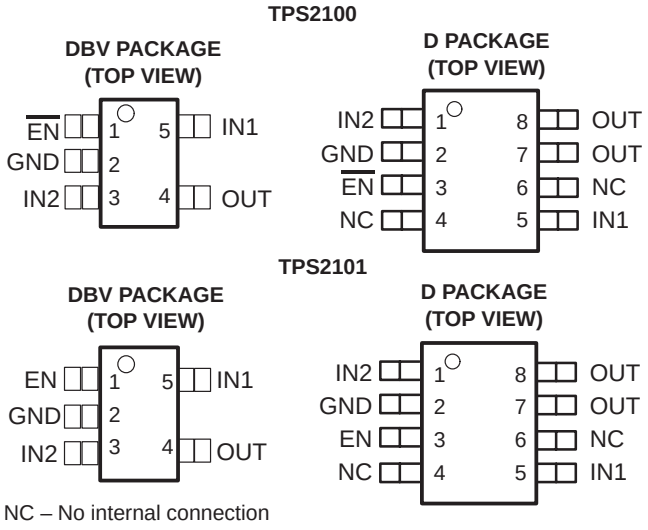


Figure 2. V_{AUX} CardBus Implementation



NC – No internal connection

Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

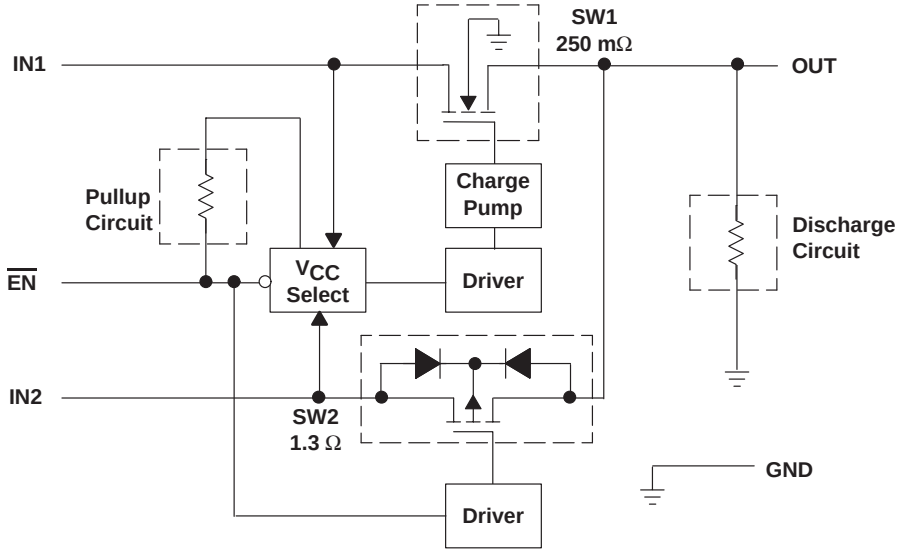
AVAILABLE OPTIONS

T_J	DEVICE	ENABLE	PACKAGED DEVICES	
			SOT-23-5 (DBV)†	SOIC-8 (D)
–40°C to 85°C	TPS2100	$\overline{EN}$	TSP2100DBV†	TPS2100D
	TPS2101	EN	TPS2101DBV†	TPS2101D

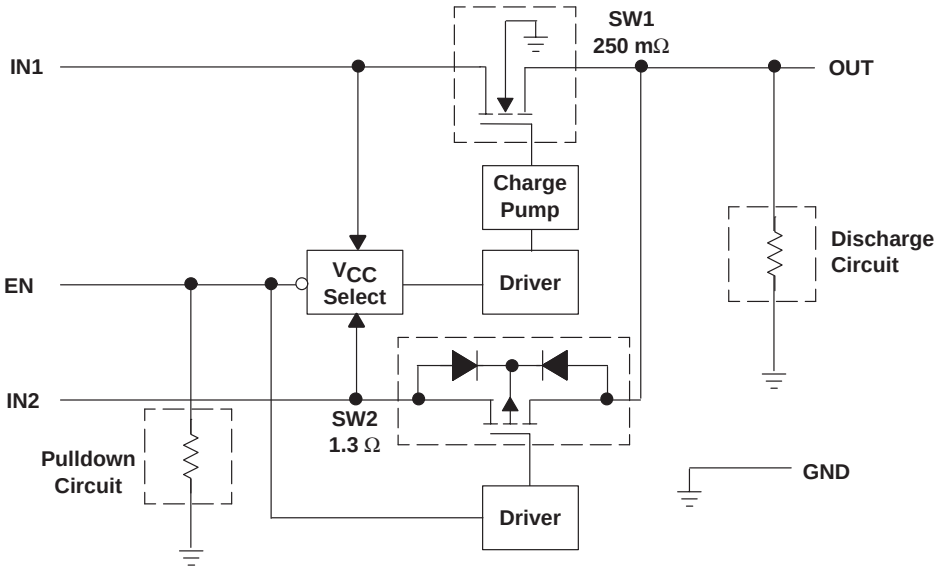
Both packages are available left-end taped and reeled. Add an R suffix to the D device type (e.g., TPS2101DR).

† Add T (e.g., TPS2100DBVT) to indicate tape and reel at order quantity of 250 parts.
Add R (e.g., TPS2100DBVR) to indicate tape and reel at order quantity of 3000 parts.

TPS2100 functional block diagram



TPS2101 functional block diagram



Function Tables

TPS2100			
VIN1	VIN2	$\overline{EN}$	OUT
0 V	0 V	XX	GND
0 V	3.3 V	L	GND
3.3 V	3.3 V	L	VIN1
3.3 V	0 V	L	VIN1
0 V	3.3 V	H	VIN2
3.3 V	0 V	H	VIN2
3.3 V	3.3 V	H	VIN2

XX = don't care

TPS2101			
VIN1	VIN2	EN	OUT
0 V	0 V	XX	GND
0 V	3.3 V	H	GND
3.3 V	3.3 V	H	VIN1
3.3 V	0 V	H	VIN1
0 V	3.3 V	L	VIN2
3.3 V	0 V	L	VIN2
3.3 V	3.3 V	L	VIN2

Terminal Functions

TERMINAL					I/O	DESCRIPTION
NAME	NO.					
	TPS2100		TPS2101			
	DBV	D	DBV	D		
EN			1	3		Active-high enable for IN1-OUT switch
$\overline{EN}$	1	3			I	Active-low enable for IN1-OUT switch
GND	2	2	2	2	I	Ground
IN1	5	5	5	5	I	Main Input voltage, NMOS drain (250 mΩ)
IN2	3	1	3	1	I	Auxilliary input voltage, PMOS drain (1.3 Ω)
OUT	4	7, 8	4	7, 8	O	Power switch output
NC		4, 6		4, 6		No connection

detailed description

power switches

n-channel MOSFET

The IN1-OUT n-channel MOSFET power switch has a typical on-resistance of 250 mΩ at 3.3-V input voltage, and is configured as a high-side switch.

p-channel MOSFET

The IN2-OUT p-channel MOSFET power switch with typical on-resistance of 1.3 Ω at 3.3-V input voltage and is configured as a high-side switch. When operating, the p-channel MOSFET quiescent current is reduced to less than 1.5 μA.

