



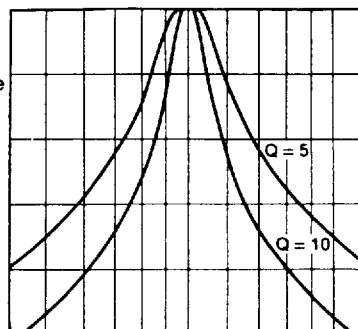
## FREQUENCY DEVICES INC

### FEATURES

- Tuneable over 1000:1 Frequency Range
- Filter  $f_0$  from 0.05Hz to 20kHz
- Constant Q over 3-Decade Tuning Range
- 2 Pole-Pair Butterworth Response
- Easy to Use and Available Now
- Stable, Well Defined Performance

### APPLICATIONS

- Contiguous Comb Filtering
- Vibration Studies
- RMS Measurements
- Harmonic Analysis
- Noise Reduction



### DESCRIPTION

Frequency Devices' 760 Series are resistive tuneable bandpass filters with versatile performance features. These units can be tuned over a 1000:1 frequency range, with each discrete center frequency externally selected by four nominal equal, matched 1% resistors. All models provide a two pole-pair Butterworth bandpass response which is independent of tuning and, therefore, capable of maintaining a constant Q over the three-decade frequency range.

Standard 760 Series models are available with tuning ranges of 0.05 to 50Hz, 0.5 to 500Hz and 20Hz to 20kHz. The filters are factory-calibrated so that, for most applications, the desired center-frequency tuning is the only adjustment needed.

These filters are ideal for the many applications requiring switch-tuneable center frequency selection across a three-decade spectrum. Applications requiring a variety of fixed center frequencies are equally well-served by the 760 Series.

### HOW TO ORDER

All 760 Series filters feature a well-defined and stable, two pole-pair **Butterworth** response. To order, simply select the model number tabulated below which provides the desired performance.

### AVAILABLE 760 SERIES FILTERS

Table 1.1

MODEL NUMBER	$f_0$ MAX Hz	TUNING RANGE (Hz)	Q	CASE
760BQ5 760BQ10	50	0.05-50	$5 \pm 10\%$ $10 \pm 10\%$	C-3
762BQ5 762BQ10	500	0.5-500	$5 \pm 10\%$ $10 \pm 10\%$	C-2
764BQ5 764BQ10	20K	20-20K	$5 \pm 10\%$ $10 \pm 10\%$	C-2



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**FREQUENCY DEVICES** 760 SERIES RESISTIVE TUNEABLE  
 2 POLE-PAIR BANDPASS  
 ACTIVE FILTERS

FREQUENCY DEVICES INC

**760 SERIES SPECIFICATIONS**

(Typical @ 25°C;  $V_S = \pm 15\text{Vdc}$ ;  $f_0 \leq f_{\text{max}}$  unless otherwise noted)

**PASSBAND CHARACTERISTICS**

**GAIN @  $f_0$**

Nominal Value  $0 \pm 0.3\text{dB}$ , Non-Inverting

Adjustment Range  $0 \text{ to } +10$  (Not in dB)

**CENTER FREQUENCY**

Tolerance<sup>1</sup>  $\pm 3\%$

Stability<sup>1</sup>  $\pm 0.03\%/^{\circ}\text{C}$

**INPUT CHARACTERISTICS**

Input Impedance  $100\text{k}\Omega$

Input Voltage Range  $\pm 10\text{V}$

**OUTPUT CHARACTERISTICS**

Linear Output Range @  $\pm 2\text{mA}$ <sup>2</sup>  $\pm 10\text{V}$

Output Noise<sup>3</sup>  $50\mu\text{V RMS}$

Output Offset Voltage<sup>4</sup>  $\pm 2\text{mV} \pm 20\mu\text{V}/^{\circ}\text{C}$

Output Resistance  $1\Omega$

**DC POWER SUPPLY ( $\pm V_S$ )**

Nominal Operating Voltage  $\pm 15\text{Vdc}$

Quiescent Current  $\pm 16\text{mA}$

Operating Voltage Range  $\pm 5\text{V to } \pm 18\text{V}$

**TEMPERATURE**

Operating  $0^{\circ}\text{C to } 70^{\circ}\text{C}$

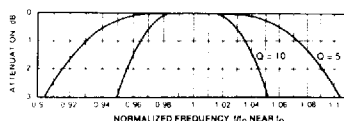
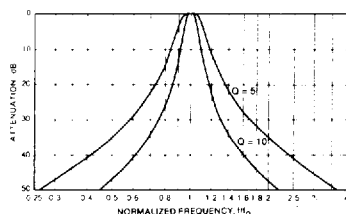
Storage  $-25^{\circ}\text{C to } +85^{\circ}\text{C}$

**NOTES**

- 1) Applicable when using matched 1%, 100ppm/ $^{\circ}\text{C}$  tuning resistors.
- 2) Output short circuit protected to ground.
- 3) Measured within a 1Hz to 100kHz bandwidth.
- 4) Offset adjustable to zero using external potentiometer and with  $R_X$  installed.


