

# LF441A/LF441 Low Power JFET Input Operational Amplifier



T-79-15

#### **General Description**

The LF441A/441 low power operational amplifier provides many of the same AC characteristics as the industry standard LM741 while greatly improving the DC characteristics of the LM741. The amplifier has the same bandwidth, slew rate, and gain (10 kΩ load) as the LM741 and only draws one tenth the supply current of the LM741. In addition, the well matched high voltage JFET input devices of the LF441A/441 reduce the input bias and offset currents by a factor of 10,000 over the LM741. A combination of careful layout design and internal trimming guarantees very low input offset voltage and voltage drift. The LF441A/441 also has a very low equivalent input noise voltage for a low power amplifier.

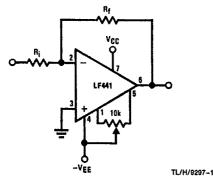
The LF441A/441 is pin compatible with the LM741, allowing an immediate 10 times reduction in power drain in many applications. The LF441A/441 should be used where low

power dissipation and good electrical characteristics are the major considerations.

#### **Features**

■ 1/10 supply current of a LM741	200 μA (max)
■ Low input bias current	50 pA (max)
■ Low input offset voltage	0.5 mV (max)
Low input offset voltage drift	10 μV/°C (max)
High gain bandwidth	1 MHz
■ High slew rate	1 V/μs
Low noise voltage for low power	35 nV/√Hz
■ Low input noise current	0.01 pA/1/Hz
■ High input impedance	1012Ω
■ High gain $V_O = \pm 10V$ , $R_L = 10k$	50k (min)

#### **Typical Connection**



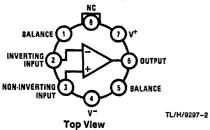
#### **Ordering Information**

#### LF441XYZ

- X indicates electrical grade
- Y Indicates temperature range "M" for military,
  - "C" for commercial
- Z indicates package type "H" or "N"

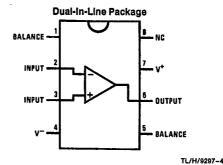
#### **Connection Diagrams**

Metal Can Package



Note: Pin 4 connected to case.

Order Number LF441AMH or LF441CH
See NS Package Number H08B



Top View Order Number LF441ACN, LF441CJ, LF441CM or LF441CN See NS Package Number J08A, M08A or N08E

A	bsol	lute	Maximum	Ratings
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If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/Distributors for availability and specifications.

Input Voltage Range (Note 1)

LF441A LF441 ±19V ±15V

Supply Voltage Differential input Voltage ±22V ±18V ±38V ±30V

165°C/W

65°C/W

LF441A

Output Short Circuit Duration

Continuous

Continuous

H Package **Power Dissipation** 670 mW (Notes 2 and 9) T<sub>j max</sub> 150°C

J Package N Package 670 mW 115°C 130°C/W

M Package

185°C/W

Board Mount in 400 LF/ min air flow  $\theta_{\rm IC}$ 

Board Mount in still air

 $\theta_{A}$ (Typical)

25°C/W Operating Temp. Range (Note 3) Storage Temp. Range  $-65^{\circ}\text{C} \le \text{T}_{\text{A}} \le 150^{\circ}\text{C}$ Lead Temperature 300°C (Soldering, 10 seconds)

(Note 3) 300°C

-65°C ≤ T<sub>A</sub> ≤ 150°C 260°C

 $\theta_{|C}$ 

Small Outline Package Vapor Phase (60 sec.)

Infrared (15 sec.)

25°C LF441A LF441 Soldering Information Dual-In-Line Package Soldering (10 sec.) 260°C 260°C

215°C

220°C

215°C

220°C

See AN-450 "Surface Mounting Methods and Their Effect on Product Reliability" for other methods of soldering surface mount devices.

ESD rating to be determined.

