

Features

- ☐ Double-conversion superhet architecture for high degree of image rejection
- ☐ FSK for digital data and FM reception for analog signal transmission
- ☐ FSK/FM demodulation with phase-coincidence demodulator
- ☐ Low current consumption in active mode and very low standby current
- ☐ Switchable LNA gain for improved dynamic range
- ☐ RSSI allows signal strength indication and ASK detection
- ☐ 32-pin Low profile Quad Flat Package (LQFP)

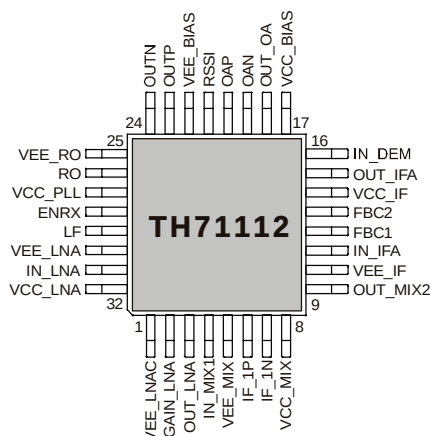
Ordering Information

Part No.	Temperature Code	Package Code	
Part Number	Temperature Code	Package Code	Delivery Form
TH71112	E (-40 °C to 85 °C)	NE (LQFP32)	250 pc/tray 2000 pc/T&R

Application Examples

- ☐ General digital data transmission
- ☐ Tire Pressure Monitoring Systems (TPMS)
- ☐ Remote Keyless Entry (RKE)
- ☐ Wireless access control
- ☐ Alarm and security systems
- ☐ Garage door openers
- ☐ Remote Controls
- ☐ Home and building automation
- ☐ Low-power telemetry systems

Pin Description



General Description

The TH71112 FSK/FM/ASK double-conversion superheterodyne receiver IC is designed for applications in the European 868 MHz industrial-scientific-medical (ISM) band, according to the EN 300 220 telecommunications standard. It can also be used for any other system with carrier frequencies ranging from 750 MHz to 990 MHz (e.g. for applications according to FCC part 15).

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1 Theory of Operation

1.1 General

With the TH71112 receiver chip, various circuit configurations can be arranged in order to meet a number of different customer requirements. For FM/FSK reception the IF tank used in the phase coincidence demodulator can also be constituted by a ceramic discriminator with a varactor diode to create an AFC circuit. In ASK configuration, the RSSI signal is fed to an ASK detector, which is constituted by the operational amplifier.

The superheterodyne configuration is double conversion where MIX1 and MIX2 are driven by the internal local oscillator signals LO1 and LO2, respectively. This allows a high degree of image rejection, achieved in conjunction with an RF front-end filter. Efficient RF front-end filtering is realized by using a SAW, ceramic or helix filter in front of the LNA and by adding an LC filter at the LNA output.

A single-conversion variant, called TH71111, is also available. Both Receiver ICs have the same die. At the TH71111 the second mixer MIX2 operates as an amplifier.

The TH71112 receiver IC consists of the following building blocks:

- PLL synthesizer (PLL SYNTH) for generation of the first and second local oscillator signals LO1 and LO2, parts of the PLL SYNTH are: the high-frequency VCO1, the feedback dividers DIV_16 and DIV_2, a phase-frequency detector (PFD) with charge pump (CP) and a crystal-based reference oscillator (RO)
- Low-noise amplifier (LNA) for high-sensitivity RF signal reception
- First mixer (MIX1) for down-conversion of the RF signal to the first IF (IF1)
- Second mixer (MIX2) for down-conversion of the IF1 to the second IF (IF2)
- IF amplifier (IFA) to amplify and limit the IF2 signal and for RSSI generation
- Phase coincidence demodulator (DEMODO) with third mixer (MIX3) to demodulate the IF signal
- Operational amplifier (OA) for data slicing, filtering and ASK detection
- Bias circuitry for bandgap biasing and circuit shutdown

1.2 Technical Data Overview

- | | |
|--|--|
| ☐ Input frequency range: 750 MHz to 990 MHz | ☐ Range of IF2: 400 kHz to 22 MHz |
| ☐ Power supply range: 2.3 V to 5.5 V @ ASK | ☐ Maximum input level: -10 dBm @ ASK |
| ☐ Temperature range: -40 °C to +85 °C | ☐ 0 dBm @ FSK |
| ☐ Standby current: 50 nA | ☐ Image rejection: > 60 dB (e.g. with 868.3 MHz SAW front-end filter and at 10.7 MHz IF2) |
| ☐ Operating current: 7.5 mA @ low gain mode | ☐ Spurious emission: < -70 dBm |
| ☐ 9.2 mA @ high gain mode | ☐ Input frequency acceptance range: up to ± 100 kHz |
| ☐ Sensitivity: -112 dBm @ ASK 1) | ☐ RSSI range: 70 dB |
| ☐ -106 dBm @ FSK 2) | ☐ FM/FSK deviation range: ± 2.5 kHz to ± 80 kHz |
| ☐ Maximum data rate: 260 kbps NRZ @ ASK | ☐ Maximum analog modulation frequency: 15 kHz |
| ☐ 180 kbps NRZ @ FSK | |
| ☐ Range of IF1: 10 MHz to 80 MHz | |

1) at 4 kbps NRZ, BER = $3 \cdot 10^{-3}$, 180 kHz IF filter BW, without SAW front-end-filter loss

2) at 4 kbps NRZ, BER = $3 \cdot 10^{-3}$, ± 20 kHz FSK deviation, 180 kHz IF filter BW, without SAW front-end-filter loss

