



Advanced
Linear
Devices

002176

ALD1702A/ALD1702B
ALD1702/ALD1703

5V RAIL TO RAIL PRECISION OPERATIONAL AMPLIFIER

GENERAL DESCRIPTION

The ALD 1702/ ALD 1703 is a monolithic operational amplifier intended primarily for a wide range of analog applications in +5 V single power supply and ± 5 V dual power supply systems as well as +6 V to +12 V battery operated systems. All device characteristics are specified for +5V single supply or ± 2.5 V dual supply systems. It is manufactured with Advanced Linear Devices' enhanced ACMOS silicon gate CMOS process.

The device is designed to offer a balanced tradeoff of performance parameters providing a wide range of desired specifications. It offers the industry pin configuration of μA 741 and ICL 7611 types.

The ALD 1702/ ALD 1703 has been developed specifically with the 5 V single supply or ± 2.5 dual supply user in mind. Several important characteristics of the device make many applications easy to implement for these supply voltages. First, the operational amplifier can operate with rail to rail input and output voltages. This feature allows numerous analog serial stages to be implemented without losing operating voltage margin. Secondly, the device was designed to accommodate mixed applications where digital and analog circuits may work off the same 5 V power supply. Thirdly, the output stage can drive up to 400 pF capacitive and 5 K Ω resistive loads in non-inverting unity gain connection and double the capacitance in the inverting unity gain mode. These features, coupled with extremely low input currents, high voltage gain, useful bandwidth of 1.5 MHz, slew rate of 2.1 V/ μs , low power dissipation, low offset voltage and temperature drift, make the ALD 1702/ALD 1703 a truly versatile, user friendly, operational amplifier.

The ALD1702/ALD1703 is designed and fabricated with silicon gate CMOS technology, and offers 1 pA typical input bias current. On-chip offset voltage trimming allows the device to be used without nulling in most applications. The device offers typical offset drift of less than 7 $\mu V/^{\circ}C$ which eliminates many trim or temperature compensation circuits. For precision applications, the 1702 is designed to settle to 0.01 % in 8 μs .

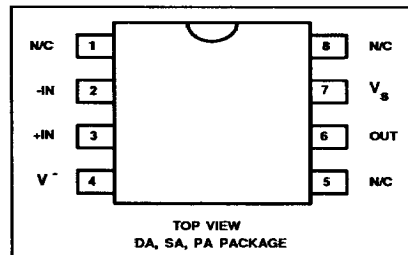
ORDERING INFORMATION

Operating Temperature Range			
	-55 $^{\circ}C$ to +125 $^{\circ}C$	0 $^{\circ}C$ to +70 $^{\circ}C$	0 $^{\circ}C$ to +70 $^{\circ}C$
+25 $^{\circ}C$ VOS (mV)	8-Pin CERDIP Package	8-Pin Small Outline Package (SOIC)	8-Pin Plastic Dip Package
0.9		ALD 1702 ASA	ALD 1702A PA
2.0	ALD 1702B DA	ALD 1702 BSA	ALD 1702B PA
4.5	ALD 1702 DA	ALD 1702 SA	ALD 1702 PA
10.0		ALD 1703 SA	ALD 1703 PA
			ALD 1703 Z (Dice)

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PIN CONFIGURATION



FEATURES

- All parameters specified for +5 V single supply or ± 2.5 V dual supply systems.
- Rail to rail input and output voltage ranges
- High load capacitance capability – 400 pF typical
- No frequency compensation required – unity gain stable
- Extremely low input bias currents – 1.0 pA typical (30 pA max.)
- Ideal for high source impedance applications
- Dual power supply ± 2.5 V to ± 5.0 V operation
- Single power supply +5 V to +12 V operation
- High voltage gain – typically 85 V/mV @ ± 2.5 V and 250 V/mV @ ± 5.0 V
- Drive as low as 2 K Ω load with 5 mA drive current
- Output short circuit protected
- Unity gain bandwidth of 1.5 MHz (1.0 MHz min.)
- Slew rate of 2.1 V/ μs (1.4 V/ μs min.)
- Low power dissipation

