

TS75C32

V.32, V.22bis, V.22, V.23, V.21, BELL 212A, BELL 103 MODEM CHIP SET

ADVANCE DATA

- CCITT V.32, V22bis, V.22, V.21, V.23, Bell 212A, Bell 103 COMPATIBLE MODEM CHIP SET
- INTEGRATED IMPLEMENTATION ON THREE DSP AND THREE MAFE CHIPS
- FULL DUPLEX OPERATION FROM 9600 TO 300BPS
- FULL IMPLEMENTATION OF THE V.32 AND V.22bis HANDSHAKE
- DYNAMIC RANGE: 43dB
- TWO SATELLITE HOPS AND FREQUENCY OFFSET CAPABILITIES (10Hz) FOR THE FAR END ECHO CANCELLER IN V.32 MODE
- TRELLIS ENCODING AND VITERBI DECOD-ING
- 12.5% ROLL-OFF RAISED COSINE TRANS-MITTER PULSE SHAPING
- HIGH PERFORMANCE PASSBAND FRAC-TIONALLY SPACED ADAPTIVE EQUALIZER
- SIGNAL QUALITY MONITORING
- PARALLEL INTERFACE TO STANDARD MI-CROPROCESSORS
- BIT RATE DATA CLOCKS PROVIDED FOR SYNCHRONOUS DATA TRANSFER
- FULL DIAGNOSTIC CAPABILITY
- DTMF GENERATION
- CALL PROGRESS TONE DETECTION
- SOFTWARE LICENSE AND DEVELOPMENT TOOLS AVAILABLE FOR EASY CUSTOMIZA-TION
- TOTAL POWER CONSUMPTION BELOW 2W

DIP24 (Plastic Package) DIP28 (Plastic Package) DIP48 (Plastic Package) PLCC28 (Plastic Chip Carrier) PLCC52 (Plastic Chip Carrier)

DESCRIPTION

The SGS-THOMSON Microelectronics TS75C32 chip set is a highly integrated modem engine, which can operate in full duplex from 9600 to 300bps. The modem hardware consists of three analog front end (MAFE) chips, three DSP processor chips and additional memory chips.

The three SGS-THOMSON analog front end chips (TS68950/1/2) are the transmit interface, the receive interface and the clock generator respectively.

The modem signal processing functions are implemented on three ST18930 programmable digital signal processors.

ORDER CODES

Part Number	Temperature Range	Package
TS75C320CP	0°C to 70°C	DIP48
TS75C321CP	0°C to 70°C	DIP48
TS75C322CP	0°C to 70°C	DIP48
TS68950CP	0°C to 70°C	DIP24
TS68951CP	0°C to 70°C	DIP28
TS68952CP	0°C to 70°C	DIP28
TS75C320CFN	0°C to 70°C	PLCC52
TS75C321CFN	0°C to 70°C	PLCC52
TS75C322CFN	0°C to 70°C	PLCC52
TS68950CFN	0°C to 70°C	PLCC28
TS68951CFN	0°C to 70°C	PLCC28
TS68952CFN	0°C to 70°C	PLCC28

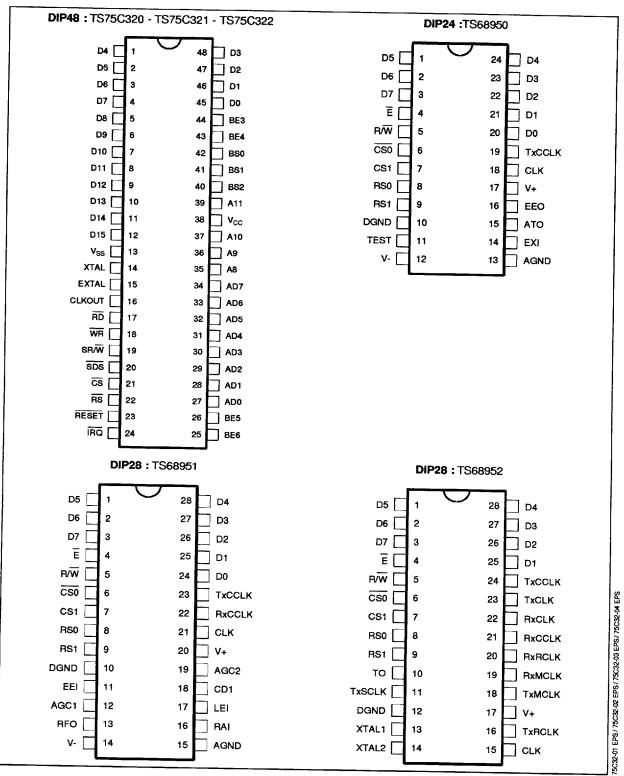
February 1995

1/40

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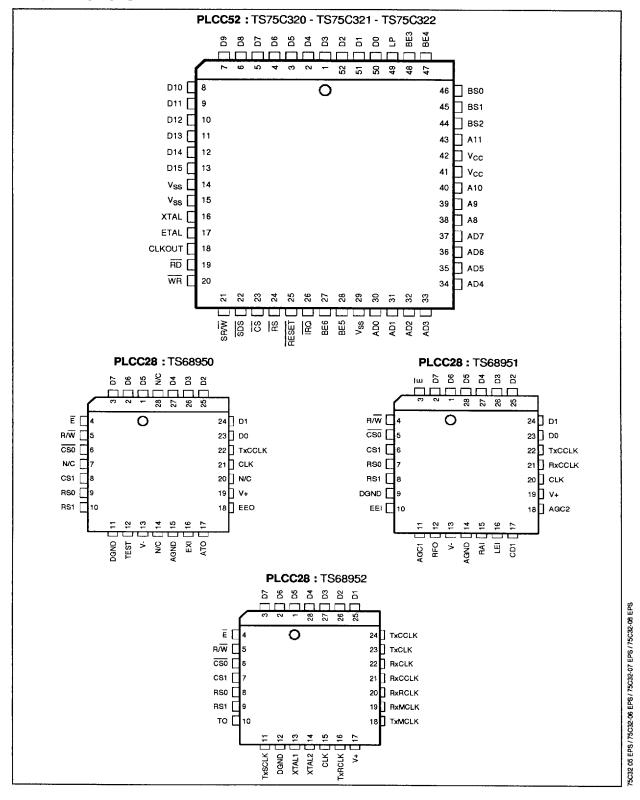
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PIN CONNECTIONS



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PIN CONNECTIONS



SGS-THOMSON MICROELECTRONICS

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PIN DESCRIPTION

Pin Name	Туре	Signal Name	Description
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SYSTEM INTERFACE

TS75C321 (DSP#1 Transmitter and Handshake)

AD0.AD7	1/0	D0H.D7H	System Data Bus: these lines are used for data transfer between the TS75C32 mailbox and the host processor
<u>cs</u>	ı	CSL	Chip Select: this input is asserted when the TS75C32 is to be accessed by the host processor
RS	ı	RSL	Register Select: this signal is used to control the data transfers between the host processor and the TS75C32 mailbox
SDS	1	DSL	System Data Strobe: synchronizes the transfer between the TS75C32 mailbox and the host processor
SR/W	1	RWL	System Read/Write: control signal for the TS75C32 mailbox operation
ĪRQ	0	INTL	Interrupt Request : signal sent to the host processor to access the TS75C32 mailbox
RESET	1	RSTL1	Master Reset of DSP#1

ANALOG INTERFACE

TS68950 (Analog Front End Transmitter)

ATO	0	ATO	Analog Transmit Output
TS68951 (An:	aloo Fron	t End Receive	ur)

RAI	-	RAI	Receive Analog Input
LEI	1	LEI	Local Echo Input. Must be grounded.

CLOCK INTERFACE

TS68952 (Clock Generator)

TxCLK	0	TxCLK	Transmit Bit Clock	
TxRCLK	0	TxRCLK	Transmit Baud Clock	
TxCCLK	0	TxCCLK	Transmit Conversion Clock	
TxMCLK	0	TxMCLK	Transmit Multiplex Clock	
RxCLK	0	RxCLK	Receive Bit Clock	
RxRCLK	0	RxRCLK	Receive Baud Clock	
RxCCLK	0	RxCCLK	Receive Conversion Clock	
RxMCLK	0	RxMCLK	Receive Multiplex Clock	Ē
TxSCLK	I.	TxSCLK	Transmit Synchro Clock: can be used to synchronize the transmitter on an external bit clock provided by the RS232C (or V.24) junction	

4/40

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ABSOLUTE MAXIMUM RATINGS

Symbol	Parameter	Value	Unit
S75C320/1	/2		
V _{CC} (1)	Supply Voltage	- 0.3, + 7.0	V
Vin (1)	Input Voltage	- 0.3, + 7.0	V
Toper	Operating Temperature	0, + 70	°C
T _{stg}	Storage Temperature	- 55, + 150	°C
S68950/1/2			
	Supply Voltage between V + and AGND or DGND	- 0.3, + 7	V
	Supply Voltage between V - and AGND or DGND	- 7, + 0.3	٧
	Voltage between AGND and DGND	~ 0.3, + 0.3	٧
	Digital Input Voltage	DGND - 0.3, V _{CC} ⁺ + 0.3	٧
	Digital Output Voltage	DGND - 0.3, V _{CC} ⁺ + 0.3	٧
	Digital Output Current	- 20, + 20	mA
	Analog Input Voltage	V _{CC} - 0.3, V _{CC} + 0.3	٧
	Analog Output Voltage	V _{CC} - 0.3, V _{CC} + 0.3	٧
	Analog Output Current	- 10, + 10	mA
Ptot	Power Dissipation	500	mW
Toper	Operating Temperature	0, + 70	°C
T_{stg}	Storage Temperature	- 65, + 150	°C

DC ELECTRICAL CHARACTERISTICS (DGND = AGND = 0V)

Symbol	Parameter	Min.	Тур.	Max.	Unit
DIGITALS	SUPPLY ($V_{CC} = 5.0V \pm 10\%$, $V_{SS} = 0$, $T_{amb} = 0$ to $+70^{\circ}$ C, unless	s otherwise specified)		L	
Vcc	Supply Voltage	4.5	5	5.5	V
VIL	Input Low Voltage	- 0.3		0.8	V
V _{IH}	Input High Voltage	2.4		V _{CC}	V
I,	Input Extal Current	- 50		+ 50	μА
I _{in}	Input Leakage Current	- 10		10	μA
VoH	Output High Voltage (I _{load} = - 300μA)	2.7			v
Vol	Output Low Voltage (Iload = 3.2mA)			0.5	v
PD	Total Power Dissipation		1.5		w
C _{in}	Input Capacitance		10		pF
ITSI	Three State (off state) Input Current	- 20		- 20	
NALOG S	SUPPLY			<u> </u>	
V ⁺	Positive Power Supply	4.75		5.25	V
V-	Negative Power Supply	- 5.25		- 4.75	V
1+	Positive Supply Current			35	mA
1-	Negative Supply Current	- 35			17,5 (

Note: Case Temperature T_C must be maintened below 100°C.

