PAL22RX8A

T-46-13-47

High Speed Programmable Array Logic

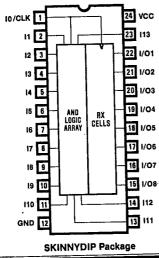
Features/Benefits

- Programmable flip-flops allow J-K, S-R, T or D-types for the most efficient use of product terms
- 8 input/output macrocells for flexibility
- Programmable registered or combinatorial outputs
- Programmable output polarity
- Global register asynchronous preset/asynchronous reset
- Automatic register reset on power up
- Preloadable output registers for testability
- High speed at 25 ns tpp, 28.5 MHz fMAX
- Space-saving 24-pin 300-mil SKINNYDIP® package or 28-pin chip carrier

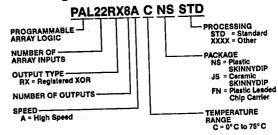
General Description

The PAL22RX8A is a high-density Programmable Array Logic (PAL®) device which implements a sum-of-products transfer function via a user-programmable AND logic array and a fixed OR logic array. Featured are eight highly flexible input/output macrocells which are user-configurable for combinatorial or registered operation. Each flip-flop can be programmed to be either a J-K, S-R, T, or D-type for optimal design of state machines and other synchronous logic. The PAL22RX8A is a functional superset of, and pin-compatible with, the PAL24A Series, but can implement many new functions due to its added features. Supplied in space-saving 300-mil-wide dual in-line packages or 28-pin chip carriers, the PAL22RX8A offers a

Pin Configurations



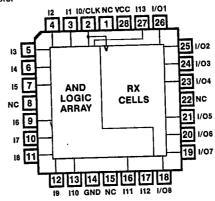
Ordering Information



powerful, space-saving alternative to SSI/MSI logic devices, while providing the advantage of instant prototyping. Security fuses defeat readout after programming and make proprietary designs difficult to copy.

The PAL22RX8A is fabricated using Monolithic Memories' advanced bipolar process for high speed and low power. TiW fuse links provide high reliability and programming yields. Special on-chip test circuits allow full AC, DC, and functional testing before programming. Preloadable output registers facilitate functional testing.

The PAL22RX8A can be programmed on standard PAL device programmers, fitted with appropriate programming modules and configuration software. Design development is supported by Monolithic Memories' PALASM® 2 software as well as by other programmable logic CAD tools available from third party vendors.



Plastic Leaded Chip Carrier

Package Drawings

(refer to PAL Device Package Outlines, page 3-179)

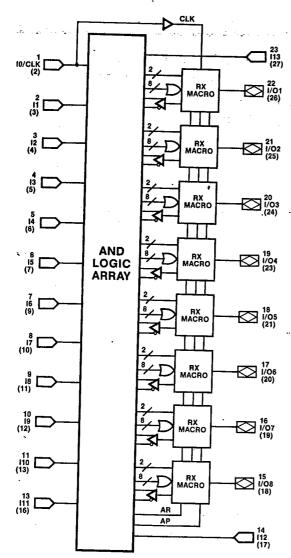
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Block Diagram



Note: PLCC pin numbers are indicated in parentheses. PLCC pins 1, 8, 15 and 22 are not connected.

Description of Architecture

The PAL22RX8A has fourteen dedicated inputs and eight programmable I/O macrocelis. Pin 1 serves either as an array input or as a clock for all flip-flops. Buffers for device inputs have complementary outputs to provide user-programmable input signal polarity. The fuse matrix implements a programmable AND logic array, which drives a fixed OR logic array.

The high level of flexibility built into each macrocell, shown in Figure 1, allows the PAL22RX8A to implement several different architecture options. Each macrocell can be individually programmed to implement a variety of combinatorial or registered logic functions.

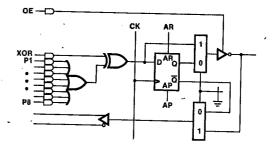


Figure 1, PAL22RX8A Macrocell

Programmable Flip-Flops

Each output macrocell contains a unique programmable flipflop consisting of a basic D-type flip-flop driven by an XOR gate. This allows the user to choose the optimal flip-flop for the design, since either J-K, S-R, or T-type flip-flops can be synthesized from such a structure without wasting product terms.

As indicated in the macrocell logic diagram, one input of the XOR gate is connected to a single product term, while the second input is connected to the output of the OR logic array. The XOR gate output feeds the input of the D flip-flop. The way in which the XOR gate is used to synthesize the different flip-flop types is described in detail below.

