Devices

Am8530H

Serial Communications Controller

DISTINCTIVE CHARACTERISTICS

- Fast data rate
 - Up to 8 MHz/2 Mb/s
- Two 0 to 2 Mb/s full duplex serial channels
 - Each channel has independent oscillator, baud-rate generator, and PLL for clock recovery, dramatically reducing external components
- Programmable protocols
 - NRZ, NRZI, and FM data encoding supported under program control

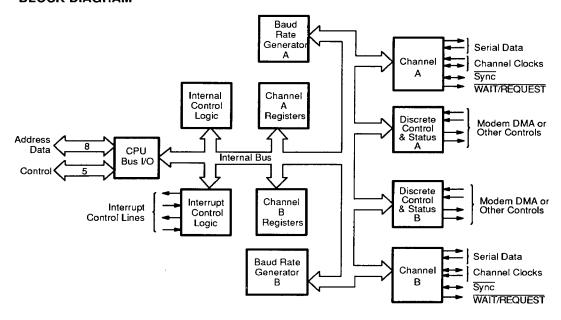
Programmable asynchronous modes

 5- to 8-bit characters with programmable stop bits, clock, break detect and error conditions

■ Programmable synchronous modes

- SDLC and HDLC and SDLC loop supported with frame control, zero insertion and deletion, abort, and residue handling. CRC-16 and CCITT generator and checkers
- Compatible with non-multiplexed bus
 - The Am8530H interfaces easily to most other CPUs

BLOCK DIAGRAM



00970F-001

Publication# 00970 Rev. G Amendment/0 Issue Date: November 1991

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GENERAL DESCRIPTION

The Serial Communications Control (SCC) is a dualchannel, multi-protocol data communications peripheral designed for use with 8- and 16-bit microprocessors. The SCC functions as a serial-to-parallel, parallel-toserial converter/controller. The SCC can be softwareconfigured to satisfy a wide variety of serial communications applications. The device contains a variety of new, sophisticated internal functions, including on-chip baud rate generators, digital phase-locked loops, and crystal oscillators, which dramatically reduce the need for external logic.

The SCC handles asynchronous formats, synchronous byte-oriented protocols, such as IBM® Bisync, and synchronous bit-oriented protocols, such as HDLC and IBM SDLC.

This versatile device supports virtually any serial data transfer application (cassette, diskette, tape drivers, etc).

The device can generate and check CRC codes in any synchronous mode and can be programmed to check data integrity in various modes. The SCC also has facilities for modem controls in both channels. In applications where these controls are not needed, the modem controls can be used for general-purpose I/O.

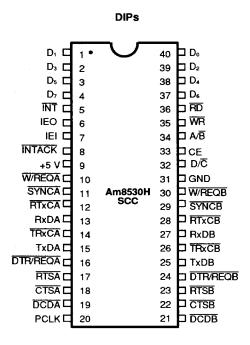
The Am8530H is designed for non-multiplexed buses and is easily interfaced with most other CPUs, such as 8080, Z80, 6800, 68000 and MULTIBUS™.

RELATED AMD PRODUCTS

Part Number	Description
Am79C12	Full Duplex 1200 bps Modem
Am7960	Coded Data Transceiver
80186	Highly Integrated 16-Bit Microprocessor
80286/80C286	High-Performance 16-Bit Microprocessor
8080A	8-Bit Microprocessor
Am9517A	DMA Controller
Am386™	High-Performance 32-Bit Microprocessor

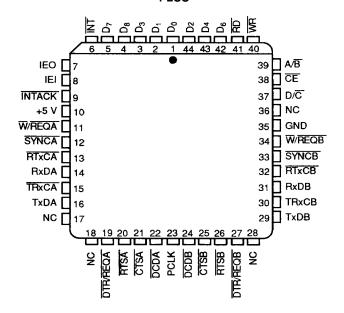
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CONNECTION DIAGRAMS Top View



00970G-028B

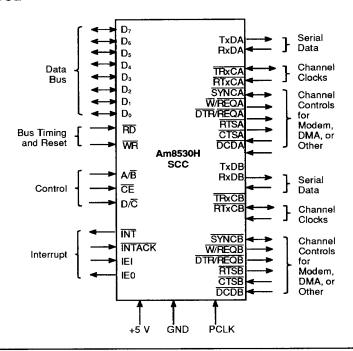
PLCC



00970G-002B

Note: Pin 1 is marked for orientation.

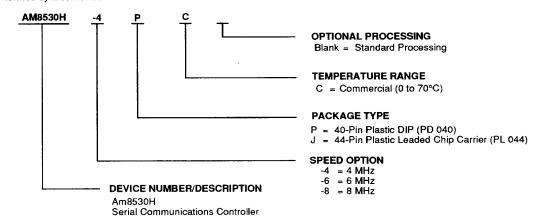
LOGIC SYMBOL



ORDERING INFORMATION

Commodity Products

AMD standard products are available in several packages and operating ranges. The order number (Valid Combination) is formed by a combination of:



Valid Combinations					
Am8530H-4					
Am8530H-6	₽C, JC				
Am8530H-8	_				

Valid Combinations

00970G-003

Valid Combinations list configurations planned to be supported in volume for this device. Consult the local AMD sales office to confirm availability of specific valid combinations, to check on newly released combinations, and to obtain additional data on AMD's standard military grade products.

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PIN DESCRIPTION

A/B

Channel A/Channel B Select (Input)

This signal selects the channel in which the Read or Write operation occurs.

CE

Chip Enable (Input; Active Low)

This signal selects the SCC for a Read or Write opera-

CTSA, CTSB

Clear to Send (Inputs; Active Low)

If these pins are programmed as Auto Enables, a Low on these inputs enables their respective transmitters. If not programmed as Auto Enables, they may be used as general-purpose inputs. Both inputs are Schmitt-trigger buffered to accommodate slow rise-time inputs. The SCC detects pulses on these inputs and may interrupt the CPU on both logic level transitions.

 D/\overline{C}

Data/Control Select (Input)

This signal defines the type of information transferred to or from the SCC. A High means data is transferred; a Low indicates a command is transferred.

DCDA, DCDB

Data Carrier Detect (Inputs; Active Low)

These pins function as receiver enables if they are programmed as Auto Enables; otherwise, they may be used as general-purpose input pins. Both are Schmittrigger buffered to accommodate slow rise-time signals. The SCC detects pulses on these pins and may interrupt the CPU on both logic level transitions.

D7-D0

Data Bus (Inputs/Outputs; Three State)

These lines carry data and commands to and from the SCC.

DTR/REQA, DTR/REQB Data Terminal Ready/Request (Outputs; Active Low)

These outputs follow the inverted state programmed into the DTR bit in WR5. They can also be used as general-purpose outputs or as Request Lines for a DMA controller.

GND

Ground

IEI

Interrupt Enable In (Input; Active High)

IEI is used with IEO to form an interrupt daisy chain when there is more than one interrupt-driven device. A High IEI indicates that no other higher priority device has an interrupt under service or is requesting an interrupt.

IEO

Interrupt Enable Out (Output; Active High)

IEO is High only if IEI is High and the CPU is not servicing an SCC interrupt or the SCC is not requesting an interrupt (interrupt acknowledge cycle only). IEO is con-

nected to the next lower priority device's IEI input and thus inhibits interrupts from lower priority devices.

INT

Interrupt Request (Output; Active Low, Open Drain)

This signal is activated when the SCC requests an interrupt.

INTACK

Interrupt Acknowledge (Input; Active Low)

This signal indicates an active interrupt acknowledge cycle. During this cycle, the SCC interrupt daisy chain settles. When $\overline{\text{RD}}$ becomes active, the SCC places an interrupt vector on the data bus (if IEI is High). $\overline{\text{INTACK}}$ is latched by the rising edge of PCLK.

PCLK

Clock (Input)

RD

Read (Input; Active Low)

This signal indicates a Read operation and, when the SCC is selected, enables the SCC's bus drivers. During the Interrupt Acknowledge cycle, this signal gates the interrupt vector onto the bus if the SCC is the highest priority device requesting an interrupt.

