

# Dual, Low-Noise, 200mA LDO Regulator

## General Description

The RT9182 is a dual-channel, low noise, and low dropout regulator supplying up to 200mA current at each channel. The output voltage ranges from 1.5V to 3.3V in 100mV increments and 2% accuracy by operating from a +2.7V to +5.5V input.

The RT9182 uses an internal PMOS as the pass device, which consumes 165μA supply current (both LDOs on) independent of load current and dropout conditions. The SHDN1 and SHDN2 pins control each output respectively. When both outputs shutdown simultaneously, the chip will be turn off and consumes nearly zero operation current which is suitable for battery-power devices. Other features include a current limiting, and over temperature protection.

## Ordering Information

RT9182□□□	
□	Package type
ES : SOT-26	
□	Operating temperature range
C : Commercial standard	
□	Output voltage
A : 2.8V (Output1), 2.8V (Output2)	
B : 3.0V (Output1), 3.0V (Output2)	
C : 2.8V (Output1), 3.0V (Output2)	

## Marking Information

For marking information, contact our sales representative directly or through a RichTek distributor located in your area, otherwise visit our website for detail.

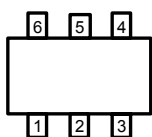
## Features

- Up to 200mA Output Current (Each LDO)
- Dual Shutdown Pins Control Each Output
- 124μV<sub>RMS</sub> Low Noise Output
- Current Limiting and Thermal Protection
- Short Circuit Protection
- 120mV Dropout at 100mA Load
- Two LDOs in SOT-26 Package

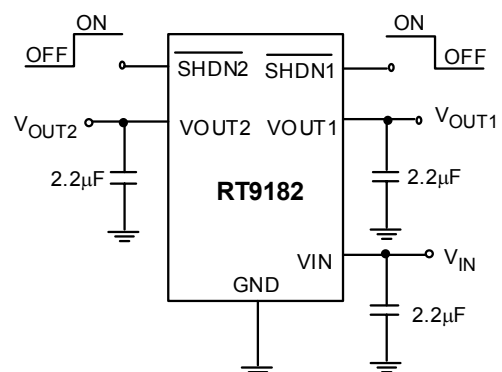
## Applications

- Cellular Phones
- Laptop, Notebook, and Palmtop Computers
- Battery-powered Equipment
- Hand-held Equipment
- Wireless LAN

## Pin Configurations

Part Number	Pin Configurations
RT9182□CES (Plastic SOT-26)	<p>TOP VIEW</p>  <ol style="list-style-type: none"> <li>1. VOUT2</li> <li>2. <u>GND</u></li> <li>3. <u>SHDN2</u></li> <li>4. SHDN1</li> <li>5. VIN</li> <li>6. VOUT1</li> </ol>

