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## 1-Gbps to 4.25-Gbps Rate-Selectable Limiting Amplifier

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### FEATURES

- Multirate Operation from 1 Gbps up to 4.25 Gbps
- Loss-of-Signal Detection (LOS)
- Two-Wire Digital Interface
- Digitally Selectable LOS Threshold
- Digitally Selectable Bandwidth
- Digitally Selectable Output Voltage
- Low Power Consumption
- Input Offset Cancellation

- CML Data Outputs With On-Chip, 50- $\Omega$  Back-Termination to  $V_{CC}$
- Single 3.3-V Supply
- Surface-Mount, Small-Footprint, 4-mm  $\times$  4-mm, 16-Terminal QFN Package

### APPLICATIONS

- Multirate SONET/SDH Transmission Systems
- 4.25-Gbps, 2.125-Gbps, and 1.0625-Gbps Fibre-Channel Receivers
- Gigabit Ethernet Receivers

### DESCRIPTION

The ONET4291PA is a versatile, high-speed, rate-selectable limiting amplifier for multiple fiber-optic applications with data rates up to 4.25 Gbps.

The device provides a two-wire interface, which allows digital bandwidth selection, digital output amplitude selection, and digital loss of signal threshold adjust.

This device provides a gain of about 43 dB, which ensures a fully differential output swing for input signals as low as 5 mV<sub>p-p</sub>.

The ONET4291PA provides loss-of-signal detection with either digital or analog threshold adjust.

The part is available in a small-footprint, 4-mm  $\times$  4-mm, 16-terminal QFN package. It requires a single 3.3-V supply.

This power-efficient, rate-selectable limiting amplifier is characterized for operation from  $-40^{\circ}\text{C}$  to  $85^{\circ}\text{C}$  ambient temperature.

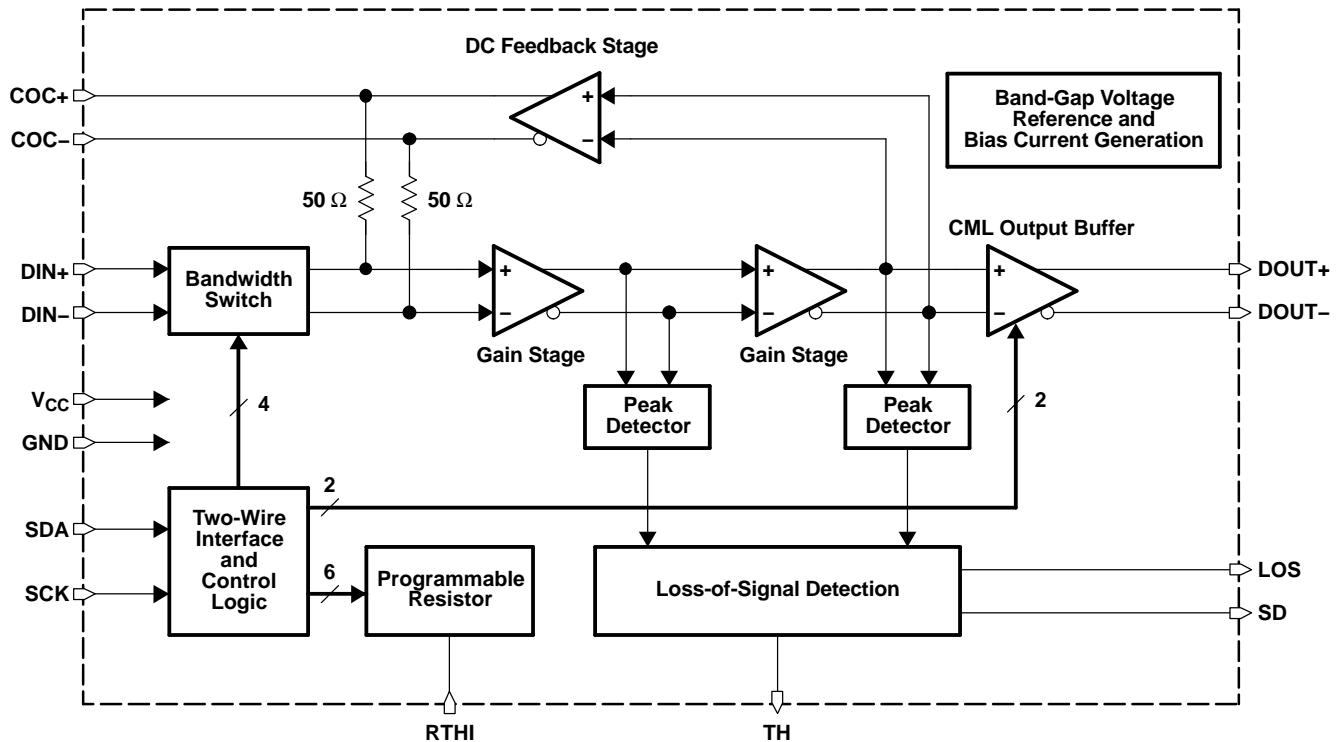


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## BLOCK DIAGRAM

A simplified block diagram of the ONET4291PA is shown in Figure 1.

This compact, 3.3-V, low-power, 1-Gbps to 4.25-Gbps rate-selectable limiting amplifier consists of a high-speed data path with offset cancellation block (dc feedback), a loss-of-signal detection block using two peak detectors, a programmable resistor, a two-wire interface and control-logic block, and a band-gap voltage reference and bias-current generation block.



B0067-01

Figure 1. Simplified Block Diagram of the ONET4291PA

## HIGH-SPEED DATA PATH

The high-speed data signal is applied to the data path by means of the input signal terminals DIN+ and DIN-. The data path consists of a digitally controllable bandwidth switch followed by two 50-Ω on-chip line termination resistors; two gain stages, which provide a typical gain of about 37 dB; and a CML output stage, which provides another 6-dB gain. The amplified data-output signal is available at the output terminals DOUT+ and DOUT-, which feature on-chip  $2 \times 50\text{-}\Omega$  back-termination to  $V_{CC}$ .

A dc feedback stage compensates for internal offset voltages and thus ensures proper operation even for small input data signals. This stage is driven by the output signal of the second gain stage. The signal is low-pass filtered, amplified, and fed back to the input of the first gain stage via the on-chip 50-Ω termination resistors. The required low-frequency cutoff is determined by an external 0.1-μF capacitor, which must be differentially connected to the COC+ and COC- terminals.

## LOSS-OF-SIGNAL DETECTION AND PROGRAMMABLE RESISTOR

The peak values of the output signals of the first and second gain stages are monitored by two peak detectors. The peak values are compared to a predefined loss-of-signal threshold voltage inside the loss-of-signal detection block. As a result of the comparison, the loss-of-signal detection block generates the SD signal, which indicates a sufficient input-signal amplitude, or the LOS signal, which indicates that the input signal amplitude is below the defined threshold level.

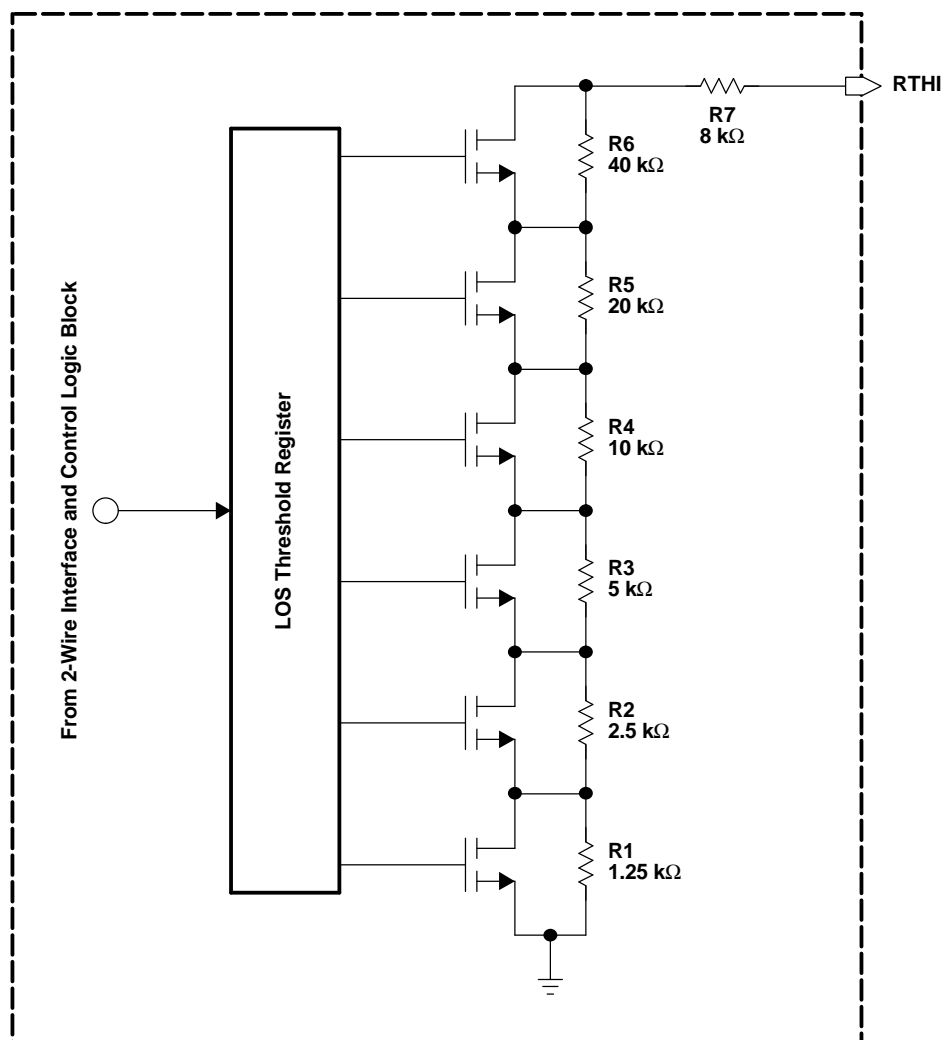
The threshold voltage can be set within a certain range by means of an external resistor connected between the TH terminal and ground (GND). Alternatively, shorting the TH and RTHI terminals causes an internal, digitally selectable resistor to be used for threshold adjustment. The resistor value is selectable using the two-wire interface.