RTSA, RTSB

Request to Send (Outputs; Active Low)

When the Request to Send (RTS) bit in Write Register 5 is set, the RTS signal goes Low. When the RTS bit is reset in the asynchronous mode and Auto Enable is on, the signal goes High after the transmitter is empty. In SYNC mode, or in asynchronous mode with Auto Enable off, the RTS pins strictly follow the inverted state of the RTS bit. Both pins can be used as general-purpose outputs.

RTxCA RTxCB

Receive/Transmit Clocks (Inputs; Active Low)

These pins can be programmed in <u>several</u> different modes of operation. In each channel, <u>RTxC</u> may supply the receive clock, the transmit clock, the clock for the baud rate generator, or the clock of the digital phase-locked loop. These pins <u>can</u> also be programmed for use with the respective <u>SYNC</u> pins as a crystal oscillator. The receive clock may be 1, 16, 32, or 64 times the data rate in asynchronous modes.

RxDA, RxDB

Receive Data (Inputs; Active High)

These input signals receive serial data at standard TTL levels.

SYNCA, SYNCB

Synchronization (Inputs/Outputs; Active Low)

These pins can act either as inputs, outputs, or part of the crystal oscillator circuit. In the Asynchronous Receive mode (crystal oscillator option not selected), these pins are inputs similar to CTS and DCD. In this mode, transitions on these lines affect the state of the Sync/



Hunt status bits in Read Register 0 but have no other function.

In External Synchronization mode with the crystal oscillator not selected, these lines also act as inputs. In this mode, SYNC must be driven Low two receive clock cycles after the last bit in the SYNC character is received. Character assembly begins on the rising edge of the receive clock immediately preceding the activation of SYNC.

In the Internal Synchronization mode (Monosync and Bisync) with the crystal oscillator not selected, these pins act as outputs and are active only during the part of the receive clock cycle in which SYNC characters are recognized. The SYNC condition is not latched, so these outputs are active each time a SYNC pattern is recognized (regardless of character boundaries). In SDLC mode, these pins act as outputs and are valid on receipt of a flag.

TRxCA, TRxCB Transmit/Receive Clocks (Inputs/Outputs; Active Low)

These pins can be <u>programmed</u> in several different modes of operation. TRxC may supply the receive clock or the transmit clock in the input mode or supply the output of the digital phase-locked loop, the crystal oscillator, the baud rate generator, or the transmit clock in the output mode.

TxDA, TxDB Transmit Data (Outputs; Active High)

These output signals transmit serial data at standard TTL levels.

This is the master SCC clock used to synchronize internal signals. PCLK is not required to have any phase relationship with the master system clock. PCLK is a TTL-level signal. Maximum transmit rate is ¹/4 PCLK.

WR

Write (Input; Active Low)

When the SCC is selected, this signal indicates a Write operation. The coincidence of $\overline{\text{RD}}$ and $\overline{\text{WR}}$ is interpreted as a reset.

W/REQA, W/REQB

Wait/Request (Outputs; Open drain when programmed for a Wait function, driven High or Low when programmed for a Request function)

These dual-purpose outputs may be programmed as Request lines for a DMA controller or as Wait lines to synchronize the CPU to the SCC data rate. The reset state is Wait.

Vcc

+ 5-V Power Supply

Am8530H

ARCHITECTURE

The Am8530H internal structure includes two full-duplex channels, two baud rate generators, internal control and interrupt logic, and a bus interface to a non-multiplexed CPU bus. Associated with each channel are a number of read and write registers for mode control and status information, as well as logic necessary to interface with modems or other external devices (Figure 1).

The logic for both channels provides formats, synchronization, and validation for data transferred to and from the channel interface. The modem control inputs are monitored by the control logic under program control. All of the modem control signals are general-purpose in nature and can optionally be used for functions other than modem control.

The register set for each channel includes ten control (write) registers, two synchronous character (write) registers, and four status (read) registers. In addition, each baud rate generator has two (read/write) registers for holding the time constant that determines the baud rate. Finally, associated with the interrupt logic is a write register for the interrupt vector accessible through either channel, a write-only Master Interrupt Control register and three read registers: one containing the vector with status information (Channel B only), one containing the vector without status (A only), and one containing the Interrupt Pending bits (A only).

The registers for each channel are designated as follows:

WR15-WR0-Write Registers 15 through 0.

RR0-RR3, RR10, RR12, RR13, RR15—Read Registers 0 through 3, 10, 12, 13, 15.

The following table lists the functions assigned to each read or write register. The SCC contains only one WR2 and WR9, but they can be accessed by either channel. All other registers are paired (one for each channel).

Data Path

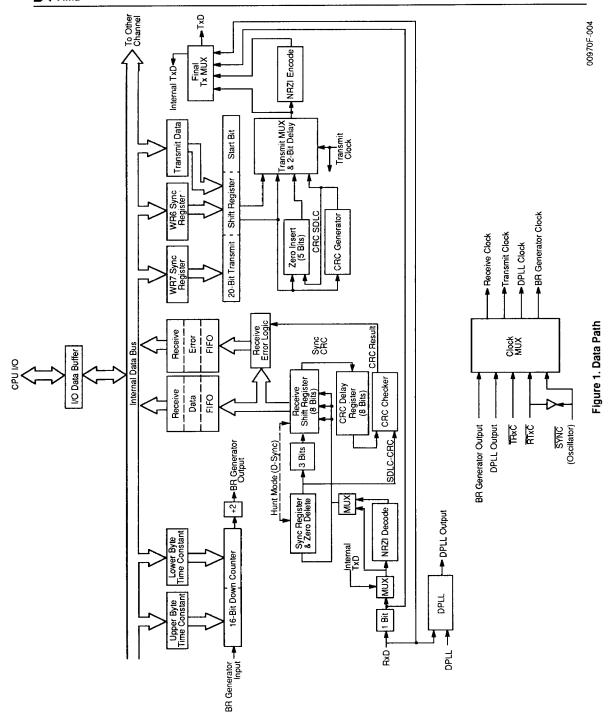
The transmit and receive data path illustrated in Figure 2 is identical for both channels. The receiver has three 8-bit buffer registers in a FIFO arrangement, in addition to the 8-bit receive shift register. This scheme creates additional time for the CPU to service an interrupt at the beginning of a block of high-speed data. Incoming data is routed through one of several paths (data or CRC) depending on the selected mode (the character length in asynchronous modes also determines the data path).

The transmitter has an 8-bit transmit data buffer register loaded from the internal data bus and a 20-bit transmit shift register that can be loaded either from the synccharacter registers or from the transmit data register. Depending on the operational mode, outgoing data is routed through one of four main paths before it is transmitted from the Transmit Data output (TxD).

Table 1. Read and Write Register Functions

Read F	Register Functions	Write Register Functions				
RR0 RR1	Transmit/receive buffer status and external status Special receive condition status	WRO	CRC initialize, initialization commands for the various modes, shift right/shift left command			
RR2	Modified interrupt vector (Channel B only)	WR1	Transmit/receive interrupt and data transfer mode definition			
	Unmodified interrupt vector (Channel A only)	WR2 WR3	Interrupt vector (accessed through either channel) Receive parameters and control			
RR3	Interrupt pending bits (Channel A only)	WR4	Transmit/receive miscellaneous parameters and modes			
RR8	Receive buffer	WR5	Transmit parameters and controls			
RR10	Miscellaneous status	WR6	Sync character or SDLC address field			
RR12	Lower byte of baud rate generator time constant	WR7	Sync character or SDLC flag			
RR13	Upper byte of baud rate generator time constant	WR8	Transmit buffer			
RR15	External/status interrupt information	WR9	Master interrupt control and reset (accessed through either channel)			
		WR10	Miscellaneous transmitter/receiver control bits			
		WR11	Clock mode control			
		WR12 WR13 WR14	Lower byte of baud rate generator time constant Upper byte of baud rate generator time constant Miscellaneous control bits			
		WR15	External/Status interrupt control			

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DETAILED DESCRIPTION

The functional capabilities of the SCC can be described from two different points of view: as a data communications device, it transmits and receives data in a wide variety of data communications protocols; as a microprocessor peripheral, it interacts with the CPU and provides vectored interrupts and handshaking signals.