## Typical Application Circuit

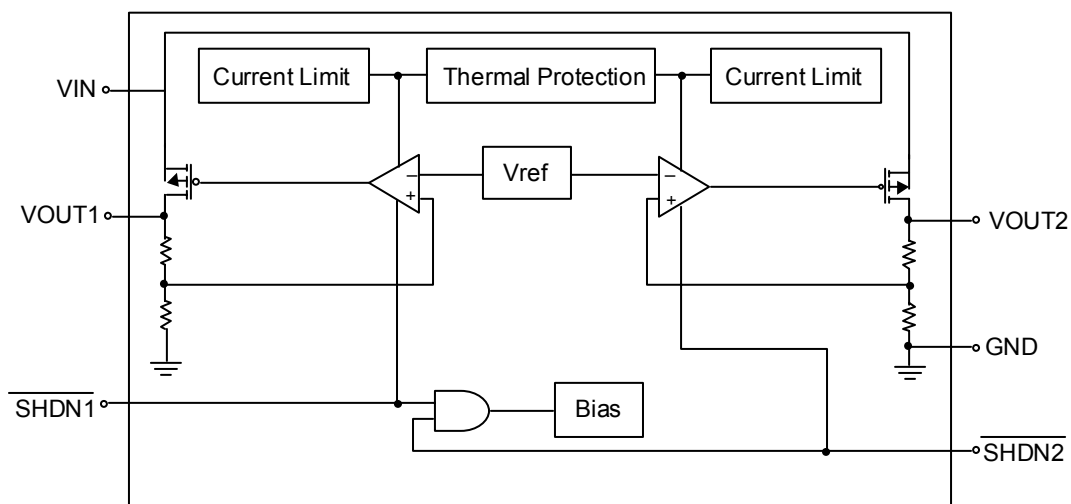


## Pin Description

Pin No.	Pin Name	Pin Function
1	VOUT2	Channel 2 Output Voltage
2	GND	Common Ground
3	$\overline{\text{SHDN2}}$ <small>Note</small>	Channel 2 Shutdown Input Logic, Active Low. If the shutdown feature is not required, connect SHDN2 to VIN.
4	$\overline{\text{SHDN1}}$ <small>Note</small>	Channel 1 Shutdown Input Logic, Active Low. If the shutdown feature is not required, connect SHDN1 to VIN.
5	VIN	Supply Input
6	VOUT1	Channel 1 Output Voltage

Note: Active Low shutdown. If  $\overline{\text{SHDN1}}$  and  $\overline{\text{SHDN2}}$  are both low, both regulators and the reference turn off.

## Function Block Diagram



## Absolute Maximum Ratings

- Supply Input Voltage ..... 6.5V
- Power Dissipation,  $P_D$  @  $T_A = 25^\circ\text{C}$   
SOT-26 ..... 748mW
- Lead Temperature (Soldering, 10 sec.) .....  $260^\circ\text{C}$
- Junction Temperature .....  $125^\circ\text{C}$
- Storage Temperature Range .....  $-65^\circ\text{C}$  to  $150^\circ\text{C}$

## Recommended Operating Conditions (Note 1)

- Supply Input Voltage ..... 2.7V to 5.5V
- Shutdown Input Voltage ..... 0V to 5.5V
- Junction Temperature Range .....  $-40^\circ\text{C}$  to  $85^\circ\text{C}$

## Electrical Characteristics

( $V_{IN} = 3.6\text{V}$ ,  $C_{IN} = C_{OUT} = 2.2\mu\text{F}$ ,  $\overline{\text{SHDN}} = \overline{\text{SHDN1}} = \overline{\text{SHDN2}} = V_{IN}$ , typical values at  $T_A = 25^\circ\text{C}$ , for each LDO unless otherwise specified.)

Parameter	Symbol	Test Conditions	Min	Typ	Max	Units
Output Voltage Accuracy (Load Regulation)	$\Delta V_{OUT}$	$I_L = 1\text{mA}$ to $200\text{mA}$	-2	--	+2	%
Maximum Output Current	$I_{MAX}$	Continuous	200	--	--	mA
Current Limit	$I_{LIMIT}$	$R_{LOAD} = 1\Omega$	500	--	1000	mA
GND Pin Current	$I_G$	No Load	--	165	260	$\mu\text{A}$
		$I_{OUT} = 100\text{mA}$ (Both LDOs)	--	165	260	$\mu\text{A}$
Dropout Voltage (Note 2)	$V_{DROP}$	$I_{OUT} = 1\text{mA}$	--	1.2	--	mV
		$I_{OUT} = 100\text{mA}$	--	120	--	mV
		$I_{OUT} = 200\text{mA}$	--	255	--	mV
Line Regulation	$\Delta V_{LINE}$	$V_{IN} = (V_{OUT} + 0.4\text{V or } 2.7\text{V})$ to $5.5\text{V}$ $I_{OUT} = 1\text{mA}$	-0.2	--	+0.2	%/V
$\overline{\text{SHDN}}$ Input High Threshold	$V_{IH}$	$V_{IN} = 2.7\text{V}$ to $5.5\text{V}$	1.6	--	--	V
$\overline{\text{SHDN}}$ Input Low Threshold	$V_{IL}$	$V_{IN} = 2.7\text{V}$ to $5.5\text{V}$	--	--	0.4	V
$\overline{\text{SHDN}}$ Input Bias Current	$I_{SD}$	$\overline{\text{SHDN}} = \text{GND or } V_{IN}$	--	--	100	nA
Shutdown Supply Current	$I_{GSD}$	$\overline{\text{SHDN1}} = \overline{\text{SHDN2}} = \text{GND}$	--	0.01	2	$\mu\text{A}$
Thermal Shutdown Temperature	$T_{SD}$		--	140	--	$^\circ\text{C}$
Thermal Shutdown Hysteresis	$\Delta T_{SD}$		--	10	--	$^\circ\text{C}$
Output Voltage Noise	$e_{NO}$	10Hz to 100kHz, $C_{OUT} = 4.7\mu\text{F}$ , $I_{LOAD} = 1\text{mA}$	--	124	--	$\mu\text{VRMS}$
Output Voltage AC PSRR		100Hz, $C_{OUT} = 4.7\mu\text{F}$ , $I_{LOAD} = 100\text{mA}$	--	62	--	dB

