

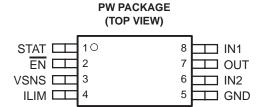
# **AUTOSWITCHING POWER MUX**

#### **FEATURES**

- Two-Input, One-Output Power Multiplexer With Low r<sub>DS(on)</sub> Switches:
  - 84 m $\Omega$  Typ (TPS2113A)
  - 120 mΩ Typ (TPS2112A)
- Reverse and Cross-Conduction Blocking
- Wide Operating Voltage Range . . . . 2.8 V to 5.5 V
- Low Standby Current . . . . 0.5-μA Typ
- Low Operating Current . . . . 55-μA Typ
- Adjustable Current Limit
- Controlled Output Voltage Transition Times, Limits Inrush Current and Minimizes Output Voltage Hold-Up Capacitance
- CMOS- and TTL-Compatible Control Inputs
- Auto-Switching Operating Mode
- Thermal Shutdown
- Available in a TSSOP-8 Package

#### **APPLICATIONS**

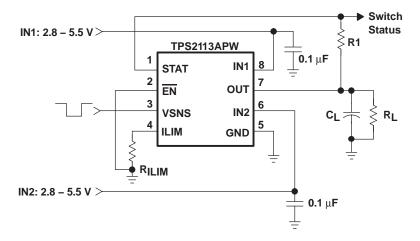
- PCs
- PDAs
- Digital Cameras
- Modems
- Cell Phones
- Digital Radios
- MP3 Players



#### **DESCRIPTION**

The TPS211xA family of power multiplexers enables seamless transition between two power supplies, such as a battery and a wall adapter, each operating at 2.8–5.5 V and delivering up to 1 A. The TPS211xA family includes extensive protection circuitry, including user-programmable current limiting, thermal protection, inrush current control, seamless supply transition, cross-conduction blocking, and reverse-conduction blocking. These features greatly simplify designing power multiplexer applications.

#### TYPICAL APPLICATION



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These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

#### **AVAILABLE OPTIONS**

FEATURE		TPS2110A	TPS2111A	TPS2112A	TPS2113A	TPS2114A	TPS2115A
Current Limit Adjustment Range		0.31-0.75A	0.63-1.25A	0.31-0.75A	0.63-1.25A	0.31-0.75A	0.63-1.25A
Cuitabia a Madaa	Manual	Yes	Yes	No	No	Yes	Yes
Switching Modes	Automatic	Yes	Yes	Yes	Yes	Yes	Yes
Switch Status Output		No	No	Yes	Yes	Yes	Yes
Package		TSSOP-8	TSSOP-8	TSSOP-8	TSSOP-8	TSSOP-8	TSSOP-8

#### **ORDERING INFORMATION**

TA	PACKAGE	ORDERING NUMBER(1)	MARKINGS
4000 1- 0500	-40°C to 85°C TSSOP-8 (PW)	TPS2112APW	2112A
-40°C to 85°C	TSSOP-8 (PW)	TPS2113APW	2113A

<sup>(1)</sup> The PW package is available taped and reeled. Add an R suffix to the device type (e.g., TPS2112APWR) to indicate tape and reel.

#### PACKAGE DISSIPATION RATINGS

PACKAGE	DERATING FACTOR ABOVE	T <sub>A</sub> ≤ 25°C	T <sub>A</sub> = 70°C	T <sub>A</sub> = 85°C
	T <sub>A</sub> = 25°C	POWER RATING	POWER RATING	POWER RATING
TSSOP-8 (PW)	3.9 mW/°C	387 mW	213 mW	155 mW

#### **ABSOLUTE MAXIMUM RATINGS**

over operating free-air temperature range unless otherwise noted(1)

		TPS2112A, TPS2113A
Input voltage range at pins IN1, IN2,	EN, VSNS, ILIM <sup>(2)</sup>	−0.3 V to 6 V
Output voltage range, VO(OUT), VO	(STAT) <sup>(2)</sup>	−0.3 V to 6 V
Output sink current, IO(STAT)		5 mA
Continuous suitaut surrent 1-	TPS2112A	0.9 A
Continuous output current, IO	TPS2113A	1.5 A
Continuous total power dissipation		See Dissipation Rating Table
Operating virtual junction temperature	e range, TJ	-40°C to 125°C
Storage temperature range, T <sub>Stg</sub>		−65°C to 150°C
Lead temperature soldering 1,6 mm	(1/16 inch) from case for 10 seconds	260°C

<sup>(1)</sup> Stresses beyond those listed under *absolute maximum ratings* may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under *recommended operating conditions* is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