1.3 Block Diagram

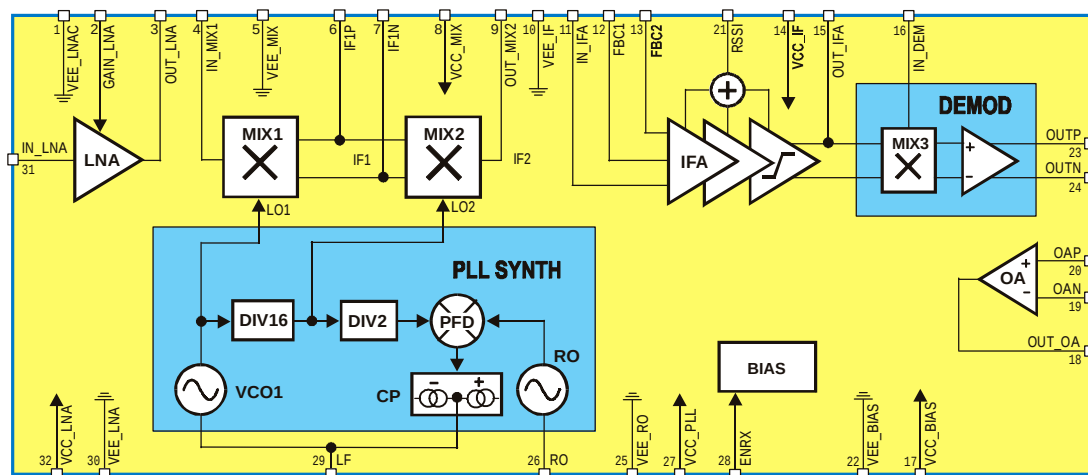


Fig. 1: TH71112 block diagram

1.4 Mode Configurations

ENRX	Mode	Description
0	RX standby	RX disabled
1	RX active	RX enable

Note: ENRX are pulled down internally

1.5 LNA GAIN Control

V _{GAIN_LNA}	Mode	Description
< 0.8 V	HIGH GAIN	LNA set to high gain
> 1.4 V	LOW GAIN	LNA set to low gain

Note: hysteresis between gain modes to ensure stability

1.6 Frequency Planning

Frequency planning is straightforward for single-conversion applications because there is only one IF that can be chosen, and then the only possible choice is low-side or high-side injection of the LO signal (which is now the one and only LO signal in the receiver).

The receiver's double-conversion architecture requires careful frequency planning. Besides the desired RF input signal, there are a number of spurious signals that may cause an undesired response at the output. Among them are the image of the RF signal (that must be suppressed by the RF front-end filter), spurious signals injected to the first IF (IF1) and their images which could be mixed down to the same second IF (IF2) as the desired RF signal (they must be suppressed by the LC filter at IF1 and/or by low-crosstalk design).

By configuring the TH71112 for double conversion and using its internal PLL synthesizer with fixed feedback divider ratios of $N1 = 16$ (DIV_16) and $N2 = 2$ (DIV_2), four types of down-conversion are possible: low-side injection of LO1 and LO2 (**low-low**), LO1 low-side and LO2 high-side (**low-high**), LO1 high-side and LO2 low-side (**high-low**) or LO1 and LO2 high-side (**high-high**). The following table summarizes some equations that are useful to calculate the crystal reference frequency (REF), the first IF (IF1) and the VCO1 or first LO frequency (LO1), respectively, for a given RF and second IF (IF2).

Injection type	high-high	low-low	high-low	low-high
REF	$(RF - IF2)/30$	$(RF - IF2)/34$	$(RF + IF2)/30$	$(RF + IF2)/34$
LO1	$32 \bullet REF$	$32 \bullet REF$	$32 \bullet REF$	$32 \bullet REF$
IF1	$LO1 - RF$	$RF - LO1$	$LO1 - RF$	$RF - LO1$
LO2	$2 \bullet REF$	$2 \bullet REF$	$2 \bullet REF$	$2 \bullet REF$
IF2	$LO2 - IF1$	$IF1 - LO2$	$IF1 - LO2$	$LO2 - IF1$

1.6.1 Selected Frequency Plans

The following table depicts crystal, LO and image signals considering the examples of 868.3 MHz and 915 MHz RF reception at $IF2 = 10.7$ MHz. The columns in bold depict the selected frequency plans to receive at 868.3 MHz and 915 MHz, respectively.

Signal type	RF = 868.3 MHz	RF = 868.3 MHz	RF = 868.3 MHz	RF = 868.3 MHz	RF = 915 MHz	RF = 915 MHz	RF = 915 MHz	RF = 915 MHz
Injection type	high-high	low-low	high-low	low-high	high-high	low-low	high-low	low-high
REF / MHz	28.58667	25.22353	29.3	25.85294	30.14333	26.59706	30.85667	27.22647
LO1 / MHz	914.77333	807.15294	937.6	827.29412	964.58667	851.10588	987.41333	871.24706
IF1 / MHz	46.47333	61.14706	69.3	41.00588	49.58667	63.89412	72.41333	43.75294
LO2 / MHz	57.17333	50.44706	58.6	51.70588	60.28667	53.19412	61.71333	54.45294
RF image/MHz	961.24667	746.00588	1006.9	786.28824	1014.17	787.21176	1059.83	827.49412
IF1 image/MHz	67.87333	39.74706	47.9	62.40588	70.98667	42.49412	51.01333	65.15294

1.6.2 Maximum Frequency Coverage

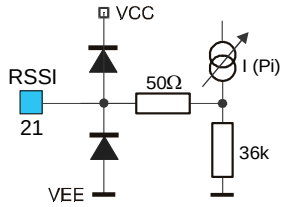
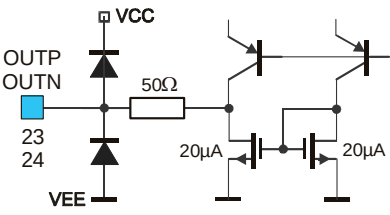
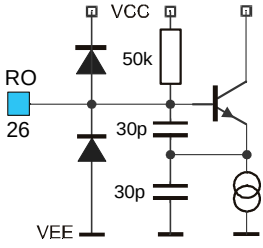
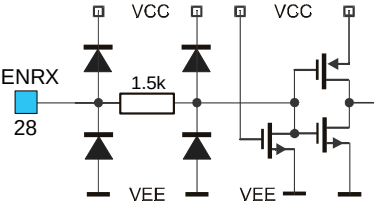
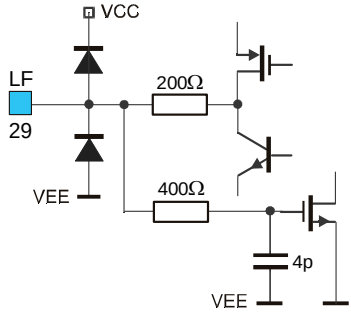
Parameter	f_{min}	f_{max}
Injection type	high-low	low-low
RF / MHz	739.3	998.825
REF / MHz	25.0	29.0625
LO1 / MHz	800	930
IF1 / MHz	60.7	68.825
LO2 / MHz	50.0	58.125
IF2/ MHz	10.7	10.7

The selection of the reference crystal frequency is based on some assumptions. As for example: the first IF and the image frequencies should not be in a radio band where strong interfering signals might occur (because they could represent parasitic receiving signals), the LO1 signal should be in the range of 800 MHz to 930 MHz (because this is the optimum frequency range of the VCO1). Furthermore the first IF should be as high as possible to achieve highest RF image rejection.