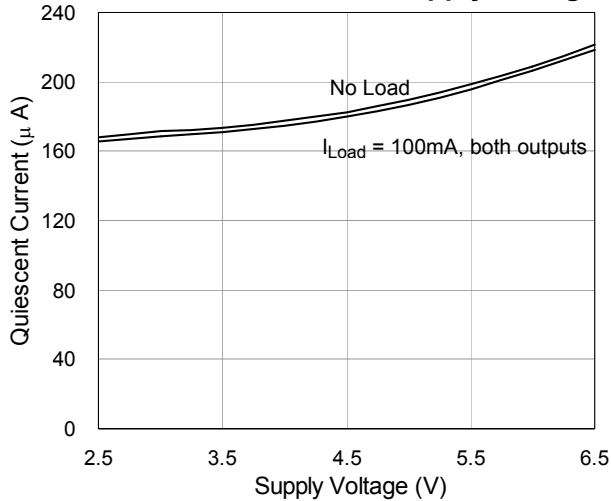
**Note 1.** The device is not guaranteed to function outside its operating conditions

**Note 2.** Dropout voltage definition:  $V_{IN} - V_{OUT}$  when  $V_{OUT}$  is 50mV below the value of  $V_{OUT}$  (normal)

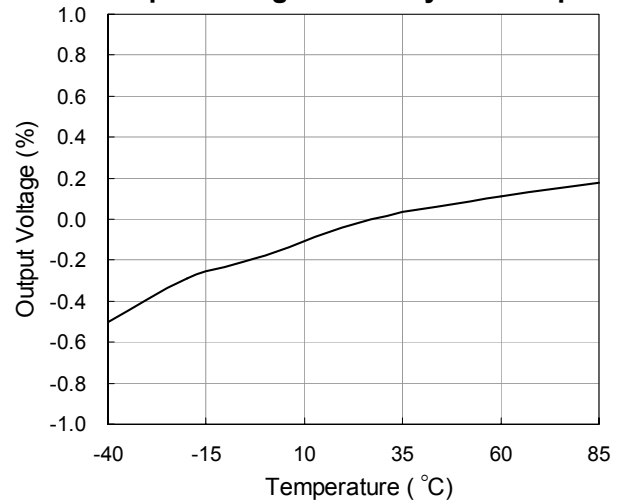
## Typical Operating Characteristics

$V_{OUT} = 2.8V$ ,  $I_{LOAD} = 100mA$ ,  $V_{IN} = 3.6V$ ,  $C_{OUT} = 4.7\mu F$ , and  $C_{IN} = 2.2\mu F$ , unless otherwise noted.

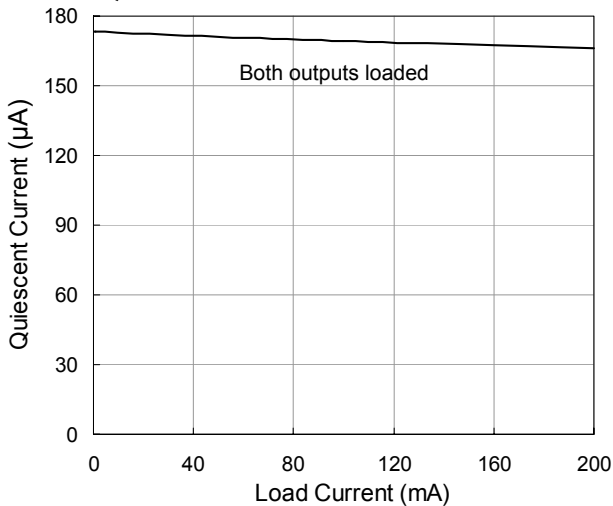
**Quiescent Current vs. Supply Voltage**