D Filp-Fiop. The D flip-flop option is implemented directly. In this configuration, the XOR gate on the input of the flip-flop can be used to program the logic polarity of the transfer function.

J-K Flip-Flop. The J-K flip-flop option can be easily synthesized with a more sophisticated manipulation of the XOR gate inputs and the D flip-flop output.

The transfer function of a J-K flip-flop can be mapped in the Karnaugh Map of Figure 2, where Q+ represents the next state of the flip-flop:

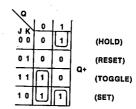


Figure 2. J-K Flip-Flop Transfer Function

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Dropping the (+) for simplicity, the equivalent Boolean expression for Q+ is:

In general, J and K can be sum-of-product expressions which are provided in the PAL architecture only in active-high form. Thus, a direct implementation of \overline{K} expressions must invoke a DeMorgan transformation, which can use excessive product terms. This can be avoided by rewriting the equation for Q without inversions on the J or K inputs.

The XOR gate can be used to construct a logically equivalent expression without any inversions on the J or K inputs. The rewritten Boolean expression is:

To check that these expressions are logically equivalent, change the XOR to its equivalent sum of products form (remember $A : +: B = A \cdot B + A \cdot B$) and reduce (using DeMorgan's theorem):

$$\begin{array}{lll} Q := Q^*(\overline{J^*\overline{Q}} + \overline{K^*Q}) & + \overline{Q}^*(\overline{J}^*\overline{Q} + \overline{Q}^*K^*Q) & \\ Q := Q^*(\overline{J^*\overline{K}} + \overline{J^*\overline{Q}} + Q^*\overline{K} + Q^*\overline{Q}) & + \overline{Q}^*J^*\overline{Q} + \overline{Q}^*K^*Q & \\ Q := \overline{J^*\overline{K}^*Q} + \overline{K^*Q} + \overline{J^*\overline{Q}} & + \overline{Q}^*J^*\overline{Q} + \overline{Q}^*K^*Q & \\ \end{array}$$

which simplifies to

Since J and K are, in general, sums of products, J and K in either expression can be substituted with (J1+J2+...+Jm) and (K1+K2+...+K8-m), where 8 is the total number of product terms associated with a given output macrocell. Thus, the total 8 product term resource is shared between the J and K control inputs (Figure 3). Note that all J terms will contain \overline{Q} and all K terms will contain \overline{Q} .

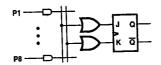


Figure 3. J-K Filp-Flop Logic Equivalent; J and K Can Also be Active-Low

The above discussions have assumed that it was most convenient to "group ones" in the Karnaugh Map. Sometimes it takes fewer product terms to "group zeros", i.e., implement the inversion of the desired function. The equations shown in Table 1 are equivalent and can be interchanged to optimize product term utilization. This can be readily proved through logic reductions similar to that above.

J and K active high	Q := Q :+: (J*Q + K*Q)
J active high, K active low	Ŏ:= 1, <u>Ö</u> + <u>K</u> ,O
J active low, K active high	Q := J•Q + K•Q
J and K active low	Q := Q :+: (J*Q + K*Q)

Note: J = sum of products J1 + J2 + ... + Jm

K = sum of products K1 + K2 + ... + K8-m

8 = total number of available product terms for a given macrocell

Table 1. J-K Filp-Flop Transfer Functions

S-R Flip-Flop. The S-R flip-flop has a truth table identical to that of the J-K flip-flop, with the exception that the J=K=1 (toggle) condition is not allowed. The S-R flip-flop implementation is identical to that of the J-K flip-flop, with J-K replaced by S-R, and the S=R=1 condition avoided.

T Flip-Flop. A T (toggle) flip-flop either holds its state or toggles, depending on the logic state of the T input. The T flip-flop is a subset of the J-K flip-flop and can be considered equivalent to a J-K type with J=K. The general transfer function and its active-low T equivalent are both given in Table 2.

Q := Q :+: T	•
Q := Q :+: T	

Note: T = sum of products T1 + T2 + T3 + ... + T8.

Table 2. T Flip-Flop Transfer Functions

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Summary

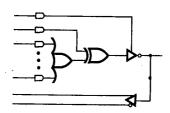
The PAL22RX8A can synthesize J-K, S-R, T, and D flip-flops, whichever is most convenient for the application, without sacrificing product terms. Additionally, the synthesized equations can use the active-high or active-low forms of the inputs, allowing the designer to minimize product term requirements.

