

Dual/Quad Over-The-Top Micropower Rail-to-Rail Input and Output Op Amps

FEATURES

- Low Input Offset Voltage: 500µV Max
- Output Swings to 10mV Max from V⁻
- Rail-to-Rail Input and Output
- Micropower: 50µA/Amplifier Max
- MSOP Package
- Over-The-TopTM Input Common Mode Range Extends 44V Above V⁻, Independent of V⁺
- Specified on 3V, 5V and ±15V Supplies
- High Output Current: 20mA
- Output Drives 10,000pF with Output Compensation
- Reverse Battery Protection to 18V
- No Supply Sequencing Problems
- High Voltage Gain: 1500V/mV
- High CMRR: 98dB
- No Phase Reversal
- Gain Bandwidth Product: 200kHz

APPLICATIONS

- Battery- or Solar-Powered Systems Portable Instrumentation Sensor Conditioning
- Supply Current Sensing
- Battery Monitoring
- Micropower Active Filters
- 4mA to 20mA Transmitters

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DESCRIPTION

The LT®1490A/LT1491A are enhanced versions of the popular LT1490/LT1491 dual and guad op amps with improved input offset voltage (500µV max) and output voltage swing (10mV max from V⁻). They are recommended as replacements for the LT1490/LT1491 in all new designs. The LT1490A/LT1491A operate on all single and split supplies with a total voltage of 2V to 44V, drawing only 40µA of quiescent current per amplifier. These amplifiers are reverse supply protected; they draw virtually no current for reverse supply up to 18V. The input range of the LT1490A/LT1491A includes both supplies and the output swings to both supplies. Unlike most micropower op amps, the LT1490A/ LT1491A can drive heavy loads; their rail-to-rail outputs drive 20mA. The LT1490A/LT1491A are unity-gain stable and drive all capacitive loads up to 10,000pF when optional $0.22\mu F$ and 150Ω compensation is used.

The LT1490A/LT1491A have a unique input stage that operates and remains high impedance when above the positive supply. The inputs take 44V both differential and common mode even when operating on a 3V supply. Built-in resistors protect the inputs for faults below the negative supply up to 15V. There is no phase reversal of the output for inputs 15V below V^- or 44V above V^- , independent of V^+ .

The LT1490A dual op amp is available in the 8-pin MSOP, PDIP and SO packages. The quad LT1491A is available in the 14-pin SO and PDIP packages.

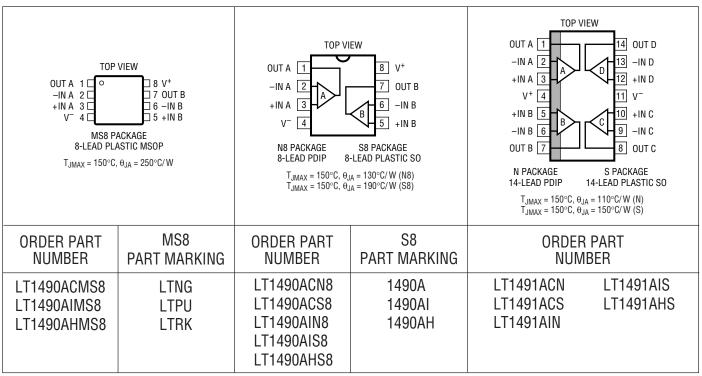
Battery Monitor TYPICAL APPLICATION Q1 2N3904 ₩. 1/4 LT1491 IBATT 1/4 LT1491A LOGIC Q2 2N3904 LOGIC HIGH (5V) = CHARGING LOGIC LOW (0V) = DISCHARGING 1/4 LT149 LOAD 1/4 LT1491 V_{OUT} V_{BATT} = 12V 90 9k V_{OUT} S1 = OPEN. GAIN = 1 $R_A = R_B$ $V_S = 5V$, 0V(R_S)(R_G/R_A)(GAIN) = CLOSED, GAIN = 10

ABSOLUTE MAXIMUM RATINGS (Note 1)

44V
44V
±12mA
Continuous
−40°C to 85°C
−40°C to 85°C
-40°C to 125°C

Specified Temperature Range (Note	e 4)
LT1490AC/LT1490AI	40°C to 85°C
LT1491AC/LT1491AI	40°C to 85°C
LT1490AH/LT1491AH	40°C to 125°C
Junction Temperature	150°C
Storage Temperature Range	65°C to 150°C
Lead Temperature (Soldering, 10 se	ec) 300°C

PACKAGE/ORDER INFORMATION



Consult factory for Military grade parts.