The principle of the digitally selectable resistor is shown in Figure 2. The complete resistor between the RTHI terminal and GND consists of seven series-connected resistors.

Six of the resistors have binary-weighted resistance values, and each can be shunted individually by means of a parallel-connected MOS transistor.

The seventh resistor defines the minimum remaining resistance in case all six MOS devices are conductive.

With the resistor values shown in Figure 2, the minimum selectable resistance is 8 k $\Omega$ , the maximum resistance is 86.75 k $\Omega$ , and the resolution is 1.25 k $\Omega$ /step.



S0098-01

Figure 2. Digitally Controllable On-Chip Resistor

## TWO-WIRE INTERFACE AND CONTROL LOGIC

The ONET4291PA uses a two-wire serial interface for digital control of the amplifier bandwidth, output amplitude, and LOS threshold. A simplified block diagram of this interface is given in [Figure 3](#).

SDA and SCK are inputs for the serial data and the serial clock, respectively, and can be driven by a microprocessor. Both inputs have 100-k $\Omega$  pullup resistors to  $V_{CC}$ . For driving these inputs, an open-drain output is recommended.

A write cycle consists of a START command, 3 address bits with MSB first, 8 data bits with MSB first, and a STOP command. In idle mode, both the SDA and SCK lines are at a high level.

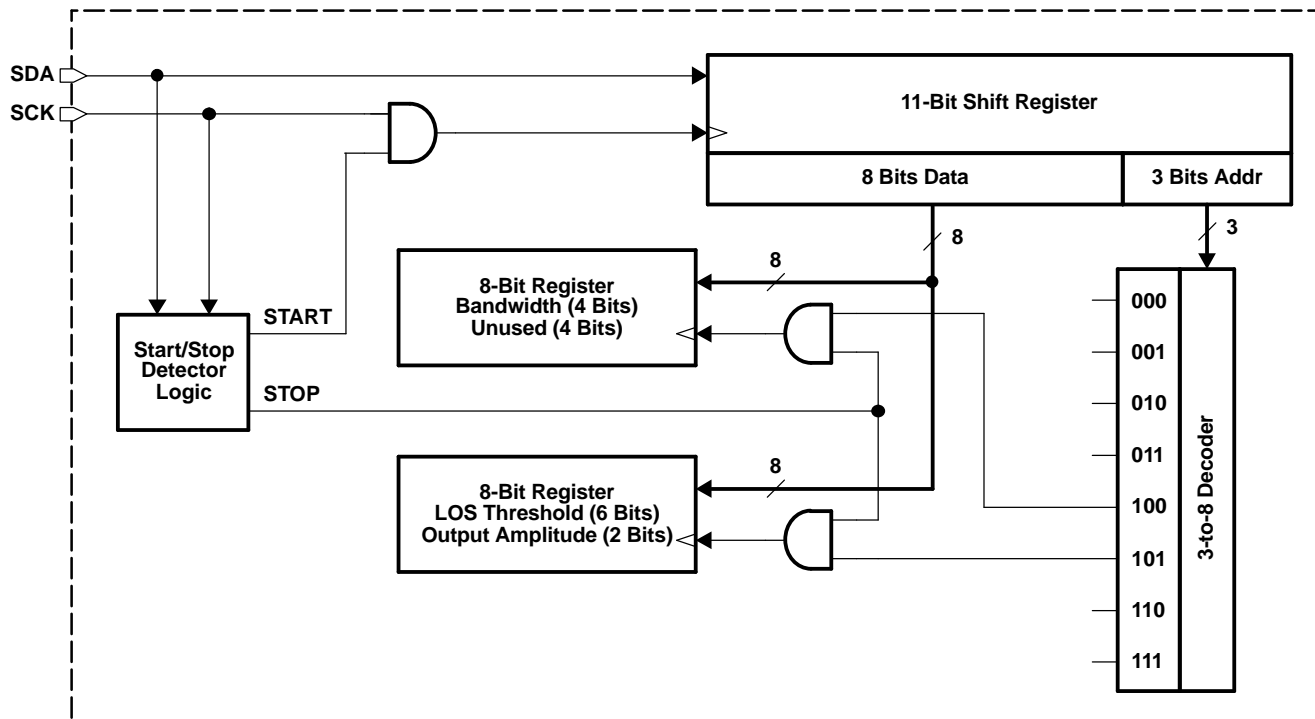
A START command is initiated by a falling edge on SDA with SCK at a high level.

Bits are clocked into an 11-bit-wide shift register while the SCK level is high.

A STOP command is detected on the rising edge of SDA after SCK has changed from a low level to a high level.

At the time of detection of a STOP command, the 8 data bits from the shift register are copied to a selected 8-bit register. Register selection occurs according to the 3 address bits in the shift register, which are decoded to 8 independent select signals using a 3-to-8 decoder block.

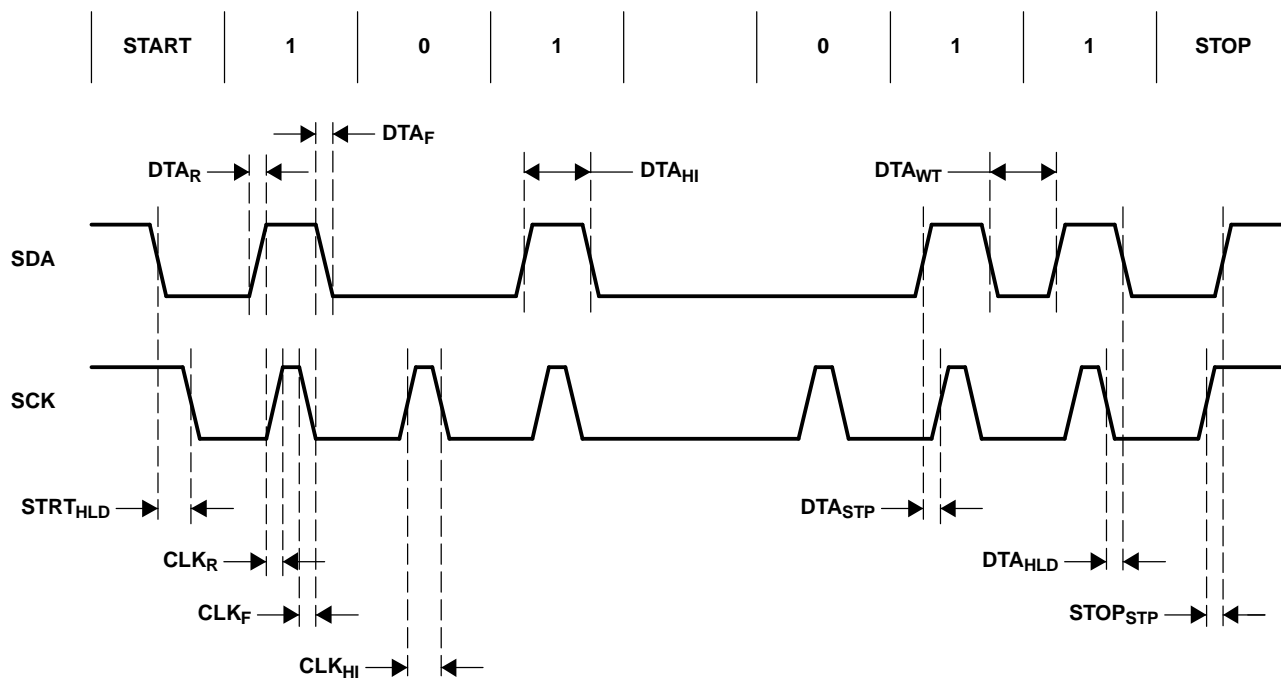
In the ONET4291PA, only addresses 4 (100b) and 5 (101b) are used.



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Figure 3. Simplified Two-Wire Interface Block Diagram

The timing definition for the serial data signal SDA and the serial clock signal SCK is shown in Figure 4.



