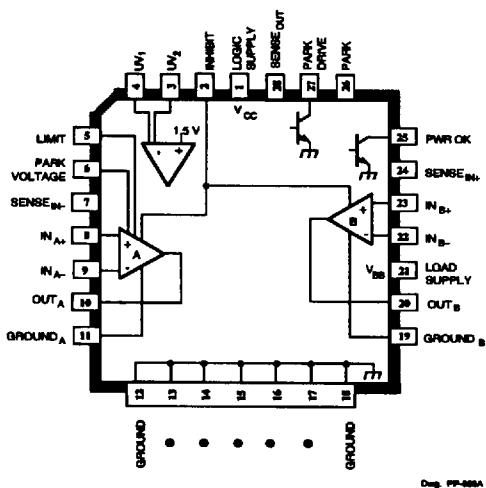


# 8958

## VOICE COIL MOTOR DRIVER

A8958CEA



### ABSOLUTE MAXIMUM RATINGS at $T_A = 25^\circ\text{C}$

Supply Voltages, $V_{BB}$ and $V_{CC}$ .....	16 V
Output Current, $I_{OUT}$ .....	$\pm 1.0 \text{ A}$
Park Drive Output Current, $I_{PARK}$	
Continuous .....	250 mA
Peak .....	1.0 A
Amplifier Input Voltage Range,	
$V_{IN}$ .....	-2.0 V to $V_{CC}$
Sense Input Voltage Range,	
$V_{SENSE\ IN}$ .....	-0.3 V to $V_{CC}$
Comparator and Digital Inputs,	
$V_{IN}$ .....	-0.3 V to 10 V
$I_{IN}$ .....	$\pm 10 \text{ mA}$
Power OK Output, $V_{CEX}$ .....	20 V
$I_C$ .....	30 mA
Output Clamp Diode Current,	
$I_F$ (pulsed) .....	1.0 A
Package Power Dissipation, $P_D$ .. See Graph	
Operating Temperature Range,	
$T_A$ .....	0°C to +70°C
Junction Temperature, $T_J$ .....	150°C*
Storage Temperature Range,	
$T_S$ .....	-55°C to +150°C

\* Fault conditions that produce excessive junction temperature will activate device thermal shutdown circuitry. These conditions can be tolerated but should be avoided.

Providing control and drive of the voice coil motor used for head positioning in disk drive applications, the A8958— is a full-bridge driver which can be configured so that its output current is a direct function of an externally applied control voltage or current. This linear current control function is supplemented by additional circuitry to protect the heads and the data disk during system failure or normal system shutdown.

The two  $\pm 800 \text{ mA}$  driver outputs provide very-low saturation voltage drops and precise current control utilizing a single current-sensing resistor connected in series with the load. Under-voltage lockout disables the system in a controlled sequence if a fault condition occurs.

When activated by the under-voltage comparator, or a park command, the output power drivers change from a controlled current to a user-determined constant park voltage. Other features include a power ok flag, a limit input to force the outputs to their maximum level in either polarity, an over-riding output disable to shut down both power amplifiers and reduce quiescent supply current, and internal thermal shutdown which disables the load (but still allowing the head to be parked) in the event of excessive junction temperatures. The load is re-enabled when the junction temperature returns to a safe level.

The A8958CEA is supplied in a 28-lead power PLCC for surface-mount applications; the A8958CLB is supplied in a 24-lead power SOIC. The copper half-batwing/batwing construction provides for maximum package power dissipation in a minimum package size. Both are rated for continuous operation over the temperature range of 0°C to +70°C.

