

Low Power, Stereo A/D Converter for Digital Audio

Features

- Single +5 V Power Supply
- 91 dB Dynamic Range
- Linear Phase Digital Anti-Alias Filtering
0.05dB Passband Ripple
73.5dB Stopband Rejection
- Low Power Dissipation: 100 mW
Power-Down Mode for Portable Applications
- Complete CMOS Stereo A/D System
Delta-Sigma A/D Converters
Digital Anti-Alias Filtering
S/H Circuitry and Voltage Reference
- Adjustable System Sampling Rates
including 32kHz, 44.1 kHz & 48kHz
- Internal 64X Oversampling
- Evaluation Board Available

General Description

The CS5345 is a complete stereo analog-to-digital converter which performs anti-alias filtering, sampling and analog-to-digital conversion generating 16-bit values for both left and right inputs in serial form. The output word rate can be up to 50 kHz per channel.

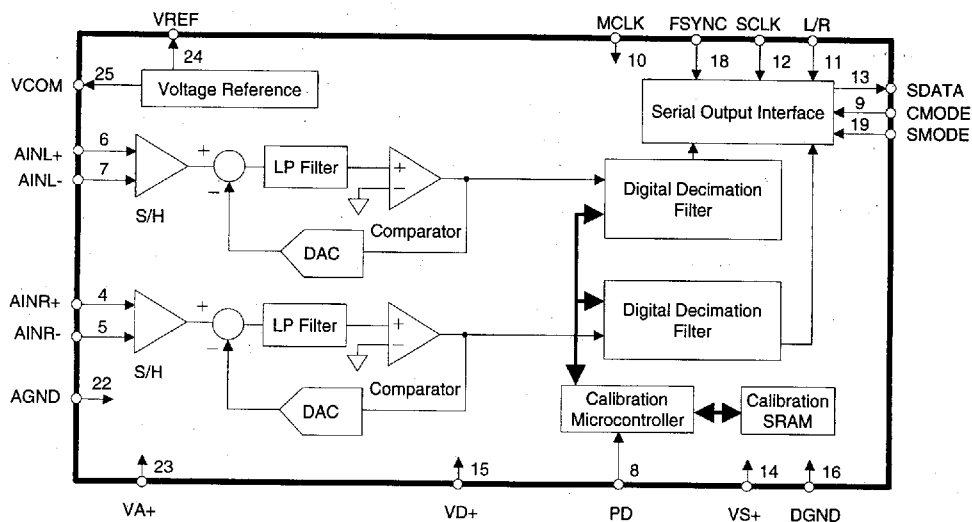
The CS5345 operates from a single +5V supply and requires only 100 mW for normal operation, making it ideal for battery-powered applications.

The ADC uses delta-sigma modulation with 64X oversampling, followed by digital filtering and decimation, which removes the need for an external anti-alias filter. The linear-phase digital filter has a passband of dc to 22 kHz, 0.05 dB passband ripple and >73.5 dB stopband rejection.

The device is available in a 0.3" wide 28-pin SOIC surface mount package.

ORDERING INFORMATION:

Model	Temp. Range	Package Type
CS5345-KS	0° to 70°C	28-pin plastic SOIC
CDB5345		Evaluation Board



Preliminary Product Information

This document contains information for a new product. Crystal Semiconductor reserves the right to modify this product without notice.

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SEPT '93
DS112PP1
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ANALOG CHARACTERISTICS ($T_A = 25^{\circ}\text{C}$; V_{A+} , V_{S+} , $V_{D+} = 5\text{V}$; Full-Scale Input Sinewave, 1kHz; Output word rate = 48 kHz; SCLK = 3.072 MHz; Source Impedance = 150Ω with 1500 pF across A_{IN+} , A_{IN-} ; Measurement Bandwidth is 10 Hz to 20 kHz unless otherwise specified, Logic 0=0V, Logic 1= V_{D+} .)

Parameter	Symbol	Min	Typ	Max	Units
Resolution		16	-	-	Bits
Dynamic Performance					
Dynamic Range	A-weighted	-	91 88	-	dB dB
Total Harmonic Distortion+Noise	0dB -20dB -60dB	THD+N	-	-86 -68 -28	dB dB dB
Total Harmonic Distortion	0dB	THD	-	.0018	%
Interchannel Phase Deviation			-	0.0001	Degree
Interchannel Isolation	(dc to 20 kHz)		-	100	dB
dc Accuracy					
Interchannel Gain Mismatch			-	0.05	dB
Gain Error			-	±2	%
Gain Drift			-	60	ppm/°C
Bipolar Offset Error	(After Calibration)		-	±5	±30 LSB
Analog Input					
Differential Input Voltage Range (Full Scale) (Note 1)	VIN	3.46	3.75	4.06	Vpp
Input Impedance	ZIN	-	85	-	kΩ
Power Supplies					
Power Supply Current	VA+	IA+	-	10.5	16 mA mA
Normal Operation	(VD+) + (VS+)	ID+	-	9.5	14
Power Supply Current	VA+	IA+	-	50	μA
Power-Down Mode	(VD+) + (VS+)	ID+	-	50	μA
Power Dissipation			-	100	150 mW mW
	Power Down		-	0.5	-

Notes: 1. The peak-to-peak input voltage range for each input is equal to $\{(V_{A+}) - V_{REF}\} \times 0.625$. The nominal value for $(V_{A+}) - V_{REF}$ is 3 Volts. The differential peak-to-peak input voltage is equal to twice the individual voltage range. (See Figure 5)

* Refer to Parameter Definitions at the end of this data sheet.

Specifications are subject to change without notice.

DIGITAL FILTER CHARACTERISTICS

($T_A = 25\text{ }^\circ\text{C}$; V_{A+} , V_{S+} , $V_{D+} = 5V \pm 5\%$; Output word rate (F_s) of 48 kHz)

Parameter	Symbol	Min	Typ	Max	Units
Passband	(-0.05 dB)	-	0 to 18	-	kHz
	(-2.8 dB)	-	0 to 22	-	kHz
	(-6.5 dB)	-	0 to 24	-	kHz
Passband Ripple		-	-	± 0.05	dB
Stopband		32	-	3040	kHz
Stopband Attenuation	(Note 2)	73.5	-	-	dB
Group Delay (OWR = Output Word Rate)	(Note 3) t_{gd}	-	8/OWR	-	s
Group Delay Variation vs. Frequency	Δt_{gd}	-	-	0.0	μs

Notes: 2. The analog modulator samples the input at 3.072MHz for an output word rate of 48 kHz. There is no rejection of input signals which are multiples of the sampling frequency ($n \times 3.072\text{MHz} \pm 18\text{kHz}$ where $n = 0, 1, 2, 3, \dots$).

3. Group delay for OWR = 48kHz, $t_{gd} = 8/48\text{kHz} = 167\mu\text{s}$

DIGITAL CHARACTERISTICS

($T_A = 25\text{ }^\circ\text{C}$; V_{A+} , V_{S+} , $V_{D+} = 5V \pm 5\%$)

Parameter	Symbol	Min	Typ	Max	Units
High-Level Input Voltage	V_{IH}	70% V_{D+}	-	-	V
Low-Level Input Voltage	V_{IL}	-	-	30% V_{D+}	V
High-Level Output Voltage at $I_o = -20\text{ }\mu\text{A}$	V_{OH}	4.4	-	-	V
Low-Level Output Voltage at $I_o = 20\text{ }\mu\text{A}$	V_{OL}	-	-	0.1	V
Input Leakage Current	I_{in}	-	1.0	-	μA

ABSOLUTE MAXIMUM RATINGS (AGND, DGND = 0V, all voltages with respect to ground.)

Parameter	Symbol	Min	Typ	Max	Units
DC Power Supplies:					
Positive Analog	V_{A+}	-0.3	-	+6.0	V
Positive Digital	V_{D+}	-0.3	-	+6.0	V
Input Current, Any Pin Except Supplies	I_{in}	-	-	± 10	mA
Analog Input Voltage	V_{INA}	-0.3	-	$(V_{A+}) + 0.3$	V
Digital Input Voltage	V_{IND}	-0.3	-	$(V_{D+}) + 0.3$	V
Ambient Temperature (power applied)	T_A	-55	-	+125	$^\circ\text{C}$
Storage Temperature	T_{stg}	-65	-	+150	$^\circ\text{C}$

WARNING: Operation at or beyond these limits may result in permanent damage to the device.
Normal operation is not guaranteed at these extremes.