ELECTRICAL CHARACTERISTICS The • denotes specifications which apply over the full operating temperature range of $-40^{\circ}\text{C} \le \text{T}_{\text{A}} \le 85^{\circ}\text{C}$, otherwise specifications are at $\text{T}_{\text{A}} = 25^{\circ}\text{C}$. $\text{V}_{\text{S}} = 3\text{V}$, OV; $\text{V}_{\text{S}} = 5\text{V}$, OV; $\text{V}_{\text{CM}} = \text{V}_{\text{OUT}} = \text{half supply unless}$ otherwise noted. (Note 4)

SYMBOL	PARAMETER	CONDITIONS		90AC/LT14 490AI/LT14 TYP		UNITS
V _{OS}	Input Offset Voltage (Note 5)	LT1490A N, S Packages $0^{\circ}\text{C} \leq \text{T}_{\text{A}} \leq 70^{\circ}\text{C}$ $-40^{\circ}\text{C} \leq \text{T}_{\text{A}} \leq 85^{\circ}\text{C}$	•	110	500 700 800	μV μV μV
		LT1490A MS8 Package, LT1491A N, S Packages $0^{\circ}C \le T_A \le 70^{\circ}C$ $-40^{\circ}C \le T_A \le 85^{\circ}C$	•	220	1000 1200 1400	μV μV μV
	Input Offset Voltage Drift (Note 9)	$-40^{\circ}\text{C} \le \text{T}_{\text{A}} \le 85^{\circ}\text{C}$	•	2	4	μV/°C



ELECTRICAL CHARACTERISTICS The • denotes specifications which apply over the full operating temperature range of $-40^{\circ}\text{C} \le \text{T}_{\text{A}} \le 85^{\circ}\text{C}$, otherwise specifications are at $\text{T}_{\text{A}} = 25^{\circ}\text{C}$. $\text{V}_{\text{S}} = 3\text{V}$, OV; $\text{V}_{\text{S}} = 5\text{V}$, OV; $\text{V}_{\text{CM}} = \text{V}_{\text{OUT}} = \text{half supply unless otherwise noted.}$ (Note 4)

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SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS		
I _{OS}	Input Offset Current	V _{CM} = 44V (Note 6)			0.2	0.8 0.8	nA μA		
I _B	Input Bias Current	V _{CM} = 44V (Note 6) V _S = 0V	•		1 3 0.3	8 10	nA μA nA		
	Input Bias Current Drift	$-40^{\circ}\text{C} \le \text{T}_{\text{A}} \le 85^{\circ}\text{C}$	•		2		pA/°C		
	Input Noise Voltage	0.1Hz to 10Hz			1		μV _{P-P}		
e _n	Input Noise Voltage Density	f = 1kHz			50		nV/√Hz		
i _n	Input Noise Current Density	f = 1kHz			0.015		pA/√Hz		
R _{IN}	Input Resistance	Differential Common Mode, V _{CM} = 0V to 44V		6 4	17 11		MΩ MΩ		
C _{IN}	Input Capacitance				4.6		pF		
	Input Voltage Range		•	0		44	V		
CMRR	Common Mode Rejection Ratio (Note 6)	V _{CM} = 0V to V _{CC} - 1V V _{CM} = 0V to 44V	•	84 80	98 98		dB dB		
A _{VOL}	Large-Signal Voltage Gain	$V_S = 3V$, $V_0 = 500$ mV to 2.5V, $R_L = 10$ k $0^{\circ}C \le T_A \le 70^{\circ}C$ $-40^{\circ}C \le T_A \le 85^{\circ}C$	•	200 133 100	1500		V/mV V/mV V/mV		
		$V_S = 5V$, $V_0 = 500$ mV to 4.5V, $R_L = 10$ k $0^{\circ}C \le T_A \le 70^{\circ}C$ $-40^{\circ}C \le T_A \le 85^{\circ}C$	•	400 250 200	1500		V/mV V/mV V/mV		
V _{OL}	Output Voltage Swing Low	$V_S = 3V$, No Load $V_S = 3V$, $I_{SINK} = 5mA$	•		3 250	10 450	mV mV		
		$V_S = 5V$, No Load $V_S = 5V$, $I_{SINK} = 5mA$ $V_S = 5V$, $I_{SINK} = 10mA$	•		3 250 330	10 500 500	mV mV mV		
V _{OH}	Output Voltage Swing High	V _S = 3V, No Load V _S = 3V, I _{SOURCE} = 5mA	•	2.95 2.55	2.978 2.6		V		
		$V_S = 5V$, No Load $V_S = 5V$, $I_{SOURCE} = 10$ mA	•	4.95 4.30	4.978 4.6		V		
I _{SC}	Short-Circuit Current (Note 2)	$V_S = 3V$, Short to GND $V_S = 3V$, Short to V_{CC}		10 10	15 30		mA mA		
		V _S = 5V, Short to GND V _S = 5V, Short to V _{CC}		15 15	25 30		mA mA		
PSRR	Power Supply Rejection Ratio	$V_S = 2.5V \text{ to } 12.5V, V_{CM} = V_0 = 1V$	•	84	98		dB		
	Minimum Operating Supply Voltage		•		2	2.5	V		
	Reverse Supply Voltage	I _S = -100μA per Amplifier	•	18	27		V		
I _S	Supply Current per Amplifier (Note 7)		•		40	50 55	μA μA		
GBW	Gain Bandwidth Product (Note 6)	$ f = 1 \text{ Hz} $ $0^{\circ}\text{C} \leq \text{T}_{\text{A}} \leq 70^{\circ}\text{C} $ $-40^{\circ}\text{C} \leq \text{T}_{\text{A}} \leq 85^{\circ}\text{C} $	•	110 100 90	180		kHz kHz kHz		
SR	Slew Rate (Note 8)	$\begin{array}{l} A_V = -1, \ R_L = \infty \\ 0^{\circ}C \le T_A \le 70^{\circ}C \\ -40^{\circ}C \le T_A \le 85^{\circ}C \end{array}$	•	0.035 0.031 0.030	0.06		V/μs V/μs V/μs		



