

The revision list can be viewed directly by clicking the title page.

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H8S/2148 Group, H8S/2144 Group, H8S/2148F-ZTAT[™], H8S/2147N F-ZTAT[™], H8S/2144F-ZTAT[™], H8S/2142F-ZTAT[™] Hardware Manual

Renesa

Renesas 16-Bit Single-Chip Microcomputer H8S Family/H8S/2100 Series

H8S/2148 HD6432148S H8S/2147N HD64F2147N HD6432148SW HD64F2147NV HD64F2148 H8S/2144 HD6432144S HD64F2148V HD64F2144 HD64F2148A HD64F2144V HD64F2148AV HD64F2144A H8S/2147 HD6432147S HD64F2144AV HD6432147SW H8S/2143 HD6432143S HD64F2147A H8S/2142 HD6432142

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HD64F2147AV

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HD64F2142R HD64F2142RV

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General Precautions on Handling of Product

1. Treatment of NC Pins

Note: Do not connect anything to the NC pins.

The NC (not connected) pins are either not connected to any of the internal circuitry or are used as test pins or to reduce noise. If something is connected to the NC pins, the operation of the LSI is not guaranteed.

2. Treatment of Unused Input Pins

Note: Fix all unused input pins to high or low level.

Generally, the input pins of CMOS products are high-impedance input pins. If unused pins are in their open states, intermediate levels are induced by noise in the vicinity, a pass-through current flows internally, and a malfunction may occur.

3. Processing before Initialization

Note: When power is first supplied, the product's state is undefined.

The states of internal circuits are undefined until full power is supplied throughout the chip and a low level is input on the reset pin. During the period where the states are undefined, the register settings and the output state of each pin are also undefined. Design your system so that it does not malfunction because of processing while it is in this undefined state. For those products which have a reset function, reset the LSI immediately after the power supply has been turned on.

4. Prohibition of Access to Undefined or Reserved Addresses

Note: Access to undefined or reserved addresses is prohibited.

The undefined or reserved addresses may be used to expand functions, or test registers may have been be allocated to these addresses. Do not access these registers; the system's operation is not guaranteed if they are accessed.

Preface

The H8S/2148 Group, H8S/2144 Group, and H8S/2147N comprise high-performance microcomputers with a 32-bit H8S/2000 CPU core, and a set of on-chip supporting functions required for system configuration.

The H8S/2000 CPU can execute basic instructions in one state, and is provided with sixteen internal 16-bit general registers with a 32-bit configuration, and a concise and optimized instruction set. The CPU can handle a 16-Mbyte linear address space (architecturally 4 Gbytes). Programs based on the high-level language C can also be run efficiently.

Single-power-supply flash memory (F-ZTAT^{TM*}) and mask ROM versions are available, providing a quick and flexible response to conditions from ramp-up through full-scale volume production, even for applications with frequently changing specifications.

On-chip peripheral functions include a 16-bit free-running timer (FRT), 8-bit timer (TMR), watchdog timer (WDT), two PWM timers (PWM and PWMX), a serial communication interface (SCI, IrDA), PS/2-compatible keyboard buffer controller, host interface (HIF), D/A converter (DAC), A/D converter (ADC), and I/O ports. An I²C bus interface (IIC) can also be incorporated as an option.

An on-chip data transfer controller (DTC) is also provided, enabling high-speed data transfer without CPU intervention.

The H8S/2148 Group has all the above on-chip supporting functions, and can also be provided with an IIC module as an option. The H8S/2144 Group comprises reduced-function versions, with fewer TMR channels, and no PWM, keyboard buffer controller, HIF, IIC, or DTC modules, and the H8S/2147N with fewer TMR channels, no DTC and some other functions.

Use of the H8S/2148 Group, H8S/2144 Group, H8S/2147N enables compact, high-performance systems to be implemented easily. The comprehensive PC-related interface functions and 16×8 matrix key-scan functions are ideal for applications such as notebook PC keyboard control and intelligent battery and power supply control, while the various timer functions and their interconnectability (timer connection), plus the interlinked operation of the I^2C bus interface and data transfer controller (DTC), in particular, make these devices ideal for use in PC monitors. In addition, the combination of F-ZTAT^{TM*} and reduced-function versions is ideal for applications such as CD-ROM drive units in which on-chip program memory is essential to meet performance requirements, product start-up times are short, and program modifications may be necessary after end-product assembly.

This manual describes the hardware of the H8S/2148 Group, H8S/2144 Group, and H8S/2147N. Refer to the H8S/2600 Series and H8S/2000 Series Software Manual for a detailed description of the instruction set.

Note: * F-ZTAT (Flexible-ZTAT) is a trademark of Renesas Technology Corp.

On-Chip Supporting Modules

Group		H8S/2148 Group	H8S/2147N	H8S/2144 Group			
Product names		H8S/2148, H8S/2147	H8S/2147N	H8S/2144, H8S/2143, H8S/2142			
Bus controller (BSC)	Available (16 bits)	Available (16 bits)	Available (16 bits)			
Data transfer co	ontroller (DTC)	Available	_	_			
8-bit PWM time	r (PWM)	×16	×16	_			
14-bit PWM tim	er (PWMX)	×2	×2	×2			
16-bit free-runn	ing timer (FRT)	×1	×1	×1			
8-bit timer (TMI	₹)	×4	×3	×3			
Timer connection	on	Available	_	_			
Watchdog time	r (WDT)	×2	×2	×2			
Serial commun	ication interface (SCI)	×3	×3	×3			
I ² C bus interfac	e (IIC)	×2 (option)	×2 (option)	_			
Keyboard buffe (PS/2 compatib		×3	×3	_			
Host interface (HIF)	×4	×4	_			
D/A converter		×2	×2	×2			
A/D converter	Analog inputs	×8	×8	×8			
	Expansion A/D inputs	×16	×16	×16			

Main Revisions for This Edition

Item	Page	Revision	(See Manual 1	for Details)		
All	_	Notifica	ation of change	in company na	ame amen	ded
		(Before	e) Hitachi, Ltd.	→ (After) Rene	sas Techr	nology Corp.
		• Produc	t naming conve	ention amende	d	
		(Before	e) H8S/2148 Se	eries → (After)	H8S/2148	Group
		(Before	e) H8S/2144 Se	eries \rightarrow (After)	H8S/2144	Group
1.1 Overview	4	Host inter	rface specificati	ion in table 1.1	amended	
Table 1.1 Overview		• 8-bit ho	st interface (IS	A) port		
	6	Product li	neup specificat	tion in table 1.1	amended	
			Product Code	_		
		Group	Mask ROM Versions	F-ZTAT Versions	ROM/RAM (Bytes)	Packages
		H8S/2148	HD6432148S	HD64F2148 HD64F2148V*2	128 k/4 k	FP-100B, TFP-100B
			HD6432148SW*1	HD64F2148A HD64F2148AV*2		
			HD6432147S	HD64F2147A	64 k/2 k	
			HD6432147SW*1	HD64F2147AV*2		
		H8S/2147N	_	HD64F2147N HD64F2147NV*2	64 k/2 k	_
		H8S/2144	HD6432144S	HD64F2144 HD64F2144V*2	128 k/4 k	
				HD64F2144A HD64F2144AV*2		
			HD6432143S	_	96 k/4 k	_
			HD6432142	HD64F2142R HD64F2142RV*2	64 k/2 k	
		Notes: 1.	W indicates the I ² C	bus option.		
			V indicates the 3-V Code Lineup.	version. Please re	efer to appen	dix F, Product
1.2 Internal Block	7	Figure 1.	1 (a) amended			
Diagram		(Before)	$STBY \rightarrow (After)$	STBY		
Figure 1.1 (a) Internal Block Diagram of H8S/2148 Group						
Figure 1.1 (b) Internal	8	Figure 1.	1 (b) amended			
Block Diagram of H8S/2147N		(Before) I	$IC \times 2ch \rightarrow (Af$	ter) IIC × 2ch (option)	

Item	Page	Revision	on (See	Manual	for D	etails)							
1.3.2 Pin Functions in	14	Mode 1	descrip	tion of p	in 35 a	amended							
Each Operating Mode						7/ĪRQ7 →							
Table 1.2 (a) Pin	·	(After)	P67/TM	OX/CIN7	/KIN7	/IRQ7							
Functions in Each Operating Mode	17	Modes	2 and 3	in single	chip	modes of pin 95 amended							
		(Before) P82 –	After) I	P82/H	IFSD							
Table 1.2 (b)	19	Mode 1	descrip	tion of p	in 35 a	amended							
H8S/2147N Pin Functions in Each		(Before) P67/C	IN/KIN7/	IRQ7	→ (After) P67/CIN7/KIN7/IRQ7							
Operating Mode	21	Modes 2 and 3 in single chip modes of pin 95 amended											
		(Before) P82 –	(After)	P82/H	IFSD							
1.3.3 Pin Functions	30	Table 1.3 amended											
Table 1.3 Pin Functions				Pin No.	_								
		Туре	Symbol	FP-100B TFP-100B	1/0	Name and Function							
		Host interface	HIRQ11 HIRQ1	52 53	Output	Host interrupt 11, 1, 12, 3, and 4 : Output pins for interrupt requests to the host.							
		(HIF)	HIRQ12 HIRQ3 HIRQ4	54 91 90									
			TIINQ4										
	33			Pin No. FP-100B	_								
		Type I/O ports	Symbol PA7 to	TFP-100B	I/O Input/	Name and Function Port A: Eight input/output pins. The data direction							
		i/O ports	PA0	21, 30, 31, 47, 48	output	of each pin can be selected in the port A data direction register (PADDR). These pins have built-in MOS input pull-ups. These are the VCCB drive pins. [H8S/2148 Group and H8S/2147N only]							
2.6.1 Overview	52	Instruct	ion in a	rithmetic	opera	tions amended							
Table 2.1 Instruction		(Before) EG →	(After) N	IEG								
Classification		`	•										
3.2.4 Serial Timer	89	Bit 3 bit	table a	mended									
Control Register (STCR)		Bit 3											
		FLSHE 0	Description Addresses		o H'(FF)F	F87 are used for power-down mode control register							
		1	- ''	rting module of		gister access (Initial value) F87 are used for flash memory control register							
		·		ZTAT version		To all deed for high memory control register							
4.5 Stack Status after Exception Handling	111	Note *	deleted	from figu	ire 4.5	(2)							
Figure 4.5 (2) Stack													
Status after Exception													
Handling (Advanced Mode)													
111040/													

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5.2.8 Address Break	124	Read	d/Write de	escription an	nended									
Control Register (ABRKCR)		Bit 7	(Before)	$R/W \rightarrow (After)$	er) R									
5.5.3 Interrupt control	140	Figu	re 5.9 am	ended										
Mode 1		(Bef	(Before) Only NMI interrupts enabled and address break \rightarrow											
Figure 5.9 Example of State Transitions in Interrupt control Mode 1		(Afte	(After) Only NMI interrupts and address break enabled											
8.1 Overview	213	Tabl	e 8.1 ame	ended										
Table 8.1 H8S/2148					Expa	nded Modes	Single-Chip Mode							
Group Port Functions		Port	Description	Pins	Mode 1	Mode 2, Mode 3 (EXPE = 1)	Mode 2, Mode 3 (EXPE = 0)							
		Port A	8-bit I/O port	PA7/A23/KIN15/	I/O port also	I/O port also	I/O port also functioning as							

CIN15/PS2CD functioning as functioning as key-sense interrupt input key-sense (KIN15 to KIN8), address output (A23 PA6/A22/KIN14/ expansion A/D converter interrupt input to A16), key-sense CIN14/PS2CC (KIN15 to input (CIN15 to CIN8), and interrupt input PA5/A21/KIN13/ KIN8), (KIN15 to KIN8), keyboard buffer controller CIN13/PS2BD expansion A/D expansion A/D input/output (PS2CD, PS2CC, PS2BD, PS2BC, PS2AD, PS2AC) PA4/A20/KIN12/ converter input converter input (CIN15 to (CIN15 to CIN8), CIN12/PS2BC CIN8), and and keyboard buffer PA3/A19/KIN11/ keyboard controller input/ CIN11/PS2AD output (PS2CD, buffer PA2/A18/KIN10/ PS2CC, PS2BD, PS2BC, PS2AD, controller CIN10/PS2AC input/output (PS2CD, PS2AC) PA1/A17/KIN9/ PS2CC, CIN9 PS2BD. PA0/A16/KIN8/ PS2BC. CIN8 PS2AD, PS2AC)

Table 8.2 H8S/2147N Port Functions

Table 8.2 amended

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			Expar	ided Modes	Single-Chip Mode
Port	Description	escription Pins		Mode 2, Mode 3 (EXPE = 1)	Mode 2, Mode 3 (EXPE = 0)
Port A	8-bit I/O port	PA7/A23/KIN15/ CIN15/PS2CD PA6/A22/KIN14/ CIN14/PS2CC PA5/A21/KIN13/ CIN13/PS2BD PA4/A20/KIN12/ CIN12/PS2BC PA3/A19/KIN11/ CIN11/PS2AD PA2/A18/KIN10/ CIN10/PS2AC PA1/A17/KIN9/ CIN9 PA0/A16/KIN8/ CIN8		I/O port also functioning as address output (A23 to A16), key-sense interrupt input (KINT5 to KINB), expansion A/D converter input (CINT5 to CINB), and keyboard buffer controller input output (PS2CD, PS2CC, PS2BD, PS2AC)	I/O port also functioning as key-sense interrupt input (KINT5 to KINB), expansion A/D converter input (CIN15 to CIN8), and keyboard buffer controller input/output (PS2CD, PS2CC, PS2BD, PS2BC, PS2AD, PS2AC)

Item	Page	Revi	ision (Se	e Manu	al for	Details	s)				
8.1 Overview	220	Tabl	e 8.3 am	ended							
Table 8.3 H8S/2144 Po	rt					Expa	nded Mo	des	Sir	ngle-Chip	Mode
Functions		Port	Description	Pins		Mode 1		2, Mode 3 XPE = 1)	М	ode 2, Mo (EXPE =	
		Port A	8-bit I/O port	PA7 to PA0/ A23 to A16/ KIN15 to KIN CIN15 to CIN	fu 8/ ke 8/ in (K K) ex	O port also nctioning as ey-sense terrupt input (IN15 to IN8), and cpansion A/E onverter input (IN15 to IN8).	to A16) interrup (KIN15) and exp t convert	ning as s output (A23 , key-sense	key-sei (KIN15 expans	t also func nse interru to KIN8) a ion A/D co CIN15 to C	pt input and onverter
8.7.2 Register Configuration	247		e 8.14 ar ore) Syst		ol rea	ister →	(Afte	er) Svst	em c	ontrol	
Table 8.14 Port 6 Registers		,	ster 2		o o g	,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,	(*	, •,•.			
8.9.3 Pin Functions	258	P81/	GA20/C	S2 select	ion m	ethod a	and pi	in funct	ion d	escrip	tion
Table 8.19 Port 8 Pin		ame	nded				•			-	
Functions		Pin	,	Selection Me	thod and	d Pin Func	tions				
. anonone		P81/G/		The pin functi operating mo							
				Operating mode	Not slav	ve mode		Sla	ave mode		
			L	FGA20E	-	_		0			1
				CS2E	-	-	()	1	-	_
				P81DDR	0	1	0	1	_	0	1
				Pin function	P81	P81	P81	P81	CS2	P81	GA20

input output input output input

pin

This pin should be used as the GA20 output pin or $\overline{\text{CS2}}$ input pin only in mode 2 or 3 (EXPE = 0).

pin

pin

pin

pin

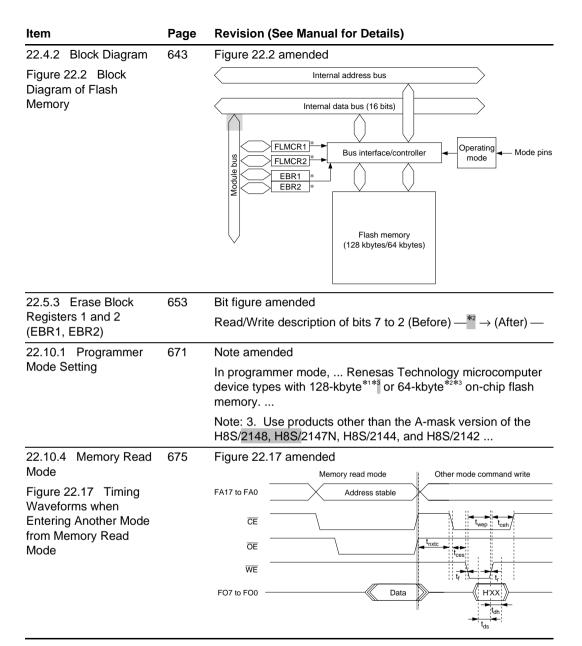
input output

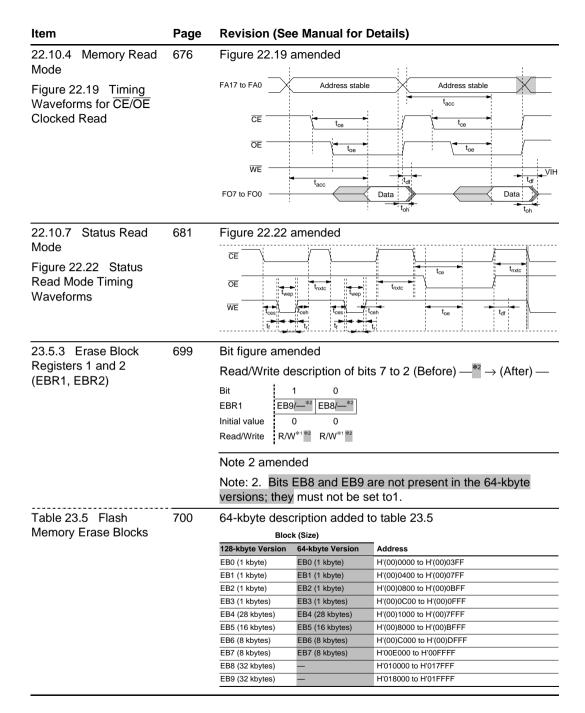
pin

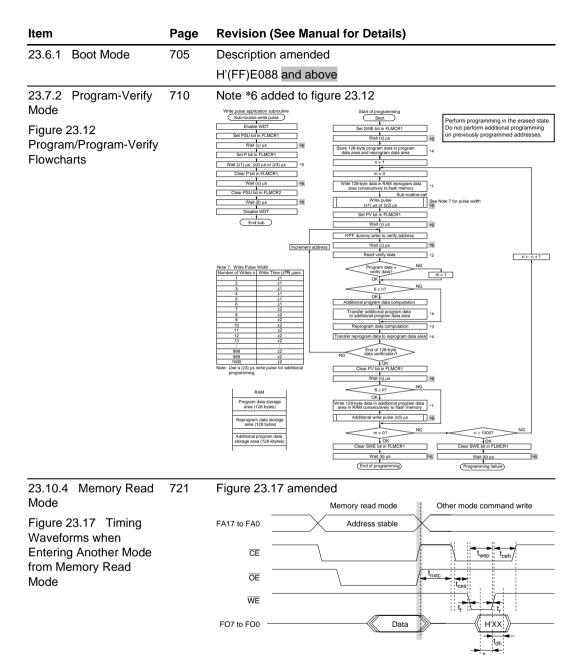
pin

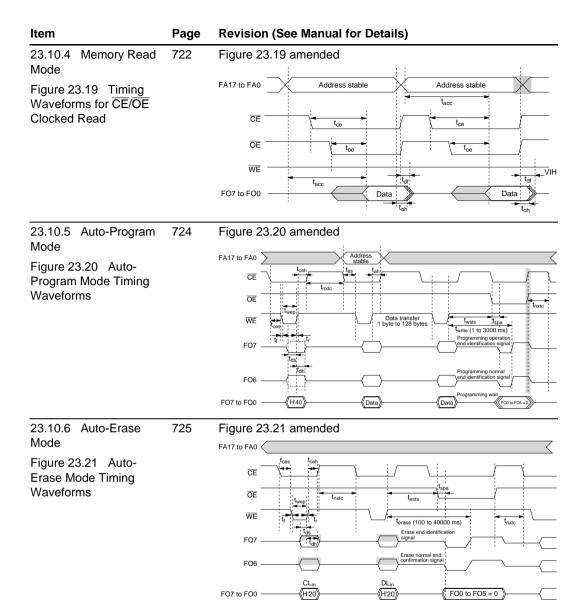
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8.11.3 Pin Functions	270	Table	8.23	ame	nded	dt										
Table 8.23 Port A Pin		Pin		Selection Method and Pin Functions												
Functions		PA1/A17 CIN9	/KIN9/			nction is switched as shown below according to the combination of node, the IOSE bit in SYSCR, and bit PA1DDR.							tion of			
				Oper	rating e		des 1, E = 0), 3		Mode 2 (EXP	E =	= 1)				
				PA1DDR		0	1	0			1					
				IOSE		_	_		0				1			
				Pin f	unction	PA1	PA1	PA1		.17			PA1			
						input pin	output pin	input pi			in	ou	tput pin			
							KIN9 inpu									
				This	oin can	always be us	ed as the KIN9	or CIN9	input pin							
		PA0/A16	6/KIN8/				ed as shown be bit in SYSCR,			he c	om	bina	tion of			
					rating	Mod	des 1, E = 0), 3		Mode 2 (EXP	E =	= 1)				
				PA0	DDR	0	1	0			•	1				
				IOSE		_	_	_		0			1			
				Pin f	unction	PA0	PA0	PA0	А	16			PA0			
						input pin	output pin	input pi			in	ou	tput pin			
							KIN8 inp									
				This	oin can	always be us	ed as the KIN8	or CIN8	input pin							
Table 9.4 Duty Cycle of Basic Pulse																
10.3 Bus Master	299	Desc	ription	ame	ende	d										
Interface		Ex	ample	2: F	Read	DADRA	\									
		MOV.	W @DA	ADR <i>I</i>	A, R	.0 ; Tra	ınsfer cor	ntents	of DA	٩D	R	A t	o R0			
10.4 Operation	303	Table	10.4	ame	nded	t										
Table 10.4 Settings and		R	esolution		Base	Conversion			Fixed I	DAD	R B	Bits	Conversio			
Operation (Examples		скѕ	Т	CFS	Cycle	Cycle	T _L (if OS :		Precision	E	Bit [Data	Cycle*			
when $\phi = 10 \text{ MHz}$)		Oito	(µs)	0.0	(µs)	(µs)	1, (11 00 1	- ',	(Bits)	3	2	1 () (µs)			
when $\phi = 10$ whiz)		0	0.1	0	6.4	1638.4	Always low level output	(or high)	14				1638.4			
							(DADR = H'O3FD)	0001 to								
				1	25.6	1638.4	1. Always low (level output (DADR = H'0 H'00FF)		14				1638.4			
		1	0.2	0	12.8	3276.8	1. Always low (level output (DADR = H'0 H'03FD)		14				3276.8			
				1	51.2	3276.8	Always low (level output (DADR = H') H'00FF)		14				3276.8			

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16.1.1 Features	492	 Automatic switching from formatless mode to I²C bus format (channel 0 only)
		Description added
16.4 Usage Notes	548	Figure 16.19 amended
Figure 16.19 Flowchart and Timing of Start Condition Instruction Issuance for Retransmission		SCL 9 SDA ACK IRIC [3] Start condition instruction issuance [1] IRIC determination of SCL = low
	550 to	Description added
	557	 Notes on WAIT Function
		 Notes on ICDR Reads and ICCR Access in Slave Transmit Mode
		 Notes on TRS Bit Setting in Slave Mode
		 Notes on Arbitration Lost in Master Mode
		 Notes on Interrupt Occurrence after ACKB Reception
		 Notes on TRS Bit Setting and ICDR Register Access
20.4.3 Input Sampling and A/D Conversion Time	628	Figure 20.5 amended
Figure 20.5 A/D	•	(1)
Conversion Time		•
		Address (2)
		Write signal









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23.10.7 Status Read	727	Figure 23.22 amended												
Mode Figure 23.22 Status Read Mode Timing Waveforms		OE tweep trixic toeh toe tdf												
24.7 Subclock Input	740	Note on Subclock Usage												
Circuit		Description added												
25.12 Usage Notes	764	Section 25.12 added												
26.2.6 Flash Memory	799	Table 26.15 amended												
Characteristics Table 26.15 Flash Memory Characteristics (Programming/erasing		Item Symbol Min Typ Max Unit Test Condition Reprogramming count N _{sec} 100** 10000** Times Times Data retention time*** t _{bsp} 10 - - Years Programming Wait time after SWE-bit setting** x 10 - - μs												
operating range)	800	Notes 8 to 10 added												
		Notes: 8. Minimum number of times for which all characteristics are guaranteed after rewriting (Guarantee rang is 1 to minimum value). 9. Reference value for 25°C (as a guideline, rewriting should normally function up to this value).												
		10. Data retention characteristic when rewriting is performed within the specification range, including the minimum value.												
		main are opcompation range, morating the miniman value.												
	819	Table 26.20 amended												
26.3.3 AC Characteristics Table 26.20 Clock Timing	819													
Characteristics Table 26.20 Clock	819	Table 26.20 amended Condition A Condition B Condition C Condition C												
Characteristics Table 26.20 Clock Timing	819	Table 26.20 amended Condition A Condition B Condition C Item Symbol Min Max Min Max Min Max Unid Test Oscillation settling time in software tool - 10 - 10 - 20 - ms Figure 26.6												
Characteristics Table 26.20 Clock Timing Table 26.25 I ² C Bus		Table 26.20 amended Condition A Condition B Condition C Item Symbol Min Max Min Max Min Max Min Max Min Max Unit Condition C Oscillation settling time at reset (crystal) to colspan="8">10 — 10 — 20 — ms Figure 26.6 Oscillation settling time in software standby (crystal) to colspan="8">Table 26.25 amended Ratings												
Characteristics Table 26.20 Clock Timing Table 26.25 I ² C Bus		Table 26.20 amended Condition A Condition B Condition C												
Characteristics Table 26.20 Clock		Table 26.20 amended Condition A Condition B Condition C Item Symbol Min Max Min Max Min Max Min Max Min Max Unit Condition C Oscillation settling time at reset (crystal) to colspan="8">10 — 10 — 20 — ms Figure 26.6 Oscillation settling time in software standby (crystal) to colspan="8">Table 26.25 amended Ratings												

Item	Page	Revision (See Manual for Details)									
26.3.4 A/D Conversion	831	Note *4 added to table condition									
Characteristics Table 26.27 A/D		Condition C: $V_{cc} = 3.0 \text{ V}$ to $3.6V^{*4}$, $AV_{cc} = 3.0 \text{ V}$ to $3.6 V^{*4}$, $AV_{ref} = 3.0 \text{ V}$ to AV_{cc} , $AV_{cc} = 3.0 \text{ V}$ to AV_{cc} , $AV_{cc} = 3.0 \text{ V}$ to $AV_{cc} = 3.0 \text{ V}$									
Conversion		Note 4 amended									
Characteristics (CIN15 to CIN0 Input: 134/266-State Conversion)		Note: 4. When using CIN, the applicable range is V_{cc} = 3.0 V to 3.6 V,									
26.3.6 Flash Memory	833	Table 26.29 amended									
Characteristics Table 26.29 Flash		Item amended (Before) Wait time after dummy write \rightarrow (After) Wait time after H'FF dummy write									
Memory Characteristics (Programming/erasing		Symbol of wait time after SWE-bit clear (Before) $\Theta \rightarrow$ (After) θ									
operating range)		Item Symbol Min Typ Max Unit Test Condition Reprogramming count N _{vec} 100** 1000** — Times Data retention time**** t _{cse} 10 — Years									
		Programming Wait time after SWE-bit setting ⁸¹ x 1 — μs									
	834	Notes 8 to 10 added									
		Notes: 8. Minimum number of times for which all characteristics are guaranteed after rewriting (Guarantee ranges 1 to minimum value).									
		9. Reference value for 25°C (as a guideline, rewriting should normally function up to this value).									
		Data retention characteristic when rewriting is performed within the specification range, including the minimum value.									
26.4.3 AC	849	Unit of t _{NMIH} amended									
Characteristics		(Before) \longrightarrow (After) ns									
Table 26.35 Control Signal Timing											
Table 26.37 Timing of On-Chip Supporting	854, 855	Units of t_{PRS} , t_{PRH} , t_{FTIS} , t_{FTCS} , t_{TMRS} , t_{TMCS} amended (Before) \longrightarrow (After) ns									
Modules (1)		Units of t_{FTCWL} , t_{TMCWL} , synchronous t_{Sove} amended									
		(Before) \longrightarrow (After) t_{cyc}									
		Unit of t _{sckf} amended									
		(Before) 1.5 \rightarrow (After) t_{cyc}									
26.4.4 A/D Conversion	860	Table condition amended									
Characteristics		Table condition A (Before), Ta = -20 to +75°C (regular									
Table 26.41 A/D Conversion Characteristics (CIN15 to CIN0 Input: 134/266-		specifications), Ta = -40 to $+85^{\circ}$ C (wide-range specifications) \rightarrow (After) Ta = -20 to $+75^{\circ}$ C									
State Conversion)											

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Item	Page	Revision (See Manual for Details)											
26.4.6 Flash Memory Characteristics Table 26.43 Flash Memory Characteristics (Programming/erasing	862												
operating range)	863	Notes 8 to 10 added Notes: 8. Minimum number of times for which all characteristics are guaranteed after rewriting (Guarantee rang is 1 to minimum value). 9. Reference value for 25°C (as a guideline, rewriting should											
		normally function up to this value). 10. Data retention characteristic when rewriting is performed within the specification range, including the minimum value.											
26.5.3 AC Characteristics Table 26.49 Control Signal Timing	877	Unit of $t_{\text{\tiny NMIH}}$ amended (Before) (blank) \rightarrow (After) ns											
26.5.6 Flash Memory Characteristics Table 26.55 Flash Memory Characteristics	885												
(Programming/erasing operating range)	886	Notes 8 to 10 added Notes: 8. Minimum number of times for which all characteristics are guaranteed after rewriting (Guarantee range is 1 to minimum value). 9. Reference value for 25°C (as a guideline, rewriting should normally function up to this value). 10. Data retention characteristic when rewriting is performed within the specification range, including the minimum value.											
26.6.3 AC Characteristics Table 26.61 Control Signal Timing	900	Unit of $t_{\text{\tiny NMIH}}$ amended (Before) (blank) \rightarrow (After) ns											

Item **Page Revision (See Manual for Details)** 26.6.6 Flash Memory 908 Table 26.67 amended Characteristics Symbol of wait time after SWE-bit clear (Before) $\Theta \rightarrow$ (After) θ Table 26.67 Flash Test Symbol Min Тур Max Unit Condition Memory Characteristics 100* 10000*9 Times Reprogramming count Ν... (Programming/erasing Data retention time 10 Years operating range) Programming Wait time after SWE-bit setting 1 909 Notes 8 to 10 added Notes: 8. Minimum number of times for which all characteristics are guaranteed after rewriting (Guarantee range is 1 to minimum value). 9. Reference value for 25°C (as a guideline, rewriting should normally function up to this value). 10. Data retention characteristic when rewriting is performed within the specification range, including the minimum value. 26.7.2 Clock Timing 912 Figure 26.6 amended Figure 26.6 Oscillation EXTAL Settling Timing t_{DEXT} V_{CC} STBY tosc. tosc₁ RES 26.7.5 Timing of On-924 Figure 26.28 amended Chip Supporting Modules Figure 26.28 I²C Bus SDA0, Interface Input/Output SDA1 Timing (Option) t_{BUF}; t_{STAH} t_{sci h} SCL0. SCL₁ S* P^* tSCLL t_{Sf} t_{of} t_{SCL}

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A.1 Instruction

Table A.1 Instruction Set

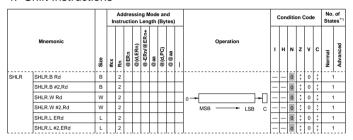
930 Table A.1 amended

2. Arithmetic Instructions

			Addressing Mode and Instruction Length (Bytes)											Con	No. of States*1					
	Mnemonic	Size	#xx	R.	@ERn	@(d,ERn)	@-ERn/@ERn+	@aa	@(q,PC)	@@aa	1	Operation	ı	н	N	z	v	С	Normal	Advanced
EXTU	EXTU.W Rd	w		2								0 → (<bits 15="" 8="" to=""> of Rd16)</bits>	_	-	0	‡	0	-	1	
	EXTU.L ERd	L		2								0 → (<bits 16="" 31="" to=""> of ERd32)</bits>	_	_	0	‡	0	_	1	
TAS	TAS @ERd*3	В			4							@ERd-0 → CCR set, (1) → (<bit 7=""> of @ERd)</bit>	-	-	1	Ţ	0	-	4	

933 Table A.1 amended

4. Shift Instructions



939 Table A.1 amended

6. Branch Instructions

Mnemonic			Addressing Mode and Instruction Length (Bytes)											Condition				•	No. of States*1	
			××#	Rn	@ERn	@(d,ERn)	@-ERn/@ERn+	@aa	@(d,PC)	@@aa	1	Operation		н	N	z	v	c	Normal	Advanced
JMP	JMP @ERn	-	Г	Г	2	Г						PC←ERn	-	-	-	-	_	-	2	
	JMP @aa:24	_						4				PC←aa:24	_	_	_	-	_	-	3	3
	JMP @@aa:8	_	Г	Г						2		PC←@aa:8	_	_	_	_	_	-	4	5
BSR	BSR d:8	-							2			PC→@-SP,PC←PC+d:8	-	-	-	-	_	-	3	4
	BSR d:16	-							4			PC→@-SP,PC←PC+d:16	_	_	_	_	_	-	4	5
JSR	JSR @ERn	_			2							PC→@-SP,PC←ERn	_	_	_	-	_	-	3	4
	JSR @aa:24	-				П		4				PC→@-SP,PC←aa:24	_	_	_	_	_	-	4	5
	JSR @@aa:8	-								2		PC→@-SP,PC←@aa:8	-	-	-	-	_	-	4	6
RTS	RTS	-									2	PC←@SP+	_	_	_	_	_	-	4	5

A.2 Instruction Codes

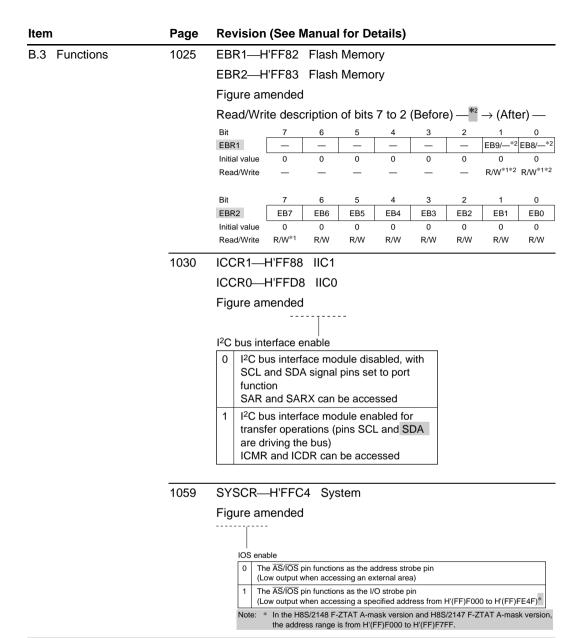
Table A.2 Instruction Codes

Table A.2 amended

949

Instruc-	Mnemonic	Size									Instruction	on Format
tion		Size	1st Byte		2nd Byte		3rd Byte		4th Byte		5th Byte	6th Byte
LDC	LDC @aa:16,CCR	W	0	1	4	0	6	В	0	0	al	bs
	LDC @aa:16,EXR	W	0	1	4	1	6	В	0	0	al	bs

Item	Page	Revision (See Manual for Details)
B.3 Functions	1013	Subheading amended
		KBCOMP—H'FEE4 IrDA/Expansion A/D
	1016	ISR—H'FEEB Interrupt Controller
		Figure amended
		IDOZ to IDOO floro
	0	RQ7 to IRQ0 flags
		Note: * When a product, in which a DTC is incorporated, is used in the following settings, the corresponding flag bit is not automatically cleared even when exception handling, which is a clear condition, is executed and the bit is held at 1.
		(1) When DTCEA3 is set to 1 (ADI is set to an interrupt source), IRQ4F flag is not automatically cleared.
		(2) When DTCEA2 is set to 1 (ICIA is set to an interrupt source), IRQ5F flag is not automatically cleared.
		(3) When DTCEA1 is set to 1 (ICIB is set to an interrupt source), IRQ6F flag is not automatically cleared.
		(4) When DTCEA0 is set to 1 (OCIA is set to an interrupt source), IRQ7F flag is not automatically cleared.
		When activation interrupt sources of DTC and IRQ interrupts are used with the above combinations, clear the interrupt flag by software in the interrupt handling routine of the corresponding IRQ.
	1019	ABRKCR—H'FEF4 Interrupt Controller
		Read/Write description amended
		Bit 7 (Before) $R/W \rightarrow (After) R$
	1021	FLMCR1—H'FF80 Flash Memory
		Initial value description amended
		Bit 7 (Before) — → (After) 1



Item	Page	Revision (See Manual for Details)
B.3 Functions	1077	TICRR—H'FFF2 TMRX
		TICRF—H'FFF3 TMRX
		Figure amended
		(Before) Stores TCNT value at fall of external trigger input \to (After) Stores TCNT value at fall of external reset input
	1080	STR1—H'FFF6 HIF
		STR2—H'FFFE HIF
		Slave R/W description amended
		Bit 0 (Before) $R \rightarrow (After) R/(W)$
C.2 Port 2 Block Diagrams Figure C.4 Port 2 Block Diagram (Pin P27)	1089	Figure C.4 amended Hardware standby Mode 1
Appendix F Product Code Lineup	1128	Package code in table F.1 amended
Table F.1 H8S/2148		HD64F2144ATE20 (Before) FP-100B → (After) TFP-100B
Group and H8S/2144 Group Product Code Lineup		HD64F2144AVFA10 (Before) TFP-100B → (After) FP-100B
Appendix G Package Dimensions	1129	Figure G.1 replaced
Figure G.1 Package Dimensions (FP-100B)		
Figure G.2 Package	1130	Figure G.2 replaced

Dimensions (TFP-100B)

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Section 1 Overview

1.1 Overview

This LSI comprise microcomputers (MCUs) built around the H8S/2000 CPU, employing Renesas Technology proprietary architecture, and equipped with supporting modules on-chip.

