

LOW-POWER DISSIPATION ADSL LINE DRIVER

FEATURES

- **Low-Power Dissipation Increases ADSL Line Card Density**
- **Low THD of -88 dBc (100 Ω , 1 MHz)**
- **Low MTPR Driving +20 dBm on the Line**
 - -76 dBc With High Bias Setting
 - -74 dBc With Low Bias Setting
- **Wide Output Swing of 44 V_{PP} Differential Into a 200- Ω Differential Load (V_{CC} = \pm 12 V)**
- **High Output Current of 600 mA (Typ)**
- **Wide Supply Voltage Range of \pm 5 V to \pm 15 V**
- **Pin Compatible with EL1503C and EL1508C**
 - Multiple Package Options
- **Multiple Power Control Modes**
 - 11 mA/ch Full Bias Mode
 - 7.5 mA/ch Mid Bias Mode
 - 4 mA/ch Low Bias Mode
 - 0.25 mA/ch Shutdown Mode
 - I_{ADJ} Pin for User Controlled Bias Current
 - Stable Operation Down to 1.8 mA/ch
- **Low Noise for Increased Receiver Sensitivity**
 - 3.2 nV/ $\sqrt{\text{Hz}}$ Voltage Noise
 - 1.5 pA/ $\sqrt{\text{Hz}}$ Noninverting Current Noise
 - 10 pA/ $\sqrt{\text{Hz}}$ Inverting Current Noise

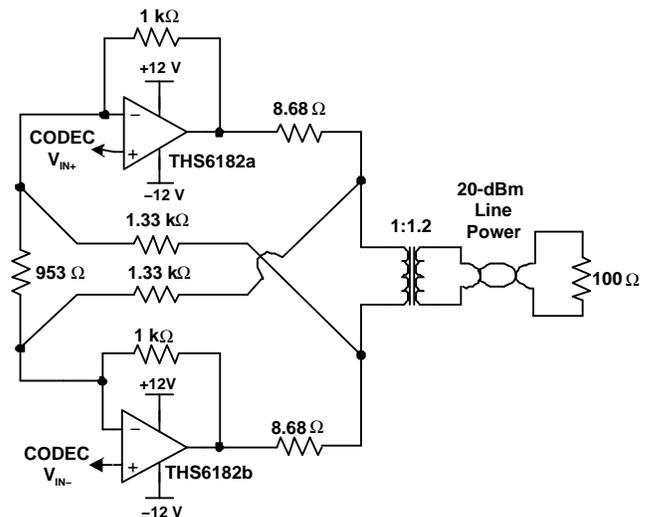
APPLICATIONS

- **Ideal for Full Rate ADSL Applications**

DESCRIPTION

The THS6182 is a current feedback differential line driver ideal for full rate ADSL systems. Its extremely low-power dissipation is ideal for ADSL systems that must achieve high densities in ADSL central office rack applications. The unique architecture of the THS6182 allows the quiescent current to be much lower than existing line drivers while still achieving high linearity without the need for excess open loop gain. Fixed multiple bias settings of the amplifiers allow for enhanced power savings for line lengths where the full performance of the amplifier is not required. To allow for even more flexibility and power savings, an I_{ADJ} pin is available to further lower the bias currents while maintaining stable operation with as little as 1.8 mA per channel. The wide output swing of 44 V_{pp} differentially with \pm 12-V power supplies allows for more dynamic headroom, keeping distortion at a minimum. With a low 3.2 nV/ $\sqrt{\text{Hz}}$ voltage noise coupled with a low 10 pA/ $\sqrt{\text{Hz}}$ inverting current noise, the THS6182 increases the sensitivity of the receive signals, allowing for better margins and reach.

TYPICAL ADSL CO-LINE DRIVER CIRCUIT USING ACTIVE IMPEDANCE



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

PowerPAD is a trademark of Texas Instruments.



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

ORDERING INFORMATION

PRODUCT	PACKAGE	PACKAGE CODE	SYMBOL	ORDER NUMBER	TRANSPORT MEDIA
THS6182RHF	Leadless 24-pin 4, mm x 5, mm PowerPAD™	RHF-24	6182	THS6182RHFR	Tape and reel (3000 devices)
				THS6182RHFT	Tape and reel (250 devices)
THS6182D	SOIC-16	D-16	THS6182	THS6182D	Tube (40 devices)
				THS6832DR	Tape and reel (2500 devices)
THS6182DW	SOIC-20	DW-20	THS6182	THS6182DW	Tube (25 devices)
				THS6182DWR	Tape and reel (2000 devices)
THS6182DWP	SOIC-20 PowerPAD	DWP-20	THS6182	THS6182DWP	Tube (25 devices)
				THS6182DWPR	Tape and reel (2000 devices)

PACKAGE DISSIPATION RATINGS⁽¹⁾

PACKAGE	PowerPAD SOLDERED ⁽²⁾ θ_{JA}	PowerPAD NOT SOLDERED ⁽³⁾ θ_{JC}	θ_{JC}
RHF-24 ⁽²⁾	32°C/W	74°C/W	1.7°C/W
D-16	--	62.9°C/W	25.7°C/W
DW-20	--	45.4°C/W	16.4°C/W
DWP-20 ⁽²⁾	21.5°C/W	43.9°C/W	0.37°C/W

- (1) θ_{JA} values shown are typical for standard test PCBs only.
- (2) For high-power dissipation applications, use of the PowerPAD package with the PowerPad on the underside of the chip. This acts as a heatsink and must be connected to a thermally dissipating plane for proper dissipation. Failure to do so may result in exceeding the maximum junction temperature which could permanently damage and/or reduce the lifetime the device. See TI technical brief SLMA002 for more information about utilizing the PowerPAD thermally enhanced package.
- (3) Use of packages without the PowerPAD or not soldering the PowerPAD to the PCB, should be limited to low-power dissipation applications.

RECOMMENDED OPERATING CONDITIONS

		MIN	NOM	MAX	UNIT
V_{CC+} to V_{CC-} Supply voltage	Dual supply	±5	±12	±15	V
	Single supply	10	24	30	

ABSOLUTE MAXIMUM RATINGS

over operating free-air temperature range unless otherwise noted⁽¹⁾

ELECTRICAL		THS6132
V_{CC}	Supply voltage	± 16.5 V
V_I	Input voltage	$\pm V_{CC}$
I_O	Output current	1000 mA
V_{IO}	Differential input voltage	± 2 V
THERMAL		
T_J	Maximum junction temperature, any condition	150°C
	Maximum junction temperature, continuous operation, long term reliability ⁽²⁾	125°C
T_{stg}	Storage temperature	65°C to 150°C
Lead temperature, 1,6 mm (1/16-inch) from case for 10 seconds		300°C
ESD		
ESD ratings	HBM	2000 V
	CDM	1500 V
	MM	200 V

- (1) The absolute maximum ratings under any condition is limited by the constraints of the silicon process. Stresses above these ratings may cause permanent damage. Exposure to absolute-maximum-rated conditions for extended periods may degrade device reliability. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those specified is not implied.
- (2) The maximum junction temperature for continuous operation is limited by package constraints. Operation above this temperature may result in reduced reliability and/or lifetime of the device.

ELECTRICAL CHARACTERISTICS

over recommended operating free-air temperature range, $T_A = 25^\circ\text{C}$, $V_{CC} = \pm 12$ V, $R_F = 2$ k Ω , Gain = +5, $I_{ADJ} = \text{Bias1} = \text{Bias2} = 0$ V, $R_L = 50$ Ω (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
NOISE/DISTORTION PERFORMANCE						
MTPR	Multitone power ratio	Gain = +9.5, 163 kHz to 1.1 MHz DMT, +20 dBm Line Power, See Figure 1 for circuit		-76		dBc
	Receive band spill-over	Gain = +5, 25 kHz to 138 kHz with MTPR signal applied, See Figure 1 for circuit		-95		dBc
HD	Harmonic distortion, $V_{O(PP)} = 2$ V $f = 1$ MHz	2 nd harmonic	Differential load = 200 Ω	-88		dBc
			Differential load = 50 Ω	-70		
		3 rd harmonic	Differential load = 200 Ω	-107		dBc
			Differential load = 50 Ω	-84		
V_n	Input voltage noise	$V_{CC} = \pm 5$ V, ± 12 V, ± 15 V, $f = 100$ kHz		3.2		nV/ $\sqrt{\text{Hz}}$
I_n	Input current noise	+Input -Input	$V_{CC} = \pm 5$ V, ± 12 V, ± 15 V, $f = 100$ kHz	1.5		pA/ $\sqrt{\text{Hz}}$
				10		
Crosstalk		$f = 1$ MHz, $V_{O(PP)} = 2$ V, $V_{CC} = \pm 5$ V, ± 12 V, ± 15 V	$R_L = 100$ Ω	-65		dBc
			$R_L = 25$ Ω	-60		dBc

ELECTRICAL CHARACTERISTICS (continued)

over recommended operating free-air temperature range, $T_A = 25^\circ\text{C}$, $V_{CC} = \pm 12\text{ V}$, $R_F = 2\text{ k}\Omega$, Gain = +5, $I_{ADJ} = \text{Bias1} = \text{Bias2} = 0\text{ V}$, $R_L = 50\ \Omega$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
OUTPUT CHARACTERISTICS							
V_O	Single-ended output voltage swing	$V_{CC} = \pm 5\text{ V}$	$R_L = 100\ \Omega$	± 3.9	± 4.1		V
			$R_L = 25\ \Omega$	± 3.7	± 3.9		
		$V_{CC} = \pm 12\text{ V}$	$R_L = 100\ \Omega$	± 10.7	± 11.0		V
			$R_L = 25\ \Omega$	± 10	± 10.6		
		$V_{CC} = \pm 15\text{ V}$	$R_L = 100\ \Omega$	± 13.5	± 13.9		V
			$R_L = 25\ \Omega$	± 12.7	± 13.4		
I_O	Output current ⁽¹⁾	$R_L = 5\ \Omega$	$V_{CC} = \pm 5\text{ V}$	± 350	± 400		mA
			$V_{CC} = \pm 12\text{ V}$	± 450	± 600		
		$R_L = 10\ \Omega$	$V_{CC} = \pm 15\text{ V}$	± 450	± 600		
$I_{(SC)}$	Short-circuit current ⁽¹⁾	$R_L = 1\ \Omega$	$V_{CC} = \pm 12\text{ V}$		1000		mA
	Output resistance	Open-loop			6		Ω
	Output resistance—terminate mode	$f = 1\text{ MHz}$,	Gain = +10		0.05		Ω
	Output resistance—shutdown mode	$f = 1\text{ MHz}$,	Open-loop		8.5		k Ω
POWER SUPPLY							
V_{CC}	Operating range	Dual supply		± 4	± 12	± 16.5	V
		Single supply		8	24	33	
I_{CC}	Quiescent current (each driver) ⁽²⁾ Full-bias mode (Bias-1 = 0, Bias-2 = 0) (Trimmed with $V_{CC} = \pm 12\text{ V}$ at 25°C)	$V_{CC} = \pm 5\text{ V}$	$T_A = 25^\circ\text{C}$		9.7	10.7	mA
			$T_A = \text{full range}$			11.7	
		$V_{CC} = \pm 12\text{ V}$	$T_A = 25^\circ\text{C}$		11	12	mA
			$T_A = \text{full range}$			12.5	
	$V_{CC} = \pm 15\text{ V}$	$T_A = 25^\circ\text{C}$		11.5	12.5	mA	
		$T_A = \text{full range}$			13		
Quiescent current (each driver) Variable bias modes, $V_{CC} = \pm 12\text{ V}$	Mid; Bias-1 = 1, Bias-2 = 0			7.5	8.5	mA	
	Low; Bias-1 = 0, Bias-2 = 1			4	5		
	Shutdown; Bias-1 = 1, Bias-2 = 1			0.25	0.9		
PSRR	Power supply rejection ratio	$V_{CC} = \pm 5\text{ V}$, $\Delta V_{CC} = \pm 0.5\text{ V}$	$T_A = 25^\circ\text{C}$	-50	-56	dB	
			$T_A = \text{full range}$	-47			
		$V_{CC} = \pm 12\text{ V}$, $\pm 15\text{ V}$, $\Delta V_{CC} = \pm 1\text{ V}$	$T_A = 25^\circ\text{C}$	-56	-60		
			$T_A = \text{full range}$	-53			
DYNAMIC PERFORMANCE							
BW	Single-ended small-signal bandwidth (-3 dB), $V_O = 0.1\text{ V}_{rms}$	$R_L = 100\ \Omega$	Gain = +1, $R_F = 1.2\text{ k}\Omega$		100	MHz	
			Gain = +2, $R_F = 1\text{ k}\Omega$		80		
			Gain = +5, $R_F = 1\text{ k}\Omega$		35		
			Gain = +10, $R_F = 1\text{ k}\Omega$		20		
		$R_L = 25\ \Omega$	Gain = +1, $R_F = 1.5\text{ k}\Omega$		65	MHz	
			Gain = +2, $R_F = 1\text{ k}\Omega$		60		
Gain = +5, $R_F = 1\text{ k}\Omega$			40				
			Gain = +10, $R_F = 1\text{ k}\Omega$		22		
SR	Single-ended slew rate ⁽³⁾	$V_O = 10\text{ V}_{pp}$,	Gain = +5		450		V/ μs

- (1) A heatsink is required to keep the junction temperature below absolute maximum rating when an output is heavily loaded or shorted. See Absolute Maximum Ratings section for more information.
- (2) Approximately 0.5 mA (total) flows from V_{CC+} to GND for internal logic control bias.
- (3) Slew rate is defined from the 25% to the 75% output levels.