charge pump

An internal charge pump supplies power to the driver circuit and provides the necessary voltage to pull the gate of the MOSFET above the source. The charge pump operates from input voltages as low as 2.7 V and requires very little supply current.

driver

The driver controls the gate voltage of the IN1-OUT and IN2-OUT power switches. To limit large current surges and reduce the associated electromagnetic interference (EMI) produced, the drivers incorporate circuitry that controls the rise times and fall times of the output voltage.

detailed description (continued)

enable

The logic enable will turn on the IN2-OUT power switch when a logic high is present on $\overline{EN}$ (TPS2100) or logic low is present on EN (TPS2101). A logic low input on $\overline{EN}$ (TPS2100) or logic high on EN (TPS2101) restores bias to the drive and control circuits and turns on the IN1-OUT power switch. The enable input is compatible with both TTL and CMOS logic levels.

the V_{AUX} application for CardBus controllers

The PC Card specification requires the support of V_{AUX} to the CardBus controller as well as to the PC Card sockets. Both are 3.3-V requirements; however the CardBus controller's current demand from the V_{AUX} supply is limited to 10 μA, whereas the PC Card may consume as much as 200 mA. In either implementation, if support of a wake-up event is required, the controller and the socket will transition from the 3.3-V V_{CC} rail to the 3.3-V V_{AUX} rail when the equipment moves into a low power mode such as D3. The transition from V_{CC} to V_{AUX} needs to be seamless in order to maintain all memory and register information in the system. If V_{AUX} is not supported, the system will lose all register information when it transitions to the D3 state.

absolute maximum ratings over operating free-air temperature (unless otherwise noted)†

Input voltage range, $V_{I(IN1)}$ (see Note1) . . . . . –0.3 V to 5 V
Input voltage range, $V_{I(IN2)}$ (see Note1) . . . . . –0.3 V to 5 V
Input voltage range, V_I at $\overline{EN}$ or EN . . . . . –0.3 V to 5 V
Output voltage range, V_O (see Note 1) . . . . . –0.3 V to 5 V
Continuous output current, $I_{O(IN1)}$ . . . . . 700 mA
Continuous output current, $I_{O(IN2)}$ . . . . . 70 mA
Continuous total power dissipation . . . . . See dissipation rating table
Operating virtual junction temperature range, T_J . . . . . –40°C to 85°C
Storage temperature range, T_{stg} . . . . . –65°C to 150°C
Lead temperature soldering 1,6 mm (1/16 inch) from case for 10 seconds . . . . . 260°C
Electrostatic discharge (ESD) protection: Human body model . . . . . 2 kV
Machine model . . . . . 200 V
Charged device model (CDM) . . . . . 750 V

† Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

NOTE 1: All voltages are with respect to GND.

DISSIPATION RATING TABLE

PACKAGE	T_A < 25°C POWER RATING	DERATING FACTOR ABOVE T_A = 25°C	T_A = 70°C POWER RATING	T_A = 85°C POWER RATING
DBV	309 mW	3.1 mW/°C	170 mW	123 mW
D	568 mW	5.7 mW/°C	313 mW	227 mW

recommended operating conditions

	MIN	MAX	UNIT
Input voltage, $V_{I(INx)}$	2.7	4	V
Input voltage, V_I at $\overline{EN}$ and EN	0	4	V
Continuous output current, $I_{O(IN1)}$		500	mA
Continuous output current, $I_{O(IN2)}$		100‡	mA
Operating virtual junction temperature, T_J	–40	85	°C

‡ The device can deliver up to 220 mA at $I_{O(IN2)}$. However, operation at the higher current levels will result in greater voltage drop across the device, and greater voltage droop when switching between IN1 and IN2.

electrical characteristics over recommended operating junction temperature range, $V_{I(IN1)} = V_{(IN2)} = 3.3$ V, I_O = rated current (unless otherwise noted)

power switch

PARAMETER			TEST CONDITIONS†	MIN	TYP	MAX	UNIT
$r_{DS(on)}$	On-state resistance	IN1-OUT	T_J = 25°C		250		mΩ
			T_J = 85°C		300	375	
		IN2-OUT	T_J = 25°C		1.3		Ω
			T_J = 85°C		1.5	2.1	

† Pulse-testing techniques maintain junction temperature close to ambient termperature; thermal effects must be taken into account separately.

enable input ($\overline{EN}$ and EN)

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
V_{IH}	High-level input voltage	2.7 V ≤ $V_{I(INx)}$ ≤ 4 V		2			V
V_{IL}	Low-level input voltage	2.7 V ≤ $V_{I(INx)}$ ≤ 4 V				0.8	V
I_I	Input current	TPS2100	$\overline{EN}$ = 0 V or $\overline{EN}$ = $V_{I(INx)}$	–0.5		0.5	μA
		TPS2101	EN = 0 V or EN = $V_{I(INx)}$	–0.5		0.5	μA

supply current

PARAMETER		TEST CONDITIONS			MIN	TYP	MAX	UNIT
I_I	Supply current	TPS2100	$\overline{EN}$ = H, IN2 selected	T_J = 25°C		0.75		μA
				–40°C ≤ T_J ≤ 85°C			1.5	
			$\overline{EN}$ = L, IN1 selected	T_J = 25°C		10		μA
				–40°C ≤ T_J ≤ 85°C			16	
		TPS2101	EN = L, IN2 selected	T_J = 25°C		0.75		μA
				–40°C ≤ T_J ≤ 85°C			1.5	
			EN = H, IN1 selected	T_J = 25°C		10		μA
				–40°C ≤ T_J ≤ 85°C			16	

switching characteristics, T_J = 25°C, $V_{I(IN1)}$ = $V_{I(IN2)}$ = 3.3 V (unless otherwise noted)†

PARAMETER			TEST CONDITIONS†		MIN	TYP	MAX	UNIT
t_r	Output rise time	IN1-OUT	$V_{I(IN2)}$ = 0	C_L = 1 µF, I_L = 500 mA		830		µs
				C_L = 10 µF, I_L = 500 mA		840		
				C_L = 1 µF, I_L = 10 mA		640		
		IN2-OUT	$V_{I(IN1)}$ = 0	C_L = 1 µF, I_L = 10 mA		5.5		
				C_L = 10 µF, I_L = 10 mA		70		
				C_L = 1 µF, I_L = 1 mA		5.5		
t_f	Output fall time	IN1-OUT	$V_{I(IN2)}$ = 0	C_L = 1 µF, I_L = 500 mA		8		µs
				C_L = 10 µF, I_L = 500 mA		93		
				C_L = 1 µF, I_L = 10 mA		23		
		IN2-OUT	$V_{I(IN1)}$ = 0	C_L = 1 µF, I_L = 10 mA		690		
				C_L = 10 µF, I_L = 10 mA		6900		
				C_L = 1 µF, I_L = 1 mA		6900		
t_{PLH}	Propagation delay time, low-to-high output	IN1-OUT	$V_{I(IN2)}$ = 0	C_L = 10 µF, I_L = 10 mA		75		µs
		IN2-OUT	$V_{I(IN1)}$ = 0			2		
t_{PHL}	Propagation delay time, high-to-low output	IN1-OUT	$V_{I(IN2)}$ = 0	C_L = 10 µF, I_L = 10 mA		3		µs
		IN2-OUT	$V_{I(IN1)}$ = 0			370		

† All timing parameters refer to Figure 3.