### **RECOMMENDED OPERATING CONDITIONS**

		M	N MAX	UNIT
	V <sub>I(IN2)</sub> ≥ 2.8 V	1	.5 5.5	
Input voltage at IN1, V <sub>I(IN1)</sub>	V <sub>I(IN2)</sub> < 2.8 V	2	.8 5.5	V
haratankan at MO V	V <sub>I(IN1)</sub> ≥ 2.8 V	1	1.5     5.5       2.8     5.5       1.5     5.5       2.8     5.5       0     5.5       0.31     0.75       0.63     1.25	
Input voltage at IN2, V <sub>I(IN2)</sub>	V <sub>I(IN1)</sub> < 2.8 V	2	.8 5.5	\ \
Input voltage, V <sub>I</sub> (EN), V <sub>I</sub> (VSNS)			0 5.5	V
	TPS2112A	0.0	31 0.75	
Current limit adjustment range, IO(OUT)	TPS2113A	0.6	3 1.25	A
Operating virtual junction temperature, TJ	•	-4	10 125	°C

# **ELECTROSTATIC DISCHARGE (ESD) PROTECTION**

	MIN	MAX	UNIT
Human body model		2	kV
CDM		500	V

<sup>(2)</sup> All voltages are with respect to GND.



# **ELECTRICAL CHARACTERISTICS**

over recommended operating junction temperature range,  $V_{I(IN1)} = V_{I(IN2)} = 5.5 \text{ V}$ ,  $R_{ILIM} = 400 \Omega$  (unless otherwise noted)

PARAMETER		TEOT	CONDITIONS	TPS2112A		TPS2113A				
		IESI	TEST CONDITIONS		TYP	MAX	MIN	TYP	MAX	UNIT
POWER SW	/ITCH									
		T 0500	$V_{I(IN1)} = V_{I(IN2)} = 5.0 \text{ V}$		120	140		84	110	
		$T_J = 25^{\circ}C,$ $I_I = 500 \text{ mA}$	$V_{I(IN1)} = V_{I(IN2)} = 3.3 \text{ V}$		120	140		84	110	$m\Omega$
rDS(on) <sup>(1)</sup>	Drain-source on-state resistance	1L = 300 HIA	$V_{I(IN1)} = V_{I(IN2)} = 2.8 \text{ V}$		120	140		84	110	
	(INx–OUT)	T 40500	$V_{I(IN1)} = V_{I(IN2)} = 5.0 \text{ V}$			220			150	
	( /	$T_J = 125^{\circ}C$ , $I_I = 500 \text{ mA}$	$V_{I(IN1)} = V_{I(IN2)} = 3.3 \text{ V}$			220			150	$m\Omega$
		IL = 000 IIIA	$V_{I(IN1)} = V_{I(IN2)} = 2.8 \text{ V}$		·	220			150	