ELECTRICAL CHARACTERISTICS (continued)

over recommended operating free-air temperature range, $T_A = 25^\circ\text{C}$, $V_{CC} = \pm 12\text{ V}$, $R_F = 2\text{ k}\Omega$,
Gain = +5, $I_{ADJ} = \text{Bias1} = \text{Bias2} = 0\text{ V}$, $R_L = 50\ \Omega$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
DC PERFORMANCE						
V_{OS}	Input offset voltage	$V_{CC} = \pm 5\text{ V}, \pm 12\text{ V}, \pm 15\text{ V}$	$T_A = 25^\circ\text{C}$	1	20	mV
			$T_A = \text{full range}$		25	
	Differential offset voltage		$T_A = 25^\circ\text{C}$	0.5	10	
			$T_A = \text{full range}$		15	
Offset drift		$T_A = \text{full range}$	50		$\mu\text{V}/^\circ\text{C}$	
I_{IB}	-Input bias current	$V_{CC} = \pm 5\text{ V}, \pm 12\text{ V}, \pm 15\text{ V}$	$T_A = 25^\circ\text{C}$	8	15	μA
			$T_A = \text{full range}$		20	
	+Input bias current		$T_A = 25^\circ\text{C}$	8	15	
			$T_A = \text{full range}$		20	
Z_{OL}	Open loop transimpedance	$R_L = 1\text{ k}\Omega, V_{CC} = \pm 12\text{ V}, \pm 15\text{ V}$		900		k Ω
INPUT CHARACTERISTICS						
V_{ICR}	Input common-mode voltage range	$V_{CC} = \pm 5\text{ V}$	$T_A = 25^\circ\text{C}$	± 2.7	± 3.0	V
			$T_A = \text{full range}$	± 2.6		
		$V_{CC} = \pm 12\text{ V}$	$T_A = 25^\circ\text{C}$	± 9.5	± 9.8	V
			$T_A = \text{full range}$	± 9.3		
		$V_{CC} = \pm 15\text{ V}$	$T_A = 25^\circ\text{C}$	± 12.4	± 12.7	V
			$T_A = \text{full range}$	± 12.1		
CMRR	Common-mode rejection ratio	$V_{CC} = \pm 5\text{ V}, \pm 12\text{ V}, \pm 15\text{ V}$	$T_A = 25^\circ\text{C}$	48	54	dB
			$T_A = \text{full range}$	44		
R_i	Input resistance	+Input		800		k Ω
		-Input		30		Ω
C_i	Input capacitance			1.7		pF
LOGIC CONTROL CHARACTERISTICS						
V_{IH}	Bias pin voltage for logic 1	Relative to GND pin voltage		2.0		V
V_{IL}	Bias pin voltage for logic 0				0.8	
I_{IH}	Bias pin current for logic 1	$V_{IH} = 3.3\text{ V}, \text{GND} = 0\text{ V}$		4	30	μA
I_{IL}	Bias pin current for logic 0	$V_{IL} = 0.5\text{ V}, \text{GND} = 0\text{ V}$		1	10	μA
	Transition time, logic 0 to logic 1 ⁽⁴⁾			1		μs
	Transition time, logic 1 to logic 0 ⁽⁴⁾			1		μs

(4) Transition time is defined as the time from when the logic signal is applied to the time when the supply current has reached half its final value.

LOGIC TABLE⁽¹⁾⁽²⁾

BIAS-1	BIAS-2	FUNCTION	DESCRIPTION
0	0	Full bias mode	Amplifiers ON with lowest distortion possible (default state)
1	0	Mid bias mode	Amplifiers ON with power savings with a reduction in distortion performance
0	1	Low bias mode	Amplifiers ON with enhanced power savings and a reduction of distortion performance
1	1	Shutdown mode	Amplifiers OFF and output has high impedance

(1) The default state for all logic pins is a logic zero (0).

(2) The GND pin useable range is from V_{CC-} to $(V_{CC+} - 4\text{ V})$.

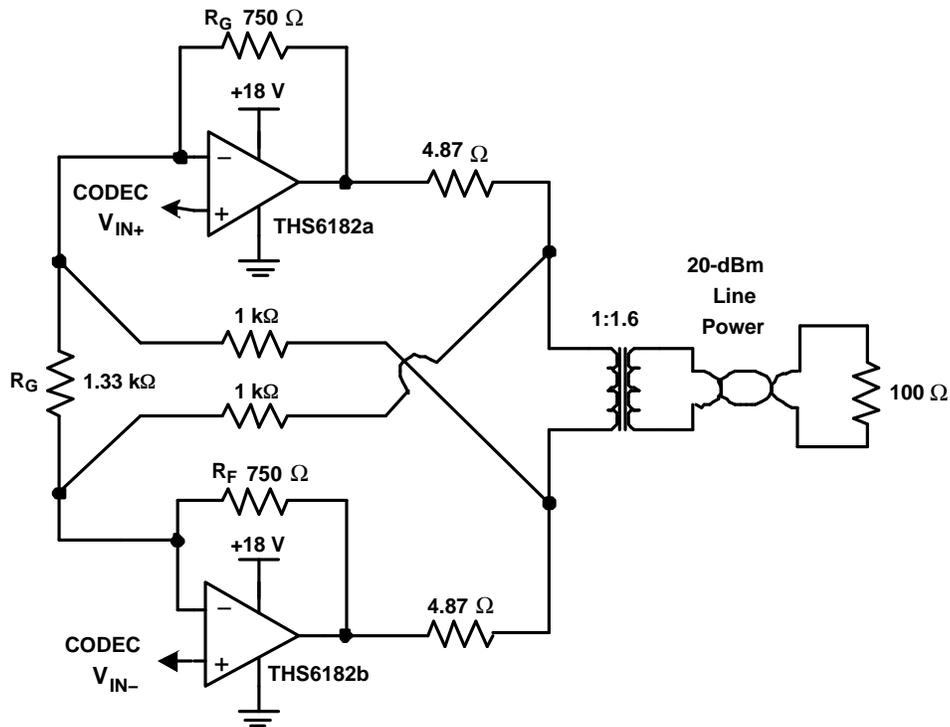
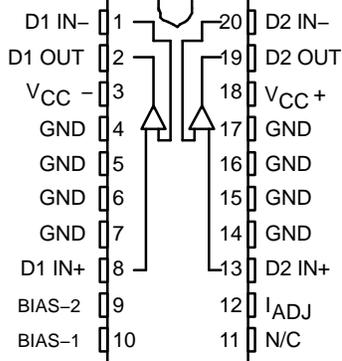


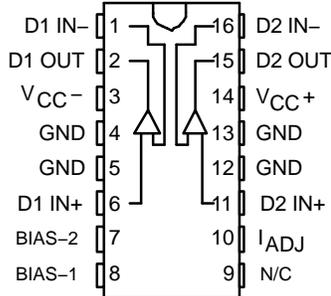
Figure 1. Single-Supply ADSL CO Line Driver Circuit Utilizing Active Impedance (SF = 4)

PIN ASSIGNMENTS

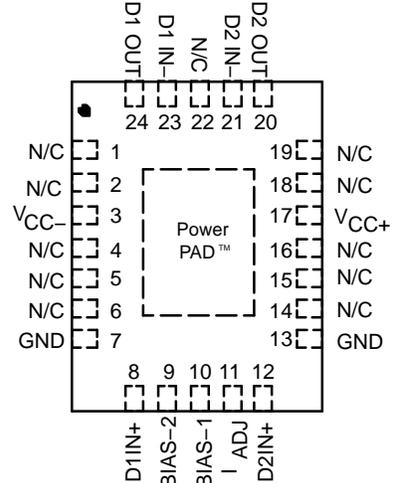
THS6182
SOIC-20 (DW) AND
SOIC-20 PowerPAD™ (DWP) PACKAGES
(TOP VIEW)



THS6182
SOIC-16 (D) PACKAGE
(TOP VIEW)



THS6182
Leadless 24-pin PowerPAD™
4 mm X 5 mm (RHF) PACKAGE
(TOP VIEW)



- A. The PowerPAD is electrically isolated from all active circuitry and pins. Connection of the PowerPAD to the PCB ground plane is highly recommended, although not required, as this plane is typically the largest copper plane on a PCB. The thermal performance will be better with a large copper plane than a small one.

TYPICAL CHARACTERISTICS

Table of Graphs

		FIGURE
Output voltage headroom	vs Output current	2
Common-mode rejection ratio	vs Frequency	3
Crosstalk	vs Frequency	4
Total quiescent current		5
Large signal output amplitude	vs Frequency	6-8
Voltage and current noise	vs Frequency	9
Overdrive recovery		10
Power supply rejection ratio	vs Frequency	11
Output amplitude	vs Frequency	12-37
Slew rate	vs Output voltage	38
Closed-loop output impedance	vs Frequency	39
Quiescent current	vs Supply voltage	40
	vs Temperature	41
Common-mode rejection ratio	vs Common-mode voltage	42
Input bias current	vs Temperature	43
Input offset voltage	vs Temperature	44
2nd Harmonic distribution	vs Frequency	45-52
3rd Harmonic distribution	vs Frequency	53-60
2nd Harmonic distribution	vs Output voltage	61-64
3rd Harmonic distribution	vs Output voltage	65-68

**OUTPUT VOLTAGE HEADROOM
vs
OUTPUT CURRENT**

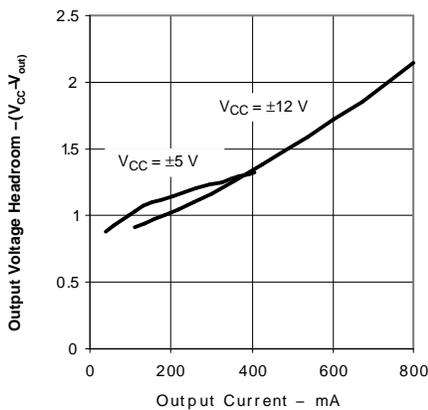


Figure 2.

**COMMON-MODE REJECTION
RATIO
vs
FREQUENCY**

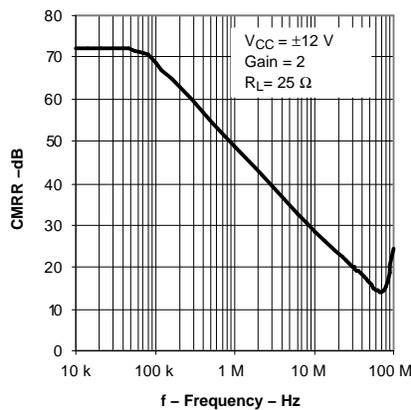


Figure 3.

**CROSSTALK
vs
FREQUENCY**

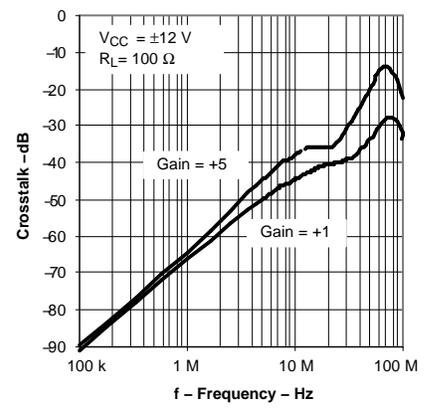


Figure 4.

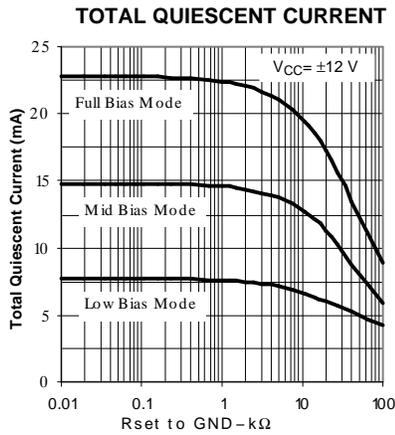


Figure 5.

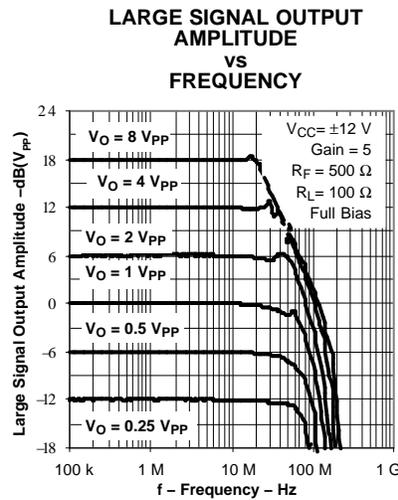


Figure 6.

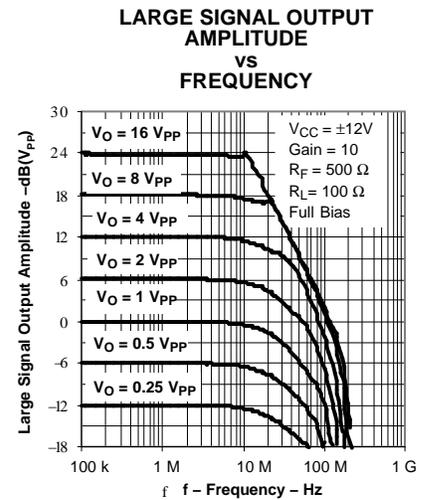


Figure 7.

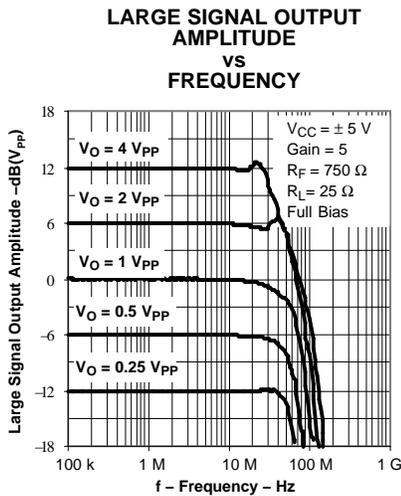


Figure 8.

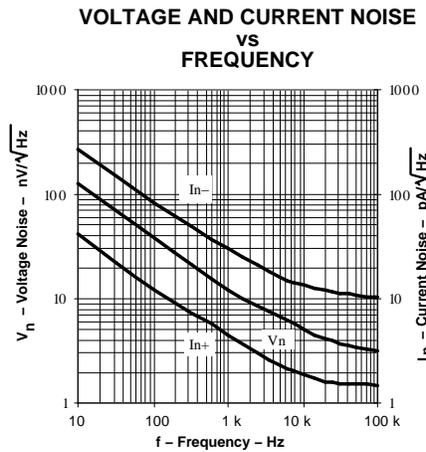


Figure 9.

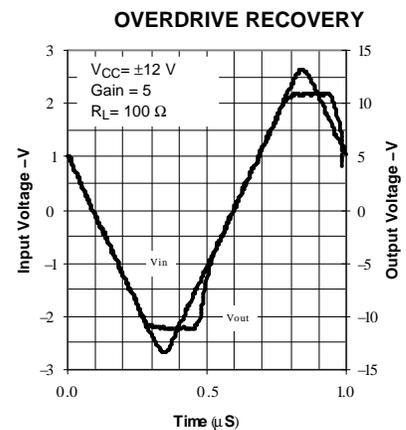


Figure 10.

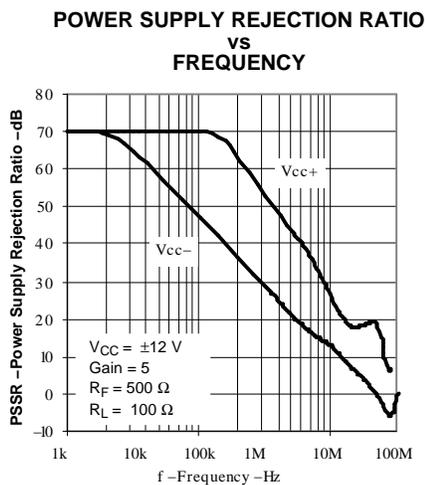


Figure 11.