PARAMETER MEASUREMENT INFORMATION

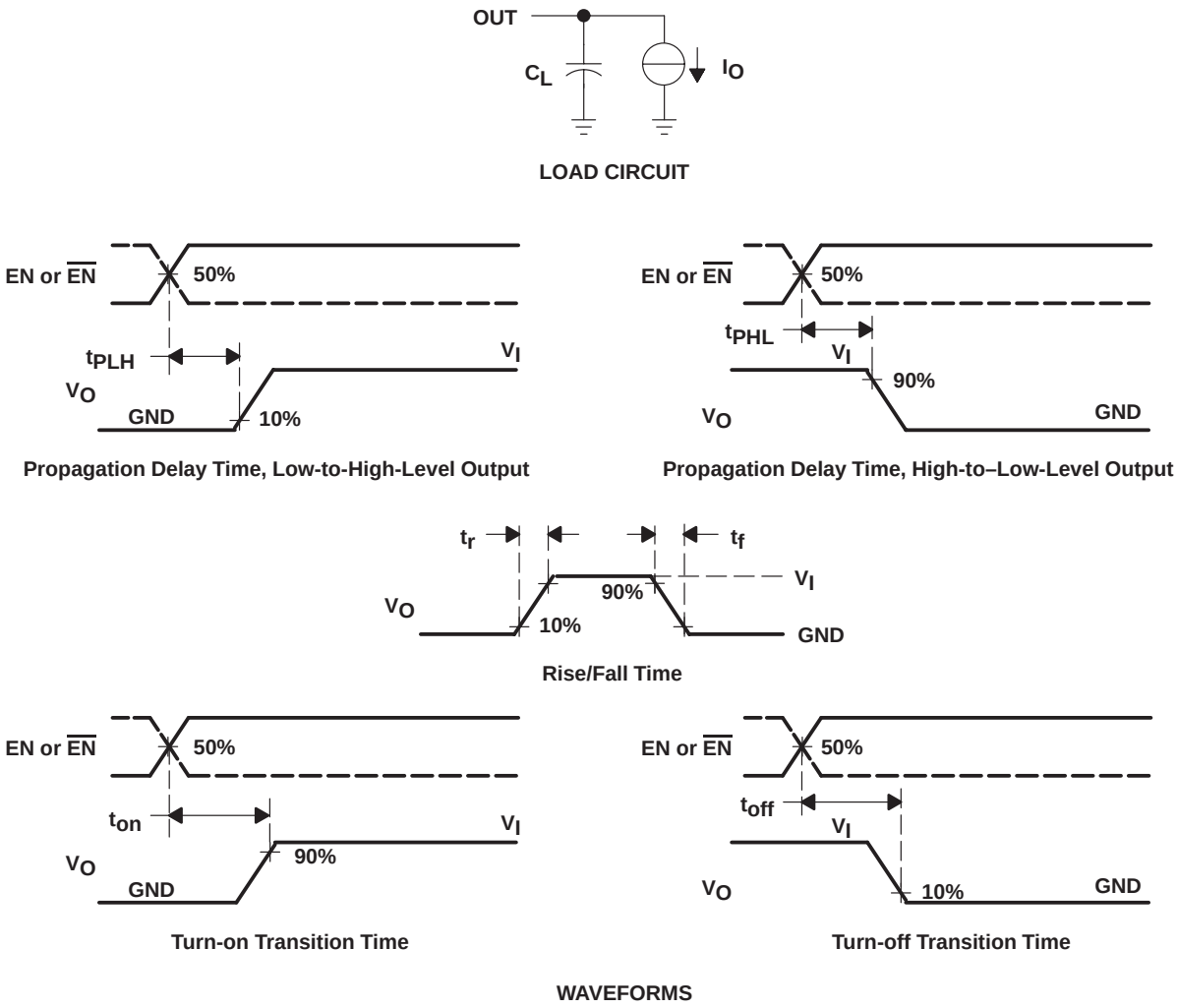


Figure 3. Test Circuit and Voltage Waveforms

Table of Timing Diagrams†

	FIGURE
Propagation Delay and Rise Time With 0.1-μF Load, IN1	4
Propagation Delay and Rise Time With 0.1-μF Load, IN2	5
Propagation Delay and Fall Time With 0.1-μF Load, IN1	6
Propagation Delay and Fall Time With 0.1-μF Load, IN2	7
Propagation Delay and Rise Time With 1-μF Load, IN1	8
Propagation Delay and Rise Time With 1-μF Load, IN2	9
Propagation Delay and Fall Time With 1-μF Load, IN1	10
Propagation Delay and Fall Time With 1-μF Load, IN2	11