### FEATURES

- Controlled-Velocity Head Parking
- Zero Deadband
- High Transconductance Bandwidth
- User-Adjustable Transconductance Gain
- $\pm 800 \text{ mA}$  Load Current
- Dual Under-Voltage Monitors with Flag and User-Selectable Trip Points
- Internal Thermal Shutdown Circuitry
- Replaces UC3175

Always order by complete part number:

Part Number	Package
A8958CEA	28-Lead Half-Batwing PLCC
A8958CLB	24-Lead Batwing SOIC

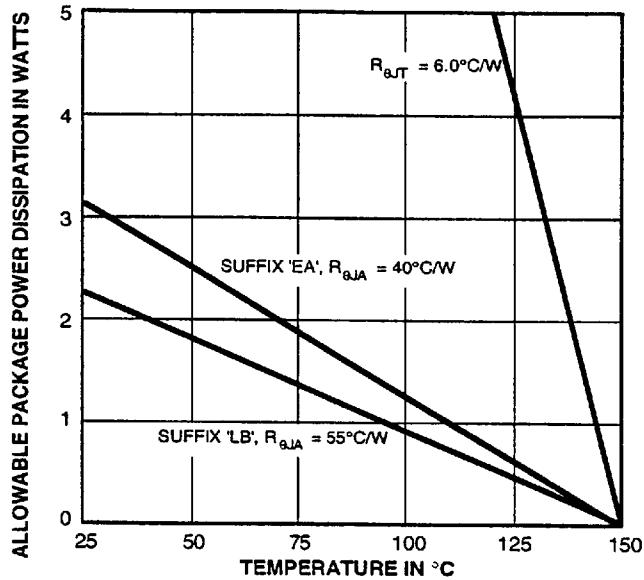
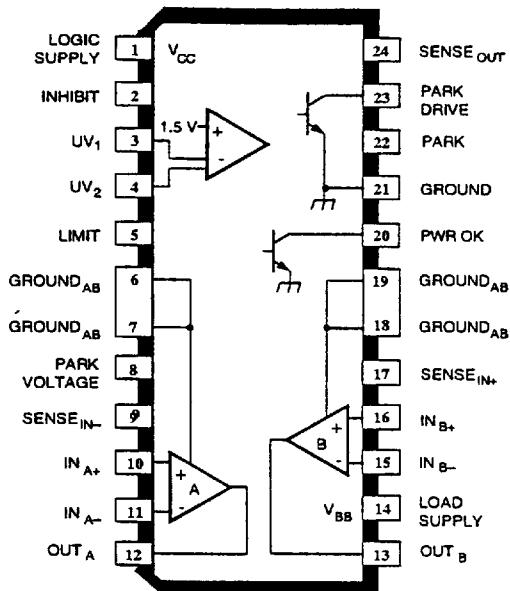


