



**CY7C1410AV18**  
**CY7C1425AV18**  
**CY7C1412AV18**  
**CY7C1414AV18**

## 36-Mbit QDR-II™ SRAM 2-Word Burst Architecture

### Features

- Separate Independent Read and Write data ports
  - Supports concurrent transactions
- 250-MHz clock for high bandwidth
- 2-Word Burst on all accesses
- Double Data Rate (DDR) interfaces on both Read and Write ports (data transferred at 500 MHz) @ 250 MHz
- Two input clocks (K and  $\bar{K}$ ) for precise DDR timing
  - SRAM uses rising edges only
- Two input clocks for output data (C and  $\bar{C}$ ) to minimize clock skew and flight-time mismatches
- Echo clocks (CQ and  $\bar{CQ}$ ) simplify data capture in high-speed systems
- Single multiplexed address input bus latches address inputs for both Read and Write ports
- Separate Port Selects for depth expansion
- Synchronous internally self timed writes
- Available in x8, x9, x18, and x36 configurations
- Full data coherency, providing most current data
- Core  $V_{DD} = 1.8V (\pm 0.1V)$ ; IO  $V_{DDQ} = 1.4V$  to  $V_{DD}$
- Available in 165-Ball FBGA package (15 x 17 x 1.4 mm)
- Offered in both Pb-free and non Pb-free packages
- Variable drive HSTL output buffers
- JTAG 1149.1 compatible test access port
- Delay Lock Loop (DLL) for accurate data placement

### Configurations

CY7C1410AV18 – 4M x 8

CY7C1425AV18 – 4M x 9

CY7C1412AV18 – 2M x 18

CY7C1414AV18 – 1M x 36

### Selection Guide

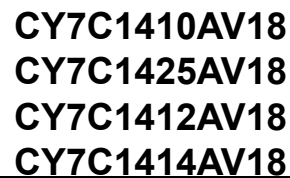
	250 MHz	200 MHz	167 MHz	Unit
Maximum Operating Frequency	250	200	167	MHz
Maximum Operating Current	1065	870	740	mA

### Functional Description

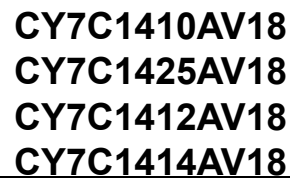
The CY7C1410AV18, CY7C1425AV18, CY7C1412AV18, and CY7C1414AV18 are 1.8V Synchronous Pipelined SRAMs, equipped with QDR™-II architecture. QDR-II architecture consists of two separate ports to access the memory array. The Read port has dedicated Data Outputs to support Read operations and the Write Port has dedicated Data Inputs to support Write operations. QDR-II architecture has separate data inputs and data outputs to completely eliminate the need to “turn-around” the data bus required with common IO devices. Access to each port is accomplished through a common address bus. The Read address is latched on the rising edge of the K clock and the Write address is latched on the rising edge of the  $\bar{K}$  clock. Accesses to the QDR-II Read and Write ports are completely independent of one another. In order to maximize data throughput, both Read and Write ports are equipped with Double Data Rate (DDR) interfaces. Each address location is associated with two 8-bit words (CY7C1410AV18) or 9-bit words (CY7C1425AV18) or 18-bit words (CY7C1412AV18) or 36-bit words (CY7C1414AV18) that burst sequentially into or out of the device. While data can be transferred into and out of the device on every rising edge of both input clocks (K and  $\bar{K}$  and C and  $\bar{C}$ ), memory bandwidth is maximized while simplifying system design by eliminating bus “turn-arounds.”

Depth expansion is accomplished with Port Selects for each port. Port selects allow each port to operate independently.

All synchronous inputs pass through input registers controlled by the K or  $\bar{K}$  input clocks. All data outputs pass through output registers controlled by the C or  $\bar{C}$  (or K or  $\bar{K}$  in a single clock domain) input clocks. Writes are conducted with on-chip synchronous self timed write circuitry.



The block diagram illustrates the internal structure of the 2M x 9-bit SRAM. It features two 2M x 9-bit arrays, each with its own write and read address decoders. The write address register is connected to the write address decoder, and the read address register is connected to the read address decoder. The control logic manages the memory's operation, receiving inputs from the address register, data register, and control signals. The data register is connected to the read data register, which outputs the data to the data bus. The control logic also manages the clock and data bus signals.

[illegible]

## Pin Configurations

### 165-Ball FBGA (15 x 17 x 1.4 mm) Pinout

#### CY7C1410AV18 (4M x 8)

	1	2	3	4	5	6	7	8	9	10	11
<b>A</b>	$\overline{\text{CQ}}$	NC/72M	A	$\overline{\text{WPS}}$	$\overline{\text{NWS}}_1$	$\overline{\text{K}}$	NC/144M	$\overline{\text{RPS}}$	A	A	CQ
<b>B</b>	NC	NC	NC	A	NC/288M	K	$\overline{\text{NWS}}_0$	A	NC	NC	Q3
<b>C</b>	NC	NC	NC	$V_{\text{SS}}$	A	A	A	$V_{\text{SS}}$	NC	NC	D3
<b>D</b>	NC	D4	NC	$V_{\text{SS}}$	$V_{\text{SS}}$	$V_{\text{SS}}$	$V_{\text{SS}}$	$V_{\text{SS}}$	NC	NC	NC
<b>E</b>	NC	NC	Q4	$V_{\text{DDQ}}$	$V_{\text{SS}}$	$V_{\text{SS}}$	$V_{\text{SS}}$	$V_{\text{DDQ}}$	NC	D2	Q2
<b>F</b>	NC	NC	NC	$V_{\text{DDQ}}$	$V_{\text{DD}}$	$V_{\text{SS}}$	$V_{\text{DD}}$	$V_{\text{DDQ}}$	NC	NC	NC
<b>G</b>	NC	D5	Q5	$V_{\text{DDQ}}$	$V_{\text{DD}}$	$V_{\text{SS}}$	$V_{\text{DD}}$	$V_{\text{DDQ}}$	NC	NC	NC
<b>H</b>	$\overline{\text{DOFF}}$	$V_{\text{REF}}$	$V_{\text{DDQ}}$	$V_{\text{DDQ}}$	$V_{\text{DD}}$	$V_{\text{SS}}$	$V_{\text{DD}}$	$V_{\text{DDQ}}$	$V_{\text{DDQ}}$	$V_{\text{REF}}$	ZQ
<b>J</b>	NC	NC	NC	$V_{\text{DDQ}}$	$V_{\text{DD}}$	$V_{\text{SS}}$	$V_{\text{DD}}$	$V_{\text{DDQ}}$	NC	Q1	D1
<b>K</b>	NC	NC	NC	$V_{\text{DDQ}}$	$V_{\text{DD}}$	$V_{\text{SS}}$	$V_{\text{DD}}$	$V_{\text{DDQ}}$	NC	NC	NC
<b>L</b>	NC	Q6	D6	$V_{\text{DDQ}}$	$V_{\text{SS}}$	$V_{\text{SS}}$	$V_{\text{SS}}$	$V_{\text{DDQ}}$	NC	NC	Q0
<b>M</b>	NC	NC	NC	$V_{\text{SS}}$	$V_{\text{SS}}$	$V_{\text{SS}}$	$V_{\text{SS}}$	$V_{\text{SS}}$	NC	NC	D0
<b>N</b>	NC	D7	NC	$V_{\text{SS}}$	A	A	A	$V_{\text{SS}}$	NC	NC	NC
<b>P</b>	NC	NC	Q7	A	A	C	A	A	NC	NC	NC
<b>R</b>	TDO	TCK	A	A	A	$\overline{\text{C}}$	A	A	A	TMS	TDI

#### CY7C1425AV18 (4M x 9)

	1	2	3	4	5	6	7	8	9	10	11
<b>A</b>	$\overline{\text{CQ}}$	NC/72M	A	$\overline{\text{WPS}}$	NC	$\overline{\text{K}}$	NC/144M	$\overline{\text{RPS}}$	A	A	CQ
<b>B</b>	NC	NC	NC	A	NC/288M	K	$\overline{\text{BWS}}_0$	A	NC	NC	Q4
<b>C</b>	NC	NC	NC	$V_{\text{SS}}$	A	A	A	$V_{\text{SS}}$	NC	NC	D4
<b>D</b>	NC	D5	NC	$V_{\text{SS}}$	$V_{\text{SS}}$	$V_{\text{SS}}$	$V_{\text{SS}}$	$V_{\text{SS}}$	NC	NC	NC
<b>E</b>	NC	NC	Q5	$V_{\text{DDQ}}$	$V_{\text{SS}}$	$V_{\text{SS}}$	$V_{\text{SS}}$	$V_{\text{DDQ}}$	NC	D3	Q3
<b>F</b>	NC	NC	NC	$V_{\text{DDQ}}$	$V_{\text{DD}}$	$V_{\text{SS}}$	$V_{\text{DD}}$	$V_{\text{DDQ}}$	NC	NC	NC
<b>G</b>	NC	D6	Q6	$V_{\text{DDQ}}$	$V_{\text{DD}}$	$V_{\text{SS}}$	$V_{\text{DD}}$	$V_{\text{DDQ}}$	NC	NC	NC
<b>H</b>	$\overline{\text{DOFF}}$	$V_{\text{REF}}$	$V_{\text{DDQ}}$	$V_{\text{DDQ}}$	$V_{\text{DD}}$	$V_{\text{SS}}$	$V_{\text{DD}}$	$V_{\text{DDQ}}$	$V_{\text{DDQ}}$	$V_{\text{REF}}$	ZQ
<b>J</b>	NC	NC	NC	$V_{\text{DDQ}}$	$V_{\text{DD}}$	$V_{\text{SS}}$	$V_{\text{DD}}$	$V_{\text{DDQ}}$	NC	Q2	D2
<b>K</b>	NC	NC	NC	$V_{\text{DDQ}}$	$V_{\text{DD}}$	$V_{\text{SS}}$	$V_{\text{DD}}$	$V_{\text{DDQ}}$	NC	NC	NC
<b>L</b>	NC	Q7	D7	$V_{\text{DDQ}}$	$V_{\text{SS}}$	$V_{\text{SS}}$	$V_{\text{SS}}$	$V_{\text{DDQ}}$	NC	NC	Q1
<b>M</b>	NC	NC	NC	$V_{\text{SS}}$	$V_{\text{SS}}$	$V_{\text{SS}}$	$V_{\text{SS}}$	$V_{\text{SS}}$	NC	NC	D1
<b>N</b>	NC	D8	NC	$V_{\text{SS}}$	A	A	A	$V_{\text{SS}}$	NC	NC	NC
<b>P</b>	NC	NC	Q8	A	A	C	A	A	NC	D0	Q0
<b>R</b>	TDO	TCK	A	A	A	$\overline{\text{C}}$	A	A	A	TMS	TDI