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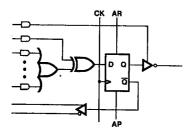
Flip-Flop Bypass

Any output in the PAL22RX8A can be configured as combinatorial by bypassing the output flip-flop. This is done by setting the output multiplexer to the appropriate state. The multiplexer is controlled by a single architecture fuse which, when intact, selects a registered output with register feedback and, when opened, selects a combinatorial output with feedback from the output pin.

Combinatorial I/O



Registered Output



Programmable I/O

Each macrocell has a three-state output buffer with programmable three-state control. Control is implemented by a single product term, allowing specification of enable/disable functions controlled by any device input or output. Each macrocell can be configured as a dedicated input by selecting the combinatorial output configuration and disabling the buffer drive capability.

Programmable Polarity

The polarity of each macrocell output can be set active high or active low.

Combinatorial Outputs. The XOR gate provides polarity control for combinatorial outputs, with the single product term to the XOR gate controlling the invert/not invert function. With all fuses intact, there is no inversion through the XOR gate, creating an active high output.

Registered Outputs. For D-type registered outputs, polarity can be set by the XOR gate, as is the case with combinatorial outputs. Using this method to set polarity, preset and reset will not be affected.

Polarity options for J-K, S-R, and T flip-flops have been discussed in the section on programmable flip-flops.

Programmable Preset and Reset

The eight macrocell flip-flops share common programmable preset and reset control for easy system initialization. The Q outputs of the register will go to the logic low state independent of the clock when the asynchronous reset (AR) product term is asserted. The register will be forced to the logic high state independent of the clock when the asynchronous preset (AP) product term is asserted.

Power-Up Reset

All flip-flops power up to a logic low for predictable system initialization. The power-up state of the flip-flop is not affected by the output polarity option chosen. Due to the output inverter, the outputs will power up to a logic high. See waveform at end of TTL/CMOS PAL Devices section.

Register Preload

The register on the PAL22RX8A can be preloaded by use of supervoltages to facilitate functional testing of complex state machine designs. This feature allows direct loading of arbitrary states, thereby making it unnecessary to cycle through long test vector sequences to reach a desired state. In addition, transitions from illegal states can be verified by preloading illegal states and observing proper recovery.

Security Fuses

After programming and verification, a PAL22RX8A design can be secured by programming the security fuses. Once programmed, these fuses defeat readback of the internal fuse pattern by a device programmer, making proprietary designs very difficult to copy.

Quality and Testability

The PAL22RX8A offers a very high level of built-in quality. Special on-chip test circuitry provides a means of verifying performance of all AC and DC parameters prior to programming. In addition, these built-in test paths verify complete functionality of each device to provide the highest post-programming functional yields in the industry.

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Absolute Maximum Ratings Supply voltage VCC			Operating	Pr	ogramming
Supply voltage VCC	•••••	 	0.5 V to 7 V),5 V to 12 V I,0 V to 12 V
Storage temperature		 		,03	C 10 + 150 C

Operating Conditions

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SYMBOL	PARAMETER		MIN	TYP	MAX	UNIT
VCC Supply voltage		4.75	5	5.25	V	
t _w Width of clock	Low	10	7		ns	
	Width of clock	High	10	7		ļ
t _{su}	Setup time from input or feedback to clock		20	16		ns
th	Hold time	*	0	-10		ns
taw	Asynchronous preset/reset width		. 15	10		ns
tar	Asynchronous preset/reset recovery time		30	25		ns
T _A	Operating free-air temperature		0	25	75	°C