† Waveforms shown in Figures 4–11 refer to TPS2100 at T_J = 25°C

PARAMETER MEASUREMENT INFORMATION

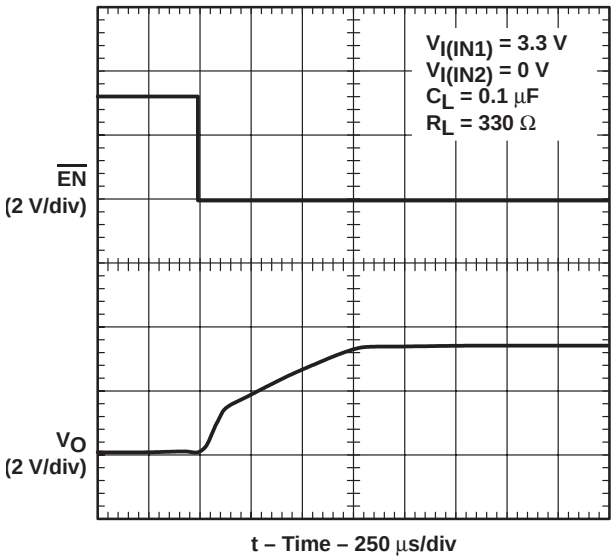


Figure 4. Propagation Delay and Rise Time With 0.1-μF Load, IN1

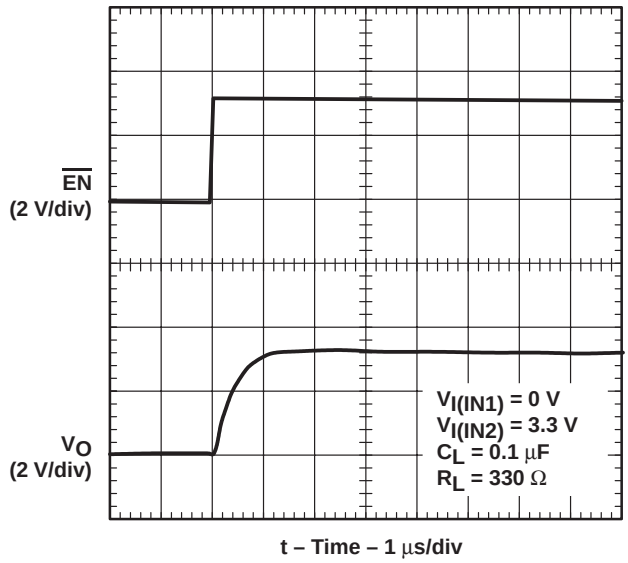


Figure 5. Propagation Delay and Fall Time With 0.1-μF Load, IN2

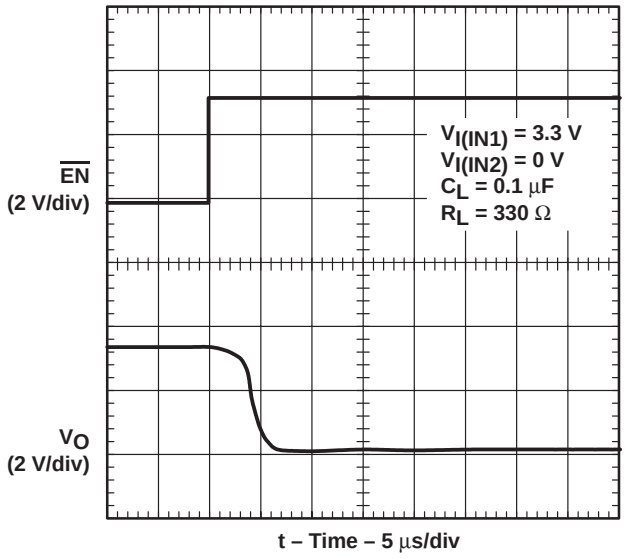


Figure 6. Propagation Delay and Fall Time With 0.1-μF Load, IN1

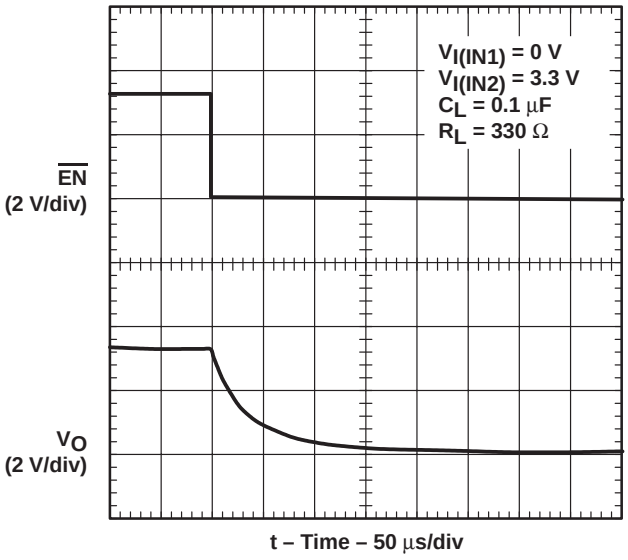


Figure 7. Propagation Delay and Fall Time With 0.1-μF Load, IN2

PARAMETER MEASUREMENT INFORMATION

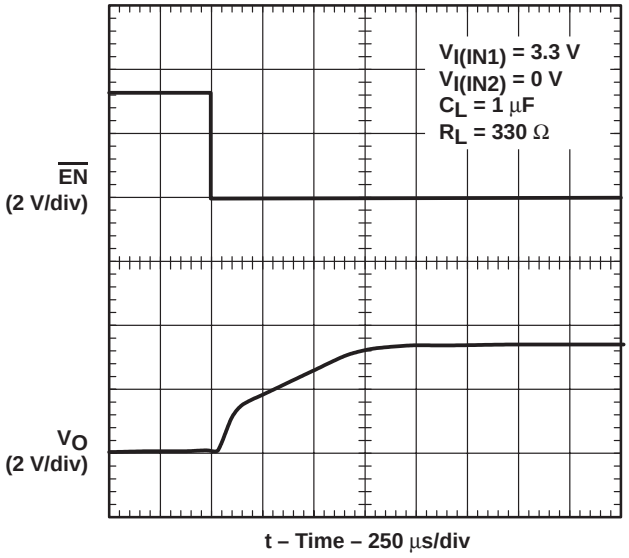


Figure 8. Propagation Delay and Rise Time With 1-μF Load, IN1

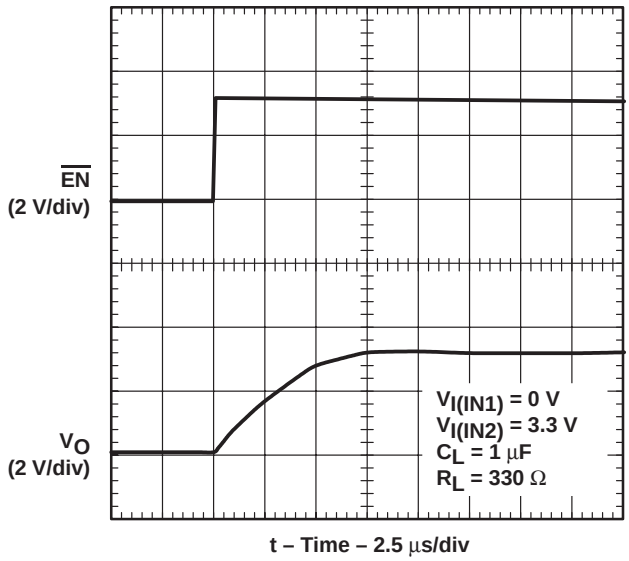


Figure 9. Propagation Delay and Rise Time With 1-μF Load, IN2

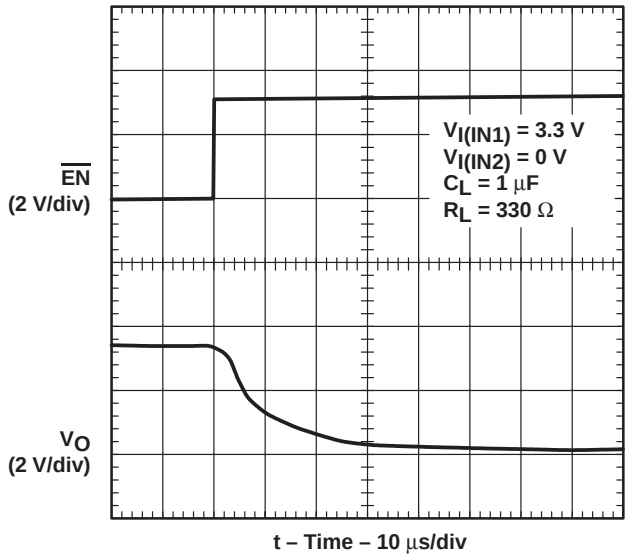


Figure 10. Propagation Delay and Fall Time With 1-μF Load, IN1

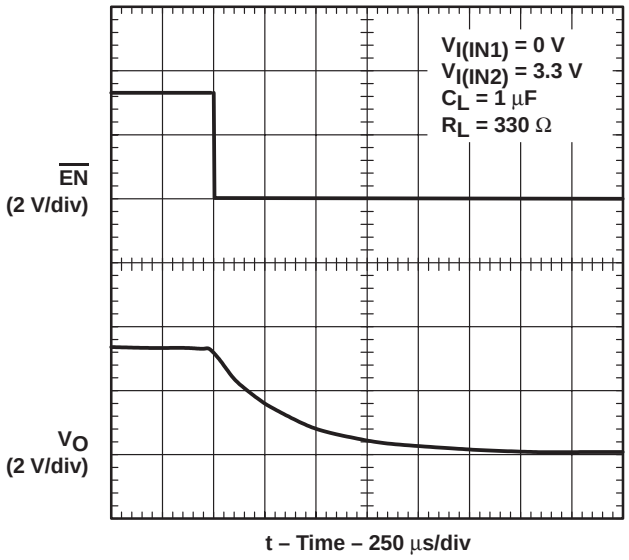


Figure 11. Propagation Delay and Fall Time With 1-μF Load, IN2

TYPICAL CHARACTERISTICS

Table of Graphs

		FIGURE
IN1 Switch Rise Time	vs Output Current	12
IN2 Switch Fall Time	vs Output Current	13
IN1 Switch Fall Time	vs Output Current	14
IN2 Switch Fall Time	vs Output Current	15
Output Voltage Droop	vs Output Current When Output Is Switched From IN2 to IN1	16
Inrush Current	vs Output Capacitance	17
IN1 Supply Current	vs Junction Temperature (IN1 Enabled)	18
IN1 Supply Current	vs Junction Temperature (IN1 Disabled)	19
IN2 Supply Current	vs Junction Temperature (IN2 Enabled)	20
IN2 Supply Current	vs Junction Temperature (IN2 Disabled)	21
IN1-OUT On-State Resistance	vs Junction Temperature	22
IN2-OUT On-State Resistance	vs Junction Temperature	23