The H8S/2000 CPU has an internal 32-bit architecture, is provided with sixteen 16-bit general registers and a concise, optimized instruction set designed for high-speed operation, and can address a 16-Mbyte linear address space. The instruction set is upward-compatible with H8/300 and H8/300H CPU instructions at the object-code level, facilitating migration from the H8/300, H8/300L, or H8/300H Series.

On-chip supporting modules required for system configuration include a data transfer controller (DTC) bus master, ROM and RAM, a 16-bit free-running timer module (FRT), 8-bit timer module (TMR), watchdog timer module (WDT), two PWM timers (PWM and PWMX), serial communication interface (SCI), PS/2-compatible keyboard buffer controller, host interface (HIF), D/A converter (DAC), A/D converter (ADC), and I/O ports. An I²C bus interface (IIC) can also be incorporated as an option.

The on-chip ROM is either flash memory (F-ZTAT^{TM*}) or mask ROM, with a capacity of 128, 96, or 64 kbytes. ROM is connected to the CPU via a 16-bit data bus, enabling both byte and word data to be accessed in one state. Instruction fetching has been speeded up, and processing speed increased.

Three operating modes, modes 1 to 3, are provided, and there is a choice of address space and single-chip mode or externally expanded modes.

The features of this LSI are shown in table 1.1.

Note: * F-ZTAT is a trademark of Renesas Technology Corp.

Table 1.1 Overview

Table 1.1	O'CL'VICW
Item	Specifications
CPU	General-register architecture
	 Sixteen 16-bit general registers (also usable as sixteen 8-bit registers or eight 32-bit registers)
	 High-speed operation suitable for real-time control
	 — Maximum operating frequency: 20 MHz/5 V, 10 MHz/3 V
	 High-speed arithmetic operations
	8/16/32-bit register-register add/subtract: 50 ns (20-MHz operation)
	16×16 -bit register-register multiply: 1000 ns (20-MHz operation)
	32 ÷ 16-bit register-register divide: 1000 ns (20-MHz operation)
	 Instruction set suitable for high-speed operation
	 Sixty-five basic instructions
	 8/16/32-bit transfer/arithmetic and logic instructions
	 Unsigned/signed multiply and divide instructions
	 — Powerful bit-manipulation instructions
	 Two CPU operating modes
	 Normal mode: 64-kbyte address space

Advanced mode: 16-Mbyte address space

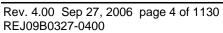
Operating modes

• Three MCU operating modes

			Externa	al Data Bus
CPU Operating Mode	Description	On-Chip ROM	Initial Value	Maximum Value
Normal	Expanded mode with on-chip ROM disabled	Disabled	8 bits	16 bits
Advanced	Single-chip mode	Enabled	None	None
	Expanded mode with on-chip ROM enabled	Enabled	8 bits	16 bits
Normal	Single-chip mode	Enabled	None	None
	Expanded mode with on-chip ROM enabled	Enabled	8 bits	16 bits
	Mode Normal Advanced	Mode Description Normal Expanded mode with on-chip ROM disabled Advanced Single-chip mode Expanded mode with on-chip ROM enabled Normal Single-chip mode Expanded mode with on-chip ROM enabled	Mode Description ROM Normal Expanded mode with on-chip ROM disabled Disabled Advanced Single-chip mode Enabled Enabled Expanded mode with on-chip ROM enabled Enabled Normal Single-chip mode Enabled Enabled Expanded mode with on-chip ROM Enabled	CPU Operating Mode Description Normal Expanded mode with on-chip ROM disabled Advanced Single-chip mode Expanded mode with on-chip ROM enabled Normal Single-chip mode Expanded mode with on-chip ROM enabled Normal Single-chip mode Expanded mode with on-chip ROM enabled Enabled None Expanded mode with on-chip ROM enabled Enabled None

Item	Specifications
Bus controller	2-state or 3-state access space can be designated for external expansion areas
	Number of program wait states can be set for external expansion areas
Data transfer	Can be activated by internal interrupt or software
controller (DTC) (H8S/2148 Group)	 Multiple transfers or multiple types of transfer possible for one activation source
	Transfer possible in repeat mode, block transfer mode, etc.
	Request can be sent to CPU for interrupt that activated DTC
16-bit free-running timer module	One 16-bit free-running counter (also usable for external event counting)
(FRT: 1 channel)	Two output compare outputs
	Four input capture inputs (with buffer operation capability)
8-bit timer module	Each channel has:
(2 channels: TMR0, TMR1)	One 8-bit up-counter (also usable for external event counting)
TWICT	Two timer constant registers
	The two channels can be connected
Timer connection and	Input/output and FRT, TMR1, TMRX, TMRY can be interconnected
8-bit timer module (TMR) (2 channels: TMRX, TMRY)	 Measurement of input signal or frequency-divided waveform pulse width and cycle (FRT, TMR1)
(Timer connection and TMRX provided in	 Output of waveform obtained by modification of input signal edge (FRT, TMR1)
H8S/2148 Group)	Determination of input signal duty cycle (TMRX)
	 Output of waveform synchronized with input signal (FRT, TMRX, TMRY)
	Automatic generation of cyclical waveform (FRT, TMRY)
Watchdog timer	Watchdog timer or interval timer function selectable
module (WDT: 2 channels)	Subclock operation capability (channel 1 only)
8-bit PWM timer	Up to 16 outputs
(PWM) (H8S/2148 Group and	 Pulse duty cycle settable from 0 to 100%
H8S/2147N)	Resolution: 1/256
	1.25 MHz maximum carrier frequency (20-MHz operation)

Item	Specifications
14-bit PWM timer	Up to 2 outputs
(PWMX)	Resolution: 1/16384
	• 312.5 kHz maximum carrier frequency (20-MHz operation)
Serial communication	Asynchronous mode or synchronous mode selectable
interface (SCI: 2 channels, SCI0 and SCI1)	Multiprocessor communication function
SCI with IrDA:	Asynchronous mode or synchronous mode selectable
1 channel (SCI2)	Multiprocessor communication function
	Compatible with IrDA specification version 1.0
	TxD and RxD encoding/decoding in IrDA format
Keyboard buffer	Compatible with PS/2 interface
controller (PS2: 3 channels)	Direct manipulation of transmission output by software
(H8S/2148 Group,	Receive data input to 8-bit shift register
H8S/2147N)	 Data receive completed interrupt, parity error detection, stop bit monitoring
Host interface (HIF)	8-bit host interface (ISA) port
(H8S/2148 Group, H8S/2147N)	 Five host interrupt requests (HIRQ11, HIRQ1, HIRQ12, HIRQ3, HIRQ4)
	Normal and fast A20 gate output
	 Four register sets (each comprising two data registers and two status registers)
Keyboard controller	Matrix keyboard control using keyboard scan with wakeup interrupt and sense port configuration
A/D converter	Resolution: 10 bits
	• Input:
	 8 channels (dedicated analog pins)
	 16 channels (same pins as keyboard sense port)
	 High-speed conversion: 6.7 µs minimum conversion time (20-MHz operation)
	Single or scan mode selectable
	Sample-and-hold function
	A/D conversion can be activated by external trigger or timer trigger





Item	Specifications					
D/A converter	Resolution: 8 bits					
	Output: 2 channels					
I/O ports	74 input/output pins	(including 24 v	vith LED drive ca	apability)		
	 8 input-only pins 					
	 VCCB (separate po 	wer supply) dri	ve pins among l	O pins (H8S/2148		
	Group and H8S/214	17N)				
Memory	Flash memory or ma	ask ROM				
	 High-speed static R 	AM				
	Product Name	ROM	RAM	<u></u>		
	H8S/2144, H8S/2148	128 kbytes	4 kbytes	<u> </u>		
	H8S/2143	96 kbytes	4 kbytes	<u> </u>		
	H8S/2142, H8S/2147, H8S/2147N	64 kbytes	2 kbytes	_		
Interrupt controller	Nine external interru44 internal interruptThree priority levels	sources	RQ0 to IRQ7)			
Power-down state	Medium-speed mod	le		-		
	Sleep mode					
	Module stop mode					
	Software standby mode					
	Hardware standby mode					
	Subclock operation					
Clock pulse generator	Built-in duty correction	on circuit				
Packages	100-pin plastic QFP (FP-100B)					
	100-pin plastic TQFP (TFP-100B)					
I ² C bus interface	Conforms to Philips I ² C bus interface standard					
(IIC: 2 channels) (option in H8S/2148	Single master mode/slave mode					
Group and	Arbitration lost condition can be identified					
H8S/2147N)	Supports two slave	addresses				

Product lineup		Product Code*2			
(preliminary)	Group	Mask ROM Versions	F-ZTAT Versions	ROM/RAM (Bytes)	Packages
	H8S/2148	HD6432148S	HD64F2148 HD64F2148V*2	128 k/4 k	FP-100B, TFP-100B
		HD6432148SW*1	HD64F2148A HD64F2148AV*2		
		HD6432147S	HD64F2147A	64 k/2 k	_
		HD6432147SW*1	HD64F2147AV*2		
	H8S/2147N	_	HD64F2147N HD64F2147NV*2	64 k/2 k	_
	H8S/2144	HD6432144S	HD64F2144 HD64F2144V*2	128 k/4 k	_
			HD64F2144A HD64F2144AV*2		
		HD6432143S	_	96 k/4 k	_
		HD6432142	HD64F2142R HD64F2142RV*2	64 k/2 k	
	Notes: 1.	W indicates the I ² C	bus option.		

1.2 Internal Block Diagram

An internal block diagram of the H8S/2148 Group is shown in figure 1.1 (a), an internal block diagram of the H8S/2147N is shown in figure 1.1 (b), and an internal block diagram of the H8S/2144 Group in figure 1.1 (c).

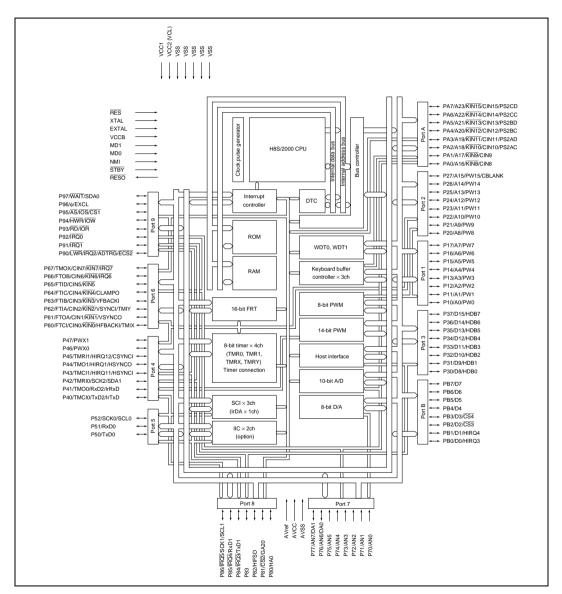


Figure 1.1 (a) Internal Block Diagram of H8S/2148 Group

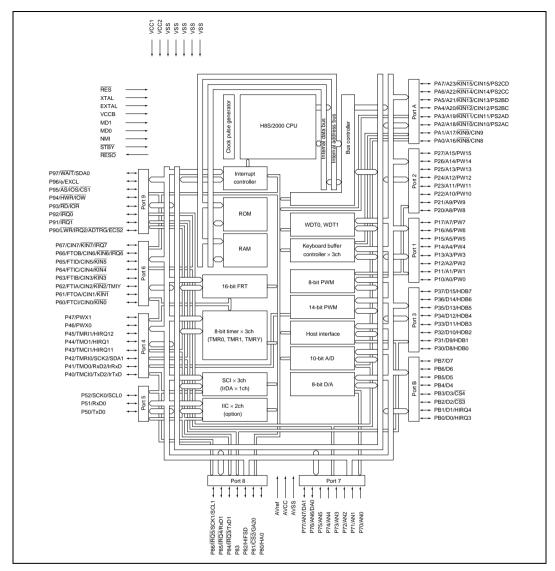


Figure 1.1 (b) Internal Block Diagram of H8S/2147N

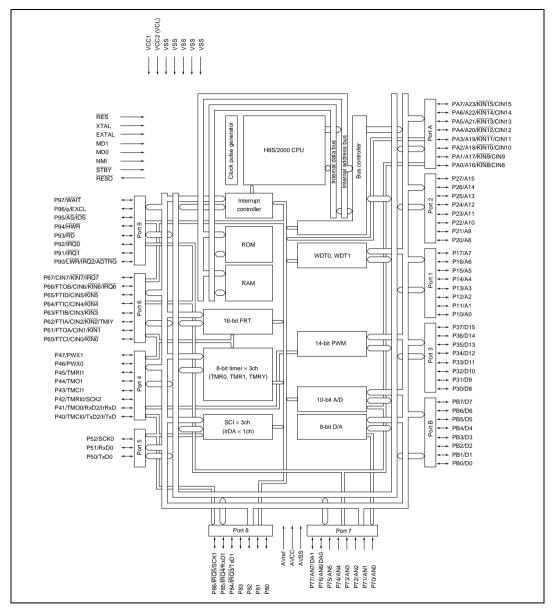


Figure 1.1 (c) Internal Block Diagram of H8S/2144 Group

1.3 Pin Arrangement and Functions

1.3.1 Pin Arrangement

The pin arrangement of the H8S/2148 Group is shown in figure 1.2 (a), the pin arrangement of the H8S/2147N is shown in figure 1.2 (b), and the pin arrangement of the H8S/2144 Group in figure 1.2 (c).

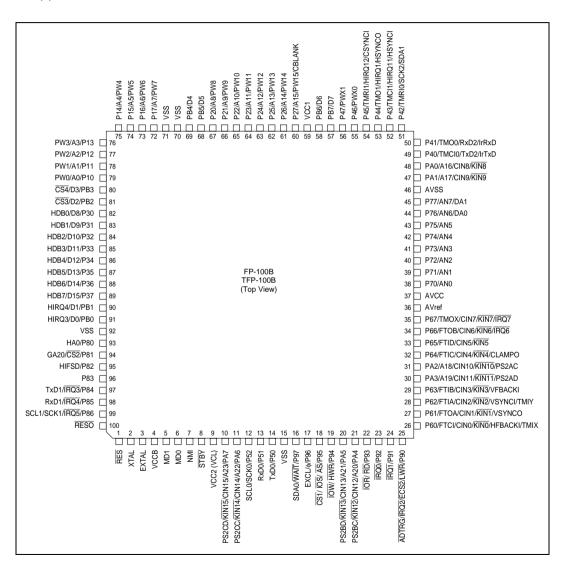


Figure 1.2 (a) Pin Arrangement of H8S/2148 Group (FP-100B, TFP-100B: Top View)

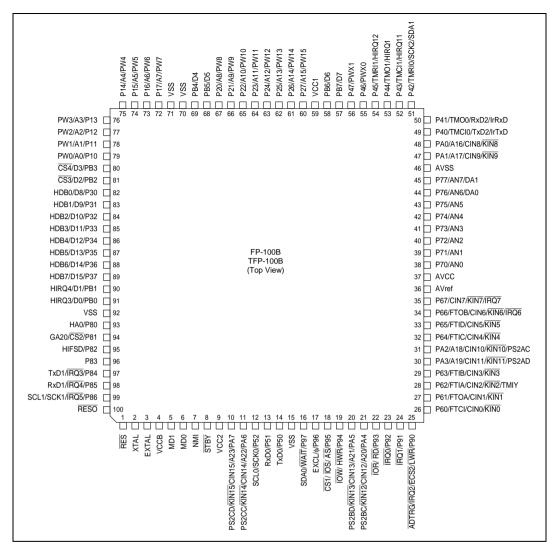


Figure 1.2 (b) Pin Arrangement of H8S/2147N (FP-100B, TFP-100B: Top View)

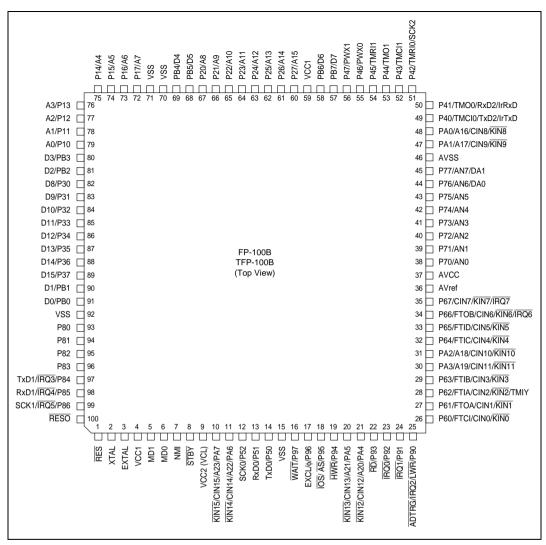


Figure 1.2 (c) Pin Arrangement of H8S/2144 Group (FP-100B, TFP-100B: Top View)

1.3.2 Pin Functions in Each Operating Mode

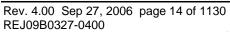
Tables 1.2 (a), (b) and (c) show the pin functions of the H8S/2148 Group, H8S/2147N, and H8S/2144 Group in each of the operating modes.

Table 1.2 (a) H8S/2148 Group Pin Functions in Each Operating Mode

Pin No.	Expanded Modes Single-Chip Modes			
FP-100B TFP-100B	Mode 1	Mode 2 (EXPE = 1) Mode 3 (EXPE = 1)	Mode 2 (EXPE = 0) Mode 3 (EXPE = 0)	Flash Memory Writer Mode
1	RES	RES	RES	RES
2	XTAL	XTAL	XTAL	XTAL
3	EXTAL	EXTAL	EXTAL	EXTAL
4	VCCB	VCCB	VCCB	VCC
5	MD1	MD1	MD1	VSS
6	MD0	MD0	MD0	VSS
7	NMI	NMI	NMI	FA9
8	STBY	STBY	STBY	VCC
9	VCC2 (VCL)	VCC2 (VCL)	VCC2 (VCL)	VCC
10	PA7/CIN15/ KIN15/PS2CD	A23/PA7/CIN15/ KIN15/PS2CD	PA7/CIN15/ KIN15/PS2CD	NC
11	PA6/CIN14/ KIN14/PS2CC	A22/PA6/CIN14/ KIN14/PS2CC	PA6/CIN14/ KIN14/PS2CC	NC
12	P52/SCK0/SCL0	P52/SCK0/SCL0	P52/SCK0/SCL0	NC
13	P51/RxD0	P51/RxD0	P51/RxD0	FA17
14	P50/TxD0	P50/TxD0	P50/TxD0	NC
15	VSS	VSS	VSS	VSS
16	P97/WAIT/SDA0	P97/WAIT/SDA0	P97/SDA0	VCC
17	P96/ø/EXCL	P96/ø/EXCL	P96/φ/EXCL	NC
18	AS/IOS	AS/IOS	P95/CS1	FA16
19	HWR	HWR	P94/IOW	FA15
20	PA5/CIN13/ KIN13/PS2BD	A21/PA5/CIN13/ KIN13/PS2BD	PA5/CIN13/ KIN13/PS2BD	NC
21	PA4/CIN12/ KIN12/PS2BC	A20/PA4/CIN12/ KIN12/PS2BC	PA4/CIN12/ KIN12/PS2BC	NC

Din	Name
- FIII	Name

Pin No.	Expand	ded Modes	Single-Chip Modes	
FP-100B TFP-100B	Mode 1	Mode 2 (EXPE = 1) Mode 3 (EXPE = 1)	Mode 2 (EXPE = 0) Mode 3 (EXPE = 0)	Flash Memory Writer Mode
22	RD	RD	P93/IOR	WE
23	P92/IRQ0	P92/IRQ0	P92/IRQ0	VSS
24	P91/IRQ1	P91/IRQ1	P91/IRQ1	VCC
25	LWR/P90/IRQ2/ ADTRG	LWR/P90/IRQ2/ ADTRG	P90/IRQ2/ADTRG/ ECS2	VCC
26	P60/FTCI/CIN0/ KIN0/TMIX/ HFBACKI	P60/FTCI/CIN0/ KIN0/TMIX/ HFBACKI	P60/FTCI/CIN0/ KIN0/TMIX/ HFBACKI	NC
27	P61/FTOA/CIN1/ KIN1/VSYNCO	P61/FTOA/CIN1/ KIN1/VSYNCO	P61/FTOA/CIN1/ KIN1/VSYNCO	NC
28	P62/FTIA/CIN2/ KIN2/TMIY/ VSYNCI	P62/FTIA/CIN2/ KIN2/TMIY/ VSYNCI	P62/FTIA/CIN2/ KIN2/TMIY/ VSYNCI	NC
29	P63/FTIB/CIN3/ KIN3/VFBACKI	P63/FTIB/CIN3/ KIN3/VFBACKI	P63/FTIB/CIN3/ KIN3/VFBACKI	NC
30	PA3/CIN11/ KIN11/PS2AD	A19/PA3/CIN11/ KIN11/PS2AD	PA3/CIN11/ KIN11/PS2AD	NC
31	PA2/CIN10/ KIN10/PS2AC	A18/PA2/CIN10/ KIN10/PS2AC	PA2/CIN10/ KIN10/PS2AC	NC
32	P64/FTIC/CIN4/ KIN4/CLAMPO	P64/FTIC/CIN4/ KIN4/CLAMPO	P64/FTIC/CIN4/ KIN4/CLAMPO	NC
33	P65/FTID/CIN5/ KIN5	P65/FTID/CIN5/ KIN5	P65/FTID/CIN5/ KIN5	NC
34	P66/FTOB/CIN6/ KIN6/IRQ6	P66/FTOB/CIN6/ KIN6/IRQ6	P66/FTOB/CIN6/ KIN6/IRQ6	NC
35	P67/TMOX/CIN7/ KIN7/IRQ7	P67/TMOX/CIN7/ KIN7/IRQ7	P67/TMOX/CIN7/ KIN7/IRQ7	VSS
36	AVref	AVref	AVref	VCC
37	AVCC	AVCC	AVCC	VCC
38	P70/AN0	P70/AN0	P70/AN0	NC
39	P71/AN1	P71/AN1	P71/AN1	NC
40	P72/AN2	P72/AN2	P72/AN2	NC
41	P73/AN3	P73/AN3	P73/AN3	NC

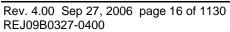




Pin No.	Expar	nded Modes	Single-Chip Modes	
FP-100B TFP-100B	Mode 1	Mode 2 (EXPE = 1) Mode 3 (EXPE = 1)	Mode 2 (EXPE = 0) Mode 3 (EXPE = 0)	Flash Memory Writer Mode
42	P74/AN4	P74/AN4	P74/AN4	NC
43	P75/AN5	P75/AN5	P75/AN5	NC
44	P76/AN6/DA0	P76/AN6/DA0	P76/AN6/DA0	NC
45	P77/AN7/DA1	P77/AN7/DA1	P77/AN7/DA1	NC
46	AVSS	AVSS	AVSS	VSS
47	PA1/CIN9/KIN9	A17/PA1/CIN9/ KIN9	PA1/CIN9/KIN9	NC
48	PA0/CIN8/KIN8	A16/PA0/CIN8/KIN8	PA0/CIN8/KIN8	NC
49	P40/TMCI0/ TxD2/IrTxD	P40/TMCI0/ TxD2/IrTxD	P40/TMCI0/ TxD2/IrTxD	NC
50	P41/TMO0/ RxD2/IrRxD	P41/TMO0/ RxD2/IrRxD	P41/TMO0/ RxD2/IrRxD	NC
51	P42/TMRI0/ SCK2/SDA1	P42/TMRI0/ SCK2/SDA1	P42/TMRI0/ SCK2/SDA1	NC
52	P43/TMCI1/ HSYNCI	P43/TMCI1/ HSYNCI	P43/TMCI1/HIRQ11/ HSYNCI	NC
53	P44/TMO1/ HSYNCO	P44/TMO1/ HSYNCO	P44/TMO1/HIRQ1/ HSYNCO	NC
54	P45/TMRI1/ CSYNCI	P45/TMRI1/ CSYNCI	P45/TMRI1/HIRQ12/ CSYNCI	NC
55	P46/PWX0	P46/PWX0	P46/PWX0	NC
56	P47/PWX1	P47/PWX1	P47/PWX1	NC
57	PB7/D7	PB7/D7	PB7	NC
58	PB6/D6	PB6/D6	PB6	NC
59	VCC1	VCC1	VCC1	VCC
60	A15	A15/P27/PW15/ CBLANK	P27/PW15/ CBLANK	CE
61	A14	A14/P26/PW14	P26/PW14	FA14
62	A13	A13/P25/PW13	P25/PW13	FA13
63	A12	A12/P24/PW12	P24/PW12	FA12
64	A11	A11/P23/PW11	P23/PW11	FA11

Din	Name

Pin No.	Expar	nded Modes	Single-Chip Modes	
FP-100B TFP-100B	Mode 1	Mode 2 (EXPE = 1) Mode 3 (EXPE = 1)	Mode 2 (EXPE = 0) Mode 3 (EXPE = 0)	Flash Memory Writer Mode
65	A10	A10/P22/PW10	P22/PW10	FA10
66	A9	A9/P21/PW9	P21/PW9	ŌĒ
67	A8	A8/P20/PW8	P20/PW8	FA8
68	PB5/D5	PB5/D5	PB5	NC
69	PB4/D4	PB4/D4	PB4	NC
70	VSS	VSS	VSS	VSS
71	VSS	VSS	VSS	VSS
72	A7	A7/P17/PW7	P17/PW7	FA7
73	A6	A6/P16/PW6	P16/PW6	FA6
74	A5	A5/P15/PW5	P15/PW5	FA5
75	A4	A4/P14/PW4	P14/PW4	FA4
76	A3	A3/P13/PW3	P13/PW3	FA3
77	A2	A2/P12/PW2	P12/PW2	FA2
78	A1	A1/P11/PW1	P11/PW1	FA1
79	A0	A0/P10/PW0	P10/PW0	FA0
80	PB3/D3	PB3/D3	PB3/CS4	NC
81	PB2/D2	PB2/D2	PB2/CS3	NC
82	D8	D8	P30/HDB0	FO0
83	D9	D9	P31/HDB1	FO1
84	D10	D10	P32/HDB2	FO2
85	D11	D11	P33/HDB3	FO3
86	D12	D12	P34/HDB4	FO4
87	D13	D13	P35/HDB5	FO5
88	D14	D14	P36/HDB6	FO6
89	D15	D15	P37/HDB7	FO7
90	PB1/D1	PB1/D1	PB1/HIRQ4	NC
91	PB0/D0	PB0/D0	PB0/HIRQ3	NC
92	VSS	VSS	VSS	VSS
93	P80	P80	P80/HA0	NC





Pin No.	Expanded Modes		Single-Chip Modes	
FP-100B TFP-100B	Mode 1	Mode 2 (EXPE = 1) Mode 3 (EXPE = 1)	Mode 2 (EXPE = 0) Mode 3 (EXPE = 0)	Flash Memory Writer Mode
94	P81	P81	P81/CS2/GA20	NC
95	P82	P82	P82/HIFSD	NC
96	P83	P83	P83	NC
97	P84/IRQ3/TxD1	P84/IRQ3/TxD1	P84/IRQ3/TxD1	NC
98	P85/IRQ4/RxD1	P85/IRQ4/RxD1	P85/IRQ4/RxD1	NC
99	P86/IRQ5/SCK1/ SCL1	P86/IRQ5/SCK1/ SCL1	P86/IRQ5/SCK1/ SCL1	NC
100	RESO	RESO	RESO	NC

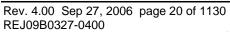
Table 1.2 (b) H8S/2147N Pin Functions in Each Operating Mode

Pin No.	Expan	ded Modes	Single-Chip Modes	
FP-100B TFP-100B	Mode 1	Mode 2 (EXPE = 1) Mode 3 (EXPE = 1)	Mode 2 (EXPE = 0) Mode 3 (EXPE = 0)	Flash Memory Writer Mode
1	RES	RES	RES	RES
2	XTAL	XTAL	XTAL	XTAL
3	EXTAL	EXTAL	EXTAL	EXTAL
4	VCCB	VCCB	VCCB	VCC
5	MD1	MD1	MD1	VSS
6	MD0	MD0	MD0	VSS
7	NMI	NMI	NMI	FA9
8	STBY	STBY	STBY	VCC
9	VCC2	VCC2	VCC2	VCC
10	PA7/CIN15/ KIN15/PS2CD	A23/PA7/CIN15/ KIN15/PS2CD	PA7/CIN15/ KIN15/PS2CD	NC
11	PA6/CIN14/ KIN14/PS2CC	A22/PA6/CIN14/ KIN14/PS2CC	PA6/CIN14/ KIN14/PS2CC	NC
12	P52/SCK0/SCL0	P52/SCK0/SCL0	P52/SCK0/SCL0	NC
13	P51/RxD0	P51/RxD0	P51/RxD0	FA17
14	P50/TxD0	P50/TxD0	P50/TxD0	NC
15	VSS	VSS	VSS	VSS
16	P97/WAIT/SDA0	P97/WAIT/SDA0	P97/SDA0	VCC
17	φ/P96/EXCL	φ/P96/EXCL	P96/φ/EXCL	NC
18	AS/IOS	AS/IOS	P95/CS1	FA16
19	HWR	HWR	P94/ IOW	FA15
20	PA5/CIN13/ KIN13/PS2BD	A21/PA5/CIN13/ KIN13/PS2BD	PA5/CIN13/ KIN13/PS2BD	NC
21	PA4/CIN12/ KIN12/PS2BC	A20/PA4/CIN12/ KIN12/PS2BC	PA4/CIN12/ KIN12/PS2BC	NC
22	RD	RD	P93/IOR	WE
23	P92/IRQ0	P92/IRQ0	P92/IRQ0	VSS
24	P91/IRQ1	P91/IRQ1	P91/IRQ1	VCC

Pin No.	Expan	ded Modes	Single-Chip Modes	
FP-100B TFP-100B	Mode 1	Mode 2 (EXPE = 1) Mode 3 (EXPE = 1)	Mode 2 (EXPE = 0) Mode 3 (EXPE = 0)	Flash Memory Writer Mode
25	LWR/P90/IRQ2/ ADTRG	LWR/P90/IRQ2/ ADTRG	P90/IRQ2/ADTRG/ ECS2	VCC
26	P60/FTCI/CIN0/ KIN0	P60/FTCI/CIN0/ KIN0	P60/FTCI/CIN0/ KIN0	NC
27	P61/FTOA/CIN1/ KIN1	P61/FTOA/CIN1/ KIN1	P61/FTOA/CIN1/ KIN1	NC
28	P62/FTIA/CIN2/ KIN2/TMIY	P62/FTIA/CIN2/ KIN2/TMIY	P62/FTIA/CIN2/ KIN2/TMIY	NC
29	P63/FTIB/CIN3/ KIN3	P63/FTIB/CIN3/ KIN3	P63/FTIB/CIN3/ KIN3	NC
30	PA3/CIN11/ KIN11/PS2AD	A19/PA3/CIN11/ KIN11/PS2AD	PA3/CIN11/ KIN11/PS2AD	NC
31	PA2/CIN10/ KIN10/PS2AC	A18/PA2/CIN10/ KIN10/PS2AC	PA2/CIN10/ KIN10/PS2AC	NC
32	P64/FTIC/CIN4/ KIN4	P64/FTIC/CIN4/ KIN4	P64/FTIC/CIN4/ KIN4	NC
33	P65/FTID/CIN5/ KIN5	P65/FTID/CIN5/ KIN5	P65/FTID/CIN5/ KIN5	NC
34	P66/FTOB/CIN6/ KIN6/IRQ6	P66/FTOB/CIN6/ KIN6/IRQ6	P66/FTOB/CIN6/ KIN6/IRQ6	NC
35	P67/CIN7/KIN7/ IRQ7	P67/CIN7/KIN7/ IRQ7	P67/CIN7/KIN7/ IRQ7	VSS
36	AVref	AVref	AVref	VCC
37	AVCC	AVCC	AVCC	VCC
38	P70/AN0	P70/AN0	P70/AN0	NC
39	P71/AN1	P71/AN1	P71/AN1	NC
40	P72/AN2	P72/AN2	P72/AN2	NC
41	P73/AN3	P73/AN3	P73/AN3	NC
42	P74/AN4	P74/AN4	P74/AN4	NC
43	P75/AN5	P75/AN5	P75/AN5	NC
44	P76/AN6/DA0	P76/AN6/DA0	P76/AN6/DA0	NC
45	P77/AN7/DA1	P77/AN7/DA1	P77/AN7/DA1	NC

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Pin No.	Expar	ided Modes	Single-Chip Modes		
FP-100B TFP-100B	Mode 1	Mode 2 (EXPE = 1) Mode 3 (EXPE = 1)	Mode 2 (EXPE = 0) Mode 3 (EXPE = 0)	Flash Memory Writer Mode	
46	AVSS	AVSS	AVSS		
47	PA1/CIN9/KIN9	A17/PA1/CIN9/ KIN9	PA1/CIN9/KIN9	NC	
48	PA0/CIN8/KIN8	A16/PA0/CIN8/KIN8	PA0/CIN8/KIN8	NC	
49	P40/TMCI0/ TxD2/IrTxD	P40/TMCI0/ TxD2/IrTxD	P40/TMCI0/ TxD2/IrTxD	NC	
50	P41/TMO0/ RxD2/IrRxD	P41/TMO0/ RxD2/IrRxD	P41/TMO0/ RxD2/IrRxD	NC	
51	P42/TMRI0/ SCK2/SDA1	P42/TMRI0/ SCK2/SDA1	P42/TMRI0/ SCK2/SDA1	NC	
52	P43/TMCI1	P43/TMCI1	P43/TMCI1/HIRQ11	NC	
53	P44/TMO1	P44/TMO1	P44/TMO1/HIRQ1	NC	
54	P45/TMRI1	P45/TMRI1	P45/TMRI1/HIRQ12	NC	
55	P46/PWX0	P46/PWX0	P46/PWX0	NC	
56	P47/PWX1	P47/PWX1	P47/PWX1	NC	
57	PB7/D7	PB7/D7	PB7	NC	
58	PB6/D6	PB6/D6	PB6	NC	
59	VCC1	VCC1	VCC1	VCC	
60	A15	A15/P27/PW15	P27/PW15	CE	
61	A14	A14/P26/PW14	P26/PW14	FA14	
62	A13	A13/P25/PW13	P25/PW13	FA13	
63	A12	A12/P24/PW12	P24/PW12	FA12	
64	A11	A11/P23/PW11	P23/PW11	FA11	
65	A10	A10/P22/PW10	P22/PW10	FA10	
66	A9	A9/P21/PW9	P21/PW9	ŌĒ	
67	A8	A8/P20/PW8	P20/PW8	FA8	
68	PB5/D5	PB5/D5	PB5	NC	
69	PB4/D4	PB4/D4	PB4	NC	
70	VSS	VSS	VSS	VSS	
71	VSS	VSS	VSS	VSS	





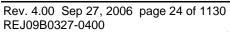
Pin No.	Expan	ded Modes	Single-Chip Modes		
FP-100B TFP-100B	Mode 1	Mode 2 (EXPE = 1) Mode 3 (EXPE = 1)	Mode 2 (EXPE = 0) Mode 3 (EXPE = 0)	Flash Memory Writer Mode	
72	A7	A7/P17/PW7	P17/PW7	FA7	
73	A6	A6/P16/PW6	P16/PW6	FA6	
74	A5	A5/P15/PW5	P15/PW5	FA5	
75	A4	A4/P14/PW4	P14/PW4	FA4	
76	A3	A3/P13/PW3	P13/PW3	FA3	
77	A2	A2/P12/PW2	P12/PW2	FA2	
78	A1	A1/P11/PW1	P11/PW1	FA1	
79	A0	A0/P10/PW0	P10/PW0	FA0	
80	PB3/D3	PB3/D3	PB3/CS4	NC	
81	PB2/D2	PB2/D2	PB2/CS3	NC	
82	D8	D8	P30/HDB0	FO0	
83	D9	D9	P31/HDB1	FO1	
84	D10	D10	P32/HDB2	FO2	
85	D11	D11	P33/HDB3	FO3	
86	D12	D12	P34/HDB4	FO4	
87	D13	D13	P35/HDB5	FO5	
88	D14	D14	P36/HDB6	FO6	
89	D15	D15	P37/HDB7	FO7	
90	PB1/D1	PB1/D1	PB1/HIRQ4	NC	
91	PB0/D0	PB0/D0	PB0/HIRQ3	NC	
92	VSS	VSS	VSS	VSS	
93	P80	P80	P80/HA0	NC	
94	P81	P81	P81/CS2/GA20	NC	
95	P82	P82	P82/HIFSD	NC	
96	P83	P83	P83	NC	
97	P84/IRQ3/TxD1	P84/IRQ3/TxD1	P84/IRQ3/TxD1	NC	
98	P85/IRQ4/RxD1	P85/IRQ4/RxD1	P85/IRQ4/RxD1	NC	
99	P86/IRQ5/SCK1/ SCL1	P86/IRQ5/SCK1/ SCL1	P86/IRQ5/SCK1/ SCL1	NC	
100	RESO	RESO	RESO	NC	

Table 1.2 (c) H8S/2144 Group Pin Functions in Each Operating Mode

Pin No.	Expar	ided Modes	Single-Chip Modes		
FP-100B TFP-100B Mode 1		Mode 2 (EXPE = 1) Mode 3 (EXPE = 1)	Mode 2 (EXPE = 0) Mode 3 (EXPE = 0)	Flash Memory Writer Mode	
1 RES		RES	RES	RES	
2	XTAL	XTAL	XTAL	XTAL	
3	EXTAL	EXTAL	EXTAL	EXTAL	
4	VCC1	VCC1	VCC1	VCC	
5	MD1	MD1	MD1	VSS	
6	MD0	MD0	MD0	VSS	
7	NMI	NMI	NMI	FA9	
8	STBY	STBY	STBY	VCC	
9	VCC2 (VCL)	VCC2 (VCL)	VCC2 (VCL)	VCC	
10	PA7/CIN15/ KIN15	A23/PA7/CIN15/ KIN15	PA7/CIN15/ KIN15	NC	
11	PA6/CIN14/ KIN14	A22/PA6/CIN14/ KIN14	PA6/CIN14/ KIN14	NC	
12	P52/SCK0	P52/SCK0	P52/SCK0	NC	
13	P51/RxD0	P51/RxD0	P51/RxD0	FA17	
14	P50/TxD0	P50/TxD0	P50/TxD0	NC	
15	VSS	VSS	VSS	VSS	
16	P97/WAIT	P97/WAIT	P97	VCC	
17	φ/P96/EXCL	φ/P96/EXCL	P96/φ/EXCL	NC	
18	AS/IOS	AS/IOS	P95	FA16	
19	HWR	HWR	P94	FA15	
20	PA5/CIN13/ KIN13	A21/PA5/CIN13/ KIN13	PA5/CIN13/ KIN13	NC	
21	PA4/CIN12/ KIN12	A20/PA4/CIN12/ KIN12	PA4/CIN12/KIN12	NC	
22	RD	RD	P93	WE	
23	P92/IRQ0	P92/IRQ0	P92/IRQ0	VSS	
24	P91/IRQ1	P91/IRQ1	P91/IRQ1	VCC	