SWITCHING CHARACTERISTICS

(T_A = 25 °C; V_A+, V_S+, V_D+ = 5V ± 5%; Inputs: Logic 0 = 0V, Logic 1 = V_A+, V_S+, V_D+; C_L = 20 pF)

Parameter	Symbol	Min	Typ	Max	Units
MCLK Period (CMODE low)	t _{clkw1}	78.13	-	781.3	ns
MCLK Low (CMODE low)	t _{clk1}	31.25	-	-	ns
MCLK High (CMODE low)	t _{clkh1}	31.25	-	-	ns
MCLK Period (CMODE high)	t _{clkw2}	52.0	-	520.8	ns
MCLK Low (CMODE high)	t _{clk2}	20.83	-	-	ns
MCLK High (CMODE high)	t _{clkh2}	20.83	-	-	ns
L/R duty cycle		25	-	75	%
SCLK Period	t _{sclkw}	312.5	-	-	ns
SCLK Pulse Width Low	t _{sclkl}	100	-	-	ns
SCLK Pulse Width High	t _{sclkh}	100	-	-	ns
SCLK falling to SDATA valid	t _{dss}	-	-	70	ns
L/R edge to MSB valid	t _{lrdss}	-	-	70	ns
PD pulse width	t _{pdw}	150	-	-	ns
PD falling to SDATA valid	t _{pcf}	-	8224/OWR	-	s
Rising SCLK to L/R edge delay	t _{slr1}	30	-	-	ns
L/R edge to rising SCLK setup time	t _{slr2}	30	-	-	ns
SCLK Falling to FSYNC delay	t _{fs}	-70	-	+70	ns

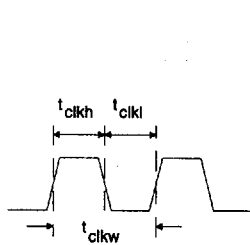
RECOMMENDED OPERATING CONDITIONS

(AGND, DGND = 0V; all voltages with respect to ground)

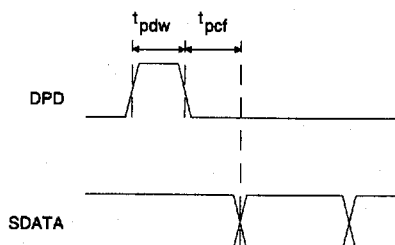
Parameter			Symbol	Min	Typ	Max	Units
DC Power Supplies:	Positive Analog (Note 5)	VA+	4.75	5.0	5.25	V	
	Positive Digital	VD+	4.75	5.0	5.25	V	
	Positive Substrate	VS+	4.75	5.0	5.25		
Differential Analog Input Voltage (Notes 1&6)			V _{IN}	-	3.75	-	V _{pp}
Analog Input Bias Voltage				-	0.5VA+	-	V

Notes: 5. V_D+ must be within 0.3V of V_A+

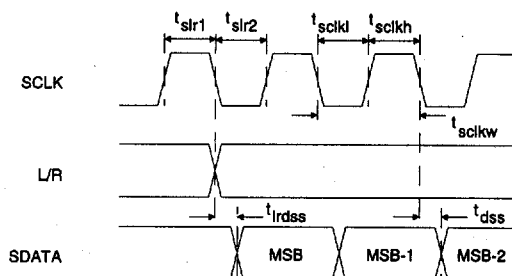
6. The output codes will clip at full scale with differential input signals >3.75 V_{pp}.



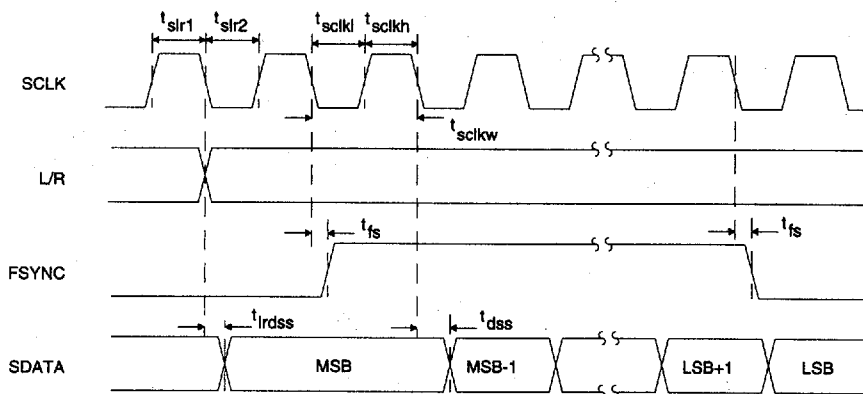
MCLK Definition



Power Down & SDATA Valid



**SCLK to L/R & SDATA-Mode 1 (SMODE High), FSYNC Ignored
Mode 2 - FSYNC High (SMODE Low)**



SCLK to L/R, SDATA & FSYNC - Mode 2 - FSYNC Controlled (SMODE Low)

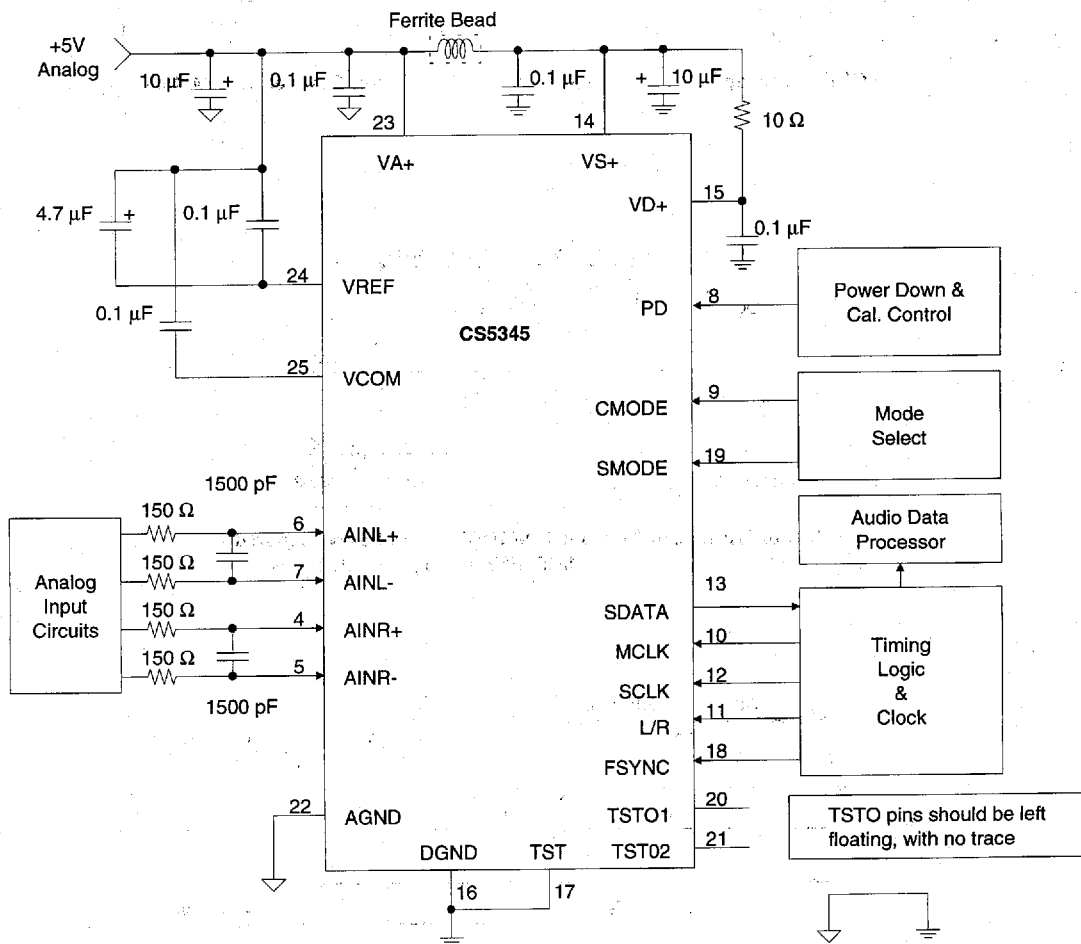


Figure 1. Typical Connection Diagram

GENERAL DESCRIPTION

The CS5345 is a 16-bit, 2-channel A/D converter designed for stereo digital audio applications that require a single +5V supply. The device uses two one-bit delta-sigma modulators which simultaneously sample the analog input signals at a 64 X sampling rate. The resulting serial bit streams are digitally filtered, yielding pairs of 16-bit values. This technique yields nearly ideal conversion performance independent of input frequency and amplitude. The converters do not require difficult-to-design or expensive anti-alias filters and do not require external sample-and-hold amplifiers or a voltage reference.