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PARAMETER	DESCRIPTION	MIN	MAX	UNIT
STRT <sub>HLD</sub>	START hold time	Time required from data falling edge to clock falling edge at START	10	ns
CLK <sub>R</sub> , DTA <sub>R</sub>	Clock and data rise time		10	ns
CLK <sub>F</sub> , DTA <sub>F</sub>	Clock and data fall time		10	ns
CLK <sub>HI</sub>	Clock high time	Minimum clock high period	50	ns
DTA <sub>HI</sub>	Data high time	Minimum data high period	100	ns
DTA <sub>STP</sub>	Data setup time	Minimum time from data rising edge to clock rising edge	10	ns
DTA <sub>WT</sub>	Data wait time	Minimum time from data falling edge to data rising edge	50	ns
DTA <sub>HLD</sub>	Data hold time	Minimum time from clock falling edge to data falling edge	10	ns
STOP <sub>STP</sub>	STOP setup time	Minimum time from clock rising edge to data rising edge at STOP	10	ns

Figure 4. Two-Wire Interface Timing Diagram

The register mapping for register addresses 4 (100b) and 5 (101b) is shown in Table 1 and Table 2, respectively.

Table 1. Register 4 (100b) Mapping

BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
BW3	BW2	BW1	BW0	–	–	–	–

Table 2. Register 5 (101b) Mapping

BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
A1	A0	R5	R4	R3	R2	R1	R0

Table 3 through Table 5 describe circuit functionality based on the register settings.

**Table 3. Bandwidth Selection**

BW3	BW2	BW1	BW0	TYP	UNIT
0	0	0	0	4.39	GHz
0	0	0	1	3.91	GHz
0	0	1	0	3.47	GHz
0	0	1	1	3.03	GHz
0	1	0	0	2.81	GHz
0	1	0	1	2.31	GHz
0	1	1	0	1.82	GHz
0	1	1	1	1.60	GHz
1	0	0	0	1.55	GHz
1	0	0	1	1.33	GHz
1	0	1	0	1.11	GHz
1	0	1	1	1.03	GHz
1	1	0	0	0.86	GHz
1	1	0	1	0.82	GHz
1	1	1	0	0.76	GHz
1	1	1	1	0.73	GHz

**Table 4. Output Amplitude Selection**

A1	A0	TYP	UNIT
0	0	400	mV <sub>p-p</sub>
0	1	600	mV <sub>p-p</sub>
1	0	800	mV <sub>p-p</sub>
1	1	1000	mV <sub>p-p</sub>

**Table 5. LOS-Threshold Digitally Controlled Resistor Selection**

R5	R4	R3	R2	R1	R0	TYP	UNIT
0	0	0	0	0	0	86.75	kΩ
0	0	0	0	0	1	85.5	kΩ
0	0	0	0	1	0	84.25	kΩ
0	0	0	0	1	1	83	kΩ
0	0	0	1	0	0	81.75	kΩ
0	0	0	1	0	1	80.5	kΩ
0	0	0	1	1	0	79.25	kΩ
0	0	0	1	1	1	78	kΩ
0	0	1	0	0	0	76.75	kΩ
0	0	1	0	0	1	75.5	kΩ
0	0	1	0	1	0	74.25	kΩ
0	0	1	0	1	1	73	kΩ
0	0	1	1	0	0	71.75	kΩ
0	0	1	1	0	1	70.5	kΩ
0	0	1	1	1	0	69.25	kΩ
0	0	1	1	1	1	68	kΩ
0	1	0	0	0	0	66.75	kΩ
0	1	0	0	0	1	65.5	kΩ
0	1	0	0	1	0	64.25	kΩ
0	1	0	0	1	1	63	kΩ
0	1	0	1	0	0	61.75	kΩ

**Table 5. LOS-Threshold Digitally Controlled Resistor Selection (continued)**

R5	R4	R3	R2	R1	R0	TYP	UNIT
0	1	0	1	0	1	60.5	kΩ
0	1	0	1	1	0	59.25	kΩ
0	1	0	1	1	1	58	kΩ
0	1	1	0	0	0	56.75	kΩ
0	1	1	0	0	1	55.5	kΩ
0	1	1	0	1	0	54.25	kΩ
0	1	1	0	1	1	53	kΩ
0	1	1	1	0	0	51.75	kΩ
0	1	1	1	0	1	50.5	kΩ
0	1	1	1	1	0	49.25	kΩ
0	1	1	1	1	1	48	kΩ
1	0	0	0	0	0	46.75	kΩ
1	0	0	0	0	1	45.5	kΩ
1	0	0	0	1	0	44.25	kΩ
1	0	0	0	1	1	43	kΩ
1	0	0	1	0	0	41.75	kΩ
1	0	0	1	0	1	40.5	kΩ
1	0	0	1	1	0	39.25	kΩ
1	0	0	1	1	1	38	kΩ
1	0	1	0	0	0	36.75	kΩ
1	0	1	0	0	1	35.5	kΩ
1	0	1	0	1	0	34.25	kΩ
1	0	1	0	1	1	33	kΩ
1	0	1	1	0	0	31.75	kΩ
1	0	1	1	0	1	30.5	kΩ
1	0	1	1	1	0	29.25	kΩ
1	0	1	1	1	1	28	kΩ
1	1	0	0	0	0	26.75	kΩ
1	1	0	0	0	1	25.5	kΩ
1	1	0	0	1	0	24.25	kΩ
1	1	0	0	1	1	23	kΩ
1	1	0	1	0	0	21.75	kΩ
1	1	0	1	0	1	20.5	kΩ
1	1	0	1	1	0	19.25	kΩ
1	1	0	1	1	1	18	kΩ
1	1	1	0	0	0	16.75	kΩ
1	1	1	0	0	1	15.5	kΩ
1	1	1	0	1	0	14.25	kΩ
1	1	1	0	1	1	13	kΩ
1	1	1	1	0	0	11.75	kΩ
1	1	1	1	0	1	10.5	kΩ
1	1	1	1	1	0	9.25	kΩ
1	1	1	1	1	1	8	kΩ

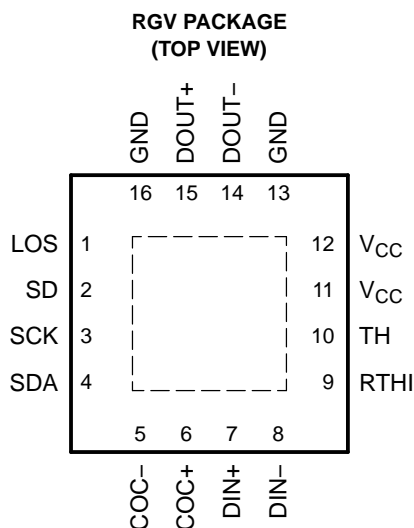
## BAND-GAP VOLTAGE AND BIAS GENERATION

The ONET4291PA limiting amplifier is supplied by a single, 3.3-V supply voltage connected to the  $V_{CC}$  terminals. This voltage is referred to GND.

On-chip band-gap voltage circuitry generates a reference voltage, independent of supply voltage, from which all other internally required voltages and bias currents are derived.

## TERMINAL ASSIGNMENTS

For the ONET4291PA, a small-footprint 4-mm × 4-mm, 16-terminal QFN package is used, with a terminal pitch of 0,65 mm.



P0030-01

## TERMINAL DESCRIPTION

TERMINAL		TYPE	DESCRIPTION
NAME	NO.		
COC+	6	Analog	Offset cancellation filter capacitor plus terminal. An external 0.1-μF filter capacitor must be connected between this terminal and COC– (terminal 5).
COC–	5	Analog	Offset cancellation filter capacitor minus terminal. An external 0.1-μF filter capacitor must be connected between this terminal and COC+ (terminal 6).
DIN+	7	Analog input	Non-inverted data input. On-chip 50-Ω terminated to COC+. Differentially 100-Ω terminated to DIN–.
DIN–	8	Analog input	Inverted data input. On-chip 50-Ω terminated to COC–. Differentially 100-Ω terminated to DIN+.
DOUT+	15	CML output	Non-inverted data output. On-chip 50-Ω back-terminated to $V_{CC}$ .
DOUT–	14	CML output	Inverted data output. On-chip 50-Ω back-terminated to $V_{CC}$ .
GND	13, 16, EP	Supply	Circuit ground. Exposed die pad (EP) must be grounded.
LOS	1	Open-drain MOS	High level indicates that the input signal amplitude is below the programmed threshold level. Open-drain output. Requires an external 10-kΩ pullup resistor to $V_{CC}$ for proper operation.
RTHI	9	Analog	Digitally controlled internal resistor to ground, which can be used for LOS threshold adjustment. A 6-bit-wide control register can be set via the two-wire interface.
SCK	3	CMOS input	Two-wire interface serial clock. Includes a 100-kΩ pullup resistor to $V_{CC}$ .
SD	2	CMOS output	High level indicates that sufficient input signal amplitude is applied to the device. Low level indicates that the input signal amplitude is below the programmed threshold level.
SDA	4	CMOS input	Two-wire interface serial data input. Includes a 100-kΩ pullup resistor to $V_{CC}$ .
TH	10	Analog input	LOS threshold adjustment with resistor to GND. For use of the internal digitally controlled resistor, connect TH with RTHI (terminal 9).
$V_{CC}$	11, 12	Supply	3.3-V, +10%/–12% supply voltage



## ABSOLUTE MAXIMUM RATINGS

over operating free-air temperature range (unless otherwise noted) <sup>(1)</sup>

$V_{CC}$	Supply voltage <sup>(2)</sup>	–0.3 V to 4 V
$V_{DIN+}$ , $V_{DIN-}$	Voltage at DIN+, DIN– <sup>(2)</sup>	0.5 V to 4 V
$V_{LOS}$ , $V_{SD}$ , $V_{SCK}$ , $V_{SDA}$ , $V_{COC+}$ , $V_{COC-}$ , $V_{RTHI}$ , $V_{TH}$ , $V_{DOUT+}$ , $V_{DOUT-}$	Voltage at LOS, SD, SCK, SDA, COC+, COC–, RTHI, TH, DOUT+, DOUT– <sup>(2)</sup>	–0.3 V to 4 V
$V_{DIN,DIFF}$	Differential voltage between DIN+ and DIN–	±1.25 V
$I_{LOS}$	Current into LOS	10 mA
$I_{DIN+}$ , $I_{DIN-}$ , $I_{DOUT+}$ , $I_{DOUT-}$	Continuous current at inputs and outputs	20 mA
ESD	ESD rating at all terminals (HBM)	4 kV
$T_{J,max}$	Maximum junction temperature	125°C
$T_{stg}$	Storage temperature range	–65°C to 85°C
$T_A$	Characterized free-air operating temperature range	–40°C to 85°C
$T_{LEAD}$	Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

- (1) Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
- (2) All voltage values are with respect to network ground terminal.