Data Communications Capabilities

The SCC provides two independent full-duplex channels programmable for use in any common asynchronous or synchronous data-communication protocol. Figure 2 and the following description briefly detail these protocols.

Asynchronous Modes

Transmission and reception can be accomplished independently on each channel with five to eight bits per character, plus optional even or odd parity. The transmitters can supply one, one-and-a-half or two stop bits per character and can provide a break output at any time. The receiver break-detection logic interrupts the CPU both at the start and at the end of a received break. Reception is protected from spikes by a transient spike-rejection mechanism that checks the signal one-half a bit time after a Low level is detected on the receive data input (RxDA or RxDB in the Am8530H Logic Symbol). If the Low does not persist (as in the case of a transient), the character assembly process does not start.

Framing errors and overrun errors are detected and buffered together with the partial character on which

they occur. Vectored interrupts allow fast servicing or error conditions using dedicated routines. Furthermore, a built-in checking process avoids the interpretation of framing error as a new start bit: a framing error results in the addition of one-half a bit time to the point at which the search for the next start bit begins.

The SCC does not require symmetric transmit and receive clock signals—a feature allowing use of the wide variety of clock sources. The transmitter and receiver can handle data at a rate of 1, ½16, ½2, or ½64 of the clock rate supplied to the receive and transmit clock inputs. In asynchronous modes, the SYNC pin may be programmed as an input used for functions, such as monitoring a ring indicator.

Synchronous Modes

The SCC supports both byte-oriented and bit-oriented synchronous communication. Synchronous byte-oriented protocols can be handled in several modes, allowing character synchronization with a 6-bit or 8-bit synchronous character (Monosync), any 12-bit synchronous pattern (Bisync), or with an external synchronous signal. Leading synchronous characters can be removed without interrupting the CPU.

5- or 7-bit synchronous characters are detected with 8or 16-bit patterns in the SCC by overlapping the larger pattern across multiple incoming synchronous characters as shown in Figure 3.

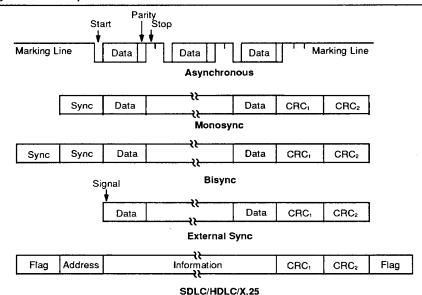


Figure 2. SCC Protocols

00970F-005



Figure 3. Detecting 5- or 7-Bit Synchronous Characters

CRC checking for Synchronous byte-oriented modes is delayed by one character time so that the CPU may disable CRC checking on specific characters. This permits the implementation of protocols, such as IBM Bisync.

Both CRC-16 ($X^{16} + X^{15} + X^2 + 1$) and CCITT ($X^{16} + X^{12} + X^5 + 1$) error checking polynomials are supported. Either polynomial may be selected in all synchronous modes. Users may preset the CRC generator and checker to all 1s or all 0s. The SCC also provides a feature that automatically transmits CRC data when no other data is available for transmission. This allows for high–speed transmissions under DMA control with no need for CPU intervention at the end of a message. When there is no data or CRC to send in synchronous modes, the transmitter inserts 6-, 8-, or 16-bit synchronous characters, regardless of the programmed character length.

The SCC supports synchronous bit-oriented protocols, such as SDLC and HDLC, by performing automatic flag sending, zero insertion, and CRC generation. A special command can be used to abort a frame in transmission. At the end of a message, the SCC automatically transmits the CRC and trailing flag when the transmitter underruns. The transmitter may also be programmed to send an idle line consisting of continuous flag characters or a steady marking condition.

If a transmit underrun occurs in the middle of a message, an external/status interrupt warns the CPU of this status change so that an abort may be issued. The SCC may also be programmed to send an abort itself in case of an underrun, reliving the CPU of this task. One to eight bits per character can be sent allowing reception of a message with no prior information about the character structure in the information field of a frame.

The receiver automatically acquires synchronization on the leading flag of a frame in SDLC or HDLC and provides a synchronization signal on the SYNC pin (an interrupt can also be programmed). The receiver can be programmed to search for frames addressed by a single byte (or four bits within a byte) of a user-selected address or to a global broadcast address. In this mode, frames not matching either the user-selected or broadcast address are ignored. The number of address bytes can be extended under software control. For receiving data, an interrupt on the first received character, or an interrupt on every character, or on special condition only

(end-of-frame) can be selected. The receiver automatically deletes all 0s inserted by the transmitter during character assembly. CRC is also calculated and is automatically checked to validate frame transmission. At the end of transmission, the status of a received frame is available in the status registers. In SDLC mode, the SCC must be programmed to use the SDLC CRC polynomial, but the generator and checker may be preset to all 1s or all 0s. The CRC is inverted before transmission and the receiver checks against the bit pattern 0001110100001111.

NRZ, NRZI or FM coding may be used in any 1X mode. The parity options available in asynchronous modes are available in synchronous modes.

The SCC can be conveniently used under DMA control to provide high-speed reception or transmission. In reception, for example, the SCC can interrupt the CPU when the first character of a message is received. The CPU then enables the DMA to transfer the message to memory. The SCC then issues an end-of-frame interrupt and the CPU can check the status of the received message. Thus, the CPU is freed for other service while the message is being received. The CPU may also enable the DMA first and have the SCC interrupt only on end-of-frame. This procedure allows all data to be transferred via the DMA.

SDLC Loop Mode

The SCC supports SDLC Loop mode in addition to normal SDLC. In an SDLC Loop, there is a primary controller station that manages the message traffic flow and any number of secondary stations. In SDLC Loop mode, the SCC performs the functions of a secondary station while an SCC operating in regular SDLC mode can act as a controller (Figure 4).

A secondary station in an SDLC Loop is always listening to the messages being sent around the loop and, in fact, must pass these messages to the rest of the loop by retransmitting them with a one-bit-time delay. The secondary station can place its own message on the loop only at specific times. The controller signals that secondary stations may transmit messages by sending a special character, called an EOP (End of Poll), around the loop. The EOP character is the bit pattern 11111110.

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Because of zero insertion during messages, this bit pattern is unique and easily recognized.

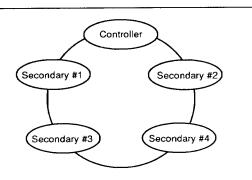


Figure 4. An SDLC Loop

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When a secondary station has a message to transmit and recognizes an EOP on the line, it changes the last binary one of the EOP to a zero before transmission. This has the effect of turning the EOP into a flag sequence. The secondary station now places its message on the loop and terminates the message with an EOP. Any secondary stations further down the loop with messages to transmit can then append their messages to the message of the first secondary station by the same process. Any secondary stations without messages to send merely echo the incoming messages and are prohibited from placing messages on the loop (except upon recognizing an EOP).

SDLC Loop mode is a programmable option in the SCC. NRZ, NRZI, and FM coding may all be used in SDLC Loop mode.

Baud Rate Generator

Each channel in the SCC contains a programmable baud rate generator. Each generator consists of two 8-bit time constant registers that form a 16-bit time constant, a 16-bit down counter, and a flip-flop on the output producing a square wave. On start-up the flip-flop on the output is set in a High state; the value in the time constant register is loaded into the counter; and the counter starts counting down. The output of the baud rate generator toggles upon reaching zero; the value in the time constant register is loaded into the counter; and the process is repeated. The time constant may be changed

at any time, but the new value does not take effect until the next load of the counter.

The output of the baud rate generator may be used as either the transmit clock, the receive clock, or both. It can also drive the digital phase-locked loop (see next section).

If the receive clock or transmit clock is not programmed to come from the \overline{TRxC} pin, the output of the baud rate generator may be echoed out via the \overline{TRxC} pin.