<sup>(1)</sup> The TPS211xA can switch a voltage as low as 1.5 V as long as there is a minimum of 2.8 V at one of the input power pins. In this specific case, the lower supply voltage has no effect on the IN1 and IN2 switch on-resistances.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
LOGIC INPUTS (EN)		•		•	
V <sub>IH</sub> High-level input voltage		2			V
V <sub>IL</sub> Low-level input voltage				0.7	V
longit gurrant	EN = High, sink current			1	^
Input current	EN = Low, source current	0.5	1.4	5	μΑ
SUPPLY AND LEAKAGE CURRE					
	$V_{I(VSNS)} = 1.5 \text{ V}, \overline{EN} = \text{Low (IN1 active)}, V_{I(IN1)} = 5.5 \text{ V}, V_{I(IN2)} = 3.3 \text{ V}, I_{O(OUT)} = 0 \text{ A}$		55	90	
Supply current from IN1	$V_{I(VSNS)} = 1.5 \text{ V}, \overline{EN} = \text{Low (IN1 active)}, V_{I(IN1)} = 3.3 \text{ V}, V_{I(IN2)} = 5.5 \text{ V}, I_{O(OUT)} = 0 \text{ A}$		1	12	
(operating)	$V_{I(VSNS)} = 0 \text{ V}, \overline{EN} = \text{Low (IN2 active)}, V_{I(IN1)} = 5.5 \text{ V}, V_{I(IN2)} = 3.3 \text{ V}, I_{O(OUT)} = 0 \text{ A}$			75	μΑ
	$V_{I(VSNS)} = 0 \text{ V}, \overline{EN} = \text{Low (IN2 active)}, V_{I(IN1)} = 3.3 \text{ V}, V_{I(IN2)} = 5.5 \text{ V}, I_{O(OUT)} = 0 \text{ A}$			1	
	$V_{I(VSNS)} = 1.5 \text{ V}, \overline{EN} = \text{Low (IN1 active)}, V_{I(IN1)} = 5.5 \text{ V}, V_{I(IN2)} = 3.3 \text{ V}, I_{O(OUT)} = 0 \text{ A}$			1	
Supply current from IN2	$V_{I(VSNS)} = 1.5 \text{ V}, \overline{EN} = \text{Low (IN1 active)}, V_{I(IN1)} = 3.3 \text{ V}, V_{I(IN2)} = 5.5 \text{ V}, I_{O(OUT)} = 0 \text{ A}$			75	μΑ
(operating)	$V_{I(VSNS)} = 0 \text{ V}, \overline{EN} = \text{Low (IN2 active)}, V_{I(IN1)} = 5.5 \text{ V}, V_{I(IN2)} = 3.3 \text{ V}, I_{O(OUT)} = 0 \text{ A}$		1	12	μΑ
	$V_{I(VSNS)} = 0 \text{ V}, \overline{EN} = \text{Low (IN2 active)}, V_{I(IN1)} = 3.3 \text{ V}, V_{I(IN2)} = 5.5 \text{ V}, I_{O(OUT)} = 0 \text{ A}$		55	90	
Quiescent current from IN1	$\overline{\text{EN}}$ = High (inactive), $V_{\text{I(IN1)}}$ = 5.5 V, $V_{\text{I(IN2)}}$ = 3.3 V, $I_{\text{O(OUT)}}$ = 0 A		0.5	2	μА
(STANDBY)	EN = High (inactive), VI(IN1) = 3.3 V, VI(IN2) = 5.5 V, IO(OUT) = 0 A			1	μΛ
Quiescent current from IN2	$\overline{\text{EN}}$ = High (inactive), $V_{\text{I(IN1)}}$ = 5.5 V, $V_{\text{I(IN2)}}$ = 3.3 V, $I_{\text{O(OUT)}}$ = 0 A			1	μA
(STANDBY)	$\overline{\text{EN}}$ = High (inactive), $V_{\text{I(IN1)}}$ = 3.3 V, $V_{\text{I(IN2)}}$ = 5.5 V, $I_{\text{O(OUT)}}$ = 0 A		0.5	2	P** 1
Forward leakage current from IN1 (measured from OUT to GND)	$\overline{\text{EN}}$ = High (inactive), $V_{\text{I(IN1)}}$ = 5.5 V, IN2 open, $V_{\text{O(OUT)}}$ = 0 V (shorted), $T_{\text{J}}$ = 25°C		0.1	5	μΑ
Forward leakage current from IN2 (measured from OUT to GND)	$\overline{\text{EN}}$ = High (inactive), $V_{\text{I(IN2)}}$ = 5.5 V, IN1 open, $V_{\text{O(OUT)}}$ = 0 V (shorted), $T_{\text{J}}$ = 25°C		0.1	5	μΑ
Reverse leakage current to INx (measured from INx to GND)	$\overline{\text{EN}} = \text{High (inactive)}, \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \$		0.3	5	μΑ
STAT OUTPUT				•	
Leakage current	V <sub>O(STAT)</sub> = 5.5 V		0.01	1	μΑ
Saturation voltage	I <sub>I</sub> (STAT) = 2 mA, IN1 switch is on		0.13	0.4	V
Deglitch time (falling edge only)			150		μs
	L				



ELECTRICAL CHARACTERISTICS (Continued) over recommended operating junction temperature range,  $V_{I(IN1)} = V_{I(IN2)} = 5.5 \text{ V}$ ,  $R_{ILIM} = 400 \Omega$  (unless otherwise noted)

	PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
CURF	RENT LIMIT CIRCUIT						
		TPS2112A	R <sub>ILIM</sub> = 400 Ω	0.51	0.63	0.80	
		R <sub>ILIM</sub> = 700 Ω	0.30	0.36	0.50	٨	
	Current limit accuracy	TPS2113A	$R_{ILIM} = 400 \Omega$	0.95	1.25	1.56	Α
		IFSZIISA	R <sub>ILIM</sub> = 700 Ω	0.47	0.71	0.99	
td	Current limit settling time(1)	•	Time for short-circuit output current to settle within 10% of its steady state value.		1		ms
	Input current at ILIM		V <sub>I</sub> (ILIM) = 0 V, I <sub>O</sub> (OUT) = 0 A	-15		0	μΑ