APPLICATIONS

- Voltage amplifier
- Voltage follower/buffer
- Charge integrator
- Photodiode amplifier
- Data acquisition systems
- High performance portable instruments
- Signal conditioning circuits
- Sensor and transducer amplifiers
- Low leakage amplifiers
- Active filters
- Sample/Hold amplifier
- Picoammeter
- Current to voltage converter
- Coaxial cable driver

ABSOLUTE MAXIMUM RATINGS

Supply voltage, V_{DD} _____ 12V
Differential input voltage range _____ -0.3V to $V_{DD} + 0.3V$
Power dissipation _____ 600 mW
Operating temperature range 1702XPA/1702XSA/1703PA/1703SA _____ 0°C to +70°C
1702XDA _____ -55°C to +125°C
Storage temperature range _____ -65°C to +150°C
Lead temperature, 10 seconds _____ +300°C

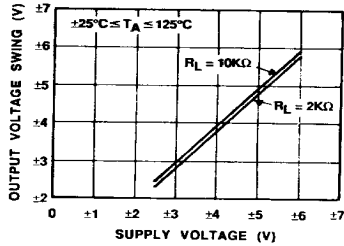
DC AND OPERATING ELECTRICAL CHARACTERISTICS

$T_A = 25^\circ\text{C}$ $V_S = \pm 2.5V$ unless otherwise specified

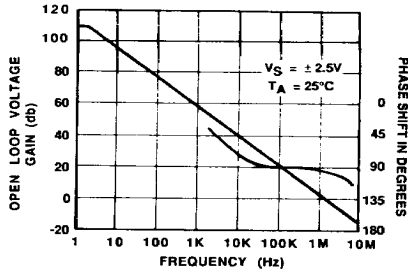
Parameter	Symbol	1702A			1702B			1702			1703			Unit	Test Conditions
		Min	Typ	Max	Min	Typ	Max	Min	Typ	Max	Min	Typ	Max		
Supply Voltage	V_S V_{DD}	± 2.0 4.0		± 6.0 (12.0)	± 2.0 4.0		± 6.0 (12.0)	± 2.0 4.0		± 6.0 (12.0)	± 2.0 4.0		± 6.0 (12.0)	V	Single Supply
Input Offset Voltage	V_{OS}			0.9 (1.7)			2.0 (2.8)			4.5 (5.3)			10.0 (11.0)	mV	$R_S \leq 100K\Omega$ $0^\circ\text{C} \leq T_A \leq +70^\circ\text{C}$
Input Offset Current	I_{OS}	1.0	25 240		1.0	25 240		1.0	25 240		1.0	30 450		pA	$T_A = 25^\circ\text{C}$ $0^\circ\text{C} \leq T_A \leq +70^\circ\text{C}$
Input Bias Current	I_B	1.0	(30) 300		1.0	(30) 300		1.0	(30) 300		1.0	(50) 600		pA	$T_A = 25^\circ\text{C}$ $0^\circ\text{C} \leq T_A \leq +70^\circ\text{C}$