An on-chip voltage reference provides for a differential input signal range of 3.75 Vpp. Offsets are internally calibrated out during a power-up self-calibration cycle. Output data is available in serial form, coded as 2's complement 16-bit numbers. Typical power consumption is 100 mW. This can be further reduced to .05 mW using the power-down mode.

For more information on delta-sigma modulation and the particular implementation inside this ADC, see the references at the end of this data sheet.

SYSTEM DESIGN

Very few external components are required to support the ADC. Normal power supply decoupling components, voltage reference bypass capacitors and a resistor and capacitor on each input for anti-aliasing are all that's required, as shown in Figure 1.

Master Clock Input

The master input clock (MCLK) into the ADC runs the digital filter and it is used to generate the delta-sigma modulator sampling clock. CMODE high will set the required MCLK frequency to 384 X the output word rate (OWR), while CMODE low will set the required MCLK frequency to 256 X OWR. Table 1 shows some common clock frequencies.

Serial Data Interface

The serial data interface has 2 possible modes of operation. L/R, SCLK and FSYNC are inputs in both interface modes. L/R must be derived from MCLK and be equal to the OWR. SCLK should also be derived from MCLK and be equal to the delta-sigma modulator sample rate (64 X OWR) to eliminate the possibility of interference tones in the output data. An SCLK frequency of 32 X OWR is possible but may cause interference tones. Data bits are clocked out via the SDATA pin using SCLK, L/R and FSYNC inputs. The serial nature of the output data results in the left and right data words being read at different times. However, the words within a L/R cycle represent simultaneously sampled analog inputs.

Mode 1 (SMODE high) is shown in Figure 2. The falling edge of SCLK causes the ADC to output each data bit. Notice the one SCLK cycle delay between L/R edges and the data MSB. FSYNC is ignored and should be tied either High or Low in this mode. Mode 1 is compatible with I²S.

L/R (kHz)	CMODE	MCLK (MHz)	SCLK (MHz)
32	low	8.192	2.048
32	high	12.288	2.048
44.1	low	11.2896	2.8224
44.1	high	16.9344	2.8224
48	low	12.288	3.072
48	high	18.432	3.072

Table 1. Common Clock Frequencies

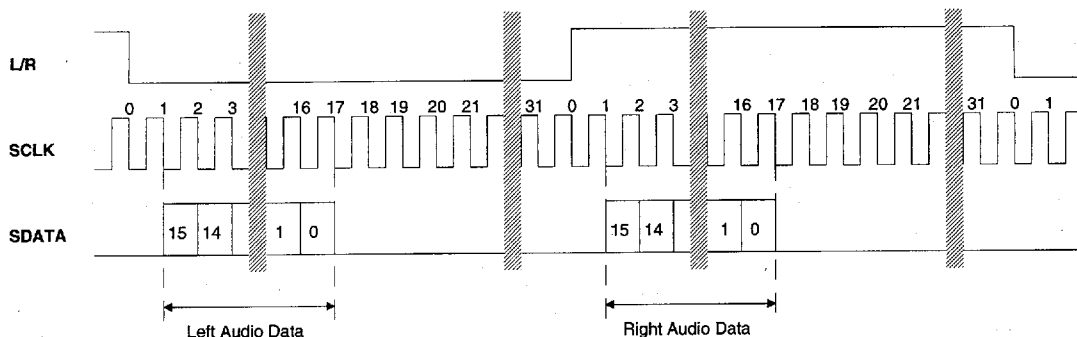


Figure 2. Data Output Timing - Mode 1 (FSYNC High or Low)

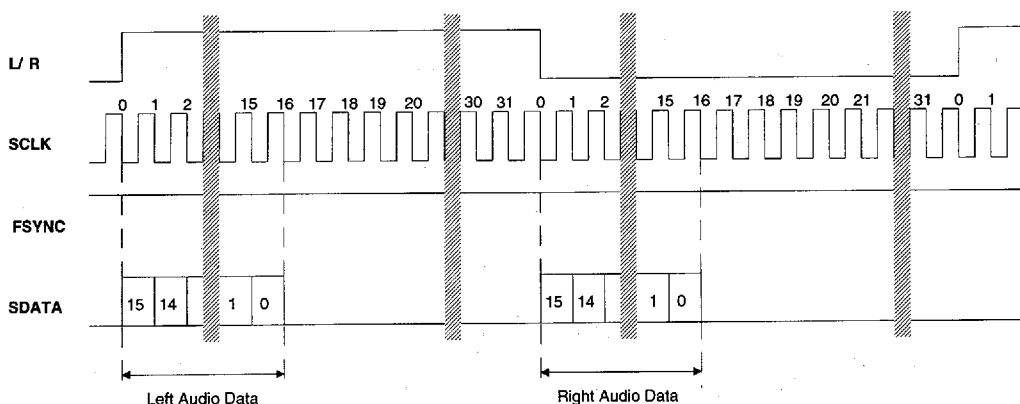


Figure 3. Data Output Timing - Mode 2 - FSYNC High

Mode 2-FSYNC High (SMODE low) is shown in Figure 3. The falling edge of SCLK causes the ADC to output each data bit with the exception of the MSB which is clocked out by the L/R edge.

In Mode 2-FSYNC Controlled, as shown in Figure 4, only the MSB is clocked out after the L/R edge with FSYNC low, SCLK is ignored. When it is desired to output data, bringing FSYNC high will enable SCLK to clock out data. This feature is particularly useful to position the data bits in time onto a common serial bus.

Analog Connections

Figure 1 shows the analog input connections. The analog inputs are presented differentially to the modulators via the AINR+, AINR- and AINL+, AINL- pins. Each analog input will accept a maximum of 1.88 Vpp centered at +2.5 V. The + and - input signals are 180° out of phase resulting in a differential input voltage of 3.75 Vpp. Figure 5 shows the input signal levels for full scale.

The CS5345 samples the analog inputs at 3.072 MHz for a 12.288 MHz MCLK (CMODE low). The digital filter rejects all noise between

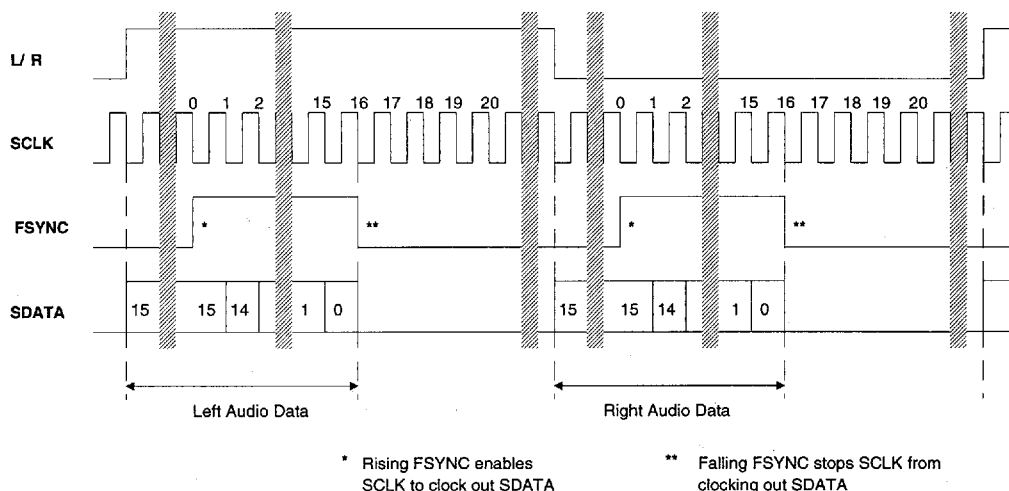


Figure 4. Data Output Timing- Mode 2 - FSYNC Controlled

32 kHz and (3.072 MHz-32 kHz). However, the filter will not reject frequencies right around 3.072 MHz (and multiples of 3.072 MHz). Most audio signals do not have significant energy at 3.072 MHz. Nevertheless, a 150 Ω resistor in series with each analog input and a 1500 pF capacitor across the inputs will attenuate any noise energy at 3.072 MHz, in addition to providing the optimum source impedance for the modulators. The use of capacitors which have a large voltage coefficient (such as general purpose ceramics) must be avoided since these will degrade signal linearity. NPO, COG and polyester film capacitors are acceptable. If active circuitry precedes the ADC, it is recommended that the above RC filter is placed between the active circuitry and the AINR and AINL pins. The above example frequencies scale linearly with output word rate.