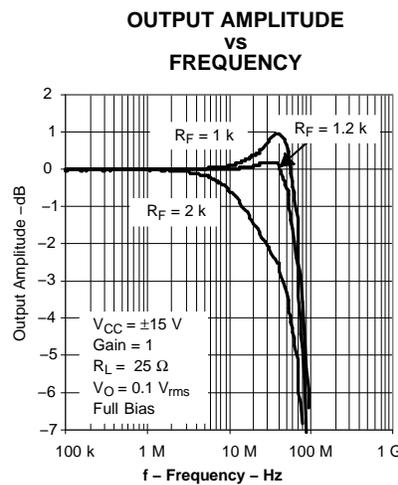


Figure 12.

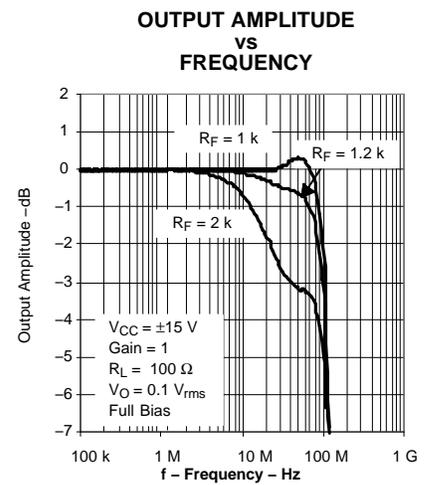


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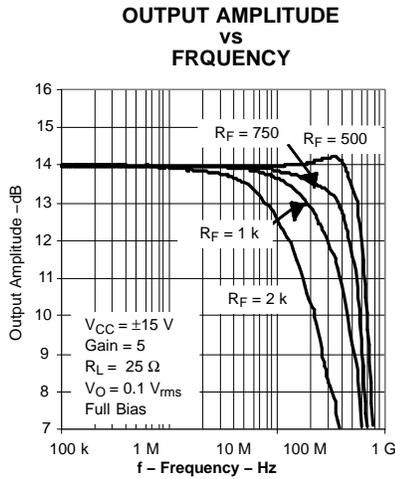


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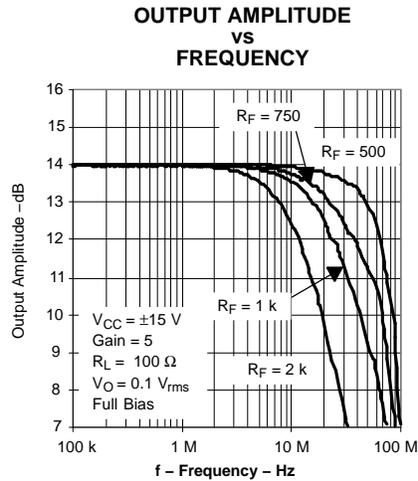


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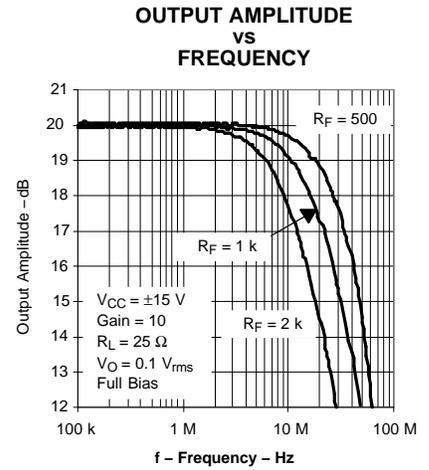


Figure 16.

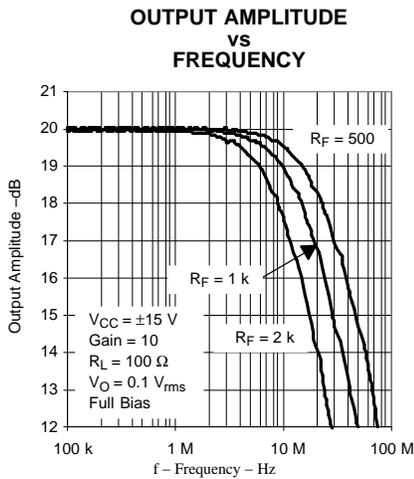


Figure 17.

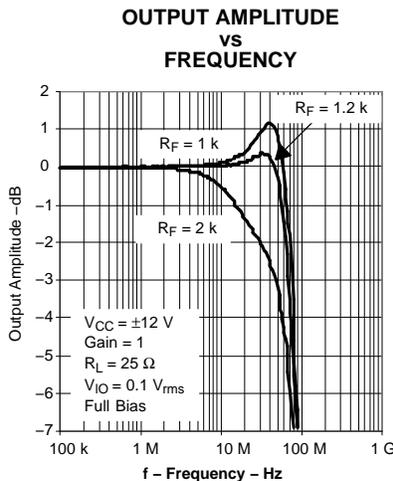


Figure 18.

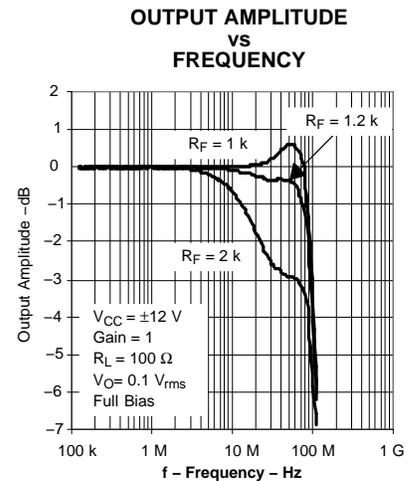


Figure 19.

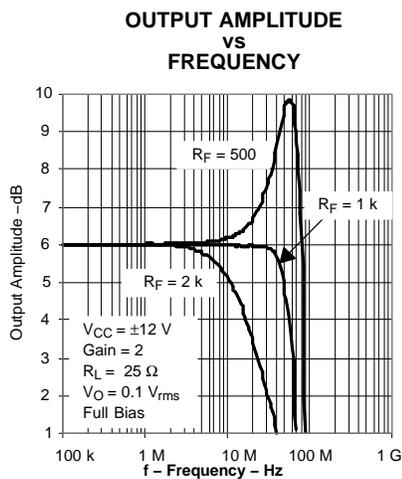


Figure 20.

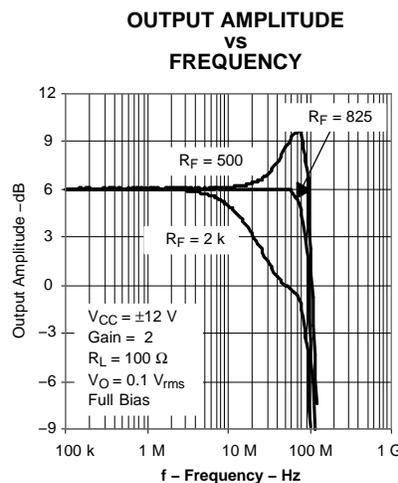


Figure 21.

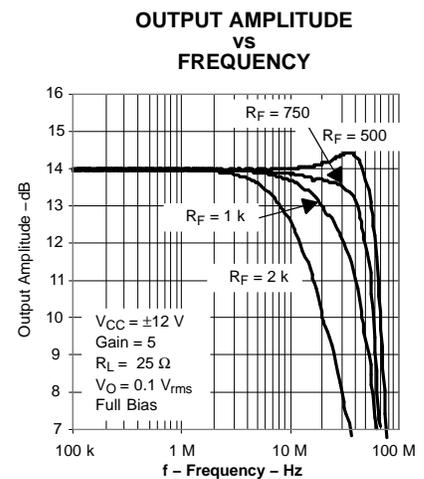


Figure 22.

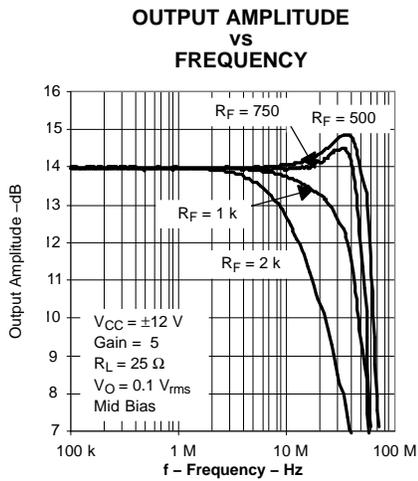


Figure 23.

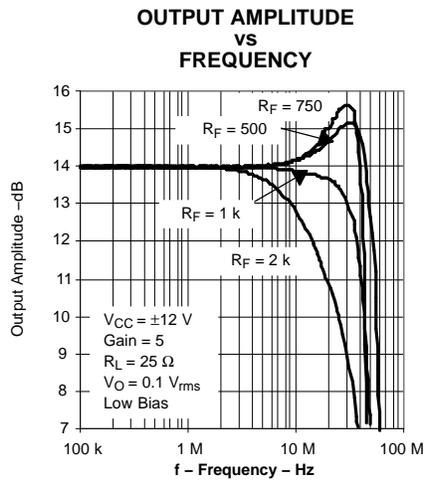


Figure 24.

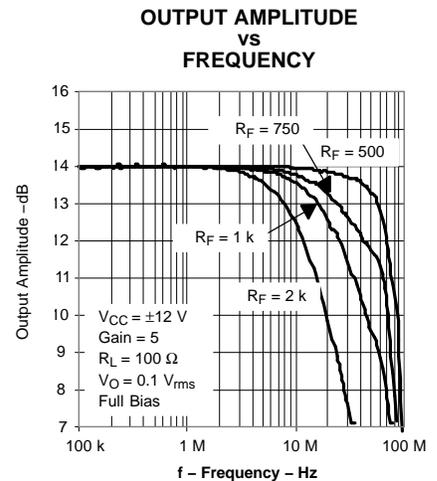


Figure 25.

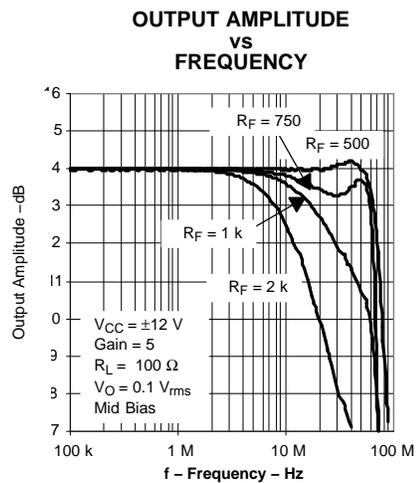


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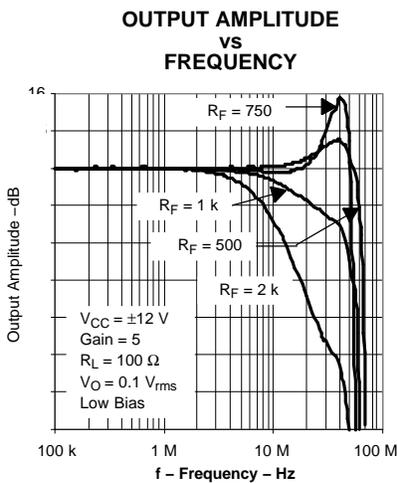


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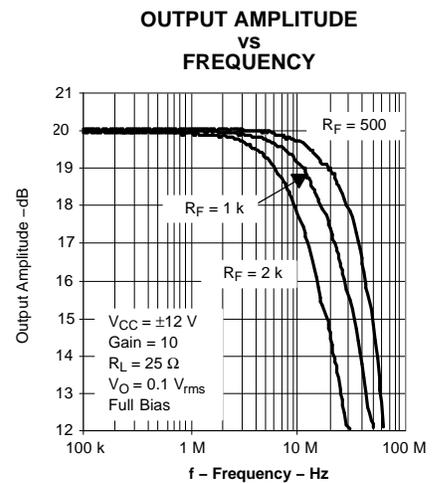


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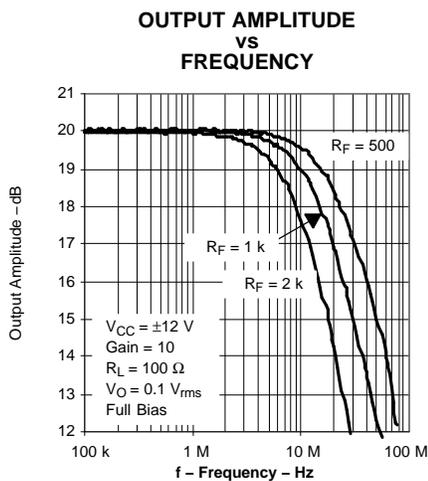


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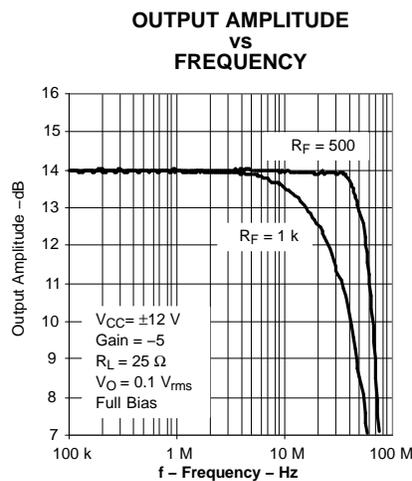


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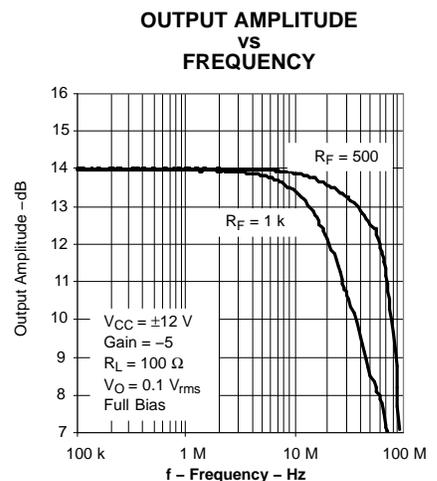


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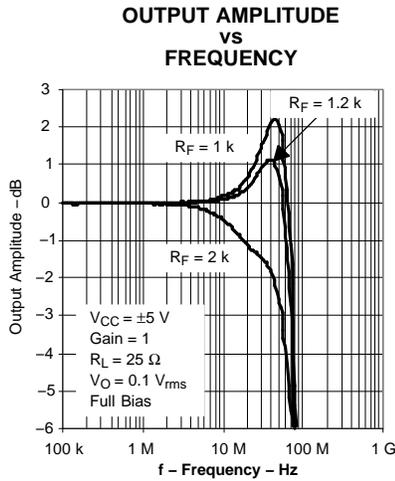


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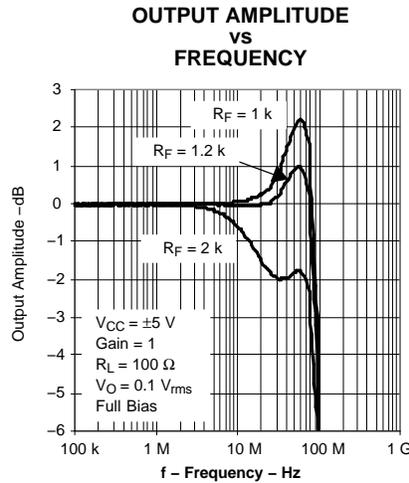


