## TPS76615, TPS76625, TPS76627 TPS76628, TPS76630, TPS76633, TPS76650, TPS76601 ULTRA LOW QUIESCENT CURRENT 250-mA LOW-DROPOUT VOLTAGE REGULATORS

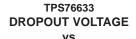
SLVS237 - AUGUST 1999

- 250-mA Low-Dropout Voltage Regulator
- Available in 1.5-V, 1.8-V, 2.5-V, 2.7-V, 2.8-V, 3.0-V, 3.3-V, 5.0-V Fixed Output and Adjustable Versions
- Dropout Voltage to 140 mV (Typ) at 250 mA (TPS76650)
- Ultra-Low 35-μA Typical Quiescent Current
- 3% Tolerance Over Specified Conditions for Fixed-Output Versions
- Open Drain Power Good
- 8-Pin SOIC Package
- Thermal Shutdown Protection

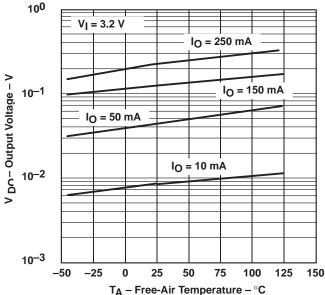
### description

This device is designed to have an ultra-low quiescent current and be stable with a 4.7-μF capacitor. This combination provides high performance at a reasonable cost.

Because the PMOS device behaves as a low-value resistor, the dropout voltage is very low (typically 230 mV at an output current of 250 mA for the TPS76650) and is directly proportional to the output current. Additionally, since the PMOS pass element is a voltage-driven device, the quiescent current is very low and independent of output loading (typically 35  $\mu$ A over the full range of output current, 0 mA to 250 mA). These two key specifications yield a significant improvement in operating life for battery-powered systems. This LDO family also features a sleep mode; applying a TTL high signal to  $\overline{\text{EN}}$  (enable) shuts down the regulator, reducing the quiescent current to less than 1  $\mu$ A (typ).



## FREE-AIR TEMPERATURE



### TPS76633 GROUND CURRENT

D PACKAGE (TOP VIEW)

NC/FB

PG

GND [] 3

EN

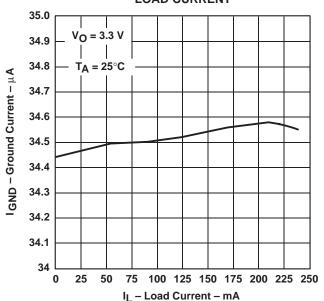
OUT

OUT

6 **∏** IN

5 **|** IN

#### vs LOAD CURRENT





Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.



#### description (continued)

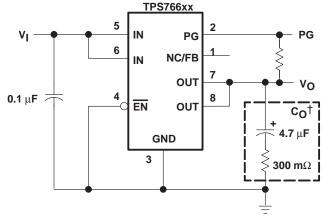
Power good (PG) is an active high output, which can be used to implement a power-on reset or a low-battery indicator.

The TPS766xx is offered in 1.5-V, 1.8-V, 2.5-V, 2.7-V, 2.8-V, 3.0-V, 3.3-V and 5.0-V fixed-voltage versions and in an adjustable version (programmable over the range of 1.25 V to 5.5 V). Output voltage tolerance is specified as a maximum of 3% over line, load, and temperature ranges. The TPS766xx family is available in 8 pin SOIC package.

#### **AVAILABLE OPTIONS**

AVAILABLE OF HORO						
Τ.	OUTPUT VOLTAGE (V)	PACKAGED DEVICES				
TJ	ТҮР	SOIC (D)				
	5.0	TPS76650D				
	3.3	TPS76633D				
	3.0	TPS76630D				
	2.8	TPS76628D				
-40°C to 125°C	2.7	TPS76627D				
10 0 10 120 0	2.5	TPS76625D				
	1.8	TPS76618D				
	1.5	TPS76615D				
	Adjustable 1.25 V to 5.5 V	TPS76601D				

The TPS76601 is programmable using an external resistor divider (see application information). The D package is available taped and reeled. Add an R suffix to the device type (e.g., TPS76601DR).

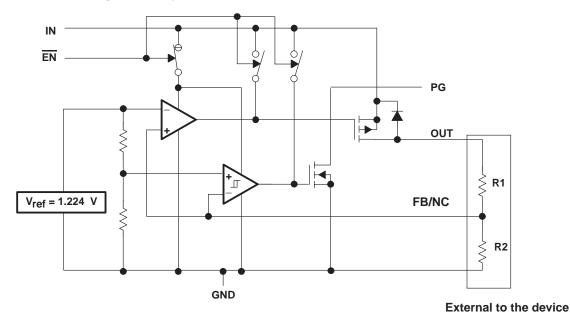


<sup>†</sup> See application information section for capacitor selection details.