2 Pin Definitions and Descriptions

Pin No.	Name	I/O Type	Functional Schematic	Description
3	OUT_LNA	analog output		LNA open-collector output, to be connected to external LC tank that resonates at RF
31	IN_LNA	analog input		LNA input, approx. 26Ω single-ended
1	VEE_LNAC	ground		ground of LNA core (cas-code)
2	GAIN_LNA	analog input		LNA gain control (input with hysteresis) RX standby: no pull-up RX active: pull-up
4	IN_MIX1	analog input		MIX1 input, approx. 33Ω single-ended
5	VEE_MIX	ground		ground of MIX1 and MIX2
6	IF1P	analog I/O		open-collector output, to be connected to external LC tank that resonates at first IF
7	IF1N	analog I/O		open-collector output, to be connected to external LC tank that resonates at first IF
8	VCC_MIX	supply		positive supply of MIX1 and MIX2
9	OUT_MIX2	analog output		MIX2 output, approx. 330Ω output impedance
10	VEE_IF	ground		ground of IFA and DEMOD

Pin No.	Name	I/O Type	Functional Schematic	Description
11	IN_IFA	analog input		IFA input, approx. $2.2k\Omega$ input impedance
12	FBC1	analog I/O		to be connected to external IFA feedback capacitor
13	FBC2	analog I/O		to be connected to external IFA feedback capacitor
14	VCC_IF	supply		positive supply of IFA and DEMOD
15	OUT_IFA	analog I/O		IFA output and MIX3 input (of DEMOD)
16	IN_DEM	analog input		DEMOM input, to MIX3 core
17	VCC_BIAS	supply		positive supply of general bias system and OA
18	OUT_OA	analog output		OA output, 40uA current drive capability
19	OAN	analog input		negative OA input
20	OAP	analog input		positive OA input

Pin No.	Name	I/O Type	Functional Schematic	Description
21	RSSI	analog output		RSSI output, for RSSI and ASK detection, approx. 36kΩ output impedance
22	VEE_BIAS	ground		ground of general bias system and OA
23	OUTP	analog output		FSK/FM positive output, output impedance of 100kΩ to 300kΩ
24	OUTN	analog output		FSK/FM negative output, output impedance of 100kΩ to 300kΩ
25	VEE_RO	ground		ground of DIV, PFD, RO and charge pump
26	RO	analog input		RO input, Colpitts type oscillator with internal feed-back capacitors
27	VCC_PLL	supply		positive supply of DIV, PFD, RO and charge pump
28	ENRX	digital input		mode control input, CMOS-compatible with internal pull-down circuit
29	LF	analog I/O		charge pump output and VCO1 control input
30	VEE_LNA	ground		ground of LNA biasing
32	VCC_LNA	supply		positive supply of LNA biasing

3 Technical Data

3.1 Absolute Maximum Ratings

Parameter	Symbol	Condition / Note	Min	Max	Unit
Supply voltage	V_{CC}		0	7.0	V
Input voltage	V_{IN}		- 0.3	$V_{CC}+0.3$	V
Input RF level	P_{IRF}	@ LNA input		10	dBm
Storage temperature	T_{STG}		-40	+125	°C
Junction temperature	T_J			+150	°C
Thermal Resistance	R_{thJA}			60	K/W
Power dissipation	P_{diss}			0.1	W
Electrostatic discharge	V_{ESD1}	human body model, 3)	-1.0	+1.0	kV
	V_{ESD2}	human body model, 4)	-0.75	+0.75	

3) all pins except OUT_LNA, IF1P and IF1N

4) pin OUT_LNA, IF1P and IF1N

3.2 Normal Operating Conditions

Parameter	Symbol	Condition	Min	Max	Unit
Supply voltage	$V_{CC, FSK}$	0 °C to 85 °C	2.5	5.5	V
		-20 °C to 85 °C	2.6	5.5	
		-40 °C to 85 °C	2.7	5.5	
	$V_{CC, ASK}$	-40 °C to 85 °C	2.3	5.5	
Operating temperature	T_A		-40	+85	°C
Input low voltage (CMOS)	V_{IL}	ENRX pin		$0.3 \cdot V_{CC}$	V
Input high voltage (CMOS)	V_{IH}	ENRX pin	$0.7 \cdot V_{CC}$		V
Input frequency range	f_i		739.3	998.8	MHz
First IF range	f_{IF1}		10	80	MHz
Second IF range	f_{IF2}		0.4	22	MHz
XOSC frequency	f_{ref}	set by the crystal	25	29.063	MHz
VCO frequency	f_{LO}	$f_{LO} = 16 \cdot f_{ref}$	800	930	MHz
Frequency deviation	Δf	at FSK or FM	± 2.5	± 80	kHz
FSK data rate	R_{FSK}	NRZ, C15 = NIP, 5)		180	kbps
ASK data rate	R_{ASK}	NRZ, C16 = NIP, 5)		260	kbps
FM bandwidth	f_m			15	kHz

5) $B_{IF} = 400$ kHz, $P_{IN} = -90$ dBm

3.3 Crystal Parameters

Parameter	Symbol	Condition	Min	Max	Unit
Crystal frequency	f_0	fundamental mode, AT	25	29.063	MHz
Load capacitance	C_L		10	15	pF
Static capacitance	C_0			7	pF
Series resistance	R_1			50	Ω

3.4 DC Characteristics

all parameters under normal operating conditions, unless otherwise stated;
typical values at $T_A = 23\text{ }^{\circ}\text{C}$ and $V_{CC} = 3\text{ V}$