## RECOMMENDED OPERATING CONDITIONS

over operating free-air temperature range (unless otherwise noted)

		MIN	NOM	MAX	UNIT
$V_{CC}$ , $V_{CCO}$	Supply voltage	2.9	3.3	3.6	V
$T_A$	Operating free-air temperature	–40		85	°C
	CMOS input high voltage	2			V
	CMOS input low voltage			0.8	V

## DC ELECTRICAL CHARACTERISTICS

over recommended operating conditions (unless otherwise noted). Typical values are at  $V_{CC} = 3.3$  V and  $T_A = 25^\circ\text{C}$ .

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
$V_{CC}$ , $V_{CCO}$	Supply voltage		2.9	3.3	3.6	V
$I_{VCC}$	Supply current <sup>(1)</sup>	$V_{OD} = 1000$ mV <sub>p-p</sub> , maximum bandwidth selected	35	50	64	mA
		$V_{OD} = 800$ mV <sub>p-p</sub> , maximum bandwidth selected	32	46	59	
		$V_{OD} = 600$ mV <sub>p-p</sub> , maximum bandwidth selected	28	41	53	
		$V_{OD} = 400$ mV <sub>p-p</sub> , maximum bandwidth selected	24	36	48	
$R_{IN}$ , $R_{OUT}$	Data input/output resistance	Single-ended		50		Ω
	CMOS output high voltage	$I_{SINK} = 1$ mA	2.3			V
	CMOS output low voltage	$I_{SOURCE} = 1$ mA			0.5	V
	LOS low voltage	$I_{SOURCE} = 1.5$ mA			0.5	V
	Optimum LOS threshold resistor		12		62	kΩ

- (1) Use of the bandwidth select switch increases current consumption. The MSB bandwidth-select bit, BW3, typically consumes 5 mA, BW2 2.6 mA, BW1 1.3 mA, and BW0 0.7 mA.

**AC ELECTRICAL CHARACTERISTICS**

over recommended operating conditions (unless otherwise noted). Typical operating condition is at  $V_{CC} = 3.3\text{ V}$  and  $T_A = 25^\circ\text{C}$ .

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
$f_{3dB-H}$	High-frequency –3-dB bandwidth	Maximum bandwidth selected (BW3 = BW2 = BW1 = BW0 = 0)	3.5	4.5	6	GHz
		Minimum bandwidth selected (BW3 = BW2 = BW1 = BW0 = 1)	0.7			
$f_{3dB-L}$	Low-frequency –3-dB bandwidth	$C_{OC} = 0.1\text{ }\mu\text{F}$		23	50	kHz
	Data rate	Maximum bandwidth selected (BW3 = BW2 = BW1 = BW0 = 0)	4.25			Gbps
$V_{IN,MIN}$	Data input sensitivity	K28.5 at 4.25 Gbps, BER < $10^{-12}$ (noise limited)		1.9	2.7	mV <sub>p-p</sub>
		$V_{OD-min} \geq 0.95 * V_{OD}$ (at $V_{IN} = 25\text{ mV}_{p-p}$ ) (gain limited)		8	14	
A	Small-signal gain		38	43	46	dB
	Small-signal gain vs temperature				2.5	dB
	Small-signal gain vs supply voltage $V_{CC}$				1	dB
$V_{IN,MAX}$	Data input overload		2000			mV <sub>p-p</sub>
DJ	Deterministic jitter	$V_{IN} = 5\text{ mV}_{p-p}$ , K28.5 at 4.25 Gbps, maximum bandwidth		10	18	ps <sub>p-p</sub>
		$V_{IN} = 10\text{ mV}_{p-p}$ , K28.5 at 4.25 Gbps, maximum bandwidth		9	17	
		$V_{IN} = 25\text{ mV}_{p-p}$ , K28.5 at 4.25 Gbps, maximum bandwidth		8	15	
RJ	Random jitter	Input = $5\text{ mV}_{p-p}$ , maximum bandwidth		3		ps <sub>RMS</sub>
		Input = $10\text{ mV}_{p-p}$ , maximum bandwidth		1.5		
$V_{OD}$	Differential-data output voltage	800-mV output amplitude selected (default), $V_{IN} > 25\text{ mV}_{p-p}$	700	850	1000	mV <sub>p-p</sub>
$t_R$	Output rise time	20% to 80%, $V_{IN} > 25\text{ mV}_{p-p}$ , maximum bandwidth		45	90	ps
$t_F$	Output fall time	20% to 80%, $V_{IN} > 25\text{ mV}_{p-p}$ , maximum bandwidth		45	90	ps
$V_{TH}$	LOS assert threshold range	K28.5 pattern at 4.25 Gbps, $R_{TH} = 62\text{ k}\Omega$		5.5		mV <sub>p-p</sub>
		K28.5 pattern at 4.25 Gbps, $R_{TH} = 12\text{ k}\Omega$		30		
	LOS threshold variation vs temperature			1		dB
	LOS threshold variation vs supply voltage $V_{CC}$				1.5	dB
	LOS hysteresis	K28.5 pattern at 4.25 Gbps	2		7.4	dB
$T_{LOS\_AST}$	LOS assert time		400		1500	ns
$T_{LOS\_DEA}$	LOS deassert time		15		80	ns

## TYPICAL CHARACTERISTICS

Typical operating condition is at  $V_{CC} = 3.3\text{ V}$  and  $T_A = 25^\circ\text{C}$ .

**FREQUENCY RESPONSE FOR  
DIFFERENT BANDWIDTH SETTINGS**

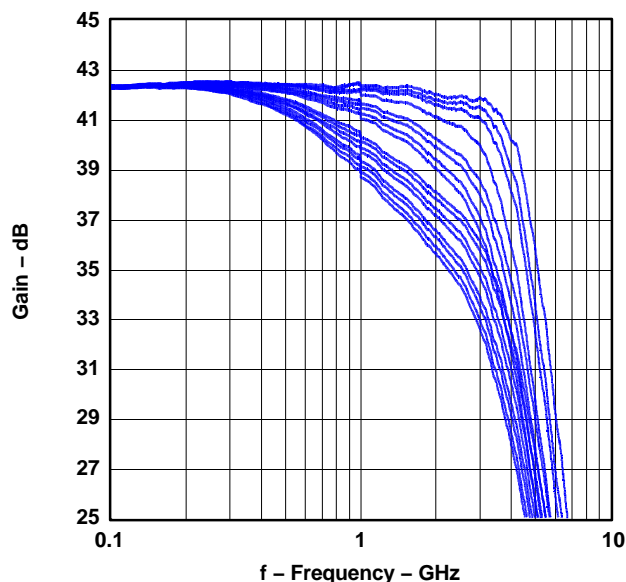


Figure 5.

**BANDWIDTH  
vs  
REGISTER-4 SETTING**

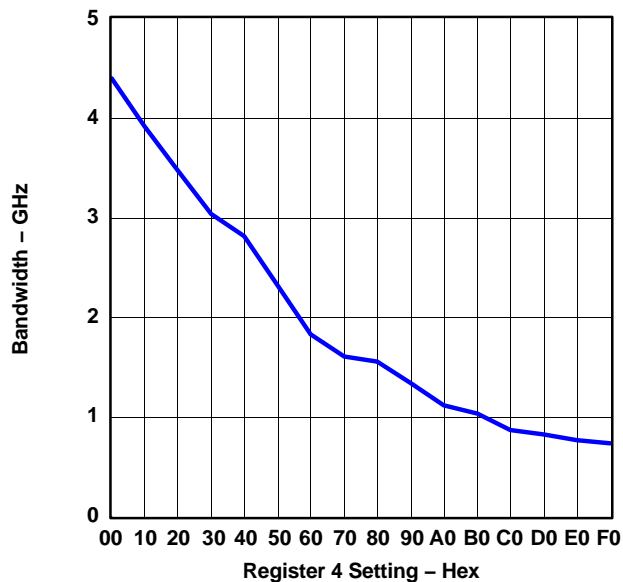


Figure 6.