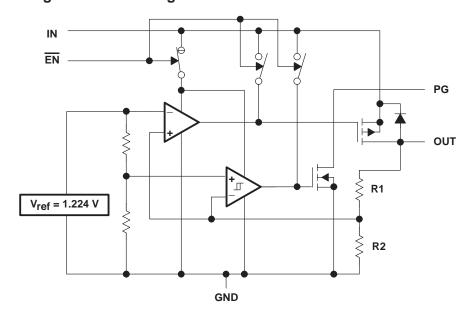
Figure 1. Typical Application Configuration for Fixed Output Options



## functional block diagram—adjustable version



## functional block diagram—fixed-voltage version



## TPS76615, TPS76618, TPS76625, TPS76627 TPS76628, TPS76630, TPS76633, TPS76650, TPS76601 **ULTRA LOW QUIESCENT CURRENT 250-mA LOW-DROPOUT VOLTAGE REGULATORS**

SLVS237 - AUGUST 1999

#### **Terminal Functions – SOIC Package**

TERMIN	IAL	1/0	DESCRIPTION			
NAME	NO.	1/0	DESCRIPTION			
EN	4	I	Enable input			
FB/NC	1	I	Feedback input voltage for adjustable device (no connect for fixed options)			
GND	3		Regulator ground			
IN	5	I	Input voltage			
IN	6	I	Input voltage			
OUT	7	0	Regulated output voltage			
OUT	8	0	Regulated output voltage			
PG	2	0	PG output			

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Input voltage range <sup>‡</sup> , V <sub>I</sub> Voltage range at EN	0.3 V to 16.5 V
Maximum PG voltage	
Peak output current	•
Continuous total power dissipation	
Output voltage, V <sub>O</sub> (OUT, FB)	
Operating virtual junction temperature range, T <sub>J</sub>	
Storage temperature range, T <sub>stg</sub>	
ESD rating, HBM	2 kV

<sup>†</sup> Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

#### **DISSIPATION RATING TABLE 1 – FREE-AIR TEMPERATURES**

PACKAGE	AIR FLOW (CFM)	T <sub>A</sub> < 25°C POWER RATING	DERATING FACTOR ABOVE T <sub>A</sub> = 25°C	T <sub>A</sub> = 70°C POWER RATING	T <sub>A</sub> = 85°C POWER RATING
D	0	568 mW	5.68 mW/°C	312 mW	227 mW
	250	904 mW	9.04 mW/°C	497 mW	361 mW

#### recommended operating conditions

	MIN	MAX	UNIT
Input voltage, V <sub>I</sub> ☆	2.7	10	V
Output voltage range, VO	1.2	5.5	V
Output current, IO (Note 1)	0	250	mA
Operating virtual junction temperature, T <sub>J</sub> (Note 1)	-40	125	°C

★ To calculate the minimum input voltage for your maximum output current, use the following equation:  $V_{I(min)} = V_{O(max)} + V_{DO(max load)}$ . NOTE 1: Continuous current and operating junction temperature are limited by internal protection circuitry, but it is not recommended that the device operate under conditions beyond those specified in this table for extended periods of time.



<sup>‡</sup> All voltage values are with respect to network terminal ground.

# electrical characteristics over recommended operating free-air temperature range, V<sub>i</sub> = V<sub>O(typ)</sub> + 1 V, I<sub>O</sub> = 10 $\mu$ A, EN = 0 V, C<sub>O</sub> = 4.7 $\mu$ F (unless otherwise noted)