ELECTRICAL CHARACTERISTICS The ullet denotes specifications which apply over the full operating temperature range of $-40^{\circ}\text{C} \le \text{T}_{A} \le 85^{\circ}\text{C}$, otherwise specifications are at $\text{T}_{A} = 25^{\circ}\text{C}$. $\text{V}_{S} = \pm 15\text{V}$; $\text{V}_{CM} = 0\text{V}$, $\text{V}_{OUT} = 0\text{V}$ unless otherwise noted. (Note 4)

				LT1	190AC/LT149 490AI/LT149	1AI	
SYMBOL	PARAMETER	CONDITIONS		MIN	ТҮР	MAX	UNITS
V _{OS}	Input Offset Voltage (Note 5)	LT1490A N, S Packages $0^{\circ}C \le T_A \le 70^{\circ}C$ $-40^{\circ}C \le T_A \le 85^{\circ}C$	•		150	700 950 1100	μV μV μV
		LT1490A MS8 Package, LT1491A N, S Packages $0^{\circ}C \le T_A \le 70^{\circ}C$ $-40^{\circ}C \le T_A \le 85^{\circ}C$	•		250	1200 1350 1500	μV μV μV
	Input Offset Voltage Drift (Note 9)	$-40^{\circ}\text{C} \le \text{T}_{\text{A}} \le 85^{\circ}\text{C}$	•		2	6	μV/°C
I _{OS}	Input Offset Current		•		0.2	0.8	nA
I _B	Input Bias Current		•		1	8	nA
	Input Bias Current Drift	$-40^{\circ}\text{C} \le \text{T}_{\text{A}} \le 85^{\circ}\text{C}$	•		5		pA/°C
	Input Noise Voltage	0.1Hz to 10Hz			1		μV _{P-P}
e _n	Input Noise Voltage Density	f = 1kHz			50		nV/√Hz
in	Input Noise Current Density	f = 1kHz			0.015		pA/√Hz
R _{IN}	Input Resistance	Differential Common Mode, V _{CM} = –15V to 14V		6	17 15000		MΩ MΩ
C _{IN}	Input Capacitance				4.6		pF
	Input Voltage Range		•	-15		29	V
CMRR	Common Mode Rejection Ratio	V _{CM} = -15V to 29V	•	80	98		dB
A _{VOL}	Large-Signal Voltage Gain	$V_0 = \pm 14V, R_L = 10k$ $0^{\circ}C \le T_A \le 70^{\circ}C$ $-40^{\circ}C \le T_A \le 85^{\circ}C$	•	100 75 50	250		V/mV V/mV V/mV
$\overline{V_0}$	Output Voltage Swing	No Load I _{OUT} = ±5mA I _{OUT} = ±10mA	•	±14.9 ±14.5 ±14.5	±14.978 ±14.750 ±14.670		V V V
I _{SC}	Short-Circuit Current (Note 2)	Short to GND $0^{\circ}C \le T_A \le 70^{\circ}C$ $-40^{\circ}C \le T_A \le 85^{\circ}C$	•	±20 ±15 ±10	±25		mA mA mA
PSRR	Power Supply Rejection Ratio	$V_S = \pm 1.25V \text{ to } \pm 22V$	•	88	98		dB
I _S	Supply Current per Amplifier		•		50	70 85	μA μA
GBW	Gain Bandwidth Product	$ f = 1 \text{kHz} $ $0^{\circ}\text{C} \leq \text{T}_{\text{A}} \leq 70^{\circ}\text{C} $ $-40^{\circ}\text{C} \leq \text{T}_{\text{A}} \leq 85^{\circ}\text{C} $	•	125 110 100	200		kHz kHz kHz
SR	Slew Rate	$A_V = -1$, $R_L = \infty$, $V_0 = \pm 10V$, Measure at $V_0 = \pm 5V$ $0^{\circ}C \le T_A \le 70^{\circ}C$ $-40^{\circ}C \le T_A \le 85^{\circ}C$	•	0.0375 0.0330 0.0300	0.07		V/µs V/µs V/µs



