

Precision Quad Supply Monitor in 6-Lead SOT-23

FEATURES

- **Ultralow Voltage Reset:** $V_{CC} = 0.5V$ Guaranteed
- **Monitor Four Inputs Simultaneously**
 3.3V, 2.5V, 1.8V, ADJ (LTC2903-A1)
 5V, 3.3V, 2.5V, 1.8V (LTC2903-B1)
 5V, 3.3V, 1.8V, $-5.2V$ (LTC2903-C1)
- **Guaranteed Threshold Accuracy:** $\pm 1.5\%$ of Monitored Voltage over Temperature
- **10% Undervoltage Monitoring**
- **Low Supply Current:** $20\mu A$ Typical
- 200ms Reset Time Delay
- Active Low Open-Drain RST Output
- Power Supply Glitch Immunity
- Low Profile (1mm) SOT-23 (ThinSOT™) Package

APPLICATIONS

- Multivoltage Systems
- Optical Networking Systems
- Cell Phone Base Stations
- Network Servers

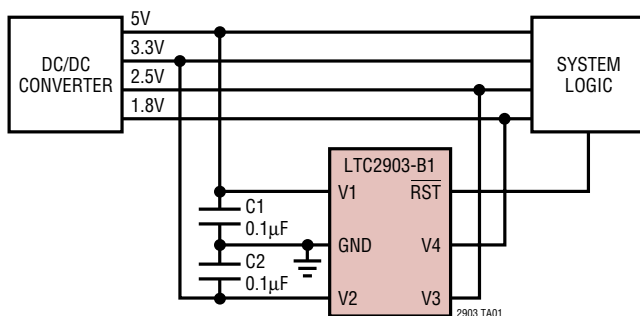
DESCRIPTION

The LTC2903-1 monitors up to four supply voltages. The common reset output remains low until all four inputs have been in compliance for 200ms. Voltage thresholds maintain $\pm 1.5\%$ accuracy over temperature (with respect to the monitored voltage). The LTC2903-1 features an open-drain RST output with a weak internal pullup.

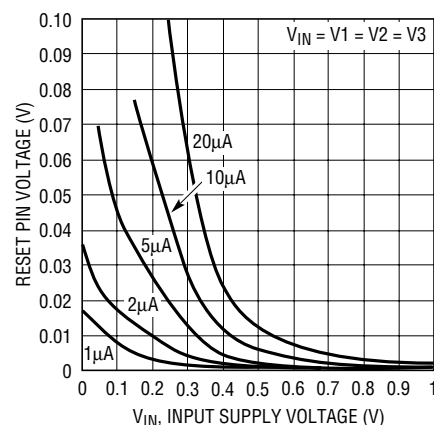
Internal supply voltage (V_{CC}) is generated from the greater voltage on the V1, V2 inputs. The RST output is guaranteed to sink at least $5\mu A$ ($V_{OL} = 0.15V$) for V1, V2 or V3 down to 0.5V and will typically conduct current down to 0V. Quiescent current is $20\mu A$ typical, making the LTC2903-1 ideal for power conscious systems. The LTC2903-1 is available in a 6-lead low profile (1mm) SOT-23 package.

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 ThinSOT is a trademark of Linear Technology Corporation.

TYPICAL APPLICATION



Low Voltage Reset Pull-Down Performance vs External Pull-Up Current and Input Supply Voltage



2903 TAO1b

ABSOLUTE MAXIMUM RATINGS

(Notes 1, 2, 3)

V1, V2	-0.3V to 6.5V
V3	2.7V or ($V_{CC} + 0.3V$)
V4 (LTC2903-A1, LTC2903-B1)	-0.3V to 6.5V
V4 (LTC2903-C1)	-6.5V to 0.3V
RST	-0.3V to 6.5V
Operating Temperature Range	
LTC2903C-X1	0°C to 70°C
LTC2903I-X1	-40°C to 85°C
Storage Temperature Range	-65°C to 150°C
Lead Temperature (Soldering, 10 sec)	300°C

PACKAGE/ORDER INFORMATION

TOP VIEW S6 PACKAGE 6-LEAD PLASTIC TSOT-23 $T_{JMAX} = 125^{\circ}\text{C}$, $\theta_{JA} = 230^{\circ}\text{C/W}$	
ORDER PART NUMBER	S6 PART MARKING
LTC2903CS6-A1	LTAJV
LTC2903CS6-B1	LTAJN
LTC2903CS6-C1	LTAJQ
LTC2903IS6-A1	LTAJW
LTC2903IS6-B1	LTAJP
LTC2903IS6-C1	LTAJR

Consult LTC Marketing for parts specified with wider operating temperature ranges.

ELECTRICAL CHARACTERISTICS

(LTC2903-A1) The ● denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^{\circ}\text{C}$. $V_{CC} = 3.3V$ unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
V_{RT33}	3.3V, 10% Reset Threshold	V1 Input Threshold	●	2.871	2.921	2.970	V
V_{RT25}	2.5V, 10% Reset Threshold	V2 Input Threshold	●	2.175	2.213	2.250	V
V_{RT18}	1.8V, 10% Reset Threshold	V3 Input Threshold	●	1.566	1.593	1.620	V
V_{RTA}	Adjustable Reset Threshold	V4 Input Threshold	●	0.492	0.500	0.508	V

(LTC2903-B1) The ● denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^{\circ}\text{C}$. $V_{CC} = 5V$ unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
V_{RT50}	5V, 10% Reset Threshold	V1 Input Threshold	●	4.350	4.425	4.500	V
V_{RT33}	3.3V, 10% Reset Threshold	V2 Input Threshold	●	2.871	2.921	2.970	V
V_{RT25}	2.5V, 10% Reset Threshold	V3 Input Threshold	●	2.175	2.213	2.250	V
V_{RT18}	1.8V, 10% Reset Threshold	V4 Input Threshold	●	1.566	1.593	1.620	V

(LTC2903-C1) The ● denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^{\circ}\text{C}$. $V_{CC} = 5V$ unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
V_{RT50}	5V, 10% Reset Threshold	V1 Input Threshold	●	4.350	4.425	4.500	V
V_{RT33}	3.3V, 10% Reset Threshold	V2 Input Threshold	●	2.871	2.921	2.970	V
V_{RT18}	1.8V, 10% Reset Threshold	V3 Input Threshold	●	1.566	1.593	1.620	V
V_{RT52N}	-5.2V, 10% Reset Threshold	V4 Input Threshold	●	-4.524	-4.602	-4.680	V