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT	
		TPS76601	$5.5 \text{ V} \ge \text{V}_{\text{O}} \ge 1.25 \text{ V},$	T <sub>J</sub> = 25°C		٧o		
		11 370001	$5.5 \text{ V} \ge \text{V}_{\text{O}} \ge 1.25 \text{ V},$	$T_J = -40^{\circ}\text{C} \text{ to } 125^{\circ}\text{C}$	0.97V <sub>O</sub>		1.03V <sub>O</sub>	
		TPS76615	T <sub>J</sub> = 25°C,	$2.7 \text{ V} < \text{V}_{IN} < 10 \text{ V}$		1.5		
		11 370013	$T_J = -40^{\circ}\text{C} \text{ to } 125^{\circ}\text{C},$	$2.7 \text{ V} < \text{V}_{1N} < 10 \text{ V}$	1.455		1.545	
		TPS76618	$T_J = 25^{\circ}C$ ,	$2.8 \text{ V} < \text{V}_{1N} < 10 \text{ V}$		1.8		
		17370010	$T_J = -40^{\circ}\text{C} \text{ to } 125^{\circ}\text{C},$	$2.8 \text{ V} < \text{V}_{1N} < 10 \text{ V}$	1.746		1.854	
		TPS76625	T <sub>J</sub> = 25°C,	$3.5 \text{ V} < \text{V}_{IN} < 10 \text{ V}$		2.5		
		11 370023	$T_J = -40^{\circ}\text{C} \text{ to } 125^{\circ}\text{C},$	$3.5 \text{ V} < \text{V}_{IN} < 10 \text{ V}$	2.425		2.575	
Output vol	tage 250 mA load)	TPS76627	T <sub>J</sub> = 25°C,	3.7 V < V <sub>IN</sub> < 10 V		2.7		V
(see Note		11 370027	$T_J = -40^{\circ}\text{C} \text{ to } 125^{\circ}\text{C},$		2.619		2.781	V
,		TPS76628	$T_J = 25^{\circ}C$ ,	$3.8 \text{ V} < \text{V}_{1N} < 10 \text{ V}$		2.8		
		11 070020	$T_J = -40^{\circ}\text{C} \text{ to } 125^{\circ}\text{C},$	3.8 V < V <sub>IN</sub> < 10 V	2.716		2.884	
		TPS76630	T <sub>J</sub> = 25°C,	$4.0 \text{ V} < \text{V}_{1N} < 10 \text{ V}$		3.0		
		11 370030	$T_J = -40^{\circ}\text{C} \text{ to } 125^{\circ}\text{C},$	$4.0 \text{ V} < \text{V}_{1N} < 10 \text{ V}$	2.910		3.090	
		TPS76633	T <sub>J</sub> = 25°C,	$4.3 \text{ V} < \text{V}_{1N} < 10 \text{ V}$		3.3		
		17 37 0033	$T_J = -40^{\circ}\text{C} \text{ to } 125^{\circ}\text{C},$	$4.3 \text{ V} < \text{V}_{1N} < 10 \text{ V}$	3.201		3.399	
		TPS76650	T <sub>J</sub> = 25°C,	$6.0 \text{ V} < \text{V}_{\text{IN}} < 10 \text{ V}$		5.0		
		11 370030	$T_J = -40^{\circ}\text{C} \text{ to } 125^{\circ}\text{C},$	$6.0 \text{ V} < \text{V}_{1N} < 10 \text{ V}$	4.850		5.150	
	current (GND current)		$10 \mu\text{A} < I_{\mbox{O}} < 250 \text{mA},$	$T_J = 25^{\circ}C$		35		μΑ
. ,	see Note 2)		$I_O = 250 \text{ mA},$	$T_J = -40^{\circ}\text{C} \text{ to } 125^{\circ}\text{C}$			50	μΛ
Output voltage line regulation (ΔV <sub>O</sub> /V <sub>O</sub> ) (see Notes 2 and 3)		$V_0 + 1 V < V_1 \le 10 V$	T <sub>J</sub> = 25°C		0.01		%/V	
Load regu	lation		$I_0 = 10 \mu\text{A} \text{ to } 250 \text{mA}$			0.5%		
Output noi	se voltage		BW = 300 Hz to 50 kH $C_O = 4.7 \mu F$ ,	z, T <sub>J</sub> = 25°C		200		μVrms
Output cur	rrent Limit		V <sub>O</sub> = 0 V			0.8	1.2	Α
Thermal s	hutdown junction temperat	ure				150		°C
			EN = V <sub>I</sub> ,	T <sub>J</sub> = 25°C, 2.7 V < V <sub>I</sub> < 10 V		1		μА
Standby c	urrent		EN = V <sub>I</sub> ,	$T_J = -40^{\circ}\text{C to } 125^{\circ}\text{C}$ 2.7 V < V <sub>I</sub> < 10 V			10	μΑ
FB input c	urrent	TPS76601	FB = 1.5 V			2		nA
High level	enable input voltage	•			2.0			V
Low level	enable input voltage						0.8	V
Power supply ripple rejection (see Note 2)		f = 1 kHz, I <sub>O</sub> = 10 μA,	$C_{O} = 4.7 \mu\text{F},$ $T_{J} = 25^{\circ}\text{C}$		63		dB	
	Minimum input voltage	Minimum input voltage for valid PG				1.1		V
	Trip threshold voltage		I <sub>O</sub> (PG) = 300μA V <sub>O</sub> decreasing		92		98	%Vo
PG	Hysteresis voltage		Measured at VO			0.5		%Vo
	Output low voltage	, ,		I <sub>O(PG)</sub> = 1mA		0.15	0.4	V
	Leakage current		$V_{I} = 2.7 \text{ V},$ $V_{(PG)} = 5 \text{ V}$	-(· -)			1	μА
			EN = 0 V		-1	0	1	6
Input curre	ent (EN)		EN = V <sub>I</sub>		<u>'</u> _1		1	μΑ
		v				'		

NOTE: 2. Minimum IN operating voltage is 2.7 V or  $V_{O(typ)}$  + 1 V, whichever is greater. Maximum IN voltage 10 V.



## TPS76615, TPS76618, TPS76625, TPS76627 TPS76628, TPS76630, TPS76633, TPS76650, TPS76601 ULTRA LOW QUIESCENT CURRENT 250-mA LOW-DROPOUT VOLTAGE REGULATORS

SLVS237 – AUGUST 1999

## electrical characteristics over recommended operating free-air temperature range, $V_i = V_{O(typ)} + 1 \text{ V}$ , $I_O = 10 \text{ }\mu\text{A}$ , EN = 0 V, $C_O = 4.7 \text{ }\mu\text{F}$ (unless otherwise noted) (continued)

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
	TPS76628	I <sub>O</sub> = 250 mA,	T <sub>J</sub> = 25°C		310		
		I <sub>O</sub> = 250 mA,	$T_J = -40^{\circ}\text{C} \text{ to } 125^{\circ}\text{C}$			540	
	TPS76630	I <sub>O</sub> = 250 mA,	T <sub>J</sub> = 25°C		270		mV
Dropout voltage		I <sub>O</sub> = 250 mA,	$T_{J} = -40^{\circ}\text{C} \text{ to } 125^{\circ}\text{C}$			470	
(See Note 4)	TPS76633	$I_O = 250 \text{ mA},$	T <sub>J</sub> = 25°C		230		
		I <sub>O</sub> = 250 mA,	$T_J = -40^{\circ}\text{C} \text{ to } 125^{\circ}\text{C}$			400	
		I <sub>O</sub> = 250 mA,	T <sub>J</sub> = 25°C		140		
		I <sub>O</sub> = 250 mA,	$T_{J} = -40^{\circ}\text{C} \text{ to } 125^{\circ}\text{C}$			250	