5/40

Notes: 1. With respect to Vss.
 Stresses above those hereby listed may cause permanent damage to the device. The ratings are stress ones only and functional operation of the device at these or any conditions beyond those indicated in the operational sections of this specification is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability. Standard MOS circuits handling procedure should be used to avoid possible damage to the device.

AC ELECTRICAL SPECIFICATIONS

CLOCK AND CONTROL PINS TIMING

(Vcc = $5.0V \pm 10\%$, $T_{amb} = 0$ to $+70^{\circ}C$, see Figure 1 - Output Load = 50pF + DC characteristics I load Reference levels : $V_{IL} = 0.8V$, $V_{IH} = 2.4V$, $V_{OL} = 1.4V$, $V_{OH} = 2.4V - t_r$, $t_f \le 5ns$ input signal)

Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Unit
t _{cex}			40 50 40	iyp.	100 200 100	ns ns ns
t _{fex}	External Clock Fall Time				5	ns
t _{rex}	External Clock Rise Time				5	ns
t _{coh}	EXTAL to CLKOUT High Delay			25		ns
t _{col}	EXTAL to CLKOUT Low Delay			25		ns
tcor	CLKOUT Rise Time		-		10	ns
tcof	CLKOUT Fall Time				10	
t dic	CLKOUT to Control Output Low (INTL)				30	ns
taho	CLKOUT to Control High (INTL)				30	ns ns

TS68952: CLOCK GENERATOR (see Figure 2)

(unless otherwise noted, Electrical Characteristics are specified over the operating range.

Typical values are given for $V^+ = 5.0V$ and $t_{amb} = 25^{\circ}C$)

Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Unit
CRYSTAL	OSCILLATOR INTERFACE		L		L	
V _{IL}	Input Low Level Voltage			Τ	1.5	V
V _{IH}	Input High Level Voltage		3.5			V
ljL	Input Low Level Current	DGND ≤ V _I ≤ V _{IL max}	-15	<u> </u>		μA
I _{iH}	Input High Level Current	$V_{IH min} \leq V_I \leq V^+$			15	μA
CLOCK W	AVEFORMS		L	L		
PC	Main Clock Period	XTAL1 Input	150	173.6		ns
twcL	Main Clock Low Level Width	XTAL1 Input	50			ns
t wcH	Main Clock High Level Width	XTAL1 Input	50			ns
t RC	Main Clock Rise Time	XTAL1 Input			50	ns
t _{FC}	Main Clock Fall Time	XTAL1 Input			50	ns
t DC	Clock Output Delay Time	All Clock Outputs C _L = 50pF			500	ns
t TC	Clock Output Transition Time	All Clock Outputs C _L = 50pF			100	ns

6/40

SGS-THOMSON MIGROLUSTRONICS

Figure 1: Clock and Control Pins Timing

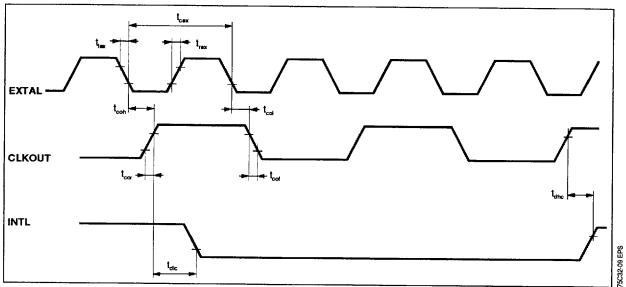
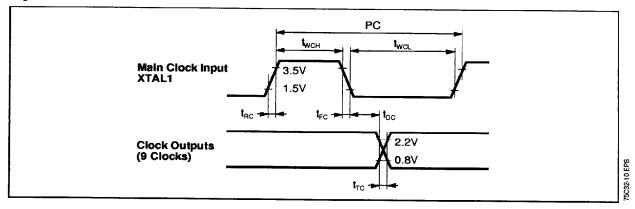


Figure 2: Clock Generator



AC ELECTRICAL SPECIFICATIONS (continued)

LOCAL BUS TIMING ($V_{CC} = 5.0V \pm 10\%$, $T_{amb} = 0$ to $+70^{\circ}$ C, see Figure 3)

Symbol	Parameter	Min.	Тур.	Max.	Unit
t⊳w	RD, WR, SOS Pulse Width	1/2 tc - 10		1/2 tc	ns
t ah	Address Hold Time	10			ns
t osw	Data Set-up Time, Write Cycle	25			ns
t DHW	Data Hold Time, Write Cycle	10			ns
tozw	DS High to Data High Impedance, Write Cycle			40	ns
t osa	Data Set-up Time, Read Cycle	20			ns
t DHR	Data Hold Time, Read Cycle	5			ns
t arw	Address Valid to WR, DS, SOS Low	1/2 tc - 25	-		ns

SYSTEM BUS TIMING ($V_{CC} = 5.0V \pm 10\%$, $T_{amb} = 0$ to $+70^{\circ}$ C, see Figure 4)

Symbol	Parameter	Min.	Тур.	Max.	Unit
tspw	SDS Pulse Width	50			ns
tsaw	SR / W, CS, RS Set-up Time	15			ns
tsah	SR / W, CS, RS Hold After SDS High	5			ns
t sdsa	Data Set-up Time, Read Cycle	15			ns
t sdhr	Data Hold Time, Read Cycle	5			ns
tsdsw	Data Set-up Time, Write Cycle			30	ns
tsdhw	Data Hold Time, Write Cycle	10		50	ns
t oshir	SDS High to IRQ High			800	ns

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8/40

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9/40

Figure 3: Local Bus Timing Diagram

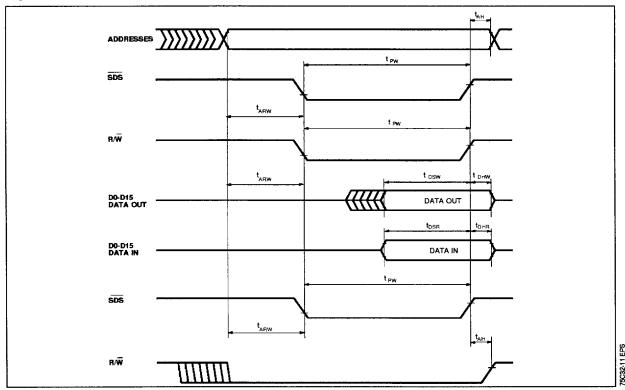
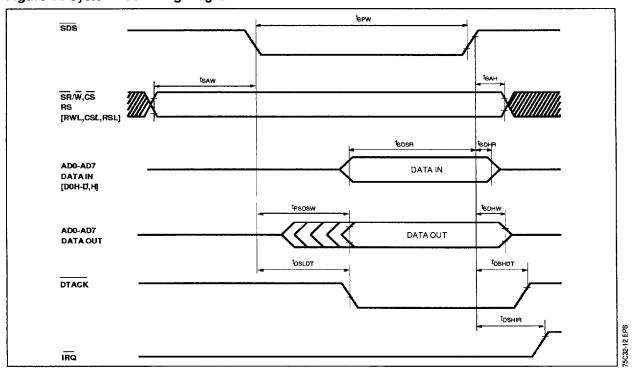


Figure 4: System Bus Timing Diagram



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FUNCTIONAL DESCRIPTION

1 - SYSTEM ARCHITECTURE

The SGS-THOMSON TS75C32 chip set is a highly integrated modem engine which provides the functionality and performance requirements for full-duplex 9600bps modem solutions at a low cost and with a small circuit area. At the heart of the modem engine are three SGS-THOMSON DSPs which implement the complete signal processing and control functions. The analog front end of the modem engine consists of the SGS-THOMSON MAFE three-chip set which is designed to meet the requirements of high-speed modem applications and particularly V.32 modems. The only other components in the modem engine are the external RAM chips used for the far-end echo canceller delay line and the Viterbi decoder.

2 - PROCESSOR AND MAFE CHIPS ARRANGEMENT

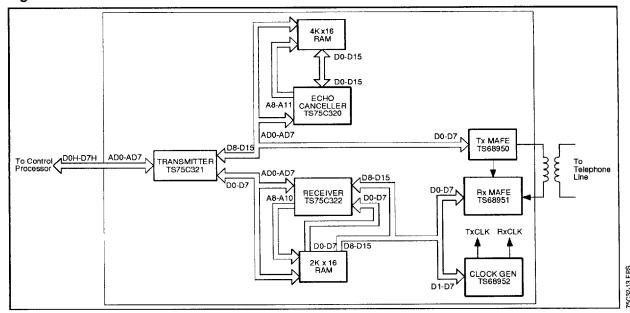
Figure 5 shows the interconnections between the MAFE and signal processors.

DSP 1 communicates with the control processor through its system bus, AD0-AD7. It is also con-

nected to the two other DSPs through its D0-D7 and D8-D15 data buses to transfer data, to pass a control command to the DSPs and to get the modem operation status and then pass it to the control processor. The transmitter, V.32 handshake and part of the receiver algorithms are implemented in this processor. DSP 0 implements the echo cancellation function. 4K x 16 of RAM are connected to this processor to implement the data delay line for the far end echo cancellation. DSP 2 implements most of the receiver functions. 2K x 16 of RAM are attached to it due to the requirements of the Viterbi decoder.

The transmitter interface chip, TS68950 (see ref 5 of Appendix D), is connected to the 8 MSB's of the DSP 1 data bus. The echo replica is sent from DSP1 to TS68950 then to the receiver interface chip, TS68951 (see ref 6 of Appendix D), after conversion to analog format. This chip and the clock generator chip, TS68952 (see ref 7 of Appendix D), are connected to the 8 MSB's of the DSP 2 data bus. The clock generator chip generates the A/D and D/A sampling clocks and the data bit and baud rate clocks.

Figure 5: Hardware Architecture



3 - OPERATION

3.1 - Modes

The modem implementation is fully compatible with many CCITT and Bell recommendation. It operates at different bit rates from 9600 to 300bps. In the 9600bps mode, the trellis encoder and the Viterbi decoder can be switched in or out. Both the bit rate and trellis options are determined during the initial modem handshake sequence.

During FSK Modes (Bell 103, V.21 and V.23) a byte of information is exchanged with the Data Pump. This byte is a sampling image (7.2kHz) of the Transmit circuit (TxD) of the junction.

3.2 - Signal Spectrum Shaping

A square root of 12.5 percent roll-off raised cosine filter is implemented in the transmitter to properly shape the transmit signal pulse. This filter is chosen based on a compromise of two considerations. First, the signal should have a narrow spectrum to avoid severe distortion on the telephone line.