ELECTRICAL CHARACTERISTICS

The ● denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^\circ\text{C}$. $V_{CC} = 3.3\text{V}$ unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
I_{V1}	V1 Input Current (Note 4)	V1 = 3.3V (LTC2903-A1)	●		20	80	μA
		V1 = 5V (LTC2903-B1, LTC2903-C1)	●		25	80	μA
I_{V2}	V2 Input Current (Note 4)	V2 = 2.5V (LTC2903-A1)	●		8	30	μA
		V2 = 3.3V (LTC2903-B1, LTC2903-C1)	●		10	30	μA
I_{V3}	V3 Input Current	V3 = 1.8V (LTC2903-A1, LTC2903-C1)	●		6	30	μA
		V3 = 2.5V (LTC2903-B1)	●		8	30	μA
I_{V4}	V4 Input Current	V4 = 0.55V (LTC2903-A1)	●			± 15	nA
		V4 = 1.8V (LTC2903-B1)	●		2	4	μA
		V4 = -5.2V (LTC2903-C1)	●		-3	-6	μA
t_{RST}	Reset Time-Out Period		●	140	200	260	ms
t_{UV}	V_X Undervoltage Detect to $\overline{RST}$	V_X Less Than Threshold V_{RTX} by More Than 1%			150		μs
V_{OH}	Output Voltage High $\overline{RST}$ (LTC2903-1) (Note 5)	$I_{RST(DN)} = -1\mu\text{A}$	●	$V_2 - 1$			V
V_{OL}	Output Voltage Low $\overline{RST}$ (Note 6)	$V_{CC} = 0.2\text{V}$, $I_{RST} = 0.1\mu\text{A}$	●		5	60	mV
		$V_{CC} = 0.5\text{V}$, $I_{RST} = 5\mu\text{A}$	●		10	150	mV
		$V_{CC} = 1\text{V}$, $I_{RST} = 200\mu\text{A}$	●		25	300	mV
		$V_{CC} = 3\text{V}$, $I_{RST} = 2500\mu\text{A}$	●		100	300	mV

Note 1: Absolute Maximum Ratings are those values beyond which the life of the device may be impaired.

Note 2: All currents into pins are positive, all voltages are referenced to GND unless otherwise noted.

Note 3: The internal supply voltage (V_{CC}) is generated from the greater voltage on the V1, V2 inputs.

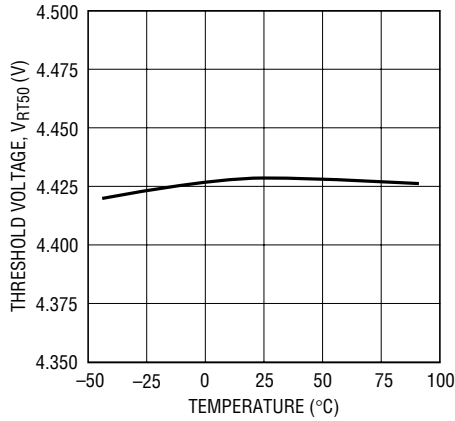
Note 4: Under typical operating conditions, quiescent current is drawn from the V1 input. When V2 exceeds V1, V2 supplies the quiescent current.

Note 5: The $\overline{RST}$ output pin on the LTC2903-1 has an internal pull-up to V_2 of typically $10\mu\text{A}$. However, for faster rise times or for V_{OH} voltages greater than V_2 , use an external pull-up resistor.

Note 6: At input voltages below 1V on V1 and V2, voltage on V3 assists pulling down the $\overline{RST}$ output.