Pin No.	Expan	ded Modes	Single-Chip Modes		
FP-100B TFP-100B	Mode 1	Mode 2 (EXPE = 1) Mode 3 (EXPE = 1)	Mode 2 (EXPE = 0) Mode 3 (EXPE = 0)	Flash Memory Writer Mode	
25	LWR/P90/IRQ2/ ADTRG	LWR/P90/IRQ2/ ADTRG	P90/IRQ2/ADTRG		
26	P60/FTCI/CIN0/ KIN0	P60/FTCI/CIN0/ KIN0	P60/FTCI/CIN0/ KIN0	NC	
27	P61/FTOA/ CIN1/KIN1	P61/FTOA/CIN1/ KIN1	P61/FTOA/CIN1/ KIN1	NC	
28	P62/FTIA/CIN2/ KIN2/TMIY	P62/FTIA/CIN2/ KIN2/TMIY	P62/FTIA/CIN2/ KIN2/TMIY	NC	
29	P63/FTIB/CIN3/ KIN3	P63/FTIB/CIN3/ KIN3	P63/FTIB/CIN3/ KIN3	NC	
30	PA3/CIN11/ KIN11	A19/PA3/CIN11/ KIN11	PA3/CIN11/ KIN11	NC	
31	PA2/CIN10/ KIN10	A18/PA2/CIN10/ KIN10	PA2/CIN10/ KIN10	NC	
32	P64/FTIC/CIN4/ KIN4	P64/FTIC/CIN4/ KIN4	P64/FTIC/CIN4/ KIN4	NC	
33	P65/FTID/CIN5/ KIN5	P65/FTID/CIN5/ KIN5	P65/FTID/CIN5/ KIN5	NC	
34	P66/FTOB/CIN6/ KIN6/IRQ6	P66/FTOB/CIN6/ KIN6/IRQ6	P66/FTOB/CIN6/ KIN6/IRQ6	NC	
35	P67/CIN7/KIN7/ IRQ7	P67/CIN7/KIN7/ IRQ7	P67/CIN7/KIN7/ IRQ7	VSS	
36	AVref	AVref	AVref	VCC	
37	AVCC	AVCC	AVCC	VCC	
38	P70/AN0	P70/AN0	P70/AN0	NC	
39	P71/AN1	P71/AN1	P71/AN1	NC	
40	P72/AN2	P72/AN2	P72/AN2	NC	
41	P73/AN3	P73/AN3	P73/AN3	NC	
42	P74/AN4	P74/AN4	P74/AN4	NC	
43	P75/AN5	P75/AN5	P75/AN5	NC	
44	P76/AN6/DA0	P76/AN6/DA0	P76/AN6/DA0	NC	
45	P77/AN7/DA1	P77/AN7/DA1	P77/AN7/DA1	NC	

Pin No.	Expan	ided Modes	Single-Chip Modes		
FP-100B TFP-100B	Mode 1	Mode 2 (EXPE = 1) Mode 3 (EXPE = 1)	Mode 2 (EXPE = 0) Mode 3 (EXPE = 0)	Flash Memory Writer Mode VSS	
46	AVSS	AVSS	AVSS		
47	PA1/CIN9/KIN9	A17/PA1/CIN9/KIN9	PA1/CIN9/KIN9	NC	
48	PA0/CIN8/KIN8	A16/PA0/CIN8/KIN8	PA0/CIN8/KIN8	NC	
49	P40/TMCI0/ TxD2/IrTxD	P40/TMCI0/ TxD2/IrTxD	P40/TMCI0/ TxD2/IrTxD	NC	
50	P41/TMO0/ RxD2/IrRxD	P41/TMO0/ RxD2/IrRxD	P41/TMO0/ RxD2/IrRxD	NC	
51	P42/TMRI0/ SCK2	P42/TMRI0/ SCK2	P42/TMRI0/ SCK2	NC	
52	P43/TMCI1	P43/TMCI1	P43/TMCI1	NC	
53	P44/TMO1	P44/TMO1	P44/TMO1	NC	
54	P45/TMRI1	P45/TMRI1	P45/TMRI1	NC	
55	P46/PWX0	P46/PWX0	P46/PWX0	NC	
56	P47/PWX1	P47/PWX1	P47/PWX1	NC	
57	PB7/D7	PB7/D7	PB7	NC	
58	PB6/D6	PB6/D6	PB6	NC	
59	VCC1	VCC1	VCC1	VCC	
60	A15	A15/P27	P27	CE	
61	A14	A14/P26	P26	FA14	
62	A13	A13/P25	P25	FA13	
63	A12	A12/P24	P24	FA12	
64	A11	A11/P23	P23	FA11	
65	A10	A10/P22	P22	FA10	
66	A9	A9/P21	P21	ŌĒ	
67	A8	A8/P20	P20	FA8	
68	PB5/D5	PB5/D5	PB5	NC	
69	PB4/D4	PB4/D4	PB4	NC	
70	VSS	VSS	VSS	VSS	
71	VSS	VSS	VSS	VSS	





Pin No.	Expan	ded Modes	Single-Chip Modes		
FP-100B TFP-100B	Mode 1	Mode 2 (EXPE = 1) Mode 3 (EXPE = 1)	Mode 2 (EXPE = 0) Mode 3 (EXPE = 0)	Flash Memory Writer Mode	
72	A7	A7/P17	P17	FA7	
73	A6	A6/P16	P16	FA6	
74	A5	A5/P15	P15	FA5	
75	A4	A4/P14	P14	FA4	
76	A3	A3/P13	P13	FA3	
77	A2	A2/P12	P12	FA2	
78	A1	A1/P11	P11	FA1	
79	A0	A0/P10	P10	FA0	
80	PB3/D3	PB3/D3	PB3	NC	
81	PB2/D2	PB2/D2	PB2	NC	
82	D8	D8	P30	FO0	
83	D9	D9	P31	FO1	
84	D10	D10	P32	FO2	
85	D11	D11	P33	FO3	
86	D12	D12	P34	FO4	
87	D13	D13	P35	FO5	
88	D14	D14	P36	FO6	
89	D15	D15	P37	FO7	
90	PB1/D1	PB1/D1	PB1	NC	
91	PB0/D0	PB0/D0	PB0	NC	
92	VSS	VSS	VSS	VSS	
93	P80	P80	P80	NC	
94	P81	P81	P81	NC	
95	P82	P82	P82	NC	
96	P83	P83	P83	NC	
97	P84/IRQ3/TxD1	P84/IRQ3/TxD1	P84/IRQ3/TxD1	NC	
98	P85/IRQ4/RxD1	P85/ IRQ4 /RxD1	P85/IRQ4/RxD1	NC	
99	P86/IRQ5/SCK1	P86/IRQ5/SCK1	P86/IRQ5/SCK1	NC	
100	RESO	RESO	RESO	NC	

1.3.3 Pin Functions

Table 1.3 summarizes the pin functions of this LSI.

Table 1.3 Pin Functions

Туре	Symbol	Pin No. FP-100B TFP-100B	_ I/O	Name and Function
Power supply	VCC1	4 [H8S/2144 Group only], 59 9*	Input	Power supply: For connection to the power supply. All VCC1 and VCC2* pins should be connected to the system power supply.
	VCL	9*	Input	Internal step-down voltage pin: A power supply pin for the product, applicable to product lines that have an internal step-down voltage. In the 5-V and 4-V versions, connect external capacitors to stabilize the internal step-down voltage between this pin and the VSS pin. Do not connect it to Vcc. In the 3-V version, connect this pin and the VCC1 pin to the power supply for the system. For details, See section 26, Electrical Characteristics.
	VCCB	4 [H8S/2148 Group and H8S/2147N only]	Input	Input/output buffer power supply: Power supply pin for the port A input/output buffer.
	VSS	15, 70 71, 92	Input	Ground: All VSS pins should be connected to the system power supply (0 V).
Clock	XTAL	2	Input	Connected to a crystal oscillator. See section 24, Clock Pulse Generator, for typical connection diagrams for a crystal oscillator and external clock input.
	EXTAL	3	Input	Connected to a crystal oscillator. The EXTAL pin can also input an external clock. See section 24, Clock Pulse Generator, for typical connection diagrams for a crystal oscillator and external clock input.
	ф	17	Output	System clock: Supplies the system clock to external devices.
	EXCL	17	Input	External subclock input: Input a 32.768 kHz external subclock.

Туре	Symbol	Pin No. FP-100B TFP-100B	 I/O	Nam	e and F	unction		
Operating mode control	MD1 MD0	5 6	Input	Mode pins: These pins set the operating mode. The relation between the settings of pins MD1 and MD0 and the operating mode is shown below. These pins should not be changed while the MCU is operating.				
				MD1	MD0	Operating Mode	Description	
				0	1	Mode 1	Normal	
							Expanded mode with on-chip ROM disabled	
				1	0	Mode 2	Advanced	
							Expanded mode with on-chip ROM enabled or single-chip mode	
				1	1	Mode 3	Normal	
							Expanded mode with on-chip ROM enabled or single-chip mode	
System control	RES	1	Input	Rese	-	: When this լ	pin is driven low, the chip is	
	RESO	100	Output	Reset output: Outputs reset signal to external device.				
	STBY	8	Input	Standby: When this pin is driven low, a transition is made to hardware standby mode.				
Address bus	A23 to A16	10, 11, 20, 21, 30, 31, 47, 48	Output	Address bus (advanced): Outputs address when 16-Mbyte space is used.				
	A15 to A0	60 to 67, 72 to 79	Output	Address bus: These pins output an address.				
Data bus	D15 to D8	89 to 82	Input/ output				ectional data bus. pper byte of 16-bit data.	
	D7 to D0	57, 58, 68, 69, 80, 81, 90, 91	Input/ output			ower): Bidire er byte of 16	ctional data bus. 6-bit data.	

		Pin No.		
Туре	Symbol	FP-100B TFP-100B	I/O	Name and Function
Bus control	WAIT	16	Input	Wait: Requests insertion of a wait state in the bus cycle when accessing external 3-state address space.
	RD	22	Output	Read: When this pin is low, it indicates that the external address space is being read.
	HWR	19	Output	High write: When this pin is low, it indicates that the external address space is being written to. The upper half of the data bus is valid.
	LWR	25	Output	Low write: When this pin is low, it indicates that the external address space is being written to. The lower half of the data bus is valid.
	AS/IOS	18	Output	Address strobe: When this pin is low, it indicates that address output on the address bus is valid.
Interrupt signals	NMI	7	Input	Nonmaskable interrupt: Requests a nonmaskable interrupt.
	IRQ0 to	23 to 25, 97 to 99, 34, 35	Input	Interrupt request 0 to 7: These pins request a maskable interrupt.
16-bit free- running	FTCI	26	Input	FRT counter clock input: Input pin for an external clock signal for the free-running counter (FRC).
timer (FRT)	FTOA	27	Output	FRT output compare A output: The output compare A output pin.
	FTOB	34	Output	FRT output compare B output: The output compare B output pin.
	FTIA	28	Input	FRT input capture A input: The input capture A input pin.
	FTIB	29	Input	FRT input capture B input: The input capture B input pin.
	FTIC	32	Input	FRT input capture C input: The input capture C input pin.
	FTID	33	Input	FRT input capture D input : The input capture D input pin.

Туре	Symbol	Pin No. FP-100B TFP-100B	_ I/O	Name and Function
8-bit timer (TMR0, TMR1,	TMO0 TMO1 TMOX	50 53 35	Output	Compare-match output: TMR0, TMR1, and TMRX compare-match output pins.
TMRX, TMRY)	TMCI0 TMCI1	49 52	Input	Counter external clock input: Input pins for the external clock input to the TMR0 and TMR1 counters.
	TMRI0 TMRI1	51 54	Input	Counter external reset input: TMR0 and TMR1 counter reset input pins.
	TMIX TMIY	26 28	Input	Counter external clock input and reset input: Dual function as TMRX and TMRY counter clock input pin and reset input pin.
PWM timer (PWM)	PW15 to PW0	60 to 67, 72 to 79	Output	PWM timer output: PWM timer pulse output pins.
14-bit PWM timer (PWMX)	PWX0 PWX1	55 56	Output	PWMX timer output: PWM D/A pulse output pins.
Serial com- munication interface	TxD0 TxD1 TxD2	14 97 49	Output	Transmit data: Data output pins.
(SCI0, SCI1, SCI2)	RxD0 RxD1 RxD2	13 98 50	Input	Receive data: Data input pins.
	SCK0	12	Input/	Serial clock: Clock input/output pins.
	SCK1 SCK2	99 51	output	The SCK0 output type is NMOS push-pull in the H8S/2148 Group and H8S/2147N, and is CMOS output in the H8S/2144 Group.
SCI with IrDA (SCI2)	IrTxD IrRxD	49 50	Output Input	IrDA transmit data/receive data: Input and output pins for data encoded for IrDA use.
Keyboard Buffer controller	PS2AC PS2BC PS2CC	31 21 11	Input/ output	PS2 clock: Keyboard buffer controller synchronization clock input/output pins.
(PS2)	PS2AD PS2BD PS2CD	30 20 10	Input/ output	PS2 data: Keyboard buffer controller data input/output pins.

Туре	Symbol	Pin No. FP-100B TFP-100B	- I/O	Name and Function					
Host interface	HDB7 to HDB0	89 to 82	Input/ output	Host interface data bus: Bidirectional 8-bit bus for accessing the host interface.					
(HIF)			Chip select 1, 2, 3, and 4: Input pins for selecting host interface channel 1 to 4.						
	ĪOR	22	Input	I/O read: Input pin that enables reading from the host interface.					
	ĪOW	19	Input	I/O write: Input pin that enables writing to the host interface.					
	HA0	93	Input	Command/data: Input pin that indicates whether an access is a data access or command access.					
	GA20	94	Output	GATE A20: A20 gate control signal output pin.					
	HIRQ11 HIRQ1 HIRQ12 HIRQ3 HIRQ4	52 53 54 91	Output	Host interrupt 11, 1, 12, 3, and 4: Output pins for interrupt requests to the host.					
	HIFSD	95	Input	Host interface shutdown: Control input pin used to place host interface input/output pins in the high-impedance/cutoff state.					
Keyboard control	KINO to KIN15	26 to 29, 32 to 35, 48, 47, 31, 30, 21, 20, 11, 10	Input	Keyboard input: Matrix keyboard input pins. Normally, P10 to P17 and P20 to P27 are used as key-scan outputs. This enables a maximum 16-output × 16-input, 256-key matrix to be configured.					
A/D converter	AN7 to AN0	45 to 38	Input	Analog input: A/D converter analog input pins.					
(ADC)	CIN0 to CIN15	26 to 29, 32 to 35, 48, 47, 31, 30, 21, 20, 11, 10	Input	Expansion A/D inputs: Expansion A/D input pins can be connected to the A/D converter, but since they are also used as digital input/output pins, precision will fall.					
	ADTRG	25	Input	A/D conversion external trigger input: Pin for input of an external trigger to start A/D conversion.					

Туре	Symbol	Pin No. FP-100B TFP-100B	_ I/O	Name and Function
D/A converter (DAC)	DA0 DA1	44 45	Output	Analog output: D/A converter analog output pins.
A/D converter	AVCC	37	Input	Analog reference voltage: The analog power supply pin for the A/D converter and D/A converter.
D/A converter				When the A/D and D/A converters are not used, this pin should be connected to the system power supply (+5 V or +3 V).
	AVref	36	Input	Analog reference voltage: The reference power supply pin for the A/D converter and D/A converter.
				When the A/D and D/A converters are not used, this pin should be connected to the system power supply (+5 V or +3 V).
	AVSS	46	Input	Analog ground: The ground pin for the A/D converter and D/A converter. This pin should be connected to the system power supply (0 V).
Timer connection	VSYNCI, HSYNCI, CSYNCI, VFBACKI, HFBACKI	-	Input	Timer connection input: Timer connection synchronous signal input pins.
	VSYNCO, HSYNCO, CLAMPO, CBLANK	53	Output	Timer connection output: Timer connection synchronous signal output pins.
I ² C bus interface (IIC)	SCL0 SCL1	12 99	Input/ output	I ² C clock input/output (channels 0 and 1): I ² C clock I/O pins. These pins have a bus drive function.
(option)				The SCL0 output form is NMOS open-drain
	SDA0 SDA1	16 51	Input/ output	I ² C data input/output (channels 0 and 1): I ² C data I/O pins. These pins have a bus drive function.
				The SDA0 output form is NMOS open-drain.

		Pin No.	_	
Туре	Symbol	FP-100B TFP-100B	I/O	Name and Function
I/O ports	P17 to P10	72 to 79	Input/ output	Port 1: Eight input/output pins. The data direction of each pin can be selected in the port 1 data direction register (P1DDR). These pins have built-in MOS input pull-ups, and also have LED drive capability.
	P27 to P20	60 to 67	Input/ output	Port 2: Eight input/output pins. The data direction of each pin can be selected in the port 2 data direction register (P2DDR). These pins have built-in MOS input pull-ups, and also have LED drive capability.
	P37 to P30	89 to 82	Input/ output	Port 3: Eight input/output pins. The data direction of each pin can be selected in the port 3 data direction register (P3DDR). These pins have built-in MOS input pull-ups, and also have LED drive capability.
	P47 to P40	56 to 49	Input/ output	Port 4: Eight input/output pins. The data direction of each pin can be selected in the port 4 data direction register (P4DDR).
	P52 to P50	12 to 14	Input/ output	Port 5: Three input/output pins. The data direction of each pin can be selected in the port 5 data direction register (P5DDR). P52 is an NMOS pushpull output in the H8S/2148 Group and H8S/2147N, and is a CMOS output in the H8S/2144 Group.
	P67 to P60	35 to 32 29 to 26	Input/ output	Port 6: Eight input/output pins. The data direction of each pin can be selected in the port 6 data direction register (P6DDR). These pins have built-in MOS input pull-ups.
	P77 to P70	45 to 38	Input	Port 7: Eight input pins.
	P86 to P80	99 to 93	Input/ output	Port 8: Seven input/output pins. The data direction of each pin can be selected in the port 8 data direction register (P8DDR).
	P97 to P90	16 to 19 22 to 25	Input/ output	Port 9: Eight input/output pins. The data direction of each pin (except P96) can be selected in the port 9 data direction register (P9DDR). P97 is an NMOS push-pull output in the H8S/2148 Group and H8S/2147N, and is a CMOS output in the H8S/2144 Group.

		Pin No.		
Туре	Symbol	FP-100B TFP-100B	I/O	Name and Function
I/O ports	PA7 to PA0	10, 11, 20, 21, 30, 31, 47, 48	Input/ output	Port A: Eight input/output pins. The data direction of each pin can be selected in the port A data direction register (PADDR). These pins have built-in MOS input pull-ups. These are the VCCB drive pins. [H8S/2148 Group and H8S/2147N only]
	PB7 to PB0	57, 58, 68, 69, 80, 81, 90, 91	Input/ output	Port B: Eight input/output pins. The data direction of each pin can be selected in the port B data direction register (PBDDR). These pins have built-in MOS input pull-ups.

Note: * In F-ZTAT and mask ROM versions of HD64F2148A, HD64F2147A, HD64F2144A, HD6432148S, HD6432148SW, HD6432147S, HD6432147SW, HD6432144S, HD6432143S pin NO.9 is VCL pin and is not VCC pin.

Section 2 CPU

2.1 Overview

The H8S/2000 CPU is a high-speed central processing unit with an internal 32-bit architecture that is upward-compatible with the H8/300 and H8/300H CPUs. The H8S/2000 CPU has sixteen 16-bit general registers, can address a 16-Mbyte (architecturally 4-Gbyte) linear address space, and is ideal for realtime control.

2.1.1 Features

The H8S/2000 CPU has the following features.

- Upward-compatible with H8/300 and H8/300H CPUs
 - Can execute H8/300 and H8/300H object programs
- General-register architecture
 - Sixteen 16-bit general registers (also usable as sixteen 8-bit registers or eight 32-bit registers)
- Sixty-five basic instructions
 - 8/16/32-bit arithmetic and logic instructions
 - Multiply and divide instructions
 - Powerful bit-manipulation instructions
- Eight addressing modes
 - Register direct [Rn]
 - Register indirect [@ERn]
 - Register indirect with displacement [@(d:16,ERn) or @(d:32,ERn)]
 - Register indirect with post-increment or pre-decrement [@ERn+ or @-ERn]
 - Absolute address [@aa:8, @aa:16, @aa:24, or @aa:32]
 - Immediate [#xx:8, #xx:16, or #xx:32]
 - Program-counter relative [@(d:8,PC) or @(d:16,PC)]
 - Memory indirect [@@aa:8]
- 16-Mbyte address space
 - Program: 16 Mbytes
 - Data: 16 Mbytes (4 Gbytes architecturally)

High-speed operation

— All frequently-used instructions execute in one or two states

— Maximum clock rate: 20 MHz

— 8/16/32-bit register-register add/subtract: 50 ns

 -8×8 -bit register-register multiply: 600 ns

— 16 ÷ 8-bit register-register divide: 600 ns

— 16×16 -bit register-register multiply: 1000 ns

— 32 ÷ 16-bit register-register divide: 1000 ns

• Two CPU operating modes

- Normal mode

Advanced mode

Power-down state

- Transition to power-down state by SLEEP instruction
- CPU clock speed selection

2.1.2 Differences between H8S/2600 CPU and H8S/2000 CPU

The differences between the H8S/2600 CPU and the H8S/2000 CPU are shown below.

• Register configuration

The MAC register is supported only by the H8S/2600 CPU.

• Basic instructions

The four instructions MAC, CLRMAC, LDMAC, and STMAC are supported only by the H8S/2600 CPU.

• Number of execution states

The number of execution states of the MULXU and MULXS instructions differ as follows.

Number of Execution States

Instruction	Mnemonic	H8S/2600	H8S/2000	
MULXU	MULXU.B Rs, Rd	3	12	
	MULXU.W Rs, ERd	4	20	
MULXS	MULXS.B Rs, Rd	4	13	
	MULXS.W Rs, ERd	5	21	

There are also differences in the address space, EXR register functions, power-down state, etc., depending on the product.



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2.1.3 Differences from H8/300 CPU

In comparison to the H8/300 CPU, the H8S/2000 CPU has the following enhancements.

- More general registers and control registers
 - Eight 16-bit extended registers, and one 8-bit control register, have been added.
- Expanded address space
 - Normal mode supports the same 64-kbyte address space as the H8/300 CPU.
 - Advanced mode supports a maximum 16-Mbyte address space.
- · Enhanced addressing
 - The addressing modes have been enhanced to make effective use of the 16-Mbyte address space.
- Enhanced instructions
 - Addressing modes of bit-manipulation instructions have been enhanced.
 - Signed multiply and divide instructions have been added.
 - Two-bit shift instructions have been added.
 - Instructions for saving and restoring multiple registers have been added.
 - A test and set instruction has been added.
- · Higher speed
 - Basic instructions execute twice as fast.

2.1.4 Differences from H8/300H CPU

In comparison to the H8/300H CPU, the H8S/2000 CPU has the following enhancements.

- Additional control register
 - One 8-bit control register has been added.
- Enhanced instructions
 - Addressing modes of bit-manipulation instructions have been enhanced.
 - Two-bit shift instructions have been added.
 - Instructions for saving and restoring multiple registers have been added.
 - A test and set instruction has been added.
- Higher speed
 - Basic instructions execute twice as fast.

2.2 **CPU Operating Modes**

The H8S/2000 CPU has two operating modes: normal and advanced. Normal mode supports a maximum 64-kbyte address space. Advanced mode supports a maximum 16-Mbyte total address space (architecturally the maximum total address space is 4 Gbytes, with a maximum of 16 Mbytes for the program area and a maximum of 4 Gbytes for the data area). The mode is selected by the mode pins of the microcontroller.

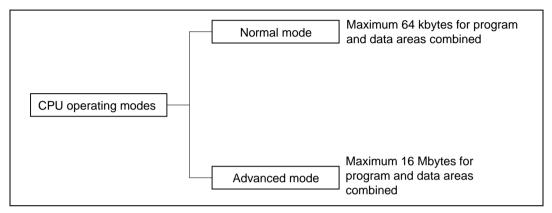


Figure 2.1 CPU Operating Modes

(1) Normal Mode

The exception vector table and stack have the same structure as in the H8/300 CPU.

Address Space: A maximum address space of 64 kbytes can be accessed.

Extended Registers (En): The extended registers (E0 to E7) can be used as 16-bit registers, or as the upper 16-bit segments of 32-bit registers. When En is used as a 16-bit register it can contain any value, even when the corresponding general register (Rn) is used as an address register. If the general register is referenced in the register indirect addressing mode with pre-decrement (@-Rn) or post-increment (@Rn+) and a carry or borrow occurs, however, the value in the corresponding extended register (En) will be affected.

Instruction Set: All instructions and addressing modes can be used. Only the lower 16 bits of effective addresses (EA) are valid.



Exception Vector Table and Memory Indirect Branch Addresses: In normal mode the top area starting at H'0000 is allocated to the exception vector table. One branch address is stored per 16 bits. The configuration of the exception vector table in normal mode is shown in figure 2.2. For details of the exception vector table, see section 4, Exception Handling.

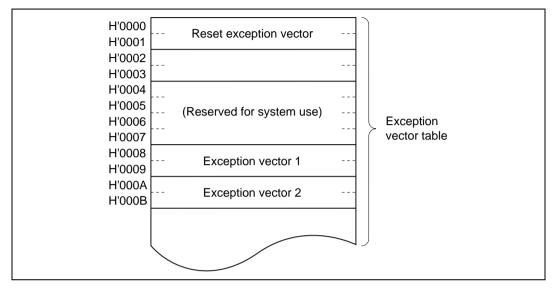


Figure 2.2 Exception Vector Table (Normal Mode)

The memory indirect addressing mode (@@aa:8) employed in the JMP and JSR instructions uses an 8-bit absolute address included in the instruction code to specify a memory operand that contains a branch address. In normal mode the operand is a 16-bit word operand, providing a 16-bit branch address. Branch addresses can be stored in the top area from H'0000 to H'00FF. Note that this area is also used for the exception vector table.

Stack Structure: When the program counter (PC) is pushed onto the stack in a subroutine call, and the PC and condition-code register (CCR) are pushed onto the stack in exception handling, they are stored as shown in figure 2.3. The extended control register (EXR) is not pushed onto the stack. For details, see section 4, Exception Handling.

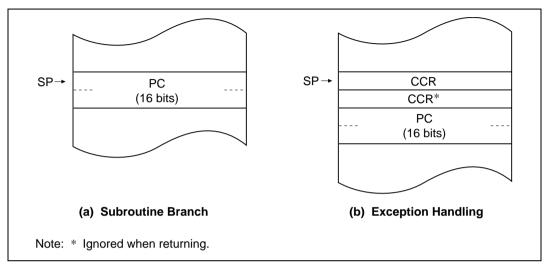


Figure 2.3 Stack Structure in Normal Mode

(2) Advanced Mode

Address Space: Linear access is provided to a 16-Mbyte maximum address space (architecturally a maximum 16-Mbyte program area and a maximum 4-Gbyte data area, with a maximum of 4 Gbytes for program and data areas combined).

Extended Registers (En): The extended registers (E0 to E7) can be used as 16-bit registers, or as the upper 16-bit segments of 32-bit registers or address registers.

Instruction Set: All instructions and addressing modes can be used.

Exception Vector Table and Memory Indirect Branch Addresses: In advanced mode the top area starting at H'00000000 is allocated to the exception vector table in units of 32 bits. In each 32 bits, the upper 8 bits are ignored and a branch address is stored in the lower 24 bits (figure 2.4). For details of the exception vector table, see section 4, Exception Handling.

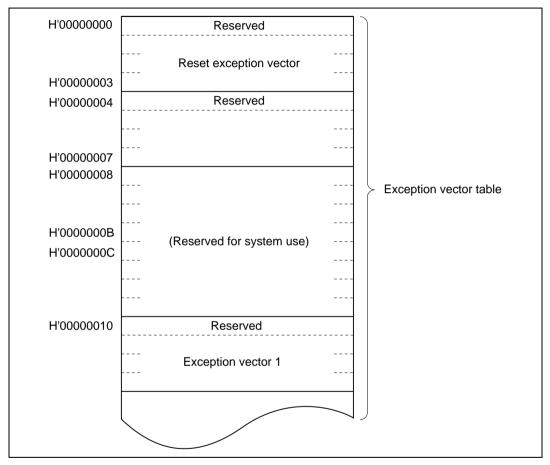


Figure 2.4 Exception Vector Table (Advanced Mode)

The memory indirect addressing mode (@@aa:8) employed in the JMP and JSR instructions uses an 8-bit absolute address included in the instruction code to specify a memory operand that contains a branch address. In advanced mode the operand is a 32-bit longword operand, providing a 32-bit branch address. The upper 8 bits of these 32 bits are a reserved area that is regarded as H'00. Branch addresses can be stored in the area from H'000000000 to H'000000FF. Note that the first part of this range is also the exception vector table.

Stack Structure: In advanced mode, when the program counter (PC) is pushed onto the stack in a subroutine call, and the PC and condition-code register (CCR) are pushed onto the stack in exception handling, they are stored as shown in figure 2.5. The extended control register (EXR) is not pushed onto the stack. For details, see section 4, Exception Handling.

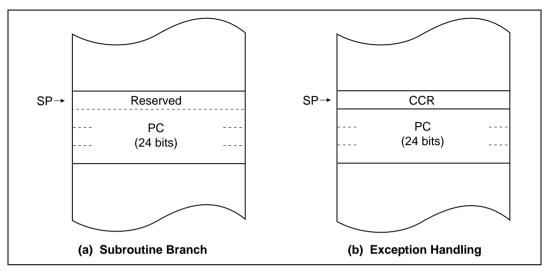


Figure 2.5 Stack Structure in Advanced Mode

2.3 Address Space

Figure 2.6 shows a memory map of the H8S/2000 CPU. The H8S/2000 CPU provides linear access to a maximum 64-kbyte address space in normal mode, and a maximum 16-Mbyte (architecturally 4-Gbyte) address space in advanced mode.

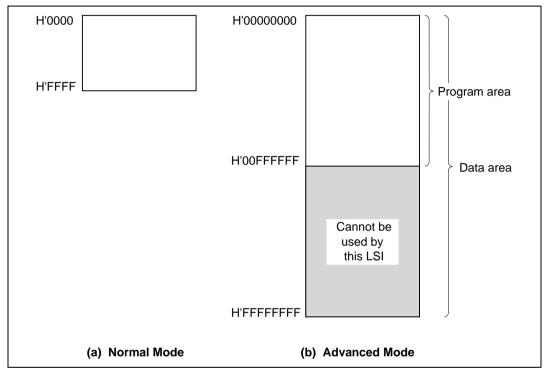


Figure 2.6 Memory Map

2.4 Register Configuration

2.4.1 Overview

The CPU has the internal registers shown in figure 2.7. There are two types of registers: general registers and control registers.

15		0 7	0	7 0
ER0	E0		R0H	R0L
ER1	E1		R1H	R1L
ER2	E2		R2H	R2L
ER3	E3		R3H	R3L
ER4	E4		R4H	R4L
ER5	E5		R5H	R5L
ER6	E6		R6H	R6L
ER7 (SP)	E7		R7H	R7L
				7 6 5 4 3 2 1 0
				CCR I UIHUNZVC
Legend:				

Figure 2.7 CPU Registers

2.4.2 General Registers

The CPU has eight 32-bit general registers. These general registers are all functionally alike and can be used as both address registers and data registers. When a general register is used as a data register, it can be accessed as a 32-bit, 16-bit, or 8-bit register. When the general registers are used as 32-bit registers or address registers, they are designated by the letters ER (ER0 to ER7).

The ER registers divide into 16-bit general registers designated by the letters E (E0 to E7) and R (R0 to R7). These registers are functionally equivalent, providing a maximum of sixteen 16-bit registers. The E registers (E0 to E7) are also referred to as extended registers.

The R registers divide into 8-bit general registers designated by the letters RH (R0H to R7H) and RL (R0L to R7L). These registers are functionally equivalent, providing a maximum of sixteen 8-bit registers.

Figure 2.8 illustrates the usage of the general registers. The usage of each register can be selected independently.

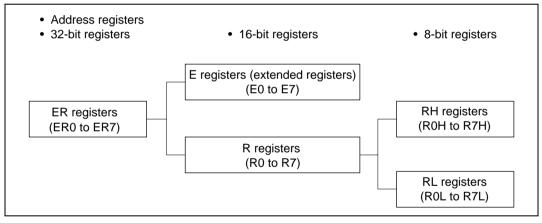


Figure 2.8 Usage of General Registers

General register ER7 has the function of stack pointer (SP) in addition to its general-register function, and is used implicitly in exception handling and subroutine calls. Figure 2.9 shows the stack.

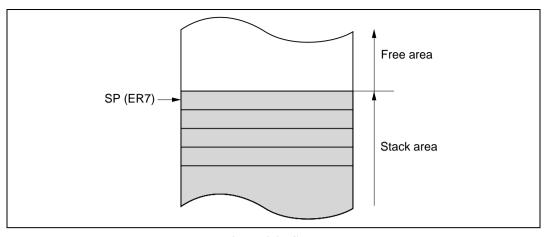


Figure 2.9 Stack

2.4.3 Control Registers

The control registers are the 24-bit program counter (PC), 8-bit extended control register (EXR), and 8-bit condition-code register (CCR).

(1) Program Counter (PC)

This 24-bit counter indicates the address of the next instruction the CPU will execute. The length of all CPU instructions is 2 bytes (one word), so the least significant PC bit is ignored. (When an instruction is fetched, the least significant PC bit is regarded as 0.)

(2) Extended Control Register (EXR)

An 8-bit register. In this LSI, this register does not affect operation.

Bit 7—Trace Bit (T): This bit is reserved. In this LSI, this bit does not affect operation.

Bits 6 to 3—Reserved: These bits are reserved. They are always read as 1.

Bits 2 to 0—Interrupt Mask Bits (I2 to I0): These bits are reserved. In this LSI, these bits do not affect operation.

(3) Condition-Code Register (CCR)

This 8-bit register contains internal CPU status information, including an interrupt mask bit (I) and half-carry (H), negative (N), zero (Z), overflow (V), and carry (C) flags.

Bit 7—Interrupt Mask Bit (I): Masks interrupts other than NMI when set to 1. (NMI is accepted regardless of the I bit setting.) The I bit is set to 1 by hardware at the start of an exception-handling sequence. For details, refer to section 5, Interrupt Controller.

Bit 6—User Bit or Interrupt Mask Bit (UI): Can be written and read by software using the LDC, STC, ANDC, ORC, and XORC instructions. This bit can also be used as an interrupt mask bit. For details, refer to section 5, Interrupt Controller.

Bit 5—Half-Carry Flag (H): When the ADD.B, ADDX.B, SUB.B, SUBX.B, CMP.B, or NEG.B instruction is executed, this flag is set to 1 if there is a carry or borrow at bit 3, and cleared to 0 otherwise. When the ADD.W, SUB.W, CMP.W, or NEG.W instruction is executed, the H flag is set to 1 if there is a carry or borrow at bit 11, and cleared to 0 otherwise. When the ADD.L, SUB.L, CMP.L, or NEG.L instruction is executed, the H flag is set to 1 if there is a carry or borrow at bit 27, and cleared to 0 otherwise.

Bit 4—User Bit (U): Can be written and read by software using the LDC, STC, ANDC, ORC, and XORC instructions.

Bit 3—Negative Flag (N): Stores the value of the most significant bit (sign bit) of data.

Bit 2—Zero Flag (Z): Set to 1 to indicate zero data, and cleared to 0 to indicate non-zero data.

Bit 1—Overflow Flag (V): Set to 1 when an arithmetic overflow occurs, and cleared to 0 otherwise.

Bit 0—Carry Flag (C): Set to 1 when a carry occurs, and cleared to 0 otherwise. Used by:

- Add instructions, to indicate a carry
- Subtract instructions, to indicate a borrow
- Shift and rotate instructions, to store the carry

The carry flag is also used as a bit accumulator by bit-manipulation instructions.

Some instructions leave some or all of the flag bits unchanged. For the action of each instruction on the flag bits, refer to appendix A.1, Instruction.

Operations can be performed on the CCR bits by the LDC, STC, ANDC, ORC, and XORC instructions. The N, Z, V, and C flags are used as branching conditions for conditional branch (Bcc) instructions.

2.4.4 Initial Register Values

Reset exception handling loads the CPU's program counter (PC) from the vector table, clears the trace bit in EXR to 0, and sets the interrupt mask bits in CCR and EXR to 1. The other CCR bits and the general registers are not initialized. In particular, the stack pointer (ER7) is not initialized. The stack pointer should therefore be initialized by an MOV.L instruction executed immediately after a reset.



2.5 Data Formats

The CPU can process 1-bit, 4-bit (BCD), 8-bit (byte), 16-bit (word), and 32-bit (longword) data. Bit-manipulation instructions operate on 1-bit data by accessing bit n (n = 0, 1, 2, ..., 7) of byte operand data. The DAA and DAS decimal-adjust instructions treat byte data as two digits of 4-bit BCD data.

2.5.1 General Register Data Formats

Figure 2.10 shows the data formats in general registers.