<sup>(1)</sup> Not tested in production.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
VSNS COMPARATOR					
VCNC three held walks are	V <sub>I</sub> (VSNS) ↑	0.78	0.8	0.82	
VSNS threshold voltage	VI(VSNS) ↓	0.735	0.755	0.775	V
VSNS comparator hysteresis(1)		30		60	mV
Deglitch of VSNS comparator (both $\uparrow\downarrow$ )(1)		90	150	220	μs
Input current	0 V ≤ V <sub>I(VSNS)</sub> ≤ 5.5 V	-1		1	μΑ
UVLO					
INTO A STATE OF THE STATE OF TH	Falling edge	1.15	1.25		.,
IN1 and IN2 UVLO	Rising edge		1.30	1.35	V
IN1 and IN2 UVLO hysteresis(1)		30	57	65	mV
Leterard M. LIM O (the bigher of INA and INO)	Falling edge	2.4	2.53		.,
Internal V <sub>DD</sub> UVLO (the higher of IN1 and IN2)	Rising edge		2.58	2.8	V
Internal V <sub>DD</sub> UVLO hysteresis(1)		30	50	75	mV
UVLO deglitch for IN1, IN2 <sup>(1)</sup>	Falling edge		110		μs

<sup>(1)</sup> Not tested in production.

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT				
REVERSE CO	REVERSE CONDUCTION BLOCKING									
ΔVO(I_block)	Minimum output-to-input voltage difference to block switching	$\overline{\text{EN}}$ = high, V <sub>I(IN1)</sub> = 3.3 V and V <sub>I(IN2)</sub> = V <sub>I</sub> (VSNS) = 0 V. Connect OUT to a 5 V supply through a series 1-k $\Omega$ resistor. Let $\overline{\text{EN}}$ = low. Slowly decrease the supply voltage until OUT connects to IN1.	80	100	120	mV				

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
THERMAL SHUTDOWN					
Thermal shutdown threshold <sup>(1)</sup>	TPS211xA is in current limit.	135			
Recovery from thermal shutdown(1)	TPS211xA is in current limit.	125			°C
Hysteresis(1)			10		
IN2-IN1 COMPARATORS					
Hysteresis of IN2–IN1 comparator		0.1		0.2	V
Deglitch of IN2–IN1 comparator (both ↑↓)(1)		10	20	50	μs

<sup>(1)</sup> Not tested in production.



# **SWITCHING CHARACTERISTICS**

over recommended operating junction temperature range,  $V_{I(IN1)} = V_{I(IN2)} = 5.5 \text{ V}$ ,  $R_{ILIM} = 400 \Omega$  (unless otherwise noted)

	242445752	TT0T 00				TPS2112A			TPS2113A		
	PARAMETER	TEST CONDITIONS			TYP	MAX	MIN	TYP	MAX	UNIT	
POWE	POWER SWITCH										
t <sub>r</sub>	Output rise time from an enable <sup>(1)</sup>	V <sub>I</sub> (IN1) = V <sub>I</sub> (IN2) = 5 V, V <sub>I</sub> (VSNS) = 1.5 V	$T_J = 25^{\circ}C, C_L = 1 \mu F,$ $I_L = 500 \text{ mA},$ See Figure 1(a)	0.5	1.0	1.5	1	1.8	3	ms	
tf	Output fall time from a disable <sup>(1)</sup>	V <sub>I</sub> (IN1) = V <sub>I</sub> (IN2) = 5 V, V <sub>I</sub> (VSNS) = 1.5 V	$T_J = 25^{\circ}C$ , $C_L = 1 \mu F$ , $I_L = 500 \text{ mA}$ , See Figure 1(a)	0.35	0.5	0.7	0.5	1	2	ms	
t <sub>t</sub>	Transition time(1)	IN1 to IN2 transition, VI(IN1) = 3.3 V, VI( <u>IN2</u> ) = 5 V, VI( <u>EN</u> ) = 0 V	$T_J$ = 125°C, $C_L$ = 10 μF, $I_L$ = 500 mA [Measure transition time as 10–90% rise time or from 3.4 V to 4.8 V on VO(OUT)], See Figure 1(b)		40	60		40	60	με	
<sup>t</sup> PLH1	Turn-on propagation delay from enable(1)	VI(IN1) = VI(IN2) = 5 V Measured from enable to 10% of VO(OUT), VI(VSNS) = 1.5 V	$T_J = 25^{\circ}C, C_L = 10 \mu F,$ $I_L = 500 \text{ mA},$ See Figure 1(a)		0.5			1		ms	
<sup>t</sup> PHL1	Turn-off propagation delay from a disable(1)	VI(IN1) = VI(IN2) = 5 V, Measured from disable to 90% of VO(OUT), VI(VSNS) = 1.5 V	$T_J = 25^{\circ}C$ , $C_L = 10 \mu F$ , $I_L = 500 \text{ mA}$ , See Figure 1(a)		3			5		ms	
<sup>t</sup> PLH2	Switch-over rising propagation delay <sup>(1)</sup>	Logic 1 to Logic 0 transition on VSNS, VI(IN1) = 1.5 V, VI( <u>IN2</u> ) = 5 V, VI( <u>EN</u> ) = 0 V, Measured from VSNS to 10% of VO(OUT)	$T_J = 25$ °C, $C_L = 10 \mu F$ , $I_L = 500 \text{ mA}$ , See Figure 1(c)		40	100		40	100	μs	
<sup>†</sup> PHL2	Switch-over falling propagation delay(1)	Logic 0 to Logic 1 transition on VSNS VI(IN1) = 1.5V, VI(IN2) = 5V, VI(EN) = 0 V, Measured from VSNS to 90% of VO(OUT)	$T_J = 25$ °C, $C_L = 10 \mu F$ , $I_L = 500 \text{ mA}$ , See Figure 1(c)	2	3	10	2	5	10	ms	

<sup>(1)</sup> Not tested in production.