NOTES: 3. If  $V_0 \le 1.8 \text{ V}$  then  $V_{imin} = 2.7 \text{ V}$ ,  $V_{imax} = 10 \text{ V}$ :

Line Reg. (mV) = 
$$(\%/V) \times \frac{V_O(V_{imax} - 2.7 \text{ V})}{100} \times 1000$$

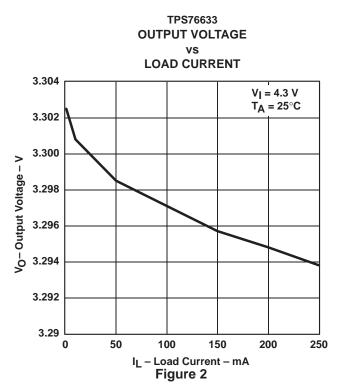
If  $V_O \ge 2.5 \text{ V}$  then  $V_{imin} = V_O + 1 \text{ V}$ ,  $V_{imax} = 10 \text{ V}$ :

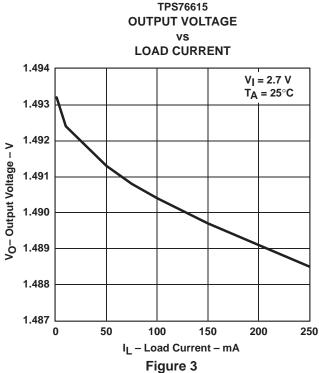
Line Reg. (mV) = 
$$(\%/V) \times \frac{V_O(V_{imax} - (V_O + 1 V))}{100} \times 1000$$

 IN voltage equals V<sub>O</sub>(Typ) – 100 mV; TPS76601 output voltage set to 3.3 V nominal with external resistor divider. TPS76615, TPS76618, TPS76625, and TPS76627 dropout voltage limited by input voltage range limitations (i.e., TPS76630 input voltage needs to drop to 2.9 V for purpose of this test).

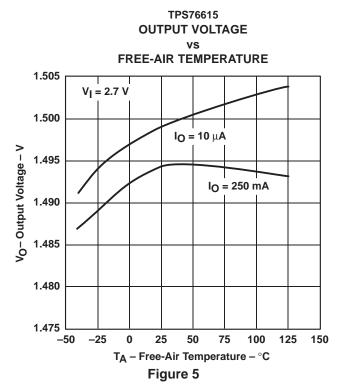
#### **Table of Graphs**

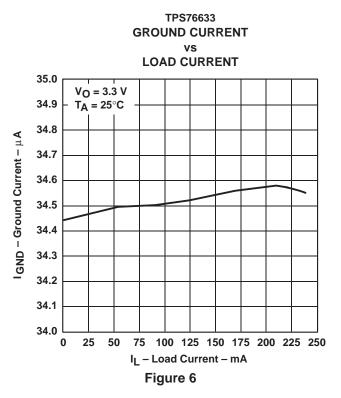
		FIGURE
Output voltage	vs Load current	2, 3
Output voltage	vs Free-air temperature	4, 5
	vs Load current	6, 7
Ground current	vs Free-air temperature	8, 9
Power supply ripple rejection	vs Frequency	10
Output spectral noise density	vs Frequency	11
Output impedance	vs Frequency	12
Dropout voltage	vs Free-air temperature	13, 14
Line transient response		15, 17
Load transient response		16, 18
Output voltage	vs Time	19
Dropout voltage	vs Input voltage	20
Equivalent series resistance (ESR)	vs Output current	21 – 24
Equivalent series resistance (ESR)	vs Added ceramic capacitance	25, 26

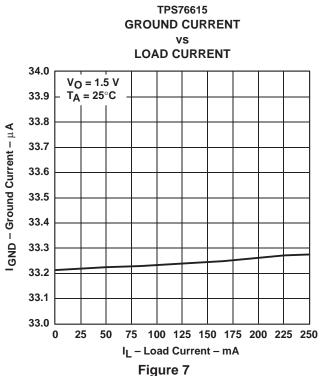


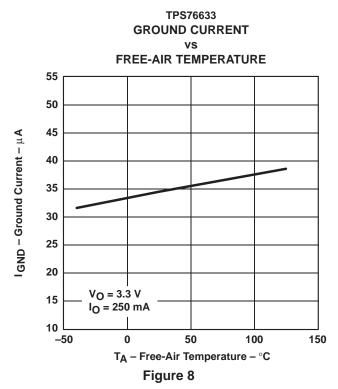


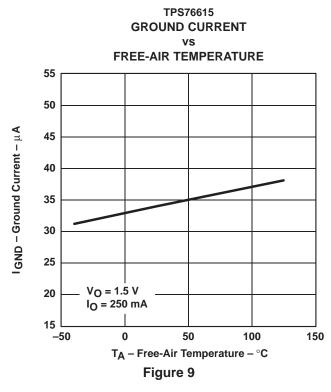
**TPS76633 OUTPUT VOLTAGE** vs FREE-AIR TEMPERATURE 3.310  $I_0 = 10 \mu A$  $V_{I} = 4.3 V$ 3.305 3.300 Vo-Output Voltage - V 3.295  $I_0 = 250 \text{ mA}$ 3.290 3.285 3.280 3.275 3.270 3.265 100 125 150 -50 -25 50 T<sub>A</sub> - Free-Air Temperature - °C Figure 4