Figure 6 is a suggested active input buffer circuit which provides a differential signal and level shifts up to +2.5 V. This circuit has been implemented on the CDB5345 evaluation board which is available from Crystal Semiconductor.

The on-chip voltage reference ((VA+) - 3.0 V) is connected to VREF (pin 24). A 4.7 μ F electrolytic capacitor in parallel with a 0.1 μ F ceramic capacitor attached between VREF and VA+ eliminates the effects of high frequency noise. Notice that VREF is decoupled to VA+ not AGND. This requirement is a result of the modulator sampling between VREF and VA+. No load current may be taken from the VREF output pin.

An additional on-chip voltage reference {(VA+) - 2.5 V} is connected to VCOM (pin 25). This output may be used to bias the analog input circuitry if a high impedance, low-bias current buffer is used.

Power-Down and Offset Calibration

The ADC has a power-down mode wherein typical consumption drops to 0.5 mW. PD is the power down pin for the device. When high, it places the analog and digital circuitry in the power-down mode. Bringing this pin low will release the power-down mode and initiate a calibration sequence.

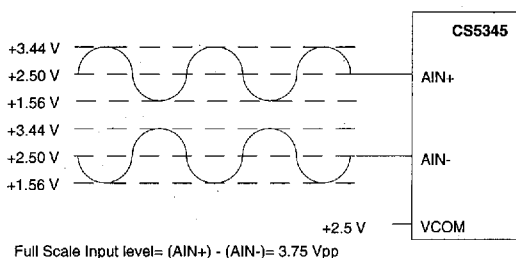


Figure 5. Full Scale Input Voltage

The delta-sigma modulator +/- inputs are internally disconnected from the AIN pins and shorted together during calibration. The digital section of the device measures and stores a value corresponding to the DC offset of each channel in the calibration registers. This calibration value is then subtracted from all future conversions during normal operation. 8224 L/R cycles are required for a calibration sequence. A short delay of approximately 18 output words will occur following calibration for the digital filter to begin accurately tracking audio band signals. During

calibration, the digital output of both channels is forced to a 2's complement zero.

Power-up Considerations

Upon initial application of power to the supply pins, the data in the calibration registers will be indeterminate. A calibration cycle should always be initiated after application of power to replace potentially large values of data in these registers with the correct values.

The modulators settle very quickly (a matter of microseconds) after the analog section is powered, either through the application of power or by exiting the power-down mode. The voltage reference takes a longer time to reach a final value due to the presence of large external capacitance on the VREF pin. The calibration period is optimized to allow the reference to settle for capacitor values of up to 4.7 μ F. The use of larger capacitors can cause erroneous calibration or make the device inoperable and is not recommended.

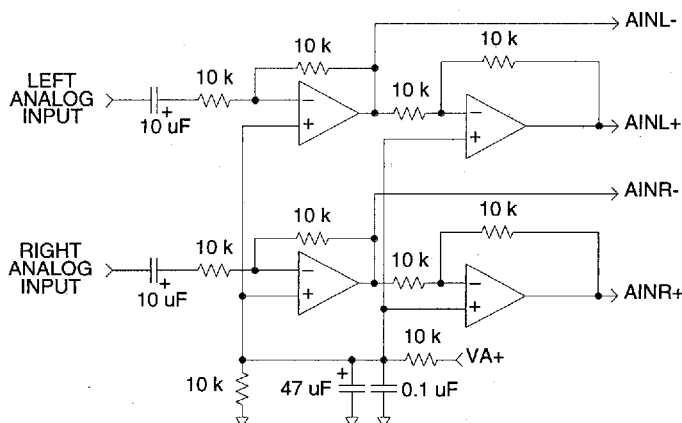


Figure 6. Example Input Buffer Circuit

Grounding and Power Supply Decoupling

As with any high resolution converter, the ADC requires careful attention to power supply and grounding arrangements if its potential performance is to be realized. Figure 1 shows the recommended power arrangements with VA+ connected to a clean +5 V supply. VS+ and VD+ should be powered from VA+ via a ferrite bead to minimize noise coupling. To further minimize noise coupling into the ADC, no additional devices should be powered from VD+. Analog ground and digital ground should be connected together near to where the supplies are brought onto the printed circuit board. Decoupling capacitors should be as near to the ADC as possible, with the low value ceramic capacitor being the nearest.

The printed circuit board layout should have separate analog and digital regions and ground planes, with the ADC straddling the boundary. All signals, especially clocks, should be kept away from the VREF pin in order to avoid unwanted coupling into the modulators. The Crystal Semiconductor application note "Layout and Design Rules for Data Converters" is available Crystal Semiconductor data books and should be considered required reading. An evaluation board is available which demonstrates the optimum layout and power supply arrangements,

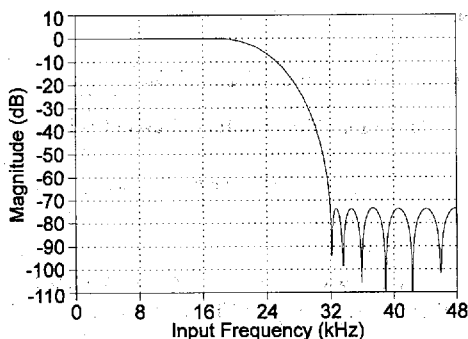


Figure 7. CS5345 Digital Filter Stopband Rejection

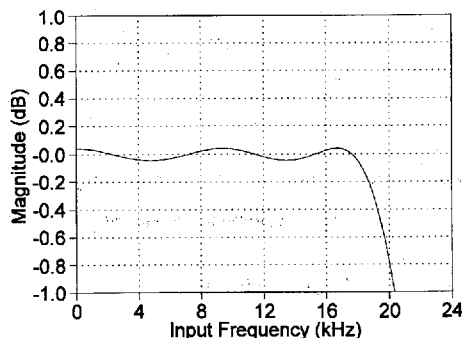


Figure 8. CS5345 Digital Filter Passband Ripple

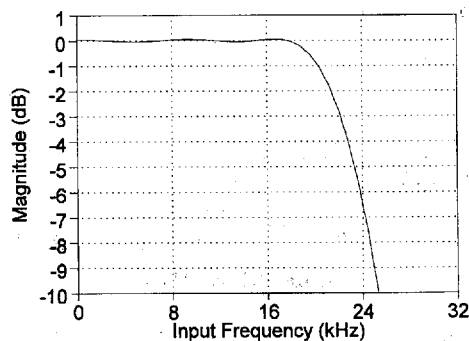


Figure 9. CS5345 Digital Filter Transition Band

as well as allowing fast evaluation of the CS5345.

To minimize digital noise, connect the ADC digital outputs only to CMOS inputs.

Synchronization of Multiple CS5345

In systems where multiple ADC's are required, care must be taken to insure that the ADC internal clocks are synchronized between converters to insure simultaneous sampling.

Synchronous sampling is achieved by connecting all PD pins to a single control signal and supplying the same L/R and MCLK to all converters.