Electrical Characteristics Over Operating Conditions

SYMBOL	PARAMETER	TEST CONI	TEST CONDITION		TYP	MAX	UNIT
V _{IL} 2	Low-level input voltage					0.8	٧
V _{IH} 2	High-level input voltage			2.0			٧
V _{IC}	Input clamp voltage	V _{CC} = MIN I ₁	= -18 mA		-0.8	-1.5	V
I _{IL} 3	Low-level input current	V _{CC} = MAX V	I = 0.4 V		-0.1	-0.25	mA
I _{IH} 3	High-level input current	V _{CC} = MAX V	1 = 2.4 V			25	μA
1 ₁	Maximum input current	V _{CC} = MAX V	= 5.5 V			100	μA
v _{OL}	Low-level output voltage	V _{CC} = MIN	OL = 24 mA		0.35	0.5	V
VOH	High-level output voltage	V _{CC} = MIN	OH = -3.2 mA	2.4	3.4		V
lozL ³		\ MAY	O = 0.4 V			-100	μΑ
IOZH ³	Off-state output current	V _{CC} = MAX	O = 2.4 V	_		100	μΑ
los ⁴	Output short-circuit current	V _{CC} = 5 V	O = 0 V	-30	-70	-130	m/
lcc	Supply current	V _{CC} = MAX			180	210	mA
		V = 00V ot 6 = 1 MUz	DIP pins 1, 13		15		_
CIN	Input capacitance	V _{IN} = 2.0 V at f = 1 MHz	All other inputs		12		pF
COUT	Output capacitance	V _{OUT} = 2.0 V at f = 1 MHz			12		

Notes. 1. The PAL22RX8A is designed to operate over the full military operating conditions. For availability and specifications, contact Monolithic Memories.

3. I/O pin teakage is the worst case of I_{IL} and I_{OZL} (or I_{IH} and I_{OZH}).

These are absolute voltages with respect to the ground pin on the device and include all overshoots due to system and/or tester noise.Do not attempt to test these values without suitable equipment.

^{4.} No more than one output should be shorted at a time and duration of the short circuit should not exceed one second.

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Switching Characteristics Over Operating Conditions

SYMBOL	PARAMETER		TEST CONDITIONS		MMER TYP	CIAL MAX	UNIT
t _{PD}	Input or feedback to output				20	25	ns
t _{CLK}	Clock to output or feedback	,,_,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,			12	15	ns
t _{EA}	Input to output enable	Input to output enable Input to output disable Asynchronous preset/reset to output			20	25	ns
tER	Input to output disable				18	25	ns
t _{AP}	Asynchronous preset/reset to				25	35	ns
f _{MAX} Maximum frequency	External feedback		28.5	40			
	waxiinum irequency	No feedback		50	55		MHz

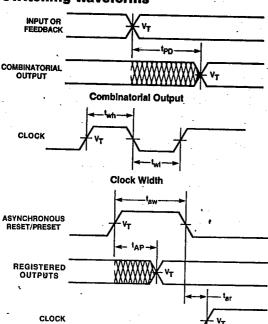
Switching Test Load (refer to page 5-164)

Power-Up Reset Waveform (refer to page 5-164)

Schematic of Inputs and Outputs

(rèfer to page 5-164)

Switching Waveforms



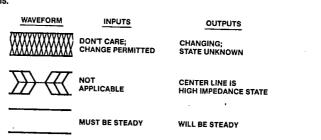
Output Register Preload

The preload function allows the register to be loaded from the output pins. This feature aids functional testing of sequential designs by allowing direct setting of output states. The procedure is:

Asynchronous Reset/Preset

- 1. Raise V_{CC} to 5.0 V.
- 2. Set pin 1 (CLK) to V_{IL}.
- 3. Disable output registers by setting pin 5 to V_{PL} (18.0 V \pm 0.5 V).
- Apply the desired level (V_{IL}/V_{IH}) to all registered output pins. Leave combinatorial outputs floating.
- 5. Pulse pin 11 to $V_{\mbox{\scriptsize PL}}$, then back to 0 V.
- 6. Remove V_{IL}/V_{IH} from all output registers.
- 7. Lower pin 5 to VIL.
- 8. Enable output registers per programmed pattern.
- 9. Verify for VOL/VOH at all registered output pins.

Key to Timing Diagrams

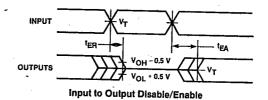


CLOCK

CLOCK

REGISTERED
OUTPUT

Registered Output

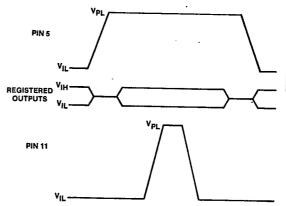


Notes:

- 1. VT = 1.5 V
- 2. Input pulse amplitude 0 V to 3.0 V
- Input rise and fall times 2-5 ns typical

Programmers/Development Systems

(refer to Programmer Reference Guide, page 3-81)



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Logic Diagram

DIP (PLCC) Pinouts

PAL22RX8A

