

# description

The TPS2834 and TPS2835 are MOSFET drivers for synchronous-buck power stages. These devices are ideal for designing a high-performance power supply using switching controllers that do not include on-chip MOSFET drivers. The drivers are designed to deliver minimum 2-A peak currents into large capacitive loads. The high-side driver can be configured as ground-reference or as floating-bootstrap. An adaptive dead-time control circuit eliminates shoot-through currents through the main power FETs during switching transitions, and provides high efficiency for the buck regulator. The TPS2834 and TPS2835 have additional control functions: ENABLE, SYNC, and CROWBAR. Both high-side and low-side drivers are off when ENABLE is low. The driver is configured as a nonsynchronous-buck driver disabling the low-side driver when SYNC is low. The CROWBAR function turns on the low-side power FET, overriding the IN signal, for overvoltage protection against faulted high-side power FETs.

The TPS2834 has a noninverting input, while the TPS2835 has an inverting input. These drivers are available in 14-terminal SOIC and thermally enhanced TSSOP PowerPAD<sup>TM</sup> packages and operate over a junction temperature range of  $-40^{\circ}$ C to  $125^{\circ}$ C.

### **Related Synchronous MOSFET Drivers**

DEVICE NAME	ADDITIONAL FEATURES	INPUTS		
TPS2830	ENABLE, SYNC, and CROWBAR	CMOS	Noninverted	
TPS2831	ENABLE, STING, and CROWBAR	CIVIOS	Inverted	
TPS2832	W/O ENABLE, SYNC, and CROWBAR	CMOS	Noninverted	
TPS2833	WO ENABLE, STINC, and CROWBAR	CIVIOS	Inverted	
TPS2836	W/O ENABLE, SYNC, and CROWBAR	тті	Noninverted	
TPS2837	W/O ENABLE, STING, and CROWBAR	' ' ' '	Inverted	



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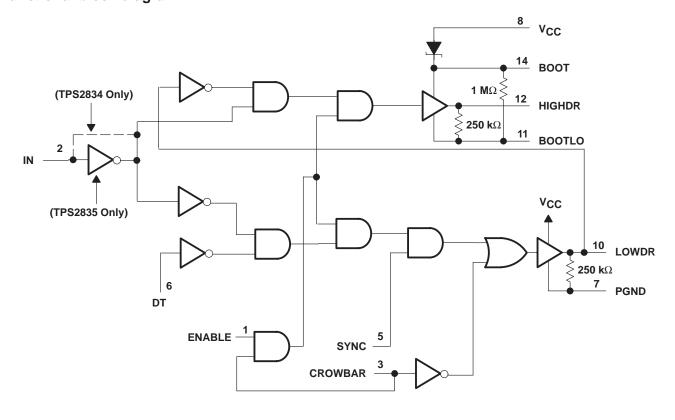


## **AVAILABLE OPTIONS**

	PACKAGED DEVICES			
ТЈ	SOIC (D)	TSSOP (PWP)		
– 40°C to 125°C	TPS2834D TPS2835D	TPS2834PWP TPS2835PWP		

The D and PWP packages are available taped and reeled. Add R suffix to device type (e.g., TPS2834DR)

# functional block diagram





### **Terminal Functions**

TERMI	NAL		DESCRIPTION
NAME	NO.	1/0	DESCRIPTION
воот	14	I	Bootstrap terminal. A ceramic capacitor is connected between BOOT and BOOTLO to develop the floating bootstrap voltage for the high-side MOSFET. The capacitor value is typically between 0.1 $\mu$ F and 1 $\mu$ F.
BOOTLO	11	0	This terminal connects to the junction of the high-side and low-side MOSFETs.
CROWBAR	3	I	CROWBAR can to be driven by an external OVP circuit to protect against a short across the high-side MOSFET. If CROWBAR is driven low, the low-side driver will be turned on and the high-side driver will be turned off, independent of the status of all other control terminals.
DT	6	ı	Dead-time control terminal. Connect DT to the junction of the high-side and low-side MOSFETs.
ENABLE	1	I	If ENABLE is low, both drivers are off.
HIGHDR	12	0	Output drive for the high-side power MOSFET
IN	2	I	Input signal to the MOSFET drivers (noninverting input for the TPS2834; inverting input for the TPS2835).
LOWDR	10	0	Output drive for the low-side power MOSFET
NC	4, 9, 13		No internal connection
PGND	7		Power ground. Connect to the FET power ground.
SYNC	5	ı	Synchronous rectifier enable terminal. If SYNC is low, the low-side driver is always off; If SYNC is high, the low-side driver provides gate drive to the low-side MOSFET.
Vcc	8	ı	Input supply. Recommended that a 1-μF capacitor be connected from V <sub>CC</sub> to PGND.