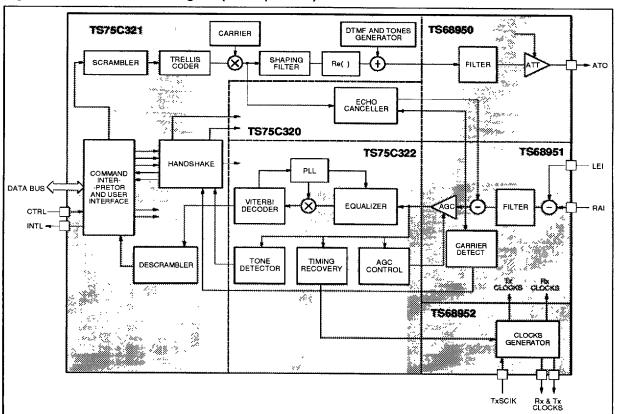
Second, the signal spectrum should be made as wide as possible to facilitate timing recovery in receiver.

3.3 - Echo Cancellation

The echo canceller is implemented on a single DSP (see ref 8 of Appendix D) with its associated external RAM. It cancels both near-end and far-end echoes even in the presence of frequency offset in the far-end echo path. The near-end echo cancellation is better than 55dB and the residual near-end echo is smaller than - 65dBm with a near-end echo level of - 10dBm at the receiver input and a far-end signal level of - 43dBm.

The combined near-end and far-end echo cancellers maintain the residual echo level 24dB below the received signal even if the far-end echo signal path introduces up to 10Hz of frequency offset. This level of cancellation is achieved when the far-end echo is 8dB below the received far-end signal.

Figure 6: Functional Block Diagram (V.32 Operation)



3.4 - Receiver Description

The incoming signal is sent to the receiver interface chip to have the echo removed before being sent to DSP 2. The timing recovery algorithm takes the signal after the echo cancellation to derive the timing error to control the sampling phase of the A/D. It is able to cope with distant modern frequency drifts up to ±2.10⁻⁴ as per CCITT rec. The A/D output samples are sent to the adaptive equalizer and the signal energy estimator for the gain control. The adaptive equalizer outputs a complex number every baud interval, which is then phase corrected by the carrier recovery loop. The Viterbi decoder makes hard decisions on the phase corrected samples for the adaptation of the equalizer and carrier recovery. It also makes soft decisions with an optimum decoding depth.

3.5 - Equalization

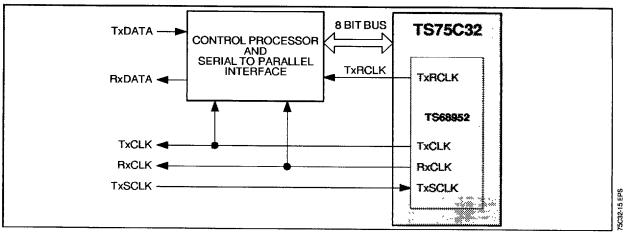
The modem receiver has a passband T/3 fraction-

ally spaced automatic adaptive equalizer which can compensate for the signal degradation caused by low quality line conditions.

3.6 - Synchronous and Asynchronous Data Transfer

The TS75C32 modem engine provides the control processor and the DTE with both the transmit and the receive bit clocks (Figure 7). These clocks are generated by the TS68952 and are independent of each other. The receive clock (RxCLK) is derived from the received data signal. The transmit clock (TxCLK) is free-running at the nominal bit rate except during Digital Loopback Mode when it is synchronous to the RxCLK. If the transmit clock is free-running and an external bit clock signal from the terminal is connected to point TxSCLK then the transmit bit clock will be synchronized to that signal. The baud clocks (TxRCLK and RxRCLK) are also available to the control processor. If the TxSCLK pin is not used, it should be tied to a fixed logic level.

Figure 7: Clock Signals for Synchronous Transmission



The control processor interface is synchronous with the transmit baud clock. Eight bits of data are transferred from the control processor to DSP 1 for each information exchange. At 9600bps, the data is transmitted every 2 bauds and the data is transmitted every 4 bauds for 4800bps. The received bits are also nominally transferred from DSP 1 to the control processor once every two transmit baud intervals. When the transmitter is not synchronized with the receiver, however, the receive baud interval may be slightly shorter or longer than the transmit baud interval. If it is shorter, it is necessary to periodically pass 16 received bits from DSP 1 to the control processor. If it is longer, then periodically, there will be no data transmitted from DSP 1 to the control processor. Since the received bits are being passed to the DTE at a fixed rate equal to the RxCLK, some buffering is necessary in the control processor.

For asynchronous transmission, the clocks are not required by the DTE. But since the control processor to DSP 1 interface is still synchronous with respect to the transmit baud clock, the control processor must implement the asynchronous to synchronous conversion (as specified in the V.22 bis recommendation, for example). This will consist of inserting or deleting stop bits as required, to ensure that the transmitted bit rate is within 0.01% of the nominal rate (9600 or 4800bps).

3.7 - Clock and Data Synchronization

Both transmit and receive clocks have the same nominal frequency value. However, except in FSK Modes, these clocks are plesiochronous. The nominal frequencies are:

Mode	TxCLK	TxRCLK	Byte/irq	Rate
V.32TCM	9.6K	2.4K	1/2	9600bps Trellis
V.32QAM	9.6K	2.4K	1/2	9600bps QAM
V.32QAM	4.8K	2.4K	1/4	4800bps
V.22Bis	2.4K	600	1/8	2400bps
V.22	1.2	600	1/16	1200bps
V.22	600	600	1/32	600bps
Bell212A	1.2	600	1/16	1200bps
V.23	7.2K	1.2K	1/3	1200/75bps
Bell103	7.2K	1.2K	1/3	300bps
V.21	7.2K	1.2K	1/3	300bps

The TxCCLK is always 7.2kHz. The TxMCLK is always set to 2.4kHz after a configuration command (cv32, cv22b...), but can be changed by a cmafe command.

During the Hansdshake (V.32 and V.22bis) the bit clock (TxCLK and RxCLK) is set to the maximum value (respectively 9600Hz and 2400Hz). At the

end of the handshake, depending of the negociation, this frequency is automatically set to the proper value.

The Irq Rate is always 2400 interruptions by second. The Byte/Irq is the nominal ratio of number of Byte of information (data) for each Irq generated by the data pump. The 1/2 ratio means that we have to send (xmiti command) 1 byte of data each two Irq, the additional Irq can be used to send an extra CCI command or must be a nop command.

3.8 - Tone Generator

The TS75C32 Engine has thirteen tone commands to quickly program the tone generators to generate the 2100Hz Answer Tone (ANSWR) and the tone pairs for DTMF digits (DTMF0, ..., DTMF9, DTMF*, DTMF#). Silence, i.e. termination of tone generation, is accomplished by the use of a fourteenth command, SLNTS. These commands provide the tones and control required for normal operation of the modem.

Some circumstances might arise where additional tones are desired. For such cases, the V.32 Engine provides the user with the ability to generate such additional tones. This special feature is achieved through use of the tone control commands.

The TS75C32 Engine maintains a pair of locations which are reserved for tone generation parameters. These locations are denoted as TONE1 and TONE2. These locations may be programmed by the use of the define tone commands, DEFT1 and DEFT2. These commands provide the two tone generators with the phase increment of the tone to be generated with respect to the 7200Hz sample rate.

The normal tone commands automatically program the tone generators. The DEFT1 and DEFT2 commands do not change the enabled or disabled state of the tone generators. If a tone is being generated when the DEFT command is received, the new tone will be generated without further action on the part of the user. If tone generation was not in progress it is not started.

Enabling the tone generators is accomplished by the tone control commands TGEN0, TGEN1, TGEN2, and TGEN12. Each of these commands affects both tone generators. TGEN0 disables both tone generators and TGEN12 enables both tone generators. To enable tone generator 1 and disable tone generator 2 the TGEN1 command is used. For the reverse condition, with generator 1 disabled and generator 2 enabled, the TGEN2 command is employed.

13/40



Refer to the command in appendix A for more detailed information.

Generation of special user tones is not part of the normal data communications operations of the modem. Use of this feature may interfere with data transfer operations. It is the responsability of the user to insure that the tone generators are used at a time when such interference will not occur and to disable both tone generators when the tone generation operations have been completed.

3.9 - Test Modes

The modem can be configured in two test modes, namely analog loop back and digital loop back modes. These loop back modes conform to the test loops 3 and 2 respectively defined in CCITT recommendation V.54.

In the local analog loop back mode, the transmitter signal is directly fed back into the local receiver input with the echo canceller enabled. The user is responsible for supplying a switch, which is controllable by the control processor, to enable or disable the analog loop back mode. The receiver descrambler is set as the inverse of the transmitter scrambler so that the receiver detects correct bits.

If the modern is configured in the digital loop back mode, the transmitter dock is locked to the receiver clock and the received bits are used as the transmitter input.

Figure 8: Functional Interconnect Diagram

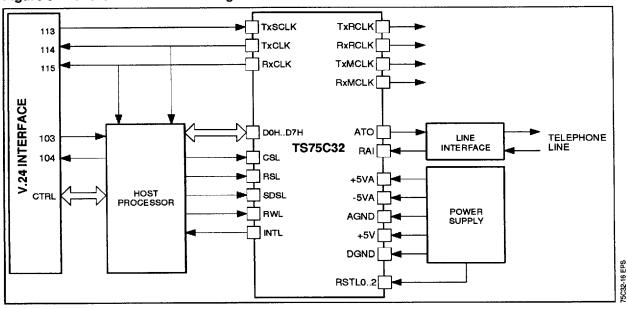
3.10 - Power-on Initialization

When the power is turned on, the transmitter interface sets the output signal attenuation to infinite. This avoids undesirable signal transmission on the telephone line (see ref 5 of Appendix D). The gain of the AGC in the receive interface is set at the lowest level to avoid signal clipping during the initial handshake. The clock generator is programmed to generate all the necessary clocks for the 9600bps mode. The clocks include the 7200Hz sampling clock, the 2400Hz baud rate clocks and the 9600bps bit rate clocks. The transmit clocks are free running when the TxSClk pin is tied to a fixed logic level. Otherwise, the transmit bit clock is synchronized to the frequency present at the TxSClk pin. DSP 1 is configured properly to receive commands from the control processor.

4 - MODEM INTERFACE (Figure 8)

4.1 - Analog Interface

The transmit signal at the tip and ring is programmable over a 22dB dynamic range by 2dB steps in TS68950. The signal level can be further scaled to any value by setting a scaling factor in the DSP. The nominal Transmit level, at the ATO pin is -5.7dBm.



14/40

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4.2 - Digital Interface

The DSP and control processor interface complies with the system bus interface of the ST18930. The interface to the control processor is managed by DSP 1 as shown in Figure 5. The DSP signals which are presented to the interface, and a brief definition of the signals are tabulated in Table 1.