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A8958

## VOICE COIL MOTOR DRIVER

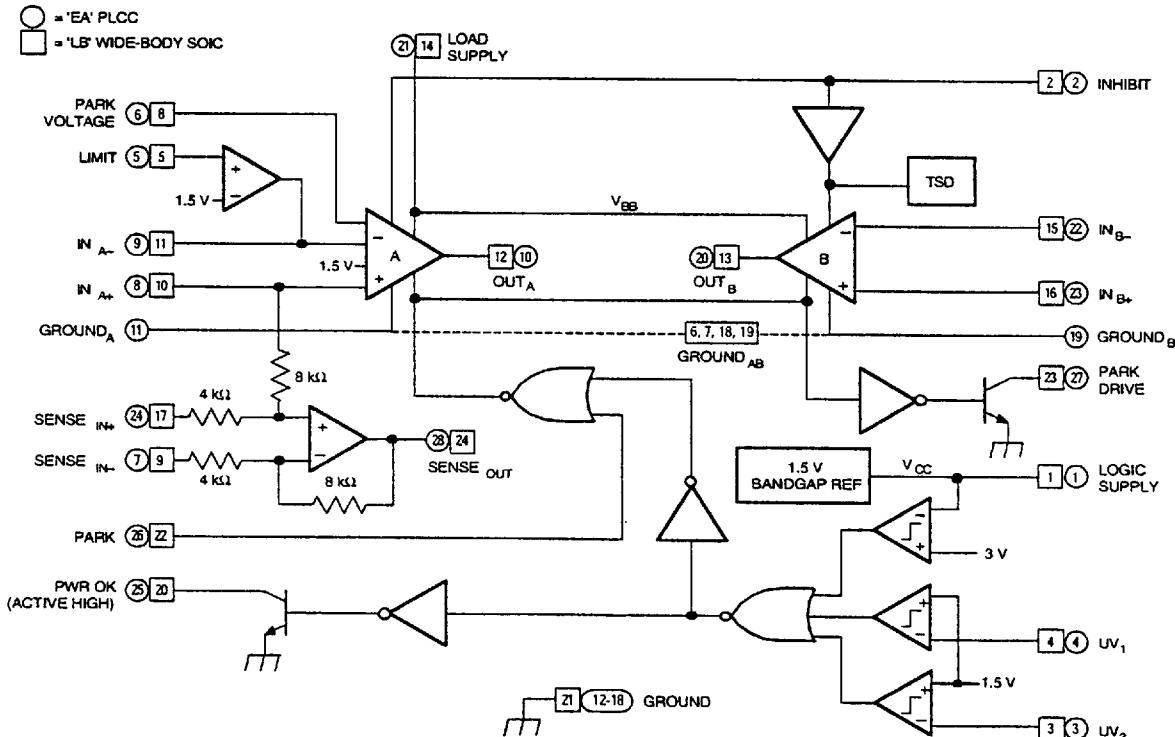
## A8958CLB



Dwg. GP-049-1

Dwg. PP-054

## FUNCTIONAL BLOCK DIAGRAM



Dwg. FP-020B

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**ELECTRICAL CHARACTERISTICS at  $T_A = +25^\circ\text{C}$ ,  $V_{CC} = V_{BB} = 12 \text{ V}$** 

Characteristic	Symbol	Test Conditions	Limits			
			Min.	Typ.	Max.	
Logic Supply Voltage Range	$V_{CC}$	Operating	8.0	12	16	V
Logic Supply UV Threshold	$V_{CC}$	High-to-low transition	—	2.8	3.0	V
Logic Supply UV Hysteresis	$\Delta V_{CC}$		—	200	—	mV
Supply Current	$I_{BB}$	$V_{OUT} = 6 \text{ V}$ , no load	—	2.0	—	mA
	$I_{CC}$		—	23	—	mA
Inhibited Supply Current	—	$I_{BB} + I_{CC}$ , $V_2 \geq 1.7 \text{ V}$	—	3.0	8.0	mA
Thermal Shutdown Temp.	$T_J$		—	165	—	$^\circ\text{C}$
Thermal Shutdown Hysteresis	$\Delta T_J$		—	8.0	—	$^\circ\text{C}$