**FREQUENCY  
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**760 SERIES RESISTIVE TUNEABLE  
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ACTIVE FILTERS**

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**FREQUENCY DEVICES INC**
**NORMALIZED THEORETICAL RESPONSE**

**NORMALIZED THEORETICAL RESPONSE VALUES**

$f_i/f_0$	$Q = 5$		$Q = 10$	
	A(dB)	$\psi(^{\circ})$	A(dB)	$\psi(^{\circ})$
0.20	-55.21	176.6	-67.25	178.3
0.24	—	—	-63.76	177.9
0.25	-50.92	175.7	—	—
0.30	-47.24	174.7	-59.28	177.3
0.34	—	—	-56.61	176.9
0.35	-43.93	173.5	—	—
0.40	-40.85	172.3	-52.89	176.1
0.44	—	—	-50.52	175.6
0.45	-37.90	170.8	—	—
0.50	-35.00	169.1	-47.04	174.6
0.54	—	—	-44.72	173.8
0.55	-32.09	167.1	—	—
0.60	-29.09	164.6	-41.12	172.4
0.64	—	—	-38.60	171.2
0.65	-25.92	161.5	—	—
0.70	-22.48	157.2	-34.50	168.8
0.74	—	—	-31.46	166.6
0.75	-18.66	151.2	—	—
0.80	-14.25	141.9	-26.14	161.7
0.84	—	—	-21.82	156.3
0.85	-9.09	125.8	—	—
0.90	-3.51	94.4	-13.19	139.2
0.94	—	—	-5.25	106.9
0.95	-0.29	44.6	—	—
1.00	0.00	0.0	0.00	-000.0

USE OF THE NORMALIZED AMPLITUDE AND PHASE DATA is simple. If the center frequency ( $f_0$ ) exceeds a frequency of interest ( $f_i$ ), the normalized frequency  $f_n = f_i/f_0$  ( $f_i < f_0$ ). Alternatively,  $f_n = f_0/f_i$  for  $f_i > f_0$ . The phase shift ( $\psi$ ) will be the tabulated value for  $f_i < f_0$  and the negative of the tabulated value for  $f_i > f_0$ . Normalization is relative to the 0dB gain point.

A Q of 5 type 760 Series filter with a 50Hz  $f_0$ , would have a gain of -36dB at both 25Hz and 100Hz, because the normalized frequency ratios are 0.5 and 2.0, respectively.



## FREQUENCY DEVICES INC

## INSTALLATION NOTES

## TUNING PROCEDURE

All 760 Series models require a set of four nominally equal tuning resistors for each selected frequency; each resistor within a set tunes one of the four filter poles. A  $\pm 1\%$  match between set-resistors yields tuning accuracy closer than  $\pm 3\%$ .

Equation 4.1 calculates the common, theoretical value of the four equal-valued tuning resistors (R) shown connected in Fig. 4.1. The variables  $f_o$  and  $f_{\max}$  refer to a desired center frequency and the maximum permissible center frequency, respectively; values of  $f_{\max}$  for assorted 760 Series models are tabulated on Table 1.1 on this specification.

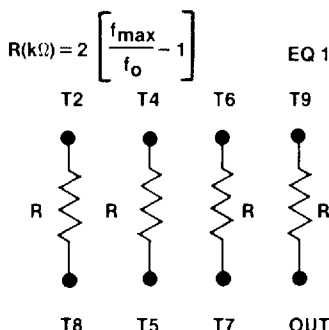


FIGURE 4.1: Connection of Resistive Tuning Components

For user convenience, Table 4.1 (below) lists the standard decade value closest to that calculated by Eq. 1.

The recommended procedure for **TROUBLE FREE START UP REQUIRES USE OF THE SHORTEST POSSIBLE CONNECTING LEADS BETWEEN FILTER TERMINALS T5, T7, T8 AND T9 AND THEIR ASSOCIATED TUNING RESISTORS (R's)**. Careful circuit layout at this juncture will provide multiple-frequency selection capability for immediate or future applications.