**DIFFERENTIAL INPUT RETURN GAIN  
vs  
FREQUENCY (MAXIMUM BANDWIDTH)**

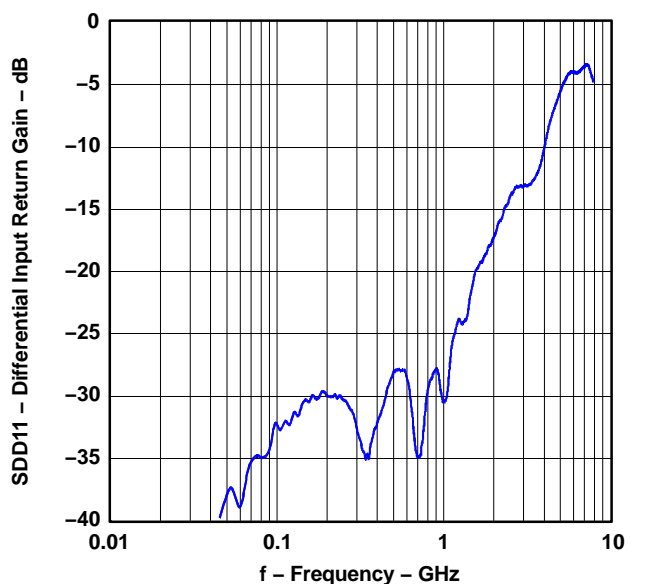


Figure 7.

**RANDOM JITTER  
vs  
INPUT AMPLITUDE (4.25 Gbps, MAXIMUM BANDWIDTH)**

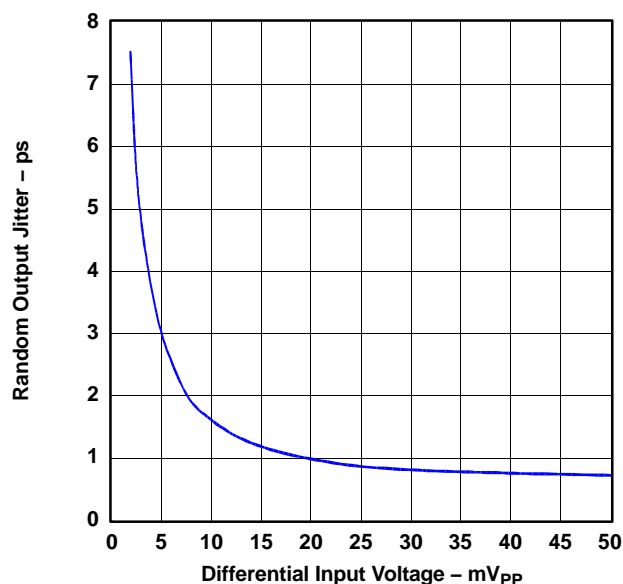


Figure 8.

**TYPICAL CHARACTERISTICS (continued)**

Typical operating condition is at  $V_{CC} = 3.3\text{ V}$  and  $T_A = 25^\circ\text{C}$ .

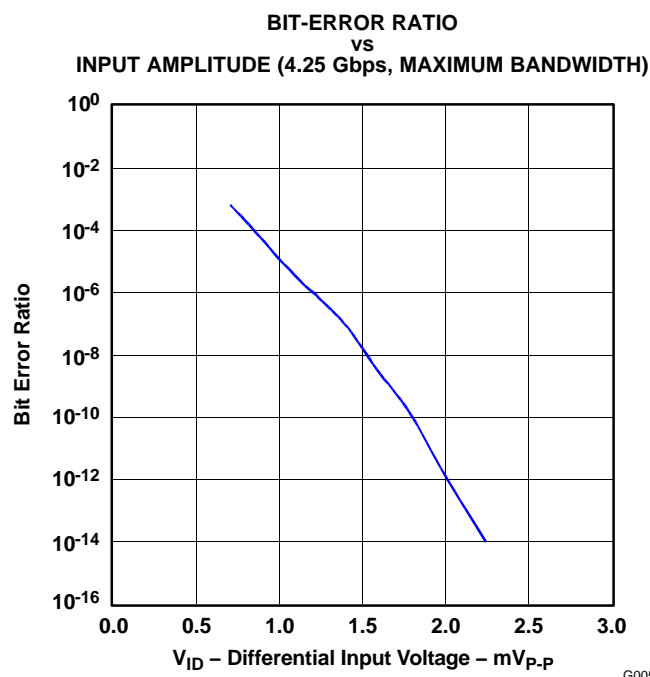


Figure 9.

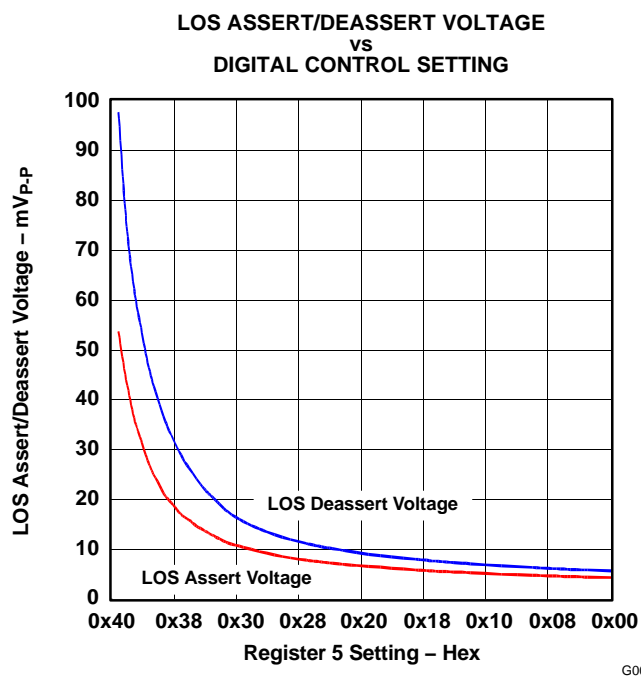


Figure 10.

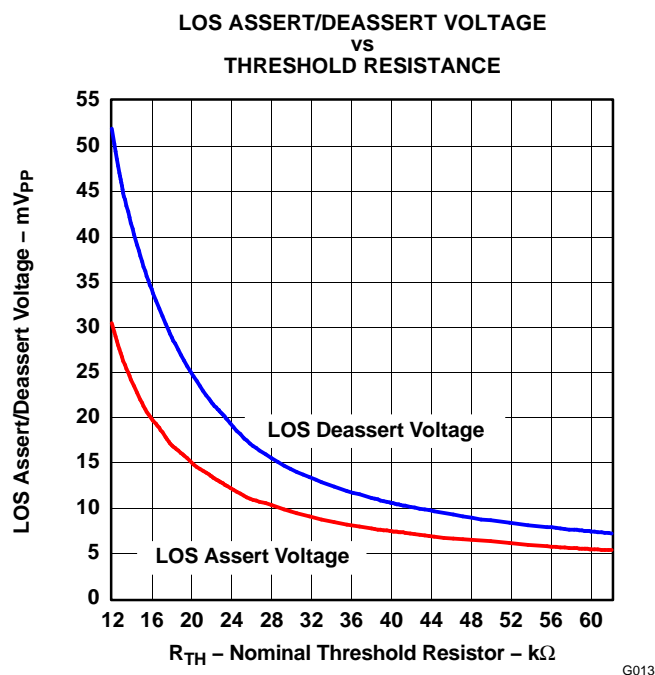


Figure 11.

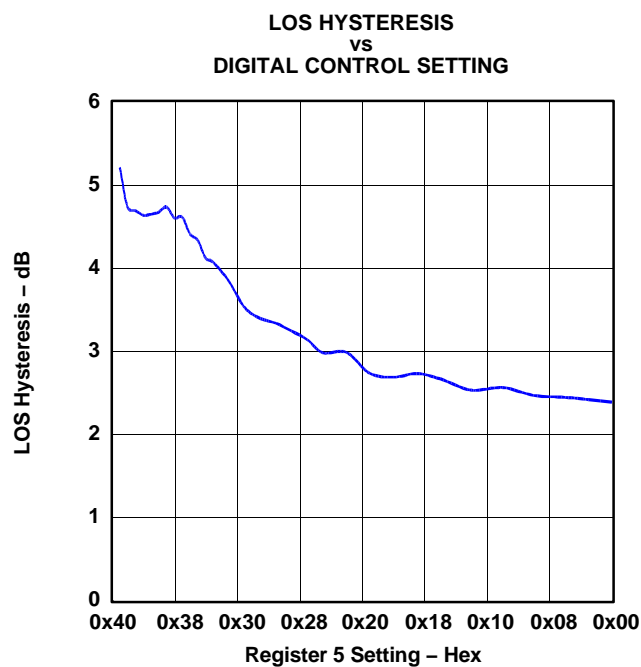


Figure 12.

## TYPICAL CHARACTERISTICS (continued)

Typical operating condition is at  $V_{CC} = 3.3\text{ V}$  and  $T_A = 25^\circ\text{C}$ .