**Output Power Drivers**

Output Saturation Voltage	$V_{SAT}$	$I_{OUT} = 250 \text{ mA}$	—	250	—	mV
		$I_{OUT} = 800 \text{ mA}$	—	450	—	mV
		$I_{OUT} = -250 \text{ mA}$	—	750	—	mV
		$I_{OUT} = -800 \text{ mA}$	—	950	—	mV
Total Saturation Voltage (Source + Sink)	$V_{SAT}$	$I_{LOAD} = 250 \text{ mA}$	—	1.0	1.4	V
		$I_{LOAD} = 800 \text{ mA}$	—	1.4	2.0	V
Input Offset Voltage	$V_{IO}$	$V_{CM} = 6 \text{ V}$	—	5.0	8.0	mV
Input Offset Drift	$\Delta V_{IO}$		—	—	25	$\mu\text{V}/^\circ\text{C}$
Input Bias Current	$I_{IN}$	Except $IN_{A+}$ , $V_{CM} = 6 \text{ V}$	—	-150	-500	nA
		$IN_{A+}$ to SENSE $_{IN+} = 12 \text{ k}\Omega$ , $T_J = 25^\circ\text{C}$	69	84	105	$\mu\text{A}/\text{V}$
Input Offset Current	$I_{IO}$	$IN_B$ only, $V_{CM} = 6 \text{ V}$	—	—	200	nA
Differential Sense Input Current	$I_{ID}$	$I_{OUT} = 5 \text{ mA}$	—	$\pm 300$	—	$\mu\text{A}$
		$I_{OUT} = 500 \text{ mA}$	—	3.0	—	mA
Large Signal Gain	$A_{VS}$	$V_{OUT} = 2 \text{ V to } 10 \text{ V}$ , $I_{OUT} = \pm 500 \text{ mA}$	1.5	5.0	—	$\text{V}/\text{mV}$
Slew Rate	SR		—	4.0	—	$\text{V}/\mu\text{s}$
Unity Gain Bandwidth	BW	Amplifier A	0.5	1.0	1.7	MHz
		Amplifier B	0.5	2.0	2.2	MHz
Common-Mode Rejection	$k_{CMR}$	$V_{CM} = 1 \text{ V to } 10 \text{ V}$	70	90	—	dB
Clamp Diode Forward Voltage	$V_F$	$I_F = 800 \text{ mA}$ , $V_2 \geq 1.7 \text{ V}$	—	1.0	1.2	V
High-Side Current Limit	$I_{OUT}$	$T_J = 25^\circ\text{C}$	—	1.0	1.2	A
Power Supply Rejection	$k_{SVR}$	$V_{CC} = 4 \text{ V to } 15 \text{ V}$ , $V_{CM} = 1.5 \text{ V}$	70	90	—	dB

Negative current is defined as coming out of (sourcing) the specified device terminal.

Continued next page...

Typical Data is for design information only.

**ELECTRICAL CHARACTERISTICS (continued)**

Characteristic	Symbol	Test Conditions	Limits			
			Min.	Typ.	Max.	
<b>Current Sense Amplifier</b>						
Input Offset Voltage	$V_{IO}$	$V_{CM} = 6 \text{ V}$	—	—	2.0	mV
Input Offset Drift	$\Delta V_{IO}$	$V_{CM} = 0 \text{ V to } 12 \text{ V}$	—	—	3000	$\mu\text{V/V}$
			—	—	8.0	$\mu\text{V}/^\circ\text{C}$
Voltage Gain	$A_{VS}$	$V_{ID} = -1 \text{ V to } +1 \text{ V}, V_{CM} = 6 \text{ V}$	1.95	2.00	2.05	—
Output Saturation Voltage	$V_{SAT}$	$V_{OUT}, I_{OUT(SINK)} = 1.5 \text{ mA}$	—	300	500	mV
		$V_{CC} - V_{OUT}, I_{OUT(SOURCE)} = -1.5 \text{ mA}$	—	400	700	mV
<b>Park Function</b>						
PARK DRIVE Leakage Current	$I_{CEX}$	$V_{CEX} = 20 \text{ V}$	—	—	100	$\mu\text{A}$
PARK DRIVE Saturation Voltage	$V_{CE(SAT)}$	$I_C = 200 \text{ mA}$	—	300	500	mV
PARK Input Threshold	$V_{PARK}$		0.7	1.1	1.7	V
PARK Input Current	$I_{PARK}$	$V_{PARK} = 1.7 \text{ V}$	—	—	100	$\mu\text{A}$
PARK VOLTAGE Input Current	$I_{PARK\ V}$		—	-150	-500	nA
<b>Under-Voltage Protection</b>						
UV Threshold	$V_{UV}$	Low-to-High Trans., Other Input = 6 V	1.48	1.50	1.52	V
UV Threshold Hysteresis	$\Delta V_{UV}$		15	25	45	mV
UV Input Current	$I_{UV}$	$V_{UV} = 1 \text{ V}$	—	-0.5	-1.5	$\mu\text{A}$
PWR OK Saturation Voltage	$V_{CE(SAT)}$	$I_C = 5 \text{ mA}$	—	—	450	mV
PWR OK Leakage Current	$I_{CEX}$	$V_{CEX} = 20 \text{ V}$	—	—	5.0	$\mu\text{A}$
<b>Auxiliary Functions</b>						
LIMIT Input Voltage	$V_{LIMIT(L)}$	OUT <sub>A</sub> forced Low	0.7	0.8	—	V
	$V_{LIMIT(H)}$	OUT <sub>A</sub> forced High	—	2.2	2.3	V
	$V_{LIMIT}$	Limit inactive	1.2	—	1.8	V
		Open circuit	1.45	1.50	1.55	V
LIMIT Input Resistance	$R_{LIMIT}$	$V_{LIMIT} = 1.2 \text{ V to } 1.8 \text{ V}$	—	10	—	k $\Omega$
INHIBIT Input Threshold	$V_2$		0.7	1.1	1.7	V
INHIBIT Input Current	$I_2$	$V_2 = 1.7 \text{ V}$	—	—	200	$\mu\text{A}$