Parameter	Symbol	Condition	Min	Typ	Max	Unit
Operating Currents						
Standby current	I_{SBY}	ENRX=0		50	100	nA
Supply current at low gain	$I_{CC, low}$	ENRX=1 GAIN_LNA=1	4.5	7.5	12.0	mA
Supply current at high gain	$I_{CC, high}$	ENRX=1 GAIN_LNA=0	5.0	9.2	14.0	mA
Digital Pin Characteristics						
Input low voltage CMOS	V_{IL}	ENRX pin	-0.3		$0.3 \cdot V_{CC}$	V
Input high voltage CMOS	V_{IH}	ENRX pin	$0.7 \cdot V_{CC}$		$V_{CC} + 0.3$	V
Pull down current ENRX pin	I_{PDEN}	ENRX=1	0.1	2	10	μA
Low level input current ENRX pin	I_{INLEN}	ENRX=0			0.05	μA
Analog Pin Characteristics						
High level input current GAIN_LNA pin	$I_{INHGAIn}$	GAIN_LNA=1			0.05	μA
Pull up current GAIN_LNA pin active	$I_{PUGAINa}$	GAIN_LNA=0 ENRX=1	0.08	0.15	0.3	μA
Pull up current GAIN_LNA pin standby	$I_{PUGAINs}$	GAIN_LNA=0 ENRX=0			0.05	μA
High gain input voltage	V_{IHGAIn}	ENRX=1			0.7	V
Low gain input voltage	V_{ILGAIn}	ENRX=1	1.5			V
Opamp Characteristics						
Opamp input offset voltage	V_{offs}		-35		35	mV
Opamp input offset current	I_{offs}	$I_{OAP} - I_{OAN}$	-50		50	nA
Opamp input bias current	I_{bias}	$0.5 \cdot (I_{OAP} + I_{OAN})$	-150		150	nA
RSSI Characteristics						
RSSI voltage at low input level	$V_{RSSI, low}$	$P_i = -65\text{ dBm}$, GAIN_LNA=1	0.5	1.0	1.5	V
RSSI voltage at high input level	$V_{RSSI, high}$	$P_i = -35\text{ dBm}$, GAIN_LNA=1	1.2	1.9	2.5	V

3.5 AC System Characteristics

all parameters under normal operating conditions, unless otherwise stated;
typical values at $T_A = 23\text{ °C}$ and $V_{CC} = 3\text{ V}$,
RF at 868.3 MHz; SAW frond-end filter loss and IF at 10.7 MHz;
all parameters based on test circuits as shown in Fig. 2, Fig.3 and Fig. 5

Parameter	Symbol	Condition	Min	Typ	Max	Unit
Receive Characteristics						
Input sensitivity – FSK (standard)	$P_{\min, ST}$	$B_{IF} = 180\text{kHz}$, $\Delta f = \pm 20\text{kHz}$, 4kbps NRZ, $BER \leq 3 \cdot 10^{-3}$, 6)		-103		dBm
Input sensitivity – FSK (narrow band)	$P_{\min, NB}$	$B_{IF} = 30\text{kHz}$, $\Delta f = \pm 5\text{kHz}$, 4kbps NRZ, $BER \leq 3 \cdot 10^{-3}$, 6)		-105		dBm
Input sensitivity – ASK	$P_{\min, ASK}$	$B_{IF} = 180\text{kHz}$, 4kbps NRZ, $BER \leq 3 \cdot 10^{-3}$, 6)		-109		dBm
Maximum input signal – FSK/FM	$P_{\max, FSK}$	$BER \leq 3 \cdot 10^{-3}$ GAIN_LNA = 1		0		dBm
Maximum input signal – ASK	$P_{\max, ASK}$	$BER \leq 3 \cdot 10^{-3}$ GAIN_LNA = 1		-10		dBm
Spurious emission	P_{spur}				-70	dBm
Image rejection	ΔP_{imag}			60		dB
Start-up Parameters						
Crystal start-up time	T_{XTL}	ENRX from 0 to 1			0.9	ms
Receiver start-up time	T_{RX}	ENRX from 0 to 1, depends on data slicer time constant, valid data at output			T_{XTL} + $R4 \cdot C17$	
PLL Parameters						
VCO gain	K_{VCO}			350		MHz/V
Charge pump current	I_{CP}			60		μA

6) incl. 3 dB loss of front-end SAW filter

4 Test Circuits

4.1 Standard FSK Reception

4.1.1 Standard FSK Application Circuit

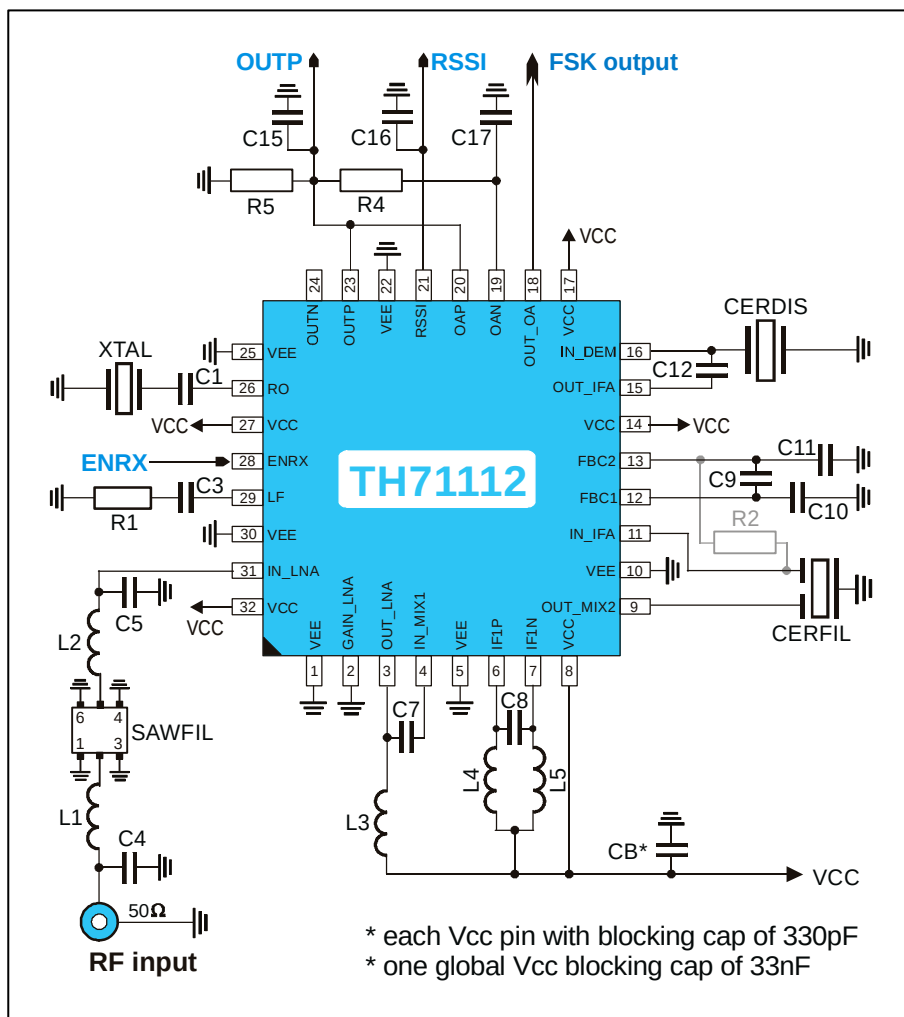


Fig. 2: Test circuit for FSK reception

Circuit Features

- Tolerates input frequency variations
- Well-suited for NRZ, Manchester and similar codes