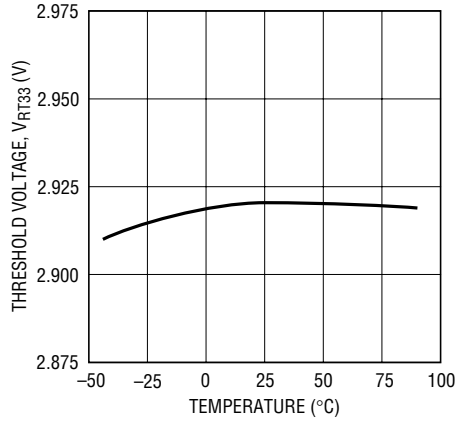
TYPICAL PERFORMANCE CHARACTERISTICS

5V Threshold Voltage vs Temperature



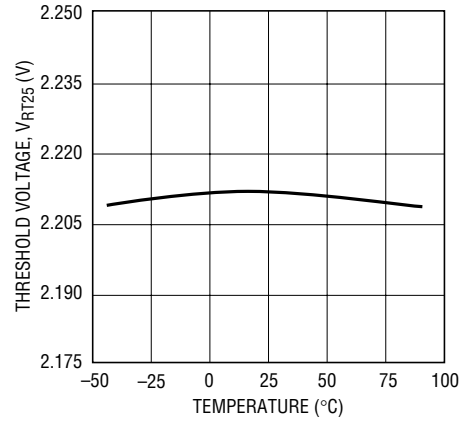
28031 G01

3.3V Threshold Voltage vs Temperature



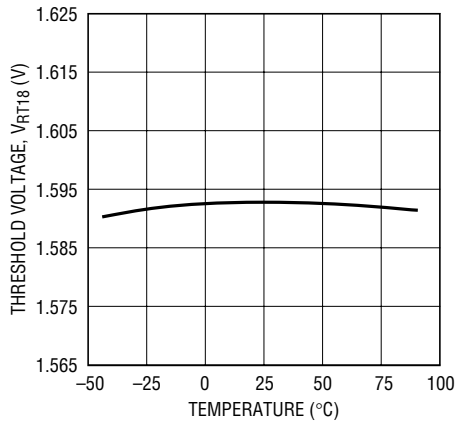
28031 G02

2.5V Threshold Voltage vs Temperature



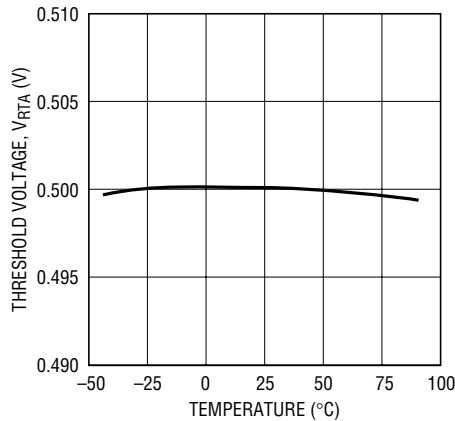
28031 G03

1.8V Threshold Voltage vs Temperature



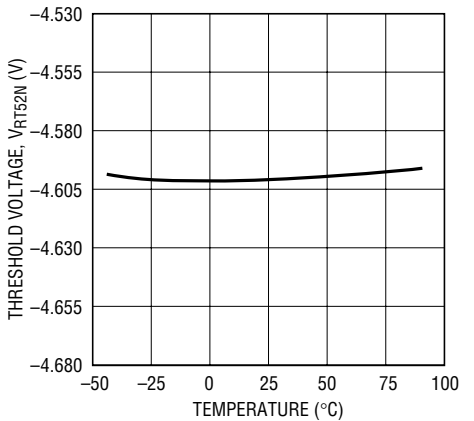
28031 G04

ADJ Threshold Voltage vs Temperature



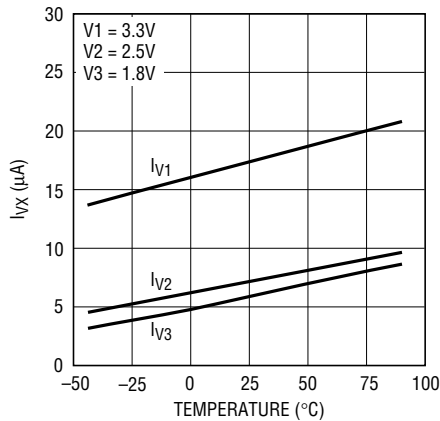
28031 G05

-5.2V Threshold Voltage vs Temperature



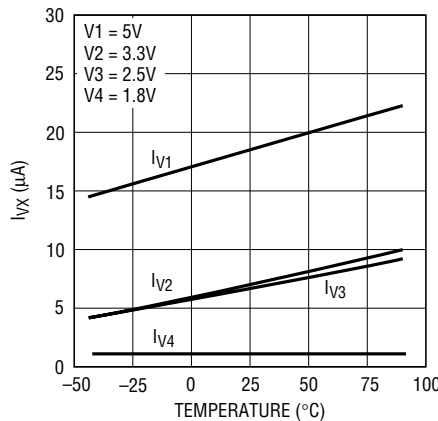
28031 G06

Supply Currents vs Temperature (LTC2903A)



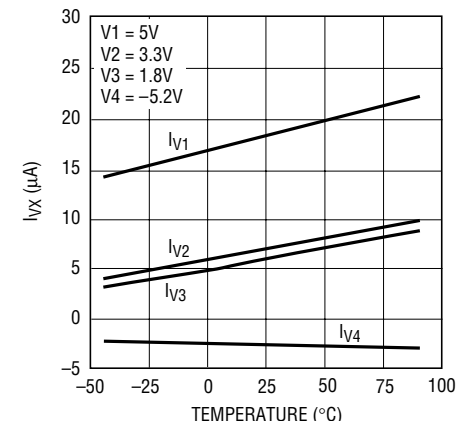
29031 G07

Supply Currents vs Temperature (LTC2903B)



29031 G08

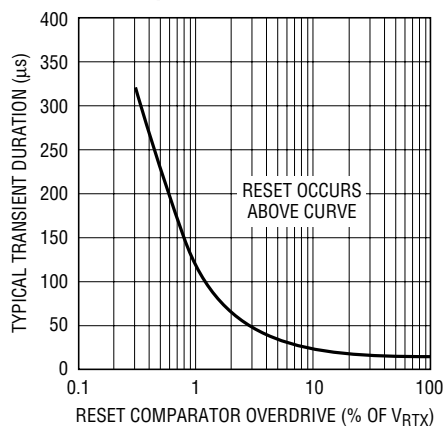
Supply Currents vs Temperature (LTC2903C)