				LT14	90AH/LT14	91AH	
SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
V _{OS}	Input Offset Voltage (Note 5)	LT1490AHS8	•		110	500 2500	μV μV
		LT1490AHMS8, LT1491AHS	•		220	1000 3000	μV μV
	Input Offset Voltage Drift (Note 9)		•		3	6	μV/°C
I _{OS}	Input Offset Current	V _{CM} = 44V (Note 6)	•			2 1.5	nA μA
I _B	Input Bias Current	V _{CM} = 44V (Note 6)	•			20 15	nA μA
	Input Voltage Range		•	0.3		44	V
CMRR	Common Mode Rejection Ratio (Note 6)	$V_{CM} = 0.3V \text{ to } V_{CC} - 1V$ $V_{CM} = 0.3V \text{ to } 44V$	•	60 74			dB dB
A _{VOL}	Large-Signal Voltage Gain	$V_S = 3V$, $V_0 = 500$ mV to 2.5V, $R_L = 10$ k	•	200 25	1500		V/mV V/MV
		$V_S = 5V$, $V_0 = 500$ mV to 4.5V, $R_L = 10$ k	•	400 50	1500		V/mV V/mV
V _{OL}	Output Voltage Swing Low	$V_S = 3V$, No Load $V_S = 3V$, $I_{SINK} = 2.5mA$	•			15 450	mV mV
		$V_S = 5V$, No Load $V_S = 5V$, $I_{SINK} = 2.5mA$	•			15 500	mV mV
V _{OH}	Output Voltage Swing High	$V_S = 3V$, No Load $V_S = 3V$, $I_{SOURCE} = 5mA$	•	2.925 2.350			V
		$V_S = 5V$, No Load $V_S = 5V$, $I_{SOURCE} = 10$ mA	•	4.925 4.100			V V
PSRR	Power Supply Rejection Ratio	$V_S = 2.5V$ to 12.5V, $V_{CM} = V_0 = 1V$	•	80			dB
	Minimum Operating Supply Voltage		•			2.5	V
	Reverse Supply Voltage	$I_S = -100\mu$ A per Amplifier	•	18			V
Is	Supply Current per Amplifier (Note 7)		•		40	50 70	μA μA
GBW	Gain Bandwidth Product (Note 6)	f = 1kHz	•	110 60	180		kHz kHz
SR	Slew Rate (Note 8)	$A_V = -1$, $R_L = \infty$	•	0.035 0.015	0.06		V/µs V/µs



ELECTRICAL CHARACTERISTICS The ullet denotes specifications which apply over the full operating temperature range of $-40^{\circ}\text{C} \le T_{\text{A}} \le 125^{\circ}\text{C}$. $V_{\text{S}} = \pm 15\text{V}$, $V_{\text{CM}} = 0\text{V}$, $V_{\text{OUT}} = 0\text{V}$ unless otherwise noted. (Note 4)

				LT14	90AH/LT14	91AH	
SYMBOL	PARAMETER Input Offset Voltage (Note 5)	CONDITIONS		MIN	TYP	MAX	UNITS
V _{OS}		LT1490AHS8	•		150	700 2700	μV μV
		LT1490AHMS8, LT1491AHS	•		250	1200 3200	μV μV
	Input Offset Voltage Drift (Note 9)		•		3	7	μV/°C
I _{OS}	Input Offset Current		•			2	nA
I _B	Input Bias Current		•			20	nA
	Input Voltage Range		•	-14.7		29	V
CMRR	Common Mode Rejection Ratio	$V_{CM} = -14.7V \text{ to } 29V$	•	72			dB
A _{VOL}	Large-Signal Voltage Gain	$V_0 = \pm 14V, R_L = 10k$	•	100 4	250		V/mV V/mV
V ₀	Output Voltage Swing	No Load I _{OUT} = ±2.5mA	•	±14.8 ±14.3			mV mV
PSRR	Power Supply Rejection Ratio	$V_S = \pm 1.25 V \text{ to } \pm 22 V$	•	84			dB
Is	Supply Current per Amplifier		•		50	70 95	μA μA
GBW	Gain Bandwidth Product	f = 1kHz	•	125 75	200		kHz kHz
SR	Slew Rate	$A_V = -1$, $R_L = \infty$	•	0.0375 0.02	0.07		V/μs V/μs

Note 1: Absolute Maximum Ratings are those values beyond which the life of the device may be impaired.

Note 2: A heat sink may be required to keep the junction temperature below absolute maximum. This depends on the power supply voltage and how many amplifiers are shorted.

Note 3: The LT1490AC/LT1491AC and LT1490AI/LT1491AI are guaranteed functional over the operating temperature range of -40° C to 85°C. The LT1490AH/LT1491AH are guaranteed functional over the operating temperature range of -40° C to 125°C.

Note 4: The LT1490AC/LT1491AC are guaranteed to meet specified performance from 0°C to 70°C. The LT1490AC/LT1491AC are designed, characterized and expected to meet specified performance from -40°C to 85°C but are not tested or QA sampled at these temperatures. The LT1490AI/LT1491AI are guaranteed to meet specified performance from -40°C to 85°C. The LT1490AH/LT1491AH are guaranteed to meet specified performance from -40°C to 125°C.