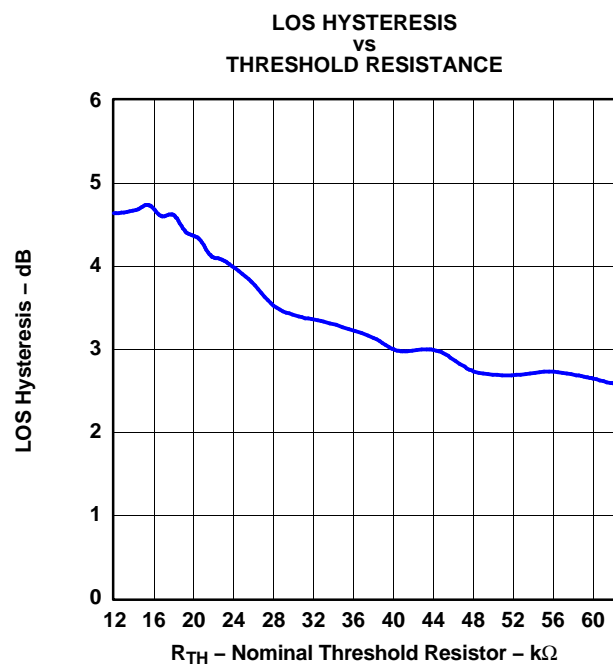


Figure 13.

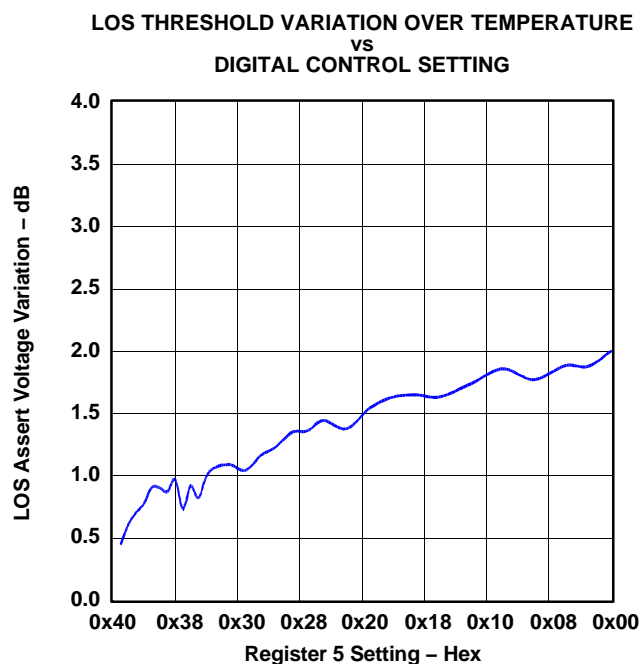


Figure 14.

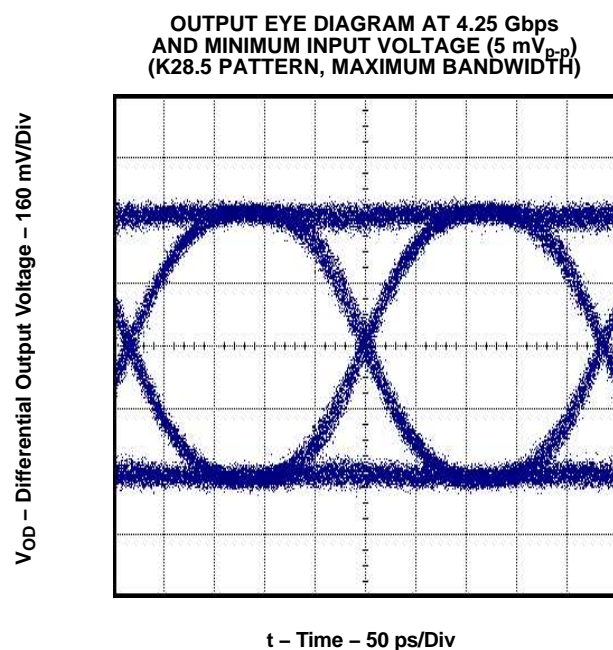


Figure 15.

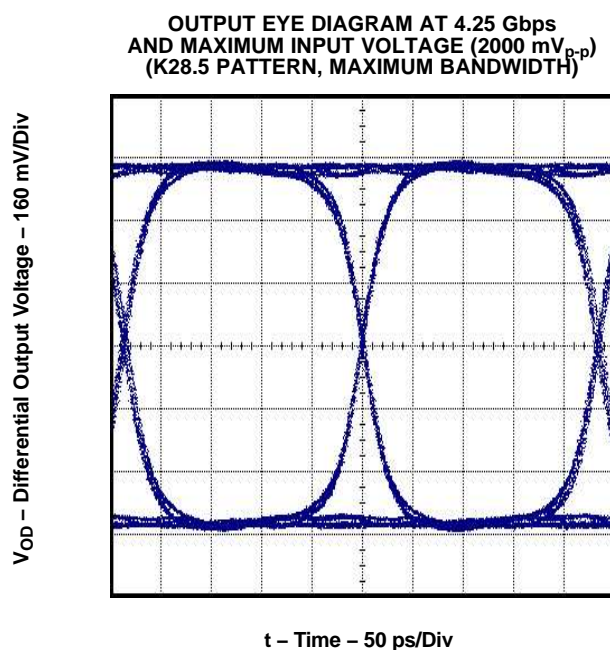


Figure 16.

**TYPICAL CHARACTERISTICS (continued)**

Typical operating condition is at  $V_{CC} = 3.3\text{ V}$  and  $T_A = 25^\circ\text{C}$ .

**OUTPUT EYE DIAGRAM AT 1.0625 Gbps  
AND MINIMUM INPUT VOLTAGE (5 mV<sub>p-p</sub>)  
(K28.5 PATTERN, REGISTER 4 SET TO 0x70)**

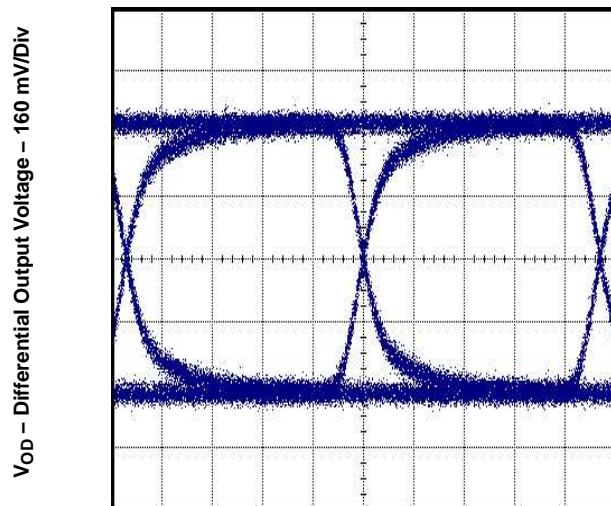


Figure 17.

G011

**OUTPUT EYE DIAGRAM AT 1.0625 Gbps  
AND MAXIMUM INPUT VOLTAGE (2000 mV<sub>p-p</sub>)  
(K28.5 PATTERN, REGISTER 4 SET TO 0x70)**

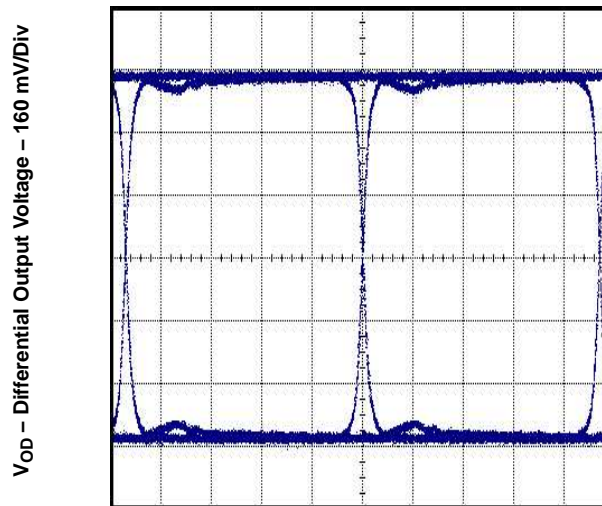
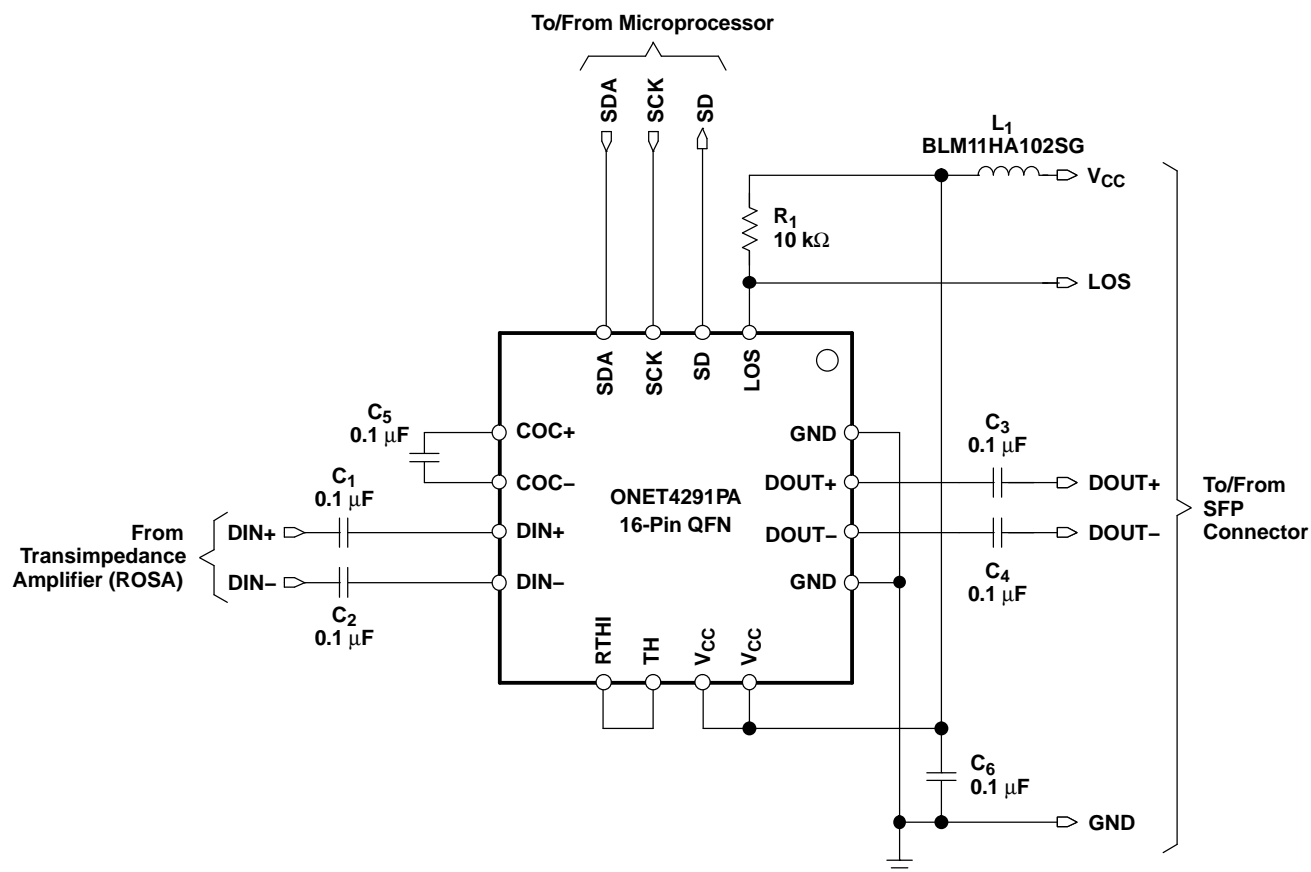