29031 G09

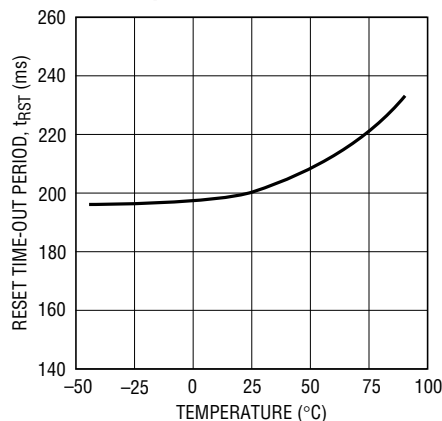
TYPICAL PERFORMANCE CHARACTERISTICS

**Transient Duration
vs Comparator Overdrive**



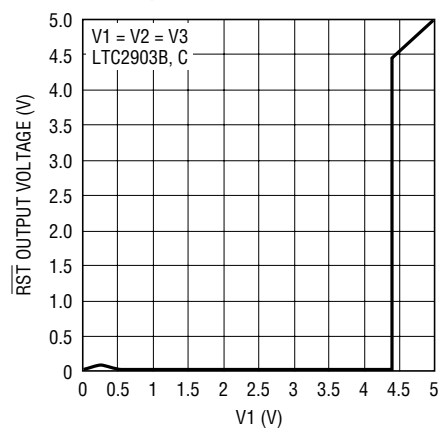
29031 G10

**Reset Time-Out Period
vs Temperature**



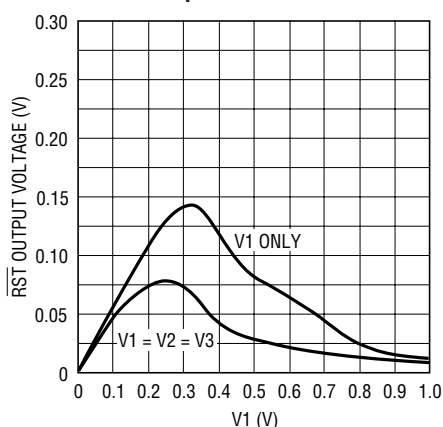
29031 G11

**RST Output Voltage with 10k
Pull-Up to V1**



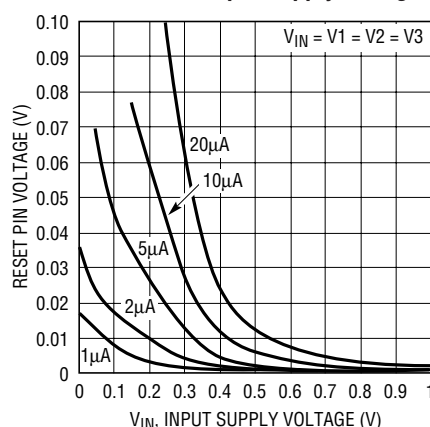
29031 G12

**RST Output Voltage with
10k Pull-Up to V1**



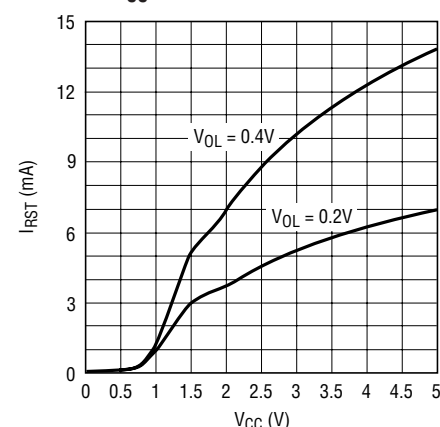
29031 G13

**Low Voltage Reset Pull-Down
Performance vs External Pull-Up
Current and Input Supply Voltage**



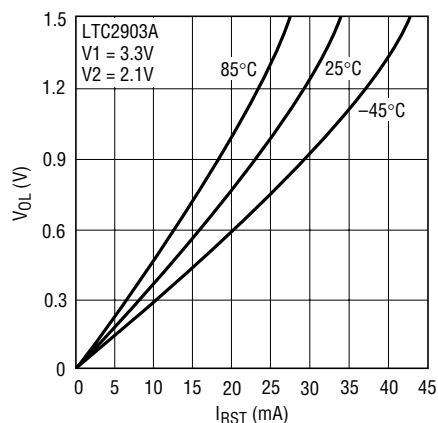
29031 G14

**RST Current Sink Capability
vs V_{CC}**



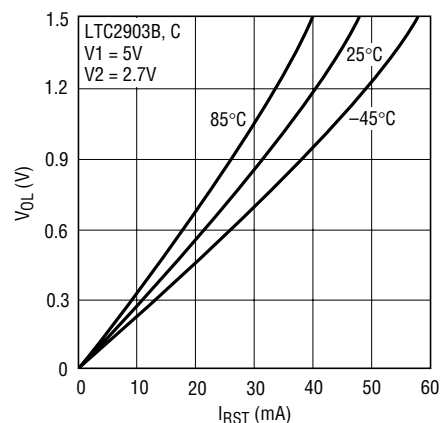
29031 G15

**RST Voltage Output Low
vs RST Sink Current**



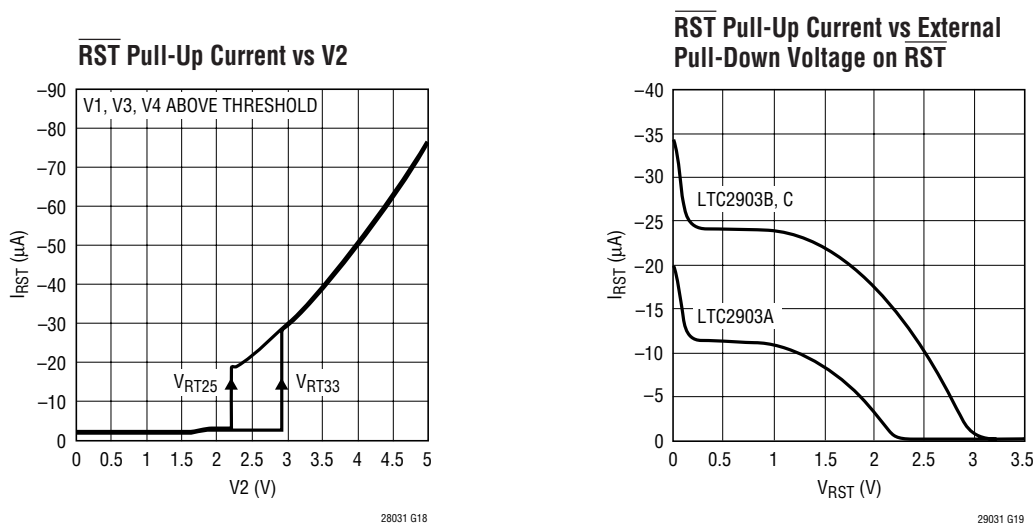
29031 G16

**RST Voltage Output Low
vs RST Sink Current**



29031 G16

TYPICAL PERFORMANCE CHARACTERISTICS



PIN FUNCTIONS

V1 (Pin 1): Voltage Input 1 (5V, 3.3V). Internal V_{CC} is generated from the greater voltage on the V1, V2 inputs. Bypass this pin to ground with a 0.1 μF (or greater) capacitor.

GND (Pin 2): Ground.

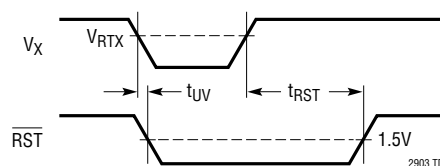
V2 (Pin 3): Voltage Input 2 (3.3V, 2.5V). Internal V_{CC} is generated from the greater voltage on the V1, V2 inputs. Bypass this pin to ground with a 0.1 μF (or greater) capacitor.

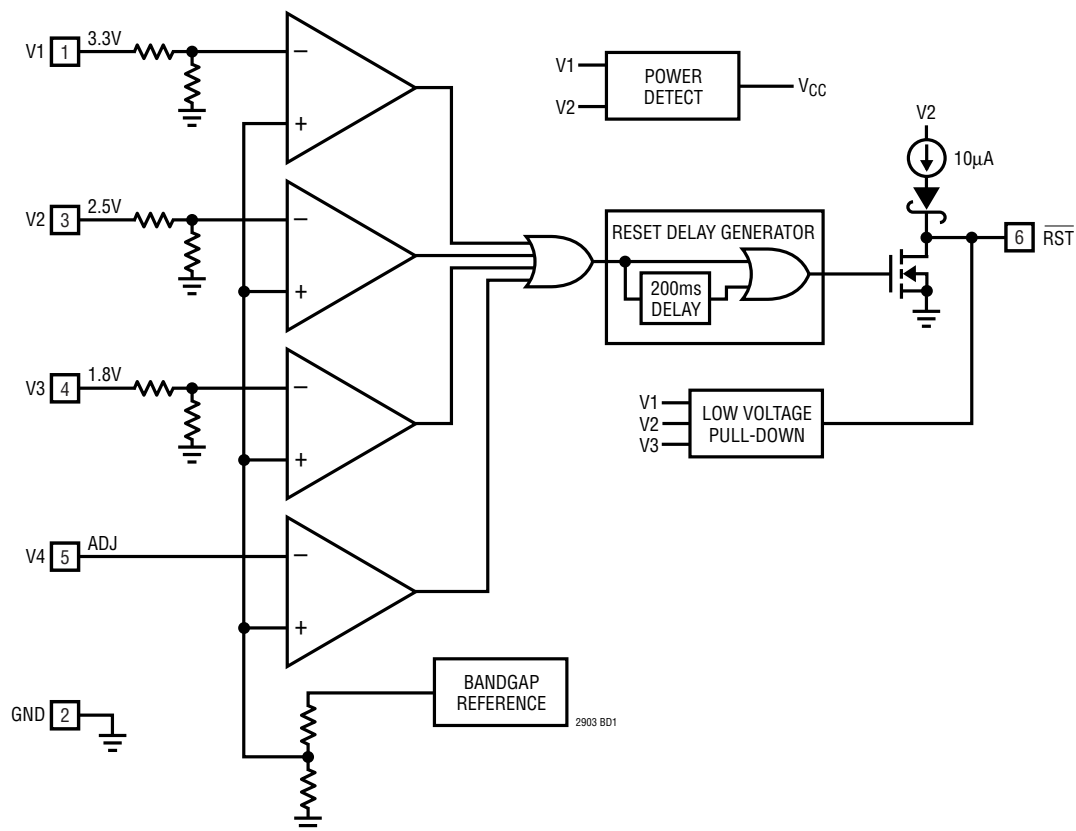
V3 (Pin 4): Voltage Input 3 (2.5V, 1.8V). This input assists the $\overline{RST}$ pull-down circuitry below 1V.