Table 1: Digital Interface Signals

Interface Signals	Signal Definition
D0H D1H D2H D3H D4H D5H D6H D7H	Data Bus (LSB) Data Bus
RWL DSL INTL CSL RSL	Write Signal Data Strobe Mailbox Handshake DSP Select Register Select
TxRCLK TxRCLK TxCLK TxCLK RxCLK TxMCLK RxMCLK TxSCLK	Transmit Baud Rate Clock Receive Baud Rate Clock Transmit Bit Rate Clock Receive Bit Rate Clock Transmit Multiplex Clock Receive Multiplex Clock Transmit Terminal Clock

All information exchanges across this interface conform to the three byte mailbox structure (see ref 4 of Appendix D) and protocol of the DSP. As may be seen in the table, the DSP generates a control signal, INTL, which defines the mailbox handshake operation.

4.3 - Control Processor/DSP Interface

As seen by the software in the user provided control processor, DSP 1 is a synchronous machine. It requires the attention of the control processor at regular intervals in order to perform properly. Any failure of the control processor to interact with the modem engine in a timely manner will result in reduced performance or improper operation.

Each interaction begins when the control processor sends a three byte command to the mailbox. Once the command has been written to the mailbox, the ownership of the mailbox is relinquished by the control processor. Upon acquisition of the mailbox, DSP 1 reads the command bytes and then sends a three byte response to the mailbox. Then, DSP 1 relinquishes the ownership of the mailbox back to the control processor. The received command is then decoded and the embedded data and/or operational parameters are extracted and acted upon

as appropriate. The modem status information will be collected for the next mailbox exchange. The control processor handles the returned information as soon as it regains the ownership of the mailbox. Because the control processor owns the mailbox initially, it may store a command at any time before it is required by DSP 1. After this, the mailbox becomes available to DSP1 and can be read by it when required.

4.4 - Mailbox Description

The mailbox located internally to the DSP contains 3-byte input (RIN) and 3-byte output (ROUT) shift registers. The DSP has an internal flag RDYOIN which indicates whether the DSP (RDYOIN = 0) or control processor (RDYOIN = 1) has access to the mailbox. The DSP can relinquish its accessability to the mailbox by setting RDYOIN but it can no longer regain access to the mailbox as RDYOIN is reset only after the control processor relinquishes its accessability to the mailbox. The access protocol and system bus transfers are controlled by an internal I/O sequencer within the DSP described as follows:

- The mailbox is made available to the control processor by the DSP program which sets RDYOIN flag to 1. This action will cause INTL mailbox handshake signal to switch to the active (low) state.
- The control processor detects INTL active and dummy reads the mailbox by forcing DSP Select (CSL) and Register Select (RSL) low along with write signal (RWL) high. The activated Data Strobe signal (DSL = 0) validates the above signals.
- The DSP detects the dummy read of its mailbox via the control signals mentioned in 2 and negates INTL mailbox handshake signal within 800ns.
- 4. The control processor detects the negation of INTL indicating that the DSP mailbox is available for data transfers. The control processor writes three 8-bit bytes and/or reads three 8-bit bytes in the mailbox shift registers RIN, ROUT respectively.
- The exchange protocol described above is terminated by the control processor performing a dummy read of the mailbox as in 2 but with RSL in the high state.
- The RDYOIN flag within the DSP is cleared to 0 by the dummy read of the mailbox in step 5 and the DSP now has access to RIN and ROUT registers within the mailbox.

15/40

SGS-THOMSON MICROPHIC BURNINGS

USER INTERFACE - COMMAND SET

The command set has the following attractive features:

- User friendly with easy to remember mnemonics
- Allows straightforward expansion with new commands to suit specific customization requirements
- Fully compatible with other SGS-THOMSON DSP-based modern products.

The command set has been designed to provide the necessary functional control of the TS75C32. Each command falls into one of several groups, based on function and the presence or absence of parameters. The length of the OP code varies with instruction type, but in all cases, a command consists of three bytes.

The commands which pass parameters or data to DSP 1 have a short OP code format. Byte 0 forms the OP code portion of the command. Bytes 1 and 2 are data and/or parameters associated with each OP code. The meaning of the last two bytes is dependent on the specific instruction.

Other instructions command the TS75C32 to perform certain specific tasks. These do not pass parameters or data to the TS75C32. These commands have an OP code which is a full 24 bits in length.

The command set of the TS75C32 is summarized below. The descriptions are of the form:

MNEMONIC (OPCODE): DESCRIPTION

For detailed information and data format specifics of each command, please refer to appendix A.

1 - COMMAND SUMMARY

1.1 - Operational Control Commands

FREZ (14): Freeze adaptive processes. Freeze the adaptive processes as specified by the data in bits 0 and 1 of byte 1. Bit 0 of byte 1 controls the adaptive equalizer. Freeze the equalizer if bit 0 of byte 1 is 1. A0 in this bit will unfreeze the equalizer. Echo canceller adaptation is controlled by bit 1 of byte 1. if bit 1 is 1, the echo canceller adaptation is frozen. The echo canceller adaptation is unfrozen by a 0 in bit 1.

HSHK (040000): Handshake. Begin the handshake sequence. The V.32 modem engine carries out all the steps defined in the CCITT recommendation. The status reported to the control processor will indicate the success or failure of the process and its progress.

INIT (0600C0): Initialize. Initialize the V.32 modem engine. Set all parameters to default values and wait for commands for the control processor.

NOP (000000): No Operation. No new operation is commanded. The state of the V.32 engine remains unaltered and a previously invoked multibaud command (such as HSHK) continues.

RTRA (050000): Retrain. Start sending the retrain sequence as defined in the CCITT recommendation.

SETGN (02): Set Gain. This command sets a global gain factor, which will be multiplied by all transmit samples before being sent to the TS68950. Bytes 1 and 2 store the gain factor.

1.2 - Data Communications Commands

XMIT (03): Transmit data. Transmit data to far end modem. The data is provided in byte 1 of the command, where the least significant bit is the first bit to be transmitted. The third byte of the command must be provided, but is not used. Hence, any value may be supplied.

XMITI (01): Transmit data and Initiate additional cycle. Transmit data and inform the DSP to accept another command at the next transmit baud. If the next command requires an answer from DSP 1, the control processor has to keep issuing this command followed by a command which does not requires an answer until the answer has been received.

1.3 - Memory Manipulation Commands

SPAC (13): Store Parameter And Count. Store parameter in addressed memory and increment the pointer. This command passes data in bytes 1 and 2, least significant byte in byte 1. It is used to write an arbitrary 16-bit value into the writable memory location currently specified by the Memory Address Register. The contents of the Memory Address Register are incremented by 1 at the completion of this command.

SPAM (12): Store parameter in Addressed Memory. This command passes data in bytes 1 and 2, least significant byte in byte 1. It is used to write an arbitrary 16-bit value into the writable momory location currently specified by the Memory Address Register.

WARP (10): Write Address and Return Parameter. This command allows the controller to read any of the XRAM, YRAM, ERAM or CROM (DSP internal memory areas) of any of the three modem DSPs without interrupting the processors. The address to the V.32 modem engine is provided in bytes 1 and 2 of the command (least significant byte first). DSP 1 stores the address in the Memory Address Register and returns the contents of the addressed location.

16/40

SGS-THOMSON MICROPLECTIONICS

USER INTERFACE - COMMAND SET (continued)

WARPX (11): Write Address and Return Parameter Complex. The address to the V.32 modem engine is provided in bytes 1 and 2 of the command (least significant byte first). DSP 1 stores the address in the Memory Address Register. The most significant bytes of the real and imaginary parts of a complex number are returned. The 8 most significant bits of the data addressed by the Memory Address Register are returned to the control processor through byte 1. Byte 2 stores the 8 most significant bits of the data at the location immediately higher. The Memory Address Register retains the address provided. (i.e. it is not incremented.)

1.4 - Configuration Control Commands

CV32 (20): Configure modem for V.32. Configure the modem as Originate / Answer, 9600/4800, Viterbi / No-Viterbi, Analog Loopback, Digital Loopback.

CFSK (1D): Configure modem for FSK modes of operation (V.23/V.21/Bell103).

CV22B (1E): Configure modem for V.22Bis/V.22/Bell212Amodes.

1.5 - MAFE Manipulation Commands

CMAFE (07): Configure MAFE. The following two bytes of this command are written directly to the MAFE chip set (TS68950/1/2). This allows the control processor to configure parameters, such as the transmit level, the receiver analog front end, and the transmit and receive clocks.

RRR1 (080000): Read Register 1. Causes the TS75C32 Engine to read and immediately return the 12 bit contents of the MAFE register RR1.

RRR2 (090000): Read Register 2. Causes the TS75C32 to read and immediately return the 12 bit contents of the MAFE register RR2.

WTR1 (0A): Write Register 1. Causes the TS75C32 to write the supplied data to the MAFE register TR1.

WTR2 (0B): Write Register 2. Causes the TS75C32 to write the supplied data to the MAFE register TR2.

1.6 - Tone Select Commands

TONE (0C): Select Tone. Program the tone generator(s) for the desired tone(s). Examples include:

- ANSWR (0C1000): Program the tone generator for the 2100Hz answer tone.
- DTMF (see appendix): Program the tone generators for the tone pair which forms the specified DTMF digit.

This command selects the tones to be transmitted, but does not enable the tone generators. To transmit the tones, the tone control commands must be issued.

1.7 - Tone Control Commands

DEFT1 (0E): Define Tone 1. Define tone 1 as specified by the parameter provided. The two data bytes following the opcode are used to program, but not enable, tone generator 1. The data for the tone is represented as a phase offset per sample. Byte 1 stores the least significant byte of the phase increment.

DEFT2 (OF): Define Tone 2. Define tone 2 as specified by the parameter provided. The two data bytes following the opcode are used to program, but not enable, tone generator 2. The data for the tone is represented as a phase offset per sample. Byte 1 stores the least significant byte of the phase increment.

SLNT (0D0000): Silence the tone generators. Discontinue tone transmissions by disabling the tone generators.

TGEN (0D): Tone Generator control. Enable or disable tone generator 1 and tone generator 2 according to parameter provided. If both tone generators are enabled, the level of tone 2 is 2dB higher than that of tone 1.

2 - STATUS REPORTING

Whenever DSP 1 owns the mailbox, it transmits the modern status to the control processor. The status consists of three bytes of information which are stored by DSP 1 in its ROUT register for access by the control processor. These three bytes may consist of received bits and modern status or they may contain the answers to the previous command, such as WARP and RRR1/2.

Data bits have higher priority than the answer to the previous command. If both data byte and command answer are ready to be sent, the data will be sent.

Byte 0 contains status flags. Refer to appendix B for the detailed format of the status response. The four most significant bits, F00, F01, F10 and F11, indicate various conditions during the call establish-ment, handshaking and the data modes. They have different meanings in different modes. The flag DAV1 and DAV2 are used to indicate the type of information contained in bytes 1 and 2. Bit H is used to indicate the condition of the handshake and bit 107 informs the control processor whether the 107 flag has to be set.