1.00	1.21	1.47	1.78	2.15	2.61	3.16	3.83	4.64	5.62	6.81	8.25
1.02	1.24	1.50	1.82	2.21	2.67	3.24	3.92	4.75	5.76	6.98	8.45
1.05	1.27	1.54	1.87	2.26	2.74	3.32	4.02	4.87	5.90	7.15	8.66
1.07	1.30	1.58	1.91	2.32	2.80	3.40	4.12	4.99	6.04	7.32	8.87
1.10	1.33	1.62	1.96	2.37	2.87	3.48	4.22	5.11	6.19	7.50	9.09
1.13	1.37	1.65	2.00	2.43	2.94	3.57	4.32	5.23	6.34	7.68	9.31
1.15	1.40	1.69	2.05	2.49	3.01	3.65	4.42	5.36	6.49	7.87	9.53
1.18	1.43	1.74	2.10	2.55	3.09	3.74	4.53	5.49	6.65	8.06	9.76

TABLE 4.1: Tabulation of 1% Resistor Standard Decade Values



## FREQUENCY DEVICES INC

### INSTALLATION NOTES, Continued

#### OPTIONAL GAIN ADJUSTMENT

The gain of all 760 Series filters can be adjusted resistively when required by the application. This enables the user to adjust the filter passband gain at the center frequency  $f_0$  to a level above or below the non-inverting, factory-preset value of  $0 \pm 0.3\text{dB}$ .

The  $20\text{k}\Omega$  cermet potentiometer of Figure 5.1 provides a linear gain ratio range of 0 (zero gain) to  $\pm 10$  (20dB). The recommended type of potentiometer provides sufficient resolution and temperature stability for the intended circuit function.

Alternatively, the fixed-resistor approach of Figure 5.2 affords a simple and economical method for incrementally controlling gain. Resistors R1 and R2 form a divider network in which R1 increases gain, R2 decreases gain. These resistors should

be 1%, 100ppm/ $^{\circ}\text{C}$  metal film components, the values of which can be determined by the following experimental method:

First, connect a decade box in place of R1 if gain is to be increased or in place of R2 if gain is to be decreased. Next, adjust the resistance until the desired passband gain is obtained. Finally, replace the decade box with the discrete resistor value called for.

When a 760 Series filter is adjusted for a gain of greater than 0dB, the input signal amplitude must be restricted accordingly. The additional gain could cause a  $\pm 10\text{V}$  input signal to drive the filter output stage beyond its linear operating region and produce a clipped output signal. Note that if the factory set gain level is satisfactory, no external gain connections are required.

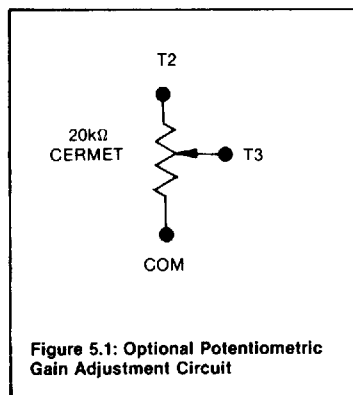


Figure 5.1: Optional Potentiometric Gain Adjustment Circuit

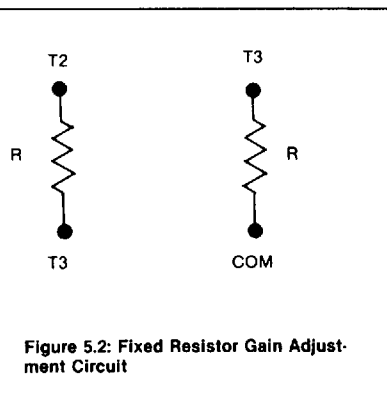


Figure 5.2: Fixed Resistor Gain Adjustment Circuit

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INSTALLATION NOTES, Continued

OPTIONAL DC OFFSET ADJUSTMENT

The filter input bias current flowing through a portion of the external tuning resistor network generates a DC offset voltage. For the typical bandpass filtering application in which DC contains no useful information and the filter output can be AC coupled, this offset presents no problem. The offset can, however, represent a significant amplitude measurement error relative to the rectified and filtered value of an AC input signal. For acceptable system accuracy, the DC offset error voltage should be much less than the DC equivalent value of the AC input signal.

Connected as shown in Figure 6.1, resistor  $R_x$  can compensate the bias current generated offset voltage as well as its temperature drift. Note that if no compensation is required,  $R_x = 0\Omega$  and **FILTER TERMINAL T10 MUST BE CONNECTED TO FILTER TERMINAL T11.**

Figure 6.2 illustrates the typical form and degree of compensation that resistive-balancing can achieve. The graph shows that the compensation resistor  $R_x$  equal to tuning resistance  $R$  provides an approximate tenfold improvement of output offset and offset drift relative to the uncompensated ( $R_x = 0$ ) case. Here,  $R$  refers to the common value of the four nominally equal tuning resistors, which must increase for lower frequencies and decrease for higher selected frequencies. For offset critical multi-frequency applications, a new value of  $R_x = R$  should accompany the  $R$  for each frequency selected.