PERFORMANCE

Digital Filter

Figures 7, 8, and 9 show the performance of the digital filter included in the ADC. All the plots assume an output word rate of 48 kHz. The filter frequency response will scale precisely with changes in output word rate. The passband ripple

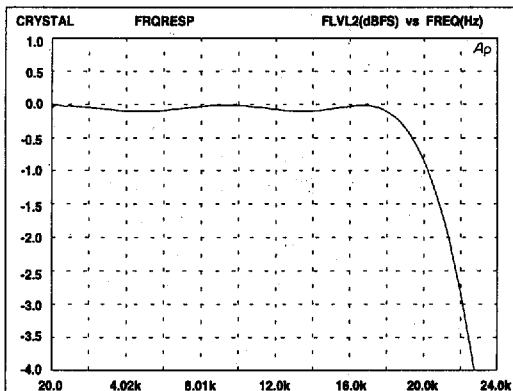


Figure 10. Frequency Response

is flat to ± 0.05 dB maximum.

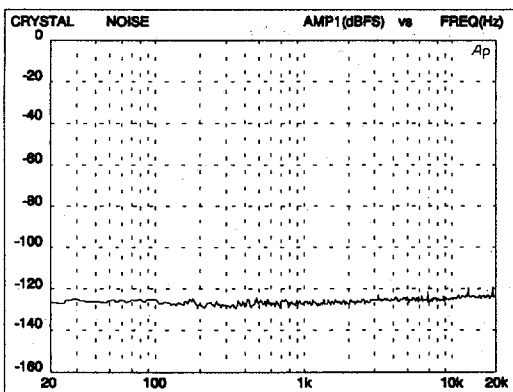


Figure 11. Noise Floor

Performance Measurements

All the following performance measurements were taken using an Audio Precision System

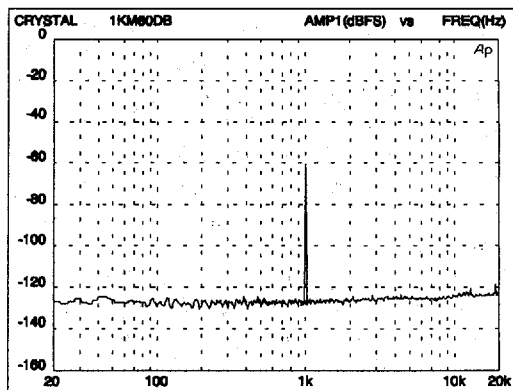


Figure 12. 1 kHz -60 dB

One Dual Domain tester. The CS5345 was in a CDB5345 evaluation board, running at 48 kHz word rate and interfaced to the System One via the digital audio interface using a CS8402 transmitter.

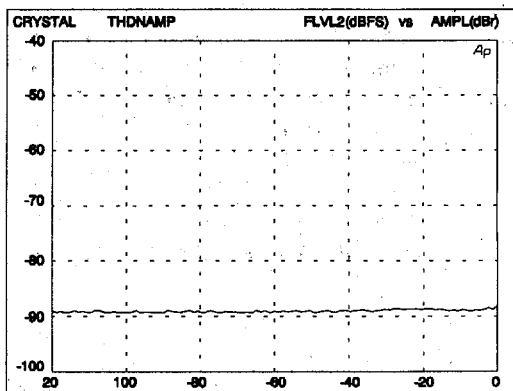


Figure 13. THD+N vs Input Level at 1kHz

Figure 10 shows the CS5345 frequency response.

Figure 11 shows the noise floor with zero input signal level. A 16 K point FFT was used.

Figure 12 shows a 1 kHz, -60 dB input signal FFT plot. Notice the lack of harmonic distortion

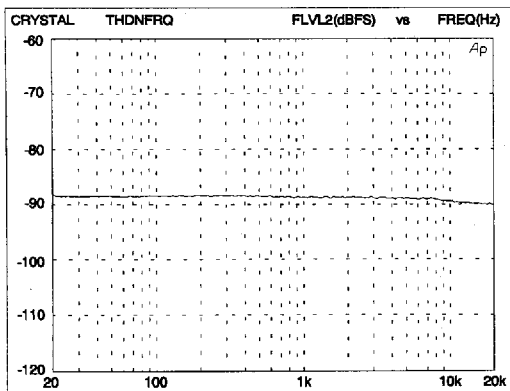


Figure 14. THD+N vs Frequency at -1dB

components. This is a direct result of the perfect differential non-linearity of the delta-sigma architecture

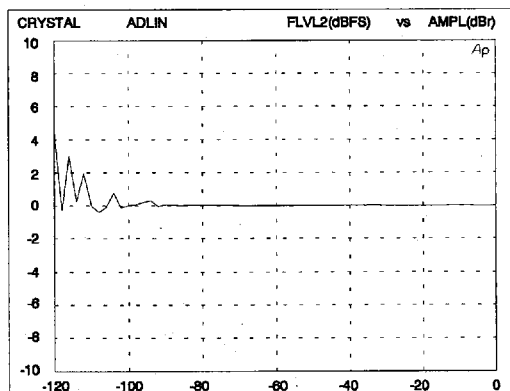


Figure 15. Output Level Error vs Input Level

Figure 13 shows the THD+N versus input level at 1 kHz. This plot indicates an unweighted dynamic range of 88 dB.

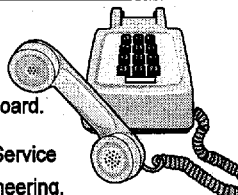
Figure 14 shows THD+N versus frequency at -1 dB input.

Figure 15 shows the linearity of the CS5345. The input signal is at 500 Hz and is varied from 0 dB (full scale) to -120 dB. At each input level, the output level is measured and compared to the perfect value. Any deviation is plotted as a deviation away from 0 dB. Notice the close conformance to perfect linearity, until the noise starts to influence the readings at about -100 dB.

3

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PIN DESCRIPTIONS

NO CONNECT	NC	1	28	NC	NO CONNECT
NO CONNECT	NC	2	27	NC	NO CONNECT
NO CONNECT	NC	3	26	NC	NO CONNECT
+ RIGHT CHANNEL ANALOG INPUT	AINR+	4	25	VCOM	VOLTAGE COMMON OUTPUT
- RIGHT CHANNEL ANALOG INPUT	AINR-	5	24	VREF	VOLTAGE REFERENCE OUTPUT
+ LEFT CHANNEL ANALOG INPUT	AINL+	6	23	VA+	POSITIVE ANALOG POWER
- LEFT CHANNEL ANALOG INPUT	AINL-	7	22	AGND	ANALOG GROUND
POWER DOWN INPUT	PD	8	21	TSTO2	TEST OUTPUT
SELECT CLOCK MODE	CMODE	9	20	TSTO1	TEST OUTPUT
MASTER CLOCK INPUT	MCLK	10	19	SMODE	SELECT SERIAL I/O MODE
LEFT/RIGHT SELECT	L/R	11	18	FSYNC	FRAME SYNC SIGNAL
SERIAL DATA CLOCK	SCLK	12	17	TST	TEST
SERIAL DATA OUTPUT	SDATA	13	16	DGND	DIGITAL GROUND
SILICON SUBSTRATE BIAS	VS+	14	15	VD+	POSITIVE DIGITAL POWER

Power Supply Connections

VA+ - Analog Power, PIN 23

Positive supply for the analog section. Nominally +5 volts.

VS+ - Digital Power, PIN 14

Positive supply for the silicon substrate. Nominally +5 volts.

VD+ - Digital Power, PIN 15

Positive supply for the digital section. Nominally +5 volts.

AGND - Analog Ground, PIN 22

Analog ground reference.

DGND - Digital Ground, PIN 16

Digital ground for the digital section.

Analog Inputs

±AINL, ±AINR - Differential Left and Right Channel Analog Inputs, PINS 6, 7, 4, 5

Analog input connections for the left and right input channels. A nominal differential input voltage of 3.75 V_{pp} will produce a full scale digital output.

Analog Outputs

VREF - Voltage Reference Output, Pin 24

Internal voltage reference output. Nominally (VA+) - 3.0 V. Must be bypassed to VA+ with a 0.1 µF ceramic capacitor in parallel with a 4.7 µF electrolytic capacitor.

VCOM - Voltage Common Output, PIN 25

Nominally (VA+) - 2.5 volts. May be used to bias the analog input circuitry if an additional buffer is used.