V4 (Pin 5): Voltage Input 4 (ADJ, 1.8V, -5.2V). See Table 1 for recommended ADJ resistor values.

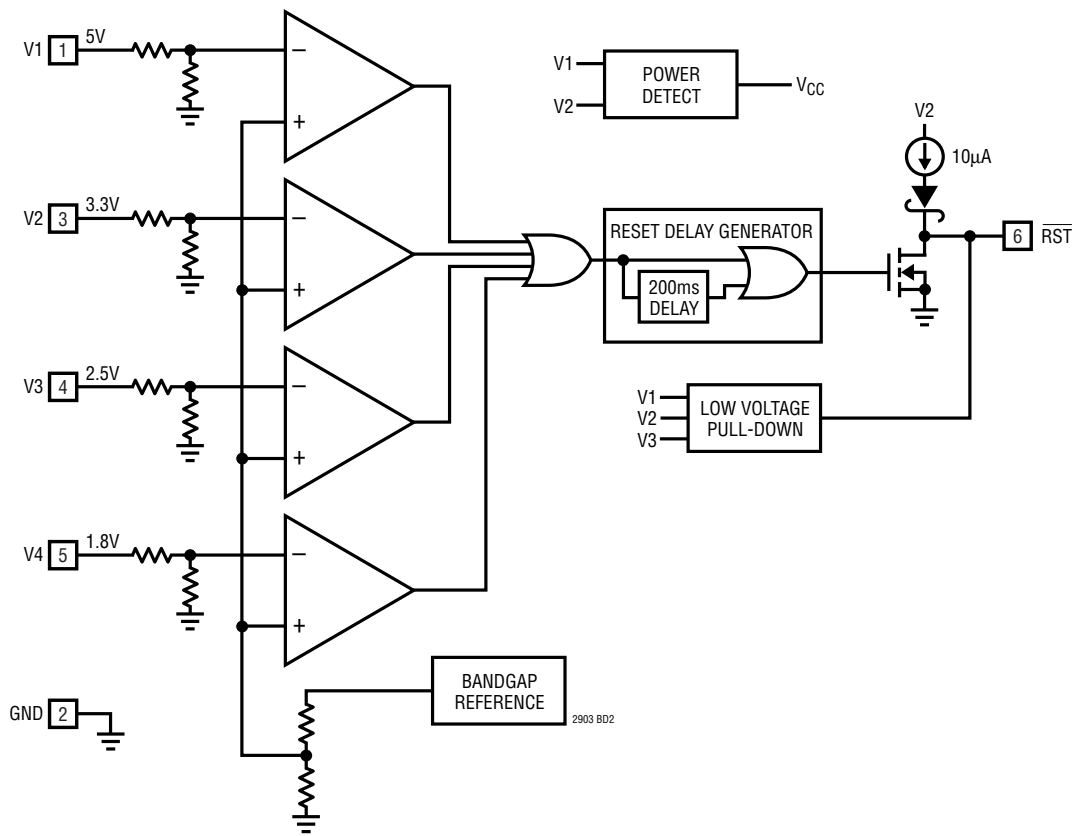
$\overline{RST}$ (Pin 6): Reset Logic Output. Pulls low when any voltage input is below reset threshold and held low for 200ms after all voltage inputs exceed threshold. The pin contains a weak pull-up to V2. Use an external pull-up for faster rise times or output voltages greater than V2.

TIMING DIAGRAM

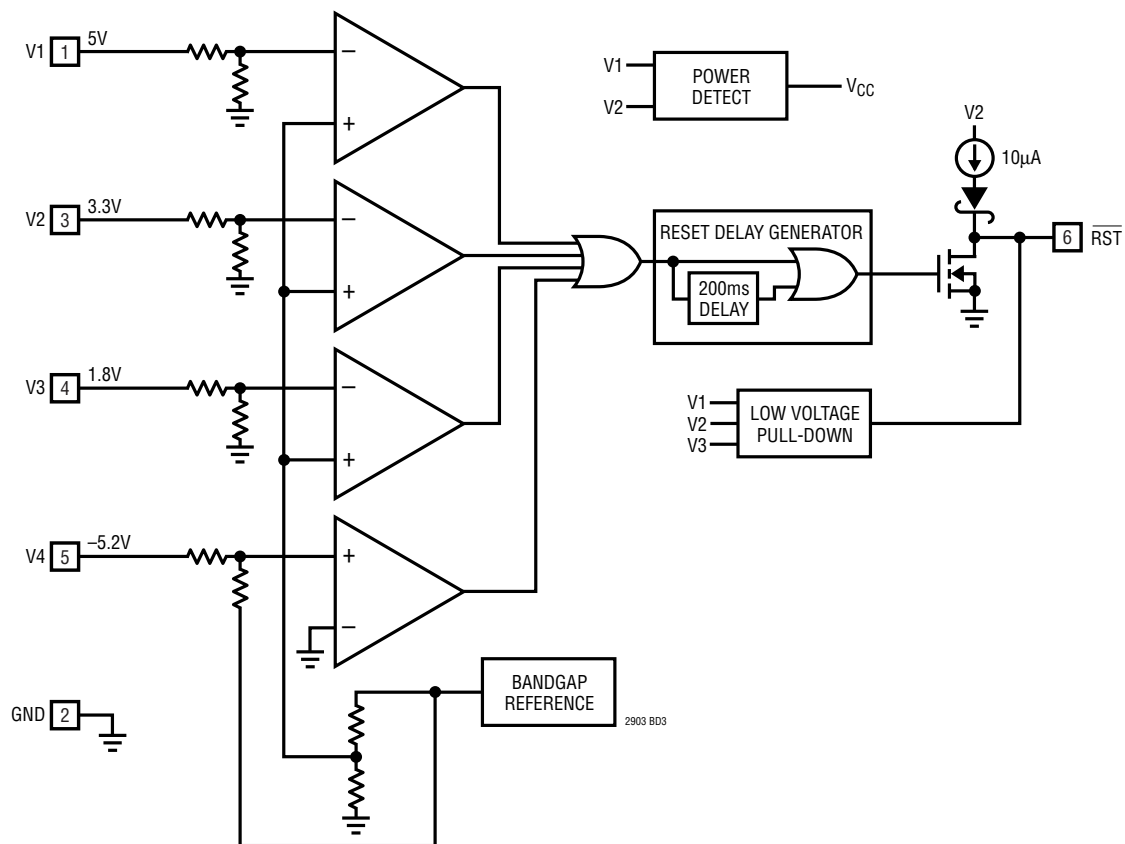


BLOCK DIAGRAMS (LTC2903-A1)

BLOCK DIAGRAMS (LTC2903-B1)



BLOCK DIAGRAMS (LTC2903-C1)



APPLICATIONS INFORMATION

Power-Up

The LTC2903-1 issues a logic low on the $\overline{\text{RST}}$ output when an input supply voltage resides below the prescribed threshold voltage. Ideally, the $\overline{\text{RST}}$ logic output would remain low with the input supply voltage down to zero volts. Most supervisors lack pull-down capability below 1V. The LTC2903-1 power supply supervisors incorporate a new low voltage pull-down circuit that can hold the $\overline{\text{RST}}$ line low with as little as 200mV of input supply voltage on V1, V2 or V3. The pull-down circuit helps maintain a low impedance path to ground, reducing the risk of floating the $\overline{\text{RST}}$ node to undetermined voltages. Such voltages may trigger external logic causing erroneous reset operation(s). Furthermore, a mid-scale voltage could cause external circuits to operate in the middle of their voltage transfer characteristic, consuming more quiescent current than normal. These conditions could cause serious system reliability problems.