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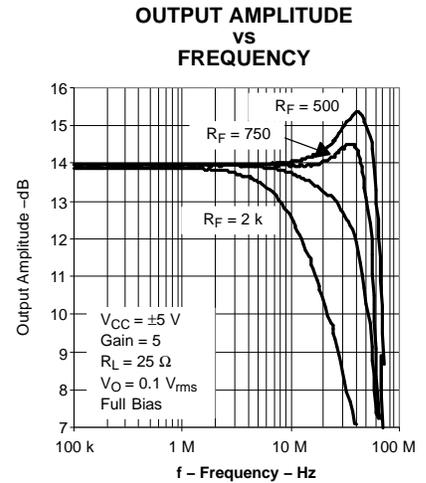


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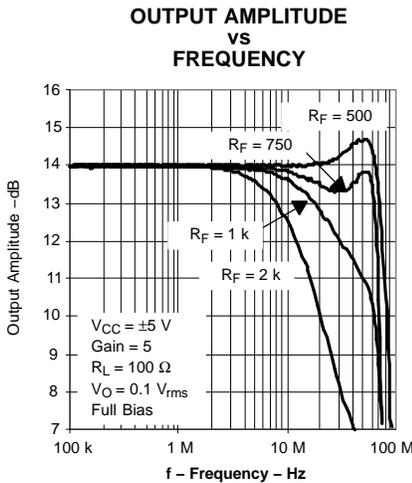


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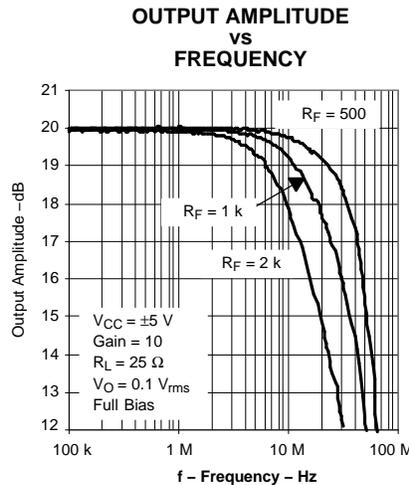


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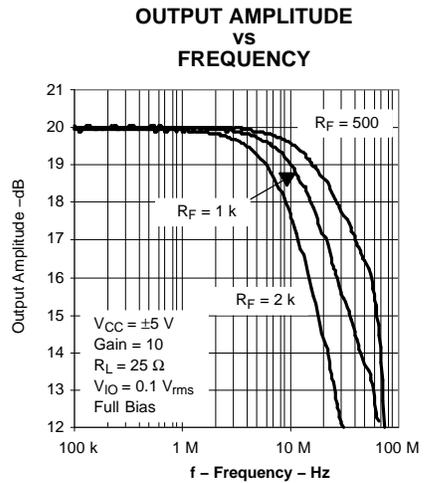


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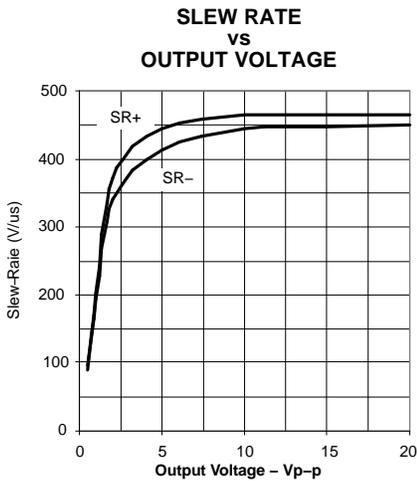


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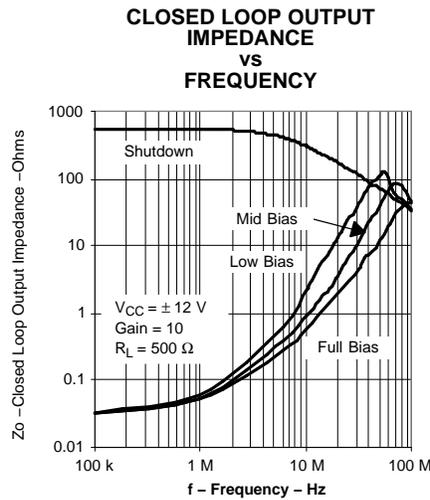


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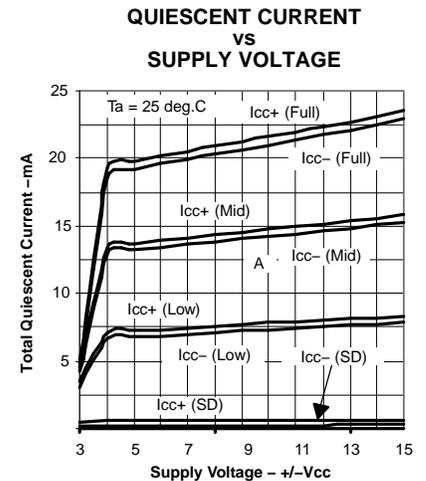


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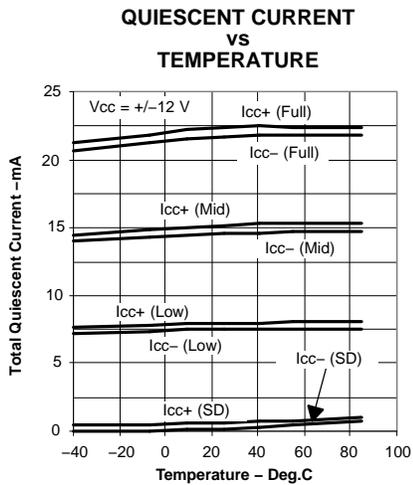


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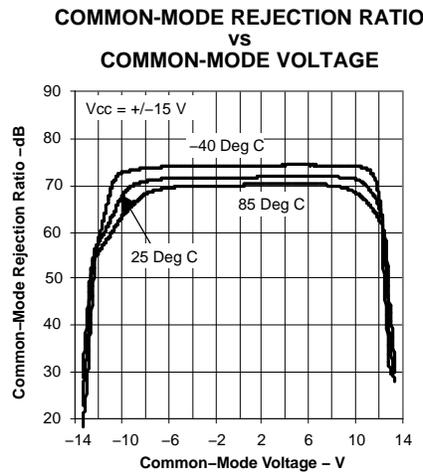


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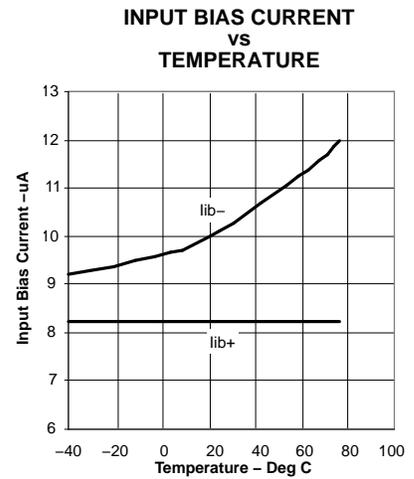


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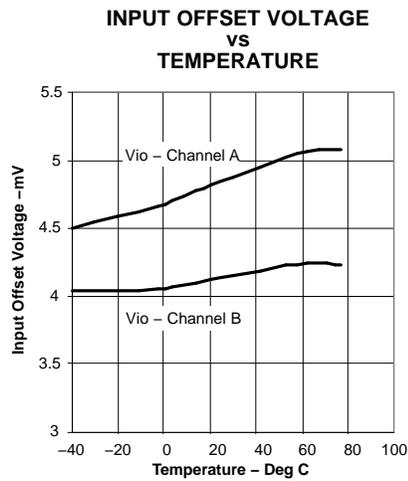


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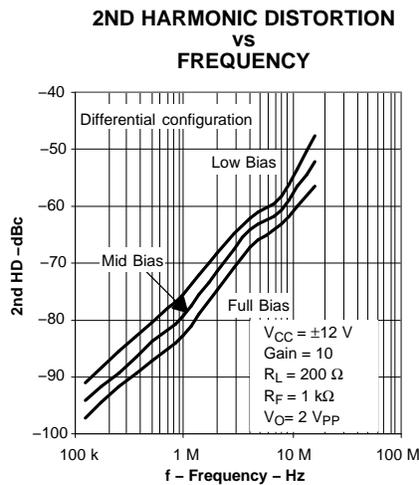


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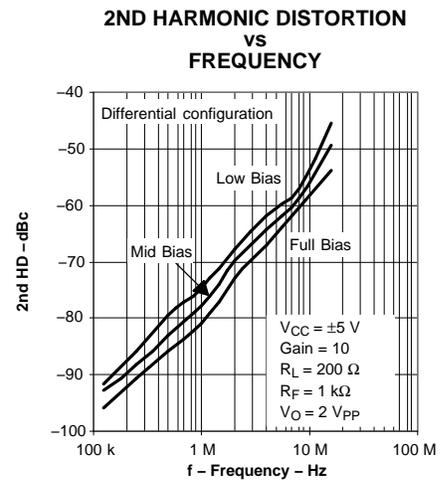


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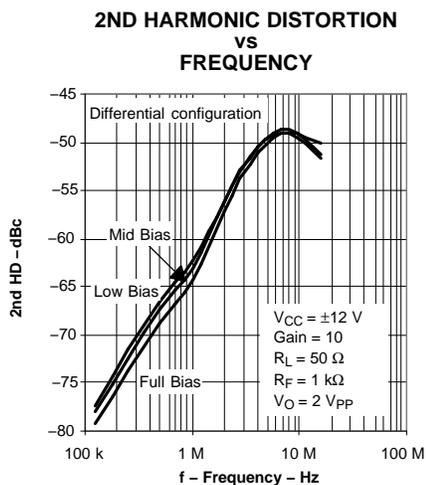


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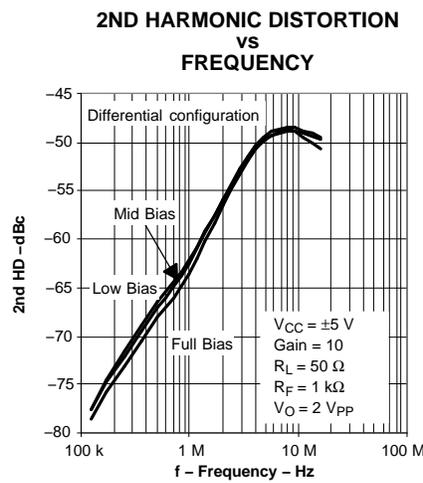


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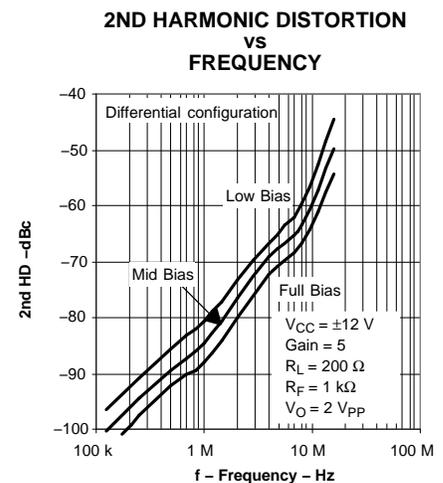


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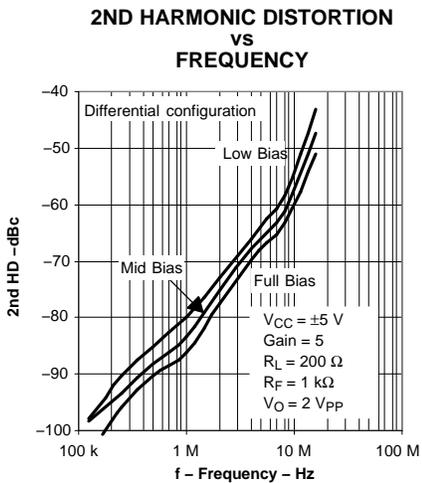


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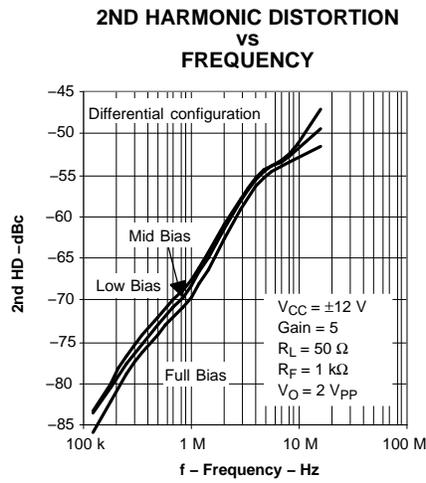


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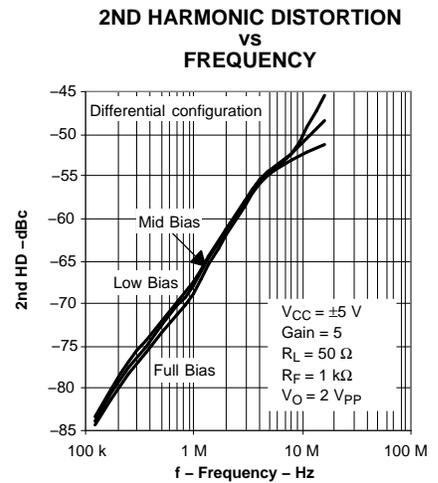


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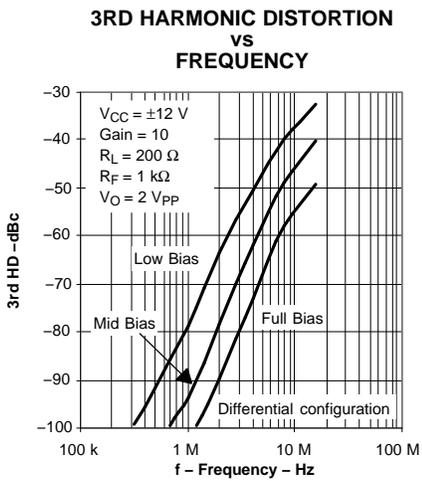


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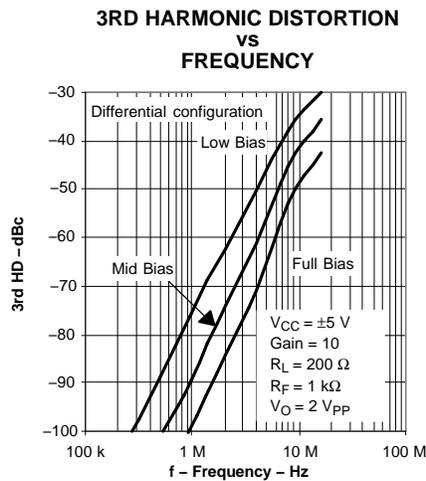