The following formula relates the time constant to the baud rate. (The baud rate is in bits/second and the BR clock period is in seconds.)

baud rate =
$$\frac{1}{2 \text{ (time constant + 2)} \cdot \text{(BR clock period)}}$$

Time Constant Values for Standard Baud Rates at BR Clock = 3.9936 MHz

Rate (Baud)		Time Constant (decimal/Hex notation)	
19200	102	(0066)	0
9600	206	(00CE)	0
7200	275	(0113)	0.12%
4800	414	(019E)	0
3600	553	(0229)	0.06%
2400	830	(033E)	0
2000	996	(03E4)	0.04%
1800	1107	(0453)	0.03%
1200	1662	(067E)	0
600	3326	(0CFE)	0
300	6654	(19FE)	0
150	13310	(33FE)	0
134.5	14844	(39FC)	0.0007%
110	18151	(46E7)	0.0015%
75	26622	(67FE)	0
50	39934	(98FE)	0

Digital Phase-Locked Loop

The SCC contains a digital phase-locked loop (DPLL) to recover clock information from a data stream with NRZI or FM encoding. The DPLL is driven by a clock that is nominally 32 (NRZI) or 16 (FM) times the data rate. The DPLL uses this clock, along with the data stream, to construct a clock for the data. This clock may then be used as the SCC receive clock, the transmit clock or both.

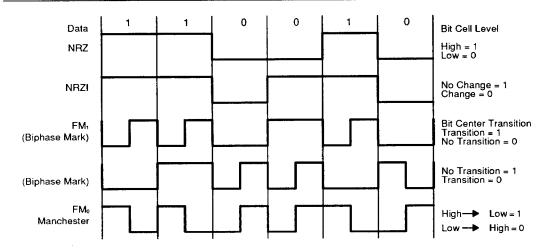


Figure 5. Data Encoding Methods

For NRZI encoding, the DPLL counts the 32X clock to create nominal bit times. As the 32X clock is counted, the DPLL is searching the incoming data stream for edges (either 1/0 or 0/1). Whenever an edge is detected, the DPLL makes a count adjustment (during the next counting cycle), producing a terminal count closer to the center of the bit cell.

For FM encoding, the DPLL still counts from 0 to 31, but with a cycle corresponding to two bit times. When the DPLL is locked, the clock edges in the data stream should occur between counts 15 and 16 and between counts 31 and 0. The DPLL looks for edges only during a time centered on the 15/16 counting transition.

The 32X clock for the DPLL can be programmed to come from either the RTxC input or the output of the baud rate generator. The DPLL output may be programmed to be echoed out of the SCC via the TRxC pin (if this pin is not being used as an input).

Data Encoding

The SCC may be programmed to encode and decode the serial data in four different ways (Figure 5). In NRZ encoding, a 1 is represented by a High level, and a 0 is represented by a Low level. In NRZI encoding, a 1 is represented by no change in level, and a 0 is represented by a change in level. In FM₁ (more properly, bi-phase mark), a transition occurs at the beginning of every bit cell. A 1 is represented by an additional transition at the center of the bit cell, and a 0 is represented by no additional transition at the center of the bit cell. In FM₀ (bi-phase space), a transition occurs at the beginning of every bit cell. A 0 is represented by an additional transition at the center of the bit cell, and a 1 is represented by no additional transition at the center of the bit cell. In ad-

dition to these four methods, the SCC can be used to decode Manchester (bi-phase level) data by using the DPLL in the FM mode and programming the receiver for NRZ data. Manchester encoding always produces a transition at the center of the bit cell. If the transition is 0/1, the bit is a 0. If the transition is 1/0, the bit is a 1.

Auto Echo and Local Loopback

The SCC is capable of automatically echoing everything it receives. This feature is useful mainly in asynchronous modes but works in synchronous and SDLC modes as well. In Auto Echo mode, TxD is RxD. Auto Echo mode can be used with NRZI or FM encoding with no additional delay, because the data stream is not decoded before retransmission. In Auto Echo mode, the CTS input is ignored as a transmitter enable (although transitions on this input can still cause interrupts if programmed to do so). In this mode, the transmitter is actually bypassed, and the programmer is responsible for disabling transmitter interrupts and WAIT/REQUEST on transmit.

The SCC is also capable of local loopback. In this mode, TxD is RxD just as in Auto Echo mode. However, in Local Loopback mode, the internal transmit data is tied to the internal receive data, and RxD is ignored (except to be echoed out via TxD). The CTS and DCD inputs are also ignored as transmit and receive enables. However, transitions on these inputs can still cause interrupts. Local Loopback works in asynchronous, synchronous and SDLC modes with NRZ, NRZI or FM coding of the data stream.

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I/O Interface Capabilities

The SCC offers the choice of Polling, Interrupt (vectored or nonvectored), and Block Transfer modes to transfer data, status, and control information to and from the CPU. The Block Transfer mode can be implemented under CPU or DMA control.

Polling

All interrupts are disabled. Three status registers in the SCC are automatically updated whenever any function is performed. For example, end-of-frame in SDLC mode sets a bit in one of these status registers. The idea behind polling is for the CPU to periodically read a status register until the register contents indicate the need for data to be transferred. Only one register needs to be read; depending on its contents, the CPU either writes data, reads data, or continues. Two bits in the register indicate the need for data transfer. An alternative is a poll of the Interrupt Pending register to determine the source of an interrupt. The status for both channels resides in one register.

Interrupts

When a SCC responds to an Interrupt Acknowledge signal (INTACK) from the CPU, an interrupt vector may be placed on the data bus. This vector is written in WR2 and may be read in RR2A or RR2B (Figures 7 and 8).

To speed interrupt response time, the SCC can modify three bits in this vector to indicate status. If the vector is read in Channel A, status is never included; if it is read in Channel B, status is always included.

Each of the six sources of interrupts in the SCC (Transmit, Receive and External/Status interrupts in both channels) has three bits associated with the interrupt source: Interrupt Pending (IP), Interrupt Under Service (IUS), and Interrupt Enable (IE). Operation of the IE bit is straightforward. If the IE bit is set for a given interrupt source, then that source can request interrupts. The exception is when the MIE (Master Interrupt Enable) bit in WR9 is reset and no interrupts may be requested. The IE bits are write-only.

The other two bits are related to the Z-Bus interrupt priority chain (Figure 6). As a Z-Bus peripheral, the SCC may request an interrupt only when no higher priority device is requesting one; e.g., when IEI is High. If the device in question requests an interrupt, it pulls down NT. The CPU then responds with NTACK, and the interrupting device places the vector on the A/D bus.

In the SCC, the IP bit signals a need for interrupt servicing. When an IP bit is 1 and the IEI input is High, the INT output is pulled Low, requesting an interrupt. In the SCC; if the IE bit is not set by enabling interrupts, then the IP for that source can never be set. The IP bits are readable in RR3A.

The IUS bits signal that an interrupt request is being serviced. If an IUS is set, all interrupt sources of lower priority in the SCC and external to the SCC are prevented from requesting interrupts. The internal interrupt sources are inhibited by the state of the internal daisy chain, while lower priority devices are inhibited by the IEO output of the SCC being pulled Low and propagated to subsequent peripherals. An IUS bit is set during an Interrupt Acknowledge cycle if there are no higher priority devices requesting interrupts.

There are three types of interrupts: Transmit, Receive and External/Status. Each interrupt type is enabled under program control with Channel A having higher priority than Channel B, and with Receive, Transmit and External/Status interrupts prioritized in that order within each channel. When the Transmit interrupt is enabled, the CPU is interrupted when the transmit buffer becomes empty. (This implies that the transmitter must have had a data character written into it so that it can become empty.) When enabled, the Receive can interrupt the CPU in one of three ways:

- Interrupt on First Receive Character or Special Receive condition.
- Interrupt on all Receive Characters or Special Receive condition.
- Interrupt on Special Receive condition only.