Negative current is defined as coming out of (sourcing) the specified device terminal.

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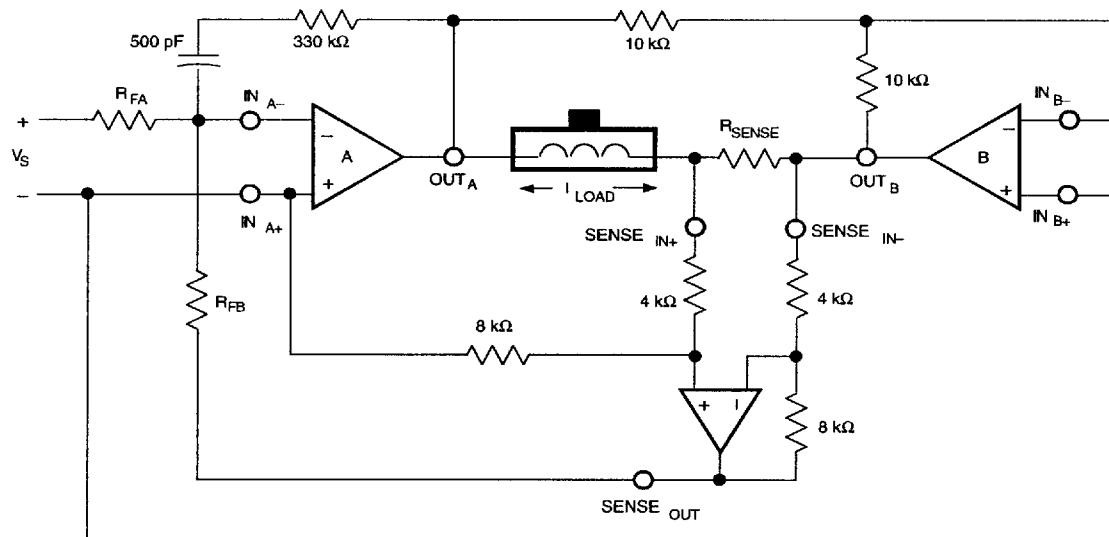
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## TERMINAL FUNCTIONS