Input Voltage Range	V_{IR}	0.0 -2.5		5.0 +2.5	0.0 -2.5		5.0 +2.5	0.0 -2.5		5.0 +2.5	0.15 -2.35		4.85 +2.35	V	$V_{DD} = +5V$ $V_S = \pm 2.5V$
Input Resistance	R_{IN}		10^{12}			10^{12}			10^{12}			10^{12}		Ω	
Input Offset Voltage Drift	TCV_{OS}		7			7			7			10		$\mu\text{V}/^\circ\text{C}$	$R_S \leq 100K\Omega$
Power Supply Rejection Ratio	PSRR	70	80		65	80		65	80		60	80		dB	$R_S \leq 100K\Omega$ $0^\circ\text{C} \leq T_A \leq +70^\circ\text{C}$
Common Mode Rejection Ratio	CMRR	70	83		65	83		65	83		60	83		dB	$R_S \leq 100K\Omega$ $0^\circ\text{C} \leq T_A \leq +70^\circ\text{C}$
Large Signal Voltage Gain	A_V	50 (20)	85 400		50 (20)	85 400		50 (20)	85 400		32 (10)	85 300		V/mV V/mV V/mV	$R_L = 10K\Omega$ $R_L \geq 1M\Omega$ $R_L = 10K\Omega$ $0^\circ\text{C} \leq T_A \leq +70^\circ\text{C}$
Output Voltage Range	V_O low		0.06	0.15		0.06	0.15		0.06	0.15		0.1	0.2	V	$R_L = 10K\Omega$ $V_{DD} = 5V$ $0^\circ\text{C} \leq T_A \leq +70^\circ\text{C}$
	V_O high	4.85	4.94		4.85	4.94		4.85	4.94		4.8	4.9		V	
	V_O low		-2.44	-2.35		-2.44	-2.35		-2.44	-2.35		-2.4	-2.3	V	$R_L = 10K\Omega$ $0^\circ\text{C} \leq T_A \leq +70^\circ\text{C}$
	V_O high	2.35	2.44		2.35	2.44		2.35	2.44		2.3	2.4		V	
Output Short Circuit Current	I_{SC}		8			8			8			8		mA	
Supply Current	I_S		1.1	2.0		1.1	2.0		1.1	2.0		1.1	2.5	mA	$V_{IN} = 0V$ No Load
Power Dissipation	P_D		5.5	10.0		5.5	10.0		5.5	10.0		5.5	12.5	mW	$V_S = \pm 2.5V$
Input Capacitance	C_{IN}		1			1			1			1		pF	
Bandwidth	B_W	(1.0)	1.5		(1.0)	1.5		(1.0)	1.5		(0.7)	1.5		MHz	
Slew Rate	S_R	(1.4)	2.1		(1.4)	2.1		(1.4)	2.1		(1.1)	2.1		V/ μS	$A_V = +1$ $R_L = 10K\Omega$
Rise time	t_r		0.2			0.2			0.2			0.2		μS	$R_L = 10K\Omega$
Overshoot Factor			10			10			10			10		%	$R_L = 10K\Omega$ $C_L = 100pF$