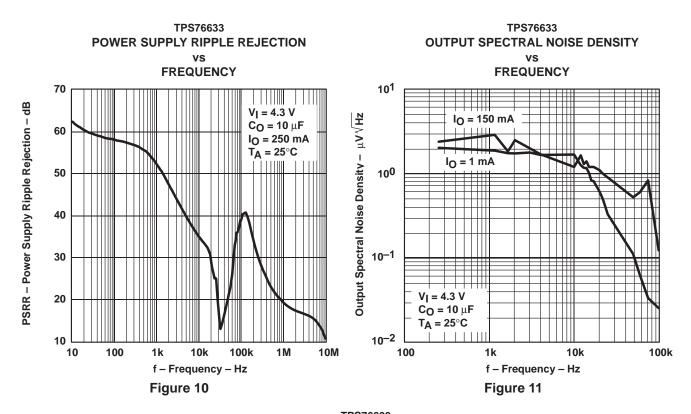


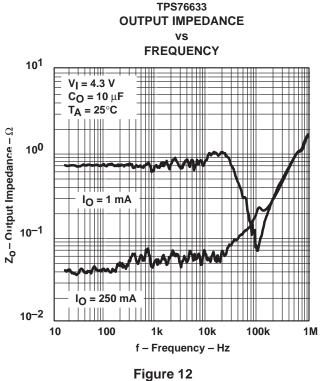














#### TYPICAL CHARACTERISTICS

#### **TPS76650 TPS76633 DROPOUT VOLTAGE DROPOUT VOLTAGE** vs vs FREE-AIR TEMPERATURE FREE-AIR TEMPERATURE 100 100 $V_{I} = 4.9 V$ $V_{I} = 3.2 \text{ V}$ I<sub>O</sub> = 250 mA $I_0 = 250 \text{ mA}$ V DO-Output Voltage - V V DO-Output Voltage-V 10-1 10-1 $I_0 = 150 \text{ mA}$ I<sub>O</sub> = 150 mA $I_O = 50 \text{ mA}$ I<sub>O</sub> = 50 mA I<sub>O</sub> = 10 mA 10-2 I<sub>O</sub> = 10 mA 10-3 10-3 125 -25 -50 -25 25 50 100 150 -50 50 75 100 125 150 T<sub>A</sub> - Free-Air Temperature - °C T<sub>A</sub> – Free-Air Temperature – °C Figure 13 Figure 14 TPS76615 **TPS76615** LINE TRANSIENT RESPONSE LOAD TRANSIENT RESPONSE 400 △VO – Change in Output Voltage – mV ∆VO – Change in Output Voltage – mV $C_{L} = 4.7 \, \mu F$ 100 200 $T_A = 25^{\circ}C$ 50 0 $C_L = 4.7 \,\mu\text{F}$ 0 -200 $T_A = 25^{\circ}C$ -50 -400 IO - Output Current - mA V<sub>I</sub> - Input Voltage - V 250 3.7 2.7 0 100 200 300 400 500 600 700 800 900 1000 100 200 300 400 500 600 700 800 900 1000 t – Time – μs $t - Time - \mu s$