4.1.2 Standard FSK Component List

Part	Size	Value @ 868.3 MHz	Tolerance	Description
C1	0805	22 pF	±5%	crystal series capacitor
C3	0805	1 nF	±10%	loop filter capacitor
C4	0603	4.7 pF	±5%	capacitor to match SAW filter input
C5	0603	2.7 pF	±5%	capacitor to match SAW filter output
C7	0603	1.0 pF	±5%	MIX1 input matching capacitor
C8	0603	22 pF	±5%	IF1 tank capacitor
C9	0805	33 nF	±10%	IFA feedback capacitor
C10	0603	1 nF	±10%	IFA feedback capacitor
C11	0603	1 nF	±10%	IFA feedback capacitor
C12	0805	10 pF	±5%	DEMODO phase-shift capacitor
C15	0805	100 pF	±5%	demodulator output low-pass capacitor, this value for data rates < 20 kbps NRZ
C16	0603	1.5 nF	±10%	RSSI output low-pass capacitor
C17	0805	10 nF	±10%	data slicer capacitor, this value for data rates > 0.8 kbps NRZ
R1	0805	10 kΩ	±5%	loop filter resistor
R2	0603	330 Ω	±5%	optional CERFIL output matching resistor
R4	0603	330 kΩ	±5%	data slicer resistor
R5	0805	220 kΩ	±5%	loading resistor
L1	0603	22 nH	±5%	SAW filter matching inductor from Würth-Elektronik (WE-KI series), or equivalent part
L2	0603	22 nH	±5%	
L3	0603	10 nH	±5%	LNA output tank inductor from Würth-Elektronik (WE-KI series), or equivalent part
L4	0603	100 nH	±5%	IF1 tank inductor from Würth-Elektronik (WE-KI series) or equivalent part
L5	0603	100 nH	±5%	
XTAL	SMD 6x3.5	25.22353 MHz @ RF = 868.3 MHz	±25ppm cal. ±30ppm temp.	fundamental-mode crystal from Telcona/Horizon or equivalent par
SAWFIL	SMD 3x3	SAFCC868MSL0X00 (f ₀ = 868.3 MHz)	B _{3dB} = 2 MHz	low-loss SAW filter from Murata, or equivalent part
CERFIL	SMD 3.45x3.1	SFECF10M7HA00	B _{3dB} = 180 kHz	ceramic filter from Murata, or equivalent part
CERDIS	SMD 4.5x2	CDSCB10M7GA135		ceramic discriminator from Murata, or equivalent part

- For component values for other frequencies, please refer to the EVB descriptions