IN1 SWTICH RISE TIME vs OUTPUT CURRENT

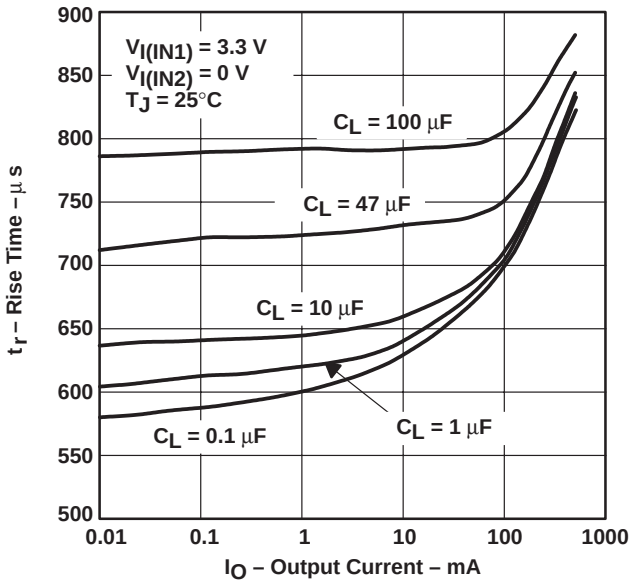


Figure 12

IN2 SWTICH RISE TIME vs OUTPUT CURRENT

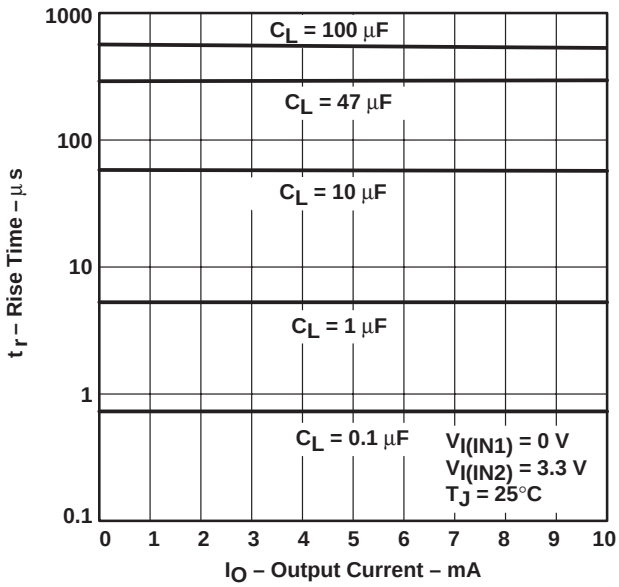


Figure 13

TYPICAL CHARACTERISTICS

IN1 SWITCH FALL TIME vs OUTPUT CURRENT

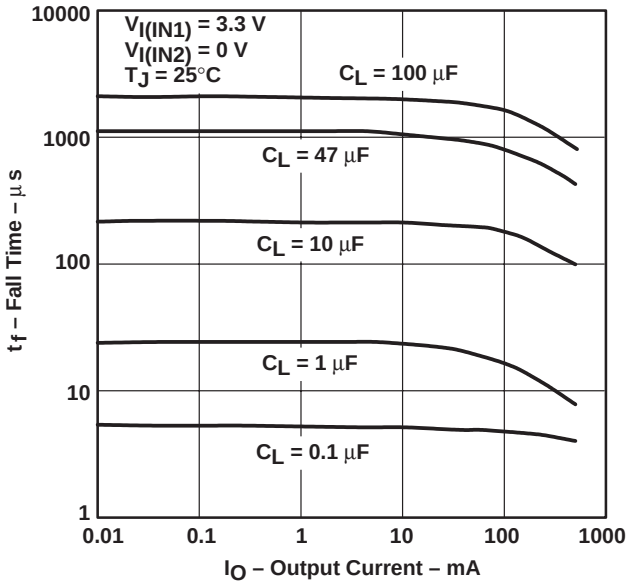


Figure 14

IN2 SWITCH FALL TIME vs OUTPUT CURRENT

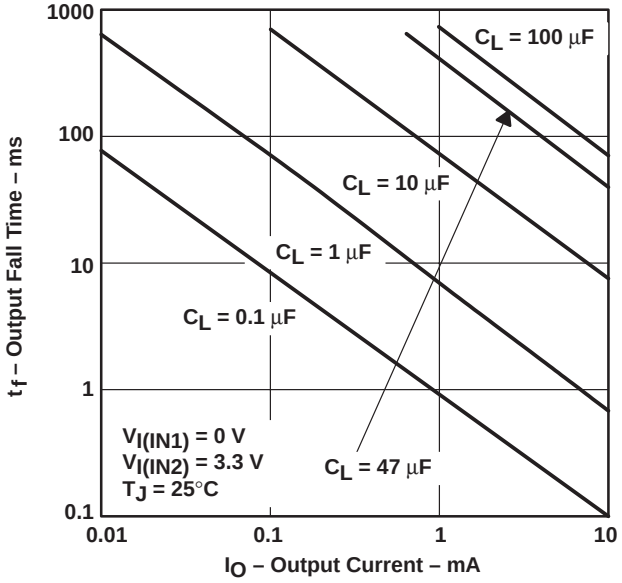


Figure 15

OUTPUT VOLTAGE DROOP vs OUTPUT CURRENT WHEN OUTPUT IS SWITCHED FROM IN2 TO IN1

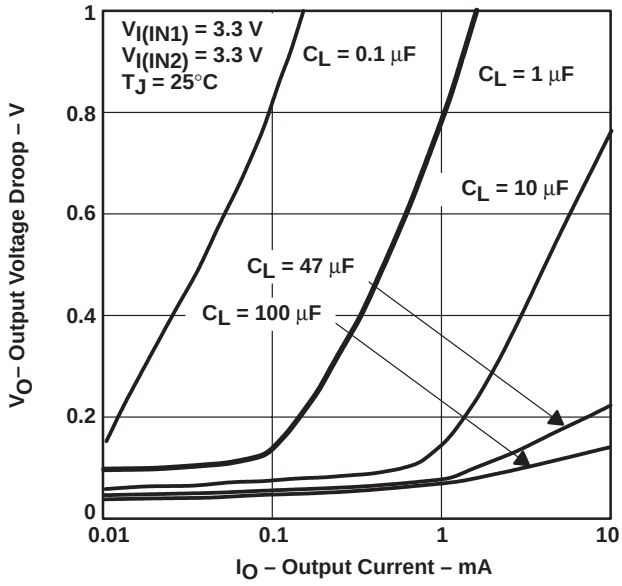