# DC Electrical Characteristics (Note 4)

Symbol	Parameter	er Conditions		LF441A			LF441			Units
				Min	Тур	Max	Min	Тур	Max	Units
Vos	input Offset Voltage	$R_S = 10  k\Omega, T_A$	= 25°C		0.3	0.5		1	5	mV
		Over Temperatur	е						7.5	mV
ΔV <sub>OS</sub> /ΔT	Average TC of Input Offset Voltage	R <sub>S</sub> = 10 kΩ (Note 5)			7	10		10		μV/°C
los	Input Offset Current	V <sub>S</sub> = ±15V	T <sub>j</sub> = 25°C		5	25		5	50	pА
		(Notes 4 and 6)	T <sub>i</sub> = 70°C			1.5			1.5	nA
			T <sub>j</sub> = 125°C			10				nA
i <sub>B</sub>	Input Blas Current	V <sub>S</sub> = ±15V (Notes 4 and 6)	T <sub>j</sub> = 25°C		10	50		10	100	pА
			T <sub>j</sub> = 70°C			3			3	nA
			T <sub>j</sub> = 125°C			20				nA
RIN	Input Resistance	T <sub>j</sub> = 25°C			1012			1012		Ω
Avol	Large Signal Voltage Gaín	$V_S = \pm 15V, V_O$ $R_L = 10 k\Omega, T_A$	= ±10V, = 25°C	50	100		25	100		V/mV
		Over Temperatur	е	25			15			V/mV
Vo	Output Voltage Swing	V <sub>S</sub> = ±15V, R <sub>L</sub>	= 10 kΩ	±12	±13		±12	±13		v
V <sub>CM</sub>	Input Common-Mode Voltage Range			±16	+ 18,	-17	±11	+ 14,	-12	٧
CMRR	Common-Mode Rejection Ratio	R <sub>S</sub> ≤ 10 kΩ	······································	80	100		70	95		dB

T	_	7	9	_	1	5
	_		_	_	_	u

DC Ele	ctrical Charac	cteristics (Note 4) (Contin	ued)						
Symbol	Parameter	Conditions	LF441A			LF441			Unita
			Min	Тур	Max	Min	Тур	Max	J.,,,to
PSRR	Supply Voltage Rejection Ratio	(Note 7)	80	100		70	90		dB
ls	Supply Current			150	200		150	250	μΑ

#### AC Electrical Characteristics (Note 4)

	symbol Parameter Conditions —	Conditions	LF441A			LF441			Units
Symbol Parameter		Min	Тур	Max	Min	Тур	Max		
SR	Slew Rate	$V_S = \pm 15V, T_A = 25^{\circ}C$	0.8	1		0.6	1		V/µs
GBW	Gain-Bandwidth Product	V <sub>S</sub> = ±15V, T <sub>A</sub> = 25°C	0.8	1		0.6	1		MHz
en	Equivalent Input Noise Voltage	$T_A = 25^{\circ}C$ , $R_S = 100\Omega$ , $f = 1 \text{ kHz}$		35			35		nV/√Hz
l <sub>n</sub>	Equivalent Input Noise Current	T <sub>A</sub> = 25°C, f = 1 kHz		0.01			0.01		pA/√Hz

Note 1: Unless otherwise specified the absolute maximum negative input voltage is equal to the negative power supply voltage.

Note 2: For operating at elevated temperature, these devices must be derated based on a thermal resistance of  $\theta_{jA}$ .

Note 3: The LF441A is available in both the commercial temperature range 0°C ≤ T<sub>A</sub> ≤ 70°C and the military temperature range —55°C ≤ T<sub>A</sub> ≤ 125°C. The LF441A/441 is available in the commercial temperature range only. The temperature range is designated by the position just before the package type in the device number. A "C" indicates the commercial temperature range and an "M" indicates the military temperature range. The military temperature range is available in "H" package only.

Note 4: Unless otherwise specified the specifications apply over the full temperature range and for V<sub>S</sub> = ±20V for the LF441A and for V<sub>S</sub> = ±15V for the LF441. V<sub>CS</sub>, I<sub>B</sub>, and I<sub>CS</sub> are measured at V<sub>CM</sub> = 0.

Note 5: The LF441A is 100% tested to this specification.