'EA' Term.	'LB' Term.	Terminal Name	Function
1	1	LOGIC SUPPLY	V <sub>CC</sub> ; logic supply voltage.
2	2	INHIBIT	An active-high logic input that inhibits the output stages without initiating a park.
3 & 4	3 & 4	UV <sub>1</sub> and UV <sub>2</sub>	Under-voltage detection inputs. If not used, these terminals must be connected to the logic supply (V <sub>CC</sub> ).
5	5	LIMIT	A tri-state input that forces the output of amplifier A into saturation in either direction, or allows normal linear operation.
—	6 & 7	GROUND <sub>AB</sub>	Power amplifiers' ground and thermal heat sink.
6	8	PARK VOLTAGE	Auxiliary inverting input to power amplifier A.
7	9	SENSE <sub>IN-</sub>	Inverting input to current sense error amplifier.
8	10	IN <sub>A+</sub>	Non-inverting input to power amplifier A.
9	11	IN <sub>A-</sub>	Inverting input to power amplifier A.
10	12	OUT <sub>A</sub>	Power amplifier A output to voice coil motor.
11	—	GROUND <sub>A</sub>	Power ground of amplifier A.
12-18	—	GROUND	Circuit reference and thermal heat sink.
19	—	GROUND <sub>B</sub>	Power ground of amplifier B.
20	13	OUT <sub>B</sub>	Power amplifier B output to voice coil motor.
21	14	LOAD SUPPLY	V <sub>BB</sub> ; load supply voltage.
22	15	IN <sub>B-</sub>	Inverting input to power amplifier B.
23	16	IN <sub>B+</sub>	Non-inverting input to power amplifier B.
24	17	SENSE <sub>IN+</sub>	Non-inverting input to current sense error amplifier.
—	18 & 19	GROUND <sub>AB</sub>	Power amplifiers' ground and thermal heat sink.
25	20	PWR OK	A logic low at this output indicates an under-voltage condition.
—	21	GROUND	Circuit reference.
26	22	PARK	An active-high logic input that activates the park function.
27	23	PARK DRIVE	Power transistor for retract current control on power down or park command.
28	24	SENSE <sub>OUT</sub>	Output of current sense error amplifier.