4.2 Narrow Band FSK Reception

4.2.1 Narrow Band FSK Application Circuit

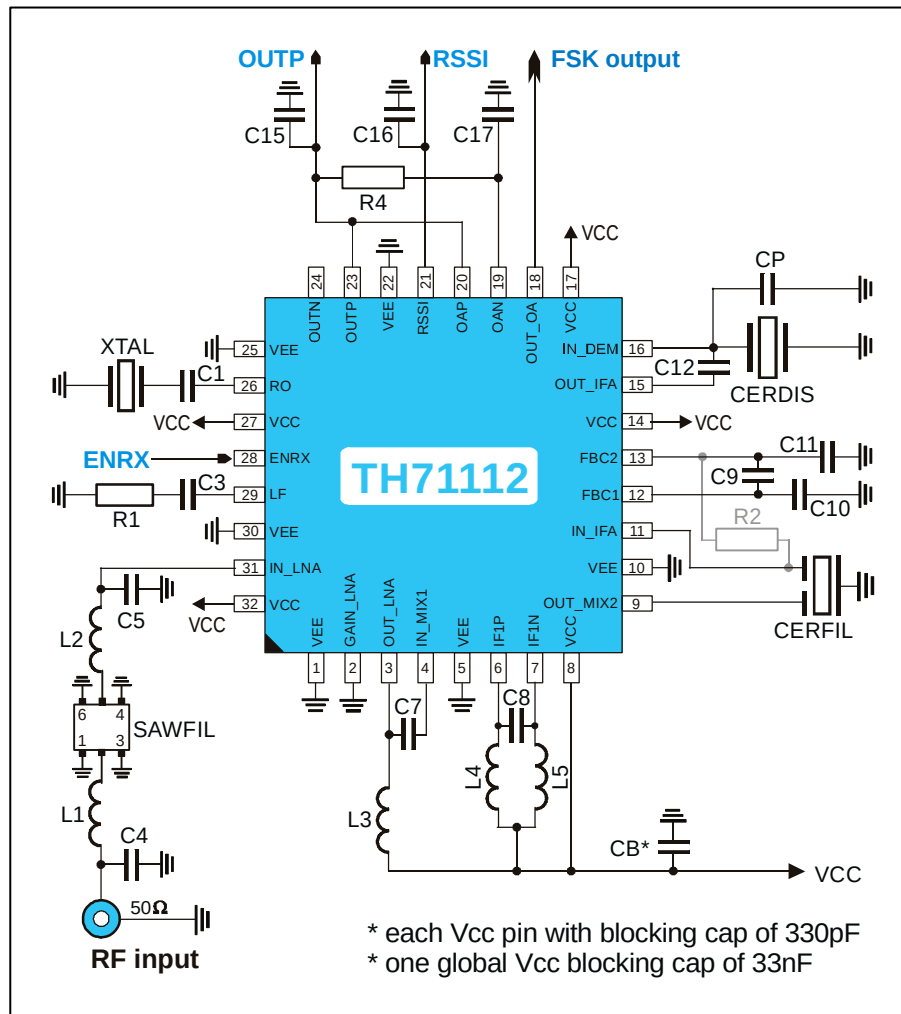


Fig. 3: Test circuit for FM reception

Circuit Features

- Applicable for narrow band FSK

4.2.2 Narrow Band FSK Component List

Part	Size	Value @ 868.3 MHz	Tolerance	Description
C1	0805	22 pF	±5%	crystal series capacitor
C3	0805	1 nF	±10%	loop filter capacitor
C4	0603	4.7 pF	±5%	capacitor to match SAW filter input
C5	0603	2.7 pF	±5%	capacitor to match SAW filter output
C7	0603	1.0 pF	±5%	MIX1 input matching capacitor
C8	0603	22 pF	±5%	IF1 tank capacitor
C9	0805	33 nF	±10%	IFA feedback capacitor
C10	0603	1 nF	±10%	IFA feedback capacitor
C11	0603	1 nF	±10%	IFA feedback capacitor
C12	0805	1.5 pF	±5%	DEMODO phase-shift capacitor
C15	0805	220 pF	±5%	demodulator output low-pass capacitor, this value for data rates < 10 kbps NRZ
C16	0603	1.5 nF	±10%	RSSI output low-pass capacitor
C17	0805	10 nF	±10%	data slicer capacitor, this value for data rates > 0.8 kbps NRZ
CP	0603	6.8 - 8.2 pF	±5%	ceramic resonator loading capacitor
R1	0805	10 kΩ	±5%	loop filter resistor
R2	0603	330 Ω	±5%	optional CERFIL output matching resistor
R4	0603	330 kΩ	±5%	data slicer resistor, this value for 0.4 to 10 kbps NRZ
L1	0603	22 nH	±5%	SAW filter matching inductor from Würth-Elektronik (WE-KI series), or equivalent part
L2	0603	22 nH	±5%	
L3	0603	10 nH	±5%	LNA output tank inductor from Würth-Elektronik (WE-KI series), or equivalent part
L4	0603	100 nH	±5%	IF1 tank inductor from Würth-Elektronik (WE-KI series) or equivalent part
L5	0603	100 nH	±5%	
XTAL	SMD 6x3.5	25.22353 MHz @ RF = 868.3 MHz	±25ppm cal. ±30ppm temp.	fundamental-mode crystal from Telcona/Horizon or equivalent par
SAWFIL	SMD 3x3	SAFCC868MSL0X00 (f ₀ = 868.3 MHz)	B _{3dB} = 2 MHz	low-loss SAW filter from Murata, or equivalent part
CERFIL	Leaded type	SFKLA10M7NL00	B _{3dB} = 30 kHz	ceramic filter from Murata, or equivalent part
		SFVLA10M7LF00	B _{3dB} = 80 kHz	optional, ceramic filter from Murata, or equivalent part
CERDIS	SMD 4.5x2	CDSCB10M7GA135		ceramic discriminator from Murata, or equivalent part

- For component values for other frequencies, please refer to the EVB descriptions