Figure 15

Figure 16

#### TYPICAL CHARACTERISTICS

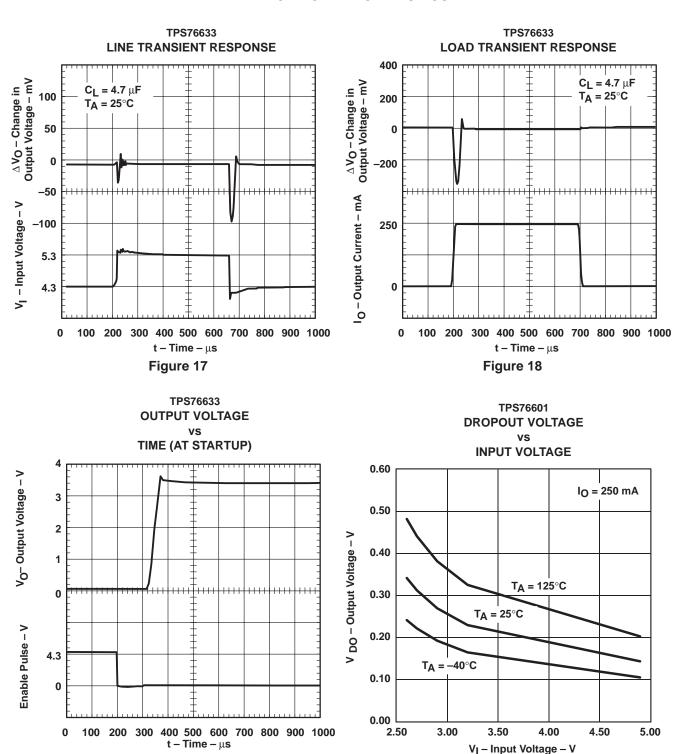




Figure 19

Figure 20

#### TYPICAL CHARACTERISTICS

#### TYPICAL REGION OF STABILITY TYPICAL REGION OF STABILITY EQUIVALENT SERIES RESISTANCE<sup>†</sup> **EQUIVALENT SERIES RESISTANCE**<sup>†</sup> vs vs **OUTPUT CURRENT OUTPUT CURRENT** 102 102 ESR – Equivalent Series Resistance – $\Omega$ **Maximum ESR** ESR – Equivalent Series Resistance – $\Omega$ **Maximum ESR** Region of Instability Region of Instability 101 101 Region of Stability Region of Stability $V_{I} = 4.3 \text{ V}$ 10<sup>0</sup> 100 $C_0 = 4.7 \, \mu F$ $V_0 = 3.3 \text{ V}$ $T_A = 25^{\circ}C$ $V_{I} = 4.3 V$ $C_0 = 4.7 \, \mu F$ Minimum ESR Minimum ESR $V_0 = 3.3 \text{ V}$ 10-1 10-1 T<sub>A</sub> = 125°C Region of Instability Region of Instability 10-2 10-2 0 50 100 150 200 250 0 50 100 150 200 250 IO - Output Current - mA IO - Output Current - mA Figure 21 Figure 22 TYPICAL REGION OF STABILITY TYPICAL REGION OF STABILITY **EQUIVALENT SERIES RESISTANCE**<sup>†</sup> **EQUIVALENT SERIES RESISTANCE**<sup>†</sup> vs vs **OUTPUT CURRENT OUTPUT CURRENT** 102 102 Maximum ESR **Maximum ESR** ESR – Equivalent Series Resistance – $\Omega$ ESR – Equivalent Series Resistance – $\Omega$ Region of Instability Region of Instability 101 101 Region of Stability 100 100 $V_{I} = 4.3 V$ $V_{I} = 4.3 \text{ V}$ Region of Stability $C_0 = 10 \, \mu F$ $C_0 = 10 \mu F$ $V_{O} = 3.3 \text{ V}$ $V_0 = 3.3 \text{ V}$ T<sub>A</sub> = 25°C $T_{\Delta} = 125^{\circ}C$ 10-1 10-1 Minimum ESR Minimum ESR Region of Instability Region of Instability 10-2 10-2 0 50 100 150 200 250 50 100 200 0 150 250 IO - Output Current - mA IO - Output Current - mA

Figure 24

Figure 23



SLVS237 - AUGUST 1999

<sup>†</sup> Equivalent series resistance (ESR) refers to the total series resistance, including the ESR of the capacitor, any series resistance added externally, and PWB trace resistance to Co.

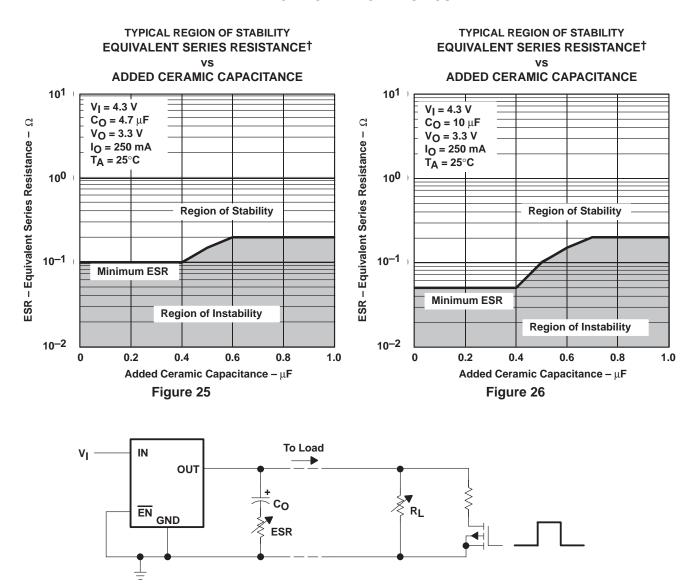


Figure 27. Test Circuit for Typical Regions of Stability (Figures 21 through 24) (Fixed Output Options)

<sup>†</sup> Equivalent series resistance (ESR) refers to the total series resistance, including the ESR of the capacitor, any series resistance added externally, and PWB trace resistance to Co.



## TPS76615, TPS76618, TPS76625, TPS76627 TPS76628, TPS76630, TPS76633, TPS76650, TPS76601 ULTRA LOW QUIESCENT CURRENT 250-mA LOW-DROPOUT VOLTAGE REGULATORS

SLVS237 - AUGUST 1999

#### **APPLICATION INFORMATION**

The TPS766xx family includes eight fixed-output voltage regulators (1.5 V, 1.8 V, 2.5 V, 2.7 V, 2.8 V, 3.0 V, 3.3 V, and 5.0 V), and an adjustable regulator, the TPS76601 (adjustable from 1.25 V to 5.5 V).