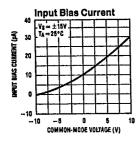
Note 6: The input bias currents are junction leakage currents which approximately double for every 10°C increase in the junction temperature,  $T_i$ . Due to limited production test time, the input bias currents measured are correlated to junction temperature. In normal operation the junction temperature rises above the ambient temperature as a result of internal power dissipation,  $P_D$ .  $T_j = T_A + \theta_{jA} P_D$  where  $\theta_{jA}$  is the thermal resistance from junction to ambient. Use of a heat sink is recommended if input bias current is to be kept to a minimum.

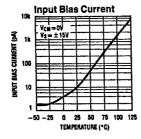
Note 7: Supply voltage rejection ratio is measured for both supply magnitudes increasing or decreasing simultaneously in accordance with common practice. From ±15V to ±5V for the LF441 and from ±20V to ±5V for the LF441A.

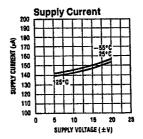
Note 8: Refer to RETS441AX for LF441AMH military specifications.

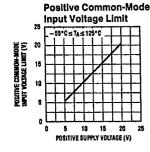
Note 9: Max. Power Dissipation is defined by the package characteristics. Operating the part near the Max. Power Dissipation may cause the part to operate outside guaranteed limits.

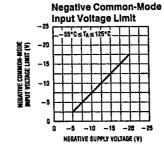
### **Typical Performance Characteristics**

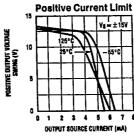




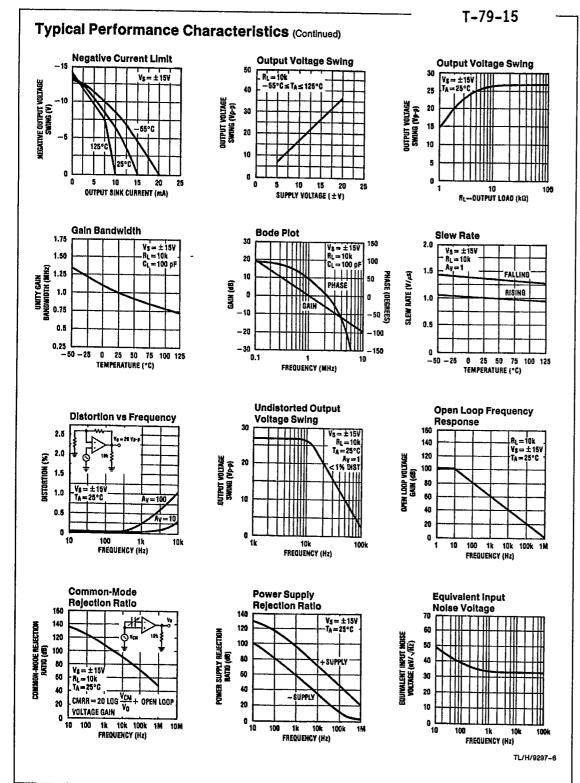






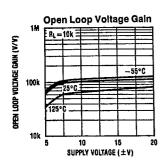


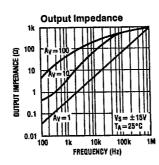
TL/H/9297-5

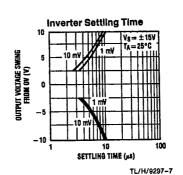


LF441A/LF441

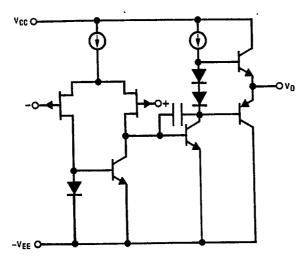
# Typical Performance Characteristics (Continued)







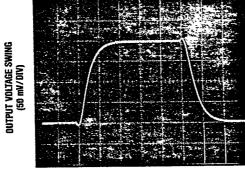
# **Simplified Schematic**



TL/H/9297-3

Pulse Response  $R_L = 10 \text{ k}\Omega$ ,  $C_L = 10 \text{ pF}$ 

#### **Small Signal Inverting**



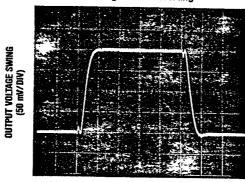
TIME (0.5 μs/DIV)