Note 5: ESD (Electrostatic Discharge) sensitive device. Extensive use of ESD protection devices are used internal to the LT1490A/LT1491A. However, high electrostatic discharge can damage or degrade the device. Use proper ESD handling precautions.

Note 6: $V_S = 5V$ limits are guaranteed by correlation to $V_S = 3V$ and $V_S = \pm 15V$ tests.

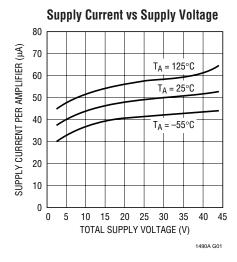
Note 7: $V_S = 3V$ limits are guaranteed by correlation to $V_S = 5V$ and $V_S = \pm 15V$ tests.

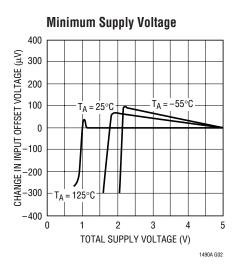
Note 8: Guaranteed by correlation to slew rate at $V_S = \pm 15V$ and GBW at $V_S = 3V$ and $V_S = \pm 15V$ tests.

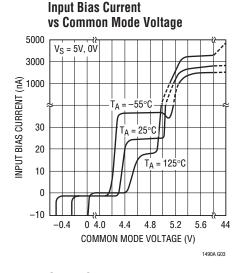
Note 9: This parameter is not 100% tested.

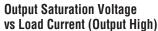


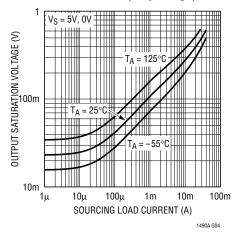
TYPICAL PERFORMANCE CHARACTERISTICS



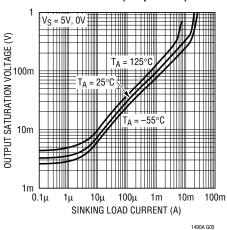




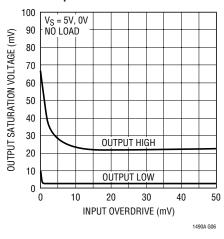




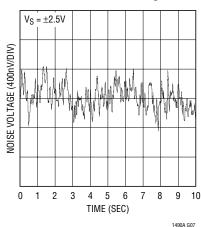




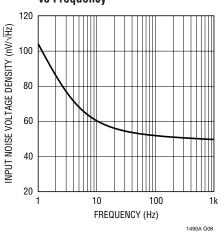
Output Saturation Voltage vs Input Overdrive