SGS-THOMSON MICROLLICTRONICS

17/40

USER INTERFACE - COMMAND SET (continued)

DAV1 and DAV2. If both DAV1 and DAV2 are set to 1, bytes 1 and 2 contain the data in response to the previous command. Refer to the relevant commands in appendix Ato get the detailed information on the interpretation of the data in bytes 1 and 2. Otherwise, they contain either the received data bits or the handshake detection status or both.

If both bits are set to 0, both byte 1 and byte 2 contain the data bits, where the bits in byte 1 are received earlier in time. The least significant bit is the first bit received. The data bits are stored in byte 1 and the modem status is stored in byte 2 when DAV2 is 1 and DAV1 is 0. When DAV1 is 1 and DAV2 is 0, the control processor should ignore the data in byte 1 and get the detection status from byte 2.

During handshake operations the TS75C32 reports the detection status regularly. When the rate sequence is received, it will be transferred in byte 1 of the response. Each bit in byte 2 indicates the detection of a specific event in the training sequence. It has different meanings for call and answer modems. For detailed information, refer to appendix B. During the data mode, byte 2 is always provided, but is used only when there are two bytes of data to transmit. This occurs occasionally when the receiver clock is running faster than the transmitter clock.

F00-F11 bits. During the call establishment operation, the V.32 Engine reports call progress tones through the F01 and F00 flags. F00 is set to 1 when the signal energy in frequency band 1 is above the threshold level. F01 is set to 1 when the signal energy in band 2 is above the threshold level. Detection of the 2100Hz answer tone is indicated by setting the F10 flag to a 1.

During handshake operations, all four bits are used to indicate the line condition and some detection

results. F00 is set to 0 if the line quality is good and 1 if it is bad. F01 is set to 1 if any segment in the training sequence is not detected within a time out. This bit can be used to indicate a non V.32 detection if either AA is not detected in the answer modem or the AC is not detected in the call modem. Both F00 and F01 are set to 1 when an illegal mode or a GSTN deardown is received in the rate sequence.

The detection of the rate sequence is reported in the flags F11 and F10. When the modem is operating at 9600 bps without trellis coding, these bits are both set to 0. With trellis coding at 9600bps, F11 is set to 1 and F10 is cleared to 0. For 4800bps, 0 and 1 will be placed in F11 and F10, respectively. When both F11 and F10 are set to 1, the modem has negotiated with the far end modem and determined that the maximum negotiated operating speed is 2400bps.

During data mode, the perceived line quality is reported in the flags F01 and F00. The line conditions are reported as either good (code 00), poor (code 01), or terrible (code 10). The code 10 should be interpreted as a local modern retrain request. Upon receipt of this code, the controller can issue the RTRAcommand to begin the retrain procedure. The code 11 is used when the remote modern begins a retrain sequence. The control processor is then res-ponsible for manipulating the appropriate data communications interface signals.

H and 107 bits. When the TS75C32 is commanded to perform the CCITT handshake sequence, the H bit will be set to 1 for the duration of the handshake operation. At the successful completion of the hand-shake operation the H flag will go to 0 and the control processor is then responsible for manipulating the appropriate data communications interface signals. e.g. 106 and 109. The 107 flag is set to a 1 to indicate that the controller should assert signal 107 on the data communications interface.

3 - COMMAND LIST

Operational Control Commands

Command Mnemonic	OP Code (HEX)	Description		
uFzec	170000	Unfreeze Echo Canceller		
Frezq	1B0000	Freeze the Equalizer Adaptation		
Frezc	160000	Freeze the Echo Canceller Adaptation		
uFzeq	1C0000	Unfreeze Equalizer		
hshk	040000	Handshake with Other Modem		
init	0600C0	Initialize Modem		
jmp	06	Force Code Execution		
nop	000000	No Operation		
rtra	050000	Retrain		
setgn	02	Set the Scaling Factor for the Transmitter		

18/40

SGS-THOMSON MICE TRONICS

7929237 0076402 804

USER INTERFACE - COMMAND SET (continued)

Data Communication Commands

Command Mnemonic	OP Code (HEX)	Description
xmit	03	Transmit Data
xmiti	01	Transmit Data and Initiate Additional Transfer

Mafe Manipulation Commands

Command Mnemonic	OP Code (HEX)	Description
cmafe rrr1 rrr2 wtr1 wtr2	07 080000 090000 0A 0B	Configure MAFE Chipset Read MAFE Reg RR1 Read MAFE Reg RR2 Write MAFE Reg TR1 Write MAFE Reg TR2

Tone Select Commands

Command Mnemonic	OP Code (HEX)	Description	
answ	0C1000	Select 2100Hz Answer Tone	
dtmf 0	0C0000	Select DTMF Digit 0	
dtmf 1	0C0100	Select DTMF Digit 1	
dtmf 2	0C0200	Select DTMF Digit2	
dtmf 3	0C0300	Select DTMF Digit3	
dtmf 4	0C0400	Select DTMF Digit 4	
dtmf 5	0C0500	Select DTMF Digit 5	
dtmf 6	0C0600	Select DTMF Digit 6	
dtmf 7	0C0700	Select DTMF Digit 7	
dtmf 8	0C0800	Select DTMF Digit8	
dtmf 9	0C0900	Select DTMF Digit 9	
dtmf *	0C0E00	Select DTMF Digit *	
dtmf #	0C0F00	Select DTMF Digit#	
tone	OC	Select Tone (s)	

Configuration Control Commands

Command Mnemonic	OP Code (HEX)	Description
cv32 CFSK cv22B cb21 cb212 cb103	20 1D 1E 26 27 28	Configure Modem for V.32 Configure Modem for FSK (V.23 / V.21 / Bell 103) Configure Modem for V.22 / V.22bis / Bell 212A Configure Modem for V.21 Configure Modem for Bell212 Configure Modem for Bell103

Memory Manipulation Commands

Command Mnemonic	OP Code (HEX)	Description
spac	13	Write MEM and Increment MEM Pointer
spam	12	Write MEM
warp	10	Write MEM Pointer & Read MEM
warpx	11	Write MEM Pointer & Read MEM & MEM + 1

Tone Control Commands

Command Mnemonic	OP Code (HEX)	Description	Description		
deft1	0E	Define Tone 1			
deft2	0F	Define Tone 2			
sint	0D0000	Transmit no Tone			
tgen 0	0D0000	Tone Generators Disabled			
tgen 1	0D0100	Tone Generator 1 Enabled			
tgen 2	0D0200	Tone Generator 2 Enabled			
tgen 3	0D0300	Tone Generators 1 & 2 Enabled			

SGS-THOMSON MICROILLECTRONICS

19/40

7929237 0076403 740

APPENDIX A: COMMAND SET DESCRIPTION

CMAFE - Configure the TS68950/1/2 Components of the V.32 Engine

INSTRUCTION TYPE

MAFE manipulation command

OPCODE

07

SYNOPSIS

CMAFE address register code data

DESCRIPTION

CMAFE is used to directly manipulate the operating parameters of the TS68950/1/2 components of the V.32 Engine. This is a low level command which allows the controller to alter such things as the transmit level, transmit timing, receive timing, and receiver parameters, etc. The command consists of a single byte OPcode followed by a byte containing the address code for the desired register and a data byte for the addressed register. The data bytes will be transferred in the order received and interpreted by the addressed device. Refer to the data sheets of the TS68950, TS68951, and TS68952 for programming specifics.

BYTE DEFINITION (OP CODE)

0	0	0	0	0	1	1	1

BYTE 1 DEFINITION

R	R	R	0	0	0	0	0

REG CODE (Refer to TS68950 Data Sheet).

BYTE 2 DEFINITION

*	*	*	*	*	*	*	*
	1						
•							•

DATA BYTE (Refer to TS68950 Data Sheet)

CFSK - Configure for FSK Modes

INSTRUCTION TYPE

Configuration control command

OPCODE

1D

SYNOPSIS

CFSK al orig atn mode II

DESCRIPTION

CFSK is used to select the FSK mode of operation.

BYTE 0 DEFINITION (OP CODE)

	0	0	0	1	1	1	0	1	
--	---	---	---	---	---	---	---	---	--

BYTE 1 DEFINITION

0	0	0	0	AL	MOD1	MOD0	ORIG

BYTDEFINITION

ALAnalog Loopback Enable
ORIGAnswer mode/Originate mode

BYTE 2 DEFINITION

Į	ATN3	ATN2	ATN1	ATN0	0	0	LL	0
---	------	------	------	------	---	---	----	---

BYTDEFINITION

ATN3-0Transmit attenuation

20/40

SGS-THOMSON MICROPLETTROPICS

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CV22B - Configure for CCITT V.22bis Mode, CCITT V.22 and Bell 212A

INSTRUCTION TYPE

Configuration control command

OPCODE

1E

SYNOPSIS

CV22B al mode orig atn LL guard fallback

DESCRIPTION

CV22B is used to select the CCITT V.22B is mode of operation. Select also, depending of Mode bits. the V.22, V.22 600bps and the Bell 212A.

BYTE 0 DEFINITION (OP CODE)

0 0 0 1 1 1 0								
	0	0	0	1	1	1	1	0

BYTE 1 DEFINITION

ſ	G1	G0	0	0	AL	MOD1	MOD0	ORIG

BYTDEFINITION

ALAnalog Loopback Enable

MOD1 O

MOD0 0V.22Bis

Selected Mode 2400bps

BYTE 2 DEFINITION

	· · · · · · · · · · · · · · · · · · ·						
ATN3	ATN2	ATN1	ATN0	0	0	LL	FB

BYTDEFINITION

ATN3-0Transmit attenuation FBFallback mode for V.22bis

CV32 - Configure for V.32 Mode

INSTRUCTION TYPE

Configuration control command

OPCODE

20

SYNOPSIS

CV32 speed ec orig atn al fc

DESCRIPTION

CV32 is used to alter the operating parameters of the V.32 Engine. The passed parameters provide a two bit speed code which selects the desired baud rate. Another parameter explicitly turns on or off the echo canceller. If the V.32 Engine is to operate in the originate mode, the originaremeter must be set. When this parameter is not set, the V.32 Engine is configured as an answer mode device. The all parameter allows the user to select the analog loopback test conditions. The transmit attenuation level is selected by the atn parameter. etc.

BYTE 0 DEFINITION (OP CODE)

0 0 1 0 0 0 0

BYTE 1 DEFINITION

RSV RSV FC EC AL SP1 SP0 ORIG	RSV	RSV	FC	EC	AL	SP1	SP0	ORIG
-------------------------------	-----	-----	----	----	----	-----	-----	------

Flag Bit Definition

BSV

FC

Reserved 0/1 Do not/Do force cleardown

EC

0/1 Echo Canceller off/on ΑL 0/1 Analog Loopback test disabled/enabled

ORIG 0/1 Answer mode / Originate mode

SPEED CODE: SP1-0

00:9600 bps

10:9600 bps treillis

01:4800 bps

11:2400 bps

BYTE 2 DEFINITION

ATN3	ATN2	ATN1	ATN0	0	0	LL	0

Bit Definition Flag

ATN3-0 Transmit attenuation

0 dB to 22 dB : codes 0000 to 1011

in 2 dB steps

Infinite: codes 1100 to 1111

RSV Reserved

21/40

7929237 0076405 513 🖿

DETF1 - Define Tone 1

INSTRUCTION TYPE

Tone control command

OPCODE

0E

SYNOPSIS

DEFT tone descriptor

DESCRIPTION

DEFT1 is a command which used to program tone generator 1. The 16 bit value provided is used as the phase offset per baud for the generator. The DEFT1 command does not enable the tone generator. See also TGEN.