**Pin Configurations** (continued)

**165-Ball FBGA (15 x 17 x 1.4 mm) Pinout**

**CY7C1412AV18 (2M x 18)**

	1	2	3	4	5	6	7	8	9	10	11
<b>A</b>	$\overline{\text{CQ}}$	NC/144M	A	$\overline{\text{WPS}}$	$\overline{\text{BWS}}_1$	$\overline{\text{K}}$	NC/288M	$\overline{\text{RPS}}$	A	NC/72M	CQ
<b>B</b>	NC	Q9	D9	A	NC	K	$\overline{\text{BWS}}_0$	A	NC	NC	Q8
<b>C</b>	NC	NC	D10	$V_{\text{SS}}$	A	A	A	$V_{\text{SS}}$	NC	Q7	D8
<b>D</b>	NC	D11	Q10	$V_{\text{SS}}$	$V_{\text{SS}}$	$V_{\text{SS}}$	$V_{\text{SS}}$	$V_{\text{SS}}$	NC	NC	D7
<b>E</b>	NC	NC	Q11	$V_{\text{DDQ}}$	$V_{\text{SS}}$	$V_{\text{SS}}$	$V_{\text{SS}}$	$V_{\text{DDQ}}$	NC	D6	Q6
<b>F</b>	NC	Q12	D12	$V_{\text{DDQ}}$	$V_{\text{DD}}$	$V_{\text{SS}}$	$V_{\text{DD}}$	$V_{\text{DDQ}}$	NC	NC	Q5
<b>G</b>	NC	D13	Q13	$V_{\text{DDQ}}$	$V_{\text{DD}}$	$V_{\text{SS}}$	$V_{\text{DD}}$	$V_{\text{DDQ}}$	NC	NC	D5
<b>H</b>	$\overline{\text{DOFF}}$	$V_{\text{REF}}$	$V_{\text{DDQ}}$	$V_{\text{DDQ}}$	$V_{\text{DD}}$	$V_{\text{SS}}$	$V_{\text{DD}}$	$V_{\text{DDQ}}$	$V_{\text{DDQ}}$	$V_{\text{REF}}$	ZQ
<b>J</b>	NC	NC	D14	$V_{\text{DDQ}}$	$V_{\text{DD}}$	$V_{\text{SS}}$	$V_{\text{DD}}$	$V_{\text{DDQ}}$	NC	Q4	D4
<b>K</b>	NC	NC	Q14	$V_{\text{DDQ}}$	$V_{\text{DD}}$	$V_{\text{SS}}$	$V_{\text{DD}}$	$V_{\text{DDQ}}$	NC	D3	Q3
<b>L</b>	NC	Q15	D15	$V_{\text{DDQ}}$	$V_{\text{SS}}$	$V_{\text{SS}}$	$V_{\text{SS}}$	$V_{\text{DDQ}}$	NC	NC	Q2
<b>M</b>	NC	NC	D16	$V_{\text{SS}}$	$V_{\text{SS}}$	$V_{\text{SS}}$	$V_{\text{SS}}$	$V_{\text{SS}}$	NC	Q1	D2
<b>N</b>	NC	D17	Q16	$V_{\text{SS}}$	A	A	A	$V_{\text{SS}}$	NC	NC	D1
<b>P</b>	NC	NC	Q17	A	A	C	A	A	NC	D0	Q0
<b>R</b>	TDO	TCK	A	A	A	$\overline{\text{C}}$	A	A	A	TMS	TDI

**CY7C1414AV18 (1M x 36)**

	1	2	3	4	5	6	7	8	9	10	11
<b>A</b>	$\overline{\text{CQ}}$	NC/288M	NC/72M	$\overline{\text{WPS}}$	$\overline{\text{BWS}}_2$	$\overline{\text{K}}$	$\overline{\text{BWS}}_1$	$\overline{\text{RPS}}$	A	NC/144M	CQ
<b>B</b>	Q27	Q18	D18	A	$\overline{\text{BWS}}_3$	K	$\overline{\text{BWS}}_0$	A	D17	Q17	Q8
<b>C</b>	D27	Q28	D19	$V_{\text{SS}}$	A	A	A	$V_{\text{SS}}$	D16	Q7	D8
<b>D</b>	D28	D20	Q19	$V_{\text{SS}}$	$V_{\text{SS}}$	$V_{\text{SS}}$	$V_{\text{SS}}$	$V_{\text{SS}}$	Q16	D15	D7
<b>E</b>	Q29	D29	Q20	$V_{\text{DDQ}}$	$V_{\text{SS}}$	$V_{\text{SS}}$	$V_{\text{SS}}$	$V_{\text{DDQ}}$	Q15	D6	Q6
<b>F</b>	Q30	Q21	D21	$V_{\text{DDQ}}$	$V_{\text{DD}}$	$V_{\text{SS}}$	$V_{\text{DD}}$	$V_{\text{DDQ}}$	D14	Q14	Q5
<b>G</b>	D30	D22	Q22	$V_{\text{DDQ}}$	$V_{\text{DD}}$	$V_{\text{SS}}$	$V_{\text{DD}}$	$V_{\text{DDQ}}$	Q13	D13	D5
<b>H</b>	$\overline{\text{DOFF}}$	$V_{\text{REF}}$	$V_{\text{DDQ}}$	$V_{\text{DDQ}}$	$V_{\text{DD}}$	$V_{\text{SS}}$	$V_{\text{DD}}$	$V_{\text{DDQ}}$	$V_{\text{DDQ}}$	$V_{\text{REF}}$	ZQ
<b>J</b>	D31	Q31	D23	$V_{\text{DDQ}}$	$V_{\text{DD}}$	$V_{\text{SS}}$	$V_{\text{DD}}$	$V_{\text{DDQ}}$	D12	Q4	D4
<b>K</b>	Q32	D32	Q23	$V_{\text{DDQ}}$	$V_{\text{DD}}$	$V_{\text{SS}}$	$V_{\text{DD}}$	$V_{\text{DDQ}}$	Q12	D3	Q3
<b>L</b>	Q33	Q24	D24	$V_{\text{DDQ}}$	$V_{\text{SS}}$	$V_{\text{SS}}$	$V_{\text{SS}}$	$V_{\text{DDQ}}$	D11	Q11	Q2
<b>M</b>	D33	Q34	D25	$V_{\text{SS}}$	$V_{\text{SS}}$	$V_{\text{SS}}$	$V_{\text{SS}}$	$V_{\text{SS}}$	D10	Q1	D2
<b>N</b>	D34	D26	Q25	$V_{\text{SS}}$	A	A	A	$V_{\text{SS}}$	Q10	D9	D1
<b>P</b>	Q35	D35	Q26	A	A	C	A	A	Q9	D0	Q0
<b>R</b>	TDO	TCK	A	A	A	$\overline{\text{C}}$	A	A	A	TMS	TDI

## Pin Definitions

Pin Name	IO	Pin Description
$D_{[x:0]}$	Input-Synchronous	<b>Data input signals, sampled on the rising edge of K and <math>\bar{K}</math> clocks during valid write operations.</b> CY7C1410AV18 - $D_{[7:0]}$ CY7C1425AV18 - $D_{[8:0]}$ CY7C1412AV18 - $D_{[17:0]}$ CY7C1414AV18 - $D_{[35:0]}$
WPS	Input-Synchronous	<b>Write Port Select, Active LOW.</b> Sampled on the rising edge of the K clock. When asserted active, a Write operation is initiated. Deasserting deselects the Write port. Deselecting the Write port causes $D_{[x:0]}$ to be ignored.
$NWS_0, NWS_1$		<b>Nibble Write Select 0, 1 – Active LOW. (CY7C1410AV18 Only)</b> Sampled on the rising edge of the K and $\bar{K}$ clocks during Write operations. Used to select which nibble is written into the device during the current portion of the Write operations. Nibbles not written remain unaltered. $NWS_0$ controls $D_{[3:0]}$ and $NWS_1$ controls $D_{[7:4]}$ . All Nibble Write Selects are sampled on the same edge as the data. Deselecting a Nibble Write Select causes the corresponding nibble of data to be ignored and not written into the device.
$\overline{BWS_0}, \overline{BWS_1}, \overline{BWS_2}, \overline{BWS_3}$	Input-Synchronous	<b>Byte Write Select 0, 1, 2 and 3 – Active LOW.</b> Sampled on the rising edge of the K and $\bar{K}$ clocks during Write operations. Used to select which byte is written into the device during the current portion of the Write operations. Bytes not written remain unaltered. CY7C1425AV18 – $\overline{BWS_0}$ controls $D_{[8:0]}$ CY7C1412AV18 – $\overline{BWS_0}$ controls $D_{[8:0]}$ , $\overline{BWS_1}$ controls $D_{[17:9]}$ CY7C1414AV18 – $\overline{BWS_0}$ controls $D_{[8:0]}$ , $\overline{BWS_1}$ controls $D_{[17:9]}$ , $\overline{BWS_2}$ controls $D_{[26:18]}$ and $\overline{BWS_3}$ controls $D_{[35:27]}$ . All the Byte Write Selects are sampled on the same edge as the data. Deselecting a Byte Write Select causes the corresponding byte of data to be ignored and not written into the device.
A	Input-Synchronous	<b>Address Inputs.</b> Sampled on the rising edge of the K (Read address) and $\bar{K}$ (Write address) clocks during active Read and Write operations. These address inputs are multiplexed for both Read and Write operations. Internally, the device is organized as 4M x 8 (2 arrays each of 2M x 8) for CY7C1410AV18, 4M x 9 (2 arrays each of 2M x 9) for CY7C1425AV18, 2M x 18 (2 arrays each of 1M x 18) for CY7C1412AV18 and 1M x 36 (2 arrays each of 512K x 36) for CY7C1414AV18. Therefore, only 21 address inputs are needed to access the entire memory array of CY7C1410AV18 and CY7C1425AV18, 20 address inputs for CY7C1412AV18 and 19 address inputs for CY7C1414AV18. These inputs are ignored when the appropriate port is deselected.
$Q_{[x:0]}$	Outputs-Synchronous	<b>Data Output signals.</b> These pins drive out the requested data during a Read operation. Valid data is driven out on the rising edge of both the C and $\bar{C}$ clocks during Read operations or K and $\bar{K}$ when in single clock mode. When the Read port is deselected, $Q_{[x:0]}$ are automatically tri-stated. CY7C1410AV18 – $Q_{[7:0]}$ CY7C1425AV18 – $Q_{[8:0]}$ CY7C1412AV18 – $Q_{[17:0]}$ CY7C1414AV18 – $Q_{[35:0]}$
RPS	Input-Synchronous	<b>Read Port Select, Active LOW.</b> Sampled on the rising edge of Positive Input Clock (K). When active, a Read operation is initiated. Deasserting causes the Read port to be deselected. When deselected, the pending access is allowed to complete and the output drivers are automatically tri-stated following the next rising edge of the C clock. Each read access consists of a burst of two sequential transfers.
C	Input-Clock	<b>Positive Input Clock for Output Data.</b> C is used in conjunction with $\bar{C}$ to clock out the Read data from the device. C and $\bar{C}$ can be used together to deskew the flight times of various devices on the board back to the controller. See application example for further details.
$\bar{C}$	Input-Clock	<b>Negative Input Clock for Output Data.</b> $\bar{C}$ is used in conjunction with C to clock out the Read data from the device. C and $\bar{C}$ can be used together to deskew the flight times of various devices on the board back to the controller. See application example for further details.

## Pin Definitions (continued)

Pin Name	IO	Pin Description
K	Input-Clock	<b>Positive Input Clock Input.</b> The rising edge of K is used to capture synchronous inputs to the device and to drive out data through $Q_{[x:0]}$ when in single clock mode. All accesses are initiated on the rising edge of K.
$\bar{K}$	Input-Clock	<b>Negative Input Clock Input.</b> $\bar{K}$ is used to capture synchronous inputs being presented to the device and to drive out data through $Q_{[x:0]}$ when in single clock mode.
CQ	Echo Clock	<b>CQ is referenced with respect to C.</b> This is a free running clock and is synchronized to the Input clock for output data (C) of the QDR-II. In the single clock mode, CQ is generated with respect to K. The timings for the echo clocks are shown in the AC Timing table.
$\bar{CQ}$	Echo Clock	<b><math>\bar{CQ}</math> is referenced with respect to <math>\bar{C}</math>.</b> This is a free running clock and is synchronized to the Input clock for output data ( $\bar{C}$ ) of the QDR-II. In the single clock mode, $\bar{CQ}$ is generated with respect to K. The timings for the echo clocks are shown in the AC Timing table.
ZQ	Input	<b>Output Impedance Matching Input.</b> This input is used to tune the device outputs to the system data bus impedance. CQ, $\bar{CQ}$ , and $Q_{[x:0]}$ output impedance are set to $0.2 \times RQ$ , where RQ is a resistor connected between ZQ and ground. Alternately, this pin can be connected directly to $V_{DDQ}$ , which enables the minimum impedance mode. This pin cannot be connected directly to GND or left unconnected.
$\overline{DOFF}$	Input	<b>DLL Turn Off, Active LOW.</b> Connecting this pin to ground turns off the DLL inside the device. The timings in the DLL turned off operation is different from those listed in this data sheet.
TDO	Output	<b>TDO for JTAG.</b>
TCK	Input	<b>TCK pin for JTAG.</b>
TDI	Input	<b>TDI pin for JTAG.</b>
TMS	Input	<b>TMS pin for JTAG.</b>
NC	N/A	<b>Not connected to the die.</b> Can be tied to any voltage level.
NC/72M	N/A	<b>Not connected to the die.</b> Can be tied to any voltage level.
NC/144M	N/A	<b>Not connected to the die.</b> Can be tied to any voltage level.
NC/288M	N/A	<b>Not connected to the die.</b> Can be tied to any voltage level.
$V_{REF}$	Input-Reference	<b>Reference Voltage Input.</b> Static input used to set the reference level for HSTL inputs and Outputs as well as AC measurement points.
$V_{DD}$	Power Supply	<b>Power supply inputs to the core of the device.</b>
$V_{SS}$	Ground	<b>Ground for the device.</b>
$V_{DDQ}$	Power Supply	<b>Power supply inputs for the outputs of the device.</b>

## Functional Overview

The CY7C1410AV18, CY7C1425AV18, CY7C1412AV18 and CY7C1414AV18 are synchronous pipelined Burst SRAMs equipped with both a Read port and a Write port. The Read port is dedicated to Read operations and the Write port is dedicated to Write operations. Data flows into the SRAM through the Write port and out through the Read port. These devices multiplex the address inputs in order to minimize the number of address pins required. By having separate Read and Write ports, the QDR-II completely eliminates the need to “turn-around” the data bus and avoids any possible data contention, thereby simplifying system design. Each access consists of two 8-bit data transfers in the case of CY7C1410AV18, two 9-bit data transfers in the case of CY7C1425AV18, two 18-bit data transfers in the case of CY7C1412AV18 and two 36-bit data transfers in the case of CY7C1414AV18, in one clock cycle.