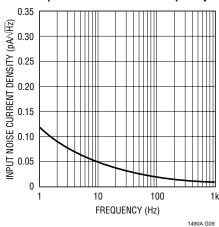
0.1Hz to 10Hz Noise Voltage



Noise Voltage Density vs Frequency

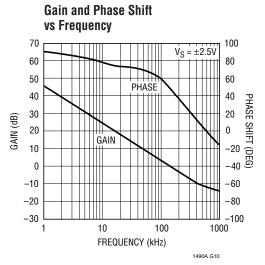


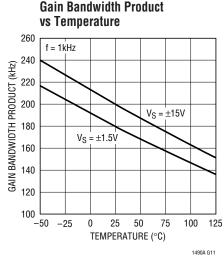
Input Noise Current vs Frequency

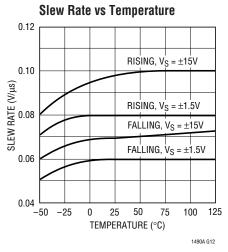


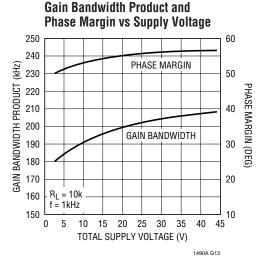


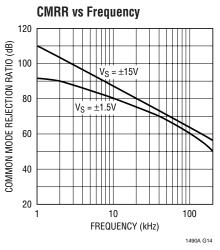
TYPICAL PERFORMANCE CHARACTERISTICS

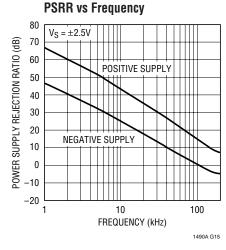


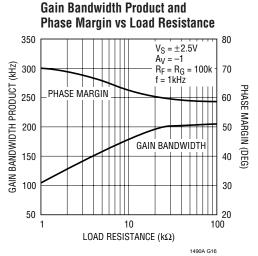


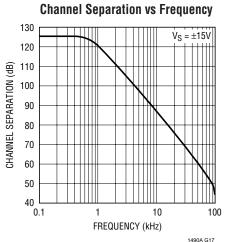


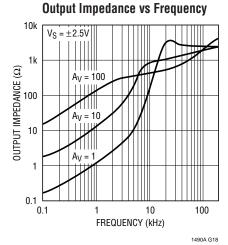






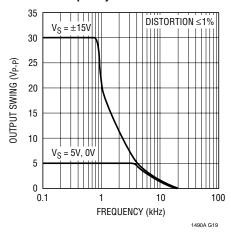




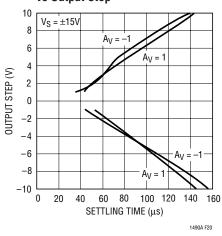


TYPICAL PERFORMANCE CHARACTERISTICS

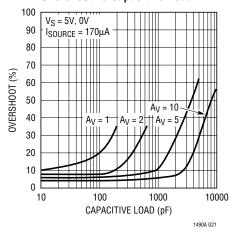
Undistorted Output Swing vs Frequency



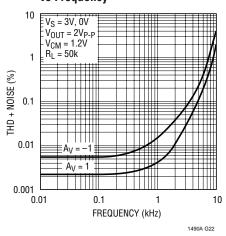
Settling Time to 0.1% vs Output Step



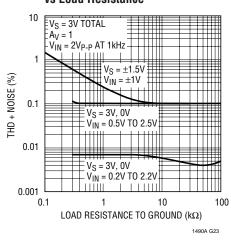
Capacitive Load Handling, **Overshoot vs Capacitive Load**



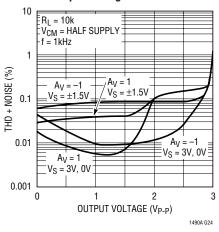
Total Harmonic Distortion + Noise vs Frequency



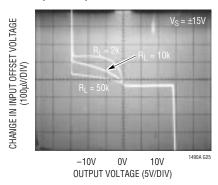
Total Harmonic Distortion + Noise vs Load Resistance



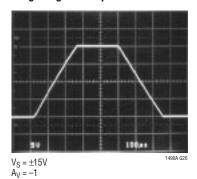
Total Harmonic Distortion + Noise vs Output Voltage



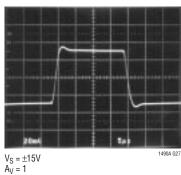
Open-Loop Gain



Large-Signal Response



Small-Signal Response





APPLICATIONS INFORMATION

Supply Voltage

The positive supply pin of the LT1490A/LT1491A should be bypassed with a small capacitor (about $0.01\mu F$) within an inch of the pin. When driving heavy loads an additional $4.7\mu F$ electrolytic capacitor should be used. When using split supplies, the same is true for the negative supply pin.

The LT1490A/LT1491A are protected against reverse battery voltages up to 18V. In the event a reverse battery condition occurs, the supply current is less than 1nA.

The LT1490A/LT1491A can be shut down by removing V $^+$. In this condition the input bias current is typically less than 0.5nA, even if the inputs are 44V above the negative supply.

When operating the LT1490A/LT1491A on total supplies of 20V or more, the supply must not rise to its final voltage in less than 1 μ s. This is especially true if low ESR bypass capacitors are used. A series RLC circuit is formed from the supply lead inductance and the bypass capacitor. A resistance of 7.5 Ω in the supply or in the bypass capacitor will dampen the tuned circuit enough to limit the rise time.

Inputs

The LT1490A/LT1491A have two input stages, NPN and PNP (see the Simplified Schematic), resulting in three distinct operating regions as shown in the Input Bias Current vs Common Mode typical performance curve.

For input voltages about 0.8V or more below V+, the PNP input stage is active and the input bias current is typically -1nA. When the input voltage is about 0.5V or less from V+, the NPN input stage is operating and the input bias current is typically 25nA. Increases in temperature will cause the voltage at which operation switches from the PNP stage to the NPN stage to move towards V+. The input offset voltage of the NPN stage is untrimmed and is typically 600μ V.

A Schottky diode in the collector of each NPN transistor of the NPN input stage allows the LT1490A/LT1491A to operate with either or both of their inputs above V^+ . At about 0.3V above V^+ the NPN input transistor is fully saturated and the input bias current is typically $3\mu A$ at room temperature. The input offset voltage is typically $700\mu V$ when

operating above V⁺. The LT1490A/LT1491A will operate with their inputs 44V above V⁻ regardless of V⁺.

The inputs are protected against excursions as much as 15V below V^- by an internal 1k resistor in series with each input and a diode from the input to the negative supply. There is no output phase reversal for inputs up to 15V below V^- . There are no clamping diodes between the inputs and the maximum differential input voltage is 44V.

Output

The output voltage swing of the LT1490A/LT1491A is affected by input overdrive as shown in the typical performance curves.

The output of the LT1490A/LT1491A can be pulled up to 18V beyond V⁺ with less than 1nA of leakage current, provided that V⁺ is less than 0.5V.

The normally reverse-biased substrate diode from the output to V^- will cause unlimited currents to flow when the output is forced below V^- . If the current is transient and limited to 100mA, no damage will occur.