Figure 16

INRUSH CURRENT vs OUTPUT CAPACITANCE

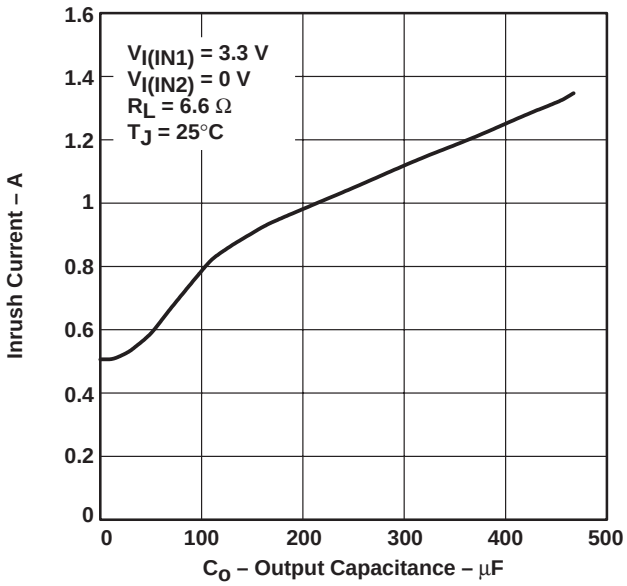


Figure 17

TYPICAL CHARACTERISTICS

IN1 SUPPLY CURRENT
vs
JUNCTION TEMPERATURE (IN1 ENABLED)

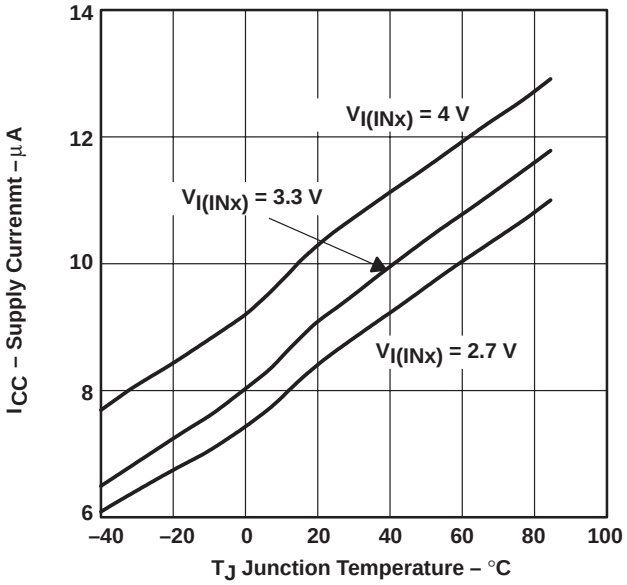


Figure 18

IN1 SUPPLY CURRENT
vs
JUNCTION TEMPERATURE (IN1 DISABLED)

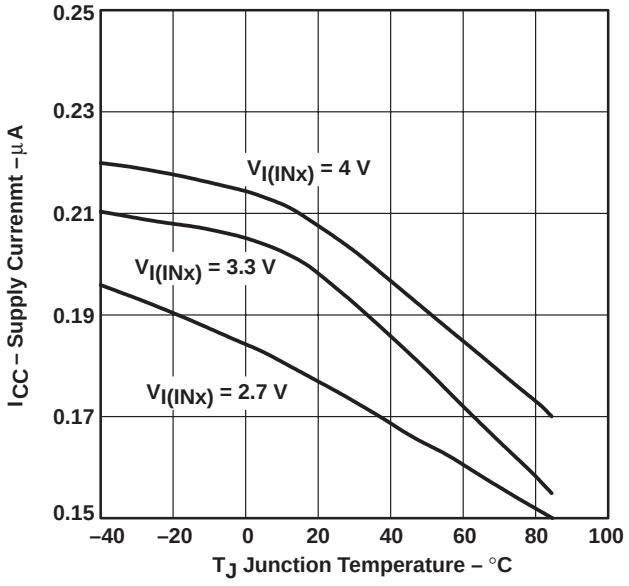


Figure 19

IN2 SUPPLY CURRENT
vs
JUNCTION TEMPERATURE (IN2 ENABLED)

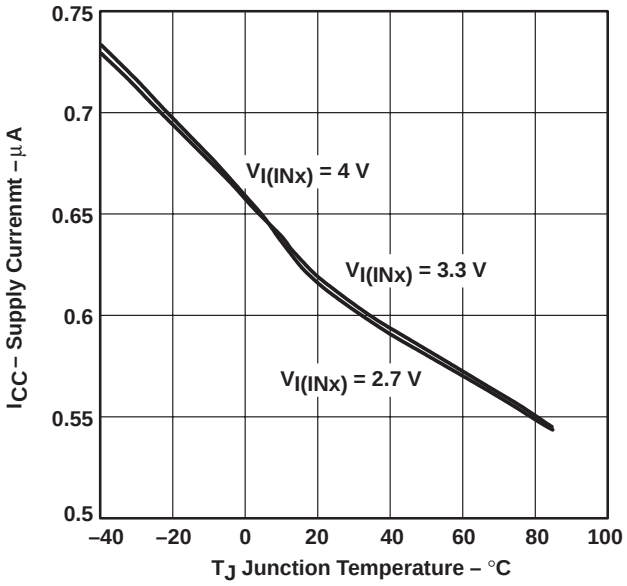


Figure 20

IN2 SUPPLY CURRENT
vs
JUNCTION TEMPERATURE (IN2 DISABLED)