Accesses for both ports are initiated on the rising edge of the positive Input Clock (K). All synchronous input timings are referenced from the rising edge of the input clocks (K and  $\bar{K}$ ) and all output timings are referenced to the rising edge of output clocks (C and  $\bar{C}$  or K and  $\bar{K}$  when in single clock mode).

All synchronous data inputs ( $D_{[x:0]}$ ) inputs pass through input registers controlled by the input clocks (K and  $\bar{K}$ ). All synchronous data outputs ( $Q_{[x:0]}$ ) outputs pass through output registers controlled by the rising edge of the output clocks (C and  $\bar{C}$  or K and  $\bar{K}$  when in single clock mode).

All synchronous control ( $\overline{RPS}$ ,  $\overline{WPS}$ ,  $\overline{BWS}_{[x:0]}$ ) inputs pass through input registers controlled by the rising edge of the input clocks (K and  $\bar{K}$ ).

CY7C1412AV18 is described in the following sections. The same basic descriptions apply to CY7C1410AV18, CY7C1425AV18, and CY7C1414AV18.



## Read Operations

The CY7C1412AV18 is organized internally as 2 arrays of 1Mx18. Accesses are completed in a burst of two sequential 18-bit data words. Read operations are initiated by asserting RPS active at the rising edge of the Positive Input Clock (K). The address is latched on the rising edge of the K Clock. The address presented to Address inputs is stored in the Read address register. Following the next K clock rise the corresponding lowest order 18-bit word of data is driven onto the  $Q_{[17:0]}$  using C as the output timing reference. On the subsequent rising edge of C, the next 18-bit data word is driven onto the  $Q_{[17:0]}$ . The requested data is valid 0.45 ns from the rising edge of the output clock (C and C or K and K when in single clock mode).

Synchronous internal circuitry automatically tri-states the outputs following the next rising edge of the Output Clocks (C/C). This allows for a seamless transition between devices without the insertion of wait states in a depth expanded memory.

## Write Operations

Write operations are initiated by asserting  $\overline{WPS}$  active at the rising edge of the Positive Input Clock (K). On the same K clock rise, the data presented to  $D_{[17:0]}$  is latched and stored into the lower 18-bit Write Data register provided  $BWS_{[1:0]}$  are both asserted active. On the subsequent rising edge of the Negative Input Clock ( $\overline{K}$ ), the address is latched and the information presented to  $D_{[17:0]}$  is stored into the Write Data register provided  $BWS_{[1:0]}$  are both asserted active. The 36 bits of data are then written into the memory array at the specified location. When deselected, the write port ignores all inputs after the pending Write operations have been completed.

## Byte Write Operations

Byte Write operations are supported by the CY7C1412AV18. A Write operation is initiated as described in the Write Operations section above. The bytes that are written are determined by  $BWS_0$  and  $BWS_1$ , which are sampled with each 18-bit data word. Asserting the appropriate Byte Write Select input during the data portion of a Write allows the data being presented to be latched and written into the device. Deasserting the Byte Write Select input during the data portion of a write allows the data stored in the device for that byte to remain unaltered. This feature can be used to simplify Read/Modify/Write operations to a Byte Write operation.

## Single Clock Mode

The CY7C1412AV18 can be used with a single clock that controls both the input and output registers. In this mode, the device recognizes only a single pair of input clocks (K and  $\overline{K}$ ) that control both the input and output registers. This operation is identical to the operation if the device had zero skew between the K/ $\overline{K}$  and C/ $\overline{C}$  clocks. All timing parameters remain the same in this mode. To use this mode of operation, the user must tie C and  $\overline{C}$  HIGH at power on. This function is a strap option and not alterable during device operation.

## Concurrent Transactions

The Read and Write ports on the CY7C1412AV18 operate completely independently of one another. While each port latches the address inputs on different clock edges, the user can Read or Write to any location, regardless of the transaction on the other port. Also, reads and writes can be started in the same clock cycle. If the ports access the same location at the same time, the SRAM delivers the most recent information associated with the specified address location. This includes forwarding data from a Write cycle that was initiated on the previous K clock rise.

## Depth Expansion

The CY7C1412AV18 has a Port Select input for each port. This allows for easy depth expansion. Both Port Selects are sampled on the rising edge of the Positive Input Clock only (K). Each port select input can deselect the specified port. Deselecting a port does not affect the other port. All pending transactions (Read and Write) are completed prior to the device being deselected.

## Programmable Impedance

An external resistor, RQ, must be connected between the ZQ pin on the SRAM and  $V_{SS}$  to allow the SRAM to adjust its output driver impedance. The value of RQ must be 5x the value of the intended line impedance driven by the SRAM. The allowable range of RQ to guarantee impedance matching with a tolerance of  $\pm 15\%$  is between 175 $\Omega$  and 350 $\Omega$ , with  $V_{DDQ} = 1.5V$ . The output impedance is adjusted every 1024 cycles upon power up to account for drifts in supply voltage and temperature.

## Echo Clocks

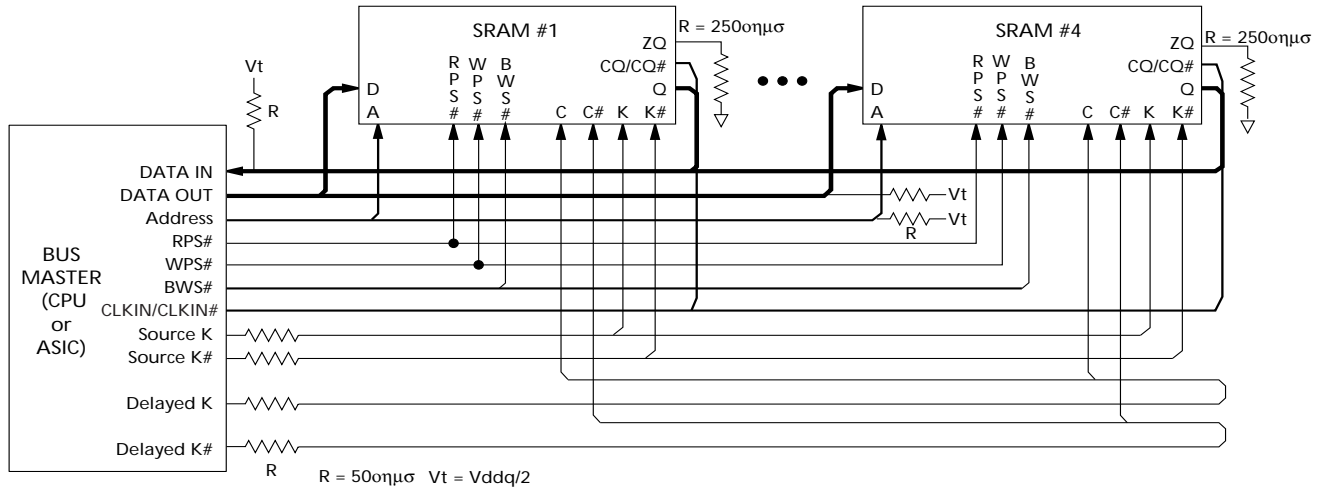
Echo clocks are provided on the QDR-II to simplify data capture on high-speed systems. Two echo clocks are generated by the QDR-II. CQ is referenced with respect to C and  $\overline{CQ}$  is referenced with respect to  $\overline{C}$ . These are free-running clocks and are synchronized to the output clock (C/ $\overline{C}$ ) of the QDR-II. In the single clock mode, CQ is generated with respect to K and  $\overline{CQ}$  is generated with respect to  $\overline{K}$ . The timings for the echo clocks are shown in the AC Timing table.

## DLL

These chips use a Delay Lock Loop (DLL) that is designed to function between 80 MHz and the specified maximum clock frequency. During power up, when the  $\overline{DOFF}$  is tied HIGH, the DLL gets locked after 1024 cycles of stable clock. The DLL can also be reset by slowing or stopping the input clock K and  $\overline{K}$  for a minimum of 30 ns. However, it is not necessary for the DLL to be specifically reset in order to lock the DLL to the desired frequency. The DLL automatically locks 1024 clock cycles after a stable clock is presented. The DLL may be disabled by applying ground to the  $\overline{DOFF}$  pin. For information refer to the application note "DLL Considerations in QDRII/DDRII/QDRII+/DDRII+".



## Application Example<sup>[1]</sup>



## Truth Table<sup>[2, 3, 4, 5, 6, 7]</sup>

Operation	K	$\overline{RPS}$	$\overline{WPS}$	DQ	DQ
Write Cycle: Load address on the rising edge of $\overline{K}$ clock; input write data on K and $\overline{K}$ rising edges.	L-H	X	L	D(A + 0) at K(t) $\uparrow$	D(A + 1) at $\overline{K}(t)$ $\uparrow$
Read Cycle: Load address on the rising edge of K clock; wait one and a half cycle; read data on C and $\overline{C}$ rising edges.	L-H	L	X	Q(A + 0) at $\overline{C}(t + 1)$ $\uparrow$	Q(A + 1) at C(t + 2) $\uparrow$
NOP: No Operation	L-H	H	H	D = X Q = High-Z	D = X Q = High-Z
Standby: Clock Stopped	Stopped	X	X	Previous State	Previous State

## Write Cycle Descriptions (CY7C1410AV18 and CY7C1412AV18) <sup>[2, 8]</sup>

$\overline{BWS}_0/\overline{NWS}_0$	$\overline{BWS}_1/\overline{NWS}_1$	K	$\overline{K}$	Comments
L	L	L-H	—	During the data portion of a write sequence: CY7C1410AV18 – both nibbles (D <sub>[7:0]</sub> ) are written into the device, CY7C1412AV18 – both bytes (D <sub>[17:0]</sub> ) are written into the device.
L	L	—	L-H	During the data portion of a write sequence: CY7C1410AV18 – both nibbles (D <sub>[7:0]</sub> ) are written into the device, CY7C1412AV18 – both bytes (D <sub>[17:0]</sub> ) are written into the device.
L	H	L-H	—	During the data portion of a write sequence: CY7C1410AV18 – only the lower nibble (D <sub>[3:0]</sub> ) is written into the device. D <sub>[7:4]</sub> remains unaltered, CY7C1412AV18 – only the lower byte (D <sub>[8:0]</sub> ) is written into the device. D <sub>[17:9]</sub> remains unaltered.

### Notes:

- The above application shows four QDR-II being used.
- X = "Don't Care," H = Logic HIGH, L = Logic LOW,  $\uparrow$  represents rising edge.
- Device powers up deselected and the outputs in a tri-state condition.
- "A" represents address location latched by the devices when transaction was initiated. A + 00, A + 01 represents the internal address sequence in the burst.
- "t" represents the cycle at which a Read/Write operation is started. t + 1 and t + 2 are the first and second clock cycles respectively succeeding the "t" clock cycle.
- Data inputs are registered at K and  $\overline{K}$  rising edges. Data outputs are delivered on C and  $\overline{C}$  rising edges, except when in single clock mode.
- It is recommended that K =  $\overline{K}$  and C =  $\overline{C}$  = HIGH when clock is stopped. This is not essential, but permits most rapid restart by overcoming transmission line charging symmetrically.
- Assumes a Write cycle was initiated according to the Write Port Cycle Description Truth Table.  $\overline{NWS}_0$ ,  $\overline{NWS}_1$ ,  $\overline{BWS}_0$ ,  $\overline{BWS}_1$ ,  $\overline{BWS}_2$  and  $\overline{BWS}_3$  can be altered on different portions of a Write cycle, as long as the setup and hold requirements are achieved.