Figure 18.

G012

## APPLICATION INFORMATION

Figure 19 shows a typical application circuit using the ONET4291PA with a microprocessor for digital control of the LOS threshold, output amplitude, and bandwidth.

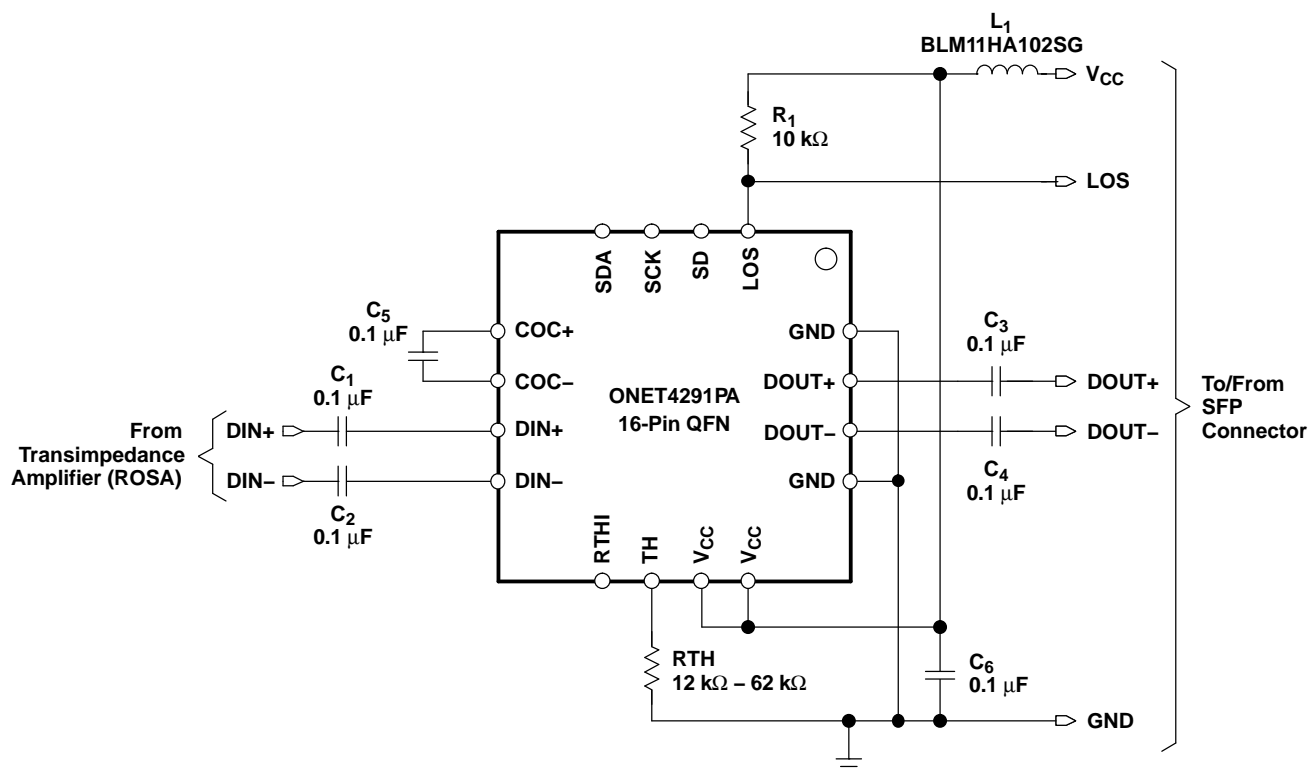


S0099-01

Figure 19. Basic Application Circuit With Digital Control

**APPLICATION INFORMATION (continued)**

Figure 20 shows a typical application without digital control. In this case, the output amplitude and bandwidth are fixed. The LOS threshold is adjusted by means of a resistor connected to the TH terminal.



S0099-02

**Figure 20. Basic Application Circuit With External LOS Threshold Resistor**



## PACKAGING INFORMATION

Orderable Device	Status <sup>(1)</sup>	Package Type	Package Drawing	Pins	Package Qty	Eco Plan <sup>(2)</sup>	Lead/Ball Finish	MSL Peak Temp <sup>(3)</sup>
ONET4291PARGVR	ACTIVE	QFN	RGV	16	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
ONET4291PARGVRG4	ACTIVE	QFN	RGV	16	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
ONET4291PARGVT	ACTIVE	QFN	RGV	16	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
ONET4291PARGVTG4	ACTIVE	QFN	RGV	16	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR

<sup>(1)</sup> The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

<sup>(2)</sup> Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check <http://www.ti.com/productcontent> for the latest availability information and additional product content details.

**TBD:** The Pb-Free/Green conversion plan has not been defined.

**Pb-Free (RoHS):** TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

**Pb-Free (RoHS Exempt):** This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

**Green (RoHS & no Sb/Br):** TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

<sup>(3)</sup> MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

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**TAPE AND REEL INFORMATION**



\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
ONET4291PARGVR	QFN	RGV	16	2500	330.0	12.4	4.3	4.3	1.5	8.0	12.0	Q2
ONET4291PARGVT	QFN	RGV	16	250	330.0	12.4	4.3	4.3	1.5	8.0	12.0	Q2