4.3 FM Reception (Varactor-tuned AFC)

4.3.1 FM Application Circuit

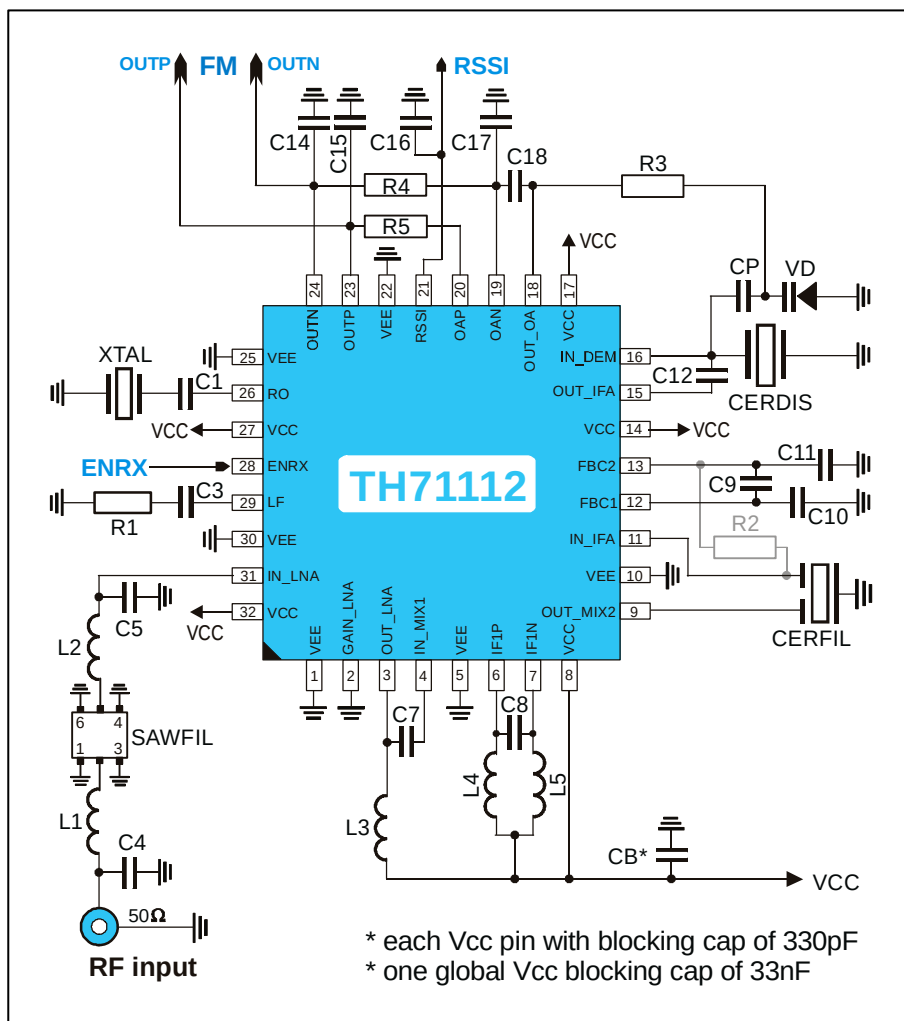


Fig. 4: Test circuit for FM reception

Circuit Features

- Tolerates input frequency variations
- Applicable for FM

4.3.2 FM Component List

Part	Size	Value @ 868.3 MHz	Tolerance	Description
C1	0805	22 pF	±5%	crystal series capacitor
C3	0805	1 nF	±10%	loop filter capacitor
C4	0603	4.7 pF	±5%	capacitor to match SAW filter input
C5	0603	2.7 pF	±5%	capacitor to match SAW filter output
C7	0603	1.0 pF	±5%	MIX1 input matching capacitor
C8	0603	22 pF	±5%	IF1 tank capacitor
C9	0805	33 nF	±10%	IFA feedback capacitor
C10	0603	1 nF	±10%	IFA feedback capacitor
C11	0603	1 nF	±10%	IFA feedback capacitor
C12	0805	1.5 pF	±5%	DEMODO phase-shift capacitor
C14	0805	1 nF	±5%	demodulator output low-pass capacitor, this value for 0.5 to 2 kHz
C15	0805	1 nF	±5%	demodulator output low-pass capacitor this value for 0.5 to 2 kHz
C16	0603	1.5 nF	±10%	RSSI output low-pass capacitor
C17	0805	10 nF	±10%	integrator capacitor
C18	0805	33 nF	±10%	integrator capacitor, this value for 0.5 to 2 kHz
CP	0805	10 pF	±5%	ceramic resonator loading capacitor
R1	0805	10 kΩ	±5%	loop filter resistor
R2	0603	330 Ω	±5%	optional CERFIL output matching resistor
R3	0805	100 kΩ	±10%	varactor diode biasing resistor
R4	0603	680 kΩ	±5%	integrator resistor
R5	0805	680 kΩ	±5%	integrator resistor
L1	0603	22 nH	±5%	SAW filter matching inductor from Würth-Elektronik (WE-KI series), or equivalent part
L2	0603	22 nH	±5%	
L3	0603	10 nH	±5%	LNA output tank inductor from Würth-Elektronik (WE-KI series), or equivalent part
L4	0603	100 nH	±5%	IF1 tank inductor from Würth-Elektronik (WE-KI series) or equivalent part
L5	0603	100 nH	±5%	
VD	SOD-323	BB535		varactor diode from Infineon
XTAL	SMD 6x3.5	25.22353 MHz @ RF = 868.3 MHz	±25ppm cal. ±30ppm temp.	fundamental-mode crystal from Telcona/Horizon or equivalent par
SAWFIL	SMD 3x3	SAFCC868MSL0X00 (f ₀ = 868.3 MHz)	B _{3dB} = 2 MHz	low-loss SAW filter from Murata, or equivalent part
CERFIL	SMD 3.45x3.1	SFECF10M7HA00	B _{3dB} = 180 kHz	ceramic filter from Murata, or equivalent part
CERDIS	SMD 4.5x2	CDSCB10M7GA135		ceramic discriminator from Murata, or equivalent part

- For component values for other frequencies, please refer to the EVB descriptions

4.4 ASK Reception

4.4.1 ASK Application Circuit

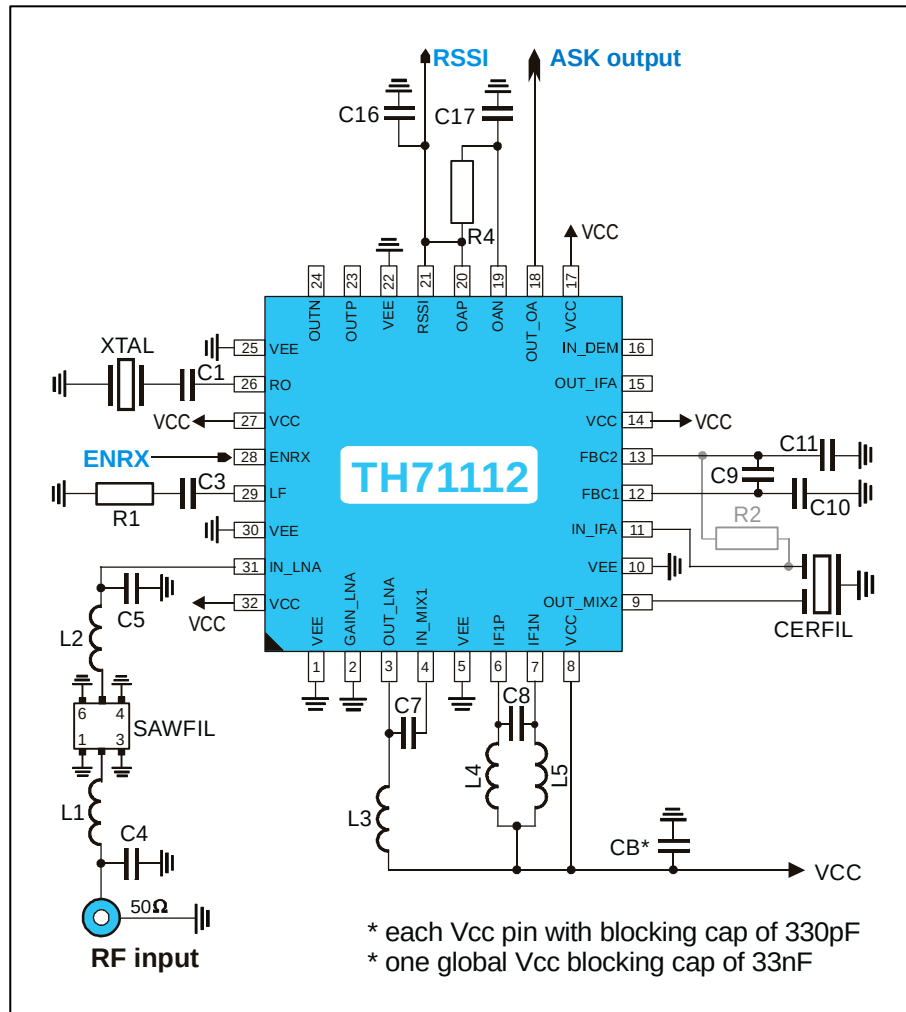


Fig. 4: Test circuit for ASK reception

4.4.2 ASK Component List

Part	Size	Value @ 868.3 MHz	Tolerance	Description
C1	0805	22 pF	±5%	crystal series capacitor
C3	0805	1 nF	±10%	loop filter capacitor
C4	0603	4.7 pF	±5%	capacitor to match SAW filter input
C5	0603	2.7 pF	±5%	capacitor to match SAW filter output
C7	0603	1.0 pF	±5%	MIX1 input matching capacitor
C8	0603	22 pF	±5%	IF1 tank capacitor
C9	0805	33 nF	±10%	IFA feedback capacitor
C10	0603	1 nF	±10%	IFA feedback capacitor
C11	0603	1 nF	±10%	IFA feedback capacitor
C16	0603	1.5 nF	±10%	RSSI output low-pass capacitor, this value for data rates < 10 kbps NRZ, for higher data rates decrease the value
C17	0805	10 nF	±10%	data slicer capacitor, this value for data rates > 0.8 kbps NRZ, for lower data rates increase the value
R1	0805	10 kΩ	±5%	loop filter resistor
R2	0603	330 Ω	±5%	optional CERFIL output matching resistor
R4	0603	330 kΩ	±5%	data slicer resistor
L1	0603	22 nH	±5%	SAW filter matching inductor from Würth-Elektronik (WE-KI series), or equivalent part
L2	0603	22 nH	±5%	
L3	0603	10 nH	±5%	LNA output tank inductor from Würth-Elektronik (WE-KI series), or equivalent part
L4	0603	100 nH	±5%	IF1 tank inductor from Würth-Elektronik (WE-KI series) or equivalent part
L5	0603	100 nH	±5%	
XTAL	SMD 6x3.5	25.22353 MHz @ RF = 868.3 MHz	±25ppm cal. ±30ppm temp.	fundamental-mode crystal from Telcona/Horizon or equivalent par
SAWFIL	SMD 3x3	SAFCC868MSL0X00 (f ₀ = 868.3 MHz)	B _{3dB} = 2 MHz	low-loss SAW filter from Murata, or equivalent part
CERFIL	SMD 3.45x3.1	SFECF10M7HA00	B _{3dB} = 180 kHz	ceramic filter from Murata, or equivalent part
	Leaded type	SFVLA10M7LF00	B _{3dB} = 80 kHz	optional, ceramic filter from Murata, or equivalent part

- For component values for other frequencies, please refer to the EVB descriptions

5 Package Description

 The device TH71112 is RoHS compliant.