BYTE 0 DEFINITION (OP CODE)

0	0	0	0	1	1	1	0

BYTE 1 DEFINITION

P7	P6	P5	P4	РЗ	P2	P1	P0
----	----	----	----	----	----	----	----

LOW BYTE OF DESCRIPTOR

BYTE 2 DEFINITION

P15	P14	P13	P12	P11	P10	P9	P8

HI BYTE OF DESCRIPTOR

DETF2 - Define Tone 2

INSTRUCTION TYPE

Tone control command

OPCODE

0F

SYNOPSIS

DEFT tone descriptor

DESCRIPTION

DEFT2 is a command which used to program tone generator 2. The 16 bit value provided is used as the phase offset per baud for the generator. The DEFT2 command does not enable the tone generator. See also TGEN.

BYTE 0 DEFINITION (OP CODE)

0	0	0	0	1	1	1	1

BYTE 1 DEFINITION

P7	P6	P5	P4	P3	P2	P1	P0

LOW BYTE OF DESCRIPTOR

BYTE 2 DEFINITION

P15	P14	P13	P12	P11	P10	P9	P8

HI BYTE OF DESCRIPTOR

22/40

SGS-THOMSON

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FREZ - Freeze the Equalizer or Echo Canceller Adaptation

INSTRUCTION TYPE

Operational control command

OPCODES

16/17/1B/1C

SYNOPSIS

FREZ freeze code

DESCRIPTION

FREZ causes the V.32 Engine to enable or disable the adaptation of the equalizer and / or the echo canceller, to the parameter provided.

FREZ includes four different opcodes, each one with a specific option:

Frezc (16) : Freeze the echo canceller adaptation

uFzec (17) : Unfreeze the echo canceller Frezq (1B) : Freeze the equalizer adaptation

uFzeq (1C) : Unfreeze the equalizer

HSHK - Begin Handshake Sequence

INSTRUCTION TYPE

Operational control command

OPCODE

040000

SYNOPSIS

HSHKEA

DESCRIPTION

HSHK is used to command the V.32 Engine to begin the handshake sequence processing. The progress of the handshake is reported to the control processor along with the data bits. For detailed information, refer to appendix B.

BYTE 0 DEFINITION (OP CODE)

	0	0	0	0	0	1	0	0
--	---	---	---	---	---	---	---	---

BYTE 1 DEFINITION (OP CODE)

_								
L	0	0	0	0	0	0	0	0

BYTE 2 DEFINITION (OP CODE)

-								
	0	0	0	0	0	0	0	0

EA: V.25bis

0 enable V.25bis generation/recognition

1 disable V.25bis

INIT - Initialize the V.32 Engine

INSTRUCTION TYPE

Operational control command

OPCODE

0600C0

SYNOPSIS

INIT

DESCRIPTION

INIT forces the V.32 Engine to reset all parameters to their default conditions and restart operations.

BYTE 0 DEFINITION (OP CODE)

	,						
0	0	0	0	0	1	1	0

BYTE 1 DEFINITION (OP CODE)

-								
	0	0	0	0	0	0	0	0

BYTE 2 DEFINITION (OP CODE)

							_
1	1	0	0	0	0	0	0

NOP - No Operation is Specified

INSTRUCTION TYPE

Operational control command

OPCODE

000000

SYNOPSIS

NOP

DESCRIPTION

NOP is used when communications with the V.32 Engine are required but no action is desired.

BYTE 0 DEFINITION (OP CODE)

0 0 0 0 0 0	0	0	0	0	0	0	0	0
-------------	---	---	---	---	---	---	---	---

BYTE 1 DEFINITION (OP CODE)

0 0 0 0	0	0	0	0
---------	---	---	---	---

BYTE 2 DEFINITION (OP CODE)

					,		
0	0	0	0	0	0	0	0

SGS-THOMSON MICROELICTRONICS

23/40

7929237 0076407 396

RRR1 - Read MAFE Register RR1

INSTRUCTION TYPE

MAFE manipulation command

OPCODE

080000

SYNOPSIS

RRR1

DESCRIPTION

RRR1 causes the V.32 Engine to read the 12 bit contents of the MAFE chipset register RR1. The data is returned in a standard three byte format. The least significant data byte is returned in byte 1, followed by the most significant data byte. Byte 0 is the standard response format (refer to appendix B) with DAV1 and DAV2 bits set to 1. Consult the data sheet of the TS68951 for the specifics of the returned data.

BYTE 0 DEFINITION (OP CODE)

ļ	0	0	0	0	1	0	0	0

BYTE 1 DEFINITION (OP CODE)

0	0	0	0	0	0	0	0

BYTE 2 DEFINITION (OP CODE)

0 0 0 0	0 0	0	0	
---------	-----	---	---	--

RRR2 - Read MAFE Register RR2

INSTRUCTION TYPE

MAFE manipulation command

OPCODE

090000

SYNOPSIS

RRR2

DESCRIPTION

RRR2 causes the V.32 Engine to read the 12 bit contents of the MAFE chipset register RR2. The data is returned in a standard three byte format. The least significant data byte is returned in byte 1, followed by the most significant data byte. Byte 0 is the standard response format (refer to appendix B) with DAV1 and DAV2 bits set to 1. Consult the data sheet of the TS68951 for the specifics of the returned data.

BYTE 0 DEFINITION (OP CODE)

0	0	0	0	1	0	0	1

BYTE 1 DEFINITION (OP CODE)

0	0	0	0	0	0	0	0

BYTE 2 DEFINITION (OP CODE)

					-		
0	0	0	0	0	0	0	0

24/40

SGS-THOMSON MICROHIECTRONIC

7929237 0076408 222

RTRA - Force a Retrain of the V.32 Engine

INSTRUCTION TYPE

Operational control command

OPCODE

050000

SYNOPSIS

RTRAfall

DESCRIPTION

RTRA is used to force the V.32 Engine to initiate a retrain sequence on the channel.

BYTE 0 DEFINITION (OP CODE)

0	0	0	0	1	0	0	1

BYTE 1 DEFINITION (OP CODE)

					-		
0	0	0	0	0	0	0	0

BYTE 2 DEFINITION (OP CODE)

0	0	0	0	0	0	0	FALL

Fallback (V.22Bis only) 1 Fallback to 1200Bps 0 goes up to 2400Bps

SETGN - Set Global Gain Factor

INSTRUCTION TYPE

Operational control command

OPCODE

02

SYNOPSIS

STEGN gain value

DESCRIPTION

SETGN is a command which used to scale the transmit samples. The 16 bit value provided is used as the multiplicative constant to be multiplied with each transmit sample.

BYTE 0 DEFINITION (OP CODE)

0	0	0	0	0	0	1	0

BYTE 1 DEFINITION

G7	G6	G5	G4	G3	G2	G1	G0

LOW BYTE OF GAIN VALUE

BYTE 2 DEFINITION

G15	G14	G13	G12	G11	G10	G9	G8

HI BYTE OF GAIN VALUE

25/40

SLNT - Disable Tone Generators

INSTRUCTION TYPE

Tone command

OPCODE

0D0000

SYNOPSIS

SLNT

DESCRIPTION

SLNT causes the V.32 Engine to disable the tone generators, thus stopping the tone output (i.e. send silence).

BYTE 0 DEFINITION (OP CODE)

0	0	0	0	1	1	0	1

BYTE 1 DEFINITION

0	0	0	0	0	0	0	0

BYTE 2 DEFINITION (OP CODE)

0	0	0	0	0	0	0	0	

SPAC - Store Parameter and Count

INSTRUCTION TYPE

Memory manipulation command

OPCODE

13

SYNOPSIS

SPAC lo-byte hi-byte

DESCRIPTION

SPAC is a command which used to write an arbitrary 16 bit value into the writable memory location currently specified by the Memory Address Register. The content of the Memory Address Register is incremented by 1 at the completion of command execution. See also WARP.

BYTE 0 DEFINITION (OP CODE)

0	0	0	1	0	0	1	1

BYTE 1 DEFINITION

D7	D6	D5	D4	D3	D2	D1	D0
----	----	----	----	----	----	----	----

LOW BYTE OF DATA

BYTE 2 DEFINITION

D15 D14 D13 C	2 D11	D10	D9	D8
---------------	-------	-----	----	----

HI BYTE OF DATA

26/40

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SPAM - Store Parameter in Addressed Memory

INSTRUCTION TYPE

Memory manipulation command

OPCODE

12

SYNOPSIS

SPAM lo-byte hi-byte

DESCRIPTION

SPAM is a command which used to write an arbitrary 16 bit value into the writable memory location currently specified by the Memory Address Register. See also WARP.

BYTE 0 DEFINITION (OP CODE)

	 _				-4	_
LU	 U	1	U	υ	1	U

BYTE 1 DEFINITION

D7	D6	D5	D4	D3	D2	D1	D0

LOW BYTE OF DATA

BYTE 2 DEFINITION

D15	D14	D13	D12	D11	D10	D9	D8

HI BYTE OF DATA

TGEN - Enable and Disable Tone Generators

INSTRUCTION TYPE

Tone control command

OPCODE

OD.

SYNOPSIS

TGEN tg code

DESCRIPTION

TGEN causes the V.32 Engine to enable or disable tone generator 1 and tone generator 2, according to the parameter provided.

BYTE 0 DEFINITION (OP CODE)

0 0 0 0 1 1 0 1								
	0	0	0	0	1	1	0	1

BYTE 1 DEFINITION (TONE CODE)

0	0	0	0	0	0	TGC2	TGC1	
	G COD		TON	IE GE	N 1	TONE GEN 2		
	00			isable	d	Disabled		
	01			nable	t	Disabled		
	10			isable		Ena	bled	
	11			Enabled			bled	

TONE - Select and Transmit Tone(s)

INSTRUCTION TYPE

Tone select and command

OPCODE

0C

SYNOPSIS

TONE tone code

DESCRIPTION

TONE causes the V.32 Engine to program the tone generators for the specified tone or tones. The tones are defined by the tone code parameter passed in the second byte of the command. See also tonetab for the predefined single and double tones, and the commands DEFT and TGEN for user definable tones and tone generator control.