The LT1490A/LT1491A are internally compensated to drive at least 200pF of capacitance under any output loading conditions. A 0.22 μ F capacitor in series with a 150 Ω resistor between the output and ground will compensate these amplifiers for larger capacitive loads, up to 10,000pF, at all output currents.

Distortion

There are two main contributors of distortion in op amps: output crossover distortion as the output transitions from sourcing to sinking current and distortion caused by nonlinear common mode rejection. Of course, if the op amp is operating inverting there is no common mode induced distortion. When the LT1490A/LT1491A switch between input stages there is significant nonlinearity in the CMRR. Lower load resistance increases the output crossover distortion, but has no effect on the input stage transition distortion. For lowest distortion the LT1490A/LT1491A should be operated single supply, with the output always sourcing current and with the input voltage swing between ground and $(V^+ - 0.8V)$. See the Typical Performance Characteristics curves.



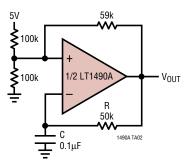
APPLICATIONS INFORMATION

Gain

The open-loop gain is almost independent of load when the output is sourcing current. This optimizes performance in single supply applications where the load is returned to ground. The typical performance photo of Open-Loop Gain for various loads shows the details.

TYPICAL APPLICATIONS

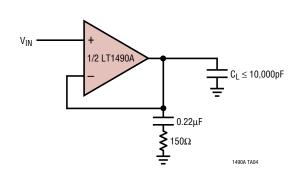
Square Wave Oscillator



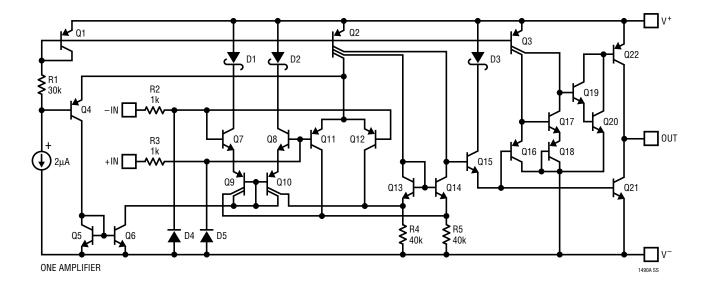
 $f = \frac{1}{2RC}$ $V_{OUT} = 5V_{P-P} \text{ WITH 5V SUPPLY }$ $I_S = 200 \mu \text{A}$

AT $V_S = 5V$, R = 50k, C = 1nFOUTPUT IS 5kHz SLEW LIMITED TRIANGLE WAVE

Optional Output Compensation for Capacitive Loads Greater Than 200pF



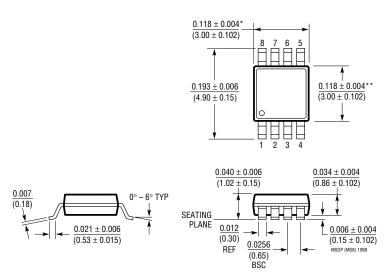
SIMPLIFIED SCHEMATIC





MS8 Package 8-Lead Plastic MSOP

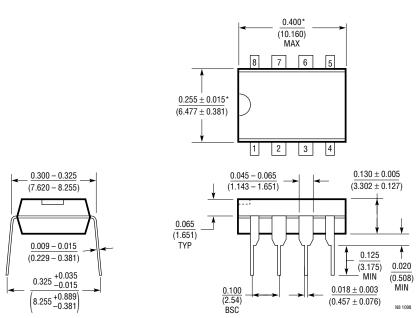
(LTC DWG # 05-08-1660)



- * DIMENSION DOES NOT INCLUDE MOLD FLASH, PROTRUSIONS OR GATE BURRS. MOLD FLASH, PROTRUSIONS OR GATE BURRS SHALL NOT EXCEED 0.006* (0.152mm) PER SIDE
- ** DIMENSION DOES NOT INCLUDE INTERLEAD FLASH OR PROTRUSIONS.
 INTERLEAD FLASH OR PROTRUSIONS SHALL NOT EXCEED 0.006" (0.152mm) PER SIDE

N8 Package 8-Lead PDIP (Narrow 0.300)

(LTC DWG # 05-08-1510)

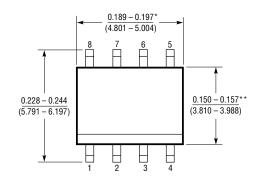


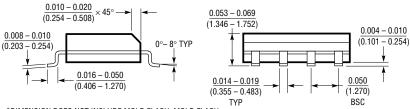
*THESE DIMENSIONS DO NOT INCLUDE MOLD FLASH OR PROTRUSIONS. MOLD FLASH OR PROTRUSIONS SHALL NOT EXCEED 0.010 INCH (0.254mm)



S8 Package 8-Lead Plastic Small Outline (Narrow 0.150)

(LTC DWG # 05-08-1610)