TL/H/9297-8

Pulse Response  $R_L=10~\text{k}\Omega,\,C_L=10~\text{pF}$  (Continued)

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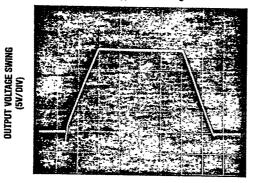
Small Signal Non-Inverting



TIME (0.5 µ\$/DIV)

TL/H/9297-9

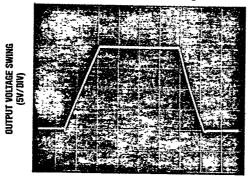
Large Signal Inverting



TIME (10 µs/DIV)

TL/H/9297~10

#### Large Signal Non-Inverting



TIME (10 µs/DIV)

TL/H/9297-11

3%

#### **Application Hints**

This device is a low power op amp with an internally trimmed input offset voltage and JFET input devices (BI-FET iI). These JFETs have large reverse breakdown voltages from gate to source and drain, eliminating the need for clamps across the inputs. Therefore, large differential input voltages can easily be accommodated without a large increase in input current. The maximum differential input voltage is independent of the supply voltages. However, neither of the input voltages should be allowed to exceed the negative supply as this will cause large currents to flow which can result in a destroyed unit.

Exceeding the negative common-mode limit on either input will force the output to a high state, potentially causing a reversal of phase to the output. Exceeding the negative common-mode limit on both inputs will force the amplifier output to a high state. In neither case does a latch occur since raising the input back within the common-mode range again puts the input stage and thus the amplifier in a normal operating mode.

Exceeding the positive common-mode limit on a single input will not change the phase of the output; however, if both inputs exceed the limit, the output of the amplifier will be forced to a high state.

The amplifier will operate with a common-mode input voltage equal to the positive supply; however, the gain bandwidth and slew rate may be decreased in this condition. When the negative common-mode voltage swings to within 3V of the negative supply, an increase in input offset voltage may occur.

The amplifier is blased to allow normal circuit operation with power supplies of  $\pm 3V$ . Supply voltages less than these may degrade the common-mode rejection and restrict the output voltage swing.

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The amplifier will drive a 10 k $\Omega$  load resistance to  $\pm 10V$  over the full temperature range.

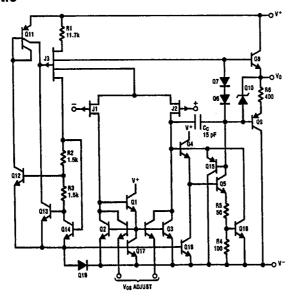
Precautions should be taken to ensure that the power supply for the integrated circuit never becomes reversed in polarity or that the unit is not inadvertently installed backwards in a socket, as an unlimited current surge through the resulting forward diode within the IC could cause fusing of the internal conductors and result in a destroyed unit.

Because this amplifier is a JFET rather than MOSFET input op amp it does not require special handling.

As with most amplifiers, care should be taken with lead dress, component placement and supply decoupling in order to ensure stability. For example, resistors from the output to an input should be placed with the body close to the input to minimize "pick-up" and maximize the frequency of the feedback pole by minimizing the capacitance from the input to ground.

A feedback pole is created when the feedback around any amplifier is resistive. The parallel resistance and capacitance from the input of the device (usually the inverting input to AC ground) set the frequency of this pole. In many instances the frequency of this pole is much greater than the expected 3 dB frequency, of the closed loop gain and consequently there is negligible effect on stability margin. However, if the feedback pole is less than approximately 6 times the expected 3 dB frequency, a lead capacitor should be placed from the output to the input of the op amp. The value of the added capacitor should be such that the RC time constant of this capacitor and the resistance it parallels is greater than or equal to the original feedback pole time constant.

#### **Detailed Schematic**



TL/H/9297-13