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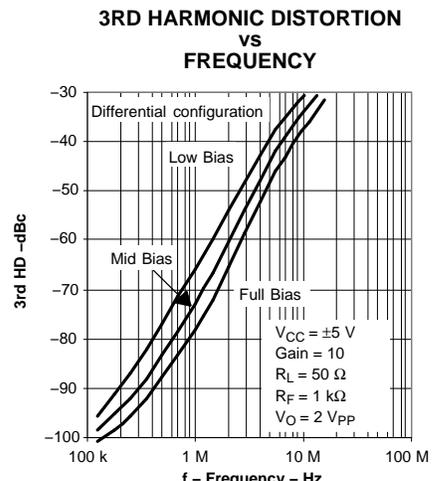


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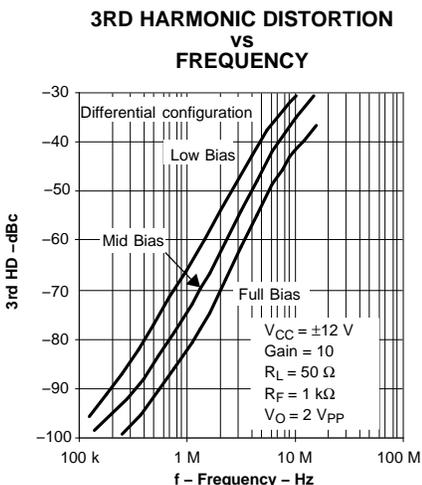


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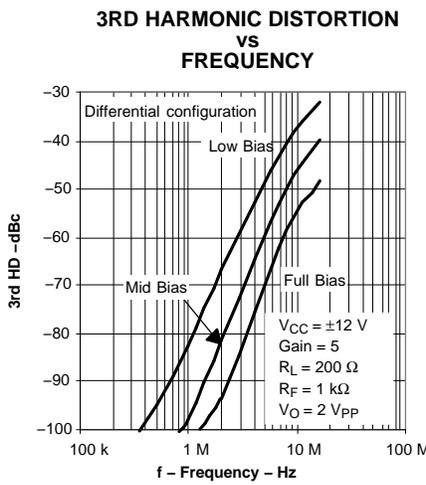


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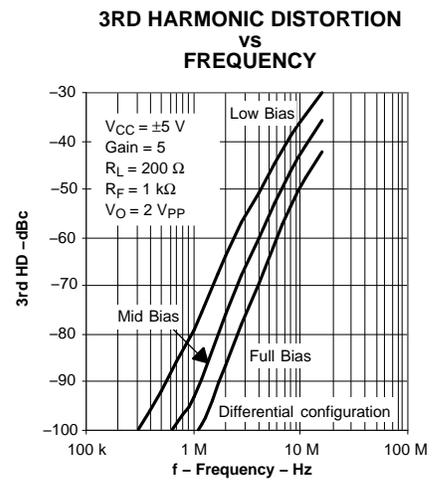


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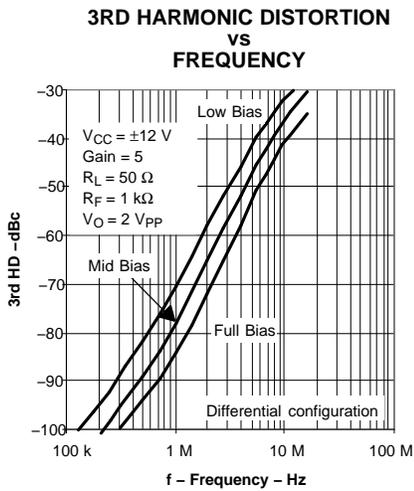


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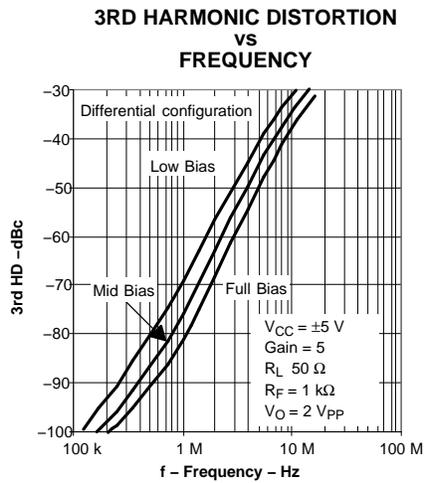


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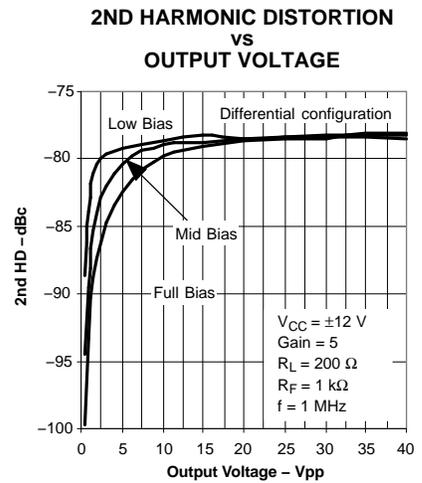


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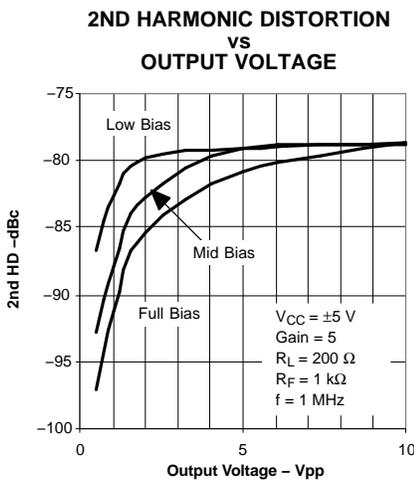


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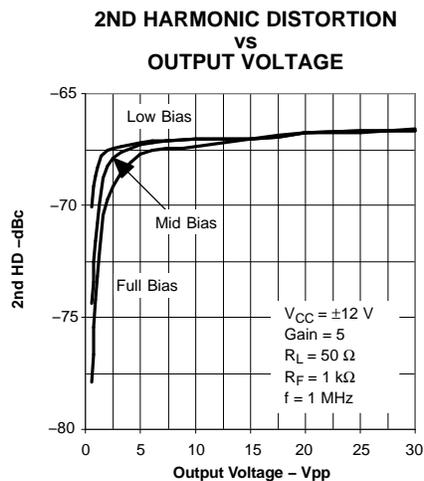


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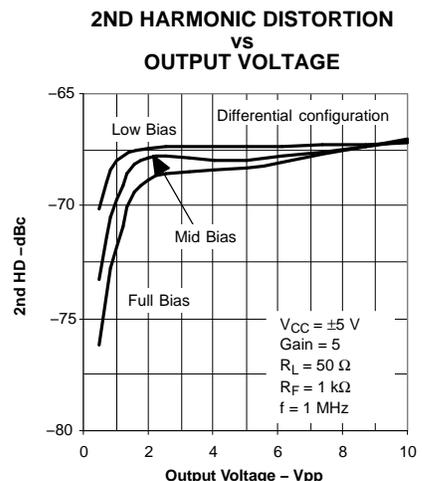


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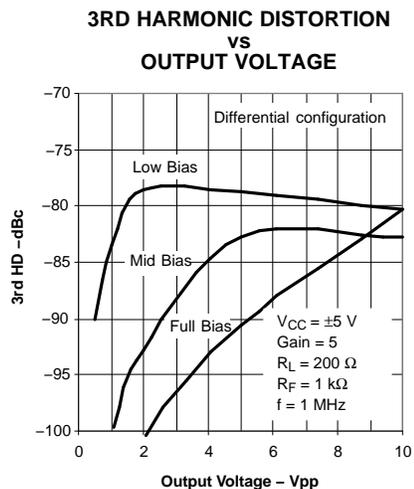


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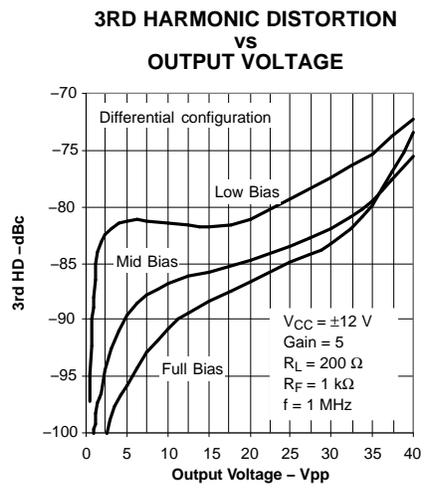


Figure 66.

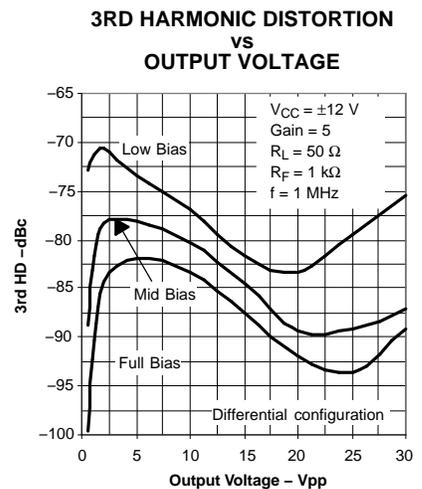


Figure 67.

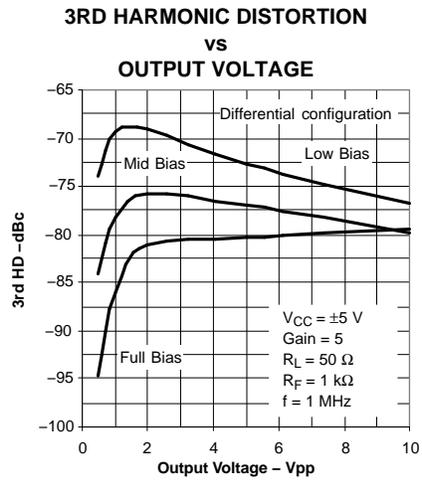


Figure 68.

APPLICATION INFORMATION

The THS6182 contains two independent operational amplifiers. These amplifiers are current feedback topology amplifiers made for high-speed operation. They have been specifically designed to deliver the full power requirements of ADSL and therefore can deliver output currents of at least 400 mA at full output voltage.

The THS6182 is fabricated using Texas Instruments 30-V complementary bipolar process, HVBiCOM. This process provides excellent isolation and high slew rates that result in the device's excellent crosstalk and extremely low distortion.

DEVICE PROTECTION FEATURE

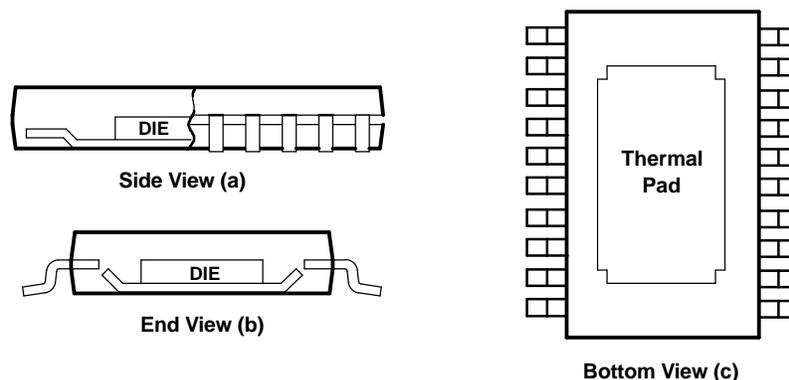
The THS6182 has a built-in thermal protection feature. Should the internal junction temperature rise above approximately 160°C, the device automatically shuts down. Such a condition could exist with improper heat sinking or if the output is shorted to ground. When the abnormal condition is fixed, the internal thermal shutdown circuit automatically turns the device back on. This occurs at approximately 145°C, junction temperature. Note that the THS6182 does not have short-circuit protection and care should be taken to minimize the output current below the absolute maximum ratings.

THERMAL INFORMATION

The THS6182 is available in a thermally-enhanced DWP and RHF package, which is a member of the PowerPAD family of packages. This package is constructed using a downset leadframe upon which the die is mounted [see [Figure 69\(a\)](#) and [Figure 69\(b\)](#), for the DWP package example]. This arrangement results in the lead frame being exposed as a thermal pad on the underside of the package [see [Figure 69\(c\)](#)]. Because this thermal pad has direct thermal contact with the die, excellent thermal performance can be achieved by providing a good thermal path away from the thermal pad. Note that the PowerPAD is electronically isolated from the active circuitry and any pins. Thus, the PowerPAD can be connected to any potential voltage within the absolute maximum voltage range. Ideally, connection of the PAD to the ground plane is preferred as the plane typically is the largest copper plane on a PCB.

The PowerPAD package allows for both assembly and thermal management in one manufacturing operation. During the surface-mount solder operation (when the leads are being soldered), the thermal pad can also be soldered to a copper area underneath the package. Through the use of thermal paths within this copper area, heat can be conducted away from the package into either a ground plane or other heat dissipating device. This is discussed in more detail in the *PCB design considerations* section of this document.

The PowerPAD package represents a breakthrough in combining the small area and ease of assembly of surface mount with the, heretofore, awkward mechanical methods of heatsinking.



- A. The thermal pad is electrically isolated from all terminals in the package.

Figure 69. Views of Thermally Enhanced DWP Package

APPLICATION INFORMATION (continued)

RECOMMENDED FEEDBACK AND GAIN RESISTOR VALUES

As with all current feedback amplifiers, the bandwidth of the THS6182 is an inversely proportional function of the value of the feedback resistor. The recommended resistors with a $\pm 12\text{-V}$ power supply for the optimum frequency response with a $25\text{-}\Omega$ load system is $1\text{ k}\Omega$ for a gain of 5. These should be used as a starting point and once optimum values are found, 1% tolerance resistors should be used to maintain frequency response characteristics.

Consistent with current feedback amplifiers, increasing the gain is best accomplished by changing the gain resistor, not the feedback resistor. This is because the bandwidth of the amplifier is dominated by the feedback resistor value and internal dominant-pole capacitor. The ability to control the amplifier gain independently of the bandwidth constitutes a major advantage of current feedback amplifiers over conventional voltage feedback amplifiers.

It is important to realize the effects of the feedback resistance on distortion. Increasing the resistance decreases the loop gain and increases the distortion. It is also important to know that decreasing load impedance increases total harmonic distortion (THD). Typically, the third order harmonic distortion increases more than the second order harmonic distortion.

Finally, in a differential configuration as shown in Figure 1, it is important to note that there is a differential gain and a common-mode gain which are different from each other. Differentially, the gain is at $1 + R_F/R_G$. While common-mode gain = 1 due to R_G being connected directly between each amplifier and not to ground. This can lead to potential problems as the stability of the amplifier is determined by R_F . Thus, R_F must be large enough to ensure the common-mode stability, even though a large differential gain may be required.