## detailed description

### low-side driver

The low-side driver is designed to drive low  $r_{DS(on)}$  N-channel MOSFETs. The current rating of the driver is 2 A, source and sink.

### high-side driver

The high-side driver is designed to drive low r<sub>DS(on)</sub> N-channel MOSFETs. The current rating of the driver is 2 A, source and sink. The high-side driver can be configured as a GND-reference driver or as a floating bootstrap driver. The internal bootstrap diode is a Schottky, for improved drive efficiency. The maximum voltage that can be applied from BOOT to ground is 30 V.

## dead-time (DT) control

Dead-time control prevents shoot-through current from flowing through the main power FETs during switching transitions by controlling the turnon times of the MOSFET drivers. The high-side driver is not allowed to turn on until the gate drive voltage to the low-side FET is low, and the low-side driver is not allowed to turn on until the voltage at the junction of the power FETs (Vdrain) is low; the TTL-compatible DT terminal connects to the junction of the power FETs.

### **ENABLE**

The ENABLE terminal enables the drivers. When enable is low, the output drivers are low. ENABLE is a TTL-compatible digital terminal.

### IN

The IN terminal is a TTL-compatible digital terminal that is the input control signal for the drivers. The TPS2834 has a noninverting input; the TPS2835 has an inverting input.



# TPS2834, TPS2835 SYNCHRONOUS-BUCK MOSFET DRIVERS WITH DEAD-TIME CONTROL

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## detailed description (continued)

### SYNC

The SYNC terminal controls whether the drivers operate in synchronous or nonsynchronous mode. In synchronous mode, the low-side FET is operated as a synchronous rectifier. In nonsynchronous mode, the low-side FET is always off. SYNC is a TTL-compatible digital terminal.

### **CROWBAR**

The CROWBAR terminal overrides the normal operation of the driver. When CROWBAR is low, the low-side FET turns on to act as a clamp, protecting the output voltage of the dc/dc converter against overvoltages due to a short across the high-side FET. VIN should be fused to protect the low-side FET. CROWBAR is a TTL-compatible digital terminal.

# absolute maximum ratings over operating free-air temperature (unless otherwise noted)

Supply voltage range, V <sub>CC</sub> (see Note 1)	0.3 V to 16 V
Input voltage range: BOOT to PGND (high-side driver ON)	
BOOTLO to PGND	0.3 V to 16 V
BOOT to BOOTLO	0.3 V to 16 V
ENABLE, SYNC, and CROWBAR	0.3 V to 16 V
IN	0.3 V to 16 V
DT	0.3 V to 30 V
Continuous total power dissipation	. See Dissipation Rating Table
Operating virtual junction temperature range, T <sub>.1</sub>	–40°C to 125°C
Storage temperature range, T <sub>stq</sub>	65°C to 150°C
Lead temperature soldering 1,6 mm (1/16 inch) from case for 10 seconds	260°C

<sup>†</sup> Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

NOTE 1: Unless otherwise specified, all voltages are with respect to PGND.