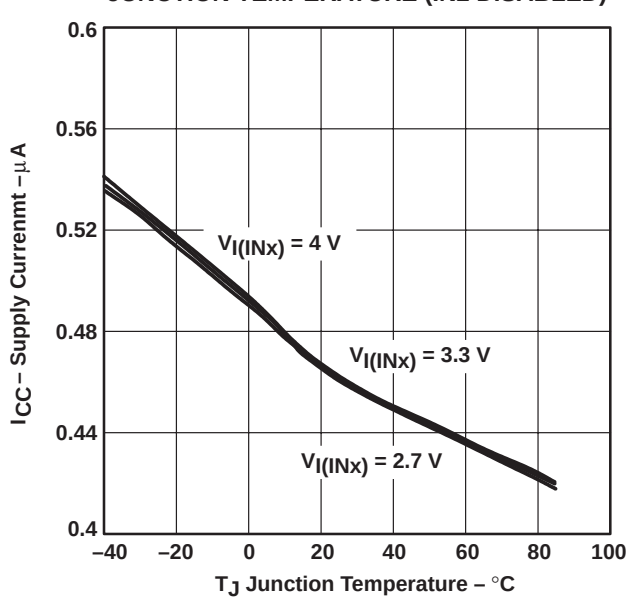


Figure 21

TYPICAL CHARACTERISTICS

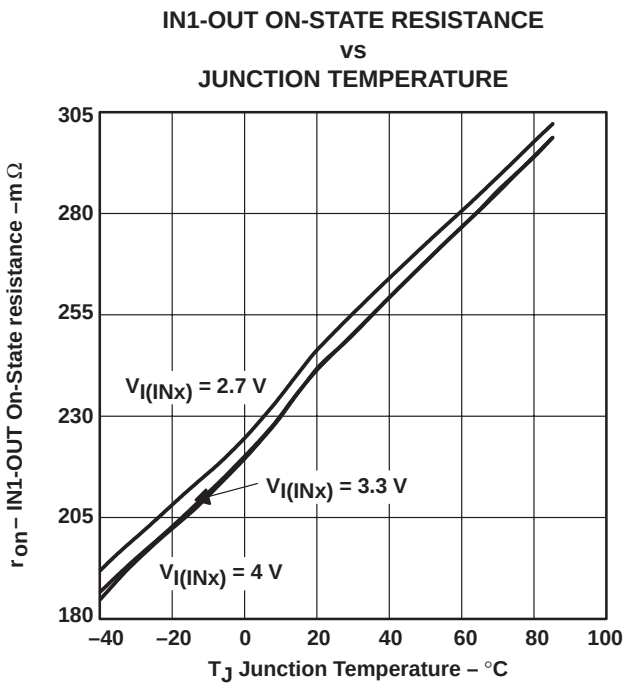


Figure 22

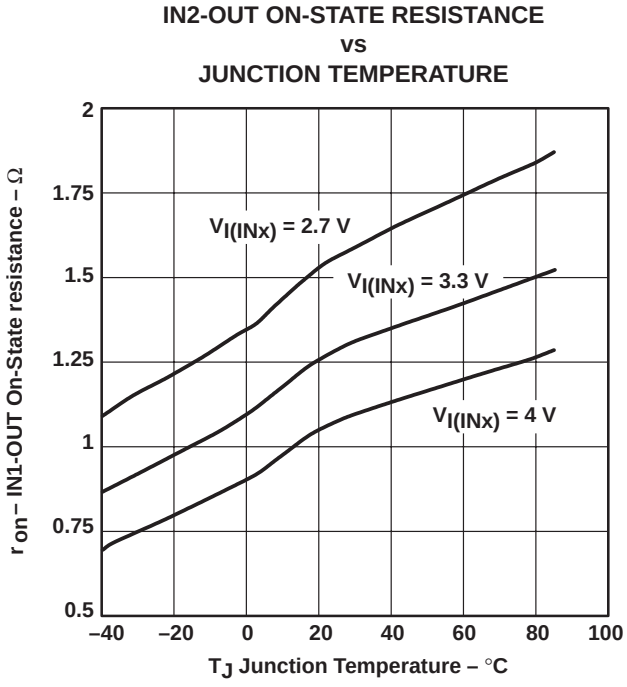


Figure 23

APPLICATION INFORMATION

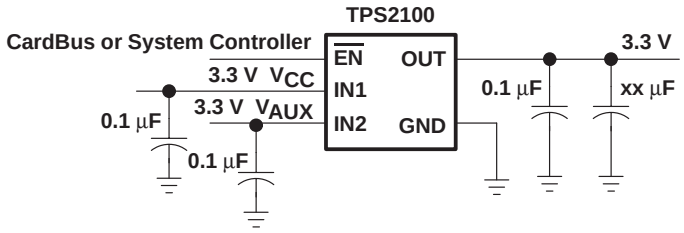


Figure 24. Typical Application

power supply considerations

A 0.01-µF to 0.1-µF ceramic bypass capacitor between IN and GND, close to the device is recommended. The output capacitor should be chosen based on the size of the load during the transition of the switch. A 47-µF capacitor is recommended for 10-mA loads. Typical output capacitors (xx µF, shown in Figure 24) required for a given load can be determined from Figure 16 which shows the output voltage droop when output is switched from IN2 to IN1. The output voltage droop is insignificant when output is switched from IN1 to IN2. Additionally, bypassing the output with a 0.01-µF to 0.1-µF ceramic capacitor improves the immunity of the device to short-circuit transients.

APPLICATION INFORMATION

power supply considerations (continued)

switch transition

The n-channel MOSFET on IN1 uses a charge-pump to create the gate-drive voltage, which gives the IN1 switch a rise time of approximately 1 ms. The p-channel MOSFET on IN2 has a simpler drive circuit that allows a rise time of approximately 8 μs. Because the device has two switches and a single enable pin, these rise times are seen as transition times, from IN1 to IN2, or IN2 to IN1, by the output. The controlled transition times help limit the surge currents seen by the power supply during switching.

thermal protection

Thermal protection provided on the IN1 switch prevents damage to the IC when heavy-overload or short-circuit faults are present for extended periods of time. The increased dissipation causes the junction temperature to rise to dangerously high levels. The protection circuit senses the junction temperature of the switch and shuts it off at approximately 125°C (T_J). The switch remains off until the junction temperature has dropped. The switch continues to cycle in this manner until the load fault or input power is removed.

undervoltage lockout

An undervoltage lockout function is provided to ensure that the power switch is in the off state at power-up. Whenever the input voltage falls below approximately 2 V, the power switch quickly turns off. This function facilitates the design of hot-insertion systems that may not have the capability to turn off the power switch before input power is removed. Upon reinsertion, the power switch will be turned on with a controlled rise time to reduce EMI and voltage overshoots.

power dissipation and junction temperature

The low on-resistance on the n-channel MOSFET allows small surface-mount packages, such as SOIC, to pass large currents. The thermal resistances of these packages are high compared to that of power packages; it is good design practice to check power dissipation and junction temperature. First, find r_{on} at the input voltage, and operating temperature. As an initial estimate, use the highest operating ambient temperature of interest and read r_{on} from Figure 22 or Figure 23. Next calculate the power dissipation using:

$$P_D = r_{on} \times I^2$$

Finally, calculate the junction temperature:

$$T_J = P_D \times R_{\theta JA} + T_A$$

Where:

T_A = Ambient temperature

$R_{\theta JA}$ = Thermal resistance

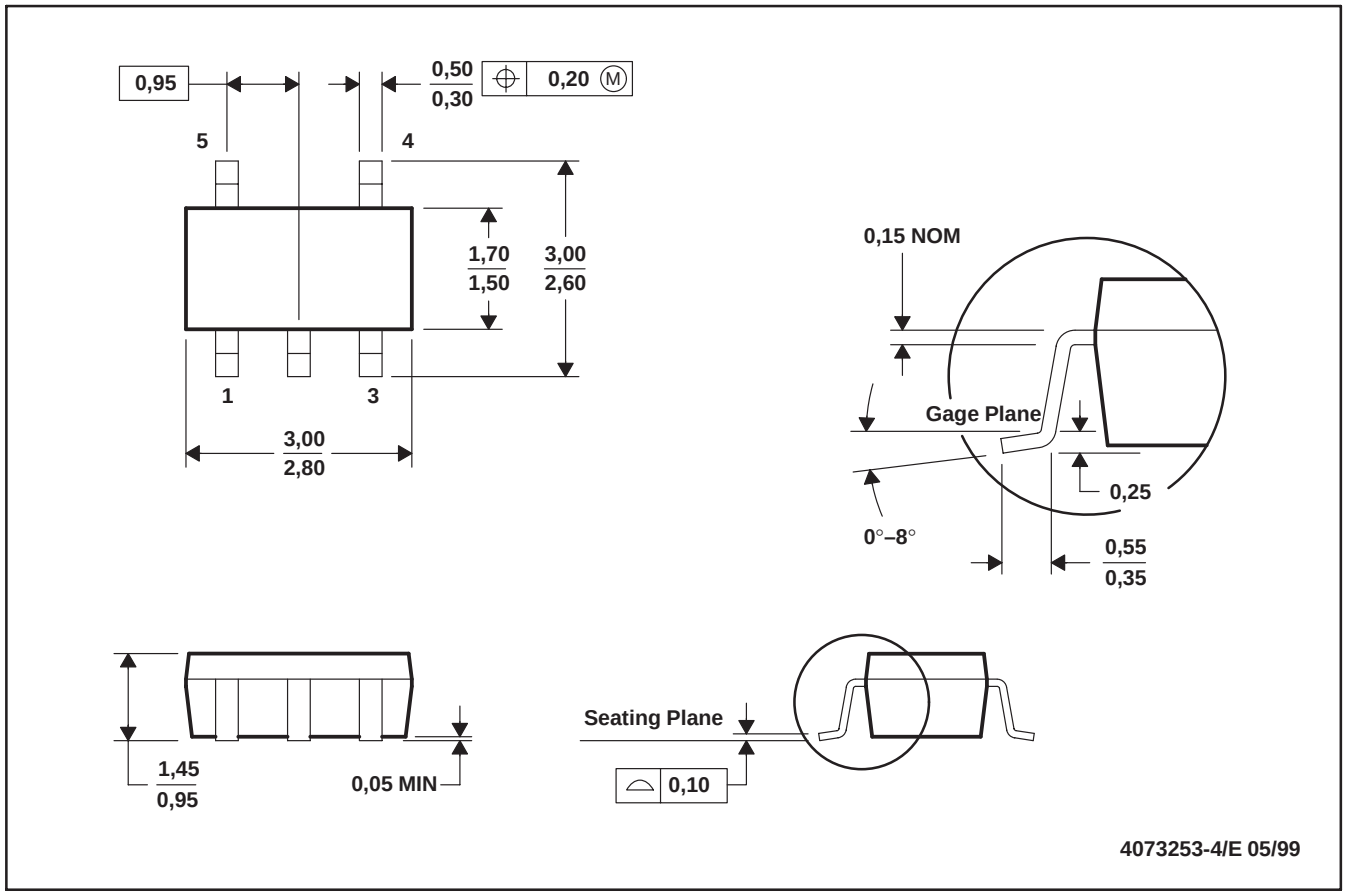
Compare the calculated junction temperature with the initial estimate. If they do not agree within a few degrees, repeat the calculation using the calculated value as the new estimate. Two or three iterations are generally sufficient to obtain a reasonable answer.

ESD protection

All TPS2100 and TPS2101 terminals incorporate ESD-protection circuitry designed to withstand a 2-kV human-body-model discharge as defined in MIL-STD-883C.

MECHANICAL DATA

DBV (R-PDSO-G5) **PLASTIC SMALL-OUTLINE**

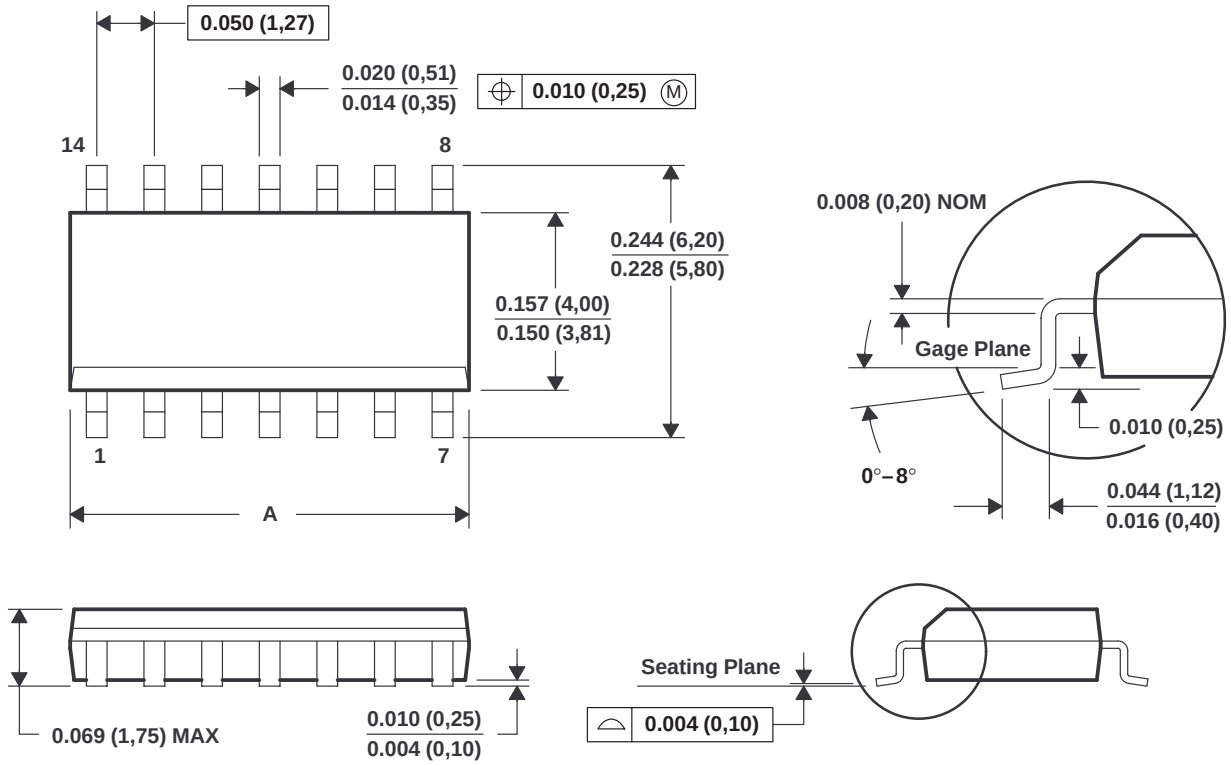


NOTES: A. All linear dimensions are in millimeters.
B. This drawing is subject to change without notice.
C. Body dimensions do not include mold flash or protrusion.
D. Falls within JEDEC MO-178

MECHANICAL DATA

D (R-PDSO-G)** **PLASTIC SMALL-OUTLINE PACKAGE**

14 PINS SHOWN



DIM \ PINS **	8	14	16
A MAX	0.197 (5,00)	0.344 (8,75)	0.394 (10,00)
A MIN	0.189 (4,80)	0.337 (8,55)	0.386 (9,80)

4040047/D 10/96

NOTES: A. All linear dimensions are in inches (millimeters).
B. This drawing is subject to change without notice.
C. Body dimensions do not include mold flash or protrusion, not to exceed 0.006 (0,15).
D. Falls within JEDEC MS-012

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