#### device operation

The TPS766xx features very low quiescent current, which remains virtually constant even with varying loads. Conventional LDO regulators use a pnp pass element, the base current of which is directly proportional to the load current through the regulator ( $I_B = I_C/\beta$ ). The TPS766xx uses a PMOS transistor to pass current; because the gate of the PMOS is voltage driven, operating current is low and invariable over the full load range.

Another pitfall associated with the pnp-pass element is its tendency to saturate when the device goes into dropout. The resulting drop in  $\beta$  forces an increase in  $I_B$  to maintain the load. During power up, this translates to large start-up currents. Systems with limited supply current may fail to start up. In battery-powered systems, it means rapid battery discharge when the voltage decays below the minimum required for regulation. The TPS766xx quiescent current remains low even when the regulator drops out, eliminating both problems.

The TPS766xx family also features a shutdown mode that places the output in the high-impedance state (essentially equal to the feedback-divider resistance) and reduces quiescent current to 1  $\mu$ A (typ). If the shutdown feature is not used,  $\overline{\text{EN}}$  should be tied to ground. Response to an enable transition is quick; regulated output voltage is reestablished in typically 160  $\mu$ s.

#### minimum load requirements

The TPS766xx family is stable even at zero load; no minimum load is required for operation.

#### FB - pin connection (adjustable version only)

The FB pin is an input pin to sense the output voltage and close the loop for the adjustable option . The output voltage is sensed through a resistor divider network to close the loop as it is shown in Figure 29. Normally, this connection should be as short as possible; however, the connection can be made near a critical circuit to improve performance at that point. Internally, FB connects to a high-impedance wide-bandwidth amplifier and noise pickup feeds through to the regulator output. Routing the FB connection to minimize/avoid noise pickup is essential.

#### external capacitor requirements

An input capacitor is not usually required; however, a ceramic bypass capacitor (0.047  $\mu$ F or larger) improves load transient response and noise rejection if the TPS766xx is located more than a few inches from the power supply. A higher-capacitance electrolytic capacitor may be necessary if large (hundreds of milliamps) load transients with fast rise times are anticipated.

Like all low dropout regulators, the TPS766xx requires an output capacitor connected between OUT and GND to stabilize the internal control loop. The minimum recommended capacitance value is 4.7  $\mu$ F and the ESR (equivalent series resistance) must be between 300-m $\Omega$  and 20- $\Omega$ . Capacitor values 4.7  $\mu$ F or larger are acceptable, provided the ESR is less than 20  $\Omega$ . Solid tantalum electrolytic, aluminum electrolytic, and multilayer ceramic capacitors are all suitable, provided they meet the requirements described previously.



#### **APPLICATION INFORMATION**

#### external capacitor requirements (continued)

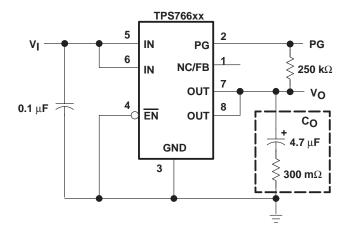


Figure 28. Typical Application Circuit (Fixed Versions)

#### programming the TPS76601 adjustable LDO regulator

The output voltage of the TPS76601 adjustable regulator is programmed using an external resistor divider as shown in Figure 29. The output voltage is calculated using:

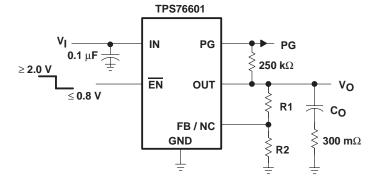
$$V_{O} = V_{ref} \times \left(1 + \frac{R1}{R2}\right) \tag{1}$$

Where

 $V_{ref} = 1.224 \text{ V typ (the internal reference voltage)}$ 

Resistors R1 and R2 should be chosen for approximately 7- $\mu$ A divider current. Lower value resistors can be used but offer no inherent advantage and waste more power. Higher values should be avoided as leakage currents at FB increase the output voltage error. The recommended design procedure is to choose R2 = 169 k $\Omega$  to set the divider current at 7  $\mu$ A and then calculate R1 using:

$$R1 = \left(\frac{V_{O}}{V_{ref}} - 1\right) \times R2 \tag{2}$$



## OUTPUT VOLTAGE PROGRAMMING GUIDE

OUTPUT VOLTAGE	R1	R2	UNIT				
2.5 V	174	169	kΩ				
3.3 V	287	169	kΩ				
3.6 V	324	169	kΩ				
4.0 V	383	169	kΩ				
5.0 V	523	169	kΩ				

Figure 29. TPS76601 Adjustable LDO Regulator Programming



#### **APPLICATION INFORMATION**

#### power-good indicator

The TPS766xx features a power-good (PG) output that can be used to monitor the status of the regulator. The internal comparator monitors the output voltage: when the output drops to between 92% and 98% of its nominal regulated value, the PG output transistor turns on, taking the signal low. The open-drain output requires a pullup resistor. If not used, it can be left floating. PG can be used to drive power-on reset circuitry or used as a low-battery indicator.

#### regulator protection

The TPS766xx PMOS-pass transistor has a built-in back diode that conducts reverse currents when the input voltage drops below the output voltage (e.g., during power down). Current is conducted from the output to the input and is not internally limited. When extended reverse voltage is anticipated, external limiting may be appropriate.