Interrupt on First Character or Special Condition and Interrupt on Special Condition Only are typically used with the Block Transfer mode. A Special Receive Condition is one of the following: receiver overrun, framing error in asynchronous mode, End-of-Frame in SDLC mode, and optionally, a parity error. The Special Receive Condition Interrupt is different from an ordinary receive character available interrupt only in the status placed in the vector during the Interrupt Acknowledge cycle. In Interrupt on First Receive Character, an interrupt can occur from Special Receive condition any time after the first receive character interrupt.



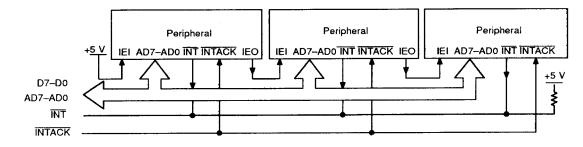


Figure 6. Z-Bus Interrupt Schedule

The main function of the External/Status Interrupt is to monitor the signal transitions of the CTS, DCD, and SYNC pins; however, an External/Status Interrupt is also caused by a Transmit Underrun condition, a zero count in the baud rate generator, the detection of a Break (asynchronous mode), Abort (SDLC mode) or EOP (SDLC Loop mode) sequence in the data stream. The interrupt caused by the Abort or EOP has a special feature allowing the SCC to interrupt when the Abort or EOP sequence is detected or terminated. This feature facilitates the proper termination of the current message, correct initialization of the next, message, and the accurate timing of the Abort condition in external logic in SDLC mode. In SDLC Loop mode, this feature allows secondary stations to recognize the wishes of the primary station to regain control of the loop during a poll sequence.

CPU/DMA Block Transfer

The SCC provides a Block Transfer mode to accommodate CPU block transfer functions and DMA controllers. The Block Transfer mode uses the WAIT/REQUEST output in conjunction with the WAIT/REQUEST bits in WR1. The WAIT/REQUEST output can be defined under software control as a WAIT line in the CPU Block Transfer mode or as a REQUEST line in the DMA Block Transfer mode.

To a DMA controller, the SCC REQUEST output indicates that the SCC is ready to transfer data to or from memory. To the CPU, the WAIT line indicates that the SCC is not ready to transfer data, thereby requesting that the CPU extend the I/O cycle. The DTR/REQUEST line allows full-duplex operation under DMA control.

PROGRAMMING INFORMATION

Each channel has 15 write registers that are individually programmed from the system bus to configure the functional personality of each channel. Each channel also has 8 read registers from which the system can read status, baud rate, or interrupt information.

In the Am8530H only the four data registers (Read or Write for Channels A and B) are directly selected by a High on the D/\overline{C} input and the appropriate levels on the \overline{RD} , \overline{WR} , and A/\overline{B} pins. All other registers are addressable indirectly by the content of Write Register 0 in conjunction with a Low on the D/\overline{C} input and the appropriate levels on the \overline{RD} , \overline{WR} , and A/\overline{B} pins. If bit D3 in WR0 is 1 and bits 5 and 6 are 0, then bits 0, 1, 2 address the higher registers 8 through 15. If bits 4, 5, 6 contain a different code, bits 0, 1, 2 address the lower registers 0 through 7 as shown in Table 2.

Writing to or reading from any register except RR0, WR0 and the Data Registers thus involves two operations.

First, write the appropriate code into WR0, then follow this by a write or read operation on the register thus specified. Bits 0 through 4 in WW0 are automatically cleared after this operation, so that WW0 then points to WR0 or RR0 again.

Channel A/Channel B selection is made by the A/D input (High = A, Low = B).

In Am8530H, the system program first issues a series of commands to initialize the basic mode of operation. This is followed by other commands to qualify conditions within the selected mode. For example, the asynchronous mode, character length, clock rate, number of stop bits, even or odd parity might be set first. Then the interrupt mode would be set and, finally, receiver or transmitter enable.

Read Registers

The SCC contains eight Read registers (actually nine, counting the receive buffer (RR8) in each channel). Four of these may be read to obtain status information (RR0, RR1, RR10, and RR15). Two registers (RR12 and RR13) may be read to learn the baud rate generator time constant. RR2 contains either the unmodified interrupt vector (Channel A) or the vector modified by status information (Channel B). RR3 contains the Interrupt

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Pending (IP) bits (Channel A). Figure 7 shows the formats for each Read register.

The status bits of RR0 and RR1 are carefully grouped to simplify status monitoring. For example, when the interrupt vector indicates a Special Receive Condition interrupt, all the appropriate error bits can be read from a single register (RR1).

ters are programmed separately to configure the functional "personality" of the channels. In addition, there are two registers (WR2 and WR9) shared by the two channels that can be accessed through either of them. WR2 contains the interrupt vector for both channels, while WR9 contains the interrupt control bits. Figure 8 shows the format of each Write register.

Write Registers

The SCC contains 15 Write registers (16 counting WR8, the transmit buffer) in each channel. These Write regis-

Table 2. Register Addressing

	D/C "Point High" Code in WR0:		D2 D1 In WR0:		Write Register	Read Register
High	Either Way	х	Х	х	Data	Data
Low	Not True	0	0	0	0	0
Low	Not True	0	0	1	1	1
Low	Not True	0	1	0	2	2
Low	Not True	0	1	1	3	3
Low	Not True	1	0	0	4	(0)
Low	Not True	i	0	1	5	(1)
Low	Not True	1	1	0	6	(2)
Low	Not True	1	1	1	7	(3)
Low	True	0	0	0	Data	Data
Low	True	0	0	1	9	-
Low	True	0	1	0	10	10
Low	True	0	1	1	11	(15)
Low	True	1	0	0	12	12
Low	True	1	0	1	13	13
Low	True	1	1	0	14	(10)
Low	True	1	1	1	15	15



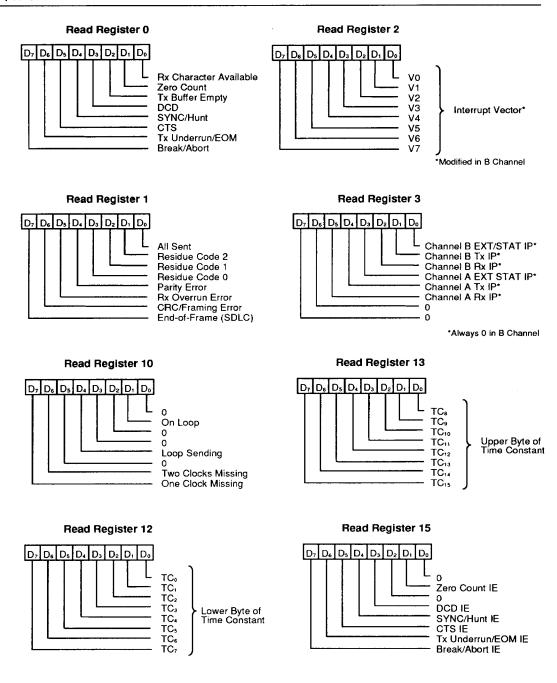


Figure 7. Read Register Bit Functions

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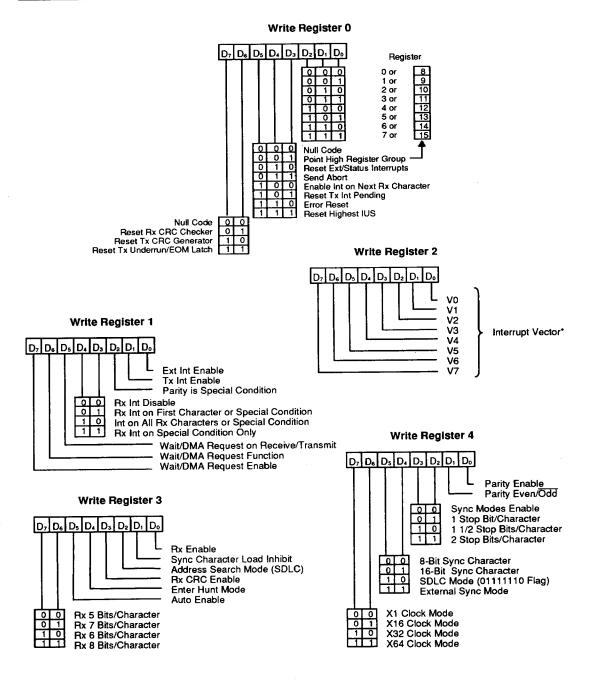
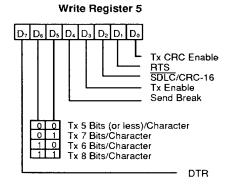