**Write Cycle Descriptions** (CY7C1410AV18 and CY7C1412AV18) (continued)<sup>[2, 8]</sup>

$\overline{\text{BWS}}_0/\overline{\text{NWS}}_0$	$\overline{\text{BWS}}_1/\overline{\text{NWS}}_1$	K	$\overline{\text{K}}$	Comments
L	H	–	L-H	During the data portion of a write sequence: CY7C1410AV18 – only the lower nibble ( $\text{D}_{[3:0]}$ ) is written into the device. $\text{D}_{[7:4]}$ remains unaltered, CY7C1412AV18 – only the lower byte ( $\text{D}_{[8:0]}$ ) is written into the device. $\text{D}_{[17:9]}$ remains unaltered.
H	L	L-H	–	During the data portion of a write sequence: CY7C1410AV18 – only the upper nibble ( $\text{D}_{[7:4]}$ ) is written into the device. $\text{D}_{[3:0]}$ remains unaltered, CY7C1412AV18 – only the upper byte ( $\text{D}_{[17:9]}$ ) is written into the device. $\text{D}_{[8:0]}$ remains unaltered.
H	L	–	L-H	During the data portion of a write sequence: CY7C1410AV18 – only the upper nibble ( $\text{D}_{[7:4]}$ ) is written into the device. $\text{D}_{[3:0]}$ remains unaltered, CY7C1412AV18 – only the upper byte ( $\text{D}_{[17:9]}$ ) is written into the device. $\text{D}_{[8:0]}$ remains unaltered.
H	H	L-H	–	No data is written into the devices during this portion of a write operation.
H	H	–	L-H	No data is written into the devices during this portion of a write operation.

**Write Cycle Descriptions** (CY7C1414AV18) <sup>[2, 8]</sup>

$\overline{\text{BWS}}_0$	$\overline{\text{BWS}}_1$	$\overline{\text{BWS}}_2$	$\overline{\text{BWS}}_3$	K	$\overline{\text{K}}$	Comments
L	L	L	L	L-H	–	During the data portion of a write sequence, all four bytes ( $\text{D}_{[35:0]}$ ) are written into the device.
L	L	L	L	–	L-H	During the data portion of a write sequence, all four bytes ( $\text{D}_{[35:0]}$ ) are written into the device.
L	H	H	H	L-H	–	During the data portion of a write sequence, only the lower byte ( $\text{D}_{[8:0]}$ ) is written into the device. $\text{D}_{[35:9]}$ remains unaltered.
L	H	H	H	–	L-H	During the data portion of a write sequence, only the lower byte ( $\text{D}_{[8:0]}$ ) is written into the device. $\text{D}_{[35:9]}$ remains unaltered.
H	L	H	H	L-H	–	During the data portion of a write sequence, only the byte ( $\text{D}_{[17:9]}$ ) is written into the device. $\text{D}_{[8:0]}$ and $\text{D}_{[35:18]}$ remains unaltered.
H	L	H	H	–	L-H	During the data portion of a write sequence, only the byte ( $\text{D}_{[17:9]}$ ) is written into the device. $\text{D}_{[8:0]}$ and $\text{D}_{[35:18]}$ remains unaltered.
H	H	L	H	L-H	–	During the data portion of a write sequence, only the byte ( $\text{D}_{[26:18]}$ ) is written into the device. $\text{D}_{[17:0]}$ and $\text{D}_{[35:27]}$ remains unaltered.
H	H	L	H	–	L-H	During the data portion of a write sequence, only the byte ( $\text{D}_{[26:18]}$ ) is written into the device. $\text{D}_{[17:0]}$ and $\text{D}_{[35:27]}$ remains unaltered.
H	H	H	L	L-H		During the data portion of a write sequence, only the byte ( $\text{D}_{[35:27]}$ ) is written into the device. $\text{D}_{[26:0]}$ remains unaltered.
H	H	H	L	–	L-H	During the data portion of a write sequence, only the byte ( $\text{D}_{[35:27]}$ ) is written into the device. $\text{D}_{[26:0]}$ remains unaltered.
H	H	H	H	L-H	–	No data is written into the device during this portion of a write operation.
H	H	H	H	–	L-H	No data is written into the device during this portion of a write operation.

**Write Cycle Descriptions** (CY7C1425AV18)

$\overline{\text{BWS}}_0$	K	$\overline{\text{K}}$	Comments
L	L-H	–	During the data portion of a write sequence: CY7C1425AV18 - the single byte ( $\text{D}_{[8:0]}$ ) is written into the device
L	–	L-H	During the data portion of a write sequence: CY7C1425AV18 - the single byte ( $\text{D}_{[8:0]}$ ) is written into the device
H	L-H	–	No data is written into the devices during this portion of a Write operation.
H	–	L-H	No data is written into the devices during this portion of a Write operation.

## IEEE 1149.1 Serial Boundary Scan (JTAG)

These SRAMs incorporate a serial boundary scan test access port (TAP) in the FBGA package. This part is fully compliant with IEEE Standard #1149.1-1900. The TAP operates using JEDEC standard 1.8V IO logic levels.

### Disabling the JTAG Feature

It is possible to operate the SRAM without using the JTAG feature. To disable the TAP controller, TCK must be tied LOW ( $V_{SS}$ ) to prevent clocking of the device. TDI and TMS are internally pulled up and may be unconnected. They may alternately be connected to  $V_{DD}$  through a pull up resistor. TDO must be left unconnected. Upon power up, the device is up in a reset state which does not interfere with the operation of the device.

### Test Access Port—Test Clock

The test clock is used only with the TAP controller. All inputs are captured on the rising edge of TCK. All outputs are driven from the falling edge of TCK.

### Test Mode Select

The TMS input is used to give commands to the TAP controller and is sampled on the rising edge of TCK. It is allowable to leave this pin unconnected if the TAP is not used. The pin is pulled up internally, resulting in a logic HIGH level.

### Test Data-In (TDI)

The TDI pin is used to serially input information into the registers and can be connected to the input of any of the registers. The register between TDI and TDO is chosen by the instruction that is loaded into the TAP instruction register. For information on loading the instruction register, see the TAP Controller State Diagram. TDI is internally pulled up and can be unconnected if the TAP is unused in an application. TDI is connected to the most significant bit (MSB) on any register.

### Test Data-Out (TDO)

The TDO output pin is used to serially clock data-out from the registers. The output is active depending upon the current state of the TAP state machine (see Instruction codes). The output changes on the falling edge of TCK. TDO is connected to the least significant bit (LSB) of any register.

### Performing a TAP Reset

A Reset is performed by forcing TMS HIGH ( $V_{DD}$ ) for five rising edges of TCK. This RESET does not affect the operation of the SRAM and may be performed while the SRAM is operating. At power up, the TAP is reset internally to ensure that TDO comes up in a high-Z state.

### TAP Registers

Registers are connected between the TDI and TDO pins and allow data to be scanned into and out of the SRAM test circuitry. Only one register can be selected at a time through the instruction registers. Data is serially loaded into the TDI pin on the rising edge of TCK. Data is output on the TDO pin on the falling edge of TCK.

#### Instruction Register

Three-bit instructions can be serially loaded into the instruction register. This register is loaded when it is placed between the TDI and TDO pins as shown in TAP Controller Block Diagram.

Upon power up, the instruction register is loaded with the IDCODE instruction. It is also loaded with the IDCODE instruction if the controller is placed in a reset state as described in the previous section.

When the TAP controller is in the Capture IR state, the two least significant bits are loaded with a binary “01” pattern to allow for fault isolation of the board level serial test path.

#### Bypass Register

To save time when serially shifting data through registers, it is sometimes advantageous to skip certain chips. The bypass register is a single-bit register that can be placed between TDI and TDO pins. This allows data to be shifted through the SRAM with minimal delay. The bypass register is set LOW ( $V_{SS}$ ) when the BYPASS instruction is executed.

#### Boundary Scan Register

The boundary scan register is connected to all of the input and output pins on the SRAM. Several no connect (NC) pins are also included in the scan register to reserve pins for higher density devices.

The boundary scan register is loaded with the contents of the RAM Input and Output ring when the TAP controller is in the Capture-DR state and is then placed between the TDI and TDO pins when the controller is moved to the Shift-DR state. The EXTEST, SAMPLE/PRELOAD and SAMPLE Z instructions can be used to capture the contents of the Input and Output ring.

The Boundary Scan Order tables show the order in which the bits are connected. Each bit corresponds to one of the bumps on the SRAM package. The MSB of the register is connected to TDI, and the LSB is connected to TDO.

#### Identification (ID) Register

The ID register is loaded with a vendor-specific, 32-bit code during the Capture-DR state when the IDCODE command is loaded in the instruction register. The IDCODE is hardwired into the SRAM and can be shifted out when the TAP controller is in the Shift-DR state. The ID register has a vendor code and other information described in the Identification Register Definitions table.

### TAP Instruction Set

Eight different instructions are possible with the three-bit instruction register. All combinations are listed in the Instruction Code table. Three of these instructions are listed as RESERVED and must not be used. The other five instructions are described in detail below.

Instructions are loaded into the TAP controller during the Shift-IR state when the instruction register is placed between TDI and TDO. During this state, instructions are shifted through the instruction register through the TDI and TDO pins. To execute the instruction after it is shifted in, the TAP controller needs to be moved into the Update-IR state.

#### IDCODE

The IDCODE instruction causes a vendor-specific, 32-bit code to be loaded into the instruction register. It also places the instruction register between the TDI and TDO pins and allows the IDCODE to be shifted out of the device when the TAP controller enters the Shift-DR state. The IDCODE instruction

is loaded into the instruction register upon power up or whenever the TAP controller is supplied a test logic reset state.

#### **SAMPLE Z**

The SAMPLE Z instruction causes the boundary scan register to be connected between the TDI and TDO pins when the TAP controller is in a Shift-DR state. The SAMPLE Z command puts the output bus into a High-Z state until the next command is supplied during the "Update IR" state.

#### **SAMPLE/PRELOAD**

SAMPLE/PRELOAD is a 1149.1 mandatory instruction. When the SAMPLE/PRELOAD instructions are loaded into the instruction register and the TAP controller is in the Capture-DR state, a snapshot of data on the inputs and output pins is captured in the boundary scan register.

The user must be aware that the TAP controller clock can only operate at a frequency up to 20 MHz, while the SRAM clock operates more than an order of magnitude faster. Because there is a large difference in the clock frequencies, it is possible that during the Capture-DR state, an input or output undergoes a transition. The TAP may then try to capture a signal while in transition (metastable state). This does not harm the device, but there is no guarantee as to the value that is captured. Repeatable results may not be possible.

To guarantee that the boundary scan register captures the correct value of a signal, the SRAM signal must be stabilized long enough to meet the TAP controller's capture setup plus hold times ( $t_{CS}$  and  $t_{CH}$ ). The SRAM clock input might not be captured correctly if there is no way in a design to stop (or slow) the clock during a SAMPLE/PRELOAD instruction. If this is an issue, it is still possible to capture all other signals and simply ignore the value of the CK and  $\overline{CK}$  captured in the boundary scan register.

After the data is captured, it is possible to shift out the data by putting the TAP into the Shift-DR state. This places the boundary scan register between the TDI and TDO pins.

PRELOAD allows an initial data pattern to be placed at the latched parallel outputs of the boundary scan register cells prior to the selection of another boundary scan test operation.

The shifting of data for the SAMPLE and PRELOAD phases can occur concurrently when required—that is, while data captured is shifted out, the preloaded data can be shifted in.

#### **BYPASS**

When the BYPASS instruction is loaded in the instruction register and the TAP is placed in a Shift-DR state, the bypass register is placed between the TDI and TDO pins. The advantage of the BYPASS instruction is that it shortens the boundary scan path when multiple devices are connected together on a board.

#### **EXTEST**

The EXTEST instruction enables the preloaded data to be driven out through the system output pins. This instruction also selects the boundary scan register to be connected for serial access between the TDI and TDO in the shift-DR controller state.

#### **EXTEST OUTPUT BUS TRI-STATE**

IEEE Standard 1149.1 mandates that the TAP controller be able to put the output bus into a tri-state mode.