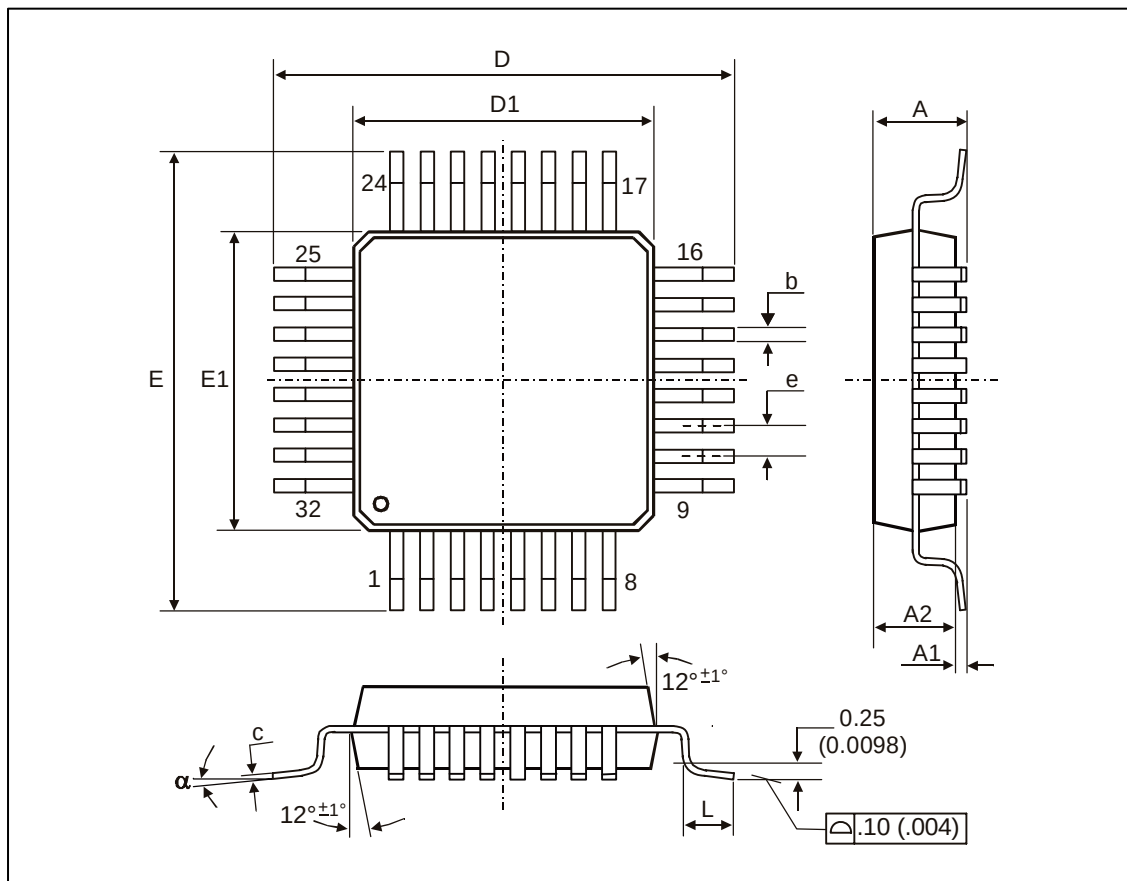


Fig. 6: LQFP32 (Low profile Quad Flat Package)

All Dimension in mm, coplanarity < 0.1mm										
	E1, D1	E, D	A	A1	A2	e	b	c	L	α
min	7.00	9.00	1.40	0.05	1.35	0.8	0.30	0.09	0.45	0°
max			1.60	0.15	1.45		0.45	0.20	0.75	7°
All Dimension in inch, coplanarity < 0.004"										
min	0.276	0.354	0.055	0.002	0.053	0.031	0.012	0.0035	0.018	0°
max			0.063	0.006	0.057		0.018	0.0079	0.030	7°

5.1 Soldering Information

- The device TH71112 is qualified for MSL3 with soldering peak temperature 260 deg C according to JEDEC J-STD-20

6 Reliability Information

This Melexis device is classified and qualified regarding soldering technology, solderability and moisture sensitivity level, as defined in this specification, according to following test methods:

Reflow Soldering SMD's (Surface Mount Devices)

- IPC/JEDEC J-STD-020
"Moisture/Reflow Sensitivity Classification for Nonhermetic Solid State Surface Mount Devices (classification reflow profiles according to table 5-2)"
- EIA/JEDEC JESD22-A113
"Preconditioning of Nonhermetic Surface Mount Devices Prior to Reliability Testing (reflow profiles according to table 2)"

Wave Soldering SMD's (Surface Mount Devices) and THD's (Through Hole Devices)

- EN60749-20
"Resistance of plastic- encapsulated SMD's to combined effect of moisture and soldering heat"
- EIA/JEDEC JESD22-B106 and EN60749-15
"Resistance to soldering temperature for through-hole mounted devices"

Iron Soldering THD's (Through Hole Devices)

- EN60749-15
"Resistance to soldering temperature for through-hole mounted devices"

Solderability SMD's (Surface Mount Devices) and THD's (Through Hole Devices)

- EIA/JEDEC JESD22-B102 and EN60749-21
"Solderability"

For all soldering technologies deviating from above mentioned standard conditions (regarding peak temperature, temperature gradient, temperature profile etc) additional classification and qualification tests have to be agreed upon with Melexis.

The application of Wave Soldering for SMD's is allowed only after consulting Melexis regarding assurance of adhesive strength between device and board.

Melexis is contributing to global environmental conservation by promoting **lead free** solutions. For more information on qualification of **RoHS** compliant products (RoHS = European directive on the Restriction Of the Use of Certain Hazardous Substances) please visit the quality page on our website:

http://www.melexis.com/quality_leadfree.aspx

7 ESD Precautions

Electronic semiconductor products are sensitive to Electro Static Discharge (ESD). Always observe Electro Static Discharge control procedures whenever handling semiconductor products.

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