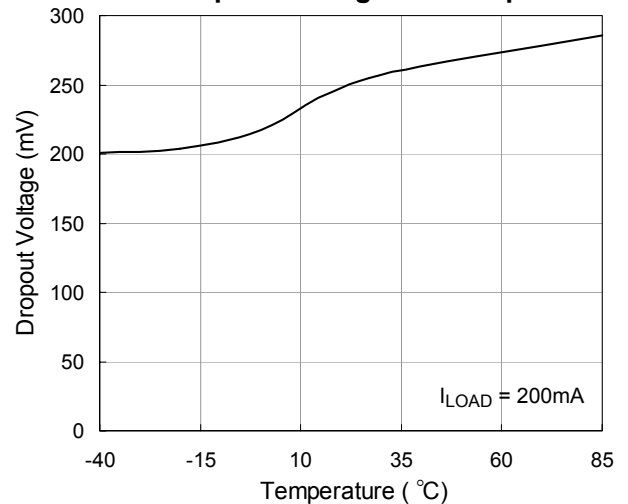
**Output Voltage Accuracy vs. Temp.**



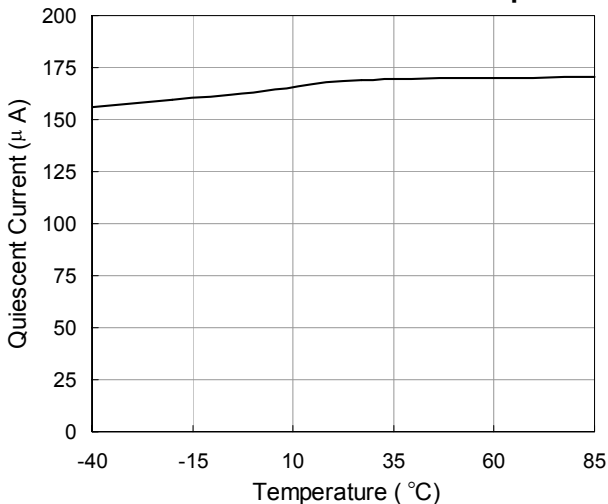
**Quiescent Current vs. Load Current**



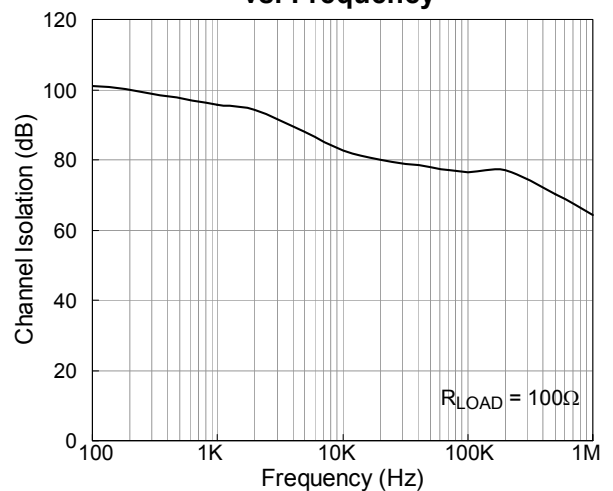
**Dropout Voltage vs. Temp.**



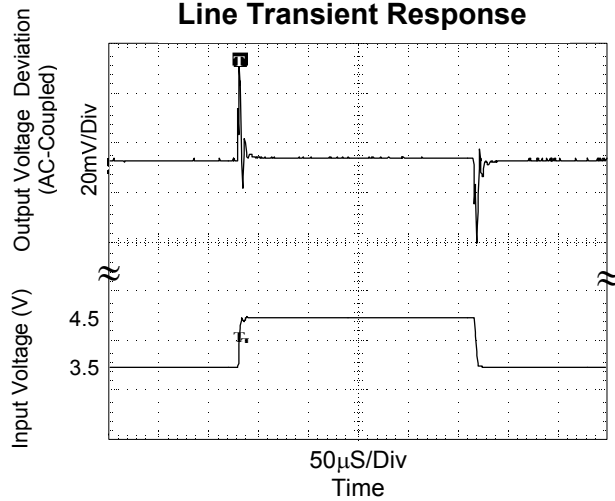
**Quiescent Current vs. Temp.**



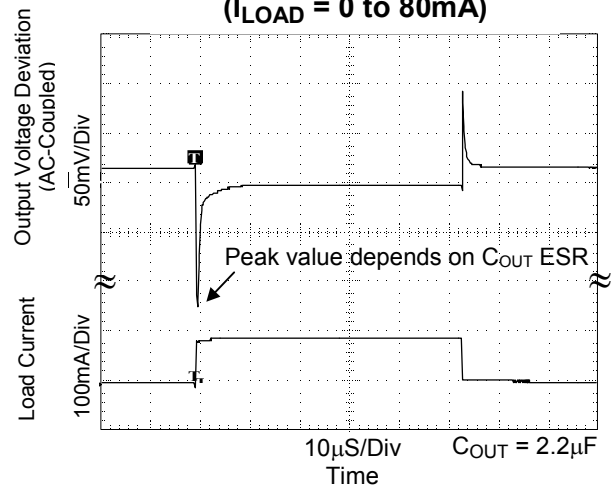
**Channel-to-Channel Isolation vs. Frequency**



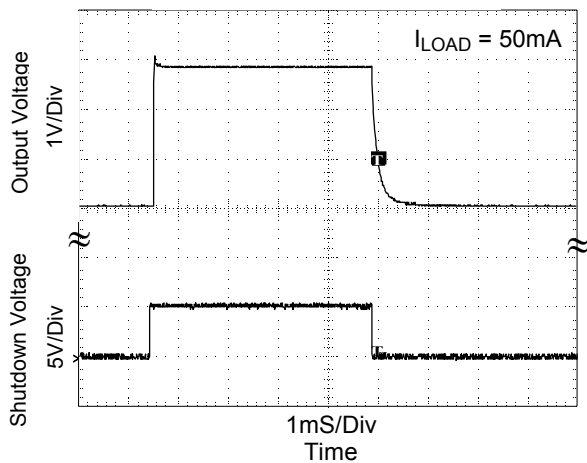
**Line Transient Response**



**Load Transient Response**  
( $I_{LOAD} = 0$  to 80mA)



**Shutdown Response**



## Functional Description

The RT9182 is integrated with two low noise, low dropout, and low quiescent current linear regulators designed primarily for battery-powered applications. Output voltages are optional ranging from 1.5V to 3.3V, and each channel can supply current up to 200mA.

### Shutdown

The RT9182 is shutdown by pulling the  $\overline{\text{SHDN}}$  input low, and turned on by driving the  $\overline{\text{SHDN}}$  input high. If this feature is not to be used, the  $\overline{\text{SHDN}}$  input should be tied to VIN to keep the regulator on at all times (the  $\overline{\text{SHDN}}$  input must **not** be left floating).

### Internal P-Channel Pass Transistor

The RT9182 features double typical  $1.5\Omega$  P-channel MOSFET pass transistors. It provides several advantages over similar designs using PNP pass transistors, including longer battery life. The P-channel MOSFET requires no base drive, which reduces quiescent current considerably. PNP-based regulators waste considerable current in dropout when the pass transistor saturates. They also use high base-drive currents under large loads. The RT9182 does not suffer from these problems and consume only  $165\mu\text{A}$  of quiescent current whether in dropout, light-load, or heavy-load applications.

### Current Limit and Thermal Protection

The RT9182 includes two independent current limit structure which monitor and control each pass transistor's gate voltage limiting the guaranteed maximum output current to 200mA minimum.