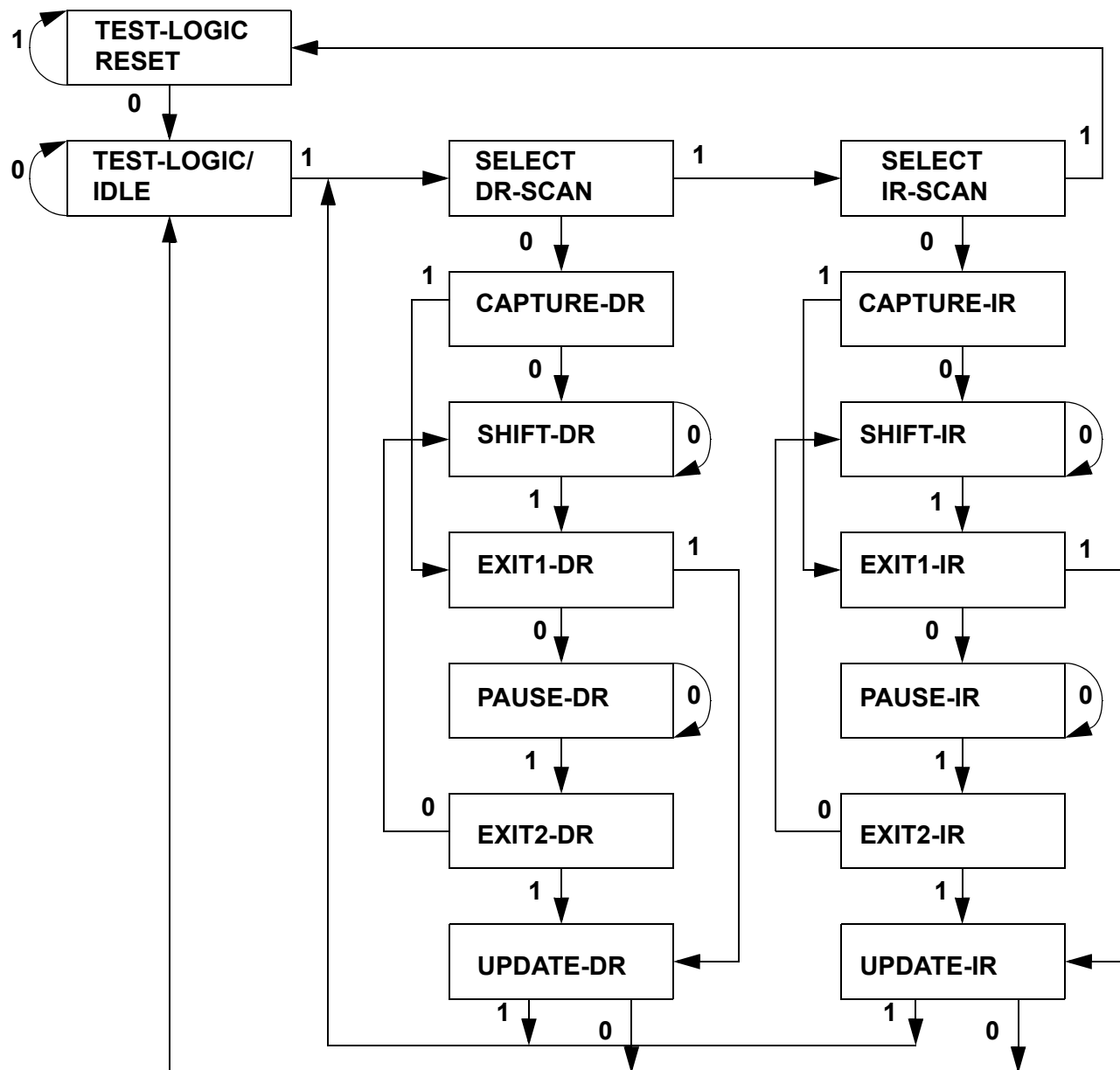
The boundary scan register has a special bit located at bit #108. When this scan cell, called the "extest output bus tri-state," is latched into the preload register during the "Update-DR" state in the TAP controller, it directly controls the state of the output (Q-bus) pins, when the EXTEST is entered as the current instruction. When HIGH, it enables the output buffers to drive the output bus. When LOW, this bit places the output bus into a High-Z condition.

This bit can be set by entering the SAMPLE/PRELOAD or EXTEST command, and then shifting the desired bit into that cell, during the "Shift-DR" state. During "Update-DR", the value loaded into that shift-register cell latches into the preload register. When the EXTEST instruction is entered, this bit directly controls the output Q-bus pins. Note that this bit is pre-set LOW to enable the output when the device is powered-up, and also when the TAP controller is in the "Test-Logic-Reset" state.

#### **Reserved**

These instructions are not implemented but are reserved for future use. Do not use these instructions.

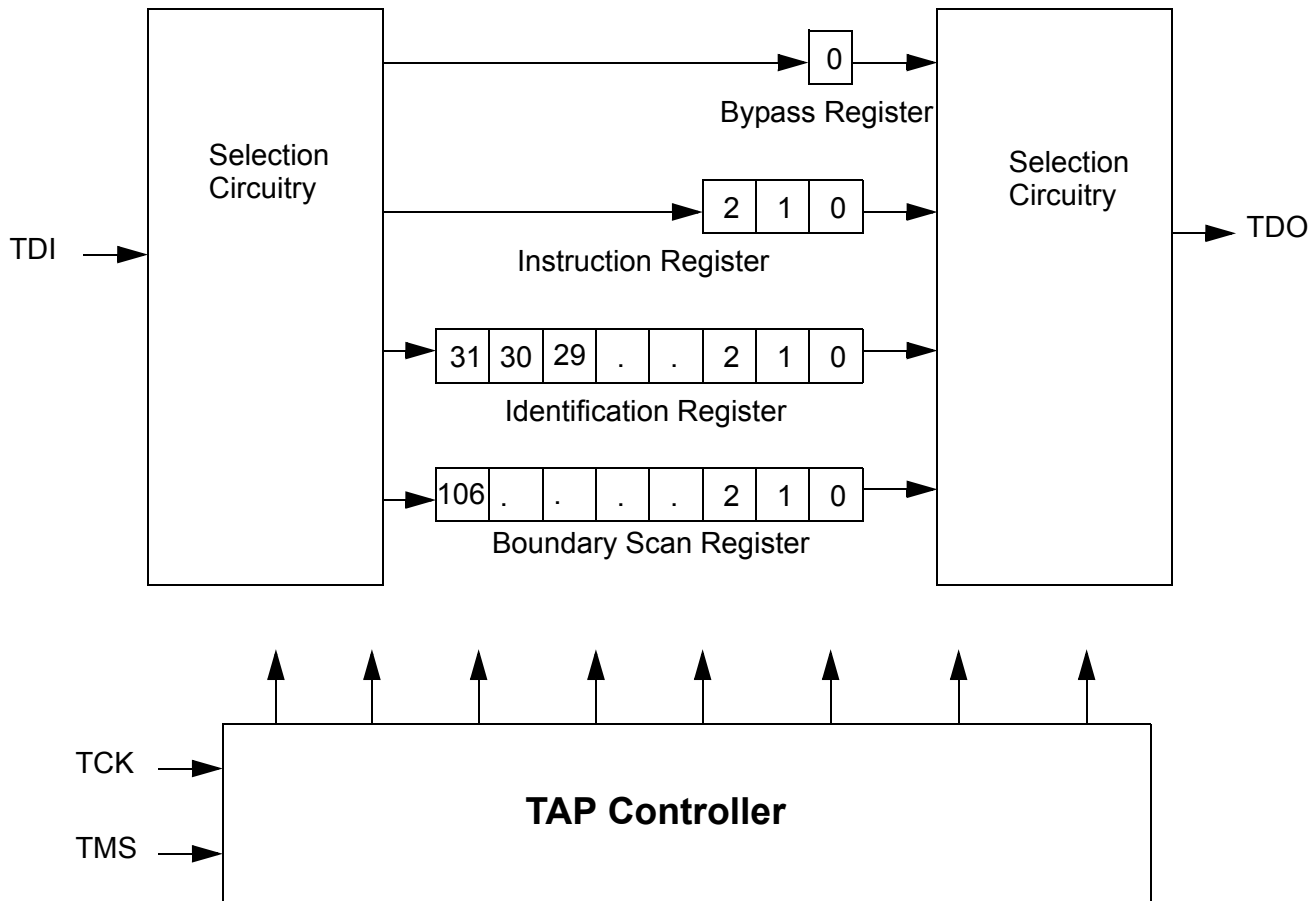
# TAP Controller State Diagram<sup>[9]</sup>



## Note:

9. The 0/1 next to each state represents the value at TMS at the rising edge of TCK.

## TAP Controller Block Diagram



## TAP Electrical Characteristics Over the Operating Range<sup>[15, 18, 10]</sup>

Parameter	Description	Test Conditions	Min	Max	Unit
V <sub>OH1</sub>	Output HIGH Voltage	I <sub>OH</sub> = -2.0 mA	1.4		V
V <sub>OH2</sub>	Output HIGH Voltage	I <sub>OH</sub> = -100 µA	1.6		V
V <sub>OL1</sub>	Output LOW Voltage	I <sub>OL</sub> = 2.0 mA		0.4	V
V <sub>OL2</sub>	Output LOW Voltage	I <sub>OL</sub> = 100 µA		0.2	V
V <sub>IH</sub>	Input HIGH Voltage		0.65V <sub>DD</sub>	V <sub>DD</sub> + 0.3	V
V <sub>IL</sub>	Input LOW Voltage		-0.3	0.35V <sub>DD</sub>	V
I <sub>X</sub>	Input and Output Load Current	GND ≤ V <sub>I</sub> ≤ V <sub>DD</sub>	-5	5	µA

**Note:**

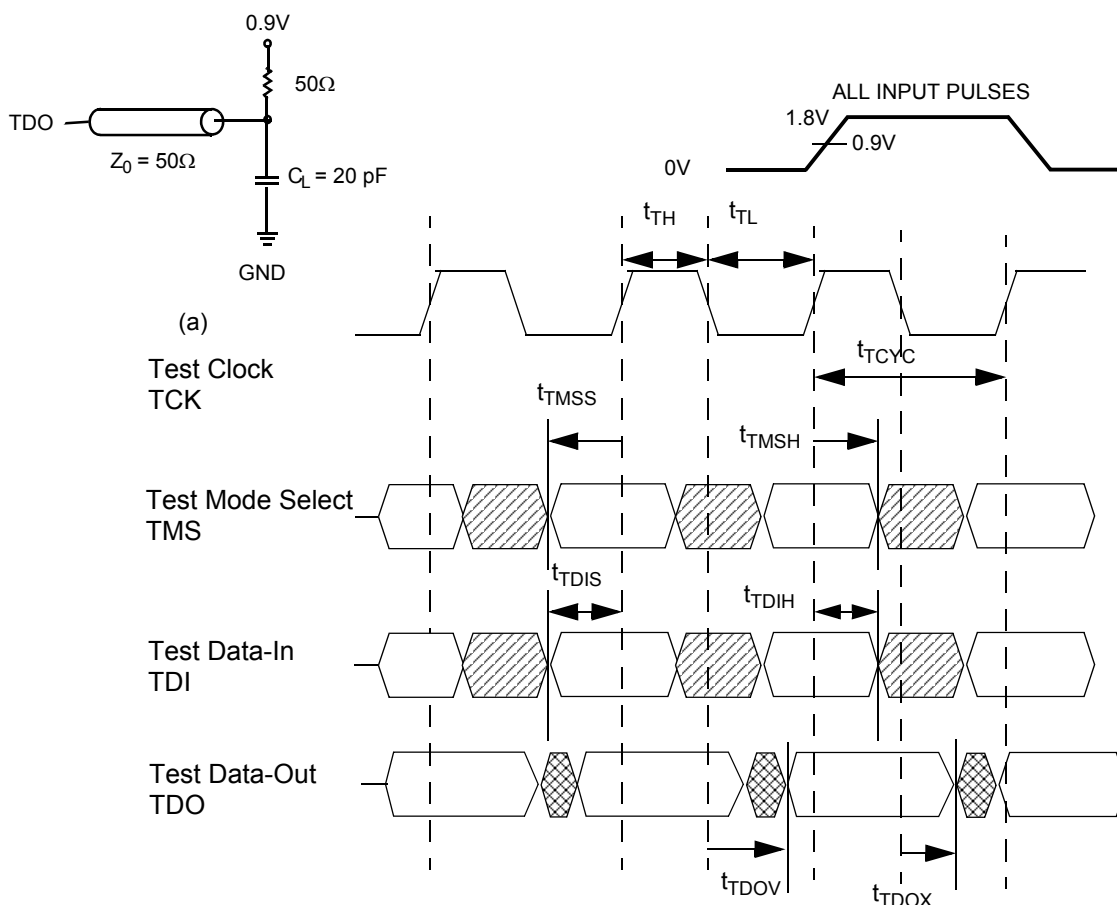
10. These characteristic pertain to the TAP inputs (TMS, TCK, TDI and TDO). Parallel load levels are specified in the Electrical Characteristics table.