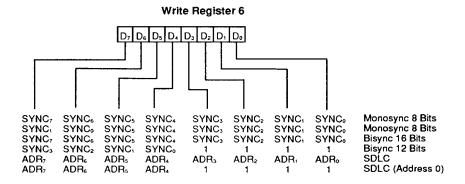
Figure 8. Write Register Bit Functions (Continued)

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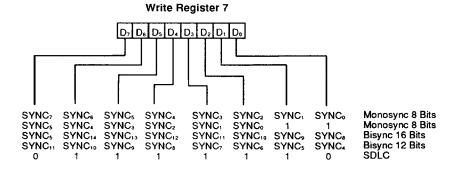
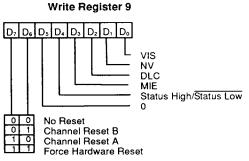


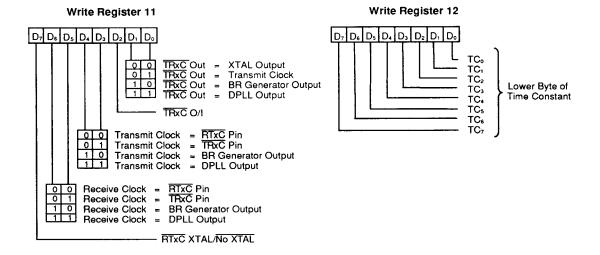
Figure 8. Write Register Bit Functions (Continued)

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^{*}Added Enhancement

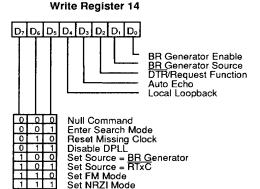


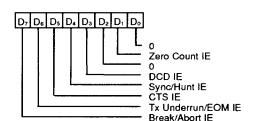
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Figure 8. Write Register Bit Functions (Continued)

Am8530H







Write Register 15

Figure 8. Write Register Bit Functions (Continued)

00970F-011

Am8530H Timing

The SCC generates internal control signals from WR and RD that are related to PCLK. Since PCLK has no phase relationship with WR and RD, the circuitry generating these internal control signals must provide time for metastable conditions to disappear. This gives rise to a recovery time related to PCLK. The recovery time applies only between bus transactions involving the SCC. The recovery time required for proper operation is specified from the rising edge of WR or RD in the first transaction involving the SCC to the falling edge of WR or RD in the second transaction involving the SCC. This time must be at least 6 PCLK cycles plus 200 ns.

Read Cycle Timing

Figure 9 illustrates Read cycle timing. Addresses on A/ \overline{B} and D/ \overline{C} and the status on INTACK must remain stable throughout the cycle. If \overline{CE} falls after \overline{RD} falls or if it rises before \overline{RD} rises, the effective \overline{RD} is shortened.

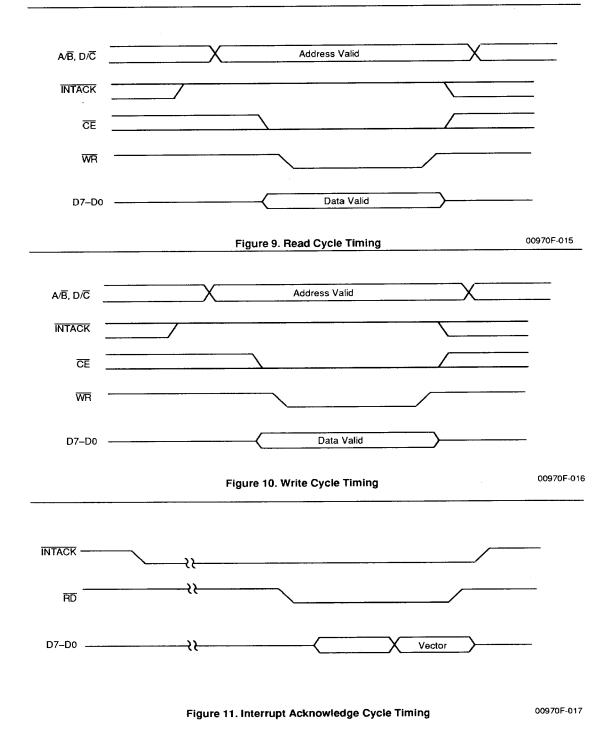
Write Cycle Timing

Figure 10 illustrates Write cycle timing. Addresses on A/B and D/C and the status on INTACK must remain stable throughout the cycle. If CE falls after WR falls or if it rises before WR rises, the effective WR is shortened.

Interrupt Acknowledge Cycle Timing

Figure 11 illustrates Interrupt Acknowledge cycle timing. Between the time $\overline{\text{INTACK}}$ goes Low and the falling edge of $\overline{\text{RD}}$, the internal and external IEI/IEO daisy chains settle. If there is an interrupt pending in the SCC and IEI is High when $\overline{\text{RD}}$ falls, the Acknowledge cycle is intended for the SCC. In this case, the SCC may be programmed to respond to $\overline{\text{RD}}$ Low by placing its interrupt vector on D7–D0; it then sets the appropriate interrupt under-service latch internally.

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ABSOLUTE MAXIMUM RATINGS

-65 to + 150°C Storage Temperature

Voltage at any Pin

-0.5 to +7.0 V Relative to Vss

Power Dissipation 1.8 W

Stresses above those listed under Absolute Maximum Ratings may cause permanent device failure. Functionality at or above these limits is not implied. Exposure to Absolute Maximum Ratings for extended periods may affect device reliability.

OPERATING RANGES

Commercial (C) Devices

Temperature (T_A)

0 to + 70°C

Supply Voltage (Vcc)

5 V ± 10%

Operating ranges define those limits between which the functionality of the device is guaranteed.

DC CHARACTERISTICS over operating range

Parameter Symbol	Parameter Description	Test Conditions	Min	Max	Unit
V _{iH}	Input High Voltage	Commercial	2.2	Vcc +0.3	V
VIL	Input Low Voltage		-0.3	0.8	V
VoH	Output High Voltage	$l_{OH} = -250 \mu A$	2.4		V
Vol	Output Low Voltage	$l_{OL} = +2.0 \text{ mA}$		0.4	V
I _{IL}	Input Leakage	$0.4 \text{ V} \leq V_{IN} \leq 2.4 \text{ V}$		±10.0	μΑ
lou	Output Leakage	$0.4 \text{ V} \leq V_{OUT} \leq 2.4 \text{ V}$		±10.0	μA
lcc	Vcc Supply Current			250	mA
CIN	Input Capacitance	Unmeasured pins returned		10	рF
Соит	Output Capacitance	to ground f = 1 MHz at		15	рF
Смо	Bidirectional Capacitance	$T_A = 25^{\circ}C$		20	рF

Standard Test Conditions

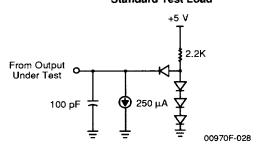
The characteristics below apply for the following standard test conditions, unless otherwise noted. All voltages are referenced to GND. Positive current flows into the referenced pin. Standard conditions are as follows:

$$+4.5 \text{ V} \le \text{V}_{CC} \le +5.5 \text{ V}$$

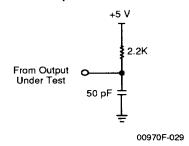
GND = 0 V
 $0^{\circ}\text{C} \le \text{T}_{A} \le 70^{\circ}\text{C}$

SWITCHING TEST CIRCUITS

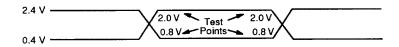
Standard Test Load



Open-Drain Test Load



Switching Test Input/Output Waveform



00970F-030

AC testing: Inputs are driven at 2.4 V for a logic 1 and 0.4 V for a logic 0. Timing measurements are made at 2.0 V for a logic 1 and 0.8 V for logic 0.