Thermal-overload protection limits total power dissipation in the RT9182. When the junction temperature exceeds  $T_J = +140^\circ\text{C}$ , the thermal sensor signals the shutdown logic turning off the pass transistor and allowing the IC to cool. The thermal sensor will turn the pass transistor on again after the IC's junction temperature cools by  $10^\circ\text{C}$ ,

resulting in a pulsed output during continuous thermal-overload conditions. Thermal-overloaded protection is designed to protect the RT9182 in the event of fault conditions. Do not exceed the absolute maximum junction-temperature rating of  $T_J = +150^\circ\text{C}$  for continuous operation. The output can be shorted to ground for an indefinite amount of time without damaging the part by cooperation of current limit and thermal protection.

### Operating Region and Power Dissipation

The maximum power dissipation of RT9182 depends on the thermal resistance of the case and circuit board, the temperature difference between the die junction and ambient air, and the rate of airflow. The power dissipation across the device is

$$P = I_{\text{OUT}} (V_{\text{IN}} - V_{\text{OUT}}).$$

The maximum power dissipation is:

$$P_{\text{MAX}} = (T_J - T_A) / \theta_{\text{JA}}$$

where  $T_J - T_A$  is the temperature difference between the RT9182 die junction and the surrounding environment,  $\theta_{\text{JA}}$  is the thermal resistance from the junction to the surrounding environment. The GND pin of the RT9182 performs the dual function of providing an electrical connection to ground and channeling heat away. Connect the GND pin to ground using a large pad or ground plane.

## Applications Information

### Capacitor Selection and Regulator Stability

Like any low-dropout regulator, the external capacitors used with the RT9182 must be carefully selected for regulator stability and performance.

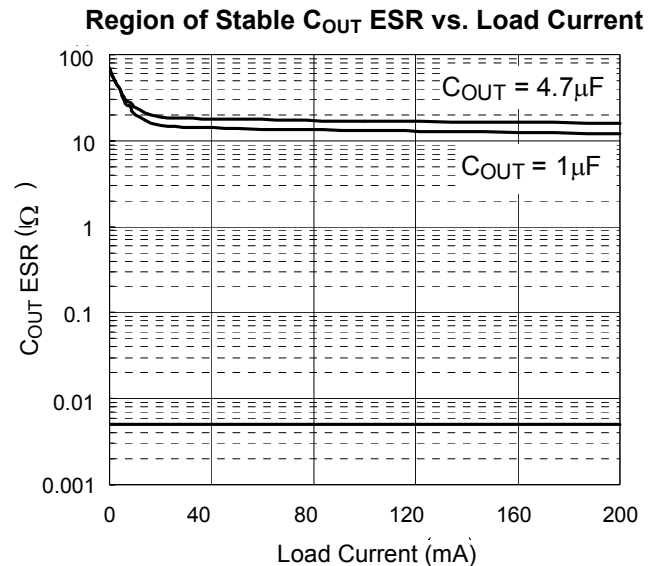
Using a capacitor whose value is  $> 1\mu\text{F}$  on the RT9182 input and the amount of capacitance can be increased without limit. The input capacitor must be located a distance of not more than 0.5" from the input pin of the IC and returned to a clean analog ground. Any good quality ceramic or tantalum can be used for this capacitor. The capacitor with larger value and lower ESR (equivalent series resistance) provides better PSRR and line-transient response.

The output capacitor must meet both requirements for minimum amount of capacitance and ESR in all LDO applications (see Fig.1). The RT9182 is designed specifically to work with low ESR ceramic output capacitor in space-saving and performance consideration. Using a ceramic capacitor whose value is at least  $1\mu\text{F}$  with ESR is  $> 5\text{m}\Omega$  on the RT9182 output ensures stability. The RT9182 still works well with output capacitor of other types due to the wide stable ESR range. Output capacitor of larger capacitance can reduce noise and improve load-transient response, stability, and PSRR. The output capacitor should be located not more than 0.5" from the  $V_{\text{OUT}}$  pin of the RT9182 and returned to a clean analog ground.

Note that some ceramic dielectrics exhibit large capacitance and ESR variation with temperature. It may be necessary to use  $2.2\mu\text{F}$  or more to ensure stability at temperatures below  $-10^\circ\text{C}$  in this case. Also, tantalum capacitors,  $2.2\mu\text{F}$  or more may be needed to maintain capacitance and ESR in the stable region for strict application environment.