## TAP AC Switching Characteristics Over the Operating Range<sup>[11, 12]</sup>

Parameter	Description	Min	Max	Unit
$t_{TCYC}$	TCK Clock Cycle Time	50		ns
$t_{TF}$	TCK Clock Frequency		20	MHz
$t_{TH}$	TCK Clock HIGH	20		ns
$t_{TL}$	TCK Clock LOW	20		ns
<b>Setup Times</b>				
$t_{TMSS}$	TMS Setup to TCK Clock Rise	5		ns
$t_{TDIS}$	TDI Setup to TCK Clock Rise	5		ns
$t_{CS}$	Capture Setup to TCK Rise	5		ns
<b>Hold Times</b>				
$t_{TMSH}$	TMS Hold after TCK Clock Rise	5		ns
$t_{TDIH}$	TDI Hold after Clock Rise	5		ns
$t_{CH}$	Capture Hold after Clock Rise	5		ns
<b>Output Times</b>				
$t_{TDOV}$	TCK Clock LOW to TDO Valid		10	ns
$t_{TDOX}$	TCK Clock LOW to TDO Invalid	0		ns

## TAP Timing and Test Conditions<sup>[12]</sup>



### Notes:

11.  $t_{CS}$  and  $t_{CH}$  refer to the setup and hold time requirements of latching data from the boundary scan register.  
12. Test conditions are specified using the load in TAP AC test conditions.  $t_R/t_F = 1$  ns.

## Identification Register Definitions

Instruction Field	Value				Description
	CY7C1410AV18	CY7C1425AV18	CY7C1412AV18	CY7C1414AV18	
Revision Number (31:29)	000	000	000	000	Version number.
Cypress Device ID (28:12)	11010011010000111	11010011010000111	11010011010010111	11010011010100111	Defines the type of SRAM.
Cypress JEDEC ID (11:1)	00000110100	00000110100	00000110100	00000110100	Unique identification of SRAM vendor.
ID Register Presence (0)	1	1	1	1	Indicates the presence of an ID register.

## Scan Register Sizes

Register Name	Bit Size
Instruction	3
Bypass	1
ID	32
Boundary Scan Cells	109

## Instruction Codes

Instruction	Code	Description
EXTEST	000	Captures the Input/Output ring contents.
IDCODE	001	Loads the ID register with the vendor ID code and places the register between TDI and TDO. This operation does not affect SRAM operation.
SAMPLE Z	010	Captures the Input/Output contents. Places the boundary scan register between TDI and TDO. Forces all SRAM output drivers to a High-Z state.
RESERVED	011	Do Not Use: This instruction is reserved for future use.
SAMPLE/PRELOAD	100	Captures the Input/Output ring contents. Places the boundary scan register between TDI and TDO. Does not affect the SRAM operation.
RESERVED	101	Do Not Use: This instruction is reserved for future use.
RESERVED	110	Do Not Use: This instruction is reserved for future use.
BYPASS	111	Places the bypass register between TDI and TDO. This operation does not affect SRAM operation.

# Boundary Scan Order

Bit #	Bump ID	Bit #	Bump ID	Bit #	Bump ID	Bit #	Bump ID
0	6R	28	10G	56	6A	84	1J
1	6P	29	9G	57	5B	85	2J
2	6N	30	11F	58	5A	86	3K
3	7P	31	11G	59	4A	87	3J
4	7N	32	9F	60	5C	88	2K
5	7R	33	10F	61	4B	89	1K
6	8R	34	11E	62	3A	90	2L
7	8P	35	10E	63	2A	91	3L
8	9R	36	10D	64	1A	92	1M
9	11P	37	9E	65	2B	93	1L
10	10P	38	10C	66	3B	94	3N
11	10N	39	11D	67	1C	95	3M
12	9P	40	9C	68	1B	96	1N
13	10M	41	9D	69	3D	97	2M
14	11N	42	11B	70	3C	98	3P
15	9M	43	11C	71	1D	99	2N
16	9N	44	9B	72	2C	100	2P
17	11L	45	10B	73	3E	101	1P
18	11M	46	11A	74	2D	102	3R
19	9L	47	10A	75	2E	103	4R
20	10L	48	9A	76	1E	104	4P
21	11K	49	8B	77	2F	105	5P
22	10K	50	7C	78	3F	106	5N
23	9J	51	6C	79	1G	107	5R
24	9K	52	8A	80	1F	108	Internal
25	10J	53	7A	81	3G		
26	11J	54	7B	82	2G		
27	11H	55	6B	83	1H		

## Power Up Sequence in QDR-II SRAM<sup>[13, 14]</sup>

QDR-II SRAMs must be powered up and initialized in a predefined manner to prevent undefined operations.

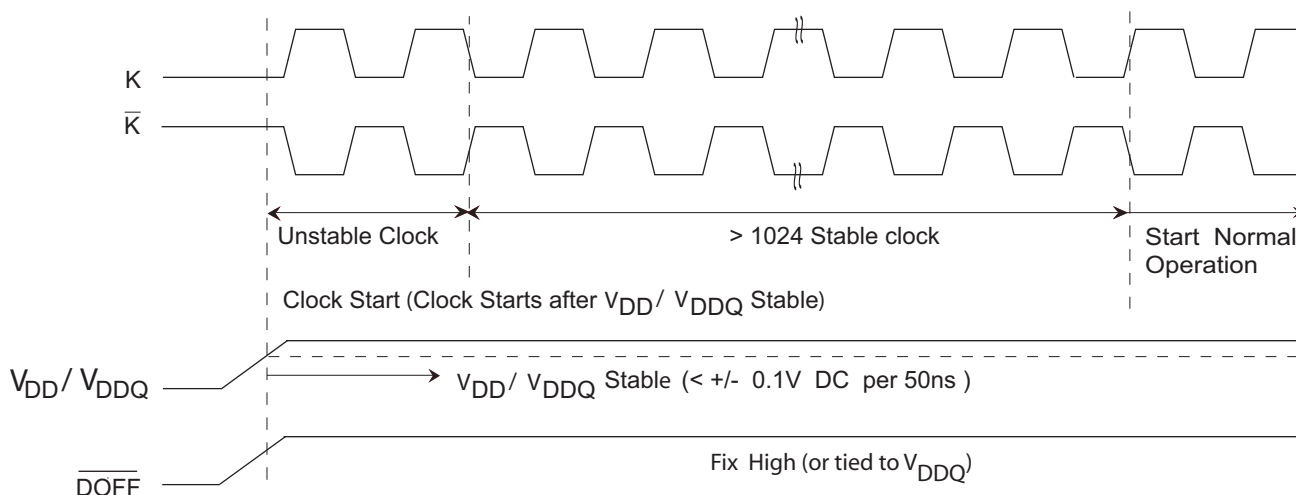
### Power Up Sequence

- Apply power and drive  $\overline{\text{DOFF}}$  LOW (All other inputs can be HIGH or LOW)
  - Apply  $V_{DD}$  before  $V_{DDQ}$
  - Apply  $V_{DDQ}$  before  $V_{REF}$  or at the same time as  $V_{REF}$
- After the power and clock (K,  $\overline{K}$ , C,  $\overline{C}$ ) are stable take  $\overline{\text{DOFF}}$  HIGH
- The additional 1024 cycles of clocks are required for the DLL to lock.

### DLL Constraints

- DLL uses either K or C clock as its synchronizing input. The input must have low phase jitter, which is specified as  $t_{KC \text{ Var}}$ .
- The DLL functions at frequencies down to 80MHz.
- If the input clock is unstable and the DLL is enabled, then the DLL may lock to an incorrect frequency, causing unstable SRAM behavior.

### Power Up Waveforms



#### Notes:

13. It is recommended that the  $\overline{\text{DOFF}}$  pin be pulled HIGH via a pull up resistor of 1 Kohm.
14. During Power Up, when the  $\overline{\text{DOFF}}$  is tied HIGH, the DLL gets locked after 1024 cycles of stable clock.

## Maximum Ratings

Exceeding maximum ratings may impair the useful life of the device. These user guidelines are not tested.

Storage Temperature ..... -65°C to +150°C

Ambient Temperature with

Power Applied ..... -55°C to +125°C

Supply Voltage on  $V_{DD}$  Relative to GND ..... -0.5V to +2.9V

Supply Voltage on  $V_{DDQ}$  Relative to GND ..... -0.5V to + $V_{DD}$

DC Voltage Applied to Outputs

in High-Z State ..... -0.5V to  $V_{DDQ} + 0.3V$

DC Input Voltage<sup>[18]</sup> ..... -0.5V to  $V_{DD} + 0.3V$

Current into Outputs (LOW) ..... 20 mA

Static Discharge Voltage ..... > 2001V  
(MIL-STD-883, Method 3015)

Latch up Current ..... > 200 mA

## Operating Range

Range	Ambient Temperature ( $T_A$ )	$V_{DD}$ <sup>[19]</sup>	$V_{DDQ}$ <sup>[19]</sup>
Com'l	0°C to +70°C	1.8 ± 0.1V	1.4V to $V_{DD}$
Ind'l	-40°C to +85°C		

## Electrical Characteristics Over the Operating Range<sup>[15, 19]</sup>

### DC Electrical Characteristics Over the Operating Range

Parameter	Description	Test Conditions	Min	Typ	Max	Unit
$V_{DD}$	Power Supply Voltage		1.7	1.8	1.9	V
$V_{DDQ}$	IO Supply Voltage		1.4	1.5	$V_{DD}$	V
$V_{OH}$	Output HIGH Voltage	Note 16	$V_{DDQ}/2 - 0.12$		$V_{DDQ}/2 + 0.12$	V
$V_{OL}$	Output LOW Voltage	Note 17	$V_{DDQ}/2 - 0.12$		$V_{DDQ}/2 + 0.12$	V
$V_{OH(LOW)}$	Output HIGH Voltage	$I_{OH} = -0.1$ mA, Nominal Impedance	$V_{DDQ} - 0.2$		$V_{DDQ}$	V
$V_{OL(LOW)}$	Output LOW Voltage	$I_{OL} = 0.1$ mA, Nominal Impedance	$V_{SS}$		0.2	V
$V_{IH}$	Input HIGH Voltage <sup>[18]</sup>		$V_{REF} + 0.1$		$V_{DDQ} + 0.3$	V
$V_{IL}$	Input LOW Voltage <sup>[18]</sup>		-0.3		$V_{REF} - 0.1$	V
$I_X$	Input Leakage Current	$GND \leq V_I \leq V_{DDQ}$	-5		5	μA
$I_{OZ}$	Output Leakage Current	$GND \leq V_I \leq V_{DDQ}$ , Output Disabled	-5		5	μA
$V_{REF}$	Input Reference Voltage <sup>[20]</sup>	Typical Value = 0.75V	0.68	0.75	0.95	V
$I_{DD}$	$V_{DD}$ Operating Supply	$V_{DD} = \text{Max.}$ , $I_{OUT} = 0$ mA, $f = f_{MAX} = 1/t_{CYC}$	167 MHz		740	mA
			200 MHz		870	mA
			250 MHz		1065	mA
$I_{SB1}$	Automatic Power down Current	Max. $V_{DD}$ , Both Ports Deselected, $V_{IN} \geq V_{IH}$ or $V_{IN} \leq V_{IL}$ , $f = f_{MAX} = 1/t_{CYC}$ , Inputs Static	167 MHz		270	mA
			200 MHz		300	mA
			250 MHz		350	mA

### AC Input Requirements Over the Operating Range

Parameter	Description	Test Conditions	Min	Typ	Max	Unit
$V_{IH}$	Input High (Logic 1) Voltage		$V_{REF} + 0.2$	-	-	V
$V_{IL}$	Input Low (Logic 0) Voltage		-	-	$V_{REF} - 0.2$	V

## Capacitance<sup>[21]</sup>

Parameter	Description	Test Conditions	Max	Unit
$C_{IN}$	Input Capacitance	$T_A = 25^\circ\text{C}$ , $f = 1$ MHz, $V_{DD} = 1.8V$ $V_{DDQ} = 1.5V$	5	pF
$C_{CLK}$	Clock Input Capacitance		4	pF
$C_O$	Output Capacitance		5	pF

### Notes:

15. All voltage referenced to Ground.

16. Output are impedance controlled.  $I_{OH} = -(V_{DDQ}/2)/(RQ/5)$  for values of  $175\Omega \leq RQ \leq 350\Omega$ .