TYPICAL PERFORMANCE CHARACTERISTICS

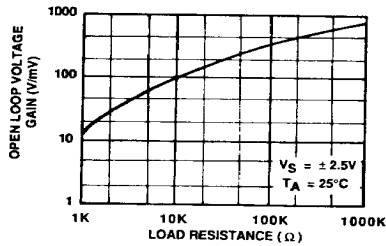
OUTPUT VOLTAGE SWING AS A FUNCTION OF SUPPLY VOLTAGE



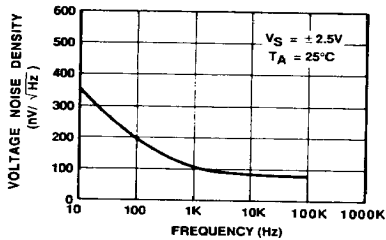
OPEN LOOP VOLTAGE GAIN AS A FUNCTION OF FREQUENCY



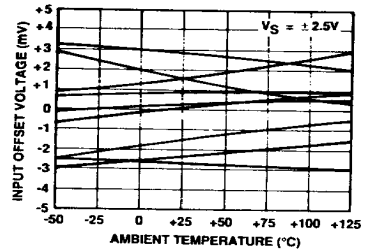
OPEN LOOP VOLTAGE GAIN AS A FUNCTION OF LOAD RESISTANCE



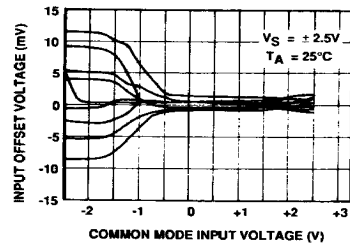
VOLTAGE NOISE DENSITY AS A FUNCTION OF FREQUENCY



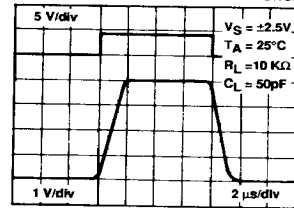
INPUT OFFSET VOLTAGE AS A FUNCTION OF AMBIENT TEMPERATURE
REPRESENTATIVE UNITS



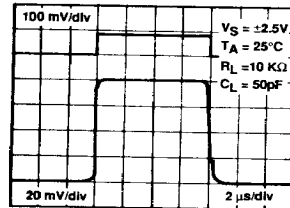
INPUT OFFSET VOLTAGE AS A FUNCTION OF COMMON MODE INPUT VOLTAGE



LARGE-SIGNAL TRANSIENT RESPONSE



SMALL-SIGNAL TRANSIENT RESPONSE



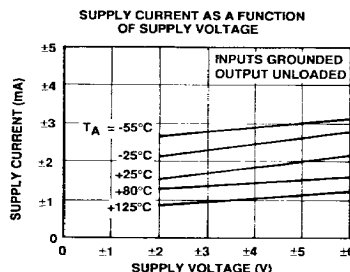
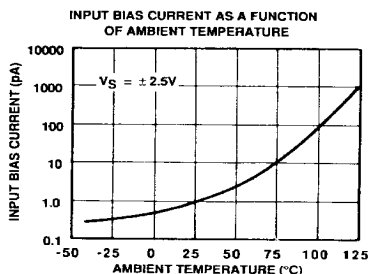
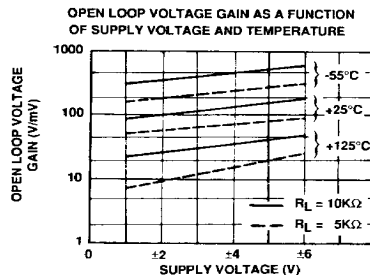
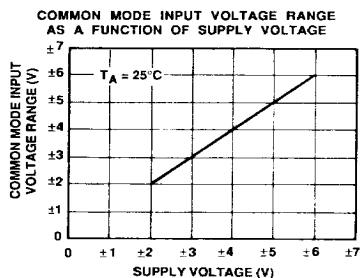
Design & Operating Notes:

1. The ALD 1702/1703 CMOS operational amplifier uses a 3 gain stage architecture and an improved frequency compensation scheme to achieve large voltage gain, high output driving capability, and better frequency stability. In a conventional CMOS operational amplifier design, compensation is achieved with a pole splitting capacitor together with a nulling resistor. This method is, however, very bias dependent and thus cannot accommodate the large range of supply voltage operation as is required from a stand alone CMOS operational amplifier. The ALD 1702 is internally compensated for unity gain stability using a novel scheme that does not use a nulling resistor. This scheme produces a clean single pole roll off in the gain characteristics while providing for more than 70 degrees of phase margin at the unity gain frequency. A unity gain buffer using the ALD 1702 will typically drive 400pF of external load capacitance without stability problems. In the inverting unity gain configuration, it can drive up to 800pF of load capacitance. Compared to other CMOS operational amplifiers, the ALD 1702 has shown itself to be more resistant to parasitic oscillations.
2. The ALD 1702/1703 has complementary p-channel and n-channel input differential stages connected in parallel to accomplish rail to rail input common mode voltage range. This means that with the ranges of common mode input voltage close to the power supplies, one of the two differential stages is switched off internally. To maintain compatibility with other operational amplifiers, this switching point has been selected to be about 1.5 V above the negative supply voltage. Since offset voltage trimming on the 1702/1703 is made when the input voltage is symmetrical to the supply voltages, this internal switching does not affect a large variety of applications such as an inverting amplifier or non-inverting amplifier with a gain larger than 2.5 (5V operation), where the common mode voltage does not make excursions below this

switching point. The user should however, be aware that this switching does take place if the operational amplifier is connected as a unity gain buffer and should make provision in his design to allow for input offset voltage variations.