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SWITCHING CHARACTERISTICS over operating range

General Timing

		D	4 MHz		lz 6 MHz		8 MHz		
No.	Parameter Symbol	Parameter Description	Min.	Max.	Min.	Max.	Min.	Max.	Unit
1	TdPC(REQ)	PCLK ↓ to W/REQ Valid Delay		250		250		250	nş
2	TdPC(W)	PCLK ↓ to Wait Inactive Delay		350		350		350	ns
3	TsRXC(PC)	RxC↑ to PCLK↑Setup Time (Notes 1 & 4)	80	Twpcl	70	Tweck	60	Twpcl	ns
4	TsRXD(RXCr)	RxD to RxC ↑ Setup Time (XI Mode) (Note 1)	0		0		0		ns
5	ThRXD(RXCr)	RxD to RxC ↑ Hold Time (XI Mode) (Note 1)	150		150		150		ns
6	TsRXD(RXCf)	RxD to RxC ↓ Setup Time (XI Mode) (Notes 1, 5)	0		0		0		ns
7	ThRXD(RXCf)	RxD to RxC ↓ Hold Time (XI Mode) (Notes 1, 5)	150		150		150		ns
8	TsSY(RXC)	SYNC to RxC ↑ Setup Time (Note 1)	-200		-200		-200		ns
9	ThSY(RXC)	SYNC to RxC ↑ Hold Time (Note 1)	3TcPC + 400		3TcPC + 320		3TcPC + 250		ns
10	TsTXC(PC)	TxC ↓ to PCLK ↑ Setup Time (Notes 2, 4)	0		0		0		ns
11	TdTXCf(TXD)	TxC ↓ to TxD Delay (XI Mode) (Note 2)		300		230		200	ns
12	TdTXCr(TXD)	TxC ↑ to TxD Delay (XI Mode) (Notes 2, 5)		300		230		200	ns
13	TdTXD(TRX)	TxD to TRxC Delay (Send Clock Echo)		200		200		200	ns
14	TwRTXh	RTxC High Width (Note 6)	180		180		150		ns
15	TwRTXI	RTxC Low Width (Note 6)	180	<u> </u>	180		150	ļ	ns
16	TcRTX	RTxC Cycle Time (Notes 6, 7)	1000		660		500		ns
17	TcRTXX	Crystal Oscillator Period (Note 3)	250	1000	165	1000	125	1000	ns
18	TwTRXh	TRxC High Width (Note 6)	180		180		150		ns
19	TwTRXI	TRxC Low Width (Note 6)	180		180	L	150		ns
20	TcTRX	TRxC Cycle Time (Notes 6, 7)	1000		660		500		ns
21	TwEXT	DCD or CTS Pulse Width	200		200		200		ns
22	TwSY	SYNC Pulse Width	200		200		200		ns

Notes:

- RXC is RTxC or TRxC, whichever is supplying the receive clock.
 TxC is TRxC or RTxC, whichever is supplying the transmit clock.
- 3. Both RTxC and SYNC have 30-pF capacitors to ground connected to them.
- 4. Parameter applies only if the data rate is one-fourth the PCLK rate. In all other cases, no phase relationship between RxC and PCLK or TxC and PCLK is required.
- 5. Parameter applies only to FM encoding/decoding.
- 6. Parameter applies only for transmitter and receiver; DPLL and baud rate generator timing requirements are identical to chip PCLK requirements.
- 7. The maximum receive or transmit data is 1/4 PCLK.

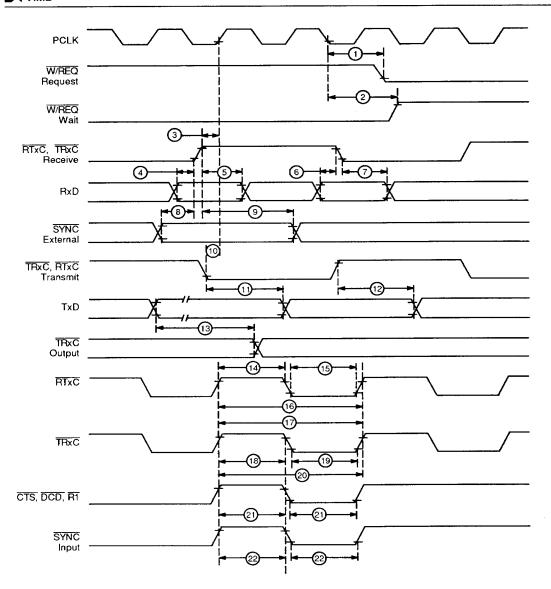


Figure 12. General Timing



SWITCHING CHARACTERISTICS over operating range (Continued) System Timing

		D		IHz	6 N	lHz	8 MHz			
No.	Parameter Symbol	Parameter Description	Min.	Мах.	Min.	Max.	Min.	Max.	Unit	
1	TdRXC(REQ)	RXC ↑ W/REQ Valid Delay (Note 2)	8	12	8	12	8	12	ТсРс	
2	TdRXC(W)	RXC ↑ to Wait Inactive Delay (Notes 1, 2)	8	14	8	14	8	14	ТсРс	
3	TdRXC(SY)	RxC 1 to SYNC Valid Delay (Note 2)	4	7	4	7	4	7	TcPc	
4	TdRXC(INT)	RxC ↑ to INT Valid Delay (Notes 1, 2)	10	16	10	16	10	16	TcPc	
5	TdTXC(REQ)	TxC ↑ to W/REQ Valid Delay (Note 3)	5	8	5	8	5	8	TcPc	
6	TdTXC(W)	TxC ↓ to Wait Inactive Delay (Notes 1, 3)	5	11	5	11	5	11	ТсРс	
7	TdTXC(DRQ)	TxC ↓ to DTR/REQ Valid Delay (Note 3)	4	7	4	7	4	7	TcPc	
8	TdTXC(INT)	TxC ↓ to INT Valid Delay (Notes 1, 3)	6	10	6	10	6	10	TcPc	
9	TdSY(INT)	SYNC Transition to INT Valid Delay (Note 1)	2	6	2	6	2	6	TcPc	
10	TdEXT(INT)	DCD or CTS Transition to INT Valid Delay (Note 1)	2	6	2	6	2	6	ТсРс	

Notes:

- 1. Open-drain, measured with open-drain test load.
- 2. RxC is RTxC or TRxC, whichever is supplying the receive clock.
- 3. TxC is TRxC or RTxC, whichever is supplying the transmit clock.

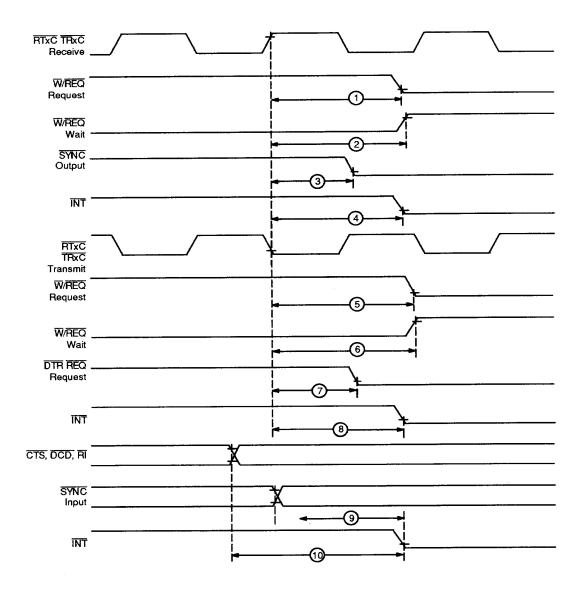


Figure 13. System Timing

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SWITCHING CHARACTERISTICS over operating rangeRead and Write Timing