BYTE 0 DEFINITION (OP CODE)

	,						
0	0	0	0	1	1	0	0

BYTE 1 DEFINITION (TONE CODE)

T 7	T6	T5	T4	T3	T2	T1	T0

BYTE 2 DEFINITION (OP CODE)

! 0	0	0	0	O	n	n	n
L		<u> </u>				U	

HI BYTE OF DATA

Tone Code	Tone Parameters
0	DTMF 0 (941 & 1336Hz)
1	DTMF 1 (697 & 1209Hz)
2	DTMF 2 (697 & 1336Hz)
3	DTMF 3 (697 & 1477Hz)
4	DTMF 4 (770 & 1209Hz)
5	DTMF 5 (770 & 1336Hz)
6	DTMF 6 (770 & 1477Hz)
7	DTMF 7 (852 & 1209Hz)
8	DTMF 8 (852 & 1336Hz)
9	DTMF 9 (852 & 1477Hz)
Α	(697 & 1633Hz)
В	(770 & 1633Hz)
С	(852 & 1633Hz)
D	(941 & 1633Hz)
Е	DTMF * (941 & 1209Hz)
F	DTMF # (941 & 1477Hz)
10	Answer tone (2100Hz)

28/40

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WARP - Write Address & Return

Parameter

INSTRUCTION TYPE

Memory manipulation command

OPCODE

10

SYNOPSIS

WARP address

DESCRIPTION

WARP is a command which is used to write the Memory Address Register of the V.32 Engine. The V.32 Engine responds with the contents of the addressed location. The data is returned in a standard three byte transfer. The least significant data byte is returned in the byte 1, followed by the most significant data byte. Byte 0 is the standard response format (refer to appendix B) with DAV1 and DAV2 bits set to 1.

BYTE 0 DEFINITION (OP CODE)

0	0	0	1	0	0	0	0

BYTE 1 DEFINITION

A7	A6	A5	A4	A3	A2	A1	A0

LOW BYTE OF DATA

BYTE 2 DEFINITION

P1	P0	M1	M0	A11	A10	A9	8A

PROC CODE P1-0	MEM CODE M1-0	ADDRESS HI A11-A8
00 : Master	00:XRAM	
10 : Receiver	01 : YRAM	
01 : Echo Canceller	10 : EMEM	

11 : CROM

WARPX - Write Address & Return Parameter Complex

INSTRUCTION TYPE

Memory manipulation command

OPCODE

11

SYNOPSIS

WARPX address

DESCRIPTION

WARPX is a command which is used to write the Memory Address Register of the V.32 Engine. The V.32 Engine responds with the contents of the most significant bytes of the addressed location and the addressed location +1. The data is returned in a stan-dard three byte transfer. Byte 0 is the standard res-ponse format (refer to appendix B) with DAV1 and DAV2 bits set to 1. Byte 1 is used to return the 8 most significant bits contained in the addressed location. The 8 most significant bits of the addressed location +1 are returned in byte 2.

BYTE 0 DEFINITION (OP CODE)

0	0	0	1	0	0	0	0

BYTE 1 DEFINITION

A7	A6	A5	A4	А3	A2	A1	A0

LOW BYTE OF DATA

BYTE 2 DEFINITION

P1	P0	M1	MO	A11	A10	A9	A8

	PROC CODE P1-0	MEM CODE M1-0	ADDRESS HI A11-A8
	00 : Master 10 : Receiver	00:XRAM 01:YRAM	
(01 : Echo Canceller	10 : EMEM 11 : CROM	

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29/40

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WTR1 - Write MAFE Register TR1

INSTRUCTION TYPE

MAFE manipulation command

OPCODE

0A

SYNOPSIS

WTR₁

DESCRIPTION

WTR1 causes the V.32 Engine to take the two supplied data bytes and write them in sequence to the MAFE chipset register TR1.

BYTE 0 DEFINITION (OP CODE)

0	0	0	0	1	0	1	0

BYTE 1 DEFINITION (DATA)

D3	D2	D1	D0	0	0	0	0

LOW BYTE OF DATA

BYTE 2 DEFINITION (DATA)

D11	D10	D9	D8	D7	D6	D5	D4

WTR2 - Write MAFE Register TR2

INSTRUCTION TYPE

MAFE manipulation command

OPCODE

0B

SYNOPSIS

WTR2

DESCRIPTION

WTR2 causes the V.32 Engine to take the two supplied data bytes and write them in sequence to the MAFE chipset register TR2.

BYTE 0 DEFINITION (OP CODE)

					-			
0	0	0	0	1	0	1	1	1

BYTE 1 DEFINITION (DATA)

DЗ	D2	D1	D0	0	0	0	0

LOW BYTE OF DATA

BYTE 2 DEFINITION (DATA)

D11	D10	D9	D8	D7	D6	D5	D4

30/40

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XMIT - Transmit Data to Other Modem

INSTRUCTION TYPE

Data communications command

OPCODE

03

SYNOPSIS

XMIT data

DESCRIPTION

XMIT is used to command the V.32 Engine to send data. The OP code for the xmit command is a single byte. The data bits to be transmitted are stored in the second byte, where D0 is the first bit to be transmitted.

BYTE 0 DEFINITION (OP CODE)

	1						
0	0	0	0	0	0	0	1

BYTE 1 DEFINITION

		r					
D7	D6	D5	D4	D3	D2	D1	D0

D0-D7 DATA BITS

BYTE 2 DEFINITION

0	0	0	0	0	0	0	0

XMITI - Transmit Data to Other Modern and Initiate Additional Cycle

INSTRUCTION TYPE

Data communications command

OPCODE

01

SYNOPSIS

XMIT data

DESCRIPTION

XMIT is used to command the V.32 Engine to send data. The OP code for the xmit command is a single byte. The data bits to be transmitted are stored in the second byte, where D0 is the first bit to be transmitted.

BYTE 0 DEFINITION (OP CODE)

0	0	0	0	0	0	1	1

BYTE 1 DEFINITION

ĺ	D7	D6	D5	D4	D3	D2	D1	D0

D0-D7 DATA BITS

BYTE 2 DEFINITION

0	0	0	0	0	0	0	0

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APPENDIX B: STATUS REPORTING DESCRIPTION

BYTE 0 DEFINITION

F11	F10	F01	F00	DAV2	DAV1	Н	CD	Ì
					L			

L				. 00	DAVE	DAVI	н	GD
Flag Code	Call Estab	lish	Hand	ishake Operat	ions	Data XFER Op	erations	
00 01 10 11	no tones Band 1 det Band 2 det Both bands	ected	line to	juality is good errible (local ret out leardown (V.32		line quality is go line quality is po line terrible (loc remote retrain s	oor al retrain reg.)	tted
F11-10 00 01 10 11	reserved Answer ton V.32 : AC d reserved		4800	bps no trellis or bps or 1200bps bps trellis bps		reserved reserved reserved reserved	- 4	
DAV1	DAV2	Definition						
0 0 1 1	0 1 0 1	No data bit	yte 1 : s and	and 2. and status word status word is i tt command is i	n byte 2.	d 2.		
FLAG	BIT	Definition						
H CD	0/1 0/1	Set circuit 1	107 off	t/is in progress f/on (V.32) /off (V.22Bis)				

BYTE 1 DEFINITION

								
RD7	RD6	RD5	RD4	RD3	RD2	RD1	RD0	
							1100	

BYTE 2 DEFINITION

0.77	·					
ST7 ST6 ST6	STA	CTO	0			
	314	ST3	ST2	ST1	OTO I	
				· · · ·	J 510	

Flag Code			Operations		Data XFER Operations
F01-00	V.	.32	V.2	2Bis	
	Answer	Originate	Answer	Originate	
ST7 ST6 ST5 ST4 ST3 ST2 ST1 ST0	AA AA-CC silence S S/S R2 R3 E	AC AC-CA CA-AC S S/S R1 E	S1 "1", 1.2 K "1", 2.4 K - - - -	Unscr. 1 S1 "1", 1.2 K "1", 2.4 K	When DAV1, DAV2 bits are 0 an additional byte of receive data is available.

32/40

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APPENDIX C

This appendix describes the interconnection between the different chips (in DIP package).

Signal Name	Chip/Pin	Description	
SYSTEM INTERF	ACE		
DOHD7H	TS75C321/2734	System Data Bus : connect to host processor	
CSL	TS75C321/21	Chip Select : connect to host processor	
RSL	TS75C321/22	Register Select : connect to host processor	
DSL	TS75C321/20	Data Strobe : connect to host	
RWL	TS75C321/19	Read/Write : connect to host	
INTL	TS75C321/24	Interrupt Request : connect to host processor	
RSTL1 RSTL2 RSTL0	TS75C321/23 TS75C322/23 TS75C320/23	Reset : connect to host processor Reset Reset	
CLOCK SIGNALS	3		
TxRCLK	TS68952/16 TS75C321/26	Transmit Baud Clock	
TxCCLK	TS68952/24 TS75C321/44 TS68950/19 TS68951/23	Transmit Conversion Clock	
RxRCLK	TS68952/20 TS75C321/43 TS75C322/44	Receive Baud Clock	
RxCCLK	TS68952/21 TS75C321/25 TS75C322/43 TS68951/22	Receive Conversion Clock	
TxSCLK	TS68952/11	If not used must be grounded.	
XTL1	TS68952/13	External Crystal Input: must be connected via a 5.76MHz crystal to XTL2	
XTL2	TS68952/14	External Crystal Input	
CLK	TS68952/15 TS68950/18 TS68951/21	Main Analog Clock: this output, in accordance with the XTL1/2 crystal, must be 1.4MHz (±7Hz).	
25 MHz	TS75C320/15 TS75C321/15 TS75C322/15	Main Digital Clock: connect to a 25MHz oscillator	
TxCCLK RxCCLK	TS68952/23 TS68952/22	Transmit Bit Clock Receive Bit Clock	
ANALOG SIGNA	LS		
ATO	TS68950/15	Analog Transmit Output : connect to DAA	
EEO	TS68950/16 TS68951/11	Analog Echo Cancelling Estimation	
LEI	TS68951/17	Local Echo Input : must be grounded	
RAI	TS68951/16	Receive Analog Input: connect to DAA	
RFO	TS68951/13	This pin must be connected throught a $1\mu\text{F}$ nonpolarised capacitor to AGC1 input.	
AGC1	TS68951/12		
AGC2	TS68951/19	Connect to the analog ground through a 1µF nonpolarised capacitor	
CD1	TS68951/18	Connect to the analog ground through a 1µF nonpolarised capacitor	

Caution: The connection between EEO (TS68950/16) and EEI (TS68951/11) must be as close as possible to avoid parasitics on echo estimate signal



33/40

APPENDIX C (continued)