### **DISSIPATION RATING TABLE**

PACKAGE	$T_{\mbox{\scriptsize A}} \leq 25^{\circ} \mbox{\scriptsize C}$	DERATING FACTOR	T <sub>A</sub> = 70°C	T <sub>A</sub> = 85°C
PWP with solder‡	2668	26.68 mW/°C	1467	1067
PWP without solder‡	1024	10.24 mW/°C	563	409
D	749	7.49 mW/°C	412	300

### JUNCTION-CASE THERMAL RESISTANCE TABLE

PWP	Junction-case thermal resistance	2.07 °C/W
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<sup>&</sup>lt;sup>‡</sup> Test Board Conditions:

- 1. Thickness: 0.062"
- 2. 3" × 3" (for packages <27 mm long)
- 3.  $4'' \times 4''$  (for packages >27 mm long)
- 4. 2-oz copper traces located on the top of the board (0.071 mm thick)
- 5. Copper areas located on the top and bottom of the PCB for soldering
- 6. Power and ground planes, 1-oz copper (0.036 mm thick)
- 7. Thermal vias, 0.33 mm diameter, 1.5 mm pitch
- 8. Thermal isolation of power plane

For more information, refer to TI technical brief literature number SLMA002.



# recommended operating conditions

		MIN	NOM MAX	UNIT
Supply voltage,	Vcc	4.5	15	V
Input voltage	BOOT to PGND	4.5	28	V

electrical characteristics over recommended operating virtual junction temperature range,  $V_{CC}$  = 6.5 V, ENABLE = High,  $C_L$  = 3.3 nF (unless otherwise noted)

## supply current

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Vcc	Supply voltage range		4.5		15	V
		V <sub>(ENABLE)</sub> = LOW, V <sub>CC</sub> =15 V			100	
		V(ENABLE) = HIGH, V <sub>CC</sub> =15 V		300	400	μΑ
VCC	Quiescent current	$ \begin{array}{lll} V(\text{ENABLE}) = \text{HIGH}, & \text{V}_{\text{CC}} = 12 \text{ V}, \\ f(\text{SWX}) = 200 \text{ kHz}, & \text{BOOTLO grounded}, \\ C(\text{HIGHDR}) = 50 \text{ pF}, & C(\text{LOWDR}) = 50 \text{ pF}, \\ \text{See Note 2} \end{array} $		3		mA

NOTE 2: Ensured by design, not production tested.

electrical characteristics over recommended operating virtual junction temperature range,  $V_{CC} = 6.5 \text{ V}$ , ENABLE = High,  $C_L = 3.3 \text{ nF}$  (unless otherwise noted) (continued)