3. The input bias and offset currents are essentially input protection diode reverse bias leakage currents, and are typically less than 1pA at room temperature. This low input bias current assures that the analog signal from the source will not be distorted by input bias currents. Normally, this extremely high input impedance of greater than $10^{12} \Omega$ would not be a problem as the source impedance would limit the node impedance. However, for applications where source impedance is very high, it may be necessary to limit noise and hum pickup through proper shielding.
4. The output stage consists of class AB complementary output drivers, capable of driving a low resistance load. The output swing is limited by the drain to source on-resistance of the output transistors as determined by the bias circuitry, and the value of the load resistor. When connected in the voltage follower configuration, the oscillation resistant feature, combined with the rail to rail input and output feature, makes an effective analog signal buffer for medium to high source impedance sensors, transducers, and other circuit networks.
5. The ALD 1702/ALD 1703 operational amplifier has been designed to provide full static discharge protection. Internally, the design has been carefully implemented to minimize latch up. However, care must be exercised when handling the device to avoid strong static fields that may degrade a diode junction, causing increased input leakage currents. In using the operational amplifier, the user is advised to power up the circuit before, or simultaneously with, any input voltages applied and to limit input voltages to not exceed 0.3V of the power supply voltage levels.

TYPICAL PERFORMANCE CHARACTERISTICS



DC AND OPERATING ELECTRICAL CHARACTERISTICS (con't)

$T_A = 25^\circ\text{C}$ $V_S = \pm 2.5\text{V}$ unless otherwise specified

Parameter	Symbol	1702A			1702B			1702			1703			Unit	Test Conditions
		Min	Typ	Max	Min	Typ	Max	Min	Typ	Max	Min	Typ	Max		
Maximum Load Capacitance	C_L		400			400			400			400		pF	
Input Noise Voltage	e_n		100			100			100			100		nV/√Hz	f=1KHz
Input Current Noise	i_n		.001			.001			.001			.001		pA/√Hz	f=10Hz
Settling Time	t_s		8.0 3.0			8.0 3.0			8.0 3.0			8.0 3.0		μs μs	0.01% 0.1% AV=-1 $R_L=5\text{K}\Omega$ $C_L=50\text{pF}$

$T_A = 25^\circ\text{C}$ $V_S = \pm 5.0\text{V}$ unless otherwise specified

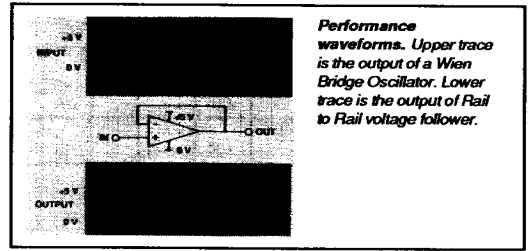
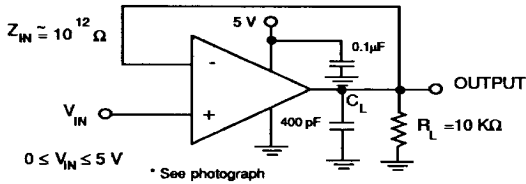
Parameter	Symbol	1702A			1702B			1702			1703			Unit	Test Conditions
		Min	Typ	Max	Min	Typ	Max	Min	Typ	Max	Min	Typ	Max		
Power Supply Rejection Ratio	PSRR		83			83			83			83		db	$R_S \leq 100\text{K}\Omega$
Common Mode Rejection Ratio	CMRR		83			83			83			83		db	$R_S \leq 100\text{K}\Omega$
Large Signal Voltage Gain	A_V		250			250			250			250		V/mV	$R_L = 10\text{K}\Omega$
Output Voltage Range	$V_{O \text{ low}}$ $V_{O \text{ high}}$	4.8	-4.9 4.93	-4.8	4.8	-4.9 4.93	-4.8	4.8	-4.9 4.93	-4.8	4.8	-4.9 4.93	-4.8	V	$R_L = 10\text{K}\Omega$
Bandwidth	B_W		1.7			1.7			1.7			1.7		MHz	
Slew Rate	S_R		2.8			2.8			2.8			2.8		V/μS	$A_V = +1$ $C_L = 50\text{pF}$