Data Type	General Register	Data Format
1-bit data	RnH	7 0 7 6 5 4 3 2 1 0 Don't care
1-bit data	RnL	7 0 Don't care 7 6 5 4 3 2 1 0
4-bit BCD data	RnH	7 4 3 0 Upper digit Lower digit Don't care
4-bit BCD data	RnL	7 4 3 0 Don't care Upper digit Lower digit
Byte data	RnH	7 0 Don't care MSB LSB
Byte data	RnL	7 0 Don't care MSB LSB

Figure 2.10 General Register Data Formats

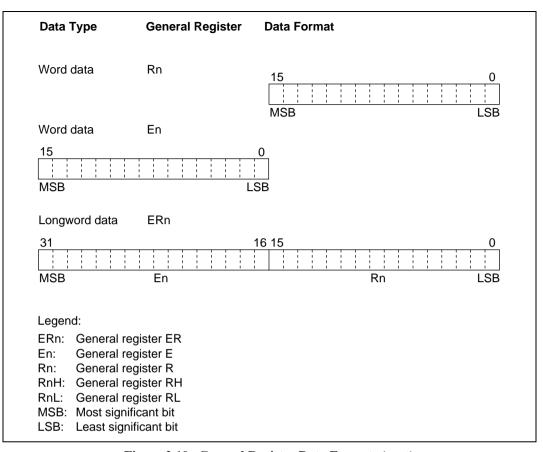


Figure 2.10 General Register Data Formats (cont)



2.5.2 Memory Data Formats

Figure 2.11 shows the data formats in memory. The CPU can access word data and longword data in memory, but word or longword data must begin at an even address. If an attempt is made to access word or longword data at an odd address, no address error occurs but the least significant bit of the address is regarded as 0, so the access starts at the preceding address. This also applies to instruction fetches.

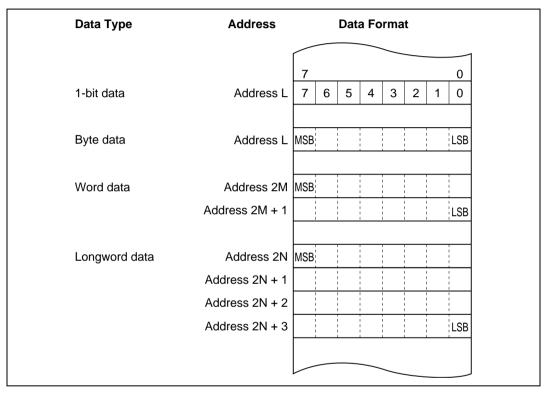


Figure 2.11 Memory Data Formats

When ER7 (SP) is used as an address register to access the stack, the operand size should be word size or longword size.

2.6 Instruction Set

2.6.1 Overview

The H8S/2000 CPU has 65 types of instructions. The instructions are classified by function in table 2.1.

Table 2.1 Instruction Classification

Function	Instructions	Size	Types
Data transfer	MOV	BWL	5
	POP*1, PUSH*1	WL	_
	LDM*5, STM*5	L	_
	MOVFPE*3, MOVTPE*3	В	_
Arithmetic	ADD, SUB, CMP, NEG	BWL	19
operations	ADDX, SUBX, DAA, DAS	В	_
	INC, DEC	BWL	_
	ADDS, SUBS	L	_
	MULXU, DIVXU, MULXS, DIVXS	BW	
	EXTU, EXTS	WL	
	TAS*4	В	_
Logic operations	AND, OR, XOR, NOT	BWL	4
Shift	SHAL, SHAR, SHLL, SHLR, ROTL, ROTR, ROTXL, ROTXR	BWL	8
Bit-manipulation	BSET, BCLR, BNOT, BTST, BLD, BILD, BST, BIST, BAND, BIAND, BOR, BIOR, BXOR, BIXOR	В	14
Branch	Bcc*2, JMP, BSR, JSR, RTS	_	5
System control	TRAPA, RTE, SLEEP, LDC, STC, ANDC, ORC, XORC, NOP	_	9
Block data transfer	EEPMOV	_	1
-			

Total: 65 types

Legend: B: Byte

W: Word L: Longword

Notes: 1. POP.W Rn and PUSH.W Rn are identical to MOV.W @SP+, Rn and MOV.W Rn, @-SP.

POP.L ERn and PUSH.L ERn are identical to MOV.L @SP+, ERn and MOV.L ERn, @-SP.

- 2. Bcc is the general name for conditional branch instructions.
- 3. Cannot be used in this LSI.
- 4. Only register ER0, ER1, ER4, or ER5 should be used when using the TAS instruction.
- 5. Only registers ER0 to ER6 should be used when using the STM/LDM instruction.

2.6.2 Instructions and Addressing Modes

Table 2.2 indicates the combinations of instructions and addressing modes that the H8S/2000 CPU can use.

Table 2.2 Combinations of Instructions and Addressing Modes

							Ad	dressi	ng Mod	les					
Function	Instruction	***	Rn	@ERn	@(d:16,ERn)	@(d:32, ERn)	@-ERn/@ERn+	@aa:8	@aa:16	@aa:24	@aa:32	@(d:8, PC)	@(d:16, PC)	@ @ aa:8	I
Data	MOV	BWL	BWL	BWL	BWL	BWL	BWL	В	BWL	_	BWL	_	_	_	_
transfer	POP, PUSH	_	_	_	_	_	_	_	_	_	_	_	_	_	WL
	LDM*3, STM*3	_	_	_	_	_	_	-	_	-	_	_	_	_	L
	MOVFPE*1, MOVTPE*1	_	_	_	_	_	_		В		_	_	_	_	_
Arithmetic	ADD, CMP	BWL	BWL	_	_	_	_	-	_	-	_	_	_	_	_
operations	SUB	WL	BWL	_	_	_	_	_	_	_	_	_	_	_	_
	ADDX, SUBX	В	В	_	_	_	_	-	_	_	_	_	_	_	_
	ADDS, SUBS	_	L	_	_	_	_	-	_	-	_	_	_	_	_
	INC, DEC	_	BWL	_	_	_	_	-	_	-	_	_	_	_	_
	DAA, DAS	_	В	_	_	_	_	_	_	_	_	_	_	_	_
	MULXU, DIVXU	_	BW	_	_	_	_		_		_	_	_	_	_
	MULXS, DIVXS	_	BW	_	_	_	_	-	_	-	_	_	_	_	_
	NEG	_	BWL	_	_	_	_	_	_	_	_	_	_	_	_
	EXTU, EXTS	_	WL	_	_	_	_	-	_	-	_	_	_	_	_
	TAS*2	_	_	В	_	_	_	_	_	_	_	_	_	_	_
Logic operations	AND, OR, XOR	BWL	BWL	_	_	_	_		_	-	_	_	_	_	_
	NOT	_	BWL	_	_	_	_	_	_	_	_	_	_	_	_
Shift	1	_	BWL	_	_	_	_	_	_	_	_	_	_	_	_
Bit-manipula	ation	_	В	В	_	_	_	В	В	_	В	_	_	_	_
Branch	Bcc, BSR	_	_	_	_	_	_	_	_	_	_	0	0	_	_
	JMP, JSR	_	_	_	_	_	_	_	_	0	_	_	_	0	_
	RTS	_	_	_	_	_	_	_	_	_	_	_	_	_	0

							Ad	dressi	ng Mod	des					
Function	Instruction	xx#	Rn	@ERn	@(d:16,ERn)	@(d:32, ERn)	@-ERn/@ERn+	@aa:8	@aa:16	@aa:24	@aa:32	@(d:8, PC)	@(d:16, PC)	@ @aa:8	ı
System	TRAPA	_	_	-	_	_	_	_	_	_	_	_	_	_	0
control	RTE	_	_	_	_	_	_	_	_	_	_	_	_	_	0
	SLEEP	_	_	_	_	_	_	_	_	_	_	_	_	_	0
	LDC	В	В	W	W	W	W	_	W	_	W	_	_	_	_
	STC	_	В	W	W	W	W	_	W	_	W	_	_	_	_
	ANDC, ORC, XORC	В	_	-	_	_	_	_	_	_	_	_	_	_	_
	NOP	_	_	_	_	_	_	_	_	_	_	_	_	_	0
Block data	transfer	_	_	_	_	_	_	_	_	_	_	_	_	_	BW
Legend:		1			1	1		ı	1	1	ı		1	1	

B: Byte W: Word

L: Longword

Notes: 1. Cannot be used in this LSI.

2. Only register ER0, ER1, ER4, or ER5 should be used when using the TAS instruction.

3. Only registers ER0 to ER6 should be used when using the STM/LDM instruction.

2.6.3 Table of Instructions Classified by Function

Table 2.3 summarizes the instructions in each functional category. The notation used in table 2.3 is defined below.

Operation Notation

-	
Rd	General register (destination)*
Rs	General register (source)*
Rn	General register*
ERn	General register (32-bit register)
(EAd)	Destination operand
(EAs)	Source operand
EXR	Extended control register
CCR	Condition-code register
N	N (negative) flag in CCR
Z	Z (zero) flag in CCR
V	V (overflow) flag in CCR
С	C (carry) flag in CCR
PC	Program counter
SP	Stack pointer
#IMM	Immediate data
disp	Displacement
+	Addition
_	Subtraction
×	Multiplication
÷	Division
^	Logical AND
<u> </u>	Logical OR
\oplus	Logical exclusive OR
\rightarrow	Move
7	NOT (logical complement)
:8/:16/:24/:32	8-, 16-, 24-, or 32-bit length

Note: * General registers include 8-bit registers (R0H to R7H, R0L to R7L), 16-bit registers (R0 to R7, E0 to E7), and 32-bit registers (ER0 to ER7).

Table 2.3 Instructions Classified by Function

Туре	Instruction	Size*1	Function
Data transfer	MOV	B/W/L	(EAs) → Rd, Rs → (EAd) Moves data between two general registers or between a general register and memory, or moves immediate data to a general register.
	MOVFPE	В	Cannot be used in this LSI.
	MOVTPE	В	Cannot be used in this LSI.
	POP	W/L	$@SP+ \rightarrow Rn$ Pops a general register from the stack.
			POP.W Rn is identical to MOV.W @SP+, Rn.
			POP.L ERn is identical to MOV.L @SP+, ERn.
	PUSH	W/L	$Rn \rightarrow @-SP$ Pushes a general register onto the stack.
			PUSH.W Rn is identical to MOV.W Rn, @-SP.
			PUSH.L ERn is identical to MOV.L ERn, @-SP.
	LDM*3	L	@SP+ → Rn (register list) Pops two or more general registers from the stack.
	STM*3	L	Rn (register list) → @−SP Pushes two or more general registers onto the stack.

Туре	Instruction	Size*1	Function
Arithmetic operations	ADD SUB	B/W/L	Rd \pm Rs \rightarrow Rd, Rd \pm #IMM \rightarrow Rd Performs addition or subtraction on data in two general registers, or on immediate data and data in a general register. (Immediate byte data cannot be subtracted from byte data in a general register. Use the SUBX or ADD instruction.)
	ADDX SUBX	В	Rd \pm Rs \pm C \rightarrow Rd, Rd \pm #IMM \pm C \rightarrow Rd Performs addition or subtraction with carry on byte data in two general registers, or on immediate data and data in a general register.
	INC DEC	B/W/L	Rd \pm 1 \rightarrow Rd, Rd \pm 2 \rightarrow Rd Increments or decrements a general register by 1 or 2. (Byte operands can be incremented or decremented by 1 only.)
	ADDS SUBS	L	Rd \pm 1 \rightarrow Rd, Rd \pm 2 \rightarrow Rd, Rd \pm 4 \rightarrow Rd Adds or subtracts the value 1, 2, or 4 to or from data in a 32-bit register.
	DAA DAS	В	Rd decimal adjust → Rd Decimal-adjusts an addition or subtraction result in a general register by referring to the CCR to produce 4-bit BCD data.
	MULXU	B/W	Rd \times Rs \rightarrow Rd Performs unsigned multiplication on data in two general registers: either 8 bits \times 8 bits \rightarrow 16 bits or 16 bits \times 16 bits \rightarrow 32 bits.
	MULXS	B/W	$Rd \times Rs \rightarrow Rd$ Performs signed multiplication on data in two general registers: either 8 bits \times 8 bits \rightarrow 16 bits or 16 bits \times 16 bits \rightarrow 32 bits.
	DIVXU	B/W	Rd \div Rs \rightarrow Rd Performs unsigned division on data in two general registers: either 16 bits \div 8 bits \rightarrow 8-bit quotient and 8-bit remainder or 32 bits \div 16 bits \rightarrow 16-bit quotient and 16- bit remainder.

Туре	Instruction	Size*1	Function
Arithmetic operations	DIVXS	B/W	Rd \div Rs \rightarrow Rd Performs signed division on data in two general registers: either 16 bits \div 8 bits \rightarrow 8-bit quotient and 8-bit remainder or 32 bits \div 16 bits \rightarrow 16-bit quotient and 16- bit remainder.
	СМР	B/W/L	Rd – Rs, Rd – #IMM Compares data in a general register with data in another general register or with immediate data, and sets CCR bits according to the result.
	NEG	B/W/L	$0-Rd \rightarrow Rd$ Takes the two's complement (arithmetic complement) of data in a general register.
	EXTU	W/L	Rd (zero extension) → Rd Extends the lower 8 bits of a 16-bit register to word size, or the lower 16 bits of a 32-bit register to longword size, by padding with zeros on the left.
	EXTS	W/L	Rd (sign extension) → Rd Extends the lower 8 bits of a 16-bit register to word size, or the lower 16 bits of a 32-bit register to longword size, by extending the sign bit.
	TAS	В	@ERd – 0, 1 \rightarrow (<bit 7=""> of @ERd)*2 Tests memory contents, and sets the most significant bit (bit 7) to 1.</bit>

Туре	Instruction	Size*1	Function
Logic operations	AND	B/W/L	$Rd \wedge Rs \rightarrow Rd$, $Rd \wedge \#IMM \rightarrow Rd$ Performs a logical AND operation on a general register and another general register or immediate data.
	OR	B/W/L	$Rd \lor Rs \to Rd$, $Rd \lor \#IMM \to Rd$ Performs a logical OR operation on a general register and another general register or immediate data.
	XOR	B/W/L	$Rd \oplus Rs \rightarrow Rd$, $Rd \oplus \#IMM \rightarrow Rd$ Performs a logical exclusive OR operation on a general register and another general register or immediate data.
	NOT	B/W/L	\neg (Rd) \rightarrow (Rd) Takes the one's complement (logical complement) of general register contents.
Shift operations	SHAL SHAR	B/W/L	Rd (shift) \rightarrow Rd Performs an arithmetic shift on general register contents. A 1-bit or 2-bit shift is possible.
	SHLL SHLR	B/W/L	Rd (shift) \rightarrow Rd Performs a logical shift on general register contents. A 1-bit or 2-bit shift is possible.
	ROTL ROTR	B/W/L	Rd (rotate) → Rd Rotates general register contents. 1-bit or 2-bit rotation is possible.
	ROTXL ROTXR	B/W/L	Rd (rotate) \rightarrow Rd Rotates general register contents through the carry flag. 1-bit or 2-bit rotation is possible.

Туре	Instruction	Size*1	Function
Bit-manipulation instructions	BSET	В	1 → (<bit-no.> of <ead>) Sets a specified bit in a general register or memory operand to 1. The bit number is specified by 3-bit immediate data or the lower three bits of a general register.</ead></bit-no.>
	BCLR	В	0 → (<bit>O → (Clears a specified bit in a general register or memory operand to 0. The bit number is specified by 3-bit immediate data or the lower three bits of a general register.</bit>
	BNOT	В	¬ (<bit-no.> of <ead>) → (<bit-no.> of <ead>) Inverts a specified bit in a general register or memory operand. The bit number is specified by 3-bit immediate data or the lower three bits of a general register.</ead></bit-no.></ead></bit-no.>
	BTST	В	¬ (<bit-no.> of <ead>) \rightarrow Z Tests a specified bit in a general register or memory operand and sets or clears the Z flag accordingly. The bit number is specified by 3-bit immediate data or the lower three bits of a general register.</ead></bit-no.>
	BAND	В	$C \wedge (\text{sbit-No.} > \text{of } < \text{EAd>}) \rightarrow C$ ANDs the carry flag with a specified bit in a general register or memory operand and stores the result in the carry flag.
	BIAND	В	$C \land \neg$ (<bit-no.> of <ead>) $\rightarrow C$ ANDs the carry flag with the inverse of a specified bit in a general register or memory operand and stores the result in the carry flag.</ead></bit-no.>
			The bit number is specified by 3-bit immediate data.
	BOR	В	$C \lor (\text{sbit-No.} > \text{of } < \text{EAd>}) \to C$ ORs the carry flag with a specified bit in a general register or memory operand and stores the result in the carry flag.
	BIOR	В	$C \lor \neg$ (<bit-no.> of <ead>) $\to C$ ORs the carry flag with the inverse of a specified bit in a general register or memory operand and stores the result in the carry flag.</ead></bit-no.>
			The bit number is specified by 3-bit immediate data.

Туре	Instruction	Size*1	Function
Bit- manipulation instructions	BXOR	В	$C \oplus (\text{-bit-No} \text{ of } \text{-EAd}) \to C$ Exclusive-ORs the carry flag with a specified bit in a general register or memory operand and stores the result in the carry flag.
	BIXOR	В	$C \oplus \neg$ (<bit-no.> of <ead>) $\rightarrow C$ Exclusive-ORs the carry flag with the inverse of a specified bit in a general register or memory operand and stores the result in the carry flag.</ead></bit-no.>
			The bit number is specified by 3-bit immediate data.
	BLD	В	(<bit-no.> of <ead>) \rightarrow C Transfers a specified bit in a general register or memory operand to the carry flag.</ead></bit-no.>
	BILD	В	\neg (<bit-no.> of <ead>) \rightarrow C Transfers the inverse of a specified bit in a general register or memory operand to the carry flag.</ead></bit-no.>
			The bit number is specified by 3-bit immediate data.
	BST	В	C → (<bit-no.> of <ead>) Transfers the carry flag value to a specified bit in a general register or memory operand.</ead></bit-no.>
	BIST	В	\neg C \rightarrow (<bit-no.> of <ead>) Transfers the inverse of the carry flag value to a specified bit in a general register or memory operand.</ead></bit-no.>
			The bit number is specified by 3-bit immediate data.

Туре	Instruction	Size*1	Function		
Branch instructions	Всс	_	Branches to a specified address if a specified condition is true. The branching conditions are listed below.		
			Mnemonic	Description	Condition
			BRA(BT)	Always (true)	Always
			BRN(BF)	Never (false)	Never
			BHI	High	C ∨ Z = 0
			BLS	Low or same	C ∨ Z = 1
			BCC(BHS)	Carry clear (high or same)	C = 0
			BCS(BLO)	Carry set (low)	C = 1
			BNE	Not equal	Z = 0
			BEQ	Equal	Z = 1
			BVC	Overflow clear	V = 0
			BVS	Overflow set	V = 1
			BPL	Plus	N = 0
			BMI	Minus	N = 1
			BGE	Greater or equal	N ⊕ V = 0
			BLT	Less than	N ⊕ V = 1
			BGT	Greater than	$Z\vee(N\oplus V)=0$
			BLE	Less or equal	Z∨(N ⊕ V) = 1
	JMP	_	Branches unconditionally to a specified address.		
	BSR	_	Branches to a subroutine at a specified address.		
	JSR	_	Branches to a	d address.	
	RTS	_	Returns from a subroutine		

Туре	Instruction	Size*1	Function
System	TRAPA	_	Starts trap-instruction exception handling.
control instructions	RTE	_	Returns from an exception-handling routine.
IIISHUCIONS	SLEEP	_	Causes a transition to a power-down state.
	LDC	B/W	(EAs) → CCR, (EAs) → EXR Moves contents of a general register or memory or immediate data to CCR or EXR. Although CCR and EXR are 8-bit registers, word-size transfers are performed between them and memory. The upper 8 bits are valid.
	STC	B/W	CCR → (EAd), EXR → (EAd) Transfers CCR or EXR contents to a general register or memory. Although CCR and EXR are 8-bit registers, word-size transfers are performed between them and memory. The upper 8 bits are valid.
	ANDC	В	CCR \land #IMM \rightarrow CCR, EXR \land #IMM \rightarrow EXR Logically ANDs the CCR or EXR contents with immediate data.
	ORC	В	CCR \vee #IMM \rightarrow CCR, EXR \vee #IMM \rightarrow EXR Logically ORs the CCR or EXR contents with immediate data.
	XORC	В	CCR \oplus #IMM \to CCR, EXR \oplus #IMM \to EXR Logically exclusive-ORs the CCR or EXR contents with immediate data.
	NOP	_	PC + 2 → PC Only increments the program counter.

Туре	Instruction	Size*1	Function
Block data transfer instructions	EEPMOV.B	_	if R4L \neq 0 then Repeat @ER5+ \rightarrow @ER6+ R4L-1 \rightarrow R4L Until R4L = 0 else next;
	EEPMOV.W	_	if R4 \neq 0 then Repeat @ER5+ \rightarrow @ER6+ R4-1 \rightarrow R4 Until R4 = 0 else next;
			Block transfer instruction. Transfers the number of data bytes specified by R4L or R4 from locations starting at the address indicated by ER5 to locations starting at the address indicated by ER6. After the transfer, the next instruction is executed.

Notes: 1. Size refers to the operand size.

B: Byte W: Word L: Longword

- 2. Only register ER0, ER1, ER4, or ER5 should be used when using the TAS instruction.
- 3. Only registers ER0 to ER6 should be used when using the STM/LDM instruction.

2.6.4 Basic Instruction Formats

The CPU instructions consist of 2-byte (1-word) units. An instruction consists of an operation field (op field), a register field (r field), an effective address extension (EA field), and a condition field (cc).

Operation Field: Indicates the function of the instruction, the addressing mode, and the operation to be carried out on the operand. The operation field always includes the first four bits of the instruction. Some instructions have two operation fields.

Register Field: Specifies a general register. Address registers are specified by 3 bits, data registers by 3 bits or 4 bits. Some instructions have two register fields. Some have no register field.

Effective Address Extension: Eight, 16, or 32 bits specifying immediate data, an absolute address, or a displacement.

Condition Field: Specifies the branching condition of Bcc instructions.

Figure 2.12 shows examples of instruction formats.

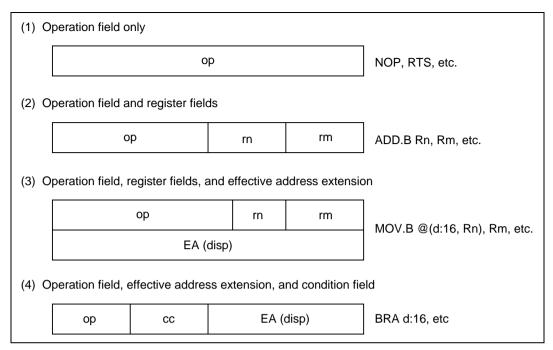


Figure 2.12 Instruction Formats (Examples)

2.6.5 Notes on Use of Bit-Manipulation Instructions

The BSET, BCLR, BNOT, BST, and BIST instructions read a byte of data, carry out bit-manipulation, then write back the byte of data. Caution is therefore required when using these instructions on a register containing write-only bits, or a port.

The BCLR instruction can be used to clear internal I/O register flags to 0. In this case, the relevant flag need not be read beforehand if it is clear that it has been set to 1 in an interrupt handling routine, etc.

2.7 Addressing Modes and Effective Address Calculation

2.7.1 Addressing Mode

The CPU supports the eight addressing modes listed in table 2.4. Each instruction uses a subset of these addressing modes. Arithmetic and logic instructions can use the register direct and immediate modes. Data transfer instructions can use all addressing modes except program-counter relative and memory indirect. Bit-manipulation instructions use register direct, register indirect, or

absolute addressing mode to specify an operand, and register direct (BSET, BCLR, BNOT, and BTST instructions) or immediate (3-bit) addressing mode to specify a bit number in the operand.

Table 2.4 Addressing Modes

No.	Addressing Mode	Symbol
1	Register direct	Rn
2	Register indirect	@ERn
3	Register indirect with displacement	@(d:16,ERn)/@(d:32,ERn)
4	Register indirect with post-increment Register indirect with pre-decrement	@ERn+ @-ERn
5	Absolute address	@aa:8/@aa:16/@aa:24/@aa:32
6	Immediate	#xx:8/#xx:16/#xx:32
7	Program-counter relative	@(d:8,PC)/@(d:16,PC)
8	Memory indirect	@ @aa:8

Register Direct—Rn

The register field of the instruction code specifies an 8-, 16-, or 32-bit general register containing the operand. R0H to R7H and R0L to R7L can be specified as 8-bit registers. R0 to R7 and E0 to E7 can be specified as 16-bit registers. ER0 to ER7 can be specified as 32-bit registers.

Register Indirect—@ERn

The register field of the instruction code specifies an address register (ERn) which contains the address of the operand in memory. If the address is a program instruction address, the lower 24 bits are valid and the upper 8 bits are all assumed to be 0 (H'00).

Register Indirect with Displacement—@(d:16, ERn) or @(d:32, ERn)

A 16-bit or 32-bit displacement contained in the instruction is added to an address register (ERn) specified by the register field of the instruction, and the sum gives the address of a memory operand. A 16-bit displacement is sign-extended when added.



Register Indirect with Post-Increment or Pre-Decrement—@ERn+ or @-ERn

- Register indirect with post-increment—@ERn+
 - The register field of the instruction code specifies an address register (ERn) which contains the address of a memory operand. After the operand is accessed, 1, 2, or 4 is added to the address register contents and the sum is stored in the address register. The value added is 1 for byte access, 2 for word access, or 4 for longword access. For word or longword access, the register value should be even.
- Register indirect with pre-decrement—@-ERn

The value 1, 2, or 4 is subtracted from an address register (ERn) specified by the register field in the instruction code, and the result becomes the address of a memory operand. The result is also stored in the address register. The value subtracted is 1 for byte access, 2 for word access, or 4 for longword access. For word or longword access, the register value should be even.

Absolute Address—@aa:8, @aa:16, @aa:24, or @aa:32

The instruction code contains the absolute address of a memory operand. The absolute address may be 8 bits long (@aa:8), 16 bits long (@aa:16), 24 bits long (@aa:24), or 32 bits long (@aa:32).

To access data, the absolute address should be 8 bits (@aa:8), 16 bits (@aa:16), or 32 bits (@aa:32) long. For an 8-bit absolute address, the upper 24 bits are all assumed to be 1 (H'FFFF). For a 16-bit absolute address the upper 16 bits are a sign extension. A 32-bit absolute address can access the entire address space.

A 24-bit absolute address (@aa:24) indicates the address of a program instruction. The upper 8 bits are all assumed to be 0 (H'00).

Table 2.5 indicates the accessible absolute address ranges.

Table 2.5 Absolute Address Access Ranges

Absolute Address		Normal Mode	Advanced Mode
Data address	8 bits (@aa:8)	H'FF00 to H'FFFF	H'FFFF00 to H'FFFFFF
	16 bits (@aa:16)	H'0000 to H'FFFF	H'000000 to H'007FFF, H'FF8000 to H'FFFFFF
	32 bits (@aa:32)		H'000000 to H'FFFFF
Program instruction address	24 bits (@aa:24)		

Immediate—#xx:8, #xx:16, or #xx:32

The instruction contains 8-bit (#xx:8), 16-bit (#xx:16), or 32-bit (#xx:32) immediate data as an operand.

The ADDS, SUBS, INC, and DEC instructions contain immediate data implicitly. Some bit-manipulation instructions contain 3-bit immediate data in the instruction code, specifying a bit number. The TRAPA instruction contains 2-bit immediate data in its instruction code, specifying a vector address.

Program-Counter Relative—@(d:8, PC) or @(d:16, PC)

This mode is used in the Bcc and BSR instructions. An 8-bit or 16-bit displacement contained in the instruction is sign-extended and added to the 24-bit PC contents to generate a branch address. Only the lower 24 bits of this branch address are valid; the upper 8 bits are all assumed to be 0 (H'00). The PC value to which the displacement is added is the address of the first byte of the next instruction, so the possible branching range is -126 to +128 bytes (-63 to +64 words) or -32766 to +32768 bytes (-16383 to +16384 words) from the branch instruction. The resulting value should be an even number.

Memory Indirect—@@aa:8

This mode can be used by the JMP and JSR instructions. The instruction code contains an 8-bit absolute address specifying a memory operand. This memory operand contains a branch address. The upper bits of the absolute address are all assumed to be 0, so the address range is 0 to 255 (H'0000 to H'00FF in normal mode, H'000000 to H'000FF in advanced mode). In normal mode the memory operand is a word operand and the branch address is 16 bits long. In advanced mode the memory operand is a longword operand, the first byte of which is assumed to be all 0 (H'00).

Note that the first part of the address range is also the exception vector area. For further details, refer to section 4, Exception Handling.



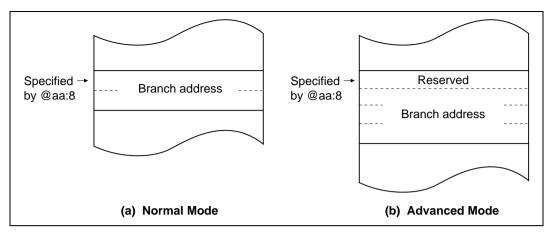


Figure 2.13 Branch Address Specification in Memory Indirect Mode

If an odd address is specified in word or longword memory access, or as a branch address, the least significant bit is regarded as 0, causing data to be accessed or an instruction code to be fetched at the address preceding the specified address. (For further information, see section 2.5.2, Memory Data Formats.)

2.7.2 Effective Address Calculation

Table 2.6 indicates how effective addresses are calculated in each addressing mode. In normal mode the upper 8 bits of the effective address are ignored in order to generate a 16-bit address.

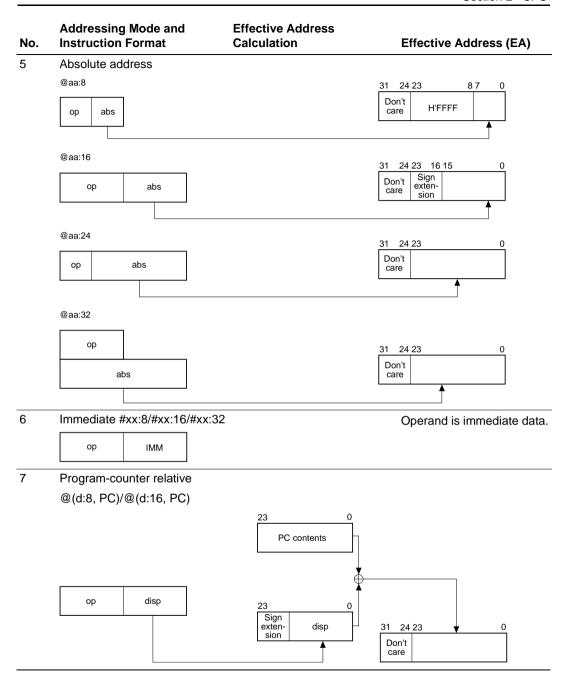
Table 2.6 Effective Address Calculation

No.	Addressing Mode and Instruction Format	Effective Address Calculation	Effective Address (EA)
1	Register direct (Rn)		Operand is general register contents.
2	Register indirect (@ERn)		
	ор г	31 0 General register contents	31 24 23 0 Don't care
3	Register indirect with displace	ement	
	@(d:16, ERn) or @(d:32, ER	General register contents 31 0 Sign extension disp	31 24 23 0 Don't care
4	Register indirect with post-in	•	
	Register indirect with pos	General register contents 1, 2, or 4	31 24 23 0 Don't care
	Register indirect with pre		
	ор г	General register contents Operand Value Size Added Byte 1 Operand Value 1, 2, or 4	31 24 23 0 Don't care

2

Word

Longword



Addressing Mode and Effective Address Instruction Format Calculation **Effective Address (EA)** No. Memory indirect @@aa:8 8 Normal mode abs ор 87 ₩ 0 31 H'000000 abs 31 24 23 16 15 0 Don't H'00 care 15 Memory contents Advanced mode ор abs 31 87 ₩ 0 H'000000 abs 31 31 24 23 Don't Memory contents care

2.8 Processing States

2.8.1 Overview

The CPU has five main processing states: the reset state, exception-handling state, program execution state, bus-released state, and power-down state. Figure 2.14 shows a diagram of the processing states. Figure 2.15 indicates the state transitions.

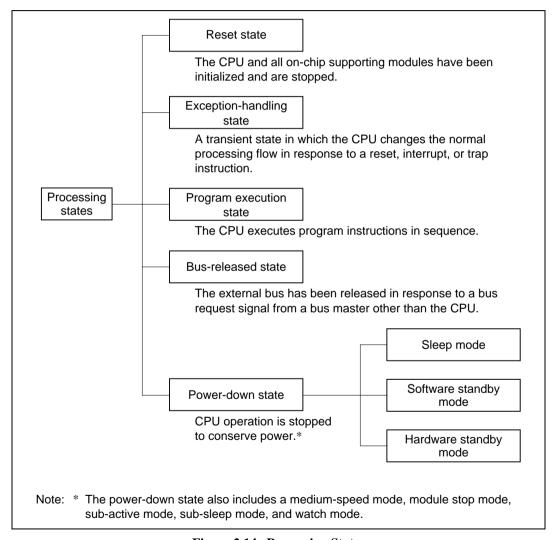


Figure 2.14 Processing States

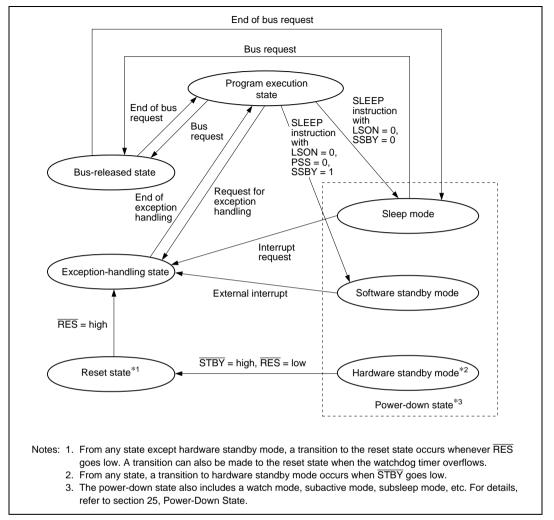


Figure 2.15 State Transitions

2.8.2 Reset State

When the \overline{RES} input goes low all current processing stops and the CPU enters the reset state. All interrupts are disabled in the reset state. Reset exception handling starts when the \overline{RES} signal changes from low to high.

The reset state can also be entered by a watchdog timer overflow. For details, refer to section 14, Watchdog Timer (WDT).

2.8.3 Exception-Handling State

The exception-handling state is a transient state that occurs when the CPU alters the normal processing flow due to a reset, interrupt, or trap instruction. The CPU fetches a start address (vector) from the exception vector table and branches to that address.

Types of Exception Handling and Their Priority

Exception handling is performed for resets, interrupts, and trap instructions. Table 2.7 indicates the types of exception handling and their priority. Trap instruction exception handling is always accepted in the program execution state.

Exception handling and the stack structure depend on the interrupt control mode set in SYSCR.

Table 2.7 Exception Handling Types and Priority

Priority	Type of Exception	Detection Timing	Start of Exception Handling
High	Reset	Synchronized with clock	Exception handling starts immediately after a low-to-high transition at the RES pin, or when the watchdog timer overflows.
	Interrupt	End of instruction execution or end of exception-handling sequence*1	When an interrupt is requested, exception handling starts at the end of the current instruction or current exception-handling sequence.
Low	Trap instruction	When TRAPA instruction is executed	Exception handling starts when a trap (TRAPA) instruction is executed.*2

Notes: 1. Interrupts are not detected at the end of the ANDC, ORC, XORC, and LDC instructions, or immediately after reset exception handling.

2. Trap instruction exception handling is always accepted in the program execution state.

Reset Exception Handling

After the \overline{RES} pin has gone low and the reset state has been entered, when \overline{RES} goes high again, reset exception handling starts. When reset exception handling starts the CPU fetches a start address (vector) from the exception vector table and starts program execution from that address. All interrupts, including NMI, are disabled during reset exception handling and after it ends.

Interrupt Exception Handling and Trap Instruction Exception Handling

When interrupt or trap-instruction exception handling begins, the CPU references the stack pointer (ER7) and pushes the program counter and other control registers onto the stack. Next, the CPU alters the settings of the interrupt mask bits in the control registers. Then the CPU fetches a start address (vector) from the exception vector table and program execution starts from that start address.

Figure 2.16 shows the stack after exception handling ends.

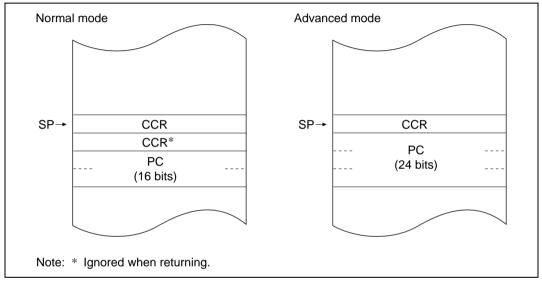


Figure 2.16 Stack Structure after Exception Handling (Examples)

2.8.4 Program Execution State

In this state the CPU executes program instructions in sequence.

2.8.5 Bus-Released State

This is a state in which the bus has been released in response to a bus request from a bus master other than the CPU. While the bus is released, all CPU internal operations are halted.

There is one other bus master in addition to the CPU: the data transfer controller (DTC).

For further details, refer to section 6, Bus Controller.

2.8.6 Power-Down State

The power-down state includes both modes in which the CPU stops operating and modes in which the CPU does not stop. There are five modes in which the CPU stops operating: sleep mode, software standby mode, hardware standby mode, subsleep mode, and watch mode. There are also three other power-down modes: medium-speed mode, module stop mode, and subactive mode. In medium-speed mode, the CPU and other bus masters operate on a medium-speed clock. Module stop mode permits halting of the operation of individual modules, other than the CPU. Subactive mode, subsleep mode, and watch mode are power-down modes that use subclock input. For details, refer to section 25, Power-Down State.

Sleep Mode

A transition to sleep mode is made if the SLEEP instruction is executed while the software standby bit (SSBY) in the standby control register (SBYCR) and the LSON bit in the low-power control register (LPWRCR) are both cleared to 0. In sleep mode, CPU operations stop immediately after execution of the SLEEP instruction. The contents of CPU registers are retained.

Software Standby Mode

A transition to software standby mode is made if the SLEEP instruction is executed while the SSBY bit in SBYCR is set to 1 and the LSON bit in LPWRCR and the PSS bit in the WDT1 timer control/status register (TCSR) are both cleared to 0. In software standby mode, the CPU and clock halt and all MCU operations stop. As long as a specified voltage is supplied, the contents of CPU registers and on-chip RAM are retained. The I/O ports also remain in their existing states.

Hardware Standby Mode

A transition to hardware standby mode is made when the \overline{STBY} pin goes low. In hardware standby mode, the CPU and clock halt and all MCU operations stop. The on-chip supporting modules are reset, but as long as a specified voltage is supplied, on-chip RAM contents are retained.