Tantalum capacitors maybe suffer failure due to surge current when it is connected to a low-impedance source of power (like a battery or very large capacitor). If a tantalum capacitor is used at the input, it must be guaranteed to have a surge current

rating sufficient for the application by the manufacture.



**Fig. 1**

### Load-Transient Considerations

The RT9182 load-transient response graphs show two components of the output response: a DC shift from the output impedance due to the load current change, and the transient response. The DC shift is quite small due to the excellent load regulation of the IC. Typical output voltage transient spike for a step change in the load current from 0mA to 50mA is tens mV, depending on the ESR of the output capacitor. Increasing the output capacitor's value and decreasing the ESR attenuates the overshoot.

### Input-Output (Dropout) Voltage

A regulator's minimum input-output voltage differential (or dropout voltage) determines the lowest usable supply voltage. In battery-powered systems, this will determine the useful end-of-life battery voltage. Because the RT9182 uses a P-channel MOSFET pass transistor, the dropout voltage is a function of drain-to-source on-resistance [ $R_{\text{DS(ON)}}$ ] multiplied by the load current.

### Reverse Current Path

The power transistor used in the RT9182 has an inherent diode connected between each regulator input and output (see Fig.2). If the output is forced above the input by more than a diode-drop, this diode will become forward biased and current will flow from the  $V_{OUT}$  terminal to  $V_{IN}$ . This diode will also be turned on by abruptly stepping the input voltage to a value below the output voltage. To prevent regulator mis-operation, a Schottky diode could be used in the applications where input/output voltage conditions can cause the internal diode to be turned on (see Fig.3). As shown, the Schottky diode is connected in parallel with the internal parasitic diode and prevents it from being turned on by limiting the voltage drop across it to about  $0.3V < 100mA$  to prevent damage to the part.

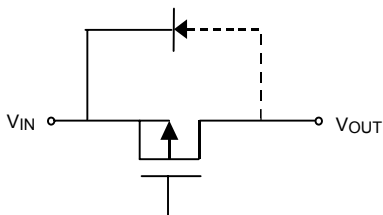


Fig. 2

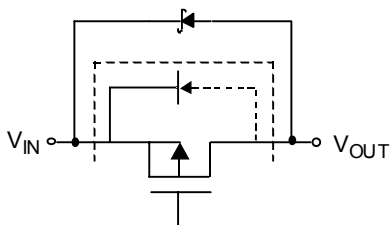
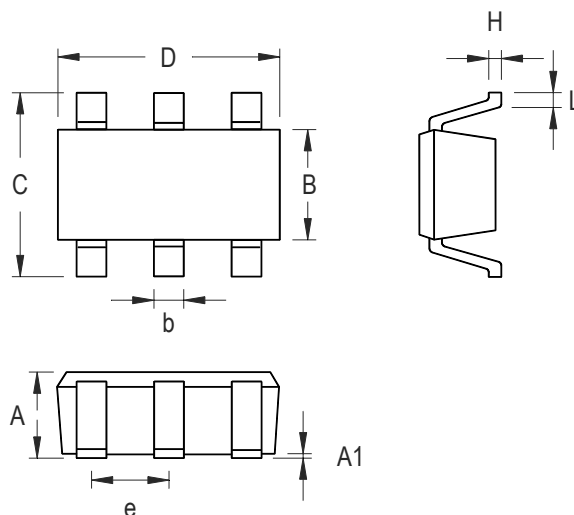


Fig. 3

**Package Information**


Symbol	Dimensions In Millimeters		Dimensions In Inches	
	Min	Max	Min	Max
A	0.889	1.295	0.035	0.051
A1	--	0.152	--	0.006
B	1.397	1.803	0.055	0.071
b	0.356	0.559	0.014	0.022
C	2.591	2.997	0.102	0.118
D	2.692	3.099	0.106	0.122
e	0.838	1.041	0.033	0.041
H	0.102	0.254	0.004	0.010
L	0.356	0.610	0.014	0.024

**SOT- 26 Surface Mount Package**

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