Digital Inputs**MCLK - Master Input Clock, PIN 10**

Sampling rates, output rates and digital filter characteristics scale to MCLK frequency. MCLK frequency is either 256 or 384 X the output word rate (see CMODE). For example, a 12.288 MHz MCLK corresponds to an output word rate of 48 kHz per channel with CMODE low.

SCLK - Serial Data Clock, PIN 12

SCLK is an input clock at any frequency from 32 X to 128 X the output word rate (64 X is preferred). Data is clocked out on the falling edge of SCLK. See the descriptions of Data Output Mode 1 and Mode 2.

L/R - Left/Right Select, PIN 11

L/R is an input which selects the left or right channel for output on SDATA. The L/R frequency must be at the output word rate. Although the outputs of each channel are transmitted at different times, the two words in an L/R cycle represent simultaneously sampled analog inputs.

Left channel data is on SDATA when L/R is low in Mode 1 (SMODE high). Right channel data is on SDATA when L/R is high. The MSB data bit appears on SDATA one SCLK cycle after L/R changes.

Left channel data is on SDATA when L/R is high in Mode 2 (SMODE low). Right channel data is on SDATA when L/R is low. The rising edge of L/R clocks out the MSB of the left channel data. The falling edge of L/R clocks out the MSB of the right channel data.

PD - Analog Power Down, PIN 8

Device power-down command. The analog and digital circuitry are in power-down mode when PD is logic high.

CMODE - Clock Mode Select, PIN 9

CMODE should be tied low to select an MCLK frequency of 256 X the output word rate. CMODE should be tied high to select an MCLK frequency of 384 X the output word rate.

SMODE - Serial Interface Mode Select, PIN 19

SMODE must be tied high to select serial interface Mode 1. SMODE must be tied low to select serial interface Mode 2. In all interface modes, L/R, FSYNC and SCLK should be derived from MCLK using external dividers.

FSYNC - Frame Synchronization Signal, PIN 18

In Mode 1 (SMODE high), FSYNC is ignored but must be connected to VD+ or DGND.

In Mode 2 (SMODE low), FSYNC is an input which controls the output of data on SDATA. FSYNC is normally tied high, which causes the data bits to be clocked out immediately following L/R transitions. If it is desired to delay the data bits from the L/R edge, then FSYNC must be low during the delay period. Bringing FSYNC high will enable SCLK to clock out of the SDATA bits. Note that the MSB will be clocked out based on the L/R edge, independent of the state of FSYNC.

Digital Outputs**SDATA - Serial Data Output, PIN 13**

Audio data bits are presented MSB first, in 2's complement format.

Miscellaneous**NC - No Connection, PINS 1, 2, 3, 26, 27, 28**

No internal connection.

TSTO1, TSTO2 - Test Pins, PIN 20, 21

These pins are factory test outputs and must not be connected to any external component or length of PC trace.

TST - Test Input, PIN 17

Allows access to the CS5345 test modes. Must be connected to digital ground for normal operation.

PARAMETER DEFINITIONS

Resolution - The total number of possible output codes is equal to 2^N , where N = the number of bits in the output word for each channel.

Dynamic Range - Full scale (RMS) signal to broadband noise ratio. The broadband noise is measured over the specified bandwidth, and with an input signal 60dB below full-scale. Units in decibels.

Total Harmonic Distortion+Noise - The ratio of the rms sum of all spectral components over the specified bandwidth (typically 10 Hz to 20 kHz), excluding signal, to the rms value of the signal.

Total Harmonic Distortion - The ratio of the rms sum of all harmonics up to 20 kHz to the rms value of the signal.

Interchannel Phase Deviation - The difference between the left and right channel sampling times.

Interchannel Isolation - A measure of crosstalk between the left and right channels. Measured for each channel at the converter's output with the input under test grounded and a full-scale signal applied to the other channel. Units in decibels.

Interchannel Gain Mismatch - The gain difference between left and right channels. Units in decibels.

Gain Error - The deviation of the measured full scale amplitude from the ideal full scale amplitude value.

Gain Drift - The change in gain value with temperature. Units in ppm/°C.

Bipolar Offset Error - The deviation of the mid-scale transition (111...111 to 000...000) from the ideal. Units in LSBs.

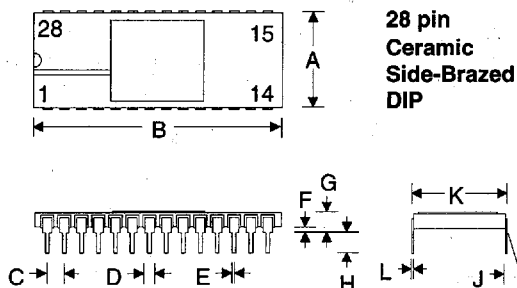
REFERENCES - All reprinted in this data book.

1) "A Stereo 16-bit Delta-Sigma A/D Converter for Digital Audio" by D.R. Welland, B.P. Del Signore, E.J. Swanson, T. Tanaka, K. Hamashita, S. Hara, K. Takasuka. Paper presented at the 85th Convention of the Audio Engineering Society, November 1988.

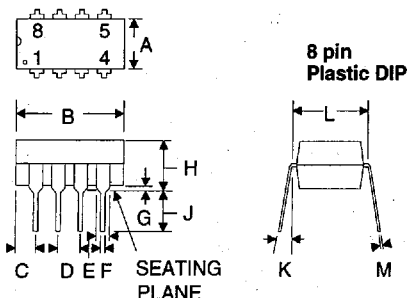
2) "The Effects of Sampling Clock Jitter on Nyquist Sampling Analog-to-Digital Converters, and on Oversampling Delta Sigma ADC's" by Steven Harris. Paper presented at the 87th Convention of the Audio Engineering Society, October 1989.

3) "An 18-Bit Dual-Channel Oversampling Delta-Sigma A/D Converter, with 19-Bit Mono Application Example" by Clif Sanchez. Paper presented at the 87th Convention of the Audio Engineering Society, October 1989.

MECHANICAL DATA



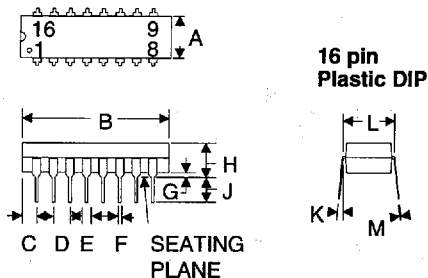
DIM	MILLIMETERS		INCHES	
	MIN	MAX	MIN	MAX
A	14.73	15.34	0.580	0.604
B	35.20	35.92	1.386	1.414
C	2.54 BSC		0.100 BSC	
D	0.76	1.40	0.030	0.055
E	0.38	0.53	0.015	0.021
F	1.02	1.52	0.040	0.060
G	2.79	4.32	0.110	0.170
H	2.54	4.57	0.100	0.180
J	-	10°	-	10°
K	14.99	15.49	0.590	0.610
L	0.20	0.30	0.008	0.012



DIM	MILLIMETERS		INCHES	
	MIN	MAX	MIN	MAX
A	6.10	6.60	0.240	0.260
B	9.14	10.2	0.360	0.400
C	0.38	1.52	0.015	0.060
D	2.54 BSC		0.100 BSC	
E	1.02	1.78	0.040	0.070
F	0.38	0.53	0.015	0.021
G	0.51	1.02	0.020	0.040
H	3.81	5.08	0.150	0.200
J	2.92	3.43	0.115	0.135
K	0°	10°	0°	10°
L	7.62BSC		0.300BSC	
M	0.20	0.38	0.008	0.015

NOTES:

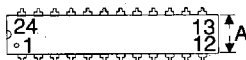
1. POSITIONAL TOLERANCE OF LEADS SHALL BE WITHIN 0.13MM (0.005") AT MAXIMUM MATERIAL CONDITION, IN RELATION TO SEATING PLANE AND EACH OTHER.
2. DIMENSION L TO CENTER OF LEADS WHEN FORMED PARALLEL.
3. DIMENSION A DOES NOT INCLUDE MOLD FLASH.