2.9 Basic Timing

2.9.1 Overview

The CPU is driven by a system clock, denoted by the symbol ϕ . The period from one rising edge of ϕ to the next is referred to as a "state." The memory cycle or bus cycle consists of one, two, or three states. Different methods are used to access on-chip memory, on-chip supporting modules, and the external address space.

2.9.2 On-Chip Memory (ROM, RAM)

On-chip memory is accessed in one state. The data bus is 16 bits wide, permitting both byte and word transfer instruction. Figure 2.17 shows the on-chip memory access cycle. Figure 2.18 shows the pin states.

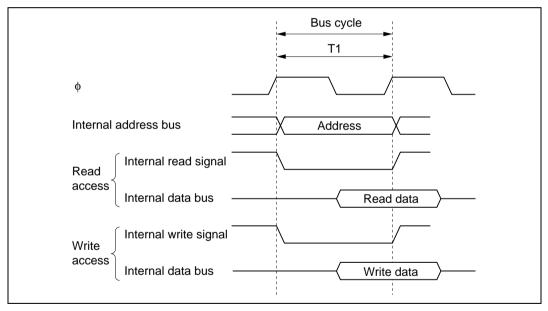


Figure 2.17 On-Chip Memory Access Cycle

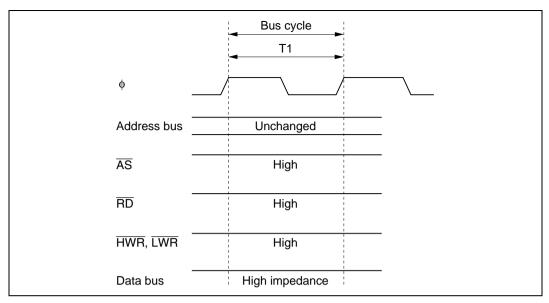


Figure 2.18 Pin States during On-Chip Memory Access

2.9.3 On-Chip Supporting Module Access Timing

The on-chip supporting modules are accessed in two states. The data bus is either 8 bits or 16 bits wide, depending on the particular internal I/O register being accessed. Figure 2.19 shows the access timing for the on-chip supporting modules. Figure 2.20 shows the pin states.

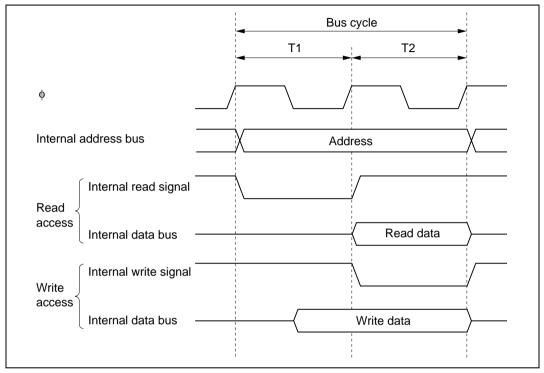


Figure 2.19 On-Chip Supporting Module Access Cycle

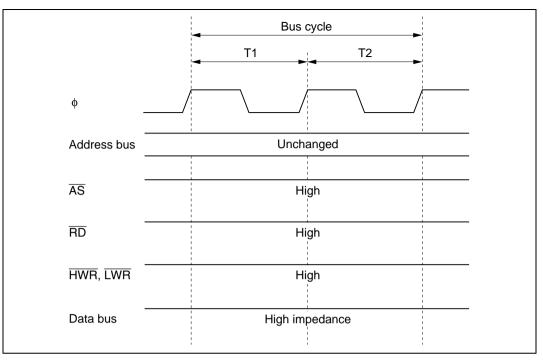


Figure 2.20 Pin States during On-Chip Supporting Module Access

2.9.4 External Address Space Access Timing

The external address space is accessed with an 8-bit or 16-bit data bus width in a two-state or three-state bus cycle. In three-state access, wait states can be inserted. For further details, refer to section 6, Bus Controller.

2.10 Usage Note

2.10.1 TAS Instruction

Only register ER0, ER1, ER4, or ER5 should be used when using the TAS instruction. The TAS instruction is not generated by the Renesas H8S and H8/300 series C/C++ compilers. If the TAS instruction is used as a user-defined intrinsic function, ensure that only register ER0, ER1, ER4, or ER5 is used

2.10.2 STM/LDM Instruction

ER7 is not used as the register that can be saved (STM)/restored (LDM) when using STM/LDM instruction, because ER7 is the stack pointer. Two, three, or four registers can be saved/restored by one STM/LDM instruction. The following ranges can be specified in the register list.

Two registers: ER0—ER1, ER2—ER3, or ER4—ER5

Three registers: ER0—ER2, or ER4—ER6

Four registers: ER0—ER3

The STM/LDM instruction including ER7 is not generated by the Renesas H8S and H8/300 series C/C++compilers.



Section 3 MCU Operating Modes

3.1 Overview

3.1.1 Operating Mode Selection

This LSI has three operating modes (modes 1 to 3). These modes enable selection of the CPU operating mode and enabling/disabling of on-chip ROM, by setting the mode pins (MD1 and MD0).

Table 3.1 lists the MCU operating modes.

Table 3.1 MCU Operating Mode Selection

MCU Operating Mode	MD1	MD0	CPU Operating Mode	Description	On-Chip ROM
0	0	0	_	_	_
1		1	Normal	Expanded mode with on-chip ROM disabled	Disabled
2	1	0	Advanced	Expanded mode with on-chip ROM enabled	Enabled
				Single-chip mode	
3	_	1	Normal	Expanded mode with on-chip ROM enabled	
				Single-chip mode	

The CPU's architecture allows for 4 Gbytes of address space, but this LSI actually access a maximum of 16 Mbytes.

Mode 1 is an externally expanded mode that allows access to external memory and peripheral devices. With modes 2 and 3, operation begins in single-chip mode after reset release, but a transition can be made to external expansion mode by setting the EXPE bit in MDCR.

This LSI can only be used in modes 1 to 3. These means that the mode pins must select one of these modes. Do not changes the inputs at the mode pins during operation.

3.1.2 Register Configuration

This LSI have a mode control register (MDCR) that indicates the inputs at the mode pins (MD1 and MD0), a system control register (SYSCR) and bus control register (BCR) that control the operation of the MCU, and a serial/timer control register (STCR) that controls the operation of the supporting modules. Table 3.2 summarizes these registers.

Table 3.2 MCU Registers

Name	Abbreviation	R/W	Initial Value	Address*
Mode control register	MDCR	R/W	Undetermined	H'FFC5
System control register	SYSCR	R/W	H'09	H'FFC4
Bus control register	BCR	R/W	H'D7	H'FFC6
Serial/timer control register	STCR	R/W	H'00	H'FFC3

Note: * Lower 16 bits of the address.

3.2 Register Descriptions

3.2.1 Mode Control Register (MDCR)

Bit	7	6	5	4	3	2	1	0	_
	EXPE	_	_	_	_	_	MDS1	MDS0	
Initial value	*	0	0	0	0	0	*	*	•
Read/Write	R/W*	_	_	_	_	_	R	R	

Note: * Determined by pins MD1 and MD0.

MDCR is an 8-bit read-only register that indicates the operating mode setting and the current operating mode of the MCU.

The EXPE bit is initialized in coordination with the mode pin states by a reset and in hardware standby mode.

Bit 7—Expanded Mode Enable (EXPE): Sets expanded mode. In mode 1, this bit is fixed at 1 and cannot be modified. In modes 2 and 3, this bit has an initial value of 0, and can be read and written.

Bit 7

EXPE	Description
0	Single chip mode is selected
1	Expanded mode is selected

Bits 6 to 2—Reserved: These bits cannot be modified and are always read as 0.

Bits 1 and 0—Mode Select 1 and 0 (MDS1, MDS0): These bits indicate the input levels at pins MD1 and MD0 (the current operating mode). Bits MDS1 and MDS0 correspond to MD1 and MD0. MDS1 and MDS0 are read-only bits—they cannot be written to. The mode pin (MD1 and MD0) input levels are latched into these bits when MDCR is read.

3.2.2 System Control Register (SYSCR)

Bit	7	6	5	4	3	2	1	0
	CS2E	IOSE	INTM1	INTM0	XRST	NMIEG	HIE	RAME
Initial value	0	0	0	0	1	0	0	1
Read/Write	R/W	R/W	R	R/W	R	R/W	R/W	R/W

SYSCR is an 8-bit readable/writable register that performs selection of system pin functions, reset source monitoring, interrupt control mode selection, NMI detected edge selection, supporting module pin location selection, supporting module register access control, and RAM address space control.

Only bits 7, 6, 3, 1, and 0 are described here. For a detailed description of these bits, refer also to the description of the relevant modules (host interface, bus controller, watchdog timer, RAM, etc.). For information on bits 5, 4, and 2, see section 5.2.1, System Control Register (SYSCR).

SYSCR is initialized to H'09 by a reset and in hardware standby mode. It is not initialized in software standby mode.

Bit 7—Chip Select 2 Enable (CS2E): Specifies the location of the host interface control pin (CS2). For details, see section 18, Host Interface. The H8S/2144 Group does not incorporate a host interface, so do not set this bit to 1 in the H8S/2144 Group.

Bit 6—IOS Enable (IOSE): Controls the function of the AS/IOS pin in expanded mode.

Bit 6

IOSE		Description	
0		The $\overline{AS}/\overline{IOS}$ pin functions as the address strobe pin (\overline{AS}) (Low output when accessing an external area)	(Initial value)
1		The AS/IOS pin functions as the I/O strobe pin (IOS) (Low output when accessing a specified address from H'(FF)F000 to H	r(FF)FE4F)*
Note:	*	In the H8S/2148 F-ZTAT A-mask version and H8S/2147 F-ZTAT A-mask address range is from H'(FF)F000 to H'(FF)F7FF.	k version, the

Bit 3—External Reset (XRST): Indicates the reset source. When the watchdog timer is used, a reset can be generated by watchdog timer overflow as well as by external reset input. XRST is a read-only bit. It is set to 1 by an external reset and cleared to 0 by watchdog timer overflow.

Bit 3

XRST	Description	
0	A reset is generated by watchdog timer overflow	
1	A reset is generated by an external reset	(Initial value)

Bit 1—Host Interface Enable (HIE): This bit controls CPU access to the host interface data registers and control registers (HICR, IDR1, ODR1, STR1, IDR2, ODR2, and STR2), the keyboard controller and MOS input pull-up control registers (KMIMR, KMPCR, and KMIMRA), the 8-bit timer (channel X and Y) data registers and control registers (TCRX/TCRY, TCSRX/TCSRY, TICRR/TCORAY, TICRF/TCORBY, TCNTX/TCNTY, TCORC/TISR, TCORAX, and TCORBX), and the timer connection control registers (TCONRI, TCONRO, TCONRS, and SEDGR).

Bit 1

HIE	Description	
0	In areas H'(FF)FFF0 to H'(FF)FFF7 and H'(FF)FFFC to H'(FF)FFFF, CPU access to 8-bit timer (channel X and Y) data registers and control registers, and timer connection control registers, is permitted	(Initial value)
1	In areas H'(FF)FFF0 to H'(FF)FFF7 and H'(FF)FFFC to H'(FF)FFFF, CPU access to host interface data registers and control registers, and keyboard controller and MOS input pull-up control registers, is permitted	

Bit 0—RAM Enable (RAME): Enables or disables the on-chip RAM. The RAME bit is initialized when the reset state is released. It is not initialized in software standby mode.

Bit 0

RAME	 Description	
0	On-chip RAM is disabled	
1	On-chip RAM is enabled	(Initial value)

3.2.3 Bus Control Register (BCR)

Bit	7	6	5	4	3	2	1	0
	ICIS1	ICIS0	BRSTRM	BRSTS1	BRSTS0	_	IOS1	IOS0
Initial value	1	1	0	1	0	1	1	1
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

BCR is an 8-bit readable/writable register that specifies the external memory space access mode, and the I/O area range when the \overline{AS} pin is designated for use as the I/O strobe. For details on bits 7 to 2, see section 6.2.1, Bus Control Register (BCR).

BCR is initialized to H'D7 by a reset and in hardware standby mode.

Bits 1 and 0—IOS Select 1 and 0 (IOS1, IOS0): These bits specify the addresses for which the $\overline{AS/IOS}$ pin output goes low when IOSE = 1.

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Bit 1	Bit 0		
IOS1	IOS0	Description	
0	0	The AS/IOS pin output goes low in accesses to addresses H'(FF)F000 to H'(FF)F03F	
	1	The AS/IOS pin output goes low in accesses to addresses H'(FF)F000 to H'(FF)F0FF	
1	0	The AS/IOS pin output goes low in accesses to addresses H'(FF)F000 to H'(FF)F3FF	
	1	The AS/IOS pin output goes low in accesses to addresses H'(FF)F000 to H'(FF)FE4F*	(Initial value)

Note: * In the H8S/2148 F-ZTAT A-mask version and H8S/2147 F-ZTAT A-mask version, the address range is from H'(FF)F000 to H'(FF)F7FF.

3.2.4 Serial Timer Control Register (STCR)

Bit	7	6	5	4	3	2	1	0
	IICS	IICX1	IICX0	IICE	FLSHE	_	ICKS1	ICKS0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

STCR is an 8-bit readable/writable register that controls register access, the IIC operating mode (when the on-chip IIC option is included), an on-chip flash memory control (in F-ZTAT versions), and also selects the TCNT input clock. For details of functions other than register access control, see the descriptions of the relevant modules. If a module controlled by STCR is not used, do not write 1 to the corresponding bit.

STCR is initialized to H'00 by a reset and in hardware standby mode.

Bits 7 to 5—I²C Control (IICS, IICX1, IICX0): These bits control the bus buffer function of the port A and the operation of the I²C bus interface when the on-chip IIC option is included. For details, see section 16.2.7, Serial/Timer Control Register (STCR).

Bit 4—I²C **Master Enable (IICE):** Controls CPU access to the I²C bus interface data registers and control registers (ICCR, ICSR, ICDR/SARX, and ICMR/SAR), the PWMX data registers and control registers (DADRAH/DACR, DADRAL, DADRBH/DACNTH, and DADRBL/DACNTL), and the SCI control registers (SMR, BRR, and SCMR).

Bit 4

IICE	Description
0	Addresses H'(FF)FF88 and H'(FF)FF89, and H'(FF)FF8E and H'(FF)FF8F, are used for SCI1 control register access (Initial value)
	Addresses H'(FF)FFA0 and H'(FF)FFA1, and H'(FF)FFA6 and H'(FF)FFA7, are used for SCI2 control register access
	Addresses H'(FF)FFD8 and H'(FF)FFD9, and H'(FF)FFDE and H'(FF)FFDF, are used for SCI0 control register access
1	Addresses H'(FF)FF88 and H'(FF)FF89, and H'(FF)FF8E and H'(FF)FF8F, are used for IIC1 data register and control register access
	Addresses H'(FF)FFA0 and H'(FF)FFA1, and H'(FF)FFA6 and H'(FF)FFA7, are used for PWMX data register and control register access
	Addresses H'(FF)FFD8 and H'(FF)FFD9, and H'(FF)FFDE and H'(FF)FFDF, are used for IIC0 data register and control register access

Bit 3—Flash Memory Control Register Enable (FLSHE): Controls CPU access to the flash memory control registers (FLMCR1, FLMCR2, EBR1, and EBR2), the power-down mode control registers (SBYCR, LPWRCR, MSTPCRH, and MSTPCRL), and the supporting module control register (PCSR and SYSCR2).

Bit 3

FLSHE	Description
0	Addresses H'(FF)FF80 to H'(FF)FF87 are used for power-down mode control register and supporting module control register access (Initial value)
1	Addresses H'(FF)FF80 to H'(FF)FF87 are used for flash memory control register access (F-ZTAT version only)

Bit 2—Reserved: Do not write 1 to this bit.

Bits 1 and 0—Internal Clock Select 1 and 0 (ICKS1, ICKS0): These bits, together with bits CKS2 to CKS0 in TCR, select the clock to be input to TCNT. For details, see section 12.2.4, Timer Control Register (TCR).

3.3 Operating Mode Descriptions

3.3.1 Mode 1

The CPU can access a 64-kbyte address space in normal mode. The on-chip ROM is disabled.

Ports 1 and 2 function as an address bus, port 3 functions as a data bus, and part of port 9 carries bus control signals. Clearing the ABW bit to 0 in the WSCR register makes port B a data bus.

3.3.2 Mode 2

The CPU can access a 16-Mbyte address space in advanced mode. The on-chip ROM is enabled. After a reset, single-chip mode is set, and the EXPE bit in MDCR must be set to 1 in order to use external addresses.

When the EXPE bit in MDCR is set to 1, ports 1, 2 and A function as input ports after a reset. They can be set to output addresses by setting the corresponding bits in the data direction register (DDR) to 1. Port 3 functions as a data bus, and part of port 9 carries bus control signals. Clearing the ABW bit to 0 in the WSCR register makes port B a data bus.

3.3.3 Mode 3

The CPU can access a 64-kbyte address space in normal mode. The on-chip ROM is enabled. After a reset, single-chip mode is set, and the EXPE bit in MDCR must be set to 1 in order to use external addresses

When the EXPE bit in MDCR is set to 1, ports 1 and 2 function as input ports after a reset. They can be set to output addresses by setting the corresponding bits in the data direction register (DDR) to 1. Port 3 functions as a data bus, and part of port 9 carries bus control signals. Clearing the ABW bit to 0 in the WSCR register makes port B a data bus.

In this operating mode, the available amount of on-chip ROM in products with 64 kbytes or more of ROM is limited to 56 kbytes.

3.4 **Pin Functions in Each Operating Mode**

The pin functions of ports 1 to 3, 9, A, and B vary depending on the operating mode. Table 3.3 shows their functions in each operating mode.

Table 3.3 **Pin Functions in Each Mode**

Port		Mode 1	Mode 2	Mode 3	
Port 1		Α	P*/A	P*/A	
Port 2		Α	P*/A	P*/A	
Port A		Р	P*/A	Р	
Port 3		D	P*/D	P*/D	
Port B		P*/D	P*/D	P*/D	
Port 9	P97	P*/C	P*/C	P*/C	
	P96	C*/P	P*/C	P*/C	
	P95 to P93	С	P*/C	P*/C	
	P92 to P91	Р	Р	Р	
	P90	P*/C	P*/C	P*/C	

Legend:

P: I/O port

A: Address bus output

D: Data bus I/O

C: Control signals, clock I/O

*: After reset

REJ09B0327-0400



3.5 **Memory Map in Each Operating Mode**

Figures 3.1 to 3.5 show memory maps for each of the operating modes.

The address space is 64 kbytes in modes 1 and 3 (normal modes), and 16 Mbytes in mode 2 (advanced mode).

The on-chip ROM capacity is 64 kbytes (H8S/2142, H8S/2147, and H8S/2147N), 96 kbytes (H8S/2143), or 128 kbytes (H8S/2144 and H8S/2148), but only 56 kbytes are available in mode 3 (normal mode).

Do not access the reserved area and addresses of modules not supported by the product. Note that normal operation is not guaranteed when these regions are accessed.

For details, see section 6, Bus Controller.

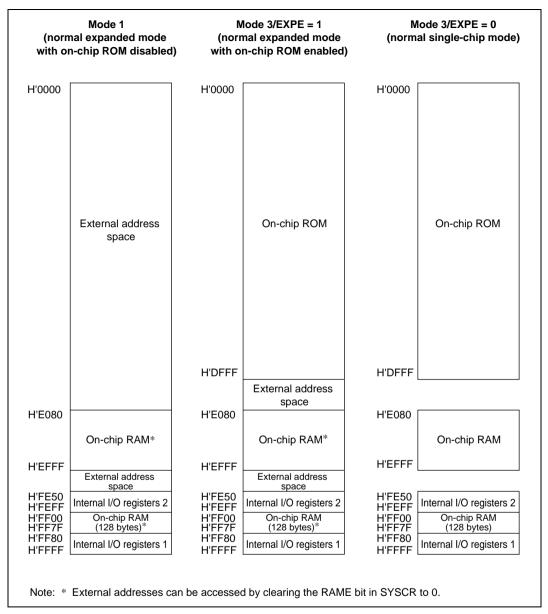


Figure 3.1 H8S/2148 (Except for F-ZTAT A-Mask Version) and H8S/2144 Memory Map in Each Operating Mode

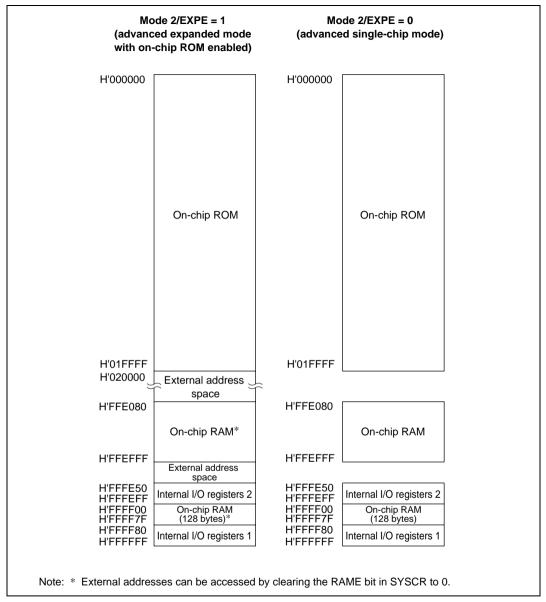


Figure 3.1 H8S/2148 (Except for F-ZTAT A-Mask Version) and H8S/2144 Memory Map in Each Operating Mode (cont)

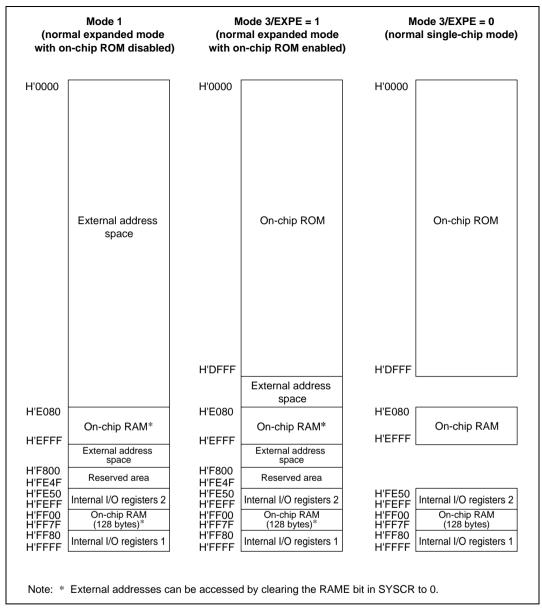


Figure 3.2 H8S/2148 F-ZTAT A-Mask Version Memory Map in Each Operating Mode

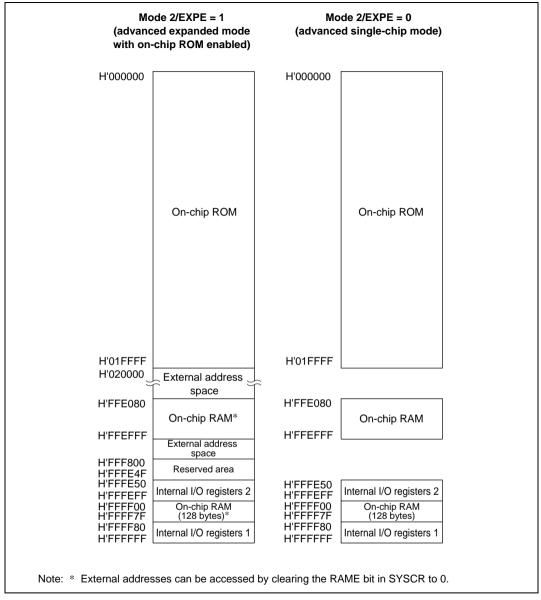


Figure 3.2 H8S/2148 F-ZTAT A-Mask Version Memory Map in Each Operating Mode (cont)

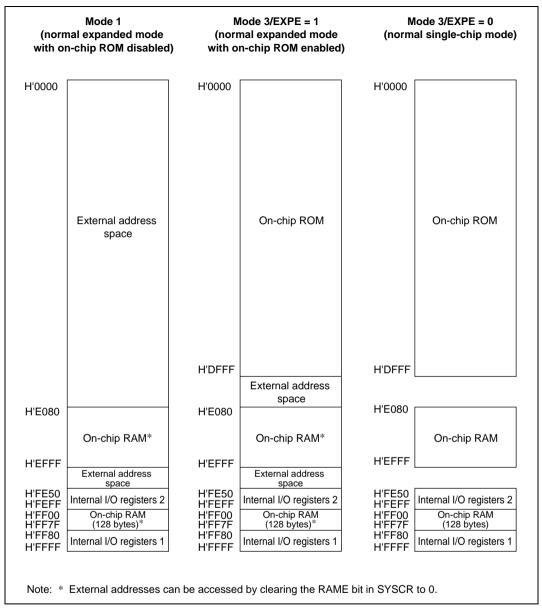


Figure 3.3 H8S/2143 Memory Map in Each Operating Mode

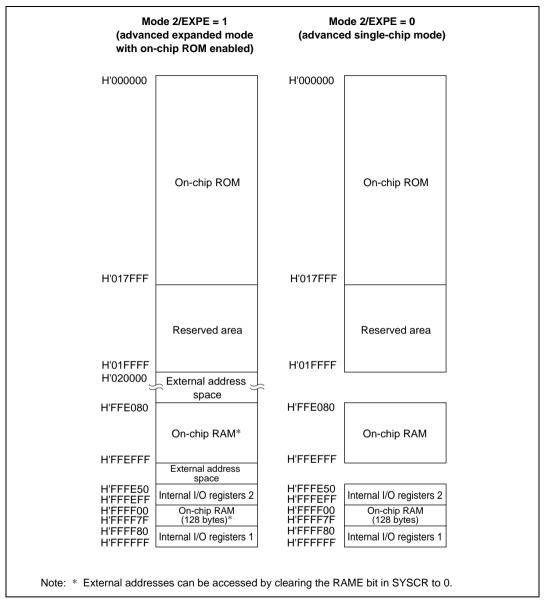


Figure 3.3 H8S/2143 Memory Map in Each Operating Mode (cont)

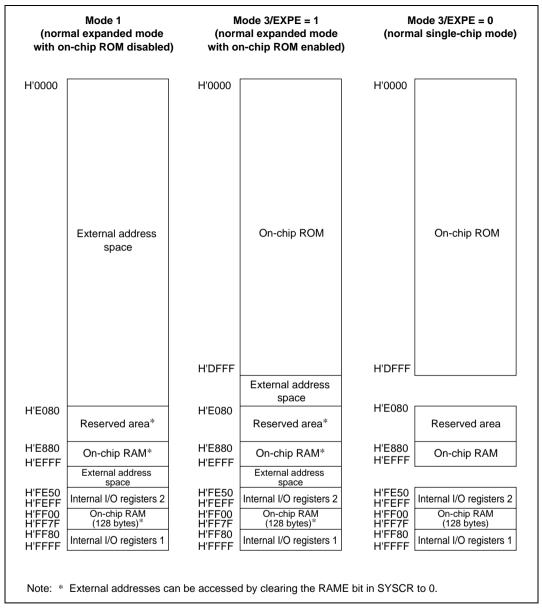


Figure 3.4 H8S/2147 (Except for F-ZTAT A-Mask Version), H8S/2147N, and H8S/2142 Memory Map in Each Operating Mode

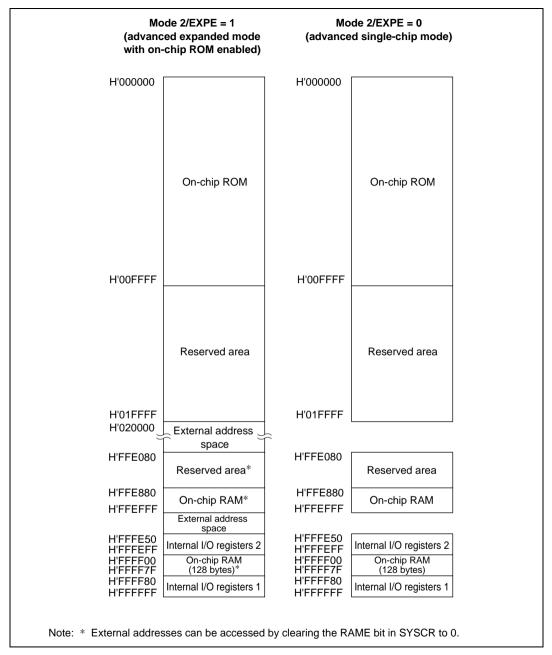


Figure 3.4 H8S/2147 (Except for F-ZTAT A-Mask Version), H8S/2147N, and H8S/2142 Memory Map in Each Operating Mode (cont)

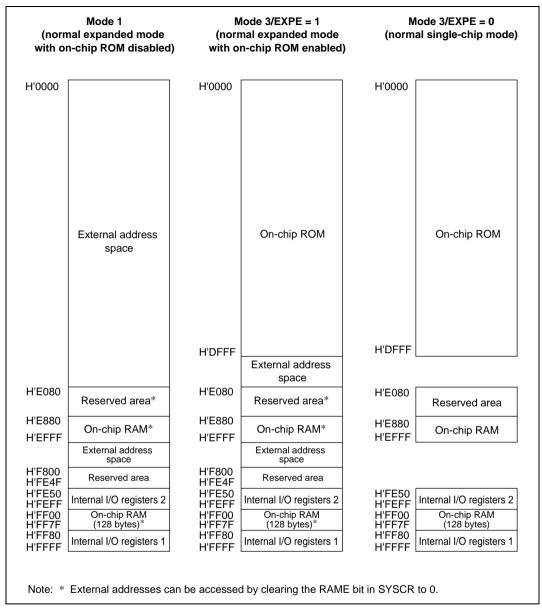


Figure 3.5 H8S/2147 F-ZTAT A-Mask Version Memory Map in Each Operating Mode

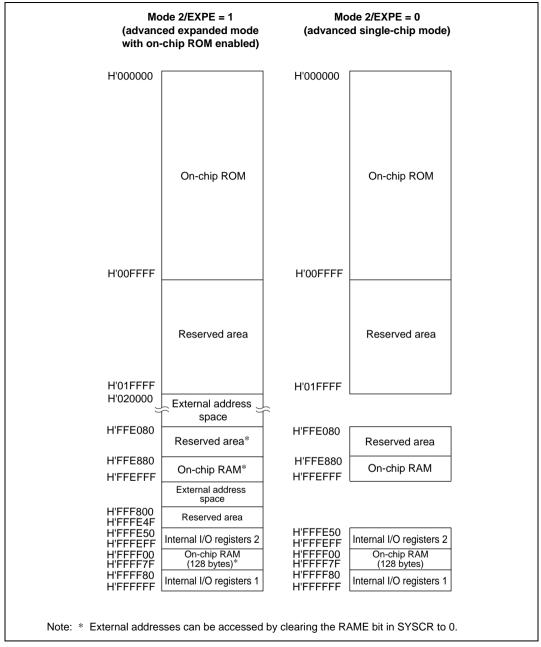


Figure 3.5 H8S/2147 F-ZTAT A-Mask Version Memory Map in Each Operating Mode (cont)

Section 4 Exception Handling

4.1 Overview

4.1.1 Exception Handling Types and Priority

As table 4.1 indicates, exception handling may be caused by a reset, trap instruction, or interrupt. Exception handling is prioritized as shown in table 4.1. If two or more exceptions occur simultaneously, they are accepted and processed in order of priority. Trap instruction exceptions are accepted at all times in the program execution state.

Exception handling sources, the stack structure, and the operation of the CPU vary depending on the interrupt control mode set by the INTM0 and INTM1 bits in SYSCR.

Table 4.1 Exception Types and Priority

Priority	Exception Type	Start of Exception Handling
High	Reset	Starts immediately after a low-to-high transition at the RES pin, or when the watchdog timer overflows.
	RES pin, or when the watchdog timer overflows. Trace*1 Starts when execution of the current instruction or exception handling ends, if the trace (T) bit is set to 1. Interrupt Starts when execution of the current instruction or exception handling ends, if an interrupt request has been issued.*2 Direct transition Started by a direct transition resulting from execution of	
	Interrupt	exception handling ends, if an interrupt request has been
	Direct transition	Started by a direct transition resulting from execution of a SLEEP instruction.
Low	Trap instruction (TRAPA)*3	Started by execution of a trap instruction (TRAPA).

 Traces are enabled only in interrupt control modes 2 and 3. (They cannot be used in this LSI.) Trace exception handling is not executed after execution of an RTE instruction.

- 2. Interrupt detection is not performed on completion of ANDC, ORC, XORC, or LDC instruction execution, or on completion of reset exception handling.
- 3. Trap instruction exception handling requests are accepted at all times in the program execution state.

4.1.2 Exception Handling Operation

Exceptions originate from various sources. Trap instructions and interrupts are handled as follows:

- 1. The program counter (PC) and condition-code register (CCR) are pushed onto the stack.
- 2. The interrupt mask bits are updated. The T bit is cleared to 0.
- 3. A vector address corresponding to the exception source is generated, and program execution starts from that address.

For a reset exception, steps 2 and 3 above are carried out.

4.1.3 Exception Sources and Vector Table

The exception sources are classified as shown in figure 4.1. Different vector addresses are assigned to different exception sources.

Table 4.2 lists the exception sources and their vector addresses.

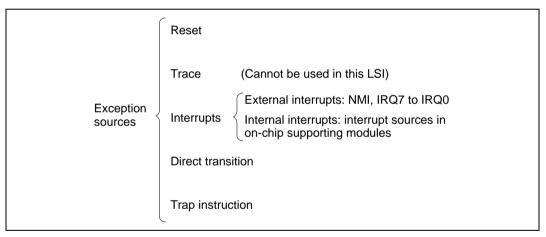


Figure 4.1 Exception Sources

Table 4.2 Exception Vector Table

Vector Address*1

Exception Source		Vector Number	Normal Mode	Advanced Mode
Reset		0	H'0000 to H'0001	H'0000 to H'0003
Reserved for system	m use	1	H'0002 to H'0003	H'0004 to H'0007
		2	H'0004 to H'0005	H'0008 to H'000B
		3	H'0006 to H'0007	H'000C to H'000F
		4	H'0008 to H'0009	H'0010 to H'0013
		5	H'000A to H'000B	H'0014 to H'0017
Direct transition		6	H'000C to H'000D	H'0018 to H'001B
External interrupt	NMI	7	H'000E to H'000F	H'001C to H'001F
Trap instruction (4	sources)	8	H'0010 to H'0011	H'0020 to H'0023
		9	H'0012 to H'0013	H'0024 to H'0027
		10	H'0014 to H'0015	H'0028 to H'002B
		11	H'0016 to H'0017	H'002C to H'002F
Reserved for system	m use	12	H'0018 to H'0019	H'0030 to H'0033
		13	H'001A to H'001B	H'0034 to H'0037
		14	H'001C to H'001D	H'0038 to H'003B
		15	H'001E to H'001F	H'003C to H'003F
External interrupt	IRQ0	16	H'0020 to H'0021	H'0040 to H'0043
	IRQ1	17	H'0022 to H'0023	H'0044 to H'0047
	IRQ2	18	H'0024 to H'0025	H'0048 to H'004B
	IRQ3	19	H'0026 to H'0027	H'004C to H'004F
	IRQ4	20	H'0028 to H'0029	H'0050 to H'0053
	IRQ5	21	H'002A to H'002B	H'0054 to H'0057
	IRQ6	22	H'002C to H'002D	H'0058 to H'005B
	IRQ7	23	H'002E to H'002F	H'005C to H'005F
Internal interrupt*2		24	H'0030 to H'0031	H'0060 to H'0063
		103	H'00CE to H'00CF	H'019C to H'019F

Notes: 1. Lower 16 bits of the address.

2. For details on internal interrupt vectors, see section 5.3.3, Interrupt Exception Vector Table.

4.2 Reset

4.2.1 Overview

A reset has the highest exception priority.

When the RES pin goes low, all processing halts and the MCU enters the reset state. A reset initializes the internal state of the CPU and the registers of on-chip supporting modules. Immediately after a reset, interrupt control mode 0 is set.

Reset exception handling begins when the \overline{RES} pin changes from low to high.

The MCUs can also be reset by overflow of the watchdog timer. For details, see section 14, Watchdog Timer (WDT).

4.2.2 Reset Sequence

The MCU enters the reset state when the \overline{RES} pin goes low.

To ensure that the chip is reset, hold the \overline{RES} pin low for at least 20 ms when powering on. To reset the chip during operation, hold the \overline{RES} pin low for at least 20 states. For pin states in a reset, see appendix D.1, Port States in Each Processing State.

When the \overline{RES} pin goes high after being held low for the necessary time, the chip starts reset exception handling as follows:

- [1] The internal state of the CPU and the registers of the on-chip supporting modules are initialized, and the I bit is set to 1 in CCR.
- [2] The reset exception vector address is read and transferred to the PC, and program execution starts from the address indicated by the PC.

Figures 4.2 and 4.3 show examples of the reset sequence.



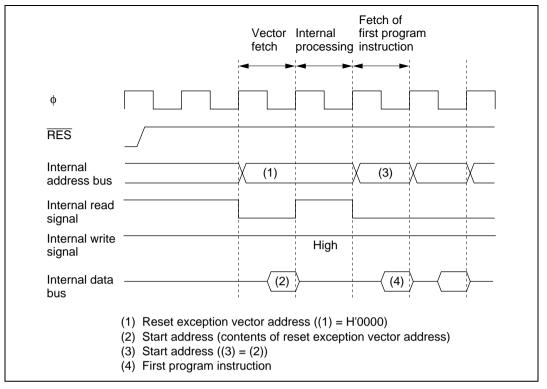


Figure 4.2 Reset Sequence (Mode 3)

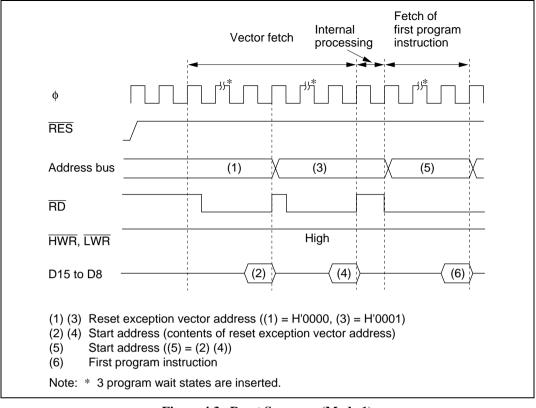


Figure 4.3 Reset Sequence (Mode 1)

4.2.3 Interrupts after Reset

If an interrupt is accepted after a reset but before the stack pointer (SP) is initialized, the PC and CCR will not be saved correctly, leading to a program crash. To prevent this, all interrupt requests, including NMI, are disabled immediately after a reset. Since the first instruction of a program is always executed immediately after the reset state ends, make sure that this instruction initializes the stack pointer (example: MOV.L #xx:32, SP).