### TRUTH TABLE

EN	V <sub>I(VSNS)</sub> > 0.8V	$V_{I(IN2)} > V_{I(IN1)}$	STAT	OUT(1)
	Yes	X	0	IN1
0	No	No	0	IN1
	No	Yes	Hi-Z	IN2
1	X	X	0	Hi-Z

# **X** = Don't care.

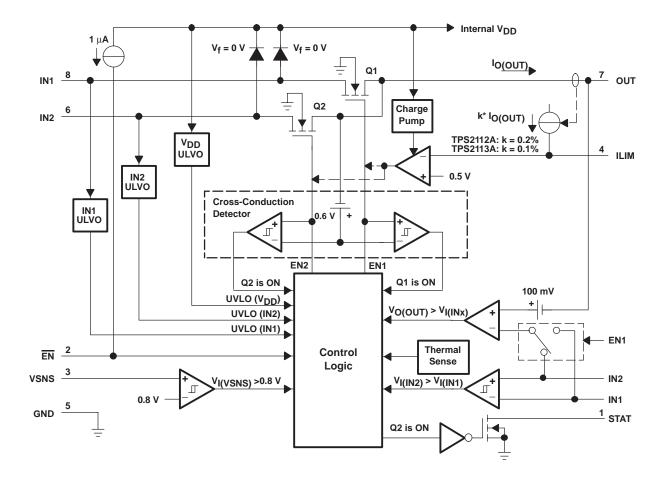
# **Terminal Functions**

TERMINAL			DECORIDATION						
NAME	NO.	I/O	DESCRIPTION						
ĒN	2	_	$\overline{\text{EN}}$ is a TTL- and CMOS-compatible input with a 1- $\mu$ A pull-up. The truth table shown above illustrates the functionality of $\overline{\text{EN}}$ .						
GND	5	I	Ground						
IN1	8	_	Primary power switch input. The IN1 switch can be enabled only if the IN1 supply is above the UVLO threshold and at least one supply exceeds the internal $V_{\mbox{DD}}$ UVLO.						
IN2	6	I	Secondary power switch input. The IN2 switch can be enabled only if the IN2 supply is above the UVLO threshold and at least one supply exceeds the internal V <sub>DD</sub> UVLO.						
ILIM	4	I	A resistor R <sub>ILIM</sub> from ILIM to GND sets the current limit I <sub>L</sub> to 250/R <sub>ILIM</sub> and 500/R <sub>ILIM</sub> for the TPS2112A and TPS2113A, respectively.						
OUT	7	0	Power switch output						
STAT	1	0	STAT is an open-drain output that is Hi-Z if the IN2 switch is ON. STAT pulls low if the IN1 switch is ON or if OUT is Hi-Z (i.e., EN is equal to logic 0)						
VSNS	3	I	An internal power FET connects OUT to IN1 if the VSNS voltage is greater than 0.8 V. Otherwise, the FET connects OUT to the higher of IN1 and IN2. The truth table shown above illustrates the functionality of VSNS.						

<sup>(1)</sup>The under-voltage lockout circuit causes the output OUT to go Hi-Z if the selected power supply does not exceed the IN1/IN2 UVLO, or if neither of the supplies exceeds the internal  $V_{\mbox{DD}}$  UVLO.