Alternatively, the circuit of Figure 6.3 can zero the DC output offset, but only for a single combination of fixed frequency and temperature. In other words, this approach does not compensate, it nulls.

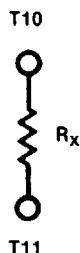


FIGURE 6.1: Connection of Offset Compensation Resistor  $R_x$

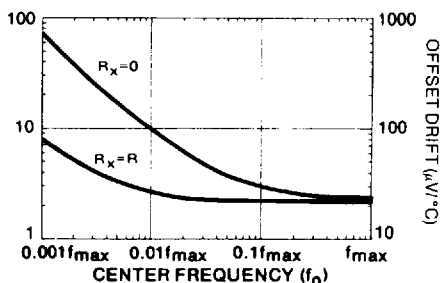


Figure 6.2: Typical Output Offset Voltage and Offset Drift vs.  $f_0$  and  $R_x$

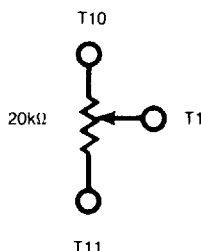


Figure 6.3: Offset Voltage Null Circuit



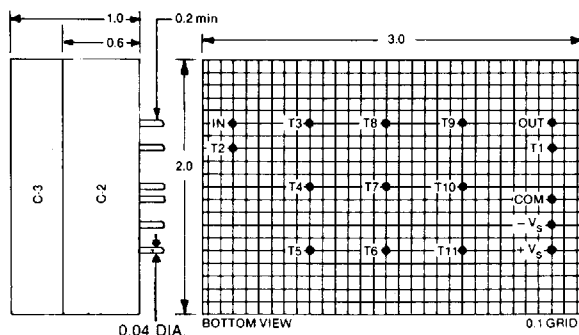
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**DIMENSIONS AND CONNECTIONS**

**FREQUENCY DEVICES INC**

**CASE DIMENSIONS**

**DIMENSIONS IN INCHES (Use S1002 Socket)**

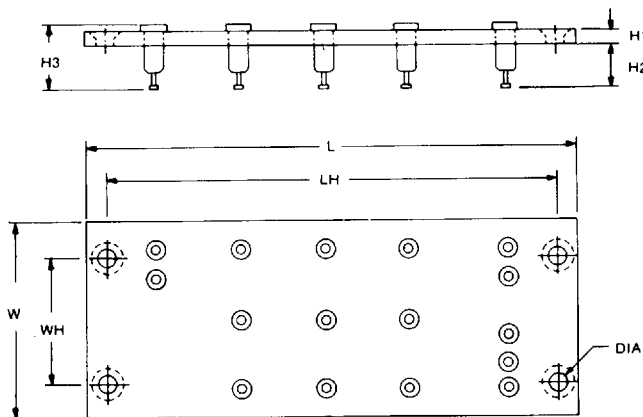


**TERMINAL KEY**

IN <sup>1</sup>	Signal Input
OUT <sup>1</sup>	Signal Output, Tuning
T1	Offset Adjust
T2 <sup>2</sup>	Gain Adjust, Tuning
T3	Gain Adjust
T4 and T5 <sup>2</sup>	Frequency Tuning Terminal — Pair No. 1
T6 and T7 <sup>2</sup>	Frequency Tuning Terminal — Pair No. 2
T2 <sup>1</sup> and T8 <sup>2</sup>	Frequency Tuning Terminal — Pair No. 3
OUT <sup>1</sup> and T9 <sup>2</sup>	Frequency Tuning Terminal — Pair No. 4
T10 and T11	Offset Compensation Terminal — Pair (for R <sub>x</sub> )
+V <sub>S</sub>	Supply Voltage, Positive
COM	Ground, Supply Common
-V <sub>S</sub>	Supply Voltage, Negative

NOTES: 1) Serves two simultaneous functions.

2) Requires shortest possible connecting leads for minimum pickup.


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**SOCKET DATA/FILTER ORDERING GUIDE**
**SOCKET DATA**
**S1002**


DIMENSION	MILLIMETERS	INCHES
L	89	3.5
LH	81	3.2
W	36	1.4
WH	23	0.9
H1	2.3	0.09
H2	7.9	0.31
H3	12	0.47
DIA	3.5	0.14

**FREQUENCY DEVICES INC**
**ORDERING GUIDE**

Any of the six standard 760 Series models can be ordered as listed in Table 1.1 on Page 1 of this specification. The S1002 socket can be ordered as a separate line item.

For special OEM requirements please contact FDI directly. Frequency Devices, Inc., can furnish standard and custom design active filters which, in OEM quantities, meet or exceed the price/performance capabilities of many competitive models.