When V1, V2 and V3 are ramped simultaneously, the reset pull-down current increases up to three times the current that may be pulled with a single input. Figure 1 demonstrates the reset pin current sinking ability for single supply and triple supply-tracking applications. Figure 2 shows a detailed view of the reset pin voltage with a 10k pull-up resistor to V1.

The LTC2903-1 supervisors derive their internal supply voltage (V_{CC}) automatically from the greater voltage on the V1 and V2 inputs. With all supply inputs above threshold, the quiescent current drawn from V_{CC} is 20 μA (typ).

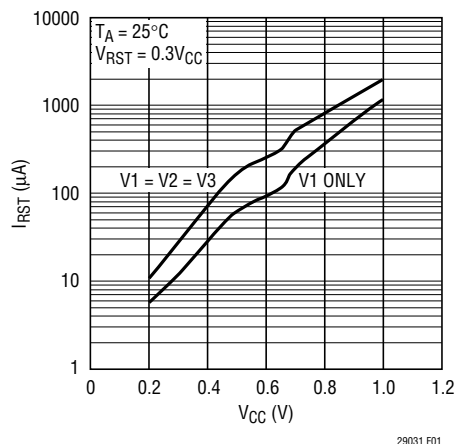


Figure 1. $\overline{\text{RST}}$ Pull-Down Current vs V_{CC}

Supply Monitoring

The LTC2903-1 accurately monitors four inputs in a small 6-lead SOT-23 package. The low voltage reset output includes an integrated 200ms reset delay timer. The reset line pulls high 200ms after all voltage inputs exceed their respective thresholds. The reset output remains low during power-up, power-down and brownout conditions on any of the voltage inputs.

For applications requiring an adjustable trip threshold, use the V4 input on the LTC2903-A1. Connect the tap point on an external resistive divider (R1, R2) placed between the positive voltage being sensed and ground, to the high impedance input on V4. The LTC2903-A1 compares the voltage on the V4 pin to the internal 0.5V reference. Figure 3 shows a generic setup for the positive adjustable application.

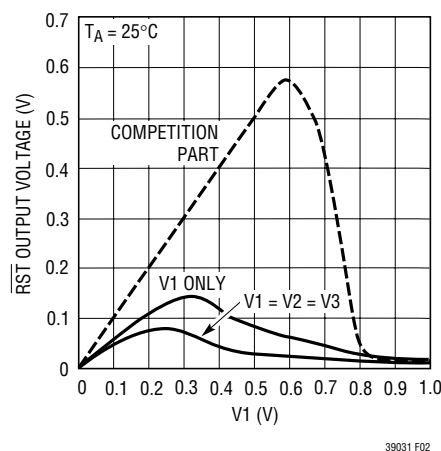


Figure 2. $\overline{\text{RST}}$ Output Voltage with a 10k Pull-Up to V1 (Enlarged Area of Detail)

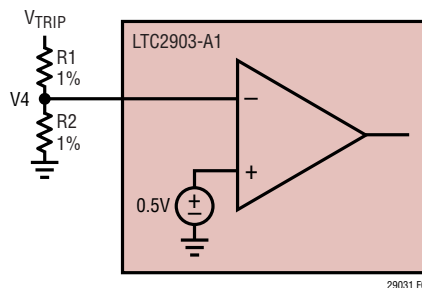


Figure 3. Setting the Positive Adjustable Trip Point

APPLICATIONS INFORMATION

Calculate the trip voltage from:

$$V_{TRIP} = 0.5V \left(1 + \frac{R1}{R2} \right)$$

Table 1 contains suggested 1% resistor values for the ADJ input to obtain nominal $\pm 11.5\%$ thresholds.

Table 1. Suggested 1% Resistor Values for the ADJ Input

V _{SUPPLY} (V)	V _{TRIP} (V)	R1 (k Ω)	R2(k Ω)
12	10.75	2050	100
10	8.95	1690	100
8	7.15	1330	100
7.5	6.7	1240	100
6	5.38	976	100
5	4.435	787	100
3.3	2.935	487	100
3	2.66	432	100
2.5	2.2	340	100
1.8	1.605	221	100
1.5	1.325	165	100
1.2	1.065	113	100
1	0.884	76.8	100
0.9	0.795	59	100

Connect unused supervisor inputs to the highest supply voltage available (typically V1). On the LTC2903-C1, the negative V4 input must always be applied.

Implications of Threshold Accuracy

Specifying system voltage margin for worst-case operation requires consideration of three factors: power supply tolerance, IC supply voltage tolerance and supervisor reset threshold accuracy. Highly accurate supervisors ease the design challenge by decreasing the overall voltage margin required for reliable system operation. Consider a 5V system with a $\pm 10\%$ power supply tolerance band. System ICs powered by this supply must operate reliably within this band (and a little more, as explained below). The bottom of the supply tolerance band, at 4.5V (5V $\pm 10\%$), is the exact voltage at which a *perfectly accurate* supervisor generates a reset. Such a perfectly accurate

supervisor does not exist—the actual reset threshold may vary over a specified band ($\pm 1.5\%$ for the LTC2903-1 supervisors). Figure 4 shows the typical relative threshold accuracy for all four inputs, over temperature.

With this variation of reset threshold in mind, the nominal reset threshold of the supervisor resides *below* the minimum supply voltage; just enough so that the reset threshold band and the power supply tolerance bands do not overlap. If the two bands overlap, the supervisor could generate a false or nuisance reset when the power supply remains within its specified tolerance band (say, at 4.6V).

Adding half of the reset threshold accuracy spread (1.5%) to the ideal 10% thresholds puts the LTC2903-1 thresholds at 11.5% (typ) below the nominal input voltage. For example, the 5V typical threshold is 4.425V, or 75mV below the ideal threshold of 4.500V. The guaranteed threshold lies in the band between 4.500V and 4.350V over temperature.