# **FUNCTIONAL BLOCK DIAGRAM**





# PARAMETER MEASUREMENT INFORMATION

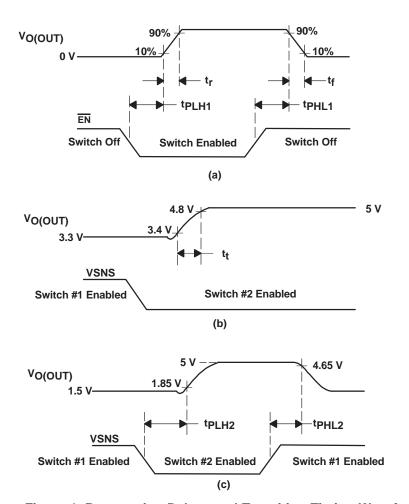


Figure 1. Propagation Delays and Transition Timing Waveforms



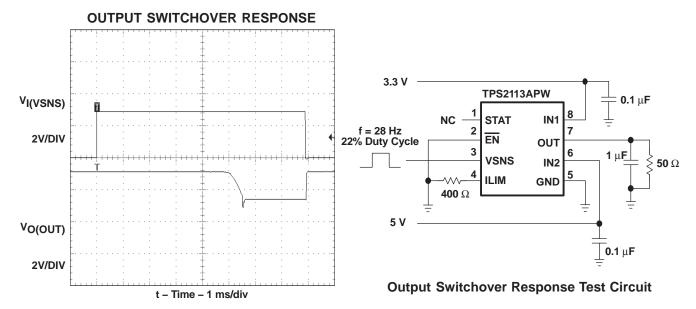


Figure 2

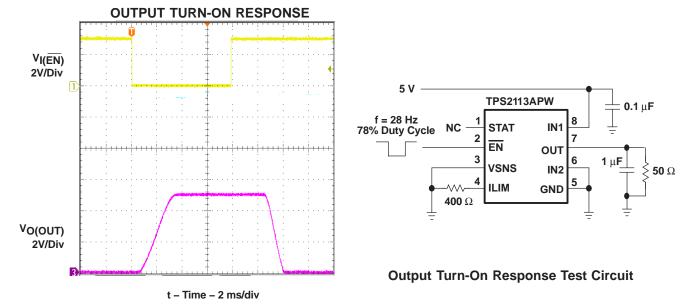
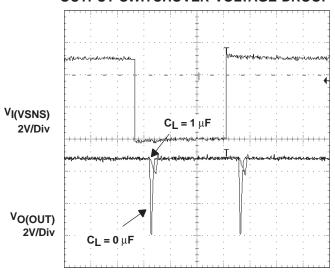


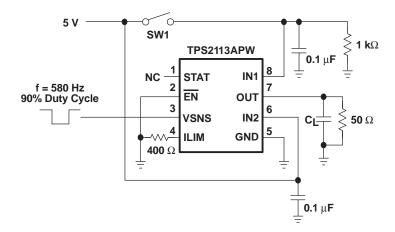
Figure 3



### **OUTPUT SWITCHOVER VOLTAGE DROOP**



t - Time - 40 μs/div



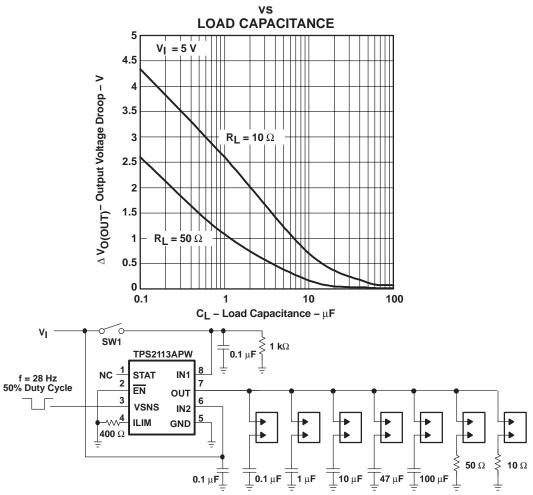
**Output Switchover Voltage Droop Test Circuit** 

Figure 4

NOTE: To initialize the TPS2113A for this test, set input VSNS equal to 0 V, turn on the 5 V supply, and then turn on switch SW1.



# **OUTPUT SWITCHOVER VOLTAGE DROOP**



# **Output Switchover Voltage Droop Test Circuit**

#### Figure 5

NOTE: To initialize the TPS2113A for this test, set input VSNS equal to 0 V, turn on the supply  $V_i$ , and then turn on switch SW1.

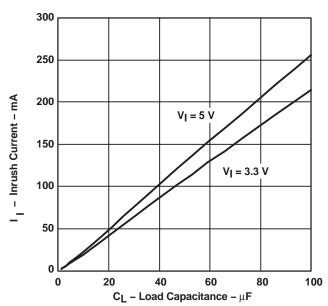


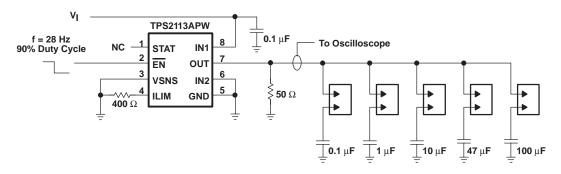
#### **AUTO SWITCHOVER VOLTAGE DROOP** $V_{I(IN1)}$ 2V/Div <u></u> \$1**k**Ω TPS2113A **0.1** μ**F** STAT IN1 2 ΕN OUT **⊙ V**оит f = 220 Hz 3 -≎ 3.3V <sub>10 μF</sub> : 20% Duty Cycle **VSNS** $\gtrsim$ 50 $\Omega$ IN2 ILIM GND = **0.1** μ**F 400**Ω § V<sub>O(OUT)</sub> 2V/Div 75% less output voltage droop compared to TPS2113 **Auto Switchover Voltage Droop Test Circuit** t - Time - 250 μs/div