$V_S = \pm 2.5\text{V}$ $-55^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$ unless otherwise specified

Parameter	Symbol	1702B DA			1702 DA			Unit	Test Conditions
		Min	Typ	Max	Min	Typ	Max		
Input Offset Voltage	V_{OS}			3.0			6.5	mV	$R_S \leq 100\text{K}\Omega$
Input Offset Current	I_{OS}			8.0			8.0	nA	
Input Bias Current	I_B			10.0			10.0	nA	
Power Supply Rejection Ratio	PSRR	60	75		60	75		db	$R_S \leq 100\text{K}\Omega$
Common Mode Rejection Ratio	CMRR	60	83		60	83		db	$R_S \leq 100\text{K}\Omega$
Large Signal Voltage Gain	A_V	10	25		7	25		V/mV	$R_L = 10\text{K}\Omega$
Output Voltage Range	$V_{O \text{ low}}$ $V_{O \text{ high}}$	4.8	0.1 4.9	0.2	4.8	0.1 4.9	0.2	V	$R_L = 10\text{K}\Omega$

TYPICAL APPLICATIONS

RAIL TO RAIL VOLTAGE FOLLOWER/BUFFER



LOW OFFSET SUMMING AMPLIFIER

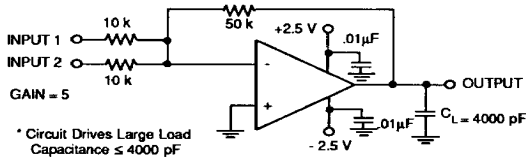
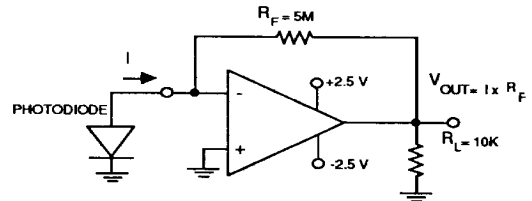
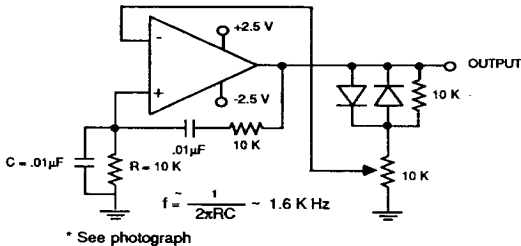


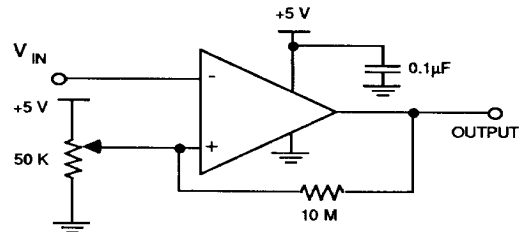
PHOTO DETECTOR CURRENT TO VOLTAGE CONVERTER



WIEN BRIDGE OSCILLATOR (RAIL TO RAIL) SINE WAVE GENERATOR



RAIL TO RAIL VOLTAGE COMPARATOR



ULTRA LONG TIME CONSTANT INTEGRATOR