OFFSET VOLTAGE

The output offset voltage, (V_{OO}) is the sum of the input offset voltage (V_{IO}) and both input bias currents (I_{IB}) times the corresponding gains. The following schematic and formula can be used to calculate the output offset voltage:

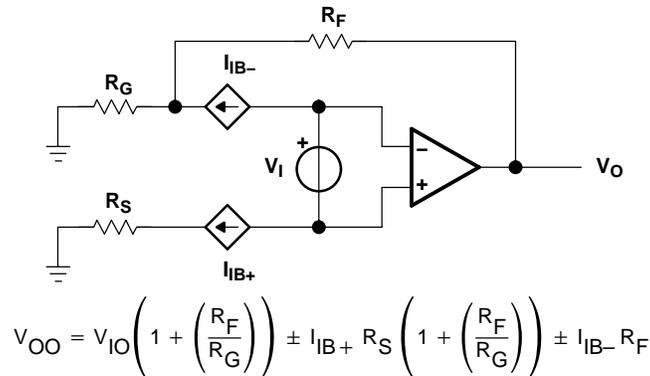
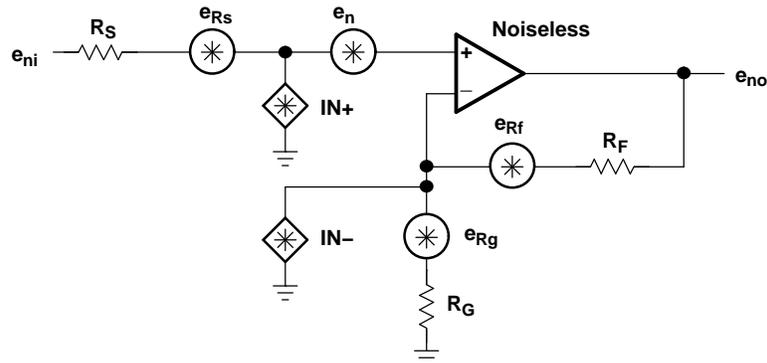


Figure 70. Output Offset Voltage Model

NOISE CALCULATIONS

Noise can cause errors on very small signals. This is especially true for the amplifying small signals. The noise model for current feedback amplifiers (CFB) is the same as voltage feedback amplifiers (VFB). The only difference between the two is that the CFB amplifiers generally specify different current noise parameters for each input while VFB amplifiers usually only specify one noise current parameter. The noise model is shown in Figure 71. This model includes all of the noise sources as follows:

- e_n = Amplifier internal voltage noise ($\text{nV}/\sqrt{\text{Hz}}$)
- $IN+$ = Noninverting current noise ($\text{pA}/\sqrt{\text{Hz}}$)
- $IN-$ = Inverting current noise ($\text{pA}/\sqrt{\text{Hz}}$)
- e_{RX} = Thermal voltage noise associated with each resistor ($e_{RX} = 4\text{ kTR}_x$)

APPLICATION INFORMATION (continued)**Figure 71. Noise Model**

The total equivalent input noise density (e_{ni}) is calculated by using the following equation:

$$e_{ni} = \sqrt{(e_n)^2 + (IN+ \times R_S)^2 + (IN- \times (R_F \parallel R_G))^2 + 4kTR_S + 4kT(R_F \parallel R_G)}$$

Where:

$$k = \text{Boltzmann's constant} = 1.380658 \times 10^{-23}$$

$$T = \text{Temperature in degrees Kelvin (273 + } ^\circ\text{C)}$$

$$R_F \parallel R_G = \text{Parallel resistance of } R_F \text{ and } R_G$$

To get the equivalent output noise of the amplifier, just multiply the equivalent input noise density (e_{ni}) by the overall amplifier gain (A_V).

$$e_{no} = e_{ni} A_V = e_{ni} \left(1 + \frac{R_F}{R_G} \right) \text{ (Noninverting Case)}$$

As the previous equations show, to keep noise at a minimum, small value resistors should be used. As the closed-loop gain is increased (by reducing R_G), the input noise is reduced considerably because of the parallel resistance term.

DRIVING A CAPACITIVE LOAD

Driving capacitive loads with high performance amplifiers is not a problem as long as certain precautions are taken. The first is to realize that the THS6182 has been internally compensated to maximize its bandwidth and slew rate performance. When the amplifier is compensated in this manner, capacitive loading directly on the output will decrease the device's phase margin leading to high frequency ringing or oscillations. Therefore, for capacitive loads of greater than 10 pF, it is recommended that a resistor be placed in series with the output of the amplifier, as shown in [Figure 72](#). A minimum value of 2 Ω should work well for most applications. For example, in 75- Ω transmission systems, setting the series resistor value to 75 Ω both isolates any capacitance loading and provides the proper line impedance matching at the source end.

APPLICATION INFORMATION (continued)

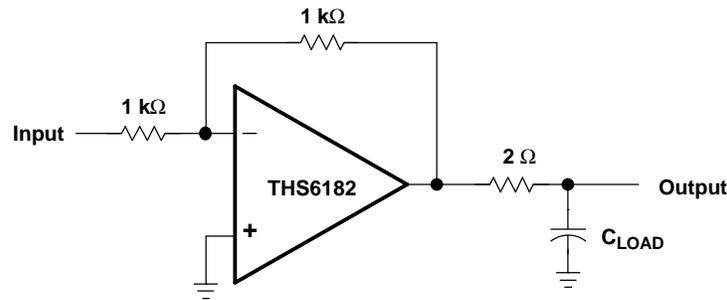


Figure 72. Driving a Capacitive Load

PCB DESIGN CONSIDERATIONS

Proper PCB design techniques in two areas are important to assure proper operation of the THS6182. These areas are high-speed layout techniques and thermal-management techniques. Because the THS6182 is a high-speed part, the following guidelines are recommended.

- Ground plane - It is essential that a ground plane be used on the board to provide all components with a low inductive ground connection. Although a ground connection directly to a terminal of the THS6012 is not necessarily required, it is recommended that the thermal pad of the package be tied to ground. This serves two functions. It provides a low inductive ground to the device substrate to minimize internal crosstalk and it provides the path for heat removal. Note that the BiCom process is a SOI process and thus, the substrate is isolated from the active circuitry.
- Input stray capacitance - To minimize potential problems with amplifier oscillation, the capacitance at the inverting input of the amplifiers must be kept to a minimum. To do this, PCB trace runs to the inverting input must be as short as possible, the ground plane should be removed under any etch runs connected to the inverting input, and external components should be placed as close as possible to the inverting input. This is especially true in the noninverting configuration.
- Proper power supply decoupling - Use a minimum of a 6.8- μF tantalum capacitor in parallel with a 0.1- μF ceramic capacitor on each supply terminal. It may be possible to share the tantalum among several amplifiers depending on the application, but a 0.1- μF ceramic capacitor should always be used on the supply terminal of every amplifier. In addition, the 0.1- μF capacitor should be placed as close as possible to the supply terminal. As this distance increases, the inductance in the connecting etch makes the capacitor less effective. The designer should strive for distances of less than 0.1 inches between the device power terminal and the ceramic capacitors.
- For a differential configuration as shown in [Figure 1](#), it is recommended that a 0.1- μF or 1- μF capacitor be added across the power supplies (from V_{CC+} to V_{CC-}) as close as possible to the THS6182. This allows for differential currents to flow properly, significantly reducing even-order harmonic distortion. The 0.1- μF capacitors to ground should also be used as previously stipulated.

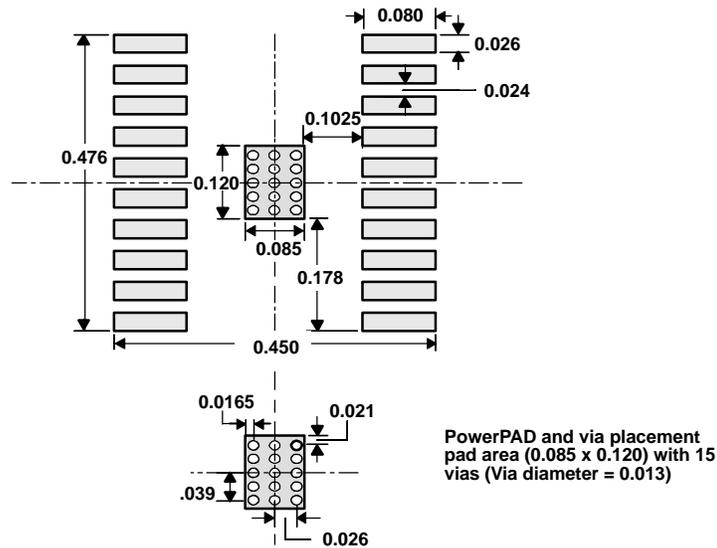
Because of its power dissipation, proper thermal management of the THS6182 is required. Although there are many ways to properly heatsink this device, the following steps illustrate one recommended approach for a multilayer PCB with an internal ground plane utilizing the 20 pin DWP PowerPAD package.

1. Prepare the PCB with a top side etch pattern as shown in [Figure 73](#). There should be etch for the leads as well as etch for the thermal pad.
2. Place 18 holes in the area of the thermal pad. These holes should be 13 mils in diameter. They are kept small so that solder wicking through the holes is not a problem during reflow.
3. It is recommended, but not required, to place six more holes under the package, but outside the thermal pad area. These holes are 25 mils in diameter. They may be larger because they are not in the area to be soldered so that wicking is not a problem.
4. Connect all 24 holes, the 18 within the thermal pad area and the 6 outside the pad area, to the internal ground plane.
5. When connecting these holes to the ground plane, do **not** use the typical web or spoke via connection methodology. Web connections have a high thermal resistance connection that is useful for slowing the heat

APPLICATION INFORMATION (continued)

transfer during soldering operations. This makes the soldering of vias that have plane connections easier. However, in this application, low thermal resistance is desired for the most efficient heat transfer. Therefore, the holes under the THS6182 package should make their connection to the internal ground plane with a complete connection around the entire circumference of the plated through hole.

6. The top-side solder mask should leave exposed the terminals of the package and the thermal pad area with its five holes. The four larger holes outside the thermal pad area, but still under the package, should be covered with solder mask.
7. Apply solder paste to the exposed thermal pad area and all of the operational amplifier terminals.
8. With these preparatory steps in place, the THS6182 DWP is simply placed in position and run through the solder reflow operation as any standard surface-mount component. This results in a part that is properly installed.



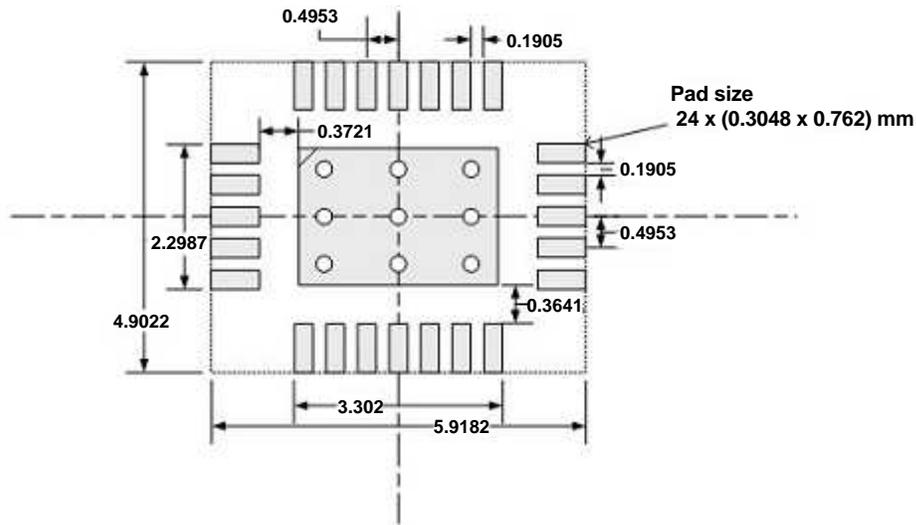
Vias should go through the board connecting the top layer PowerPad to any and all ground planes. (The larger the ground plane, the larger the area to distribute the heat.) Solder resist should be used on the bottom side ground plane in order to prevent wicking of the solder through the vias during the reflow process.

All Units in Inches

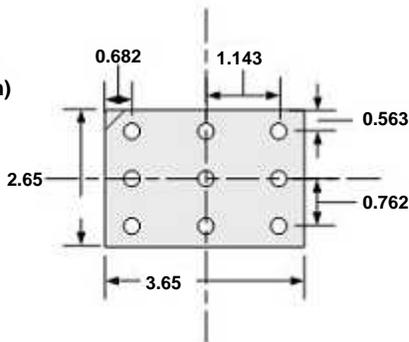
Figure 73. 20-Pin DWP PowerPAD PCB Etch and Via Pattern

The RHF package is similar to the DWP package with respect to PCB mounting procedures. The recommended PCB layout is as shown in [Figure 74](#).

APPLICATION INFORMATION (continued)



PowerPAD and Via layout
(Pad size 3.65 mm x 2.65 mm ,
9 Vias with diameter = 0.254 mm)



Vias should go through the board connecting the top layer PowerPAD to any and all ground planes. The larger the ground plane, the more area to distribute the heat.

Solder resist should be used on the bottom side ground plane to prevent wicking of the solder through the vias during the reflow process.

Figure 74. Suggested PCB Layout

The actual thermal performance achieved with the THS6182 in the 20-pin DWP PowerPAD package depends on the application. In the previous example, if the size of the internal ground plane is approximately 3 inches x 3 inches, then the expected thermal coefficient, Θ_{JA} , is about 21.5°C/W. (See the Package Dissipation Ratings Table for all other package metrics.) For a given Θ_{JA} , the maximum power dissipation is calculated by the following formula:

APPLICATION INFORMATION (continued)

$$P_D = \left(\frac{T_{MAX} - T_A}{\theta_{JA}} \right)$$

Where:

P_D = Maximum power dissipation of THS6182 (watts)

T_{MAX} = Absolute maximum operating junction temperature (125°C)

T_A = Free-ambient air temperature (°C)

θ_{JA} = $\theta_{JC} + \theta_{CA}$

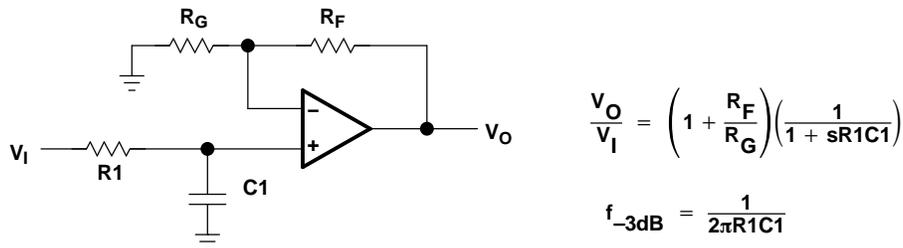
θ_{JC} = Thermal coefficient from junction to case. See the Package Dissipation Ratings table.

θ_{CA} = Thermal coefficient from case to ambient determined by PCB layout and construction.

More complete details of the PowerPAD installation process and thermal management techniques can be found in the Texas Instruments Technical Brief, *PowerPAD Thermally Enhanced Package*. This document can be found at the TI web site (www.ti.com) by searching on the key word PowerPAD. The document can also be ordered through your local TI sales office. Refer to literature number [SLMA002](#) when ordering.