## output drivers

	PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT	
		Duty cycle < 2%, t <sub>pw</sub> < 100 µs (see Note 2)	V(BOOT) - V(BOOTLO) = 4.5 V, V(HIGHDR) = 4 V	0.7	1.1			
	High-side sink (see Note 3)		V(BOOT) - V(BOOTLO) = 6.5 V, V(HIGHDR) = 5 V	1.1	1.5		Α	
		(666 116.6 2)	V(BOOT) - V(BOOTLO) = 12 V, V(HIGHDR) = 10.5 V	2	2.4			
			V(BOOT) - V(BOOTLO) = 4.5 V, V(HIGHDR) = 0.5V	1.2	1.4			
Peak output current	High-side source (see Note 3)	Duty cycle < 2%, t <sub>pw</sub> < 100 μs (see Note 2)	V(BOOT) - V(BOOTLO) = 6.5 V, V(HIGHDR) = 1.5 V	1.3	1.6		А	
		(666 11616 2)	V(BOOT) - V(BOOTLO) = 12 V, V(HIGHDR) = 1.5 V	2.3	2.7			
		Duty cycle < 2%,	$V_{CC} = 4.5 \text{ V}, V_{(LOWDR)} = 4 \text{ V}$	1.3	1.8			
	Low-side sink (see Note 3)	t <sub>pw</sub> < 100 μs	V <sub>CC</sub> = 6.5 V, V <sub>(LOWDR)</sub> = 5 V	2	2.5		A	
		(see Note 2)	$V_{CC} = 12 \text{ V}, V_{(LOWDR)} = 10.5 \text{ V}$	3	3.5			
	Low-side source (see Note 3)	Duty cycle < 2%,	V <sub>CC</sub> = 4.5 V, V <sub>LOWDR</sub> )) = 0.5V	1.4	1.7			
		t <sub>pw</sub> < 100 μs (see Note 2)	V <sub>CC</sub> = 6.5 V, V <sub>(LOWDR))</sub> = 1.5 V	2	2.4			
			V <sub>CC</sub> = 12 V, V <sub>(LOWDR0)</sub> = 1.5 V	2.5	3			
			V(BOOT) - V(BOOTLO) = 4.5 V, V(HIGHDR) = 0.5 V			5		
	High-side sink (see Note 3)		V(BOOT) - V(BOOTLO) = 6.5 V, V(HIGHDR) = 0.5 V			5	Ω	
			V(BOOT) - V(BOOTLO) = 12 V, V(HIGHDR) = 0.5 V			5		
			V(BOOT) - V(BOOTLO) = 4.5 V, V(HIGHDR) = 4 V			75		
Output resistance	High-side source (see Note 3	s)	V(BOOT) - V(BOOTLO) = 6.5 V, V(HIGHDR)= 6 V			75	Ω	
			V(BOOT) - V(BOOTLO) = 12 V, V(HIGHDR) =11.5 V			75		
			$V_{(DRV)} = 4.5 \text{ V}, V_{(LOWDR)} = 0.5 \text{ V}$			9		
	Low-side sink (see Note 3)		V(DRV) = 6.5  V, V(LOWDR) = 0.5  V			7.5	Ω	
			$V_{(DRV)} = 12 \text{ V}, V_{(LOWDR)} = 0.5 \text{ V}$			6		
			$V_{(DRV)} = 4.5 \text{ V}, V_{(LOWDR)} = 4 \text{ V}$			75		
	Low-side source (see Note 3)	)	V <sub>(DRV)</sub> = 6.5 V, V <sub>(LOWDR)</sub> = 6 V			75	Ω	
			$V_{(DRV)} = 12 \text{ V}, V_{(LOWDR)} = 11.5 \text{ V}$			75		

NOTES: 2: Ensured by design, not production tested.

3. The pullup/pulldown circuits of the drivers are bipolar and MOSFET transistors in parallel. The peak output current rating is the combined current from the bipolar and MOSFET transistors. The output resistance is the r<sub>DS(on)</sub> of the MOSFET transistor when the voltage on the driver output is less than the saturation voltage of the bipolar transistor.



electrical characteristics over recommended operating virtual junction temperature range,  $V_{CC}$  = 6.5 V, ENABLE = High,  $C_L$  = 3.3 nF (unless otherwise noted) (continued)

## dead-time control

PARAMETER			TEST CONDITIONS	MIN	TYP	MAX	UNIT
٧ <sub>IH</sub>	High-level input voltage	LOWDR	Over the Valar range (see Note 2)	0.7V <sub>CC</sub>			V
$V_{IL}$	Low-level input voltage	LOWDK	Over the V <sub>CC</sub> range (see Note 2)			1	V
٧ <sub>IH</sub>	High-level input voltage	DT	Over the Vele range	2			V
V <sub>IL</sub>	Low-level input voltage	וטו	Over the V <sub>CC</sub> range			1	V

NOTE 2: Ensured by design, not production tested.

## digital control terminals (IN, CROWBAR, SYNC, ENABLE)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
VIH	High-level input voltage	Over the Vee range	2			V
$V_{IL}$	Low-level input voltage	Over the V <sub>CC</sub> range			1	V

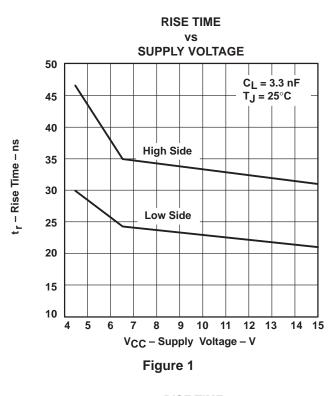
# switching characteristics over recommended operating virtual junction temperature range, ENABLE = High, $C_L$ = 3.3 nF (unless otherwise noted)