The powered system must work reliably down to the lowest voltage in the threshold band or risk malfunction before the reset line falls. In the 5V example, using the 1.5% accurate supervisor, the system ICs must work down to 4.35V. System ICs working with a $\pm 2.5\%$ accurate supervisor must operate down to 4.25V, increasing the required system voltage margin and the probability of system malfunction.

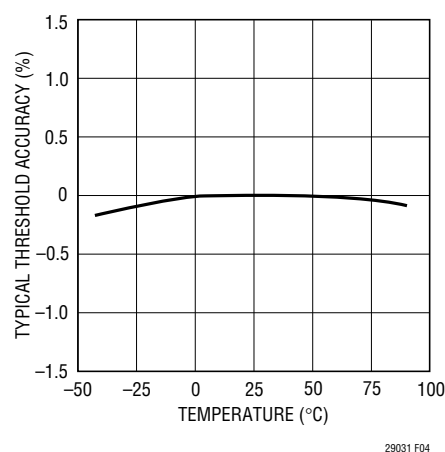


Figure 4. LTC2903 Typical Threshold Accuracy vs Temperature

APPLICATIONS INFORMATION

In any supervisory application, supply noise riding on the monitored DC voltage can cause spurious resets, particularly when the monitored voltage approaches the reset threshold. A less than desirable but commonly used technique used to mitigate this problem adds hysteresis to the input comparator. The amount of added hysteresis, usually specified as a percentage of the trip threshold, effectively degrades the advertised accuracy of the part. To maintain high accuracy, the LTC2903-1 does not use hysteresis.

To minimize spurious resets while maintaining threshold accuracy, the LTC2903-1 employs two forms of noise filtering. The first line of defense incorporates proprietary tailoring of the comparator transient response. Transient events receive electronic integration in the comparator and must exceed a certain magnitude and duration to cause the comparator to switch. Figure 5 illustrates the typical transient duration versus comparator overdrive (as a percentage of the trip threshold V_{RT}) required to trip the comparators. Once any comparator is switched, the reset line pulls low. The reset time-out counter starts once all inputs return above threshold. The nominal reset delay time is 200ms. The counter clears whenever any input

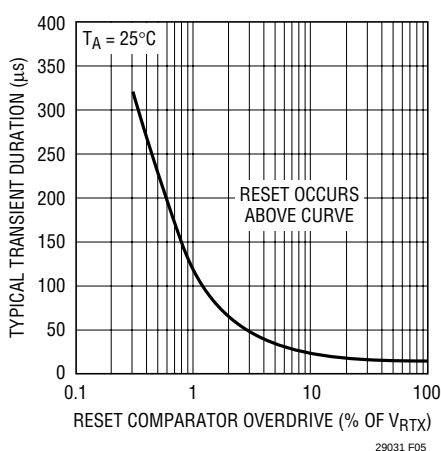


Figure 5. Typical Transient Duration vs Overdrive Required to Trip Comparator

drops back below threshold. This reset delay time effectively provides further filtering of the voltage inputs. A noisy input with frequency components of sufficient magnitude above $f = 1/t_{RST} = 5\text{Hz}$ holds the reset line low, preventing oscillatory behavior on the reset line.

Although all four comparators have built-in glitch filtering, use bypass capacitors on the V1 and V2 inputs because the greater of V1 or V2 supplies the V_{CC} for the part (a $0.1\mu\text{F}$ ceramic capacitor satisfies most applications). Apply filter capacitors on the V3 and V4 inputs in extremely noisy situations.

Reset Output Rise and Fall Time Estimation

The reset output line contains a weak pull-up current source to the V2 supply. Use an external pull-up resistor when the output needs to pull to another voltage and/or when the reset output needs a faster rise time. The open-drain output allows for wired-OR connections when more than one signal needs to pull down on the reset line. Estimate output rise time for the open-drain output without an external pull-up using:

$$t_{RISE} \approx 2.2 \cdot R_{PU} \cdot C_{LOAD}$$

where R_{PU} is the on-resistance of the pull-up transistor and C_{LOAD} is the external load capacitance on the pin. At room temperature, the average R_{PU} is approximately $50\text{k}\Omega$. When externally pulling up to voltages higher than V2, an internal network automatically protects the weak pull-up circuitry from reverse currents.

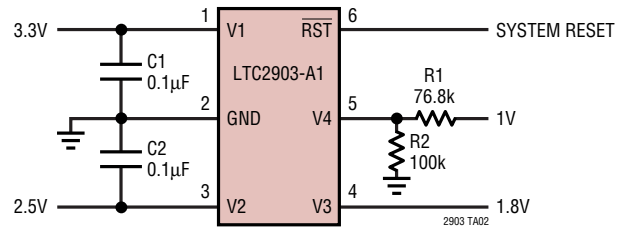
The reset output has very strong pull-down capability. Estimate the output fall time using:

$$t_{FALL} \approx 2.2 \cdot R_{PD} \cdot C_{LOAD}$$

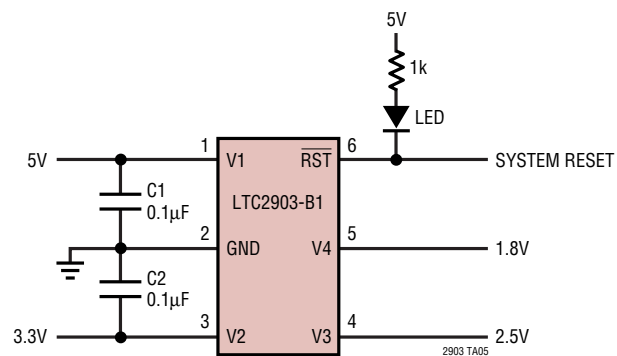
where R_{PD} is the on-resistance of the pull-down transistor and C_{LOAD} is the external load capacitance on the pin. At room temperature, the average R_{PD} is approximately 40Ω . With a 150pF load capacitance the reset line can pull down in about 13ns .