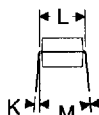
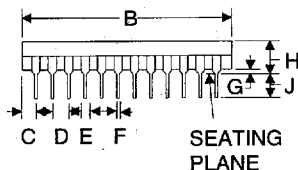
DIM	MILLIMETERS		INCHES	
	MIN	MAX	MIN	MAX
A	6.10	6.60	0.240	0.260
B	18.80	19.30	0.740	0.760
C	1.32	2.89	0.015	0.035
D	2.54 BSC		0.100 BSC	
E	1.02	1.78	0.040	0.070
F	0.38	0.53	0.015	0.021
G	0.51	1.02	0.020	0.040
H	3.81	5.08	0.150	0.200
J	2.92	3.43	0.115	0.135
K	0°	10°	0°	10°
L	7.62BSC		0.300BSC	
M	0.20	0.38	0.008	0.015

NOTES:

1. POSITIONAL TOLERANCE OF LEADS SHALL BE WITHIN 0.13MM (0.005") AT MAXIMUM MATERIAL CONDITION, IN RELATION TO SEATING PLANE AND EACH OTHER.
2. DIMENSION L TO CENTER OF LEADS WHEN FORMED PARALLEL.
3. DIMENSION A DOES NOT INCLUDE MOLD FLASH.



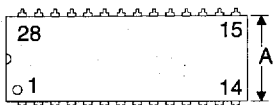
24 pin
Plastic
Skinny DIP



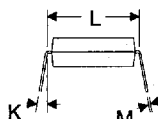
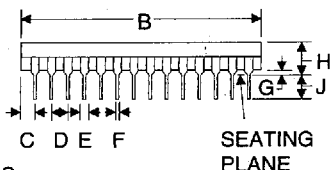
NOTES:

1. POSITIONAL TOLERANCE OF LEADS SHALL BE WITHIN 0.25MM (0.010") AT MAXIMUM MATERIAL CONDITION, IN RELATION TO SEATING PLANE AND EACH OTHER.
2. DIMENSION L TO CENTER OF LEADS WHEN FORMED PARALLEL.
3. DIMENSION A DOES NOT INCLUDE MOLD FLASH.

DIM	MILLIMETERS		INCHES	
	MIN	MAX	MIN	MAX
A	6.10	6.60	0.240	0.260
B	31.37	32.13	1.235	1.265
C	1.65	2.16	0.065	0.085
D	2.54 BSC		0.100 BSC	
E	1.02	1.52	0.040	0.060
F	0.36	0.56	0.014	0.022
G	0.51	1.02	0.020	0.040
H	3.94	4.57	0.155	0.180
J	2.92	3.43	0.115	0.135
K	0°	15°	0°	15°
L	7.62 BSC		0.300 BSC	
M	0.20	0.38	0.008	0.015



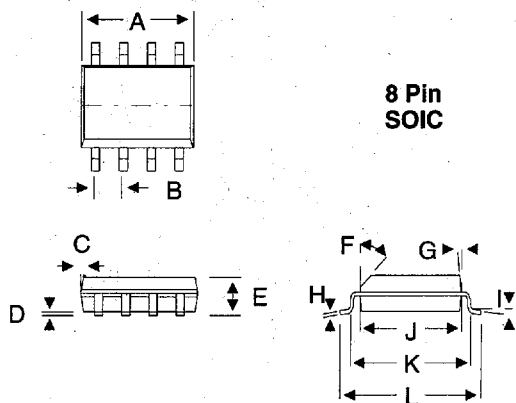
28 pin
Plastic DIP



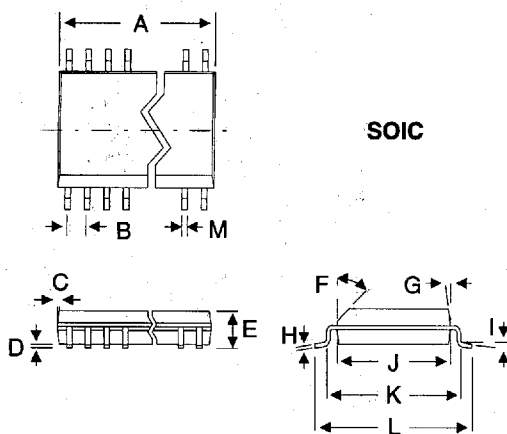
NOTES:

1. POSITIONAL TOLERANCE OF LEADS SHALL BE WITHIN 0.25MM (0.010") AT MAXIMUM MATERIAL CONDITION, IN RELATION TO SEATING PLANE AND EACH OTHER.
2. DIMENSION L TO CENTER OF LEADS WHEN FORMED PARALLEL.
3. DIMENSION A DOES NOT INCLUDE MOLD FLASH.

DIM	MILLIMETERS		INCHES	
	MIN	MAX	MIN	MAX
A	13.72	14.22	0.540	0.560
B	36.45	37.21	1.435	1.465
C	1.65	2.16	0.065	0.085
D	2.54 BSC		0.100 BSC	
E	1.02	1.52	0.040	0.060
F	0.36	0.56	0.014	0.022
G	0.51	1.02	0.020	0.040
H	3.94	5.08	0.155	0.200
J	2.92	3.43	0.115	0.135
K	0°	15°	0°	15°
L	15.24 BSC		0.600 BSC	
M	0.20	0.38	0.008	0.015



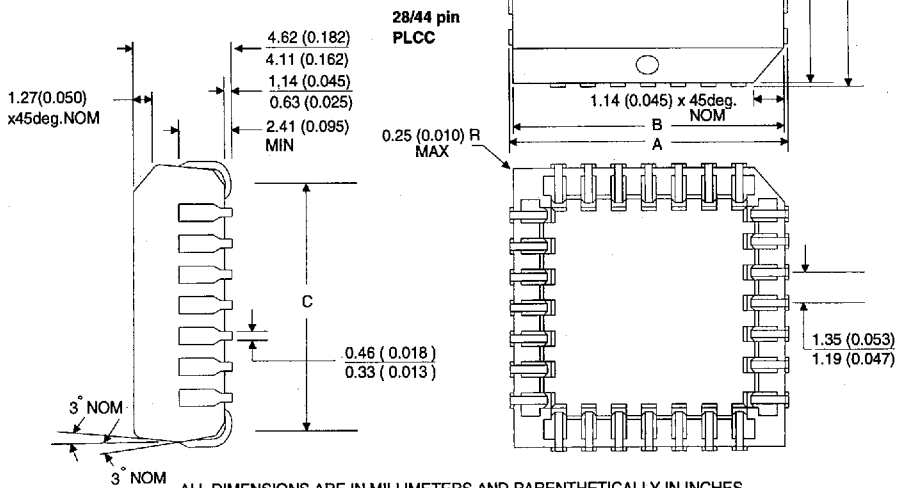
DIM	MILLIMETERS		INCHES	
	MIN	MAX	MIN	MAX
A	5.25	5.30	0.207	0.209
B	1.27 TYP		0.050 TYP	
C	7° NOM		7° NOM	
D	0.120	0.180	0.005	0.007
E	1.80	1.86	0.071	0.073
F	45° NOM		45° NOM	
G	7° NOM		7° NOM	
H	0.195	0.205	0.0078	0.0082
I	2°	4°	2°	4°
J	-	-	-	-
K	6.57	6.63	0.259	0.261
L	7.85	7.95	0.308	0.312



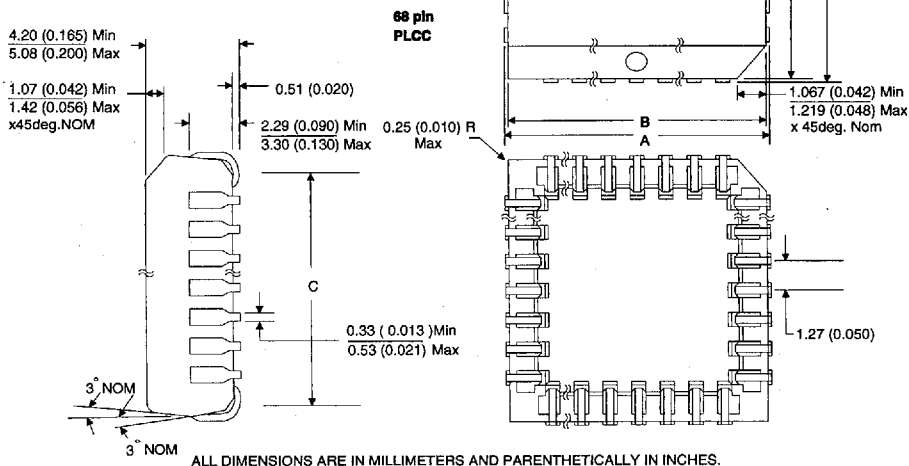
pins	MILLIMETERS		INCHES	
	MIN	MAX	MIN	MAX
16	9.91	10.41	0.390	0.410
20	12.45	12.95	0.490	0.510
24	14.99	15.50	0.590	0.610
28	17.53	18.03	0.690	0.710