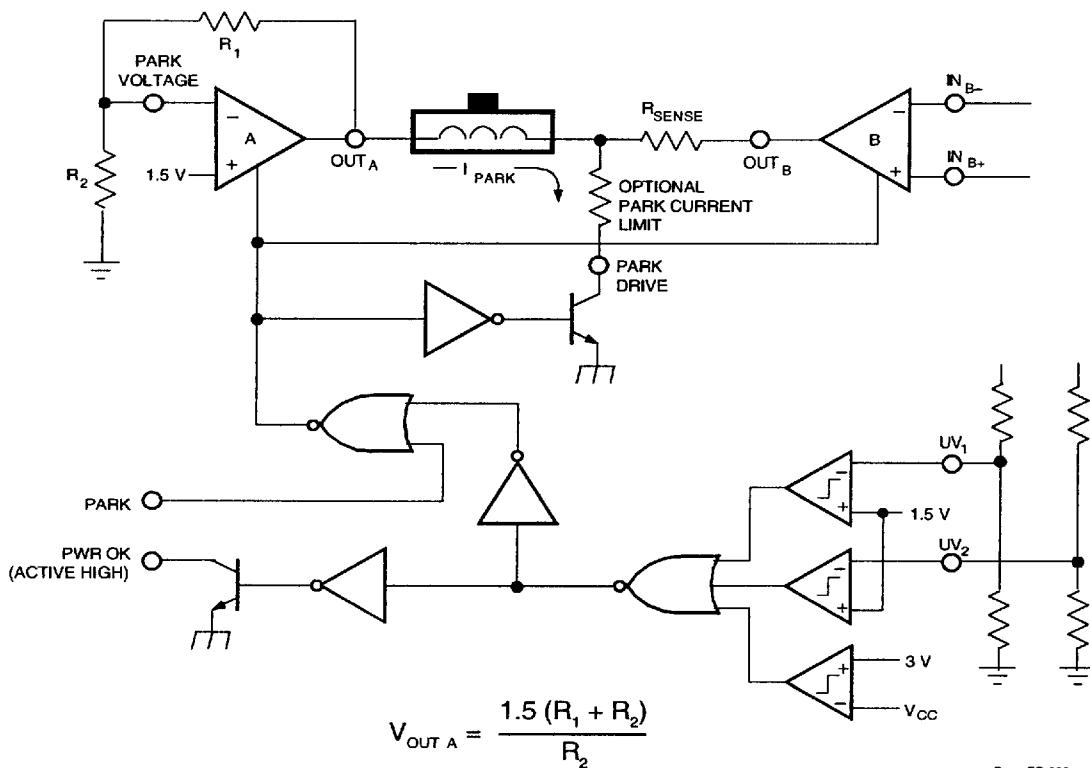
# 3053 VOICE COIL MOTOR DRIVER

## CURRENT SENSING



Dwg. EP-034

## PARKING FUNCTION



Dwg. EP-039



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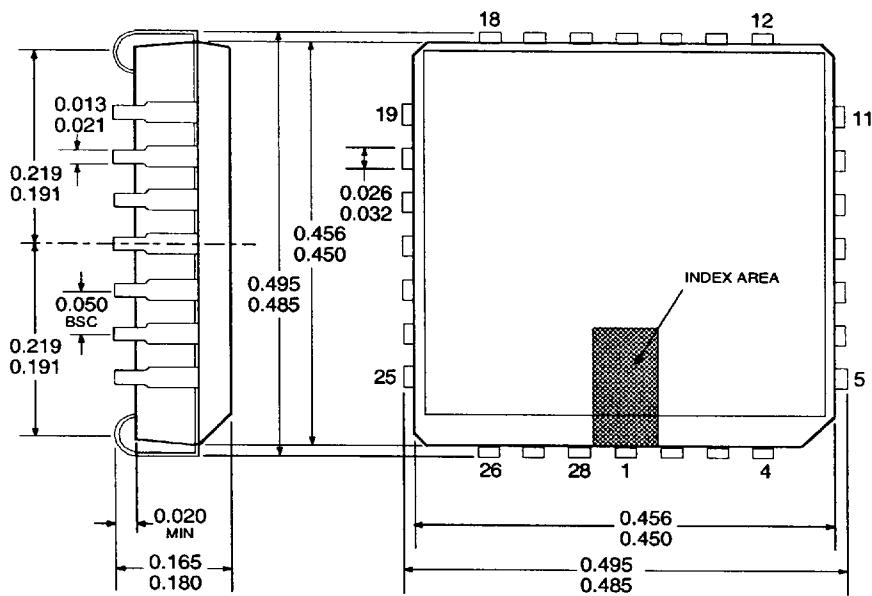
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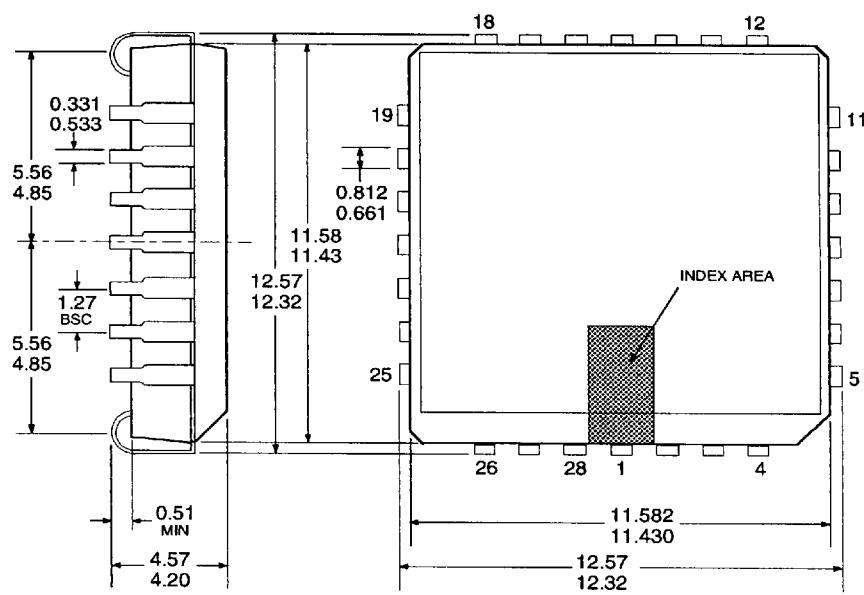
VOICE COIL MOTOR DRIVER