			4 MHz		6 MHz		8 MHz		
No.	Parameter Symbol	Parameter Description	Min.	Max.	Min.	Max.	Min.	Max.	Unit
1	TwPCI	PCLK Low Width	105	2000	70	1000	50	1000	ns
2	TwPCh	PCLK High Width	105	2000	70	1000	50	1000	ns
3	TfPC	PCLK Fall Time		20		10		10	ns
4	TrPC	PCLK Rise Time		20		10		10	ns
5	TcPC	PCLK Cycle Time	250	4000	165	2000	125	2000	ns
6	TsA(WR)	Address to WR ↓ Set-Up Time	80		80		70		ns
7	ThA(WR)	Address to WR ↑ Hold Time	0		0		0		ns
8	TsA(RD)	Address to RD ↓ Set-Up Time	80		80		70		ns
9	ThA(RD)	Address to RD ↑ Hold Time	0		0		0		ns
10	TsIA(PC)	INTACK to PCLK 1 Set-Up Time	10		10		10		ns
11	TsAi(WR)	INTACK to WR ↓ Set-Up Time (Note 1)	200		160		145		ns
12	ThIA(WR)	INTACK to WR ↑ Hold Time	0		0		0		กร
13	TsIAi(WR)	INTACK to RD ↓ Set-Up Time (Note 1)	200		160		145		ns
14	ThIA(RD)	INTACK to RD ↑ Hold Time	0		0		0		ns
15	TsCE(WR)	CE Low to WR ↓ Set-Up Time	80		80		70		ns
16	ThCE(WR)	CE to WR ↑ Hold Time	0		0		0		ns
17	TsCEh(WR)	CE High to WR ↓ Set-Up Time	100		70		60		ns
18	TsCE(WR)	CE High to WR ↓ Set-Up Time	100		70		60		ns
19	TsCEI(RD)	CE Low to RD ↓ Set-Up Time (Note 1)	0		0		0		ns
20	ThCE(RD)	CE to RD ↑ Hold Time	0		0		0		ns
21	TsCEh(RD)	CE High to RD ↓ Set-Up Time (Note 1)	100		70		60		ns
22	TwRDI	RD Low Width (Note 1)	240		200		150		ns
23	TdRD(DRA)	RD ↓ to Read Data Active Delay	0		0		0		ns
24	TdRDr(DR)	RD ↑ to Read Data Not							
		Valid Delay	0		0		0		ns
25	TdRDf(DR)	RD ↓ to Read Data Valid Delay		250	I	180		140	ns
26	TdRD(DRz)	RD ↑ to Read Data Float Delay (Note 2)		70		45		140	ns

Notes:

- 1. Parameter does not apply to Interrupt Acknowledge transactions.
- 2. Float delay is defined as the time required for a ±0.5 V change in the output with a maximum DC load and minimum AC load.

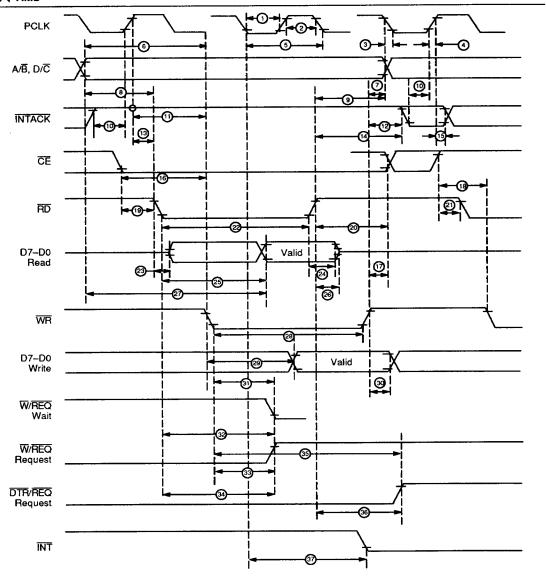


Figure 14. Read and Write Timing

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SWITCHING CHARACTERISTICS over operating range Interrupt Acknowledge Timing, Reset Timing, Cycle Timing

				MHz	61	VIHz	8 MHz		
No.	Parameter Symbol	Parameter Description	Min.	Max.	Min.	Max.	Min.	Max.	Unit
27	TdA(DR)	Address Required Valid to Read Data Valid Delay		300		280		220	ns
28	TwWRI	WR Low Width	240		200		150		ns
29	TsDW(WR)	Write Data to WR ↓ Set-Up Time	10		10		10		ns
30	ThDW(WR)	Write Data to WR ↑ Hold Time	0		0		0		ns
31	TdWR(W)	WR ↓ to Wait Valid Delay (Note 4)		240		200		170	ns
32	TdRD(W)	RD ↓ to Wait Valid Delay (Note 4)		240		200		170	ns
33	TdWRf(REQ)	WR ↓ to W/REQ Not Valid Delay		240		200		170	ns
34	TdRDf(REQ)	RD ↓ to W/REQ Not Valid Delay		240		200		170	ns
35	TdWRr(REQ)	WR ↑ DTR/REQ Not Valid Delay		4TcPC		4TcPC		4TcPC	ns
36	TdRDr(REQ)	RD ↑ DTR/REQ Not Valid Delay		4TcPC		4TcPC		4TcPC	ns
37	TdPC(INT)	PCLK ↓ to INT Valid Delay (Note 4)		500		5.00		500	ns
38	TdlAi(RD)	INTACK to RD ↓ (Acknowledge) Delay (Note 5)	250		200	·	150		ns
39	TwRDA	RD (Acknowledge) Width	250		200		150		ns
40	TdRDA(DR)	RD ↓ (Acknowledge) to Read Data Valid Delay		250		180		140	ns
41	TsIEI(RDA)	IEI to RD ↓ (Acknowledge) Set-Up Time	120		100		95		ns
42	ThIEI(RDA)	IEI to RD ↓ (Acknowledge) Hold Time	0		0		0		ns
43	TdIEI(IEO)	IEI to IEO Delay Time		120		100		95	ns
44	TdPC(IEO)	PCLK ↑ to IEO Delay		250		250		200	ns
45	TdRDA(INT)	RD ↓ to INT Inactive Delay (Note 4)		500		500		450	ns
46	TdRD(WRQ)	RD ↑ to WR ↓ Delay for No Reset	30		15		15		ns
47	TdWRQ(RD)	WR ↑ to RD ↓ Delay for No Reset	30		30		20		ns
48	TwRES	WR and RD Coincident Low for Reset	250		200		150		ns
49	Trc	Valid Access Recovery Time (Note 3)	4TcPC		4TcPC		4TcPC		ns

Notes:

- 3. Parameter applies only between transactions involving the SCC.
- 4. Open-drain output, measured with open-drain test load.
- 5. Parameter is system dependent. For any SCC in the daisy chain, TdIAI(RD) must be greater than the sum of TdPC(IEO) for the highest priority device in the daisy chain, TsIEI(RDA) for the SCC, and TdIEIf(IEO) for each device separating them in the daisy chain.

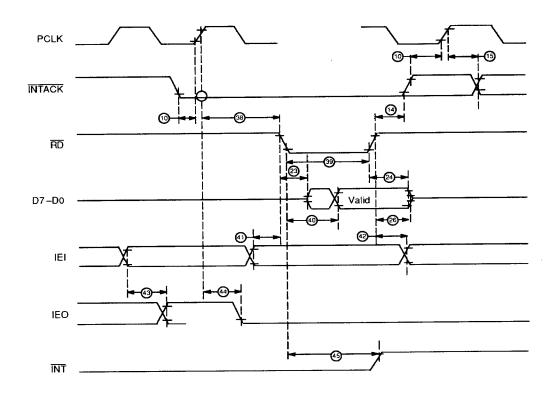


Figure 15. Interrupt Acknowledge Timing

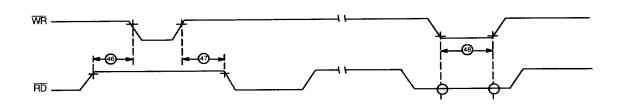


Figure 16. Reset Timing

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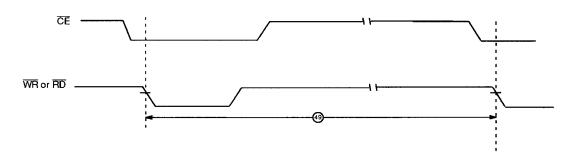


Figure 17. Cycle Timing

00970G-027