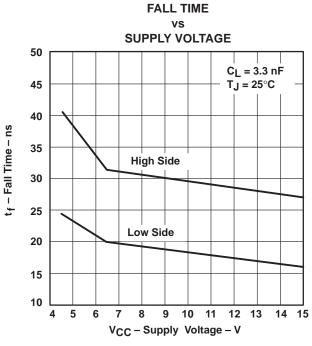
PA	RAMETER	TEST CONDITIONS	MIN	TYP MAX	UNIT	
		$V_{(BOOT)} = 4.5 \text{ V},  V_{(BOOTLO)} = 0 \text{ V}$		60		
	HIGHDR output (see Note 2)	V(BOOT) = 6.5  V,  V(BOOTLO) = 0  V		50	ns	
Rise time		$V_{(BOOT)} = 12 \text{ V},  V_{(BOOTLO)} = 0 \text{ V}$		50		
		V <sub>CC</sub> = 4.5 V		40		
	LOWDR output (see Note 2)	V <sub>CC</sub> = 6.5 V		30	ns	
		V <sub>CC</sub> = 12 V		30		
i		$V_{(BOOT)} = 4.5 \text{ V},  V_{(BOOTLO)} = 0 \text{ V}$		50	]	
	HIGHDR output (see Note 2)	$V_{(BOOT)} = 6.5 \text{ V},  V_{(BOOTLO)} = 0 \text{ V}$		40	ns	
Fall time		$V_{(BOOT)} = 12 \text{ V},  V_{(BOOTLO)} = 0 \text{ V}$		40		
	LOWDR output (see Note 2)	V <sub>CC</sub> = 4.5 V		40		
		V <sub>CC</sub> = 6.5 V		30	ns	
		V <sub>CC</sub> = 12 V		30		
	HIGHDR going low (excluding dead time) (see Note 2)	$V_{(BOOT)} = 4.5 \text{ V},  V_{(BOOTLO)} = 0 \text{ V}$		95	ns	
		$V_{(BOOT)} = 6.5 \text{ V},  V_{(BOOTLO)} = 0 \text{ V}$		80		
Propagation delay time		$V_{(BOOT)} = 12 \text{ V},  V_{(BOOTLO)} = 0 \text{ V}$		70		
Propagation delay time	LOWDR going high (excluding	$V_{(BOOT)} = 4.5 \text{ V},  V_{(BOOTLO)} = 0 \text{ V}$		80		
	dead time) (see Note 2)	$V_{(BOOT)} = 6.5 \text{ V},  V_{(BOOTLO)} = 0 \text{ V}$		70	ns	
		$V_{(BOOT)} = 12 \text{ V},  V_{(BOOTLO)} = 0 \text{ V}$		60		
		V <sub>CC</sub> = 4.5 V		80		
Propagation delay time	LOWDR going low (excluding dead time) (see Note 2)	V <sub>CC</sub> = 6.5 V		70	ns	
	dead time; (see Note 2)	V <sub>CC</sub> = 12 V		60	]	
	DT / LOWDD LLOWDD	V <sub>CC</sub> = 4.5 V	40	170		
Driver nonoverlap time	DT to LOWDR and LOWDR to HIGHDR (see Note 2)	V <sub>CC</sub> = 6.5 V	25	135	ns	
	THISTIDK (See Note 2)	V <sub>CC</sub> = 12 V	15	85		

NOTE 2: Ensured by design, not production tested.