**A8958CEA**

**Dimensions in Inches**



**Dimensions in Millimeters  
(Based on 1" = 25.4 mm)**



- NOTES:**
1. Webbed lead frame. Leads 12-18 are internally one piece.
  2. Lead spacing tolerance is non-cumulative.
  3. Exact body and lead configuration at vendor's option within limits shown.

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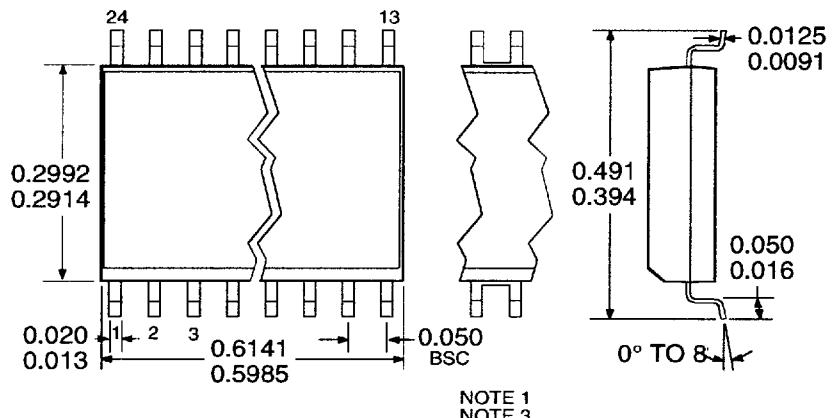
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8958

# VOICE COIL MOTOR DRIVER

## A8958CLB

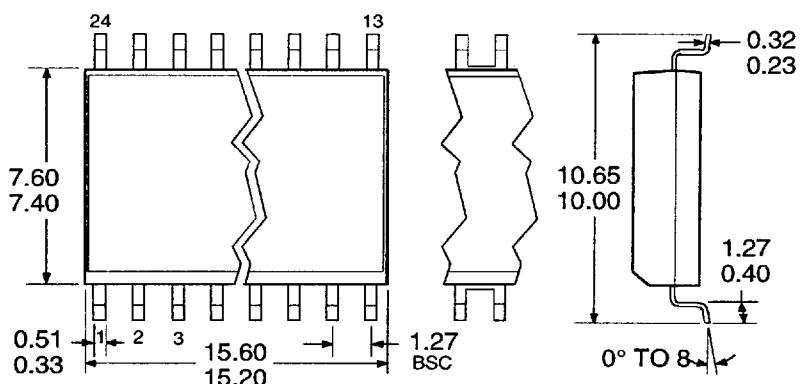
Dimensions in Inches  
(Based on 1 mm = 0.03937")



NOTE 1  
NOTE 3

Dwg. MA-008-25 In

## Dimensions in Millimeters



NOTE 1  
NOTE 3

Dwg. MA-008-25A mm

- Allegro MicroSystems, Inc. reserves the right to make, from time to time, such departures from the detail specifications as may be required to permit improvements in the design of its products. Components made under military approvals will be in accordance with the approval requirements.
- The information included herein is believed to be accurate and reliable. However, Allegro MicroSystems, Inc. assumes no responsibility for its use; nor for any infringements of patents or other rights of third parties which may result from its use.
- NOTES:
1. Webbed lead frame. Leads 6, 7, 18, and 19 are internally one piece.
  2. Lead spacing tolerance is non-cumulative.
  3. Exact body and lead configuration at vendor's option within limits shown.

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