^{*}DIMENSION DOES NOT INCLUDE MOLD FLASH. MOLD FLASH SHALL NOT EXCEED 0.006" (0.152mm) PER SIDE

S08 1298



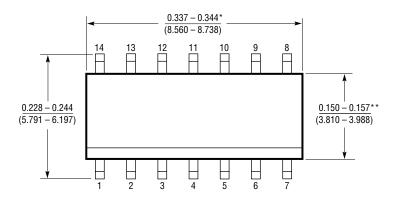
^{**}DIMENSION DOES NOT INCLUDE INTERLEAD FLASH. INTERLEAD FLASH SHALL NOT EXCEED 0.010* (0.254mm) PER SIDE

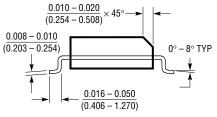
N Package 14-Lead PDIP (Narrow 0.300) (LTC DWG # 05-08-1510)

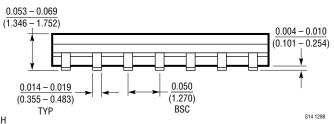
0.770* (19.558)MAX 14 13 12 11 10 9 8 $0.255 \pm 0.015*$ (6.477 ± 0.381) 7 1 2 3 4 5 6 0.300 - 0.3250.045 - 0.065 0.130 ± 0.005 (7.620 - 8.255) $\overline{(3.302 \pm 0.127)}$ (1.143 - 1.651)0.020 (0.508)0.065 MIN 0.009 - 0.015(1.651)(0.229 - 0.381)TYP $0.325 \, {}^{+0.035}_{-0.015}$ 0.005 0.125 0.018 ± 0.003 $\overline{(0.125)}$ 8.255 +0.889 (3.175) (0.457 ± 0.076) MIN 0.100 (2.54)*THESE DIMENSIONS DO NOT INCLUDE MOLD FLASH OR PROTRUSIONS. MOLD FLASH OR PROTRUSIONS SHALL NOT EXCEED 0.010 INCH (0.254mm) BSC

N14 1098

S Package 14-Lead Plastic Small Outline (Narrow 0.150) (LTC DWG # 05-08-1610)





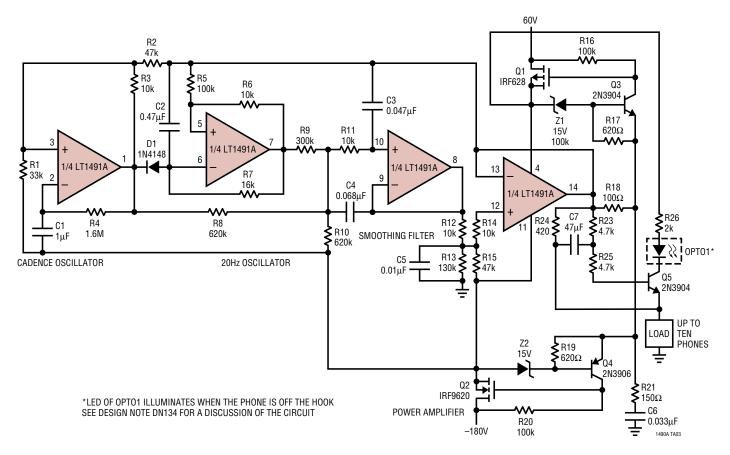


*DIMENSION DOES NOT INCLUDE MOLD FLASH. MOLD FLASH SHALL NOT EXCEED 0.006" (0.152mm) PER SIDE

^{**}DIMENSION DOES NOT INCLUDE INTERLEAD FLASH. INTERLEAD FLASH SHALL NOT EXCEED 0.010" (0.254mm) PER SIDE

TYPICAL APPLICATION

Ring-Tone Generator



RELATED PARTS

PART NUMBER	DESCRIPTION	COMMENTS
LT1366/LT1367	Dual/Quad Precision, Rail-to-Rail Input and Output Op Amps	475μV V _{OS(MAX)} , 500V/mV A _{VOL(MIN)} , 400kHz GBW
LT1636	Single Over-The-Top Micropower Rail-to-Rail Input and Output Op Amp	55μA Supply Current, V _{CM} Extends 44V above V _{EE} , Independent of V _{CC} , MSOP Package, Shutdown Function
LT1638/LT1639	Dual/Quad 1.2MHz Over-The-Top Micropower, Rail-to-Rail Input and Output Op Amps	0.4V/μs Slew Rate, 230μA Supply Current per Amplifier
LT1782	Micropower, Over-The-Top, SOT-23, Rail-to-Rail Input and Output Op Amp	SOT-23, 800µV V _{OS(MAX)} , I _S =55µA (Max), Gain-Bandwidth = 200kHz, Shutdown Pin
LT1783	1.2MHz, Over-The-Top, Micropower, Rail-to-Rail Input and Output Op Amp	SOT-23, 800μV V _{OS(MAX)} , I _S =300μA (Max), Gain-Bandwidth = 1.2MHz, Shutdown Pin