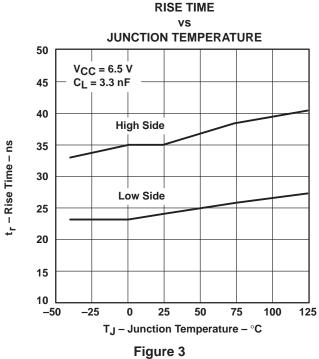


### TYPICAL CHARACTERISTICS









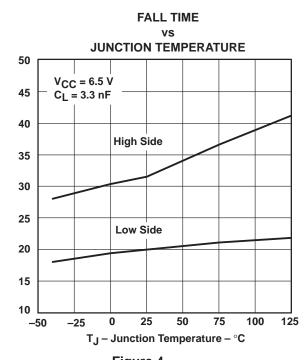


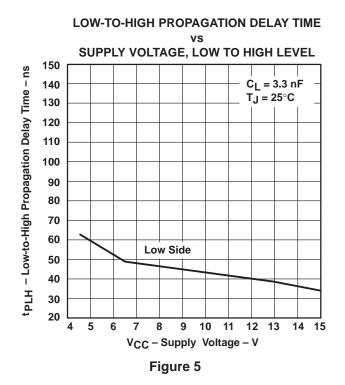
Figure 4

tf - Fall Time - ns

**HIGH-TO-LOW PROPAGATION DELAY TIME** 

SUPPLY VOLTAGE, HIGH TO LOW LEVEL

### TYPICAL CHARACTERISTICS



#### tpHL - High-to-Low Propagation Delay Time - ns $C_L = 3.3 \text{ nF}$ T<sub>J</sub> = 25°C **High Side** Low Side 14 15 V<sub>CC</sub> - Supply Voltage - V Figure 6

## JUNCTION TEMPERATURE PLH - Low-to-High Propagation Delay Time - ns $V_{CC} = 6.5 \text{ V}$ $C_L = 3.3 \text{ nF}$ **High Side** Low Side