Signal Name	Chip/pi n	Description		
INTER DSP AND EX	TERNAL MEMORY CONNECTION			
0D00D15	TS75C320/4548,112 RAM0/IO0IO15	Data Bus		
1D81D15	TS75C321/512 - TS75C320/2734 TS68950/2024,13 RAM0/AD0AD7	Data and Address Buses		
1D01D7	TS75C321/4548,14 - TS75C322/2734 RAM2/AD0AD7	Data and Address Buses		
2D92D15	TS75C322/612 - TS68951/2528,13 TS68952/2528,13 RAM2/IO9IO15	Data Bus		
2D8	TS75C322/5 - TS68951/24 RAM2/IO8	Data Pin		
2D02D7	TS75C322/4548,14 RAM2/IO0IO7	Data Bus		
1A11	TS75C321/39 - TS75C320/21 TS75C322/21 - TS68950/7	Address Line		
1A10	TS75C321/37 - TS68950/6	Adress Line		
1A9	TS75C321/36 - TS68950/9 - TS75C320/22	Address Line		
1A8	TS75C321/35 - TS68950/8 - TS75C322/22	Address Line		
1RWL	TS75C321/18 - TS75C320/19 TS75C322/19 - TS68950/5	Control Line		
1DSL	TS75C321/17 - TS75C320/20 TS75C322/20 - TS68950/4	Control Line		
0A80A11	TS75C320/3537,39 RAM0/AD8AD11	Address Line		
0DSL	TS75C320/17 RAM0/CEL	Control Line		
0RWL	TS75C320/18 RAM0/WEL	Control Line		
2A82A11	TS75C322/3537,39 - TS68951/8,9,6,7 TS68952/8,9,6,7 RAM2/A8A10,CEL	Address Line		
2DSL	TS75C322/17 - TS68951/4 - TS68952/4 RAM2/OEL	Control Line		
2RWL	TS75C322/18 - TS68951/5 - TS68952/5 RAM2/WEL	Control Line		
0IRQL	TS75C320/24 - TS75C321/42	Synchro Line		
2IRQL	TS75C322/24 - TS75C321/41	Synchro Line		

Note: RAMO Refer to DSP0 4Kx16 External memory. RAM2 Refer to DSP2 2Kx16 External memory.

Where: IO is bidirectionel data bus
AD is address line
WEL is Write Enable (active low)
CEL is Chip Select (active low)
OEL is Output Enable (Optional)

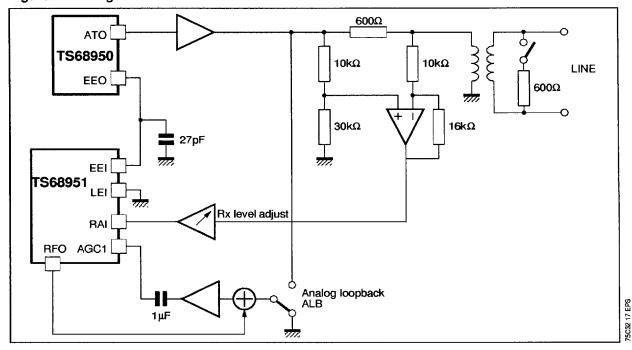
34/40



APPENDIX C (continued)

Signal Name	Chip/Pin	Description
OWER SUPPLY	ANS MISCELLANEOUS	
+5VA	T\$68951/20 - T\$68950/17	Positive Analog Power Supply
-5VA	TS68951/14 - TS68950/12	Negative Analog Power Supply
AGND	TS68950/13 - TS68951/15	Analog Ground
Vcc	TS75C320/38 - TS75C321/38 - TS75C322/38 - TS68952/17	Main Digital Power Supply
DGND	TS75C320/13 - TS75C321/13 - TS75C322/13 TS68950/10 - TS68951/10 - TS68952/12	Digital Ground Power Supply
xtal	TS75C320/14 - TS75C321/14 - TS75C322/14	Not Connected (must be left open)
Clkout	TS75C320/16 - TS75C321/16 - TS75C322/16	Not Connected (25MHz/4)
ТО	TS68952/10	Not Connected (must be left open)
AGND	TS68950/14	Auxiliary Input
DGND	TS75C321/40 - TS75C320/4044 - TS75C320/2526 TS75C322/4042 - TS75C322/2526 - TS68950/11	Not Used

Figure 9: Analog Path



APPENDIX D: REFERENCES

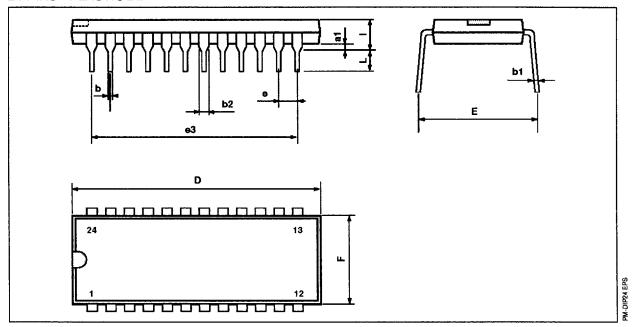
- 1. CCITT recommendation V.32.
- 2. CCITT recommendation V.54.
- 3. Data sheet of the TS75320 , V.32 modem echo canceller, SGS-THOMSON.
- Data sheet of the TS68930, TS68931 programmable signal processor, SGS-THOMSON.
- Data sheet of the TS68950 transmitter interface chip, SGS-THOMSON.
- 6. Data sheet of the TS68951 receiver interface chip, SGS-THOMSON.
- 7. Data sheet of the TS68952 clock generation chip, SGS-THOMSON.
- Application guide: Using the TS75320 Echo canceller in V.32 modems, SGS-THOMSON.

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35/40

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24 PINS - PLASTIC DIP



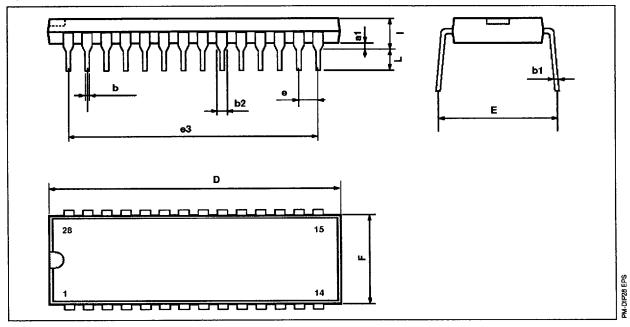
Dimensions		Millimeters				
Dimensions	Min.	Тур.	Max.	Min.	Тур.	Max.
a1	,	0.63			0.025	
b		0.45			0.018	
b1	0.23		0.31	0.009		0.012
b2		1.27			0.050	
D			32.2			1.268
E	15.2		16.68	0.598		0.657
е		2.54			0.100	
e3		27.94			1.100	
F			14.1			0.555
i		4.445			0.175	
L	•	3.3			0.130	

36/40

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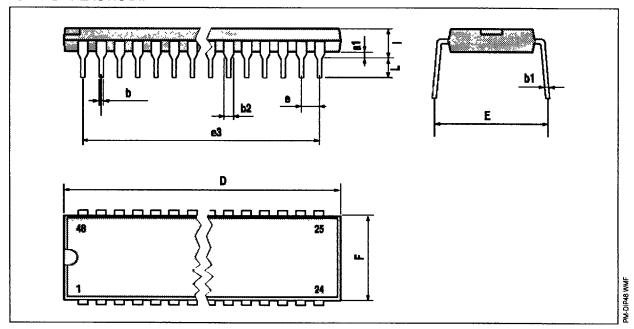
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28 PINS - PLASTIC DIP



B		Millimeters			Inches	
Dimensions	Min.	Тур.	Max.	Min.	Тур.	Max.
a1		0.63			0.025	
b		0.45			0.018	
b1	0.23		0.31	0.009		0.012
b2		1.27			0.050	
D			37.4			1.470
E	15.2		16.68	0.598		0.657
е		2.54			0.100	
e3		33.02			1.300	
F			14.1			0.555
i		4.445			0.175	
L		3.3			0.130	

48 PINS - PLASTIC DIP

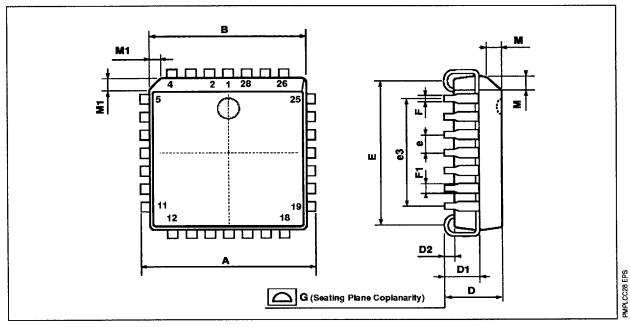


Dimensions		Millimeters		Inches			
Dilliensions	Min.	Тур.	Max.	Min.	Тур.	Max.	
a1		0.63			0.025		
b		0.45			0.018		
b1	0.23		0.31	0.009		0.012	
b2		1.27			0.050		
D			62.74			2.470	
E	15.2		16.68	0.598		0.657	
е		2.54			0.100		
e3		58.42			2.300		
F			14.1			0.555	
i		4.445			0.175		
L		3.3			0.130		

38/40

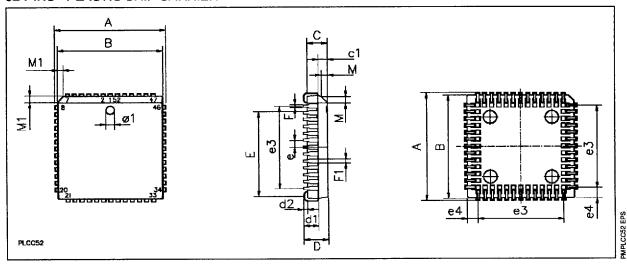
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28 PINS - PLASTIC CHIP CARRIER



D		Millimeters			Inches	
Dimensions	Min.	Тур.	Max.	Min.	Тур.	Max.
Α	12.32		12.57	0.485		0.495
В	11.43		11.58	0.450		0.456
D	4.2		4.57	0.165		0.180
D1	2.29		3.04	0.090		0.120
D2	0.51			0.020		
E	9.91		10.92	0.390		0.430
е		1.27			0.050	
e3		7.62			0.300	
F		0.46			0.018	
F1		0.71			0.028	
G			0.101			0.004
М		1.24			0.049	
M1		1.143			0.045	

52 PINS - PLASTIC CHIP CARRIER



Dimensions		Millimeters			inches	
Dimensions	Min.	Тур.	Max.	Min.	Тур.	Max.
Α		4.20	5.08		0.165	0.20
A1		0.51			0.020	
A3		2.29	3.30		0.090	0.13
В		0.33	0.53		0.013	0.021
B1		0.66	0.81		0.026	0.032
С	0.25			0.01		
D		19.94	20.19		0.785	0.795
D1		19.05	19.20		0.750	0.756
D2		17.53	18.54		0.690	0.730
D3	15.24			0.60		
E		19.94	20.19		0.785	0.795
E1		19.05	19.20		0.750	0.756
E2		17.53	18.54		0.690	0.730
E3	15.24			0.60		
е	1.27			0.05		
L		0.64			0.025	
L1		1.53			0.060	
М		1.07	1.22		0.042	0.048
M1		1.07	1.42		0.042	0.056

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40/40

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