TYPICAL APPLICATIONS

Quad Supply Monitor with Adjustable Input

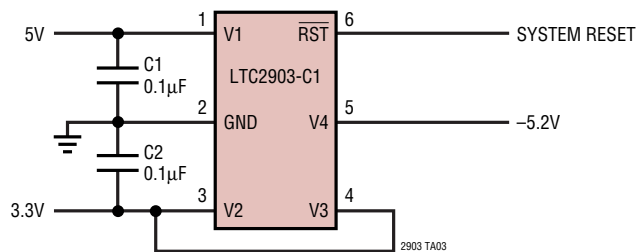


Fixed Quad Supply Monitor with LED Indication on $\overline{\text{RST}}$



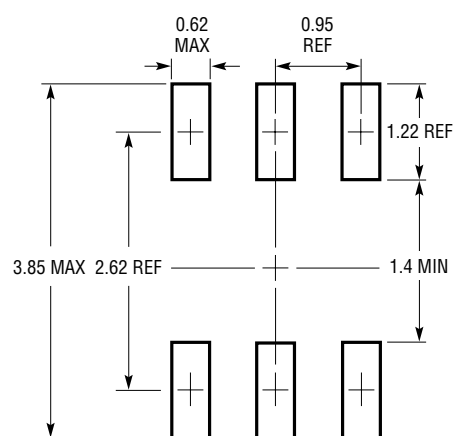
TYPICAL APPLICATIONS

Quad Supply Monitor with Unused Input Pulled Above Threshold

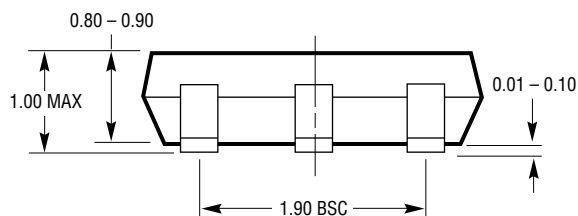
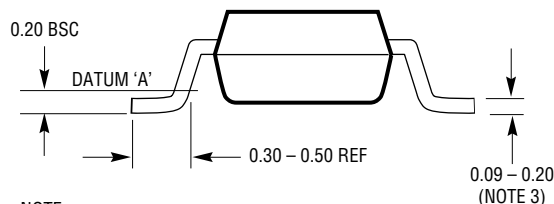
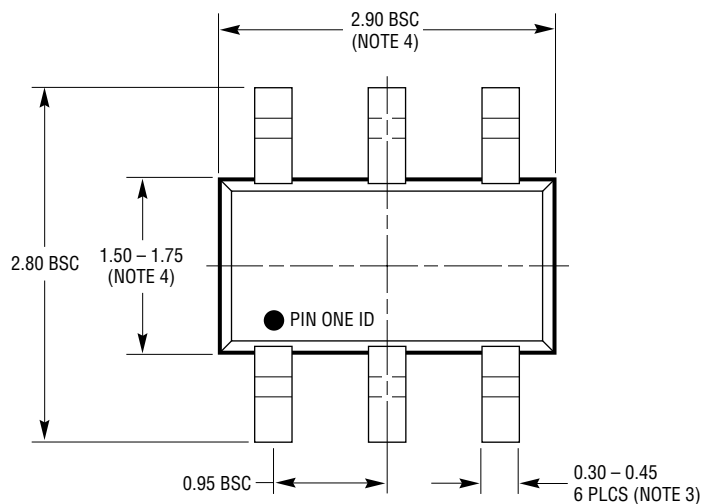


PACKAGE DESCRIPTION

S6 Package
6-Lead Plastic TSOT-23
 (Reference LTC DWG # 05-08-1636)



RECOMMENDED SOLDER PAD LAYOUT
 PER IPC CALCULATOR



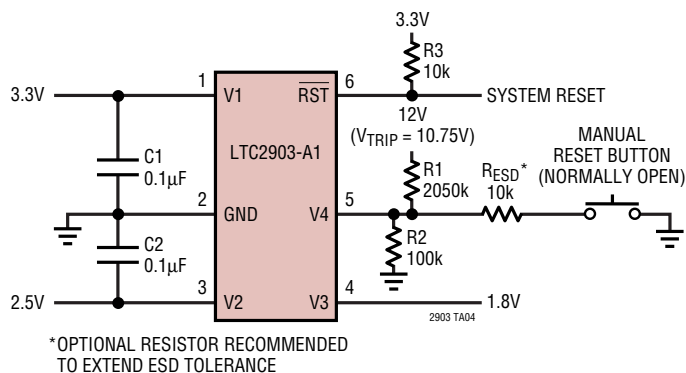
S6 TSOT-23 0302

NOTE:

1. DIMENSIONS ARE IN MILLIMETERS
2. DRAWING NOT TO SCALE
3. DIMENSIONS ARE INCLUSIVE OF PLATING
4. DIMENSIONS ARE EXCLUSIVE OF MOLD FLASH AND METAL BURR
5. MOLD FLASH SHALL NOT EXCEED 0.254mm
6. JEDEC PACKAGE REFERENCE IS MO-193

TYPICAL APPLICATION

Quad Supply Monitor with Manual Reset Button



RELATED PARTS

PART NUMBER	DESCRIPTION	COMMENTS
LTC690	5V Supply Monitor, Watchdog Timer and Battery Backup	4.65 Threshold
LTC694-3.3	3.3V Supply Monitor, Watchdog Timer and Battery Backup	2.9V Threshold
LTC699	5V Supply Monitor and Watchdog Timer	4.65 Threshold
LTC1232	5V Supply Monitor, Watchdog Timer and Pushbutton Reset	4.37V/4.62V Threshold
LTC1326	Micropower Precision Triple Supply Monitor for 5V, 3.3V and ADJ	4.725V, 3.118V, 1V Thresholds ($\pm 0.75\%$)
LTC1326-2.5	Micropower Precision Triple Supply Monitor for 2.5V, 3.3V and ADJ	2.363V, 3.118V, 1V Thresholds ($\pm 0.75\%$)
LTC1536	Precision triple Supply Monitor for PCI Applications	Meets PCI t_{FAIL} Timing Specifications
LTC1726-2.5	Micropower Triple Supply Monitor for 2.5V, 3.3V and ADJ	Adjustable $\overline{\text{RESET}}$ and Watchdog Time Outs
LTC1726-5	Micropower Triple Supply Monitor for 5V, 3.3V and ADJ	Adjustable $\overline{\text{RESET}}$ and Watchdog Time Outs
LTC1727-2.5/LTC1727-5	Micropower Triple Supply Monitors with Open-Drain Reset	Individual Monitor Outputs in MSOP
LTC1728-1.8/LTC1728-3.3	Micropower Triple Supply Monitors with Open-Drain Reset	5-Lead SOT-23 Package
LTC1728-2.5/LTC1728-5	Micropower Triple Supply Monitors with Open-Drain Reset	5-Lead SOT-23 Package
LTC1985-1.8	Micropower Triple Supply Monitor with Push-Pull Reset Output	5-Lead SOT-23 Package
LTC2900	Quad Voltage Monitor in MSOP	16 User Selectable Combinations, $\pm 1.5\%$ Threshold Accuracy
LTC2901	Quad Voltage Monitor with Watchdog	16 User Selectable Combinations, Adjustable Timers
LTC2902	Quad Voltage Monitor with $\overline{\text{RST}}$ Disable	16 User Selectable Combinations, Adjustable Tolerance
LTC2920-1/LTC2920-2	Single/Dual Power Supply Margining Controller	$< 0.4\%$ Margin Voltage Precision
LTC2921/LTC2922	Power Supply Trackers with Input Monitors	3 (LTC2921) and 5 (LTC2922) Remote Sense Switches
LTC2923	Power Supply Tracking Controller	Tracks Up and Down, Supply Sequencing