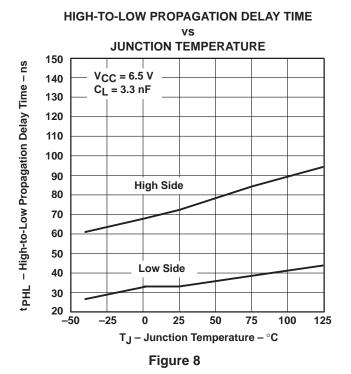
Figure 7

T<sub>J</sub> - Junction Temperature - °C

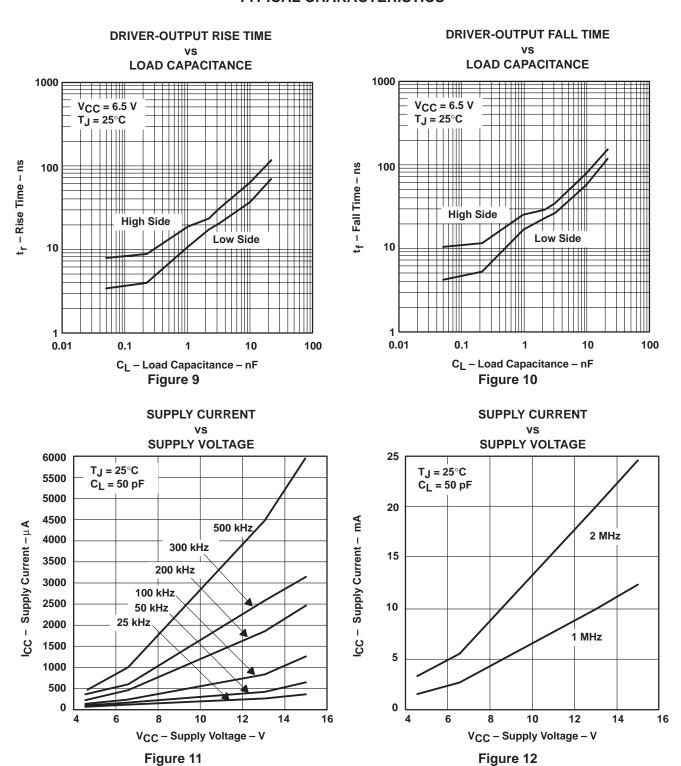
-50

-25

**LOW-TO-HIGH PROPAGATION DELAY TIME** 

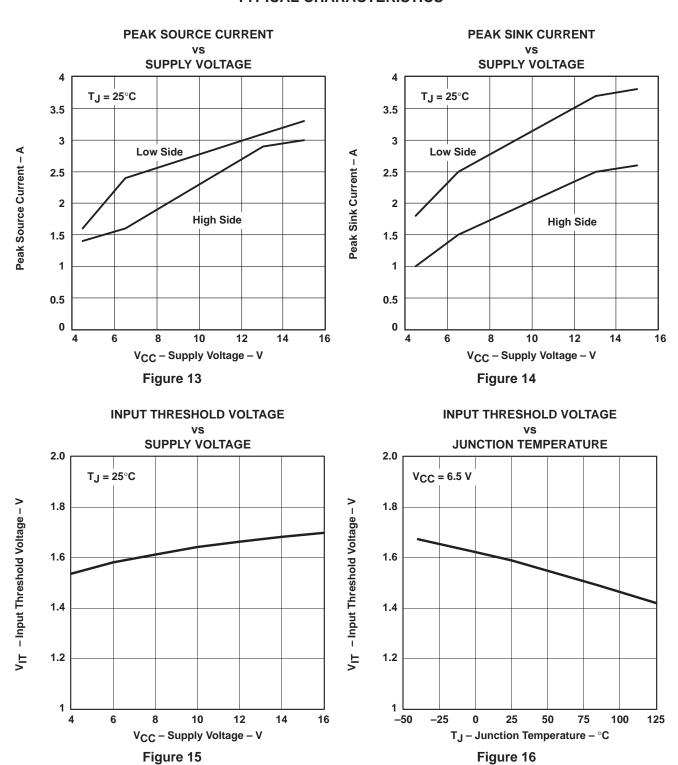


### TYPICAL CHARACTERISTICS





## **TYPICAL CHARACTERISTICS**



### **APPLICATION INFORMATION**

Figure 17 shows the circuit schematic of a 100-kHz synchronous-buck converter implemented with a TL5001A pulse-width-modulation (PWM) controller and a TPS2835 driver. The converter operates over an input range from 4.5 V to 12 V and has a 3.3-V output. The circuit can supply 3 A continuous load. The converter achieves an efficiency of 94% for  $V_{IN} = 5$  V,  $I_{load} = 1$  A, and 93% for  $V_{IN} = 5$  V,  $I_{load} = 3$  A.

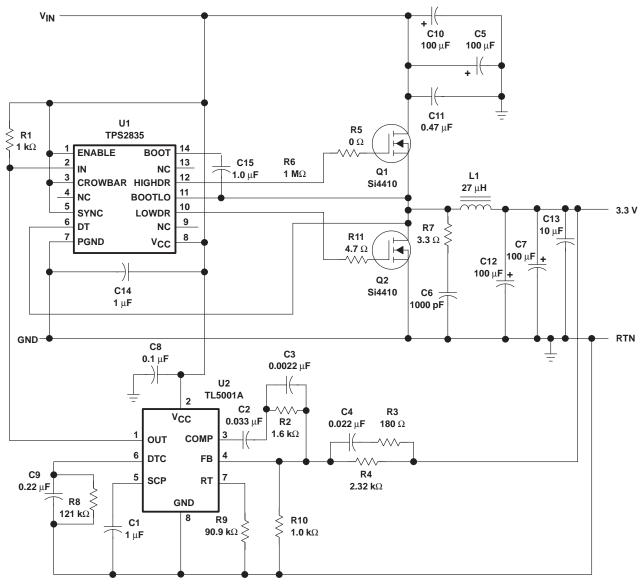


Figure 17. 3.3-V 3-A Synchronous-Buck Converter Circuit



# TPS2834, TPS2835 SYNCHRONOUS-BUCK MOSFET DRIVERS WITH DEAD-TIME CONTROL

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### APPLICATION INFORMATION

Great care should be taken when laying out the PC board. The power-processing section is the most critical and will generate large amounts of EMI if not properly configured. The junction of Q1, Q2, and L1 should be very tight. The connection from Q1 drain to the positive sides of C5, C10, and C11 and the connection from Q2 source to the negative sides of C5, C10, and C11 should be as short as possible. The negative terminals of C7 and C12 should also be connected to Q2 source.

Next, the traces from the MOSFET driver to the power switches should be considered. The BOOTLO signal from the junction of Q1 and Q2 carries the large gate drive current pulses and should be as heavy as the gate drive traces. The bypass capacitor (C14) should be tied directly across  $V_{CC}$  and PGND.