4.3 Interrupts

Interrupt exception handling can be requested by nine external sources (NMI and IRQ7 to IRQ0) from 23 input pins (NMI, $\overline{IRQ7}$ to $\overline{IRQ0}$, and $\overline{KIN15}$ to $\overline{KIN0}$), and internal sources in the on-chip supporting modules. Figure 4.4 shows the interrupt sources and the number of interrupts of each type.

The on-chip supporting modules that can request interrupts include the watchdog timer (WDT), 16-bit free-running timer (FRT), 8-bit timer (TMR), serial communication interface (SCI), data transfer controller (DTC), A/D converter (ADC), host interface (HIF), keyboard buffer controller (PS2), and I²C bus interface (option). Each interrupt source has a separate vector address.

NMI is the highest-priority interrupt. Interrupts are controlled by the interrupt controller. The interrupt controller has two interrupt control modes and can assign interrupts other than NMI and address break to either three priority/mask levels to enable multiplexed interrupt control.

For details on interrupts, see section 5, Interrupt Controller.

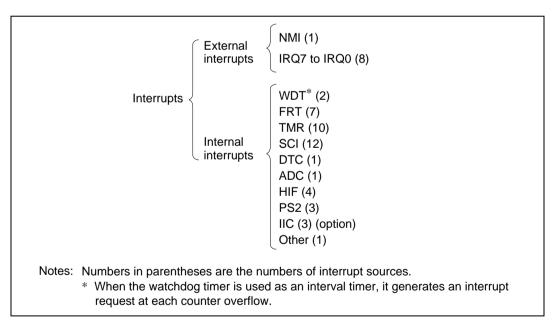


Figure 4.4 Interrupt Sources and Number of Interrupts

4.4 **Trap Instruction**

Trap instruction exception handling starts when a TRAPA instruction is executed. Trap instruction exception handling can be executed at all times in the program execution state.

The TRAPA instruction fetches a start address from a vector table entry corresponding to a vector number from 0 to 3, as specified in the instruction code.

Table 4.3 shows the status of CCR and EXR after execution of trap instruction exception handling.

Table 4.3 Status of CCR and EXR after Trap Instruction Exception Handling

	CCR		EXR		
Interrupt Control Mode	I	UI	12 to 10	Т	
0	1	_	_	_	
1	1	1	_	_	

Legend:

1: Set to 1

0: Cleared to 0

—: Retains value prior to execution.



4.5 Stack Status after Exception Handling

Figure 4.5 shows the stack after completion of trap instruction exception handling and interrupt exception handling.

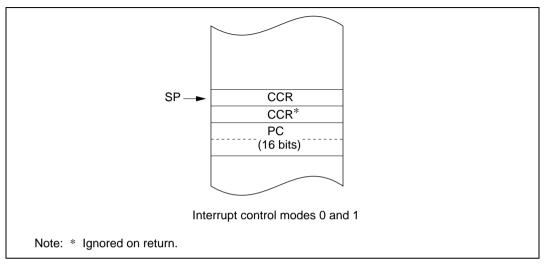


Figure 4.5 (1) Stack Status after Exception Handling (Normal Mode)

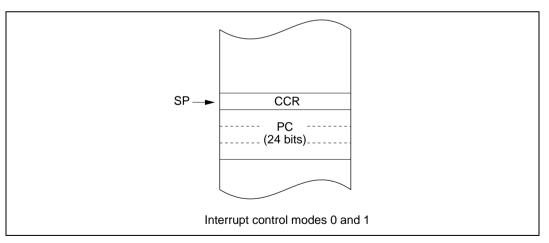


Figure 4.5 (2) Stack Status after Exception Handling (Advanced Mode)

4.6 Notes on Use of the Stack

When accessing word data or longword data, this LSI assumes that the lowest address bit is 0. The stack should always be accessed by word transfer instruction or longword transfer instruction, and the value of the stack pointer (SP: ER7) should always be kept even. Use the following instructions to save registers:

```
PUSH.W Rn (or MOV.W Rn, @-SP)

PUSH.L ERn (or MOV.L ERn, @-SP)
```

Use the following instructions to restore registers:

```
POP.W Rn (or MOV.W @SP+, Rn)
POP.L ERn (or MOV.L @SP+, ERn)
```

Setting SP to an odd value may lead to a malfunction. Figure 4.6 shows an example of what happens when the SP value is odd.

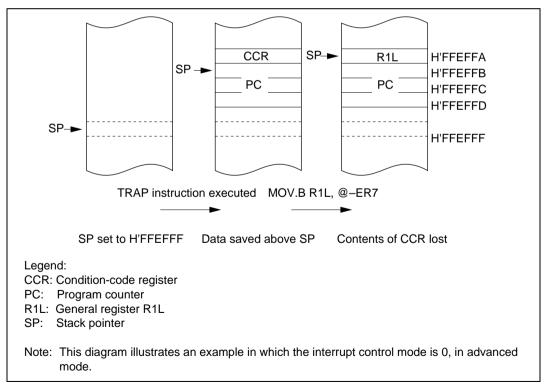


Figure 4.6 Operation when SP Value Is Odd

Section 5 Interrupt Controller

5.1 Overview

5.1.1 Features

This LSI control interrupts by means of an interrupt controller. The interrupt controller has the following features:

- Two interrupt control modes
 - Either of two interrupt control modes can be set by means of the INTM1 and INTM0 bits in the system control register (SYSCR).
- Priorities settable with ICR
 - An interrupt control register (ICR) is provided for setting interrupt priorities. Three priority levels can be set for each module for all interrupts except NMI and address break.
- Independent vector addresses
 - All interrupt sources are assigned independent vector addresses, making it unnecessary for the source to be identified in the interrupt handling routine.
- Twenty-three external interrupt pins (nine external sources)
 - NMI is the highest-priority interrupt, and is accepted at all times. A rising or falling edge at the NMI pin can be selected for the NMI interrupt.
 - Falling edge, rising edge, or both edge detection, or level sensing, at pins $\overline{IRQ7}$ to $\overline{IRQ0}$ can be selected for interrupts IRQ7 to IRQ0.
 - The IRQ6 interrupt is shared by the interrupt from the $\overline{IRQ6}$ pin and eight external interrupt inputs ($\overline{KIN7}$ to $\overline{KIN0}$), and the IRQ7 interrupt is shared by the interrupt from the $\overline{IRQ7}$ pin and eight external interrupt inputs ($\overline{KIN15}$ to $\overline{KIN8}$). $\overline{KIN15}$ to $\overline{KIN0}$ can be masked individually by the user program.
- DTC control
 - DTC activation is controlled by means of interrupts.

5.1.2 Block Diagram

A block diagram of the interrupt controller is shown in figure 5.1.

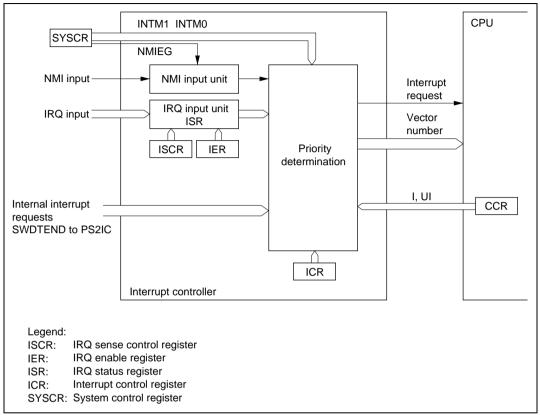


Figure 5.1 Block Diagram of Interrupt Controller

5.1.3 Pin Configuration

Table 5.1 summarizes the pins of the interrupt controller.

Table 5.1 Interrupt Controller Pins

Name	Symbol	I/O	Function
Nonmaskable interrupt	NMI	Input	Nonmaskable external interrupt; rising or falling edge can be selected
External interrupt requests 7 to 0	ĪRQ7 to ĪRQ0	Input	Maskable external interrupts; rising, falling, or both edges, or level sensing, can be selected.
Key input interrupt requests 15 to 0	KIN15 to KIN0	Input	Maskable external interrupts: falling edge or level sensing can be selected.

5.1.4 Register Configuration

Table 5.2 summarizes the registers of the interrupt controller.

Table 5.2 Interrupt Controller Registers

Name	Abbreviation	R/W	Initial Value	Address*1
System control register	SYSCR	R/W	H'09	H'FFC4
IRQ sense control register H	ISCRH	R/W	H'00	H'FEEC
IRQ sense control register L	ISCRL	R/W	H'00	H'FEED
IRQ enable register	IER	R/W	H'00	H'FFC2
IRQ status register	ISR	R/(W)*2	H'00	H'FEEB
Keyboard matrix interrupt mask register	KMIMR	R/W	H'BF	H'FFF1*3
Keyboard matrix interrupt mask register A	KMIMRA	R/W	H'FF	H'FFF3*3
Interrupt control register A	ICRA	R/W	H'00	H'FEE8
Interrupt control register B	ICRB	R/W	H'00	H'FEE9
Interrupt control register C	ICRC	R/W	H'00	H'FEEA
Address break control register	ABRKCR	R/W	H'00	H'FEF4
Break address register A	BARA	R/W	H'00	H'FEF5
Break address register B	BARB	R/W	H'00	H'FEF6
Break address register C	BARC	R/W	H'00	H'FEF7

Notes: 1. Lower 16 bits of the address.

- 2. Only 0 can be written, for flag clearing.
- 3. When setting KMIMR and KMIMRA, the HIE bit in SYSCR must be set to 1, and also MSTP2 bit in MSTPCRL must be set to 0.

5.2 Register Descriptions

5.2.1 System Control Register (SYSCR)

Bit	7	6	5	4	3	2	1	0
	CS2E	IOSE	INTM1	INTM0	XRST	NMIEG	HIE	RAME
Initial value	0	0	0	0	1	0	0	1
Read/Write	R/W	R/W	R	R/W	R	R/W	R/W	R/W

SYSCR is an 8-bit readable/writable register that selects the interrupt control mode, and the detected edge for NMI, among other functions.

Only bits 5, 4, and 2 are described here; for details on the other bits, see section 3.2.2, System Control Register (SYSCR).

SYSCR is initialized to H'09 by a reset and in hardware standby mode. It is not initialized in software standby mode.

Bits 5 and 4—Interrupt Control Mode 1 and 0 (INTM1, INTM0): These bits select one of four interrupt control modes for the interrupt controller. The INTM1 bit must not be set to 1.

Bit 5	Bit 4	Interrupt	
INTM1	INTM0	Control Mode	Description
0	0	0	Interrupts are controlled by I bit (Initial value)
	1	1	Interrupts are controlled by I and UI bits and ICR
1	0	2	Cannot be used in this LSI
	1	3	Cannot be used in this LSI

Bit 2—NMI Edge Select (NMIEG): Selects the input edge for the NMI pin.

Bit 2

NMIEG	Description	
0	Interrupt request generated at falling edge of NMI input	(Initial value)
1	Interrupt request generated at rising edge of NMI input	

5.2.2 Interrupt Control Registers A to C (ICRA to ICRC)

Bit	7	6	5	4	3	2	1	0
	ICR7	ICR6	ICR5	ICR4	ICR3	ICR2	ICR1	ICR0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W							

The ICR registers are three 8-bit readable/writable registers that set the interrupt control level for interrupts other than NMI and address break.

The correspondence between ICR settings and interrupt sources is shown in table 5.3.

The ICR registers are initialized to H'00 by a reset and in hardware standby mode.

Bit n—Interrupt Control Level (ICRn): Sets the control level for the corresponding interrupt source.

Bit n

ICRn	 Description	
0	Corresponding interrupt source is control level 0 (non-priority)	(Initial value)
1	Corresponding interrupt source is control level 1 (priority)	

Note: n = 7 to 0

Table 5.3 Correspondence between Interrupt Sources and ICR Settings

				В	its			
Register	7	6	5	4	3	2	1	0
ICRA	IRQ0	IRQ1	IRQ2	IRQ4	IRQ6	DTC	-	Watchdog
			IRQ3	IRQ5	IRQ7		timer 0	timer 1
ICRB	A/D converter	Free- running timer	_	_	8-bit timer channel 0	8-bit timer channel 1	8-bit timer channels X, Y	HIF, Keyboard buffer controller
ICRC	SCI channel 0	SCI channel 1	SCI channel 2	IIC channel 0 (option)	IIC channel 1 (option)	_	_	_

RENESAS

5.2.3 IRO Enable Register (IER)

Bit	7	6	5	4	3	2	1	0
	IRQ7E	IRQ6E	IRQ5E	IRQ4E	IRQ3E	IRQ2E	IRQ1E	IRQ0E
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W							

IER is an 8-bit readable/writable register that controls enabling and disabling of interrupt requests IRQ7 to IRQ0.

IER is initialized to H'00 by a reset and in hardware standby mode.

Bits 7 to 0—IRQ7 to IRQ0 Enable (IRQ7E to IRQ0E): These bits select whether IRQ7 to IRQ0 are enabled or disabled.

Bit n

IRQnE	Description	
0	IRQn interrupt disabled	(Initial value)
1	IRQn interrupt enabled	

Note: n = 7 to 0

IRQ Sense Control Registers H and L (ISCRH, ISCRL) 5.2.4

ISCRH

Bit	15	14	13	12	11	10	9	8
	IRQ7SCB	IRQ7SCA	IRQ6SCB	IRQ6SCA	IRQ5SCB	IRQ5SCA	IRQ4SCB	IRQ4SCA
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W							

ISCRL

Bit	7	6	5	4	3	2	1	0
	IRQ3SCB	IRQ3SCA	IRQ2SCB	IRQ2SCA	IRQ1SCB	IRQ1SCA	IRQ0SCB	IRQ0SCA
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W							

ISCRH and ISCRL are 8-bit readable/writable registers that select rising edge, falling edge, or both edge detection, or level sensing, for the input at pins $\overline{IRQ7}$ to $\overline{IRQ0}$.

Each of the ISCR registers is initialized to H'00 by a reset and in hardware standby mode.

ISCRH Bits 7 to 0, ISCRL Bits 7 to 0: IRQ7 Sense Control A and B (IRQ7SCA, IRQ7SCB) to IRQ0 Sense Control A and B (IRQ0SCA, IRQ0SCB)

ISCRH Bits 7 to 0
ISCRL Bits 7 to 0

IRQ7SCB to IRQ0SCB	IRQ7SCA to IRQ0SCA	Description
0	0	Interrupt request generated at $\overline{IRQ7}$ to $\overline{IRQ0}$ input low level (Initial value)
	1	Interrupt request generated at falling edge of IRQ7 to IRQ0 input
1	0	Interrupt request generated at rising edge of IRQ7 to IRQ0 input
	1	Interrupt request generated at both falling and rising edges of IRQ7 to IRQ0 input

5.2.5 IRQ Status Register (ISR)

Bit	7	6	5	4	3	2	1	0
	IRQ7F	IRQ6F	IRQ5F	IRQ4F	IRQ3F	IRQ2F	IRQ1F	IRQ0F
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/(W)*							

Note: $\,^*\,$ Only 0 can be written, to clear the flag.

ISR is an 8-bit readable/writable register that indicates the status of IRQ7 to IRQ0 interrupt requests.

ISR is initialized to H'00 by a reset and in hardware standby mode.

Bits 7 to 0—IRQ7 to IRQ0 Flags (IRQ7F to IRQ0F): These bits indicate the status of IRQ7 to IRQ0 interrupt requests.

Bit n

IRQnF Description

0 [Clearing conditions]

(Initial value)

- Cleared by reading IRQnF when set to 1, then writing 0 in IRQnF
- When interrupt exception handling is executed while low-level detection is set (IRQnSCB = IRQnSCA = 0) and IRQn input is high*
- When IRQn interrupt exception handling is executed while falling, rising, or both-edge detection is set (IRQnSCB = 1 or IRQnSCA = 1)*
- 1 [Setting conditions]
 - When IRQn input goes low when low-level detection is set (IRQnSCB = IRQnSCA = 0)
 - When a falling edge occurs in IRQn input while falling edge detection is set (IRQnSCB = 0, IRQnSCA = 1)
 - When a rising edge occurs in IRQn input while rising edge detection is set (IRQnSCB = 1, IRQnSCA = 0)
 - When a falling or rising edge occurs in IRQn input while both-edge detection is set (IRQnSCB = IRQnSCA = 1)

Notes: n = 7 to 0

- * When a product, in which a DTC is incorporated, is used in the following settings, the corresponding flag bit is not automatically cleared even when exception handling, which is a clear condition, is executed and the bit is held at 1.
 - (1) When DTCEA3 is set to 1 (ADI is set to an interrupt source), IRQ4F flag is not automatically cleared.
 - (2) When DTCEA2 is set to 1 (ICIA is set to an interrupt source), IRQ5F flag is not automatically cleared.
 - (3) When DTCEA1 is set to 1 (ICIB is set to an interrupt source), IRQ6F flag is not automatically cleared.
 - (4) When DTCEA0 is set to 1 (OCIA is set to an interrupt source), IRQ7F flag is not automatically cleared.

When activation interrupt sources of DTC and IRQ interrupts are used with the above combinations, clear the interrupt flag by software in the interrupt handling routine of the corresponding IRQ.

5.2.6 Keyboard Matrix Interrupt Mask Register (KMIMR)

Bit	7	6	5	4	3	2	1	0
	KMIMR7	KMIMR6	KMIMR5	KMIMR4	KMIMR3	KMIMR2	KMIMR1	KMIMR0
Initial value	1	0	1	1	1	1	1	1
Read/Write	R/W							

KMIMR is an 8-bit readable/writable register that performs mask control for the keyboard matrix interrupt inputs (pins $\overline{\text{KIN7}}$ to $\overline{\text{KIN0}}$). To enable key-sense input interrupts from multiple pin inputs in keyboard matrix scanning/sensing, clear the corresponding mask bits to 0.

KMIMR is initialized to H'BF by a reset and in hardware standby mode and only $\overline{IRQ6}$ ($\overline{KIN6}$) input is enabled.

Bits 7 to 0—Keyboard Matrix Interrupt Mask (KMIMR7 to KMIMR0): These bits control key-sense input interrupt requests (KIN7 to KIN0).

Bits 7 to 0

KMIMR7 to	Description
0	Key-sense input interrupt requests enabled
1	Key-sense input interrupt requests disabled (Initial value)*
Note: *	However, the initial value of KMIMR6 is 0 because the KMIMR6 bit controls both IRQ6 interrupt request masking and key-sense input enabling.

5.2.7 Keyboard Matrix Interrupt Mask Register (KMIMRA)

Bit	7	6	5	4	3	2	1	0
	KMIMR15	KMIMR14	KMIMR13	KMIMR12	KMIMR11	KMIMR10	KMIMR9	KMIMR8
Initial value	1	1	1	1	1	1	1	1
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

KMIMRA is an 8-bit readable/writable register that performs mask control for the keyboard matrix interrupt inputs (pins $\overline{\text{KIN15}}$ to $\overline{\text{KIN8}}$). To enable key-sense input interrupts from multiple pin inputs in keyboard matrix scanning/sensing, clear the corresponding mask bits to 0.

KMIMRA is initialized to H'FF by a reset and in hardware standby mode.

Bits 7 to 0—Keyboard Matrix Interrupt Mask (KMIMR15 to KMIMR8): These bits control key-sense input interrupt requests (KIN15 to KIN8).

Bits 7 to 0

KMIMR15 to KMIMR8 Description

0	Key-sense input interrupt requests enabled
1	Key-sense input interrupt requests disabled (Initial value)

Figure 5.2 shows the relationship between interrupts IRQ7 and IRQ6, interrupts KIN15 to KIN0, and registers KMIMR and KMIMRA.

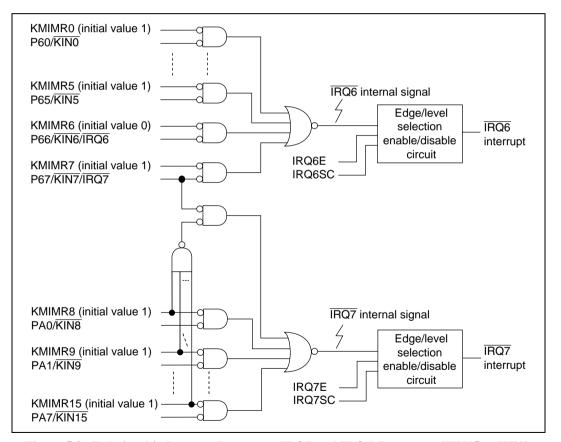


Figure 5.2 Relationship between Interrupts IRQ7 and IRQ6, Interrupts KIN15 to KIN0, and Registers KMIMR and KMIMRA

If any of bits KMIMR15 to KMIMR8 is cleared to 0, interrupt input from the IRO7 pin will be ignored. When pins $\overline{\text{KIN7}}$ to $\overline{\text{KIN0}}$ or $\overline{\text{KIN15}}$ to $\overline{\text{KIN8}}$ are used as key-sense interrupt input pins, either low-level sensing or falling-edge sensing must be designated as the interrupt sense condition for the corresponding interrupt source (IRO6 or IRO7).

5.2.8 Address Break Control Register (ABRKCR)

Bit	7	6	5	4	3	2	1	0
	CMF	_	_	_	_	_	_	BIE
Initial value	0	0	0	0	0	0	0	0
Read/Write	R	_	_	_	_	_	_	R/W

ABRKCR is an 8-bit readable/writable register that performs address break control.

ABRKCR is initialized to H'00 by a reset and in hardware standby mode. It is not initialized in software standby mode.

Bit 7—Condition Match Flag (CMF): This is the address break source flag, used to indicate that the address set by BAR has been prefetched. When the CMF flag and BIE flag are both set to 1, an address break is requested.

Bit 7

CMF	Description	
0	[Clearing condition]	
	When address break interrupt exception handling is executed	(Initial value)
1	[Setting condition]	
	When address set by BARA to BARC is prefetched while BIE = 1	

Bits 6 to 1—Reserved: These bits cannot be modified and are always read as 0.

Bit 0—Break Interrupt Enable (BIE): Selects address break enabling or disabling.

Bit 0

BIE	Description	
0	Address break disabled	(Initial value)
1	Address break enabled	



5.2.9 Break Address Registers A, B, C (BARA, BARB, BARC)

Bit	7	6	5	4	3	2	1	0
BARA	A23	A22	A21	A20	A19	A18	A17	A16
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W							
Bit	7	6	5	4	3	2	1	0
BARB	A15	A14	A13	A12	A11	A10	A9	A8
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W							
Bit	7	6	5	4	3	2	1	0
BARC	A7	A6	A5	A4	А3	A2	A1	_
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	_						

BAR consists of three 8-bit readable/writable registers (BARA, BARB, and BARC), and is used to specify the address at which an address break is to be executed.

Each of the BAR registers is initialized to H'00 by a reset and in hardware standby mode. They are not initialized in software standby mode.

BARA Bits 7 to 0—Address 23 to 16 (A23 to A16) BARB Bits 7 to 0—Address 15 to 8 (A15 to A8) BARC Bits 7 to 1—Address 7 to 1 (A7 to A1)

These bits specify the address at which an address break is to be executed. BAR bits A23 to A1 are compared with internal address bus lines A23 to A1, respectively.

The address at which the first instruction byte is located should be specified as the break address. Occurrence of the address break condition may not be recognized for other addresses.

In normal mode, no comparison is made with address lines A23 to A16.

BARC Bit 0—Reserved: This bit cannot be modified and is always read as 0.

5.3 Interrupt Sources

Interrupt sources comprise external interrupts (NMI and IRQ7 to IRQ0) and internal interrupts.

5.3.1 External Interrupts

There are nine external interrupt sources from 25 input pins (23 actual pins): NMI, $\overline{IRQ7}$ to $\overline{IRQ0}$, and $\overline{KIN15}$ to $\overline{KIN0}$. KIN15 to KIN8 share the IRQ7 interrupt source, and KIN7 to KIN0 share the IRQ6 interrupt source. Of these, NMI, IRQ7, IRQ6, and IRQ2 to IRQ0 can be used to restore the H8S/2148 Group or H8S/2144 Group chip from software standby mode.

NMI Interrupt

NMI is the highest-priority interrupt, and is always accepted by the CPU regardless of the interrupt control mode and the status of the CPU interrupt mask bits. The NMIEG bit in SYSCR can be used to select whether an interrupt is requested at a rising edge or a falling edge on the NMI pin.

The vector number for NMI interrupt exception handling is 7.

IRQ7 to **IRQ0** Interrupts

Interrupts IRQ7 to IRQ0 are requested by an input signal at pins $\overline{IRQ7}$ to $\overline{IRQ0}$. Interrupts IRQ7 to IRQ0 have the following features:

- Using ISCR, it is possible to select whether an interrupt is generated by a low level, falling edge, rising edge, or both edges, at pins $\overline{IRQ7}$ to $\overline{IRQ0}$.
- Enabling or disabling of interrupt requests IRQ7 to IRQ0 can be selected with IER.
- The interrupt control level can be set with ICR.
- The status of interrupt requests IRQ7 to IRQ0 is indicated in ISR. ISR flags can be cleared to 0
 by software.

A block diagram of interrupts IRQ7 to IRQ0 is shown in figure 5.3.



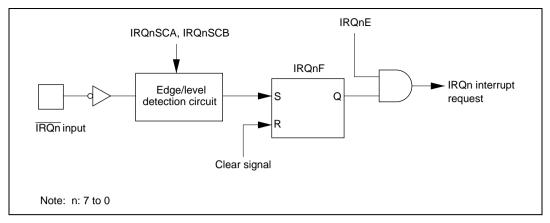


Figure 5.3 Block Diagram of Interrupts IRQ7 to IRQ0

Figure 5.4 shows the timing of IRQnF setting.

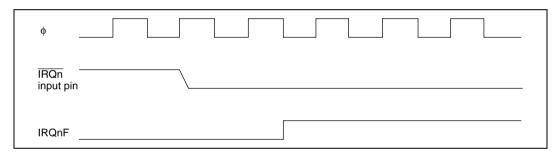


Figure 5.4 Timing of IRQnF Setting

The vector numbers for IRQ7 to IRQ0 interrupt exception handling are 23 to 16.

Detection of IRQ7 to IRQ0 interrupts does not depend on whether the relevant pin has been set for input or output. Therefore, when a pin is used as an external interrupt input pin, clear the corresponding DDR bit to 0 and do not use the pin as an I/O pin for another function. When the $\overline{\text{IRQ6}}$ pin is assigned as the IRQ6 interrupt input pin, then set the KMIMR6 bit to 0.

When the $\overline{IRQ7}$ pin is used as the IRQ7 interrupt input pin, bits KMIMR15 to KMIMR8 must all be set to 1. If any of these bits is cleared to 0, interrupt input from the $\overline{IRQ7}$ pin will be ignored.

As interrupt request flags IRQ7F to IRQ0F are set when the setting condition is met, regardless of the IER setting, only the necessary flags should be referenced.

Interrupts KIN15 to KIN0

Interrupts KIN15 to KIN0 are requested by input signals at pins KIN15 to KIN 0. When any of pins KIN15 to KIN0 are used as key-sense inputs, the corresponding KMIMR bits should be cleared to 0 to enable those key-sense input interrupts. The remaining unused key-sense input KMIMR bits should be set to 1 to disable those interrupts. Interrupts KIN15 to KIN8 correspond to the IRQ7 interrupt, and interrupts KIN7 to KIN0 correspond to the IRQ6 interrupt. Interrupt request generation pin conditions, interrupt request enabling, interrupt control level setting, and interrupt request status indications, are all in accordance with the IRQ7 and IRQ6 interrupt settings.

When pins $\overline{\text{KIN7}}$ to $\overline{\text{KIN0}}$ or $\overline{\text{KIN15}}$ to $\overline{\text{KIN8}}$ are used as key-sense interrupt input pins, either low-level sensing or falling-edge sensing must be designated as the interrupt sense condition for the corresponding interrupt source (IRQ6 or IRQ7).

5.3.2 Internal Interrupts

There are 43 sources for internal interrupts from on-chip supporting modules, plus one software interrupt source (address break).

- For each on-chip supporting module there are flags that indicate the interrupt request status, and enable bits that select enabling or disabling of these interrupts. If any one of these is set to 1, an interrupt request is issued to the interrupt controller.
- The interrupt control level can be set by means of ICR.
- The DTC can be activated by an FRT, TMR, SCI, or other interrupt request. When the DTC is activated by an interrupt, the interrupt control mode and interrupt mask bits have no effect.

5.3.3 Interrupt Exception Vector Table

Table 5.4 shows interrupt exception handling sources, vector addresses, and interrupt priorities. For default priorities, the lower the vector number, the higher the priority.

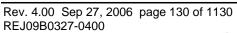
Priorities among modules can be set by means of ICR. The situation when two or more modules are set to the same priority, and priorities within a module, are fixed as shown in table 5.4.



Table 5.4 Interrupt Sources, Vector Addresses, and Interrupt Priorities

	Origin of		Vecto	or Address		
Interrupt Source	Interrupt Source	Vector Number	Normal Mode	Advanced Mode	ICR	Priority
NMI	External	7	H'000E	H'00001C		High
IRQ0	pin	16	H'0020	H'000040	ICRA7	_ 🛉
IRQ1		17	H'0022	H'000044	ICRA6	_
IRQ2		18	H'0024	H'000048	ICRA5	_
IRQ3		19	H'0026	H'00004C		
IRQ4		20	H'0028	H'000050	ICRA4	_
IRQ5		21	H'002A	H'000054		
IRQ6, KIN7 to KIN0		22	H'002C	H'000058	ICRA3	_
IRQ7, KIN15 to KIN8		23	H'002E	H'00005C		
SWDTEND (software activation interrupt end)	DTC	24	H'0030	H'000060	ICRA2	
WOVI0 (interval timer)	Watchdog timer 0	25	H'0032	H'000064	ICRA1	_
WOVI1 (interval timer)	Watchdog timer 1	26	H'0034	H'000068	ICRA0	
Address break (PC break)	_	27	H'0036	H'00006C		_
ADI (A/D conversion end)	A/D	28	H'0038	H'000070	ICRB7	_
Reserved	_	29	H'003A	H'000074		_
		to 47	to H'005E	to H'0000BC		_
ICIA (input capture A)	Free-running	48	H'0060	H'0000C0	ICRB6	
ICIB (input capture B)	timer	49	H'0062	H'0000C4		
ICIC (input capture C)		50	H'0064	H'0000C8		
ICID (input capture D)		51	H'0066	H'0000CC		
OCIA (output compare A)		52	H'0068	H'0000D0		
OCIB (output compare B)		53	H'006A	H'0000D4		
FOVI (overflow)		54	H'006C	H'0000D8		
Reserved		55	H'006E	H'0000DC		_
Reserved	_	56 to 63	H'0070 to H'007E	H'0000E0 to H'0000FC		Low

	Origin of	Vector Address				
Interrupt Source	Interrupt Source	Vector Number	Normal Mode	Advanced Mode	ICR	Priority
CMIA0 (compare-match A)	8-bit timer	64	H'0080	H'000100	ICRB3	High
CMIB0 (compare-match B)	channel 0	65	H'0082	H'000104		↑
OVI0 (overflow)		66	H'0084	H'000108		
Reserved		67	H'0086	H'00010C		
CMIA1 (compare-match A)	8-bit timer	68	H'0088	H'000110	ICRB2	_
CMIB1 (compare-match B)	channel 1	69	H'008A	H'000114		
OVI1 (overflow)		70	H'008C	H'000118		
Reserved		71	H'008E	H'00011C		
CMIAY (compare-match A)	8-bit timer	72	H'0090	H'000120	ICRB1	_
CMIBY (compare-match B)	channels Y, X	73	H'0092	H'000124		
OVIY (overflow)	Ι, Λ	74	H'0094	H'000128		
ICIX (input capture X)		75	H'0096	H'00012C		
IBF1 (IDR1 reception completed)	Host	76	H'0098	H'000130	ICRB0	_
IBF2 (IDR2 reception completed)	interface	77	H'009A	H'000134		
IBF3 (IDR3 reception completed)		78	H'009C	H'000138		
IBF4 (IDR4 reception completed)		79	H'009E	H'00013C		
ERI0 (receive error 0)	SCI	80	H'00A0	H'000140	ICRC7	_
RXI0 (reception completed 0)	channel 0	81	H'00A2	H'000144		
TXI0 (transmit data empty 0)		82	H'00A4	H'000148		
TEI0 (transmission end 0)		83	H'00A6	H'00014C		
ERI1 (receive error 1)	SCI	84	H'00A8	H'000150	ICRC6	_
RXI1 (reception completed 1)	channel 1	85	H'00AA	H'000154		
TXI1 (transmit data empty 1)		86	H'00AC	H'000158		
TEI1 (transmission end 1)		87	H'00AE	H'00015C		
ERI2 (receive error 2)	SCI	88	H'00B0	H'000160	ICRC5	_
RXI2 (reception completed 2)	channel 2	89	H'00B2	H'000164		
TXI2 (transmit data empty 2)		90	H'00B4	H'000168		
TEI2 (transmission end 2)		91	H'00B6	H'00016C		
IICI0 (1-byte transmission/ reception completed)	IIC channel 0 (option)	92	H'00B8	H'000170	ICRC4	
DDCSWI (format switch)		93	H'00BA	H'000174		Low





	Origin of		Vector Address			
Interrupt Source	Interrupt Source	Vector Number	Normal Mode	Advanced Mode	ICR	Priority
IICI1 (1-byte transmission/ reception completed)	IIC channel 1 (option)	94	H'00BC	H'000178	ICRC3	High ↑
Reserved		95	H'00BE	H'00017C		
PS2IA (reception completed A)	Keyboard	96	H'00C0	H'000180	ICRB0	_
PS2IB (reception completed B)	buffer controller (PS2)	97	H'00C2	H'000184		
PS2IC (reception completed C)		98	H'00C4	H'000188		
Reserved	,	99	H'00C6	H'00018C		
Reserved	_	100 to 103	H'00C8 to H'00CE	H'000190 to H'00019C		Low

5.4 Address Breaks

5.4.1 Features

With this LSI, it is possible to identify the prefetch of a specific address by the CPU and generate an address break interrupt, using the ABRKCR and BAR registers. When an address break interrupt is generated, address break interrupt exception handling is executed.

This function can be used to detect the beginning of execution of a bug location in the program, and branch to a correction routine.

5.4.2 Block Diagram

A block diagram of the address break function is shown in figure 5.5.

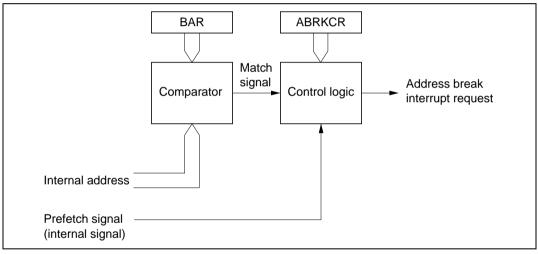


Figure 5.5 Block Diagram of Address Break Function



5.4.3 Operation

ABRKCR and BAR settings can be made so that an address break interrupt is generated when the CPU prefetches the address set in BAR. This address break function issues an interrupt request to the interrupt controller when the address is prefetched, and the interrupt controller determines the interrupt priority. When the interrupt is accepted, interrupt exception handling is started on completion of the currently executing instruction. With an address break interrupt, interrupt mask control by the I and UI bits in the CPU's CCR is ineffective.

The register settings when the address break function is used are as follows.

- 1. Set the break address in bits A23 to A1 in BAR.
- 2. Set the BIE bit in ABRKCR to 1 to enable address breaks. An address break will not be requested if the BIE bit is cleared to 0.

When the setting condition occurs, the CMF flag in ABRKCR is set to 1 and an interrupt is requested. If necessary, the source should be identified in the interrupt handling routine.

5.4.4 Usage Notes

- With the address break function, the address at which the first instruction byte is located should be specified as the break address. Occurrence of the address break condition may not be recognized for other addresses.
- In normal mode, no comparison is made with address lines A23 to A16.
- If a branch instruction (Bcc, BSR), jump instruction (JMP, JSR), RTS instruction, or RTE instruction is located immediately before the address set in BAR, execution of this instruction will output a prefetch signal for that address, and an address break may be requested. This can be prevented by not making a break address setting for an address immediately following one of these instructions, or by determining within the interrupt handling routine whether interrupt handling was initiated by a genuine condition occurrence.
- As an address break interrupt is generated by a combination of the internal prefetch signal and
 address, the timing of the start of interrupt exception handling depends on the content and
 execution cycle of the instruction at the set address and the preceding instruction. Figure 5.6
 shows some address timing examples.

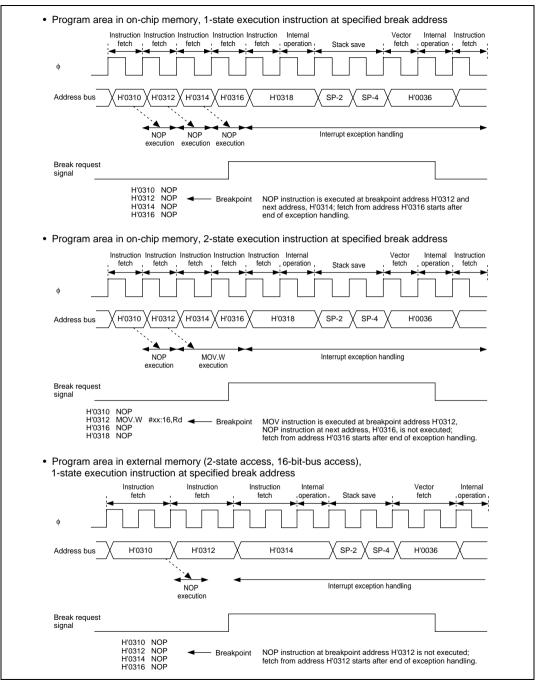


Figure 5.6 Examples of Address Break Timing

5.5 Interrupt Operation

5.5.1 Interrupt Control Modes and Interrupt Operation

Interrupt operations in this LSI differ depending on the interrupt control mode.