DIM	MILLIMETERS		INCHES	
	MIN	MAX	MIN	MAX
A	see table above			
B	1.27	BSC	0.050	BSC
C	7° NOM		7° NOM	
D	0.127	0.330	0.005	0.013
E	2.41	2.67	0.095	0.105
F	45° NOM		45° NOM	
G	7° NOM		7° NOM	
H	0.203	0.381	0.008	0.015
I	2°	8°	2°	8°
J	7.42	7.59	0.292	0.298
K	8.76	9.02	0.345	0.355
L	10.16	10.67	0.400	0.420
M	0.33	0.51	0.013	0.020

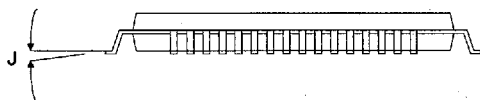
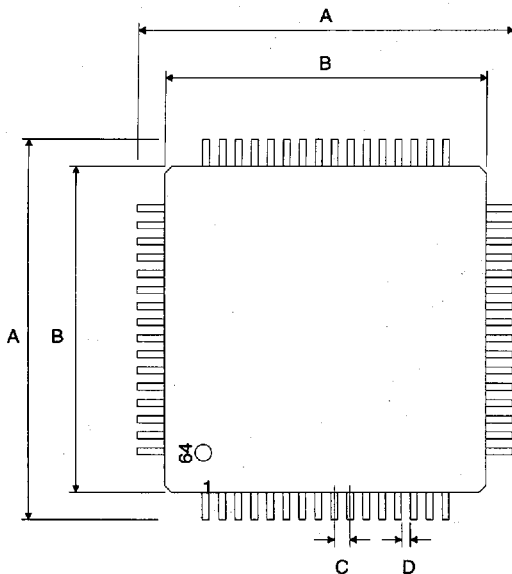
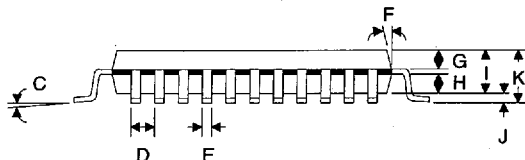
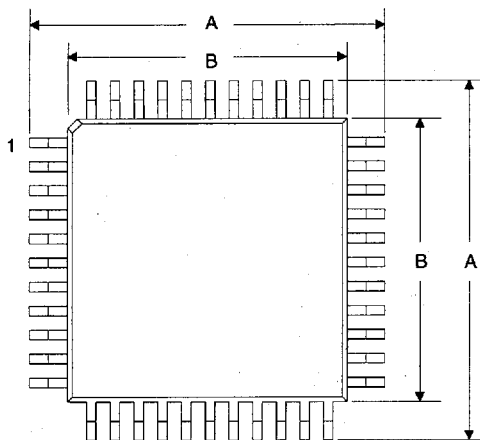
NO. OF TERMINAL	A		B		C	
	MIN	MAX	MIN	MAX	MIN	MAX
28	12.32 (0.485)	12.57 (0.495)	11.43 (0.450)	11.58 (0.456)	9.91 (0.390)	10.92 (0.430)
44	17.40 (0.685)	17.65 (0.695)	16.51 (0.650)	16.66 (0.656)	14.98 (0.590)	16.00 (0.630)



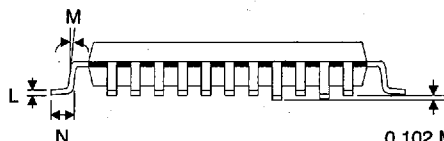
	A		B		C	
	MIN	MAX	MIN	MAX	MIN	MAX
68	25.02 (0.985)	25.27 (0.995)	24.13 (0.950)	24.33 (0.958)	22.61 (0.890)	23.62 (0.930)



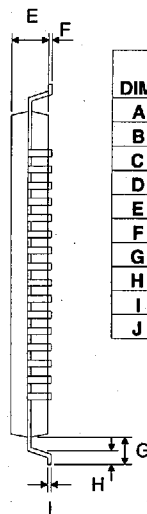
44 PIN QUAD FLATPACK



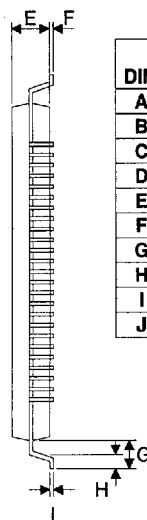
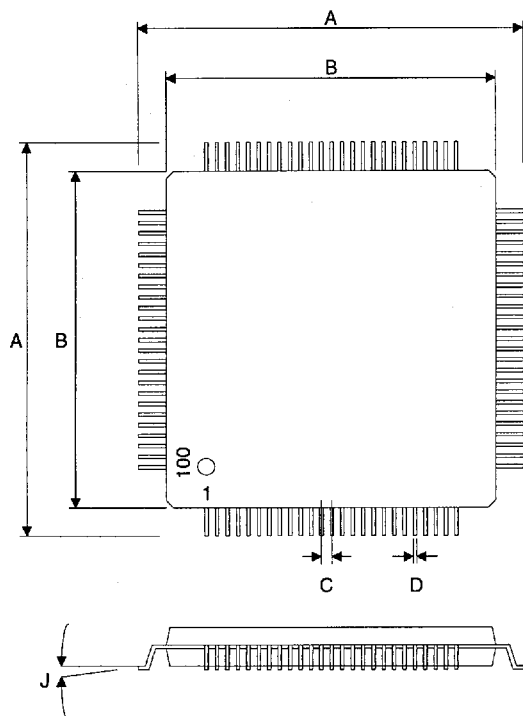
44 Pin TQFP				
1.4 mm Package Thickness				
DIM	MILLIMETERS		INCHES	
	MIN	MAX	MIN	MAX
A	11.75	12.25	0.463	0.482
B	9.90	10.10	0.390	0.398
C	0°	7°	0°	7°
D	0.80 BSC		0.031 BSC	
E	0.35 BSC		0.014 BSC	
F		120		120
G	0.54	0.74	0.021	0.029
H	0.54	0.74	0.021	0.029
I	1.35	1.50	0.053	0.059
J	0.05		0.002	
K		1.60		0.063
L		0.17		0.007
M	2°	10°	2°	10°
N	0.35	0.65	0.014	0.026



0.102 MAX
Lead Coplanarity



64 Pin TQFP				
DIM	MILLIMETERS		INCHES	
	MIN	MAX	MIN	MAX
A	12.00 BSC		0.472 BSC	
B	10.00 BSC		0.393 BSC	
C	0.50 BSC		0.020 BSC	
D	0.14	0.30	0.005	0.012
E	0.95	1.12	0.037	0.044
F	0.05	0.15	0.002	0.006
G	1.00 BSC		0.039 BSC	
H	0.45	0.75	0.018	0.030
I	0.09	0.18	0.003	0.007
J	0°	7°	0°	7°



100-pin TQFP

DIM	MILLIMETERS		INCHES	
	MIN	MAX	MIN	MAX
A	15.75	16.25	0.620	0.640
B	13.90	14.10	0.547	0.555
C	0.50 BSC		0.020 BSC	
D	0.10	0.20	0.004	0.012
E	1.25	1.55	0.049	0.061
F	0.00	0.20	0.000	0.008
G	1.00 BSC		0.039 BSC	
H	0.35	0.65	0.014	0.026
I	0.077	0.177	0.003	0.007
J	0°	10°	0°	10°