The next most sensitive node is the FB node on the controller (terminal 4 on the TL5001A). This node is very sensitive to noise pickup and should be isolated from the high-current power stage and be as short as possible. The ground around the controller and low-level circuitry should be tied to the power ground as the output. If these three areas are properly laid out, the rest of the circuit should not have other EMI problems and the power supply will be relatively free of noise.

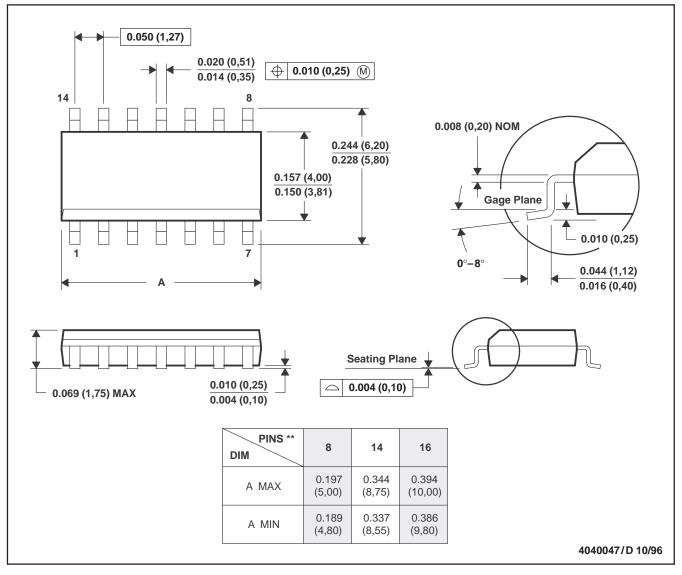


### **MECHANICAL DATA**

## D (R-PDSO-G\*\*)

### 14 PIN SHOWN

### PLASTIC SMALL-OUTLINE PACKAGE



NOTES: A. All linear dimensions are in inches (millimeters).

B. This drawing is subject to change without notice.

C. Body dimensions do not include mold flash or protrusion, not to exceed 0.006 (0,15).

D. Falls within JEDEC MS-012

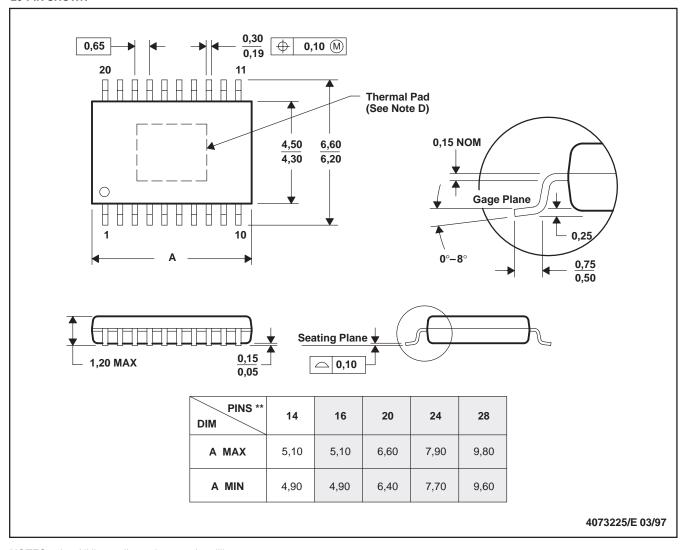


### **MECHANICAL DATA**

## PWP (R-PDSO-G\*\*)

## PowerPAD™ PLASTIC SMALL-OUTLINE PACKAGE

### **20-PIN SHOWN**



## NOTES: A. All linear dimensions are in millimeters.

- B. This drawing is subject to change without notice.
- C. Body dimensions do not include mold flash or protrusions.
- D. The package thermal performance may be enhanced by bonding the thermal pad to an external thermal plane. This pad is electrically and thermally connected to the backside of the die and possibly selected leads.
- E. Falls within JEDEC MO-153

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