Figure 6



# INRUSH CURRENT vs LOAD CAPACITANCE

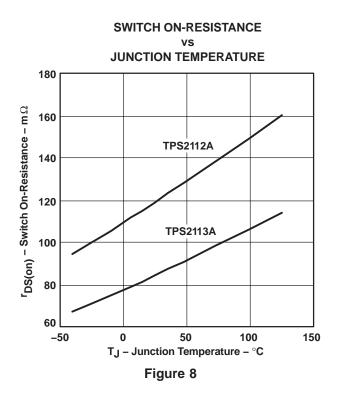


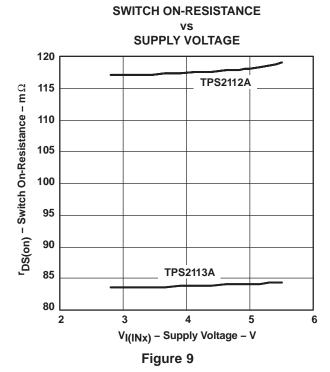


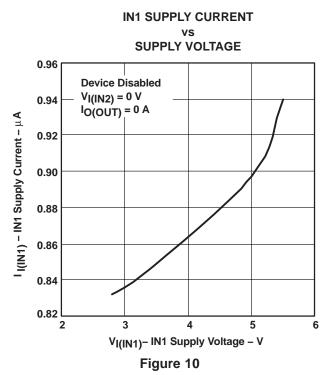
**Output Capacitor Inrush Current Test Circuit** 

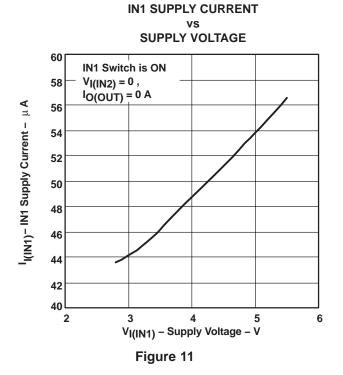
Figure 7



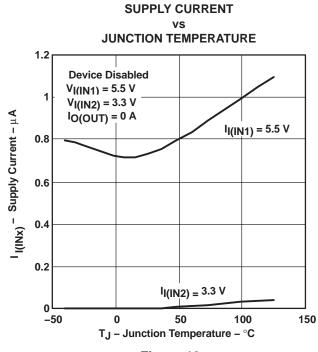




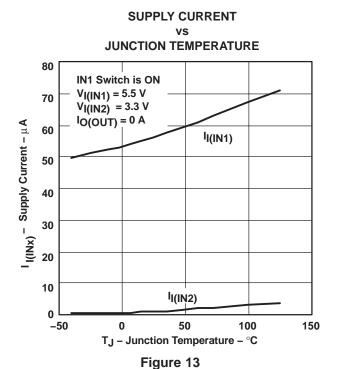














#### APPLICATION INFORMATION.

Some applications have two energy sources, one of which should be used in preference to another. Figure 14 shows a circuit that will connect IN1 to OUT until the voltage at IN1 falls below a user-specified value. Once the voltage on IN1 falls below this value, the TPS2112A/3A will select the higher of the two supplies. This usually means that the TPS2112A/3A will swap to IN2.

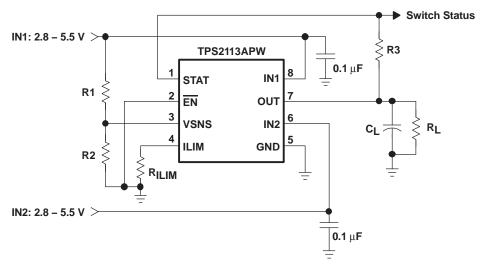


Figure 14. Auto-Selecting for a Dual Power Supply Application

In Figure 15, the multiplexer selects between two power supplies based upon the  $\overline{\text{EN}}$  logic signal. OUT connects to IN1 if  $\overline{\text{EN}}$  is logic 1; otherwise, OUT connects to IN2. The logic thresholds for the  $\overline{\text{EN}}$  terminal are compatible with both TTL and CMOS logic.