The TPS766xx also features internal current limiting and thermal protection. During normal operation, the TPS766xx limits output current to approximately  $0.8\,\mu\text{A}$  (typ). When current limiting engages, the output voltage scales back linearly until the overcurrent condition ends. While current limiting is designed to prevent gross device failure, care should be taken not to exceed the power dissipation ratings of the package. If the temperature of the device exceeds  $150^{\circ}\text{C}(\text{typ})$ , thermal-protection circuitry shuts it down. Once the device has cooled below  $130^{\circ}\text{C}(\text{typ})$ , regulator operation resumes.

#### power dissipation and junction temperature

Specified regulator operation is assured to a junction temperature of  $125^{\circ}$ C; the maximum junction temperature should be restricted to  $125^{\circ}$ C under normal operating conditions. This restriction limits the power dissipation the regulator can handle in any given application. To ensure the junction temperature is within acceptable limits, calculate the maximum allowable dissipation,  $P_{D(max)}$ , and the actual dissipation,  $P_D$ , which must be less than or equal to  $P_{D(max)}$ .

The maximum-power-dissipation limit is determined using the following equation:

$$P_{D(max)} = \frac{T_J max - T_A}{R_{\theta, IA}}$$

Where

T<sub>I</sub>max is the maximum allowable junction temperature

 $R_{\theta JA}$  is the thermal resistance junction-to-ambient for the package, i.e., 176°C/W for the 8-terminal SOIC

T<sub>A</sub> is the ambient temperature.

The regulator dissipation is calculated using:

$$P_D = (V_I - V_O) \times I_O$$

Power dissipation resulting from quiescent current is negligible. Excessive power dissipation will trigger the thermal protection circuit.

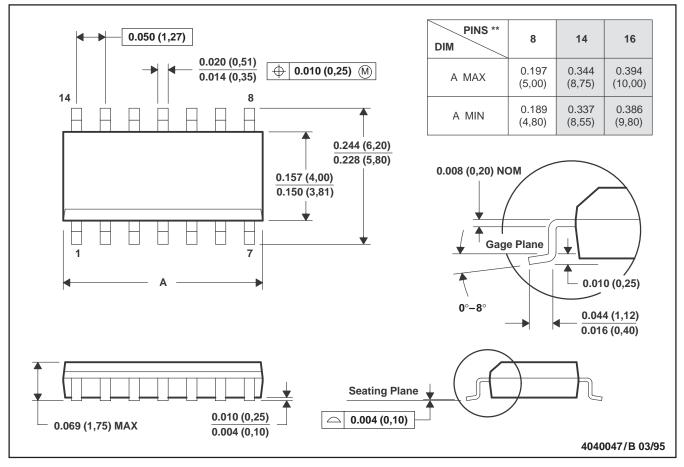


#### **MECHANICAL DATA**

#### D (R-PDSO-G\*\*)

#### 14 PIN SHOWN

#### PLASTIC SMALL-OUTLINE PACKAGE



NOTES: A. All linear dimensions are in inches (millimeters).

- B. This drawing is subject to change without notice.
- C. Body dimensions do not include mold flash or protrusion, not to exceed 0.006 (0,15).
- D. Four center pins are connected to die mount pad.
- E. Falls within JEDEC MS-012

#### **IMPORTANT NOTICE**

Texas Instruments and its subsidiaries (TI) reserve the right to make changes to their products or to discontinue any product or service without notice, and advise customers to obtain the latest version of relevant information to verify, before placing orders, that information being relied on is current and complete. All products are sold subject to the terms and conditions of sale supplied at the time of order acknowledgement, including those pertaining to warranty, patent infringement, and limitation of liability.

TI warrants performance of its semiconductor products to the specifications applicable at the time of sale in accordance with TI's standard warranty. Testing and other quality control techniques are utilized to the extent TI deems necessary to support this warranty. Specific testing of all parameters of each device is not necessarily performed, except those mandated by government requirements.

CERTAIN APPLICATIONS USING SEMICONDUCTOR PRODUCTS MAY INVOLVE POTENTIAL RISKS OF DEATH, PERSONAL INJURY, OR SEVERE PROPERTY OR ENVIRONMENTAL DAMAGE ("CRITICAL APPLICATIONS"). TI SEMICONDUCTOR PRODUCTS ARE NOT DESIGNED, AUTHORIZED, OR WARRANTED TO BE SUITABLE FOR USE IN LIFE-SUPPORT DEVICES OR SYSTEMS OR OTHER CRITICAL APPLICATIONS. INCLUSION OF TI PRODUCTS IN SUCH APPLICATIONS IS UNDERSTOOD TO BE FULLY AT THE CUSTOMER'S RISK.

In order to minimize risks associated with the customer's applications, adequate design and operating safeguards must be provided by the customer to minimize inherent or procedural hazards.

TI assumes no liability for applications assistance or customer product design. TI does not warrant or represent that any license, either express or implied, is granted under any patent right, copyright, mask work right, or other intellectual property right of TI covering or relating to any combination, machine, or process in which such semiconductor products or services might be or are used. TI's publication of information regarding any third party's products or services does not constitute TI's approval, warranty or endorsement thereof.

Copyright © 1999, Texas Instruments Incorporated