## TAPE AND REEL BOX DIMENSIONS

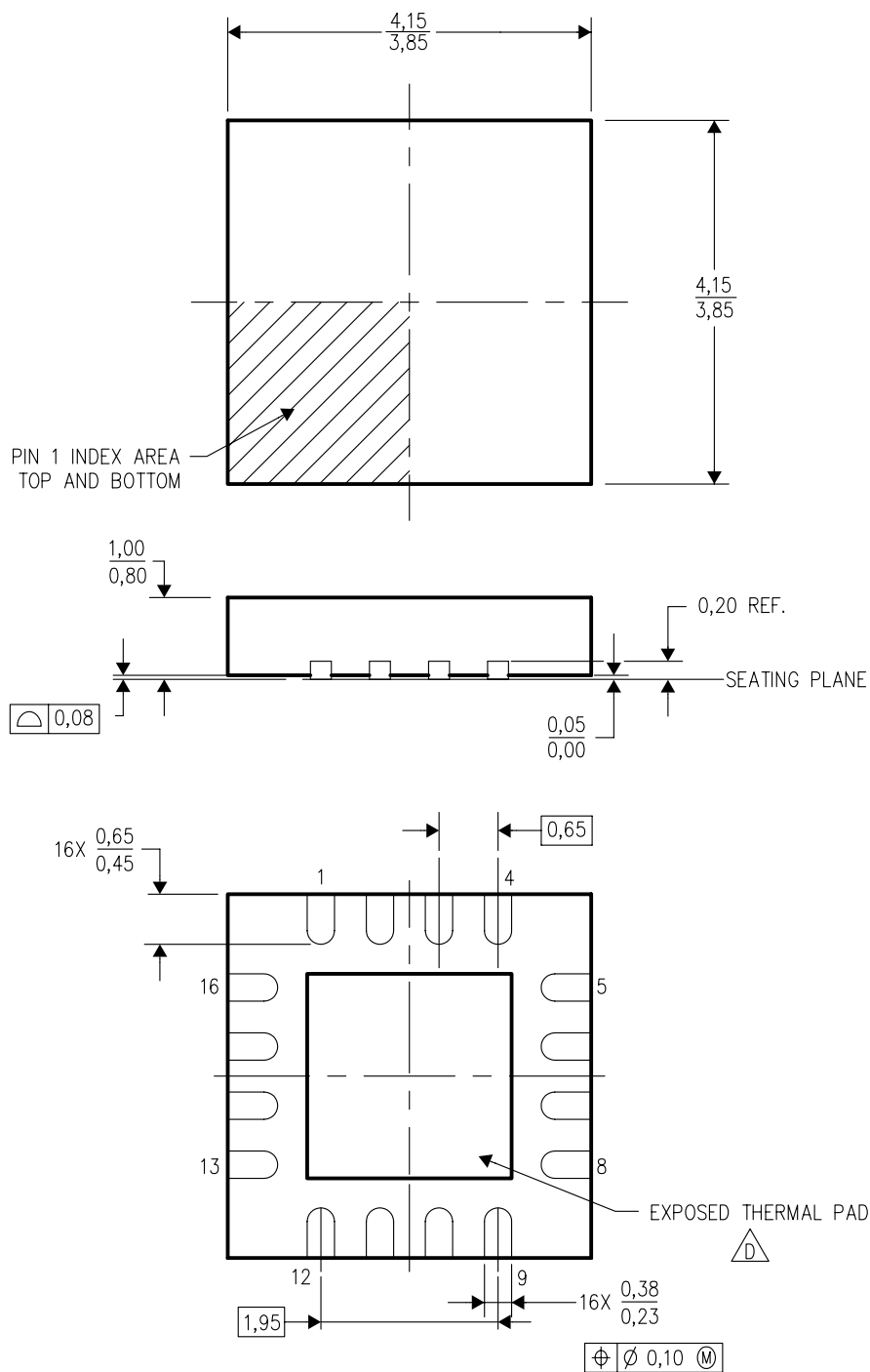


\*All dimensions are nominal


Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
ONET4291PARGVR	QFN	RGV	16	2500	340.5	333.0	20.6
ONET4291PARGVT	QFN	RGV	16	250	340.5	333.0	20.6

## RGV (S-PQFP-N16)

## PLASTIC QUAD FLATPACK



4203497/D 11/04

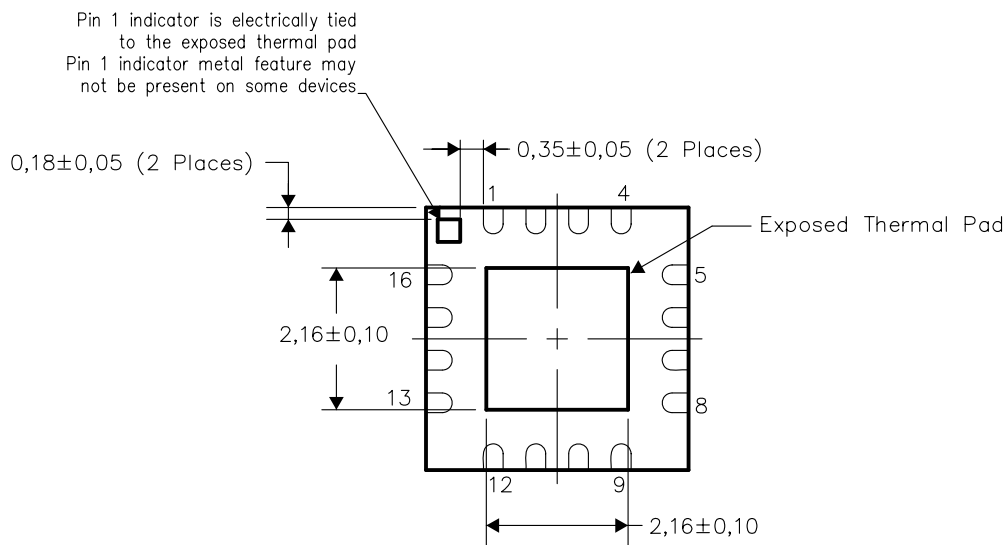
- NOTES:
- A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M-1994.
  - B. This drawing is subject to change without notice.
  - C. Quad Flatpack, No-leads (QFN) package configuration.
  -  D. The package thermal pad must be soldered to the board for thermal and mechanical performance. See the Product Data Sheet for details regarding the exposed thermal pad dimensions.
  - E. Falls within JEDEC MO-220.

## THERMAL INFORMATION

This package incorporates an exposed thermal pad that is designed to be attached directly to an external heatsink. The thermal pad must be soldered directly to the printed circuit board (PCB). After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For information on the Quad Flatpack No-Lead (QFN) package and its advantages, refer to Application Report, Quad Flatpack No-Lead Logic Packages, Texas Instruments Literature No. SCBA017. This document is available at [www.ti.com](http://www.ti.com).

The exposed thermal pad dimensions for this package are shown in the following illustration.

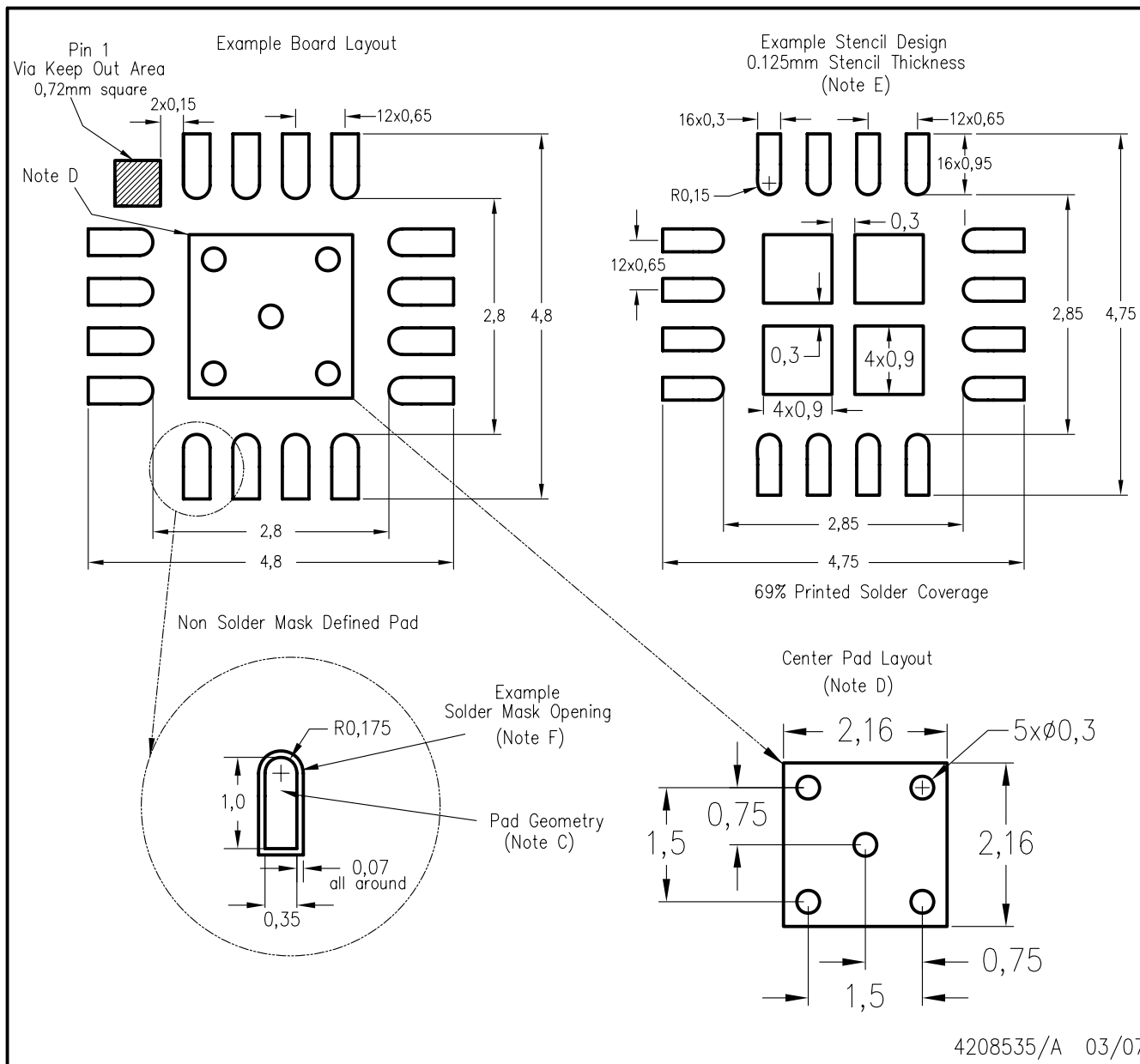


Bottom View

NOTE: All linear dimensions are in millimeters

## Exposed Thermal Pad Dimensions

# RGV (S-PQFP-N16)



4208535/A 03/07

- NOTES:
- All linear dimensions are in millimeters.
  - This drawing is subject to change without notice.
  - Publication IPC-7351 is recommended for alternate designs.
  - This package is designed to be soldered to a thermal pad on the board. Refer to Application Note, QFN Packages, Texas Instruments Literature No. SCBA017, SLUA271, and also the Product Data Sheets for specific thermal information, via requirements, and recommended board layout. These documents are available at [www.ti.com](http://www.ti.com) <<http://www.ti.com>>.
  - Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC 7525 for stencil design considerations.
  - Customers should contact their board fabrication site for solder mask tolerances.

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