NMI and address break interrupts are accepted at all times except in the reset state and the hardware standby state. In the case of IRQ interrupts and on-chip supporting module interrupts, an enable bit is provided for each interrupt. Clearing an enable bit to 0 disables the corresponding interrupt request. Interrupt sources for which the enable bits are set to 1 are controlled by the interrupt controller.

Table 5.5 shows the interrupt control modes.

The interrupt controller performs interrupt control according to the interrupt control mode set by the INTM1 and INTM0 bits in SYSCR, the priorities set in ICR, and the masking state indicated by the I and UI bits in the CPU's CCR.

Table 5.5 Interrupt Control Modes

Interrupt	SYSCR		Priority Setting	Interrupt	
Control Mode	INTM1	INTM0	Register	Mask Bits	Description
0	0	0	ICR	I	Interrupt mask control is performed by the I bit
					Priority can be set with ICR
1	_	1	ICR	I, UI	3-level interrupt mask control is performed by the I and UI bits
					Priority can be set with ICR

Figure 5.7 shows a block diagram of the priority decision circuit.

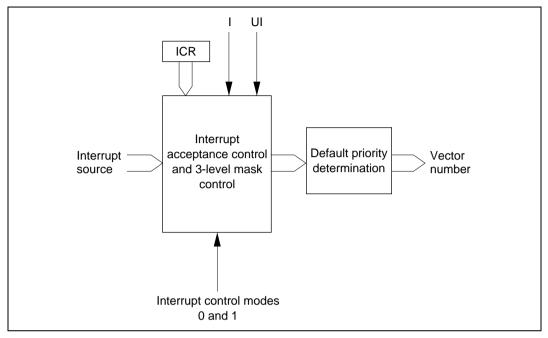


Figure 5.7 Block Diagram of Interrupt Control Operation

Interrupt Acceptance Control and 3-Level Control

In interrupt control modes 0 and 1, interrupt acceptance control and 3-level mask control is performed by means of the I and UI bits in CCR, and ICR (control level).

Table 5.6 shows the interrupts selected in each interrupt control mode.

Table 5.6 Interrupts Selected in Each Interrupt Control Mode

Interr		upt Mask Bits		
Interrupt Control Mode	I UI		Selected Interrupts	
0	0	*	All interrupts (control level 1 has priority)	
	1	*	NMI and address break interrupts	
1	0	*	All interrupts (control level 1 has priority)	
	1	0	NMI, address break and control level 1 interrupts	
		1	NMI, and address break interrupts	

Legend:

*: Don't care

Default Priority Determination

The priority is determined for the selected interrupt, and a vector number is generated.

If the same value is set for ICR, acceptance of multiple interrupts is enabled, and so only the interrupt source with the highest priority according to the preset default priorities is selected and has a vector number generated.

Interrupt sources with a lower priority than the accepted interrupt source are held pending.

Table 5.7 shows operations and control signal functions in each interrupt control mode.

Table 5.7 Operations and Control Signal Functions in Each Interrupt Control Mode

Interrupt Control Mode	Set	ting	Interrupt Acceptance Control 3-Level Control			Default Priority Determination	T (Trace)	
Control wode	INTM1	INTM0		ı	UI	ICR	Determination (Trace	
0	0	0	0	IM	_	PR	0	_
1	0	1	0	IM	IM	PR	0	_

Legend:

O: Interrupt operation control performed

IM: Used as interrupt mask bit

PR: Sets priority

—: Not used

5.5.2 **Interrupt Control Mode 0**

Enabling and disabling of IRQ interrupts and on-chip supporting module interrupts can be set by means of the I bit in the CPU's CCR, and ICR. Interrupts are enabled when the I bit is cleared to 0, and disabled when set to 1. Control level 1 interrupt sources have higher priority.

Figure 5.8 shows a flowchart of the interrupt acceptance operation in this case.

- 1. If an interrupt source occurs when the corresponding interrupt enable bit is set to 1, an interrupt request is sent to the interrupt controller.
- 2. When interrupt requests are sent to the interrupt controller, a control level 1 interrupt, according to the control level set in ICR, has priority for selection, and other interrupt requests are held pending. If a number of interrupt requests with the same control level setting are generated at the same time, the interrupt request with the highest priority according to the priority system shown in table 5.4 is selected.
- 3. The I bit is then referenced. If the I bit is cleared to 0, the interrupt request is accepted. If the I bit is set to 1, only NMI and address break interrupt are accepted, and other interrupt requests are held pending.
- 4. When an interrupt request is accepted, interrupt exception handling starts after execution of the current instruction has been completed.
- 5. The PC and CCR are saved to the stack area by interrupt exception handling. The PC saved on the stack shows the address of the first instruction to be executed after returning from the interrupt handling routine.
- 6. Next, the I bit in CCR is set to 1. This disables all interrupts except NMI and address break.
- 7. A vector address is generated for the accepted interrupt, and execution of the interrupt handling routine starts at the address indicated by the contents of that vector address.



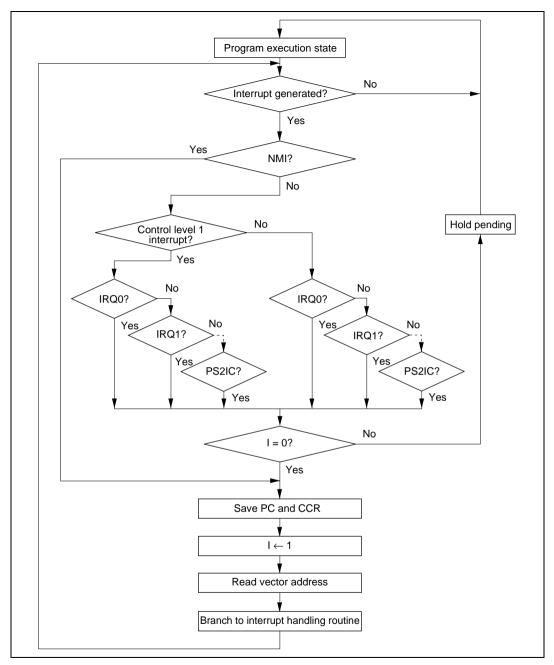


Figure 5.8 Flowchart of Procedure Up to Interrupt Acceptance in Interrupt Control Mode 0

5.5.3 **Interrupt Control Mode 1**

Three-level masking is implemented for IRQ interrupts and on-chip supporting module interrupts by means of the I and UI bits in the CPU's CCR, and ICR.

- Control level 0 interrupt requests are enabled when the I bit is cleared to 0, and disabled when set to 1.
- Control level 1 interrupt requests are enabled when the I bit or UI bit is cleared to 0, and disabled when both the I bit and the UI bit are set to 1.

For example, if the interrupt enable bit for an interrupt request is set to 1, and H'20, H'00, and H'00 are set in ICRA, ICRB, and ICRC, respectively, (i.e. IRQ2 and IRQ3 interrupts are set to control level 1 and other interrupts to control level 0), the situation is as follows:

- When I = 0, all interrupts are enabled (Priority order: NMI > IRO2 > IRO3 > address break > IRO0 > IRO1 ...)
- When I = 1 and UI = 0, only NMI, IRO2, IRO3 and address break interrupts are enabled
- When I = 1 and UI = 1, only NMI and address break interrupts are enabled

Figure 5.9 shows the state transitions in these cases.

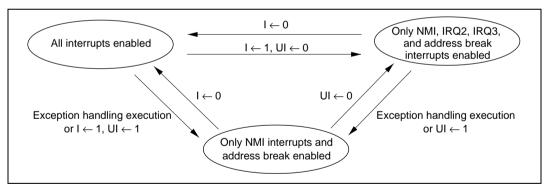


Figure 5.9 Example of State Transitions in Interrupt Control Mode 1

Figure 5.10 shows a flowchart of the interrupt acceptance operation in this case.

- 1. If an interrupt source occurs when the corresponding interrupt enable bit is set to 1, an interrupt request is sent to the interrupt controller.
- 2. When interrupt requests are sent to the interrupt controller, a control level 1 interrupt, according to the control level set in ICR, has priority for selection, and other interrupt requests are held pending. If a number of interrupt requests with the same control level setting are generated at the same time, the interrupt request with the highest priority according to the priority system shown in table 5.4 is selected.
- 3. The I bit is then referenced. If the I bit is cleared to 0, the UI bit has no effect.
 - An interrupt request set to interrupt control level 0 is accepted when the I bit is cleared to 0. If the I bit is set to 1, only an NMI and address break interrupts are accepted, and other interrupt requests are held pending.
 - An interrupt request set to interrupt control level 1 has priority over an interrupt request set to interrupt control level 0, and is accepted if the I bit is cleared to 0, or if the I bit is set to 1 and the UI bit is cleared to 0.
 - When both the I bit and the UI bit are set to 1, only an NMI and address break interrupts are accepted, and other interrupt requests are held pending.
- 4. When an interrupt request is accepted, interrupt exception handling starts after execution of the current instruction has been completed.
- 5. The PC and CCR are saved to the stack area by interrupt exception handling. The PC saved on the stack shows the address of the first instruction to be executed after returning from the interrupt handling routine.
- 6. Next, the I and UI bits in CCR are set to 1. This disables all interrupts except NMI and address break.
- 7. A vector address is generated for the accepted interrupt, and execution of the interrupt handling routine starts at the address indicated by the contents of that vector address.

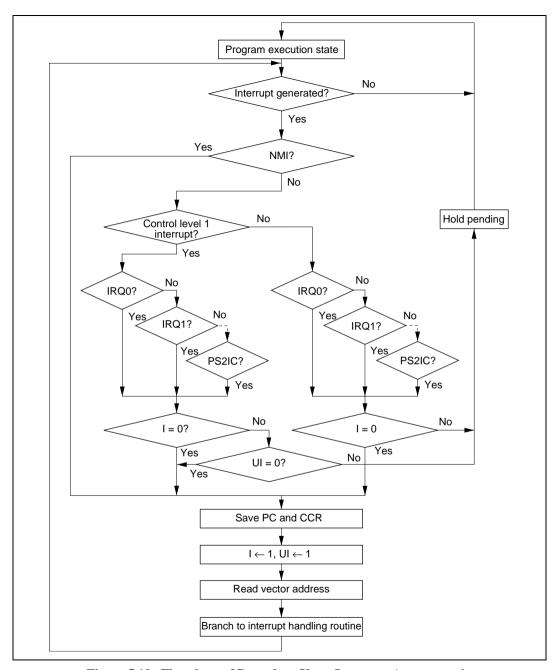


Figure 5.10 Flowchart of Procedure Up to Interrupt Acceptance in Interrupt Control Mode 1

Interrupt Exception Handling Sequence 5.5.4

Figure 5.11 shows the interrupt exception handling sequence. The example shown is for the case where interrupt control mode 0 is set in advanced mode, and the program area and stack area are in on-chip memory.

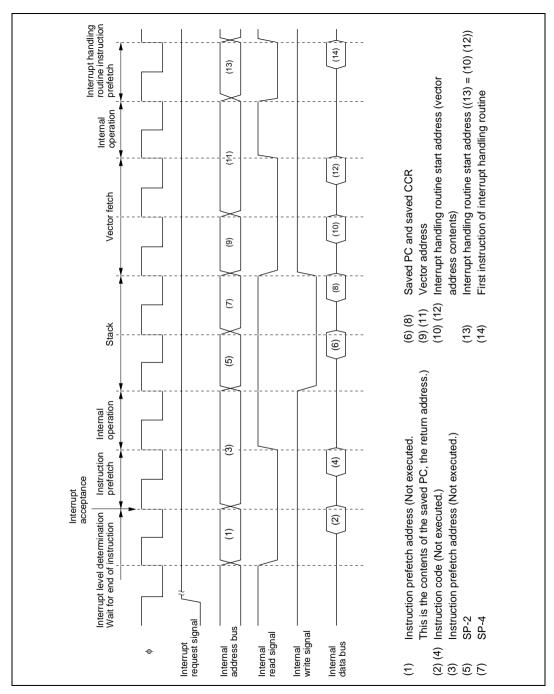


Figure 5.11 Interrupt Exception Handling

5.5.5 Interrupt Response Times

This LSI are capable of fast word access to on-chip memory, and high-speed processing can be achieved by providing the program area in on-chip ROM and the stack area in on-chip RAM.

Table 5.8 shows interrupt response times—the interval between generation of an interrupt request and execution of the first instruction in the interrupt handling routine. The symbols used in table 5.8 are explained in table 5.9.

Table 5.8 Interrupt Response Times

		Number of States				
No.	Item	Normal Mode	Advanced Mode			
1	Interrupt priority determination*1	3	3			
2	Number of wait states until executing instruction ends*2	1 to 19+2•S _i	1 to 19+2•S ₁			
3	PC, CCR stack save	2•S _K	2•S _K			
4	Vector fetch	Sı	2•S ₁			
5	Instruction fetch*3	2 - S ₁	2•S ₁			
6	Internal processing*4	2	2			
Total	(using on-chip memory)	11 to 31	12 to 32			

Notes: 1. Two states in case of internal interrupt.

- 2. Refers to MULXS and DIVXS instructions.
- 3. Prefetch after interrupt acceptance and interrupt handling routine prefetch.
- 4. Internal processing after interrupt acceptance and internal processing after vector fetch.

Table 5.9 Number of States in Interrupt Handling Routine Execution

				Object	of Access	
		•		Exteri	nal Device	
			8-1	Bit Bus	16-	Bit Bus
	Symbol	Internal Memory	2-State Access	3-State Access	2-State Access	3-State Access
Instruction fetch	Sı	1	4	6+2m	2	3+m
Branch address read	S _J	_				
Stack manipulation	S _K	_				
Logond:						

Legend:

m: Number of wait states in an external device access

5.6 Usage Notes

5.6.1 Contention between Interrupt Generation and Disabling

When an interrupt enable bit is cleared to 0 to disable interrupts, the disabling becomes effective after execution of the instruction.

In other words, when an interrupt enable bit is cleared to 0 by an instruction such as BCLR or MOV, if an interrupt is generated during execution of the instruction, the interrupt concerned will still be enabled on completion of the instruction, and so interrupt exception handling for that interrupt will be executed on completion of the instruction. However, if there is an interrupt request of higher priority than that interrupt, interrupt exception handling will be executed for the higher-priority interrupt, and the lower-priority interrupt will be ignored.

The same also applies when an interrupt source flag is cleared to 0.

Figure 5.12 shows an example in which the CMIEA bit in 8-bit timer register TCR is cleared to 0.

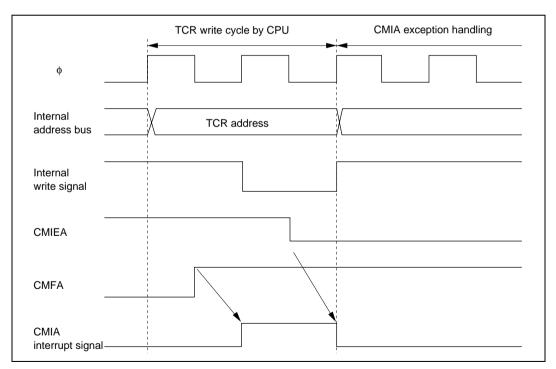


Figure 5.12 Contention between Interrupt Generation and Disabling

The above contention will not occur if an enable bit or interrupt source flag is cleared to 0 while the interrupt is masked.

5.6.2 **Instructions That Disable Interrupts**

Instructions that disable interrupts are LDC, ANDC, ORC, and XORC. After any of these instructions is executed, all interrupts except NMI are disabled and the next instruction is always executed. When the I bit or UI bit is set by one of these instructions, the new value becomes valid two states after execution of the instruction ends.

5.6.3 **Interrupts during Execution of EEPMOV Instruction**

Interrupt operation differs between the EEPMOV.B instruction and the EEPMOV.W instruction.

With the EEPMOV.B instruction, an interrupt request (including NMI) issued during the transfer is not accepted until the move is completed.

With the EEPMOV.W instruction, if an interrupt request is issued during the transfer, interrupt exception handling starts at a break in the transfer cycle. The PC value saved on the stack in this case is the address of the next instruction.

Therefore, if an interrupt is generated during execution of an EEPMOV.W instruction, the following coding should be used.

L1: EEPMOV.W

> MOV.W R4,R4

> > L1

BNE

5.7 DTC Activation by Interrupt

5.7.1 Overview

The DTC can be activated by an interrupt. In this case, the following options are available:

- Interrupt request to CPU
- Activation request to DTC
- Both of the above

For details of interrupt requests that can be used to activate the DTC, see section 7, Data Transfer Controller (DTC).

5.7.2 Block Diagram

Figure 5.13 shows a block diagram of the DTC and interrupt controller.

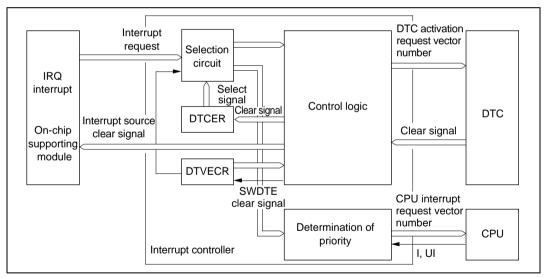


Figure 5.13 Interrupt Control for DTC

5.7.3 Operation

The interrupt controller has three main functions in DTC control.

Selection of Interrupt Source: It is possible to select DTC activation request or CPU interrupt request with the DTCE bit of DTCERA to DTCERE in the DTC.

After a DTC data transfer, the DTCE bit can be cleared to 0 and an interrupt request sent to the CPU in accordance with the specification of the DISEL bit of MRB in the DTC.

When the DTC performs the specified number of data transfers and the transfer counter reaches 0, following the DTC data transfer the DTCE bit is cleared to 0 and an interrupt request is sent to the CPU.

Determination of Priority: The DTC activation source is selected in accordance with the default priority order, and is not affected by mask or priority levels. See section 7.3.3, DTC Vector Table, for the respective priorities.

Operation Order: If the same interrupt is selected as a DTC activation source and a CPU interrupt source, the DTC data transfer is performed first, followed by CPU interrupt exception handling.

Table 5.10 summarizes interrupt source selection and interrupt source clearance control according to the settings of the DTCE bit of DTCERA to DTCERE in the DTC and the DISEL bit of MRB in the DTC.

Table 5.10 Interrupt Source Selection and Clearing Control

Sottings

DTC		Interrupt Source Selection/Clearing Contro			
DTCE	DISEL	DTC	CPU		
0	*	×	Δ		
1	0	Δ	×		
	1	0	Δ		

Legend:

- Δ : The relevant interrupt is used. Interrupt source clearing is performed. (The CPU should clear the source flag in the interrupt handling routine.)
- O: The relevant interrupt is used. The interrupt source is not cleared.
- x: The relevant bit cannot be used.
- *: Don't care

Usage Note

SCI, IIC, and A/D converter interrupt sources are cleared when the DTC reads or writes to the prescribed register, and are not dependent upon the DISEL bit.

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Section 6 Bus Controller

6.1 Overview

This LSI have a built-in bus controller (BSC) that allows external address space bus specifications, such as bus width and number of access states, to be set.

The bus controller also has a bus arbitration function, and controls the operation of the internal bus masters: the CPU and data transfer controller (DTC).

6.1.1 Features

The features of the bus controller are listed below.

- Basic bus interface
 - 2-state access or 3-state access can be selected
 - Program wait states can be inserted
- Burst ROM interface
 - External space can be designated as ROM interface space
 - 1-state or 2-state burst access can be selected
- Idle cycle insertion
 - An idle cycle can be inserted when an external write cycle immediately follows an external read cycle
- Bus arbitration function
 - Includes a bus arbiter that arbitrates bus mastership between the CPU and DTC

6.1.2 Block Diagram

Figure 6.1 shows a block diagram of the bus controller.

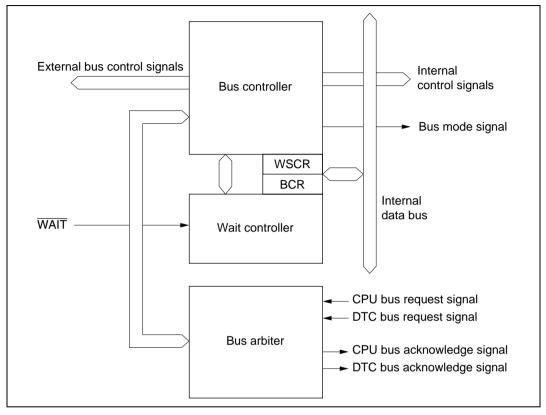


Figure 6.1 Block Diagram of Bus Controller

6.1.3 Pin Configuration

Table 6.1 summarizes the pins of the bus controller.

Table 6.1 Bus Controller Pins

Name	Symbol	I/O	Function
Address strobe	ĀS	Output	Strobe signal indicating that address output on address bus is enabled (when IOSE bit is 0)
I/O select	ĪŌS	Output	I/O select signal (when IOSE bit is 1)
Read	RD	Output	Strobe signal indicating that external space is being read
High write	HWR	Output	Strobe signal indicating that external space is being written to, and that the upper data bus (D15 to D8) is enabled
Low write	LWR	Output	Strobe signal indicating that external space is being written to, and that the lower data bus (D7 to D0) is enabled
Wait	WAIT	Input	Wait request signal when external 3-state access space is accessed

6.1.4 Register Configuration

Table 6.2 summarizes the registers of the bus controller.

Table 6.2 Bus Controller Registers

Name	Abbreviation	R/W	Initial Value	Address*
Bus control register	BCR	R/W	H'D7	H'FFC6
Wait state control register	WSCR	R/W	H'33	H'FFC7

Note: * Lower 16 bits of the address.

6.2 Register Descriptions

6.2.1 Bus Control Register (BCR)

Bit	7	6	5	4	3	2	1	0
	ICIS1	ICIS0	BRSTRM	BRSTS1	BRSTS0	_	IOS1	IOS0
Initial value	1	1	0	1	0	1	1	1
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

BCR is an 8-bit readable/writable register that specifies the external memory space access mode, and the extent of the I/O area when the I/O strobe function has been selected for the \overline{AS} pin.

BCR is initialized to H'D7 by a reset and in hardware standby mode. It is not initialized in software standby mode.

Bit 7—Idle Cycle Insert 1 (ICIS1): Reserved. Do not write 0 to this bit.

Bit 6—Idle Cycle Insert 0 (ICIS0): Selects whether or not a one-state idle cycle is to be inserted between bus cycles when successive external read and external write cycles are performed.

Bit 6

ICIS0	Description
0	Idle cycle not inserted in case of successive external read and external write cycles
1	Idle cycle inserted in case of successive external read and external write cycles (Initial value)

Bit 5—Burst ROM Enable (BRSTRM): Selects whether external space is designated as a burst ROM interface space. The selection applies to the entire external space.

Bit 5

BRSTRM	Description	
0	Basic bus interface	(Initial value)
1	Burst ROM interface	

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Bit 4—Burst Cycle Select 1 (BRSTS1): Selects the number of burst cycles for the burst ROM interface.

Bit 4

BRSTS1	Description	
0	Burst cycle comprises 1 state	
1	Burst cycle comprises 2 states	(Initial value)

Bit 3—Burst Cycle Select 0 (BRSTS0): Selects the number of words that can be accessed in a burst ROM interface burst access.

Bit 3

BRSTS0	Description	
0	Max. 4 words in burst access	(Initial value)
1	Max. 8 words in burst access	

Bit 2—Reserved: Do not write 0 to this bit.

Bits 1 and 0—IOS Select 1 and 0 (IOS1, IOS0): See table 6.4.

6.2.2 Wait State Control Register (WSCR)

Bit	7	6	5	4	3	2	1	0
	RAMS	RAM0	ABW	AST	WMS1	WMS0	WC1	WC0
Initial value	0	0	1	1	0	0	1	1
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

WSCR is an 8-bit readable/writable register that specifies the data bus width, number of access states, wait mode, and number of wait states for external memory space. The on-chip memory and internal I/O register bus width and number of access states are fixed, irrespective of the WSCR settings.

WSCR is initialized to H'33 by a reset and in hardware standby mode. It is not initialized in software standby mode.

Bit 7—RAM Select (RAMS)/Bit 6—RAM Area Setting (RAM0): Reserved bits. Always write 0 when writing to these bits in the A-mask version.

Bit 5—Bus Width Control (ABW): Specifies whether the external memory space is 8-bit access space or 16-bit access space.

Bit 5

ABW		
0	External memory space is designated as 16-bit access space	
1	External memory space is designated as 8-bit access space	(Initial value)

Bit 4—Access State Control (AST): Specifies whether the external memory space is 2-state access space or 3-state access space, and simultaneously enables or disables wait state insertion.

Bit 4

AST	Description	
0	External memory space is designated as 2-state access space Wait state insertion in external memory space accesses is disabled	
1	External memory space is designated as 3-state access space Wait state insertion in external memory space accesses is enabled	(Initial value)

Bits 3 and 2—Wait Mode Select 1 and 0 (WMS1, WMS0): These bits select the wait mode when external memory space is accessed while the AST bit is set to 1.

Bit 3	Bit 2		
WMS1	WMS0	Description	
0	0	Program wait mode	(Initial value)
	1	Wait-disabled mode	
1	0	Pin wait mode	
	1	Pin auto-wait mode	

Bits 1 and 0—Wait Count 1 and 0 (WC1, WC0): These bits select the number of program wait states when external memory space is accessed while the AST bit is set to 1.

Bit 1	Bit 0	
WC1	WC0	Description
0	0	No program wait states are inserted
	1	1 program wait state is inserted in external memory space accesses
1	0	2 program wait states are inserted in external memory space accesses
	1	3 program wait states are inserted in external memory space accesses (Initial value)

6.3 Overview of Bus Control

6.3.1 Bus Specifications

The external space bus specifications consist of three elements: bus width, number of access states, and wait mode and number of program wait states.

The bus width and number of access states for on-chip memory and internal I/O registers are fixed, and are not affected by the bus controller.

Bus Width: A bus width of 8 or 16 bits can be selected with the ABW bit.

Number of Access States: Two or three access states can be selected with the AST bit. When 2-state access space is designated, wait insertion is disabled. The number of access states on the burst ROM interface is determined without regard to the AST bit setting.

Wait Mode and Number of Program Wait States: When 3-state access space is designated by the AST bit, the wait mode and the number of program wait states to be inserted automatically is selected with WMS1, WMS0, WC1, and WC0. From 0 to 3 program wait states can be selected.

Table 6.3 shows the bus specifications for each basic bus interface area.

Table 6.3 Bus Specifications for Each Area (Basic Bus Interface)

				WC1 WC0		bus specific	ations (basi	ions (basic bus interrace)	
ABW	AST	WMS1	WMS0		WC0	Bus Width	Access States	Program Wait States	
0	0	_	_	_	_	16	2	0	
1	1	0	1	_	_	16	3	0	
		*	*	0	0		3	0	
					1	_		1	
				1	0			2	
					1			3	
1	0	_	_	_	_	8	2	0	
	1	0	1	_	_	8	3	0	
		*	*	0	0	<u>—</u>	3	0	
					1			1	
				1	0			2	
					1	<u> </u>		3	

Rus Specifications (Rasic Rus Interface)

Note: * Except when WMS1 = 0 and WMS0 = 1

6.3.2 Advanced Mode

The initial state of the external space is basic bus interface, three-state access space. In ROM-enabled expanded mode, the space excluding the on-chip ROM, on-chip RAM, and internal I/O registers is external space. The on-chip RAM is enabled when the RAME bit in the system control register (SYSCR) is set to 1; when the RAME bit is cleared to 0, the on-chip RAM is disabled and the corresponding space becomes external space.

6.3.3 Normal Mode

The initial state of the external memory space is basic bus interface, three-state access space. In ROM-disabled expanded mode, the space excluding the on-chip RAM and internal I/O registers is external space. In ROM-enabled expanded mode, the space excluding the on-chip ROM, on-chip RAM, and internal I/O registers is external space. The on-chip RAM is enabled when the RAME bit in the system control register (SYSCR) is set to 1; when the RAME bit is cleared to 0, the on-chip RAM is disabled and the corresponding space becomes external space.

6.3.4 I/O Select Signal

In this LSI, an I/O select signal ($\overline{\text{IOS}}$) can be output, with the signal output going low when the designated external space is accessed.

Figure 6.2 shows an example of $\overline{\text{IOS}}$ signal output timing.

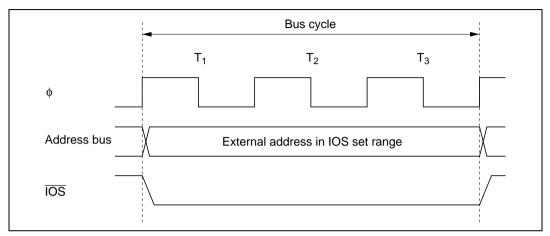


Figure 6.2 **IOS** Signal Output Timing

Enabling or disabling of \overline{IOS} signal output is controlled by the setting of the IOSE bit in SYSCR. In expanded mode, this pin operates as the \overline{AS} output pin after a reset, and therefore the IOSE bit in SYSCR must be set to 1 in order to use this pin as the \overline{IOS} signal output. See section 8, I/O Ports, for details.

The range of addresses for which the \overline{IOS} signal is output can be set with bits IOS1 and IOS0 in BCR. The \overline{IOS} signal address ranges are shown in table 6.4.

 Table 6.4
 IOS Signal Output Range Settings

IOS1	IOS0	IOS Signal Output Range	
0	0	H'(FF)F000 to H'(FF)F03F	_
	1	H'(FF)F000 to H'(FF)F0FF	
1	0	H'(FF)F000 to H'(FF)F3FF	
	1	H'(FF)F000 to H'(FF)FE4F*	(Initial value)

Note: * In the H8S/2148 and H8S/2147 F-ZTAT A-mask version, the address range is from H'(FF)F000 to H'(FF)F7FF.

6.4 **Basic Bus Interface**

6.4.1 Overview

The basic bus interface enables direct connection of ROM, SRAM, and so on.

The bus specifications can be selected with the ABW bit, the AST bit, and the WMS1, WMS0, WC1, and WC0 bits (see table 6.3).

6.4.2 **Data Size and Data Alignment**

Data sizes for the CPU and other internal bus masters are byte, word, and longword. The bus controller has a data alignment function, and when accessing external space, controls whether the upper data bus (D15 to D8) or lower data bus (D7 to D0) is used according to the bus specifications for the area being accessed (8-bit access space or 16-bit access space) and the data size.

8-Bit Access Space

Figure 6.3 illustrates data alignment control for the 8-bit access space. With the 8-bit access space, the upper data bus (D15 to D8) is always used for accesses. The amount of data that can be accessed at one time is one byte: a word access is performed as two byte accesses, and a longword access, as four byte accesses.

		Upper da _L D15	nta bus Lower o	data bus
Byte size				
Word size	1st bus cycle 2nd bus cycle			
Longword size	1st bus cycle 2nd bus cycle 3rd bus cycle 4th bus cycle			

Figure 6.3 Access Sizes and Data Alignment Control (8-Bit Access Space)

16-Bit Access Space

Figure 6.4 illustrates data alignment control for the 16-bit access space. With the 16-bit access space, the upper data bus (D15 to D8) and lower data bus (D7 to D0) are used for accesses. The amount of data that can be accessed at one time is one byte or one word, and a longword access is executed as two word accesses.

In byte access, whether the upper or lower data bus is used is determined by whether the address is even or odd. The upper data bus is used for an even address, and the lower data bus for an odd address.

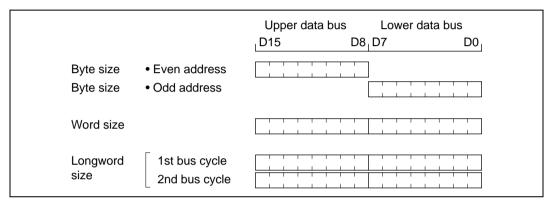


Figure 6.4 Access Sizes and Data Alignment Control (16-Bit Access Space)

6.4.3 Valid Strobes

Table 6.5 shows the data buses used and valid strobes for the access spaces.

In a read, the \overline{RD} signal is valid without discrimination between the upper and lower halves of the data bus.

In a write, the \overline{HWR} signal is valid for the upper half of the data bus, and the \overline{LWR} signal for the lower half.

Table 6.5 Data Buses Used and Valid Strobes

Area	Access Size	Read/ Write	Address	Valid Strobe	Upper Data Bus (D15 to D8)	Lower Data Bus (D7 to D0)
8-bit access	Byte	Read	_	RD	Valid	Port, etc.
space		Write	_	HWR	_	Port, etc.
16-bit access	Byte	Read	Even	RD	Valid	Invalid
space			Odd	_	Invalid	Valid
		Write	Even	HWR	Valid	Undefined
			Odd	LWR	Undefined	Valid
	Word	Read	_	RD	Valid	Valid
		Write	_	HWR, LWR	Valid	Valid

Legend:

Undefined: Undefined data is output.

Invalid: Input state; input value is ignored.

Pins are used as port or on-chip supporting module input/output pins, and not as data Port, etc.:

bus pins.



6.4.4 Basic Timing

8-Bit 2-State Access Space

Figure 6.5 shows the bus timing for an 8-bit 2-state access space. When an 8-bit access space is accessed, the upper half (D15 to D8) of the data bus is used.

Wait states cannot be inserted.

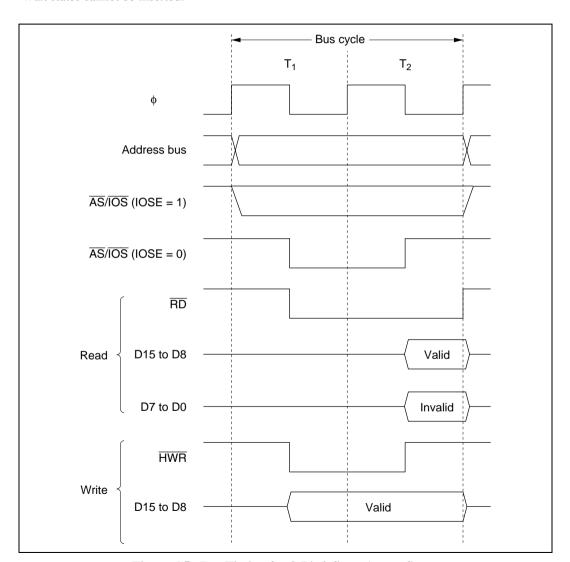


Figure 6.5 Bus Timing for 8-Bit 2-State Access Space

8-Bit 3-State Access Space

Figure 6.6 shows the bus timing for an 8-bit 3-state access space. When an 8-bit access space is accessed, the upper half (D15 to D8) of the data bus is used.

Wait states can be inserted.

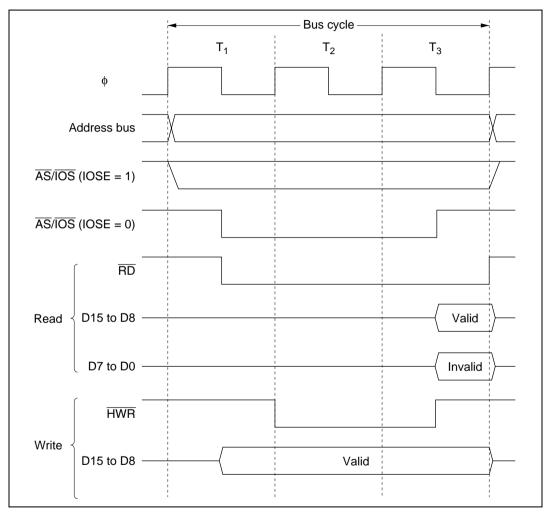


Figure 6.6 Bus Timing for 8-Bit 3-State Access Space

16-Bit, 2-State Access Space

Figures 6.7 to 6.9 show the bus timing for 16-bit, 2-state access space. When 16-bit access space is accessed, the upper data bus (D15 to D8) is used for even addresses and the lower data bus (D7 to D0) for odd addresses.

Wait states cannot be inserted.

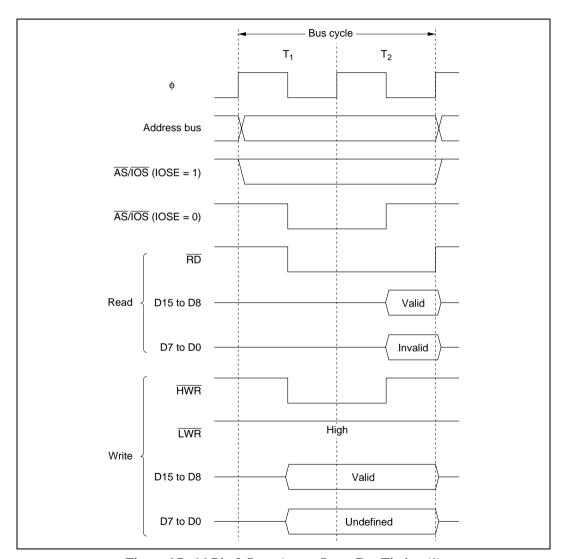


Figure 6.7 16-Bit, 2-State Access Space Bus Timing (1) (Even Address Byte Access)

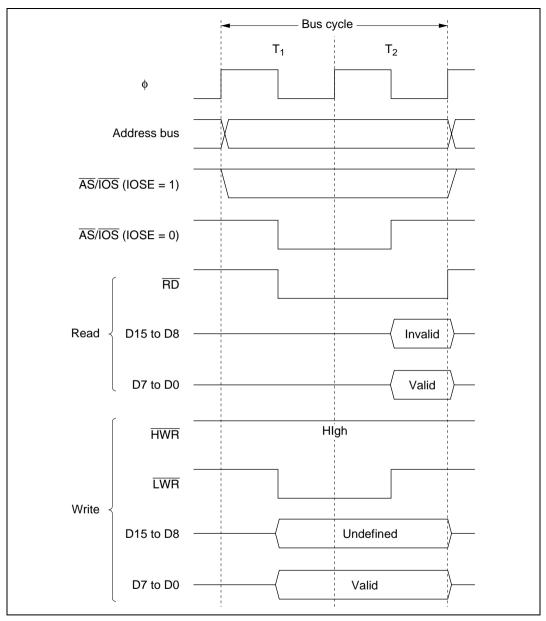


Figure 6.8 16-Bit, 2-State Access Space Bus Timing (2) (Odd Address Byte Access)

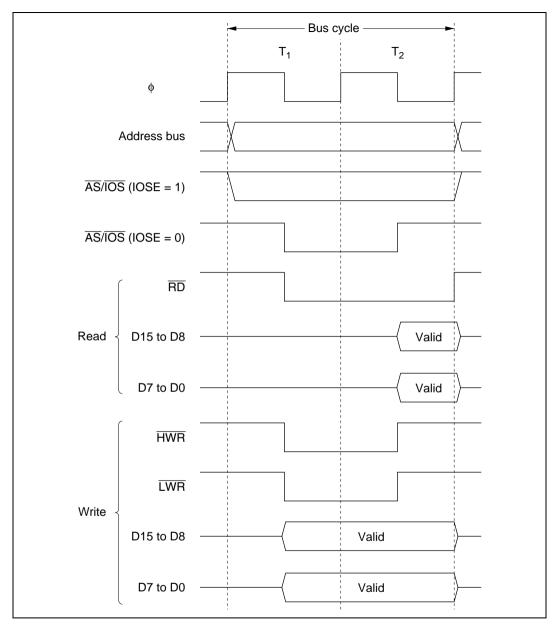


Figure 6.9 16-Bit, 2-State Access Space Bus Timing (3) (Word Access)

16-Bit, 3-State Access Space

Figures 6.10 to 6.12 show the bus timing for 16-bit, 3-state access space. When 16-bit access space is accessed, the upper data bus (D15 to D8) is used for even addresses and the lower data bus (D7 to D0) for odd addresses.