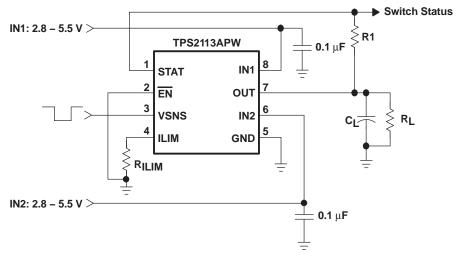


Figure 15. Manually Switching Power Sources



#### **DETAILED DESCRIPTION**

#### **AUTO-SWITCHING MODE**

The TPS2112A/3A only supports the auto-switching mode. In this mode, OUT connects to IN1 if  $V_{I(VSNS)}$  is greater than 0.8 V, otherwise OUT connects to the higher of IN1 and IN2.

The VSNS terminal includes hysteresis equal to 3.75–7.5% of the threshold selected for transition from the primary supply to the higher of the two supplies. This hysteresis helps avoid repeated switching from one supply to the other due to resistive drops.

#### **N-CHANNEL MOSFETs**

Two internal high-side power MOSFETs implement a single-pole double-throw (SPDT) switch. Digital logic selects the IN1 switch, IN2 switch, or no switch (Hi-Z state). The MOSFETs have no parallel diodes so output-to-input current cannot flow when the FET is off. An integrated comparator prevents turn-on of a FET switch if the output voltage is greater than the input voltage.

#### **CROSS-CONDUCTION BLOCKING**

The switching circuitry ensures that both power switches will never conduct at the same time. A comparator monitors the gate-to-source voltage of each power FET and allows a FET to turn on only if the gate-to-source voltage of the other FET is below the turn-on threshold voltage.

#### REVERSE-CONDUCTION BLOCKING

When the TPS211xA switches from a higher-voltage supply to a lower-voltage supply, current can potentially flow back from the load capacitor into the lower-voltage supply. To minimize such reverse conduction, the TPS211xA will not connect a supply to the output until the output voltage has fallen to within 100 mV of the supply voltage. Once a supply has been connected to the output, it will remain connected regardless of output voltage.

#### **CHARGE PUMP**

The higher of supplies IN1 and IN2 powers the internal charge pump. The charge pump provides power to the current limit amplifier and allows the output FET gate voltage to be higher than the IN1 and IN2 supply voltages. A gate voltage that is higher than the source voltage is necessary to turn on the N-channel FET.

#### **CURRENT LIMITING**

A resistor  $R_{ILIM}$  from ILIM to GND sets the current limit to  $250/R_{ILIM}$  and  $500/R_{ILIM}$  for the TPS2112A and TPS2113A, respectively. Setting resistor  $R_{ILIM}$  equal to zero is not recommended as that disables current limiting.

#### **OUTPUT VOLTAGE SLEW-RATE CONTROL**

The TPS2112A/3A slews the output voltage at a slow rate when OUT switches to IN1 or IN2 from the Hi-Z state (see *Truth Table*). A slow slew rate limits the inrush current into the load capacitor. High inrush currents can glitch the voltage bus and cause a system to hang up or reset. It can also cause reliability issues—like pit the connector power contacts, when hot-plugging a load such as a PCI card. The TPS2112A/3A slews the output voltage at a much faster rate when OUT switches between IN1 and IN2. The fast rate minimizes the output voltage droop and reduces the output voltage hold-up capacitance requirement.



#### PACKAGE OPTION ADDENDUM

4-Mar-2005

#### **PACKAGING INFORMATION**

Orderable Device	Status <sup>(1)</sup>	Package Type	Package Drawing	Pins	Package Qty	Eco Plan <sup>(2)</sup>	Lead/Ball Finish	MSL Peak Temp <sup>(3)</sup>
TPS2112APW	ACTIVE	TSSOP	PW	8	150	None	CU NIPDAU	Level-1-220C-UNLIM
TPS2112APWR	ACTIVE	TSSOP	PW	8	2000	None	CU NIPDAU	Level-1-220C-UNLIM
TPS2113APW	ACTIVE	TSSOP	PW	8	150	None	CU NIPDAU	Level-1-220C-UNLIM
TPS2113APWR	ACTIVE	TSSOP	PW	8	2000	None	CU NIPDAU	Level-1-220C-UNLIM
TPS2113APWRG4	PREVIEW	TSSOP	PW	8	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM

<sup>(1)</sup> The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

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**OBSOLETE:** TI has discontinued the production of the device.

(2) Eco Plan - May not be currently available - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

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(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDECindustry standard classifications, and peak solder temperature.

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# PW (R-PDSO-G\*\*)

#### 14 PINS SHOWN

# PLASTIC SMALL-OUTLINE PACKAGE



NOTES: A. All linear dimensions are in millimeters.

B. This drawing is subject to change without notice.

C. Body dimensions do not include mold flash or protrusion not to exceed 0,15.

D. Falls within JEDEC MO-153

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