17. Output are impedance controlled.  $I_{OL} = (V_{DDQ}/2)/(RQ/5)$  for values of  $175\Omega \leq RQ \leq 350\Omega$ .

18. Overshoot:  $V_{IH(AC)} < V_{DDQ} + 0.85V$  (Pulse width less than  $t_{CYC}/2$ ), Undershoot:  $V_{IL(AC)} > -1.5V$  (Pulse width less than  $t_{CYC}/2$ ).

19. Power up: Assumes a linear ramp from 0V to  $V_{DD}(\text{min.})$  within 200 ms. During this time  $V_{IH} < V_{DD}$  and  $V_{DDQ} \leq V_{DD}$ .

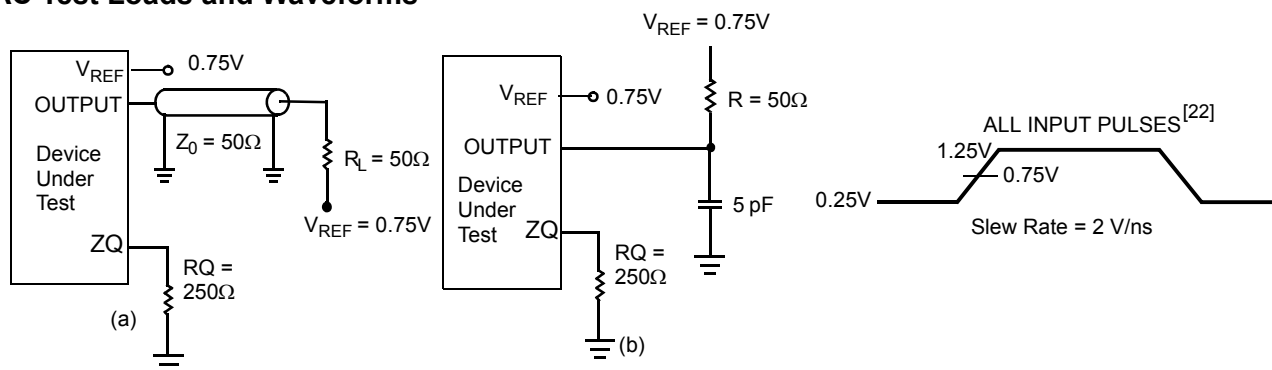
20.  $V_{REF}(\text{Min.}) = 0.68V$  or  $0.46V_{DDQ}$ , whichever is larger,  $V_{REF}(\text{Max.}) = 0.95V$  or  $0.54V_{DDQ}$ , whichever is smaller.

21. Tested initially and after any design or process change that may affect these parameters.

## Thermal Resistance<sup>[21]</sup>

Parameter	Description	Test Conditions	165 FBGA Package	Unit
$\Theta_{JA}$	Thermal Resistance (Junction to Ambient)	Test conditions follow standard test methods and procedures for measuring thermal impedance, according to EIA/JESD51.	17.2	°C/W
$\Theta_{JC}$	Thermal Resistance (Junction to Case)		3.2	°C/W

## AC Test Loads and Waveforms



### Note:

22. Unless otherwise noted, test conditions assume signal transition time of 2V/ns, timing reference levels of 0.75V, Vref = 0.75V,  $R_Q = 250\Omega$ ,  $V_{DDQ} = 1.5V$ , input pulse levels of 0.25V to 1.25V, and output loading of the specified  $I_{OL}/I_{OH}$  and load capacitance shown in (a) of AC Test Loads.



## Switching Characteristics Over the Operating Range [22, 23]

Cypress Parameter	Consortium Parameter	Description	250 MHz		200 MHz		167 MHz		Unit
			Min	Max	Min	Max	Min	Max	
$t_{POWER}$		$V_{DD}$ (Typical) to the first Access <sup>[26]</sup>	1		1		1		ms
$t_{CYC}$	$t_{KHKH}$	K Clock and C Clock Cycle Time	4.0	6.3	5.0	7.9	6.0	8.4	ns
$t_{KH}$	$t_{KHKL}$	Input Clock (K/ $\bar{K}$ and C/ $\bar{C}$ ) HIGH	1.6	–	2.0	–	2.4	–	ns
$t_{KL}$	$t_{KLKH}$	Input Clock (K/ $\bar{K}$ and C/ $\bar{C}$ ) LOW	1.6	–	2.0	–	2.4	–	ns
$t_{KH\bar{K}H}$	$t_{KH\bar{K}H}$	K Clock Rise to $\bar{K}$ Clock Rise and C to $\bar{C}$ Rise (rising edge to rising edge)	1.8	–	2.2	–	2.7	–	ns
$t_{KHCH}$	$t_{KHCH}$	K/ $\bar{K}$ Clock Rise to C/ $\bar{C}$ Clock Rise (rising edge to rising edge)	0.0	1.8	0.0	2.2	0.0	2.7	ns
<b>Setup Times</b>									
$t_{SA}$	$t_{AVKH}$	Address Setup to K Clock Rise	0.35	–	0.4	–	0.5	–	ns
$t_{SC}$	$t_{IVKH}$	Control Setup to K Clock Rise ( $\overline{RPS}$ , $\overline{WPS}$ )	0.35	–	0.4	–	0.5	–	ns
$t_{SCDDR}$	$t_{IVKH}$	Double Data Rate Control Setup to Clock (K, $\bar{K}$ ) Rise ( $BWS_0$ , $BWS_1$ , $BWS_3$ , $BWS_4$ )	0.35	–	0.4	–	0.5	–	ns
$t_{SD}$	$t_{DVKH}$	$D_{[X:0]}$ Setup to Clock (K/ $\bar{K}$ ) Rise	0.35	–	0.4	–	0.5	–	ns
<b>Hold Times</b>									
$t_{HA}$	$t_{KHAX}$	Address Hold after K Clock Rise	0.35	–	0.4	–	0.5	–	ns
$t_{HC}$	$t_{KHIX}$	Control Hold after K Clock Rise ( $\overline{RPS}$ , $\overline{WPS}$ )	0.35	–	0.4	–	0.5	–	ns
$t_{HCDDR}$	$t_{KHIX}$	Double Data Rate Control Hold after Clock (K, $\bar{K}$ ) Rise ( $BWS_0$ , $BWS_1$ , $BWS_3$ , $BWS_4$ )	0.35	–	0.4	–	0.5	–	ns
$t_{HD}$	$t_{KHDX}$	$D_{[X:0]}$ Hold after Clock (K/ $\bar{K}$ ) Rise	0.35	–	0.4	–	0.5	–	ns
<b>Output Times</b>									
$t_{CO}$	$t_{CHQV}$	C/ $\bar{C}$ Clock Rise (or K/ $\bar{K}$ in Single Clock Mode) to Data Valid	–	0.45	–	0.45	–	0.50	ns
$t_{DOH}$	$t_{CHQX}$	Data Output Hold after Output C/ $\bar{C}$ Clock Rise (Active to Active)	–0.45	–	–0.45	–	–0.50	–	ns
$t_{CCQO}$	$t_{CHCQV}$	C/ $\bar{C}$ Clock Rise to Echo Clock Valid	–	0.45	–	0.45	–	0.50	ns
$t_{CQOH}$	$t_{CHCQX}$	Echo Clock Hold after C/ $\bar{C}$ Clock Rise	–0.45	–	–0.45	–	–0.50	–	ns
$t_{CQD}$	$t_{CQHQV}$	Echo Clock High to Data Valid	–	0.30	–	0.35	–	0.40	ns
$t_{CQDOH}$	$t_{CQHQX}$	Echo Clock High to Data Invalid	–0.30	–	–0.35	–	–0.40	–	ns
$t_{CHZ}$	$t_{CHQZ}$	Clock (C/ $\bar{C}$ ) Rise to High-Z (Active to High-Z) <sup>[24,25]</sup>	–	0.45	–	0.45	–	0.50	ns
$t_{CLZ}$	$t_{CHQX1}$	Clock (C/ $\bar{C}$ ) Rise to Low-Z <sup>[24,25]</sup>	–0.45	–	–0.45	–	–0.50	–	ns
<b>DLL Timing</b>									
$t_{KC Var}$	$t_{KC Var}$	Clock Phase Jitter	–	0.20	–	0.20	–	0.20	ns
$t_{KC lock}$	$t_{KC lock}$	DLL Lock Time (K, C)	1024	–	1024	–	1024	–	cycles
$t_{KC Reset}$	$t_{KC Reset}$	K Static to DLL Reset	30	–	30	–	30	–	ns

### Notes:

23. All devices can operate at clock frequencies as low as 119 MHz. When a part with a maximum frequency above 133 MHz is operating at a lower clock frequency, it requires the input timings of the frequency range in which it is being operated and outputs data with the output timings of that frequency range.

24.  $t_{CHZ}$ ,  $t_{CLZ}$ , are specified with a load capacitance of 5 pF as in part (b) of AC Test Loads. Transition is measured  $\pm 100$  mV from steady-state voltage.

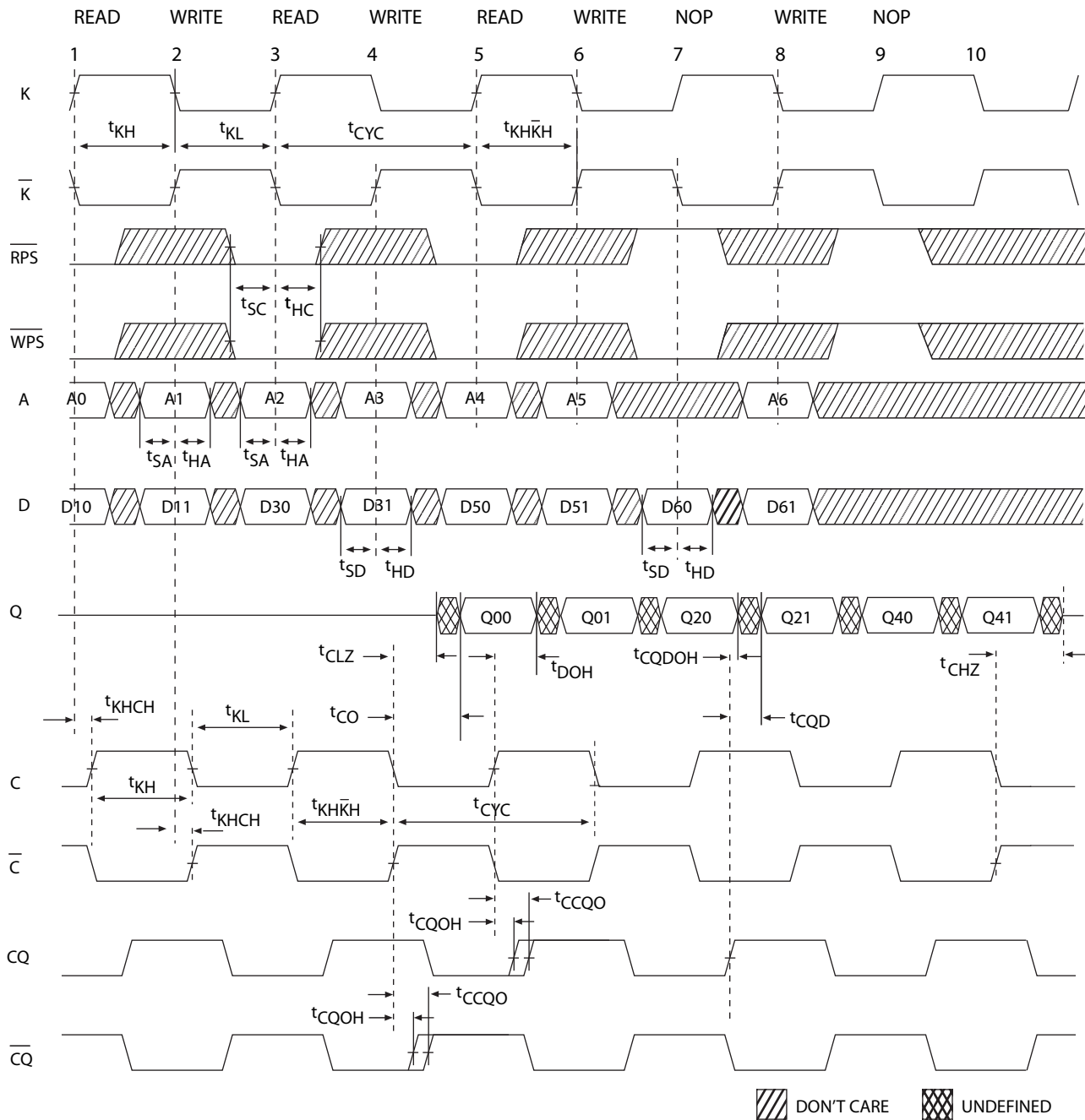
25. At any voltage and temperature  $t_{CHZ}$  is less than  $t_{CLZ}$  and  $t_{CHZ}$  less than  $t_{CO}$ .