GENERAL CONFIGURATIONS

A common error for the first-time CFB user is to create a unity gain buffer amplifier by shorting the output directly to the inverting input. A CFB amplifier in this configuration oscillates and is **not** recommended. The THS6182, like all CFB amplifiers, **must** have a feedback resistor for stable operation. Additionally, placing capacitors directly from the output to the inverting input is not recommended. This is because, at high frequencies, a capacitor has a very low impedance. This results in an unstable amplifier and should not be considered when using a current-feedback amplifier. Because of this, integrators and simple low-pass filters, which are easily implemented on a VFB amplifier, have to be designed slightly differently. If filtering is required, simply place an RC-filter at the noninverting terminal of the operational-amplifier (see [Figure 75](#)).

**Figure 75. Single-Pole Low-Pass Filter**

If a multiple pole filter is required, the use of a Sallen-Key filter can work very well with CFB amplifiers. This is because the filtering elements are not in the negative feedback loop and stability is not compromised. Because of their high slew-rates and high bandwidths, CFB amplifiers can create very accurate signals and help minimize distortion. An example is shown in [Figure 76](#).

APPLICATION INFORMATION (continued)

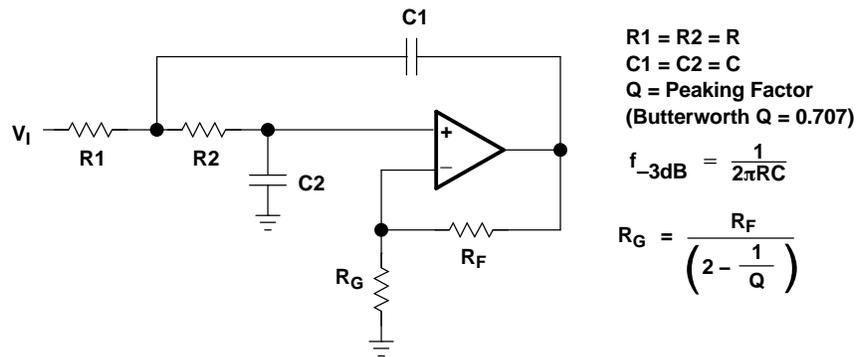


Figure 76. 2-Pole Low-Pass Sallen-Key Filter

EVALUATION BOARD

An evaluation board is available for the THS6182. This board has been configured for proper thermal management of the THS6182. The circuitry has been designed for a typical ADSL application as shown previously in this document. For more detailed information, refer to the *THS6182EVM User's Guide* (literature number [SLOU152](#)). To order the evaluation board contact your local TI sales office or distributor.

PACKAGING INFORMATION

Orderable Device	Status ⁽¹⁾	Package Type	Package Drawing	Pins	Package Qty	Eco Plan ⁽²⁾	Lead/Ball Finish	MSL Peak Temp ⁽³⁾
THS6182D	ACTIVE	SOIC	D	16	40	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
THS6182DG4	ACTIVE	SOIC	D	16	40	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
THS6182DR	ACTIVE	SOIC	D	16	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
THS6182DRG4	ACTIVE	SOIC	D	16	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
THS6182DW	ACTIVE	SOIC	DW	20	25	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
THS6182DWG4	ACTIVE	SOIC	DW	20	25	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
THS6182DWP	ACTIVE	SO Power PAD	DWP	20	25	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
THS6182DWPG4	ACTIVE	SO Power PAD	DWP	20	25	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
THS6182DWPR	ACTIVE	SO Power PAD	DWP	20	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
THS6182DWPRG4	ACTIVE	SO Power PAD	DWP	20	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
THS6182DWR	ACTIVE	SOIC	DW	20	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
THS6182DWRG4	ACTIVE	SOIC	DW	20	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
THS6182RHFR	ACTIVE	QFN	RHF	24	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
THS6182RHFRG4	ACTIVE	QFN	RHF	24	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
THS6182RHFT	ACTIVE	QFN	RHF	24	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
THS6182RHFTG4	ACTIVE	QFN	RHF	24	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR

⁽¹⁾ The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

⁽²⁾ Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check <http://www.ti.com/productcontent> for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS

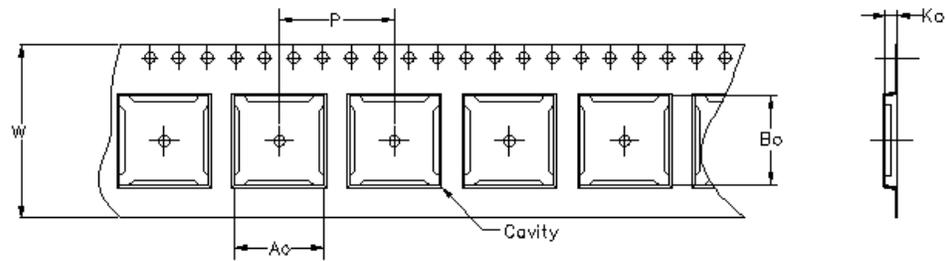
compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

⁽³⁾ MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

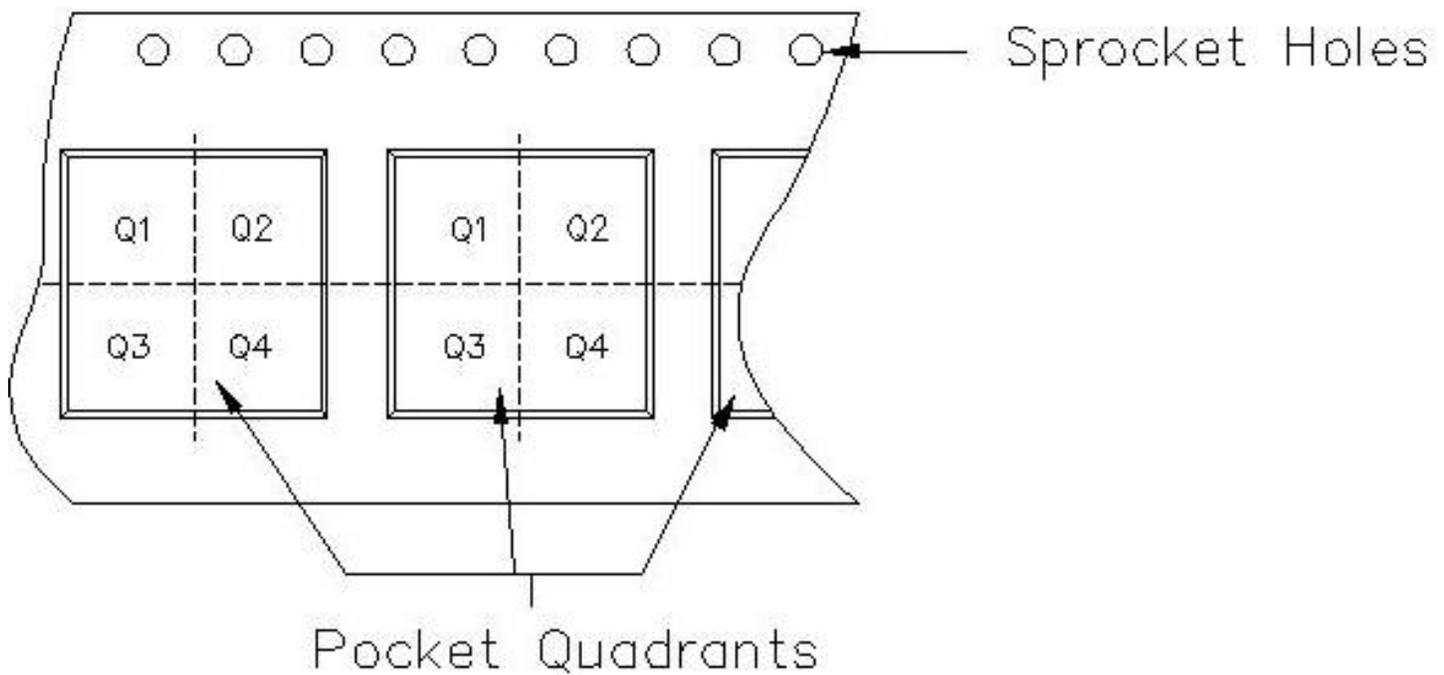
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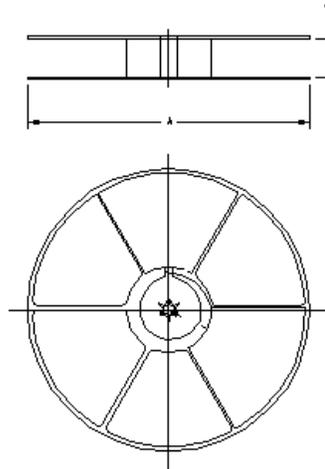
Carrier tape design is defined largely by the component length, width, and thickness.

A_o = Dimension designed to accommodate the component width.
B_o = Dimension designed to accommodate the component length.
K_o = Dimension designed to accommodate the component thickness.
W = Overall width of the carrier tape.
P = Pitch between successive cavity centers.



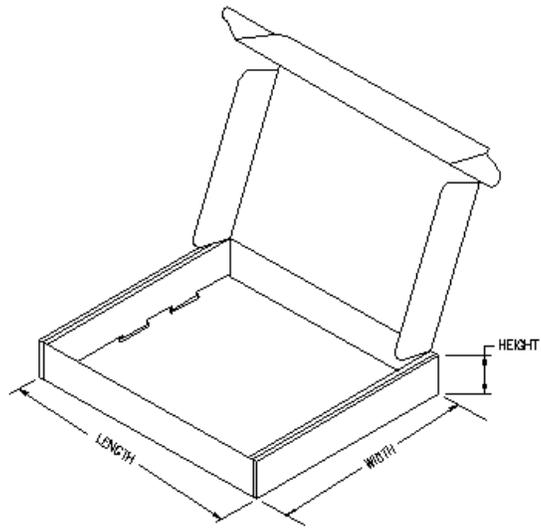
TAPE AND REEL INFORMATION

Device	Package	Pins	Site	Reel Diameter (mm)	Reel Width (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
THS6182DR	D	16	TAI	330	16	6.5	10.3	2.5	8	16	Q1
THS6182DWPR	DWP	20	TAI	330	24	10.8	13.3	2.7	12	24	Q1
THS6182DWR	DW	20	TAI	330	24	10.8	13.1	2.65	12	24	Q1
THS6182RHFR	RHF	24	TAI	330	12	4.3	5.3	1.3	8	12	NONE



TAPE AND REEL BOX INFORMATION

Device	Package	Pins	Site	Length (mm)	Width (mm)	Height (mm)
THS6182DR	D	16	TAI	346.0	346.0	33.0
THS6182DWPR	DWP	20	TAI	0.0	0.0	0.0
THS6182DWR	DW	20	TAI	346.0	346.0	41.0
THS6182RHFR	RHF	24	TAI	342.9	336.6	20.64

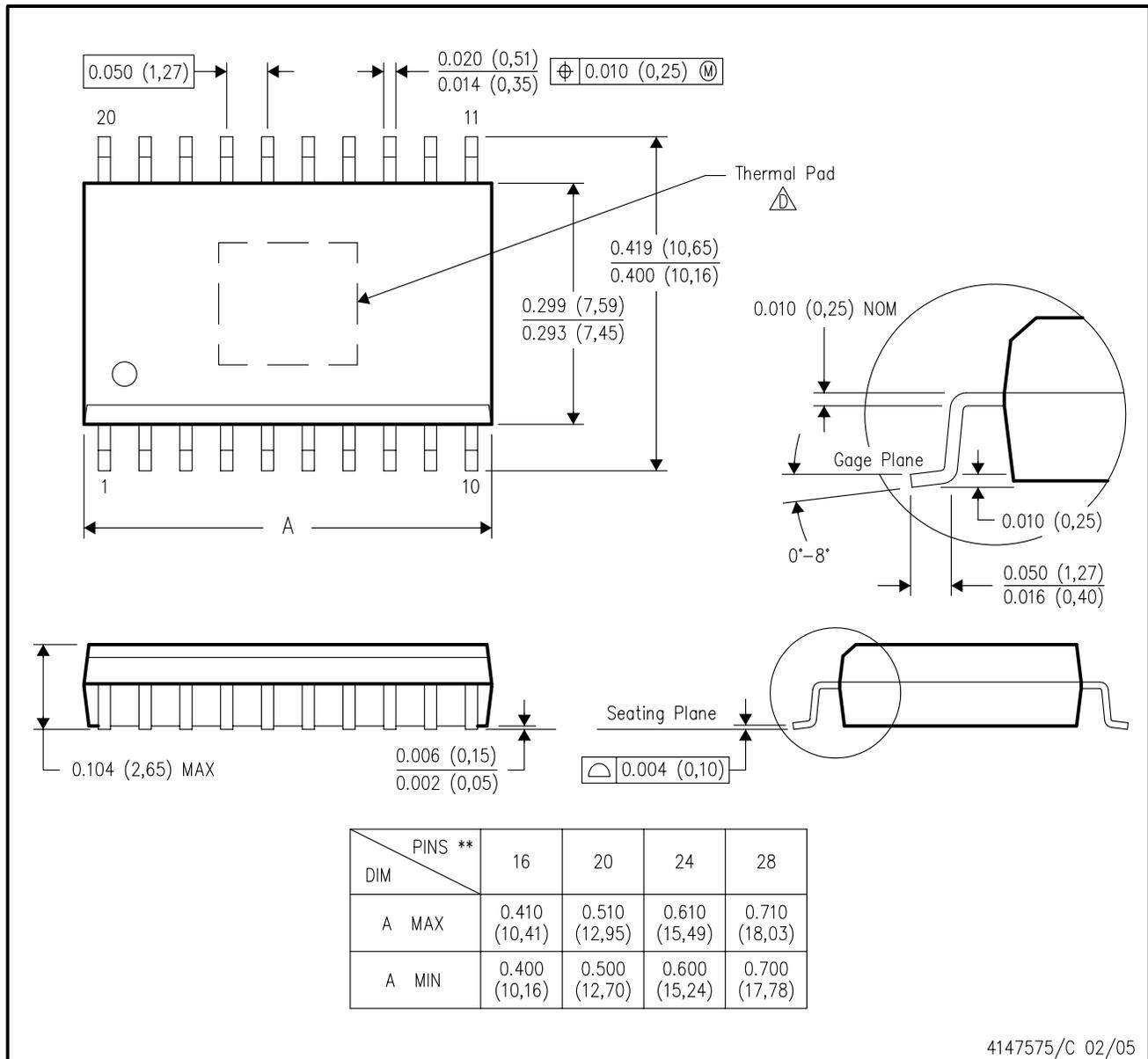


MECHANICAL DATA

DWP (R-PDSO-G**)

PowerPAD™ PLASTIC SMALL-OUTLINE PACKAGE

20 PINS SHOWN



- NOTES:
- All linear dimensions are in inches (millimeters).
 - This drawing is subject to change without notice.
 - Body dimensions do not include mold flash or protrusion not to exceed 0.006 (0,15).
- Δ This package is designed to be soldered to a thermal pad on the board. Refer to Technical Brief, PowerPAD Thermally Enhanced Package, Texas Instruments Literature No. SLMA002 for information regarding recommended board layout. This document is available at www.ti.com <<http://www.ti.com>>. See the product data sheet for details regarding the exposed thermal pad dimensions.