26. This part has a voltage regulator internally;  $t_{POWER}$  is the time that the power needs to be supplied above  $V_{DD}$  minimum initially before a read or write operation can be initiated.

27. For D2 data signal on CY7C1425AV18 device,  $t_{SD}$  is 0.5 ns for 200 MHz, and 250 MHz frequencies.

## Switching Waveforms<sup>[28, 29, 30]</sup>

### Read/Write/Deselect Sequence



#### Notes:

28. Q00 refers to output from address A0. Q01 refers to output from the next internal burst address following A0, i.e., A0 + 1.

29. Output are disabled (High-Z) one clock cycle after a NOP.

30. In this example, if address A0 = A1, then data Q00 = D10 and Q01 = D11. Write data is forwarded immediately as read results. This note applies to the whole diagram.

## Ordering Information

Not all of the speed, package and temperature ranges are available. Please contact your local sales representative or visit [www.cypress.com](http://www.cypress.com) for actual products offered.

Speed (MHz)	Ordering Code	Package Diagram	Package Type	Operating Range
167	CY7C1410AV18-167BZC	51-85195	165-Ball Fine Pitch Ball Grid Array (15 x 17 x 1.4 mm)	Commercial
	CY7C1425AV18-167BZC			
	CY7C1412AV18-167BZC			
	CY7C1414AV18-167BZC			
	CY7C1410AV18-167BZXC	51-85195	165-Ball Fine Pitch Ball Grid Array (15 x 17 x 1.4 mm) Pb-Free	
	CY7C1425AV18-167BZXC			
	CY7C1412AV18-167BZXC			
	CY7C1414AV18-167BZXC			
	CY7C1410AV18-167BZI	51-85195	165-Ball Fine Pitch Ball Grid Array (15 x 17 x 1.4 mm)	Industrial
	CY7C1425AV18-167BZI			
	CY7C1412AV18-167BZI			
	CY7C1414AV18-167BZI			
	CY7C1410AV18-167BZXI	51-85195	165-Ball Fine Pitch Ball Grid Array (15 x 17 x 1.4 mm) Pb-Free	
	CY7C1425AV18-167BZXI			
	CY7C1412AV18-167BZXI			
	CY7C1414AV18-167BZXI			
200	CY7C1410AV18-200BZC	51-85195	165-Ball Fine Pitch Ball Grid Array (15 x 17 x 1.4 mm)	Commercial
	CY7C1425AV18-200BZC			
	CY7C1412AV18-200BZC			
	CY7C1414AV18-200BZC			
	CY7C1410AV18-200BZXC	51-85195	165-Ball Fine Pitch Ball Grid Array (15 x 17 x 1.4 mm) Pb-Free	
	CY7C1425AV18-200BZXC			
	CY7C1412AV18-200BZXC			
	CY7C1414AV18-200BZXC			
	CY7C1410AV18-200BZI	51-85195	165-Ball Fine Pitch Ball Grid Array (15 x 17 x 1.4 mm)	Industrial
	CY7C1425AV18-200BZI			
	CY7C1412AV18-200BZI			
	CY7C1414AV18-200BZI			
	CY7C1410AV18-200BZXI	51-85195	165-Ball Fine Pitch Ball Grid Array (15 x 17 x 1.4 mm) Pb-Free	
	CY7C1425AV18-200BZXI			
	CY7C1412AV18-200BZXI			
	CY7C1414AV18-200BZXI			
250	CY7C1410AV18-250BZC	51-85195	165-Ball Fine Pitch Ball Grid Array (15 x 17 x 1.4 mm)	Commercial
	CY7C1425AV18-250BZC			
	CY7C1412AV18-250BZC			
	CY7C1414AV18-250BZC			
	CY7C1410AV18-250BZXC	51-85195	165-Ball Fine Pitch Ball Grid Array (15 x 17 x 1.4 mm) Pb-Free	
	CY7C1425AV18-250BZXC			
	CY7C1412AV18-250BZXC			
	CY7C1414AV18-250BZXC			

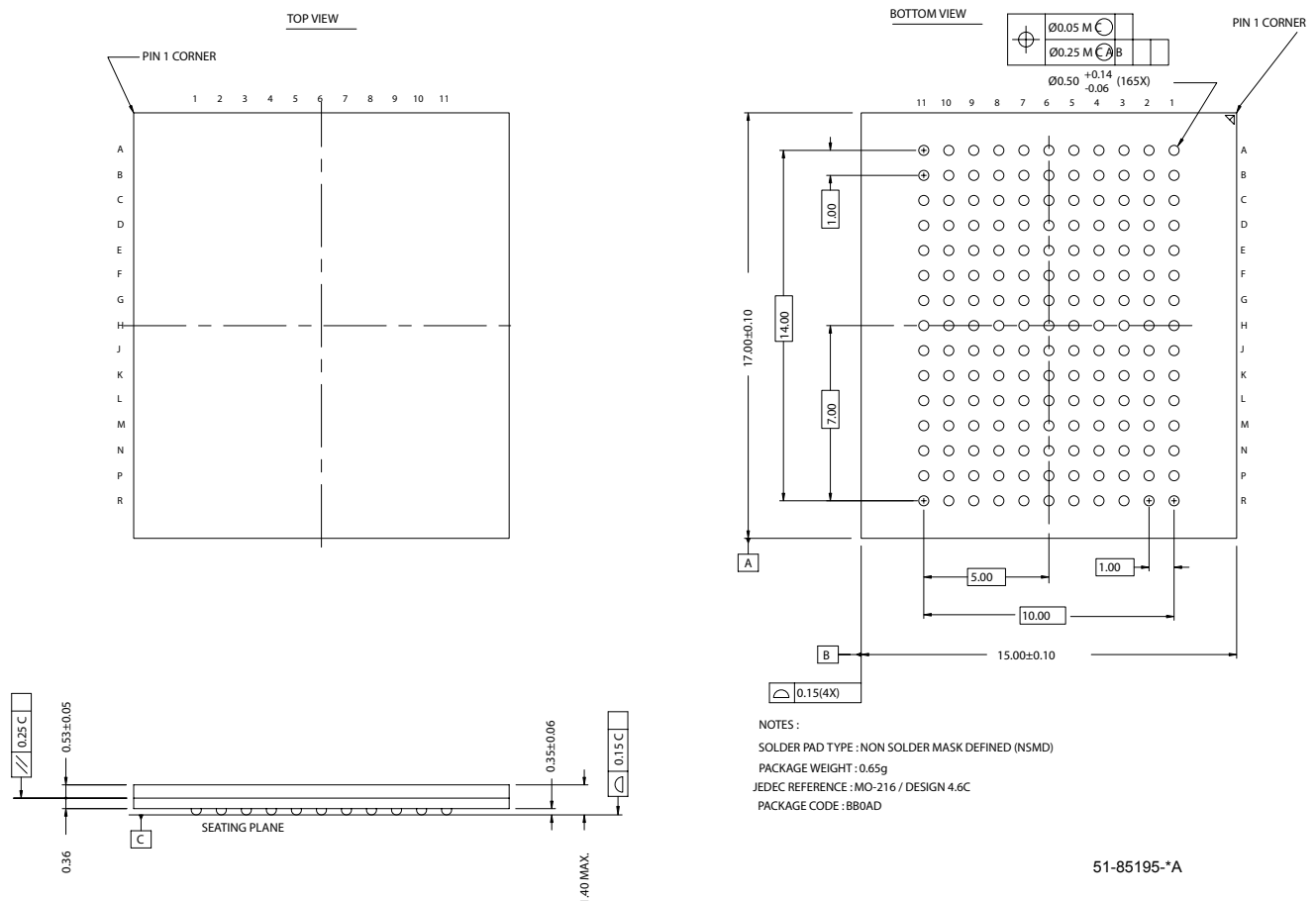
## Ordering Information (continued)

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Speed (MHz)	Ordering Code	Package Diagram	Package Type	Operating Range
250	CY7C1410AV18-250BZI	51-85195	165-Ball Fine Pitch Ball Grid Array (15 x 17 x 1.4 mm)	Industrial
	CY7C1425AV18-250BZI			
	CY7C1412AV18-250BZI			
	CY7C1414AV18-250BZI			
	CY7C1410AV18-250BZXI	51-85195	165-Ball Fine Pitch Ball Grid Array (15 x 17 x 1.4 mm) Pb-Free	
	CY7C1425AV18-250BZXI			
	CY7C1412AV18-250BZXI			
	CY7C1414AV18-250BZXI			

## Package Diagram

**165-Ball FBGA (15 x 17 x 1.40 mm) (51-85195)**



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## Document History Page

Document Title: CY7C1410AV18/CY7C1425AV18/CY7C1412AV18/CY7C1414AV18 36-Mbit QDR-II™ SRAM 2-Word Burst Architecture Document Number: 38-05615				
REV.	ECN No.	Issue Date	Orig. of Change	Description of Change
**	247331	See ECN	SYT	New Data Sheet
*A	326519	See ECN	SYT	Removed CY7C1425AV18 from the title Included 300 MHz Speed grade Replaced TBDs with their respective values for $I_{DD}$ and $I_{SB1}$ Added Industrial temperature grade Replaced the TBDs on the Thermal Characteristics Table to $\Theta_{JA} = 17.2^{\circ}\text{C/W}$ and $\Theta_{JC} = 3.2^{\circ}\text{C/W}$ Replaced TBDs in the Capacitance Table to their respective values for the 165 FBGA Package Changed typo of bit # 47 to bit # 108 under the EXTEST OUTPUT BUS TRI-STATE on Page 16 Added Pb-free Product Information Updated the Ordering Information by Shading and Unshading MPNs according to availability
*B	413953	See ECN	NXR	Converted from preliminary to final. Added CY7C1425AV18 part number to title. Removed 300-MHz speed Bin. Changed address of Cypress Semiconductor Corporation on Page# 1 from "3901 North First Street" to "198 Champion Court" Changed C, C Description in Feature Section and Pin Description. Added Power up sequence and Wave form on page# 19 Added foot notes # 13, 14, 15 on page# 19 Replaced Three-state with Tri-state. Changed the description of $I_X$ from Input Load Current to Input Leakage Current on page# 20 Modified the $I_{DD}$ and $I_{SB}$ values. Modified test condition in Footnote # 20 on page# 20 from $V_{DDQ} < V_{DD}$ to $V_{DDQ} \leq V_{DD}$ . Replaced Package Name column with Package Diagram in the Ordering Information table. Updated Ordering Information Table.
*C	468029	See ECN	NXR	Modified the ZQ Definition from Alternately, this pin can be connected directly to $V_{DD}$ to Alternately, this pin can be connected directly to $V_{DDQ}$ . Included Maximum Ratings for Supply Voltage on $V_{DDQ}$ Relative to GND Changed the Maximum Ratings for DC Input Voltage from $V_{DDQ}$ to $V_{DD}$ . Changed $t_{TH}$ and $t_{TL}$ from 40 ns to 20 ns, changed $t_{TMSS}$ , $t_{TDIS}$ , $t_{CS}$ , $t_{TMSH}$ , $t_{TDIH}$ , $t_{CH}$ from 10 ns to 5 ns and changed $t_{TDOV}$ from 20 ns to 10 ns in TAP AC Switching Characteristics table Modified Power Up waveform Changed the Maximum rating of Ambient Temperature with Power Applied from $-10^{\circ}\text{C}$ to $+85^{\circ}\text{C}$ to $-55^{\circ}\text{C}$ to $+125^{\circ}\text{C}$ Added additional notes in the AC parameter section Changed the $t_{SC}$ and $t_{HC}$ value for 250 MHz from 0.5 ns to 0.35 ns, for 200 MHz from 0.6 ns to 0.4 ns, and for 167 MHz from 0.7 ns to 0.5 ns. Modified AC Switching Waveform. Corrected the typo In the AC Switching Characteristics Table. Updated the Ordering Information Table.
*D	1274725	See ECN	VKN/AESA	Modified footnote# 30