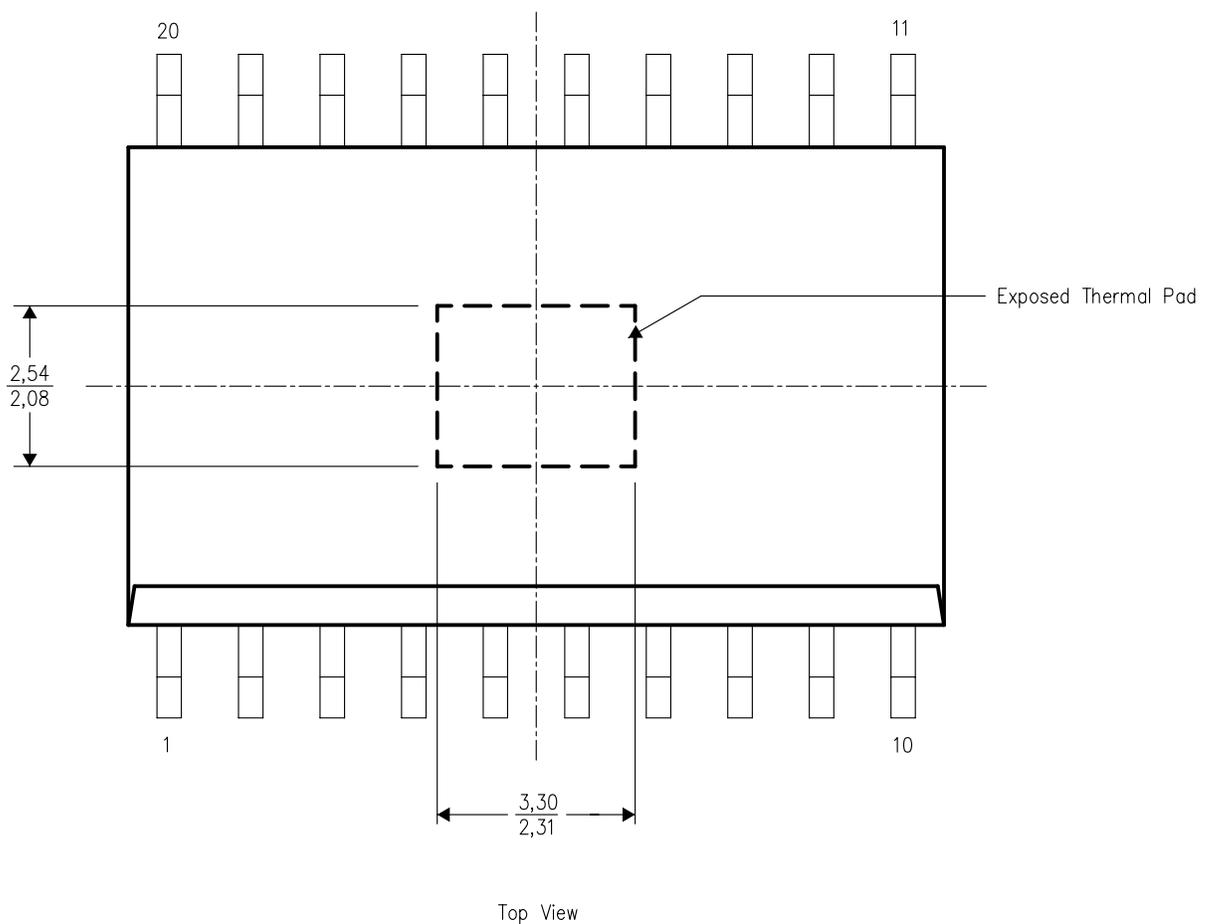
PowerPAD is a trademark of Texas Instruments.

THERMAL INFORMATION

This PowerPAD™ package incorporates an exposed thermal pad that is designed to be attached directly to an external heatsink. When the thermal pad is soldered directly to the printed circuit board (PCB), the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to a ground plane or special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For additional information on the PowerPAD package and how to take advantage of its heat dissipating abilities, refer to Technical Brief, PowerPAD Thermally Enhanced Package, Texas Instruments Literature No. SLMA002 and Application Brief, PowerPAD Made Easy, Texas Instruments Literature No. SLMA004. Both documents are available at www.ti.com.

The exposed thermal pad dimensions for this package are shown in the following illustration.

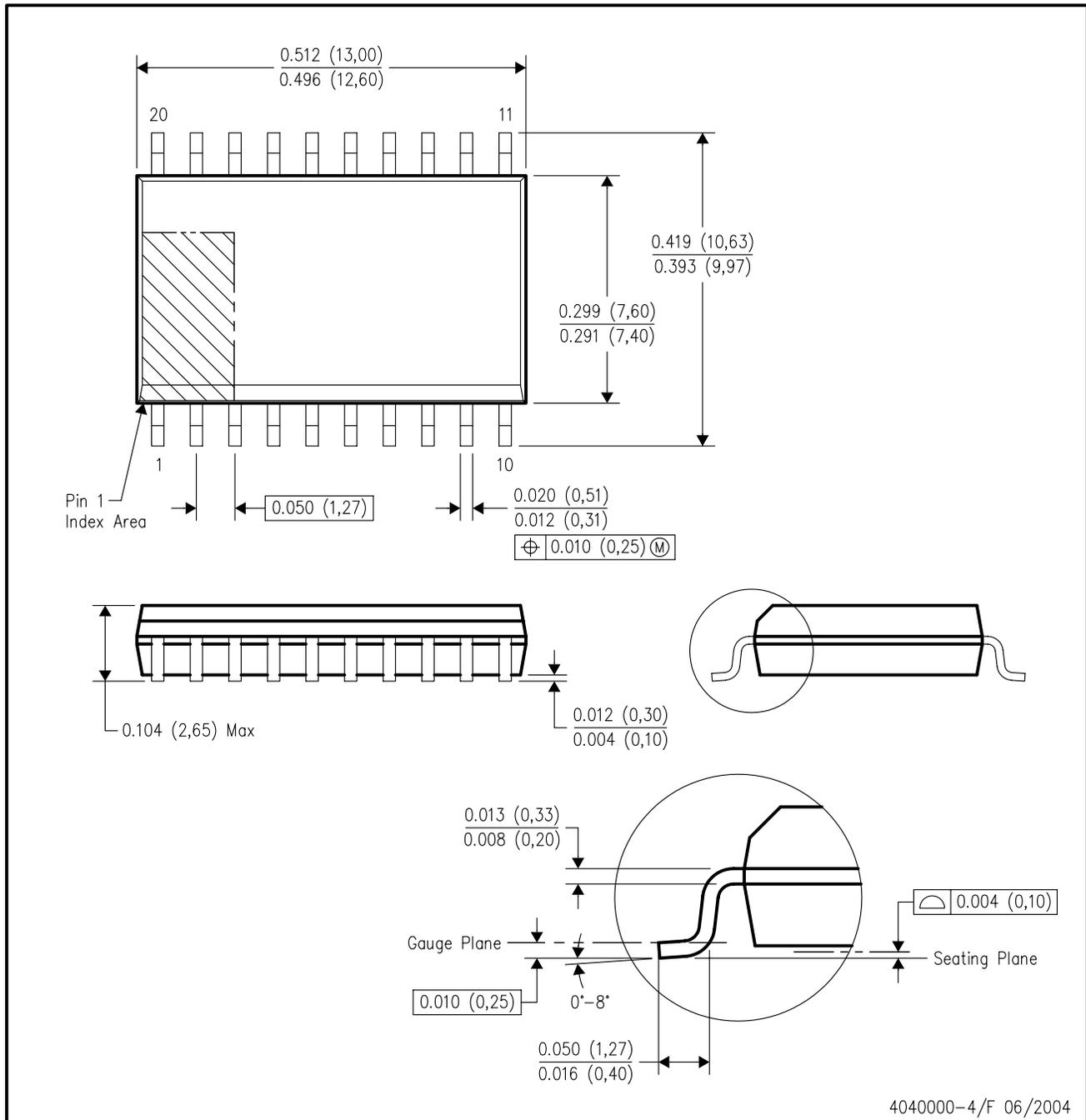


NOTE: All linear dimensions are in millimeters

Exposed Thermal Pad Dimensions

DW (R-PDSO-G20)

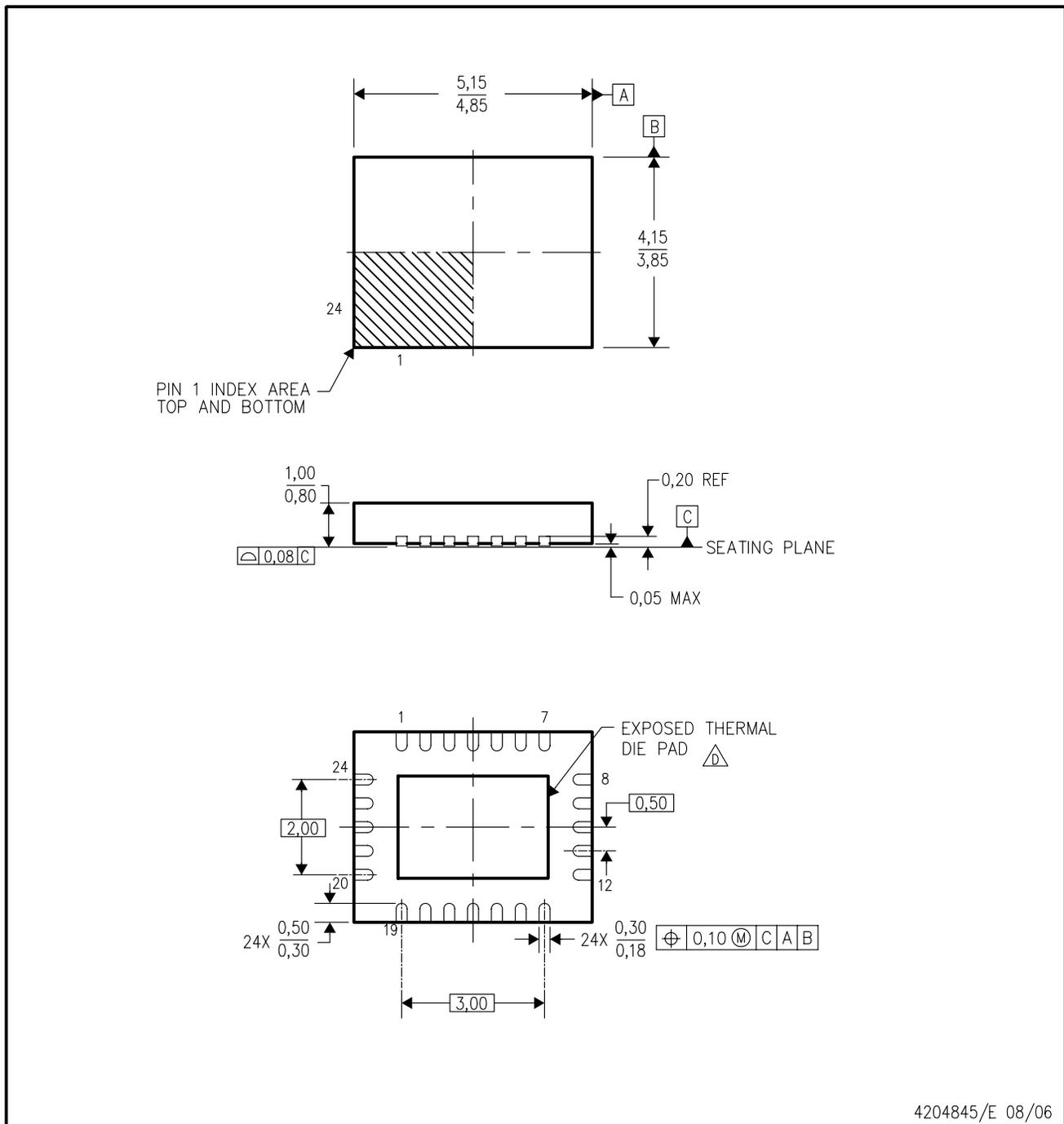
PLASTIC SMALL-OUTLINE PACKAGE



- NOTES:
- A. All linear dimensions are in inches (millimeters).
 - B. This drawing is subject to change without notice.
 - C. Body dimensions do not include mold flash or protrusion not to exceed 0.006 (0,15).
 - D. Falls within JEDEC MS-013 variation AC.

RHF (R-PQFP-N24)

PLASTIC QUAD FLATPACK



4204845/E 08/06

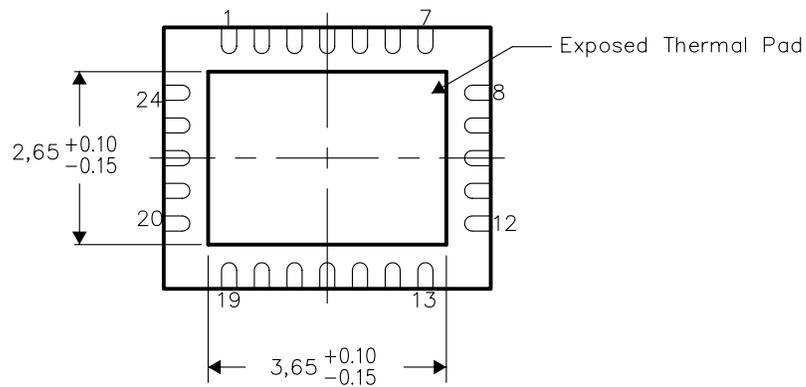
- NOTES:
- A. All linear dimensions are in millimeters.
 - B. This drawing is subject to change without notice.
 - C. QFN (Quad Flatpack No-Lead) Package configuration.
 - \triangle The package thermal pad must be soldered to the board for thermal and mechanical performance. See the Product Data Sheet for details regarding the exposed thermal pad dimensions.
 - E. Falls within JEDEC MO-220.

THERMAL INFORMATION

This package incorporates an exposed thermal pad that is designed to be attached directly to an external heatsink. The thermal pad must be soldered directly to the printed circuit board (PCB). After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to a ground or power plane (whichever is applicable), or alternatively, a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For information on the Quad Flatpack No-Lead (QFN) package and its advantages, refer to Application Report, Quad Flatpack No-Lead Logic Packages, Texas Instruments Literature No. SCBA017. This document is available at www.ti.com.

The exposed thermal pad dimensions for this package are shown in the following illustration.



Bottom View

NOTE: All linear dimensions are in millimeters

Exposed Thermal Pad Dimensions

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