Wait states can be inserted.

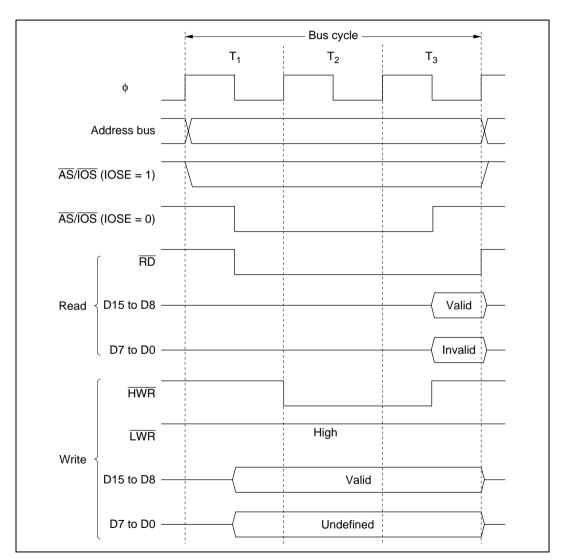


Figure 6.10 16-Bit, 3-State Access Space Bus Timing (1) (Even Address Byte Access)

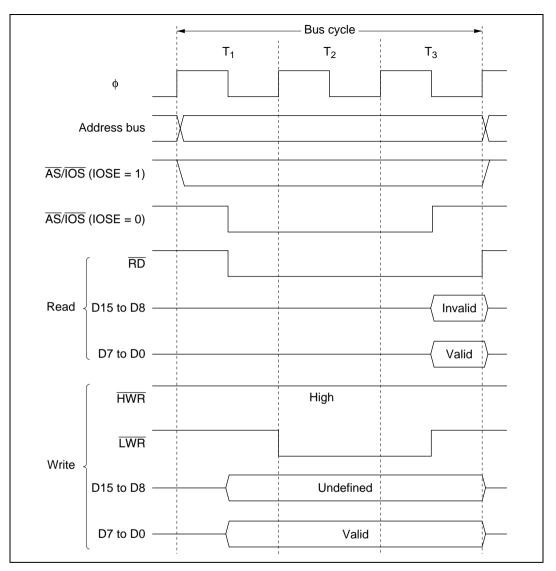


Figure 6.11 16-Bit, 3-State Access Space Bus Timing (2) (Odd Address Byte Access)

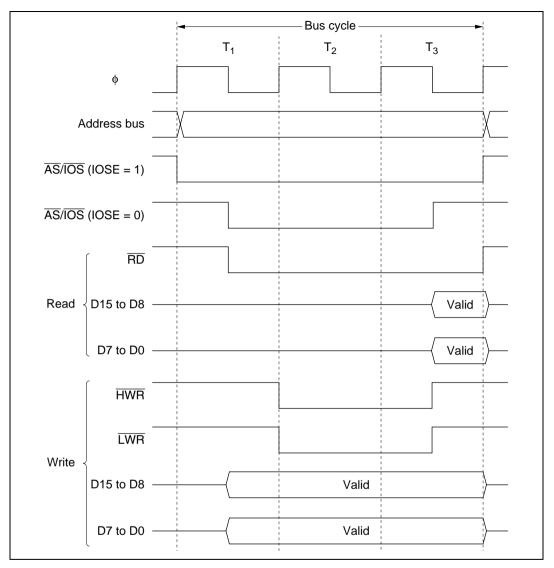


Figure 6.12 16-Bit, 3-State Access Space Bus Timing (3) (Word Access)

6.4.5 Wait Control

When accessing external space, the MCU can extend the bus cycle by inserting one or more wait states (T_w) . There are three ways of inserting wait states: program wait insertion, pin wait insertion using the \overline{WAIT} pin, and a combination of the two.

Program Wait Mode: In program wait mode, the number of T_w states specified by bits WC1 and WC0 are always inserted between the T_2 and T_3 states when external space is accessed.

Pin Wait Mode: In pin wait mode, the number of T_w states specified by bits WC1 and WC0 are always inserted between the T_2 and T_3 states when external space is accessed. If the \overline{WAIT} pin is low at the fall of ϕ in the last T_2 or T_w state, another T_w state is inserted. If the \overline{WAIT} pin is held low, T_w states are inserted until it goes high.

Pin wait mode is useful for inserting four or more wait states, or for changing the number of T_w states for different external devices.

Pin Auto-Wait Mode: In pin auto-wait mode, if the \overline{WAIT} pin is low at the fall of ϕ in the T_2 state, the number of T_w states specified by bits WC1 and WC0 are inserted when external space is accessed. No additional T_w states are inserted even if the \overline{WAIT} pin remains low. Pin auto-wait mode can be used for an easy interface to low-speed memory, simply by routing the chip select signal to the \overline{WAIT} pin.

Figure 6.13 shows an example of wait state insertion timing.

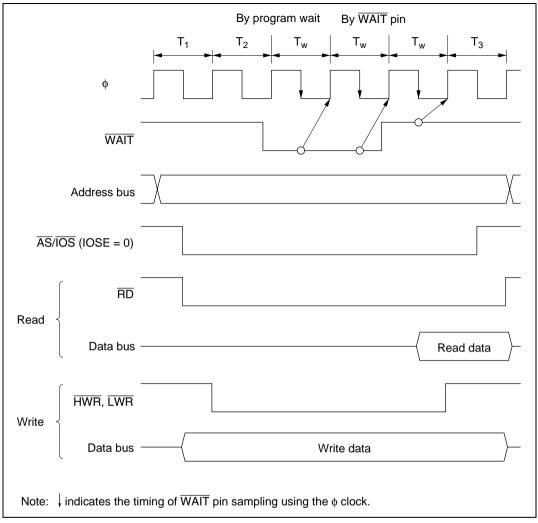


Figure 6.13 Example of Wait State Insertion Timing

The settings after a reset are: 3-state access, insertion of 3 program wait states, and WAIT input disabled.

6.5 Burst ROM Interface

6.5.1 Overview

With this LSI, external space area 0 can be designated as burst ROM space, and burst ROM interfacing can be performed.

External space can be designated as burst ROM space by means of the BRSTRM bit in BCR. Consecutive burst accesses of a maximum of 4 words or 8 words can be performed for CPU instruction fetches only. One or two states can be selected for burst access.

6.5.2 Basic Timing

The number of states in the initial cycle (full access) of the burst ROM interface is in accordance with the setting of the AST bit. Also, when the AST bit is set to 1, wait state insertion is possible. One or two states can be selected for the burst cycle, according to the setting of the BRSTS1 bit in BCR. Wait states cannot be inserted.

When the BRSTS0 bit in BCR is cleared to 0, burst access of up to 4 words is performed; when the BRSTS0 bit is set to 1, burst access of up to 8 words is performed.

The basic access timing for burst ROM space is shown in figures 6.14 (a) and (b). The timing shown in figure 6.14 (a) is for the case where the AST and BRSTS1 bits are both set to 1, and that in figure 6.14 (b) is for the case where both these bits are cleared to 0.

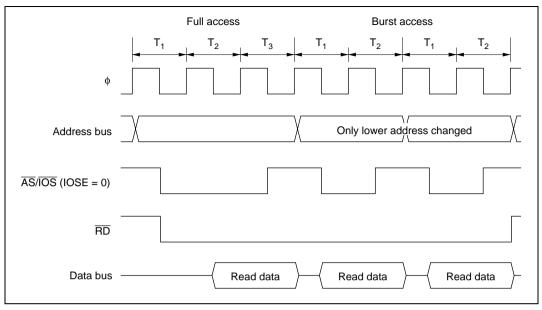


Figure 6.14 (a) Example of Burst ROM Access Timing (when AST = BRSTS1 = 1)

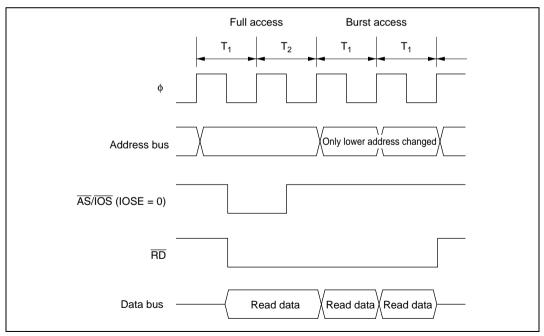


Figure 6.14 (b) Example of Burst ROM Access Timing (when AST = BRSTS1 = 0)

6.5.3 Wait Control

As with the basic bus interface, either program wait insertion or pin wait insertion using the $\overline{\text{WAIT}}$ pin can be used in the initial cycle (full access) of the burst ROM interface. See section 6.4.5, Wait Control.

Wait states cannot be inserted in a burst cycle.

6.6 Idle Cycle

6.6.1 Operation

When this LSI chip accesses external space, it can insert a 1-state idle cycle (T₁) between bus cycles when a write cycle occurs immediately after a read cycle. By inserting an idle cycle it is possible, for example, to avoid data collisions between ROM, with a long output floating time, and high-speed memory, I/O interfaces, and so on.

If an external write occurs after an external read while the ICISO bit in BCR is set to 1, an idle cycle is inserted at the start of the write cycle. This is enabled in advanced mode and normal mode.

Figure 6.15 shows an example of the operation in this case. In this example, bus cycle A is a read cycle from ROM with a long output floating time, and bus cycle B is a CPU write cycle. In (a), an idle cycle is not inserted, and a collision occurs in cycle B between the read data from ROM and the CPU write data. In (b), an idle cycle is inserted, and a data collision is prevented.

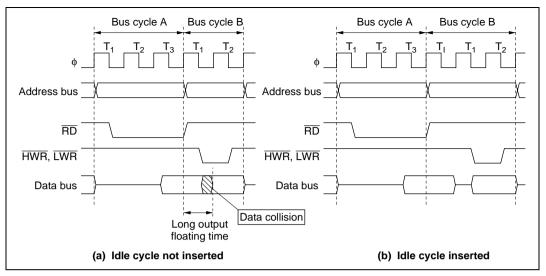


Figure 6.15 Example of Idle Cycle Operation

Pin States in Idle Cycle 6.6.2

Table 6.6 shows pin states in an idle cycle.

Table 6.6 Pin States in Idle Cycle

Pins	Pin State
A23 to A0, IOS	Contents of next bus cycle
D15 to D0	High impedance
ĀS	High
RD	High
HWR, LWR	High



6.7 Bus Arbitration

6.7.1 Overview

This LSI have a bus arbiter that arbitrates bus master operations.

There are two bus masters, the CPU and the DTC, which perform read/write operations when they have possession of the bus. Each bus master requests the bus by means of a bus request signal. The bus arbiter determines priorities at the prescribed timing, and permits use of the bus by means of a bus request acknowledge signal. The selected bus master then takes possession of the bus and begins its operation.

6.7.2 Operation

The bus arbiter detects the bus masters' bus request signals, and if the bus is requested, sends a bus request acknowledge signal to the bus master making the request. If there are bus requests from both bus masters, the bus request acknowledge signal is sent to the one with the higher priority. When a bus master receives the bus request acknowledge signal, it takes possession of the bus until that signal is canceled.

The order of priority of the bus masters is as follows:

(High) DTC > CPU (Low)

6.7.3 **Bus Transfer Timing**

Even if a bus request is received from a bus master with a higher priority than that of the bus master that has acquired the bus and is currently operating, the bus is not necessarily transferred immediately. There are specific times at which each bus master can relinquish the bus.

CPU

The CPU is the lowest-priority bus master, and if a bus request is received from the DTC, the bus arbiter transfers the bus to the DTC. The timing for transfer of the bus is as follows:

- The bus is transferred at a break between bus cycles. However, if a bus cycle is executed in discrete operations, as in the case of a longword-size access, the bus is not transferred between the operations.
 - See appendix A.5, Bus States during Instruction Execution, for timings at which the bus is not transferred
- If the CPU is in sleep mode, it transfers the bus immediately.

DTC

The DTC sends the bus arbiter a request for the bus when an activation request is generated.

The DTC does not release the bus until it has completed a series of processing operations.



Section 7 Data Transfer Controller (DTC)

Provided in the H8S/2148 Group; not provided in the H8S/2144 Group and H8S/2147N.

7.1 Overview

The H8S/2148 Group includes a data transfer controller (DTC). The DTC can be activated by an interrupt or software, to transfer data.

7.1.1 Features

- Transfer possible over any number of channels
 - Transfer information is stored in memory
 - One activation source can trigger a number of data transfers (chain transfer)
- Wide range of transfer modes
 - Normal, repeat, and block transfer modes available
 - Incrementing, decrementing, and fixing of transfer source and destination addresses can be selected
- Direct specification of 16-Mbyte address space possible
 - 24-bit transfer source and destination addresses can be specified
- Transfer can be set in byte or word units
- A CPU interrupt can be requested for the interrupt that activated the DTC
 - An interrupt request can be issued to the CPU after one data transfer ends
 - An interrupt request can be issued to the CPU after all specified data transfers have ended
- Activation by software is possible
- Module stop mode can be set
 - The initial setting enables DTC registers to be accessed. DTC operation is halted by setting module stop mode

7.1.2 Block Diagram

Figure 7.1 shows a block diagram of the DTC.

The DTC's register information is stored in the on-chip RAM*. A 32-bit bus connects the DTC to the on-chip RAM (1 kbyte), enabling 32-bit/1-state reading and writing of the DTC register information.

Note: * When the DTC is used, the RAME bit in SYSCR must be set to 1.

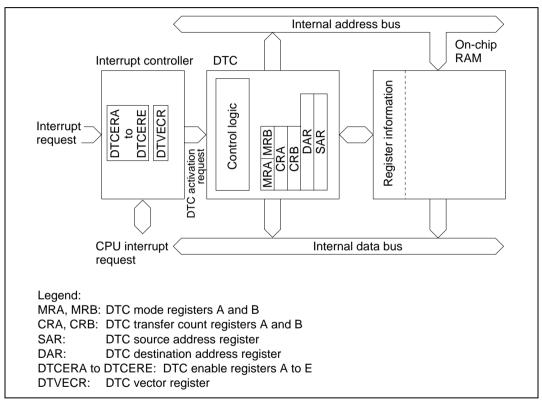


Figure 7.1 Block Diagram of DTC

7.1.3 Register Configuration

Table 7.1 summarizes the DTC registers.

Table 7.1 DTC Registers

Name	Abbreviation	R/W	Initial Value	Address*1
DTC mode register A	MRA	*2	Undefined	*3
DTC mode register B	MRB	*2	Undefined	*3
DTC source address register	SAR	*2	Undefined	*3
DTC destination address register	DAR	*2	Undefined	*3
DTC transfer count register A	CRA	*2	Undefined	*3
DTC transfer count register B	CRB	*2	Undefined	*3
DTC enable registers	DTCER*4	R/W	H'00	H'FEEE to H'FEF2
DTC vector register	DTVECR*4	R/W	H'00	H'FEF3
Module stop control register	MSTPCRH	R/W	H'3F	H'FF86
	MSTPCRL	R/W	H'FF	H'FF87

Notes: 1. Lower 16 bits of the address.

- 2. Registers within the DTC cannot be read or written to directly.
- 3. Allocated to on-chip RAM addresses H'EC00 to H'EFFF as register information. They cannot be located in external memory space.
 - When the DTC is used, do not clear the RAME bit in SYSCR to 0.
- The H8S/2144 Group and H8S/2147N do not include an on-chip DTC, and therefore the DTCER and DTVECR register addresses should not be accessed by the CPU.

7.2 **Register Descriptions**

DTC Mode Register A (MRA) 7.2.1

Bit	7	6	5	4	3	2	1	0
	SM1	SM0	DM1	DM0	MD1	MD0	DTS	Sz
Initial value	Unde-							
	fined							
Read/Write					_	_		_

MRA is an 8-bit register that controls the DTC operating mode.

Bits 7 and 6—Source Address Mode 1 and 0 (SM1, SM0): These bits specify whether SAR is to be incremented, decremented, or left fixed after a data transfer.

Bit 7	Bit 6	
SM1	SM0	Description
0	_	SAR is fixed
1	0	SAR is incremented after a transfer (by 1 when Sz = 0; by 2 when Sz = 1)
	1	SAR is decremented after a transfer (by 1 when Sz = 0; by 2 when Sz = 1)

Bits 5 and 4—Destination Address Mode 1 and 0 (DM1, DM0): These bits specify whether DAR is to be incremented, decremented, or left fixed after a data transfer.

Bit 5	Bit 4	
DM1	DM0	Description
0	_	DAR is fixed
1	0	DAR is incremented after a transfer (by 1 when Sz = 0; by 2 when Sz = 1)
	1	DAR is decremented after a transfer (by 1 when Sz = 0; by 2 when Sz = 1)

Bits 3 and 2—DTC Mode (MD1, MD0): These bits specify the DTC transfer mode.

Bit 3	Bit 2							
MD1	MD0	Description						
0	0	Normal mode						
	1	Repeat mode						
1	0	Block transfer mode						
	1	_						

Bit 1—DTC Transfer Mode Select (DTS): Specifies whether the source side or the destination side is set to be a repeat area or block area, in repeat mode or block transfer mode.

Bit 1

DTS	Description
0	Destination side is repeat area or block area
1	Source side is repeat area or block area

Bit 0—DTC Data Transfer Size (Sz): Specifies the size of data to be transferred.

Bit 0

Sz	Description	
0	Byte-size transfer	
1	Word-size transfer	

7.2.2 DTC Mode Register B (MRB)

Bit	7	6	5	4	3	2	1	0
	CHNE	DISEL	_				_	_
Initial value	Unde- fined							
Read/Write	_	_	_	_	_	_	_	_

MRB is an 8-bit register that controls the DTC operating mode.

Bit 7—DTC Chain Transfer Enable (CHNE): Specifies chain transfer. In chain transfer, multiple data transfers can be performed consecutively in response to a single transfer request. With data transfer for which CHNE is set to 1, there is no determination of the end of the specified number of transfers, clearing of the interrupt source flag, or clearing of DTCER.

Bit 7

CHNE	Description
0	End of DTC data transfer (activation waiting state is entered)
1	DTC chain transfer (new register information is read, then data is transferred)

Bit 6—DTC Interrupt Select (DISEL): Specifies whether interrupt requests to the CPU are disabled or enabled after a data transfer.

Bit 6

DISEL	Description
0	After a data transfer ends, the CPU interrupt is disabled unless the transfer counter is 0 (the DTC clears the interrupt source flag of the activating interrupt to 0)
1	After a data transfer ends, the CPU interrupt is enabled (the DTC does not clear the interrupt source flag of the activating interrupt to 0)

Bits 5 to 0—Reserved: In the H8S/2148 Group these bits have no effect on DTC operation, and should always be written with 0.

7.2.3 DTC Source Address Register (SAR)

Bit	23	22	21	20	19	 4	3	2	1	0
Initial value	Unde- Unde- Unde-				Unde-	 Unde-	Unde-	Unde-	Unde-	Unde-
	fined	fined	fined	fined	fined	fined	fined	fined	fined	fined
Read/write	_	_	_	_	_	 _	_	_	_	_

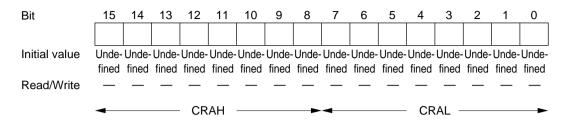
SAR is a 24-bit register that designates the source address of data to be transferred by the DTC. For word-size transfer, specify an even source address.

7.2.4 DTC Destination Address Register (DAR)

Bit	23	22	21	20	19	 4	3	2	1	0
Initial value	Unde-	Unde-	Unde-	Unde-	Unde-	 Unde-	Unde-	Unde-	Unde-	Unde-
	fined	fined	fined	fined	fined	fined	fined	fined	fined	fined
Read/write	_	_	_	_	_	 _	_	_	_	_

DAR is a 24-bit register that designates the destination address of data to be transferred by the DTC. For word-size transfer, specify an even destination address.

7.2.5 DTC Transfer Count Register A (CRA)



CRA is a 16-bit register that designates the number of times data is to be transferred by the DTC.

In normal mode, the entire CRA register functions as a 16-bit transfer counter (1 to 65,536). It is decremented by 1 every time data is transferred, and transfer ends when the count reaches H'0000.

In repeat mode or block transfer mode, CRA is divided into two parts: the upper 8 bits (CRAH) and the lower 8 bits (CRAL). CRAH holds the number of transfers while CRAL functions as an 8-bit transfer counter (1 to 256). CRAL is decremented by 1 every time data is transferred, and the contents of CRAH are transferred when the count reaches H'00. This operation is repeated.

7.2.6 DTC Transfer Count Register B (CRB)

Bit	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Initial value	Unde-															
	fined															
Read/Write	_	_	_	_	_	_	_	_	_	_	_	_	_	_	_	_

CRB is a 16-bit register that designates the number of times data is to be transferred by the DTC in block transfer mode. It functions as a 16-bit transfer counter (1 to 65,536) that is decremented by 1 every time data is transferred, and transfer ends when the count reaches H'0000.

7.2.7 DTC Enable Registers (DTCER)

Bit	7	6	5	4	3	2	1	0
	DTCE7	DTCE6	DTCE5	DTCE4	DTCE3	DTCE2	DTCE1	DTCE0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W							

The DTC enable registers comprise five 8-bit readable/writable registers, DTCERA to DTCERE, with bits corresponding to the interrupt sources that can activate the DTC. These bits enable or disable DTC service for the corresponding interrupt sources.

The DTC enable registers are initialized to H'00 by a reset and in hardware standby mode.

Bit n—DTC Activation Enable (DTCEn)

Bit n

DTCEn	 Description				
0	DTC activation by interrupt is disabled	(Initial value)			
	[Clearing conditions]				
	When data transfer ends with the DISEL bit set to 1				
	 When the specified number of transfers end 				
1	DTC activation by interrupt is enabled				
	[Holding condition]				
	When the DISEL bit is 0 and the specified number of transfers have no	ot ended			

Note: n = 7 to 0

A DTCE bit can be set for each interrupt source that can activate the DTC. The correspondence between interrupt sources and DTCE bits is shown in table 7.4, together with the vector number generated by the interrupt controller in each case.

For DTCE bit setting, read/write operations must be performed using bit-manipulation instructions such as BSET and BCLR. For the initial setting only, however, when multiple activation sources are set at one time, it is possible to disable interrupts and write after executing a dummy read on the relevant register.

7.2.8 DTC Vector Register (DTVECR)

Bit	7	6	5	4	3	2	1	0
	SWDTE	DTVEC6	DTVEC5	DTVEC4	DTVEC3	DTVEC2	DTVEC1	DTVEC0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/(W)*	R/W						

Note: * A value of 1 can always be written to the SWDTE bit, but 0 can only be written after 1 is read

DTVECR is an 8-bit readable/writable register that enables or disables DTC activation by software, and sets a vector number for the software activation interrupt.

DTVECR is initialized to H'00 by a reset and in hardware standby mode.

Bit 7—DTC Software Activation Enable (SWDTE): Specifies enabling or disabling of DTC software activation. To clear the SWDTE bit by software, read SWDTE when set to 1, then write 0 in the bit.

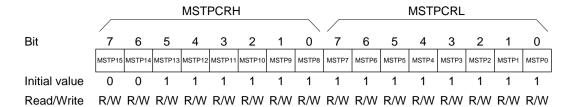
Bit 7

SWDTE	Description					
0	DTC software activation is disabled (Initial value)					
	[Clearing condition]					
	When the DISEL bit is 0 and the specified number of transfers have not ended					
1	DTC software activation is enabled					
	[Holding conditions]					
	 When data transfer ends with the DISEL bit set to 1 	When data transfer ends with the DISEL bit set to 1				
	When the specified number of transfers end					
	During software-activated data transfer					

Bits 6 to 0—DTC Software Activation Vectors 6 to 0 (DTVEC6 to DTVEC0): These bits specify a vector number for DTC software activation.

The vector address is H'0400 + (vector number) << 1 (where << 1 indicates a 1-bit left shift). For example, if DTVEC6 to DTVEC0 = H'10, the vector address is H'0420.

7.2.9 Module Stop Control Register (MSTPCR)



MSTPCR, comprising two 8-bit readable/writable registers, performs module stop mode control.

When the MSTP14 bit in MSTPCR is set to 1, the DTC operation stops at the end of the bus cycle and a transition is made to module stop mode. Note that 1 cannot be written to the MSTP14 bit when the DTC is being activated. For details, see section 25.5, Module Stop Mode.

MSTPCR is initialized to H'3FFF by a reset and in hardware standby mode. It is not initialized in software standby mode.

MSTPCRH Bit 6—Module Stop (MSTP14): Specifies the DTC module stop mode.

MSTPCRH Bit 6

MSTP14	Description	
0	DTC module stop mode is cleared	(Initial value)
1	DTC module stop mode is set	

7.3 Operation

7.3.1 Overview

When activated, the DTC reads register information that is already stored in memory and transfers data on the basis of that register information. After the data transfer, it writes updated register information back to memory. Pre-storage of register information in memory makes it possible to transfer data over any required number of channels. Setting the CHNE bit to 1 makes it possible to perform a number of transfers with a single activation.

Figure 7.2 shows a flowchart of DTC operation.

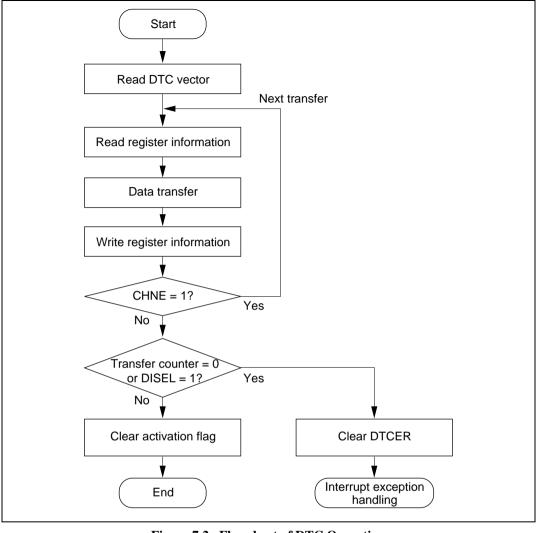


Figure 7.2 Flowchart of DTC Operation

The DTC transfer mode can be normal mode, repeat mode, or block transfer mode.

The 24-bit SAR designates the DTC transfer source address and the 24-bit DAR designates the transfer destination address. After each transfer, SAR and DAR are independently incremented, decremented, or left fixed.

Table 7.2 outlines the functions of the DTC.

Table 7.2 DTC Functions

				Addres	s Registers
Tr	ansfer Mode	A	ctivation Source	Transfer Source	Transfer Destination
•	Normal mode	•	IRQ	24 bits	24 bits
	 One transfer request transfers one byte or one word Memory addresses are incremented 	•	FRT ICI, OCI 8-bit timer CMI Host interface IBF		
	or decremented by 1 or 2 — Up to 65,536 transfers possible	•	SCI TXI or RXI A/D converter ADI		
•	Repeat mode One transfer request transfers one byte or one word	•	IIC IICI Software		
	 Memory addresses are incremented or decremented by 1 or 2 				
	 After the specified number of transfers (1 to 256), the initial state resumes and operation continues 				
•	Block transfer mode				
	 One transfer request transfers a block of the specified size 				
	 Block size is from 1 to 256 bytes or words 				
	— Up to 65,536 transfers possible				
	 A block area can be designated at eithe the source or destination 	r			

7.3.2 Activation Sources

The DTC operates when activated by an interrupt or by a write to DTVECR by software (software activation). An interrupt request can be directed to the CPU or DTC, as designated by the corresponding DTCER bit. The interrupt request is directed to the DTC when the corresponding bit is set to 1, and to the CPU when the bit is cleared to 0.

At the end of one data transfer (or the last of the consecutive transfers in the case of chain transfer) the interrupt source or the corresponding DTCER bit is cleared. Table 7.3 shows activation sources and DTCER clearing.

The interrupt source flag for RXIO, for example, is the RDRF flag in SCIO.

Table 7.3 Activation Sources and DTCER Clearing

Activation Source	When DISEL Bit Is 0 and Specified Number of Transfers Have Not Ended	When DISEL Bit Is 1 or Specified Number of Transfers Have Ended
Software	SWDTE bit cleared to 0	SWDTE bit held at 1
activation		 Interrupt request sent to CPU
Interrupt	Corresponding DTCER bit held	Corresponding DTCER bit cleared to 0
activation	at 1	 Activation source flag held at 1
	 Activation source flag cleared to 0 	 Activation source interrupt request sent to CPU

Figure 7.3 shows a block diagram of activation source control. For details see section 5, Interrupt Controller.

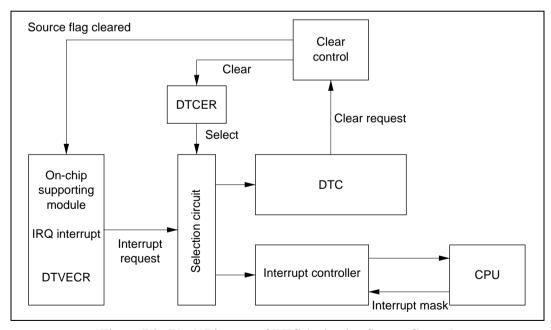


Figure 7.3 Block Diagram of DTC Activation Source Control

When an interrupt has been designated a DTC activation source, existing CPU mask level and interrupt controller priorities have no effect. If there is more than one activation source at the same time, the DTC is activated in accordance with the default priorities.

7.3.3 DTC Vector Table

Figure 7.4 shows the correspondence between DTC vector addresses and register information.

Table 7.4 shows the correspondence between activation sources, vector addresses, and DTCER bits. When the DTC is activated by software, the vector address is obtained from: H'0400 + DTVECR[6:0] << 1 (where << 1 indicates a 1-bit left shift). For example, if DTVECR is H'10, the vector address is H'0420.

The DTC reads the start address of the register information from the vector address set for each activation source, and then reads the register information from that start address. The register information can be placed at predetermined addresses in the on-chip RAM. The start address of the register information should be an integral multiple of four.

The configuration of the vector address is the same in both normal and advanced modes, a 2-byte unit being used in both cases. These two bytes specify the lower bits of the address in the on-chip RAM.

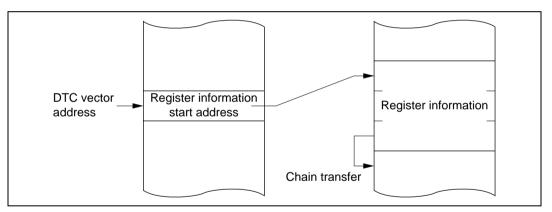


Figure 7.4 Correspondence between DTC Vector Address and Register Information

Table 7.4 Interrupt Sources, DTC Vector Addresses, and Corresponding DTCEs

Interrupt Source	Origin of Interrupt Source	Vector Number	Vector Address	DTCE*	Priority
Write to DTVECR	Software	DTVECR (Decimal indication)	H'0400 + DTVECR [6:0] << 1	_	High •
IRQ0	External pin	16	H'0420	DTCEA7	_
IRQ1	_	17	H'0422	DTCEA6	_
IRQ2	_	18	H'0424	DTCEA5	_
IRQ3	_	19	H'0426	DTCEA4	_
ADI (A/D conversion end)	A/D	28	H'0438	DTCEA3	_
ICIA (FRT input capture A)	FRT	48	H'0460	DTCEA2	_
ICIB (FRT input capture B)	_	49	H'0462	DTCEA1	_
OCIA (FRT output compare A)	_	52	H'0468	DTCEA0	_
OCIB (FRT output compare B)	_	54	H'046A	DTCEB7	_
CMIA0 (TMR0 compare-match A)	TMR0	64	H'0480	DTCEB2	_
CMIB0 (TMR0 compare-match B)	_	65	H'0482	DTCEB1	_
CMIA1 (TMR1 compare-match A)	TMR1	68	H'0488	DTCEB0	_
CMIB1 (TMR1 compare-match B)	_	69	H'048A	DTCEC7	_
CMIAY (TMRY compare-match A)	TMRY	72	H'0490	DTCEC6	_
CMIBY (TMRY compare-match B)	_	73	H'0492	DTCEC5	_
IBF1 (IDR1 reception completed)	HIF	76	H'0498	DTCEC4	_
IBF2 (IDR2 reception completed)	_	77	H'049A	DTCEC3	_
RXI0 (reception completed 0)	SCI channel 0	81	H'04A2	DTCEC2	_
TXI0 (transmit data empty 0)	_	82	H'04A4	DTCEC1	_
RXI1 (reception completed 1)	SCI channel 1	85	H'04AA	DTCEC0	_
TXI1 (transmit data empty 1)	_	86	H'04AC	DTCED7	_
RXI2 (reception completed 2)	SCI channel 2	89	H'04B2	DTCED6	_
TXI2 (transmit data empty 2)	_	90	H'04B4	DTCED5	_
IICI0 (IIC0 1-byte transmission/ reception completed)	IIC0 (option)	92	H'04B8	DTCED4	_
IICI1 (IIC1 1-byte transmission/reception completed)	IIC1 (option)	94	H'04BC	DTCED3	Low

Note: * DTCE bits with no corresponding interrupt are reserved, and should be written with 0.

7.3.4 Location of Register Information in Address Space

Figure 7.5 shows how the register information should be located in the address space.

Locate the MRA, SAR, MRB, DAR, CRA, and CRB registers, in that order, from the start address of the register information (vector address contents). In chain transfer, locate the register information in consecutive areas.

Locate the register information in the on-chip RAM (addresses: H'FFEC00 to H'FFEFFF).

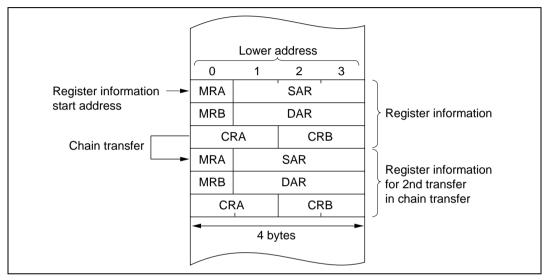


Figure 7.5 Location of DTC Register Information in Address Space

7.3.5 Normal Mode

In normal mode, one operation transfers one byte or one word of data.

From 1 to 65,536 transfers can be specified. Once the specified number of transfers have ended, a CPU interrupt can be requested.

Table 7.5 lists the register information in normal mode and figure 7.6 shows memory mapping in normal mode.

Table 7.5 Register Information in Normal Mode

Name	Abbreviation	Function
DTC source address register	SAR	Transfer source address
DTC destination address register	DAR	Transfer destination address
DTC transfer count register A	CRA	Transfer count
DTC transfer count register B	CRB	Not used

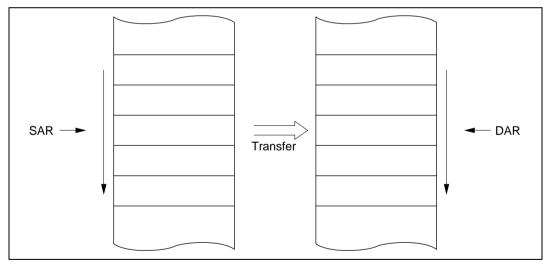


Figure 7.6 Memory Mapping in Normal Mode

7.3.6 Repeat Mode

In repeat mode, one operation transfers one byte or one word of data.

From 1 to 256 transfers can be specified. Once the specified number of transfers have ended, the initial address register state specified by the transfer counter and repeat area resumes and transfer is repeated. In repeat mode the transfer counter does not reach H'00, and therefore CPU interrupts cannot be requested when DISEL = 0.

Table 7.6 lists the register information in repeat mode and figure 7.7 shows memory mapping in repeat mode.

Table 7.6 Register Information in Repeat Mode

Name	Abbreviation	Function
DTC source address register	SAR	Transfer source address
DTC destination address register	DAR	Transfer destination address
DTC transfer count register AH	CRAH	Holds number of transfers
DTC transfer count register AL	CRAL	Transfer count
DTC transfer count register B	CRB	Not used

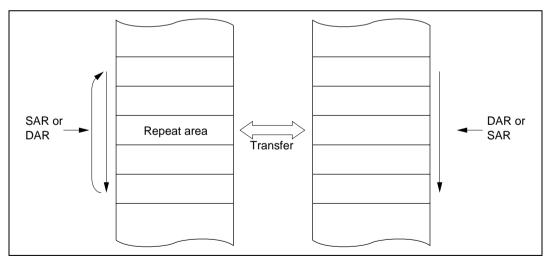


Figure 7.7 Memory Mapping in Repeat Mode

7.3.7 Block Transfer Mode

In block transfer mode, one operation transfers one block of data. Either the transfer source or the transfer destination is specified as a block area.

The block size is 1 to 256. When the transfer of one block ends, the initial state of the block size counter and the address register specified in the block area is restored. The other address register is successively incremented or decremented, or left fixed.

From 1 to 65,536 transfers can be specified. Once the specified number of transfers have ended, a CPU interrupt is requested.

Table 7.7 lists the register information in block transfer mode and figure 7.8 shows memory mapping in block transfer mode.

Table 7.7 Register Information in Block Transfer Mode

Name	Abbreviation	Function		
DTC source address register	SAR	Transfer source address		
DTC destination address register	DAR	Transfer destination address		
DTC transfer count register AH	CRAH	Holds block size		
DTC transfer count register AL	CRAL	Block size count		
DTC transfer count register B	CRB	Transfer counter		

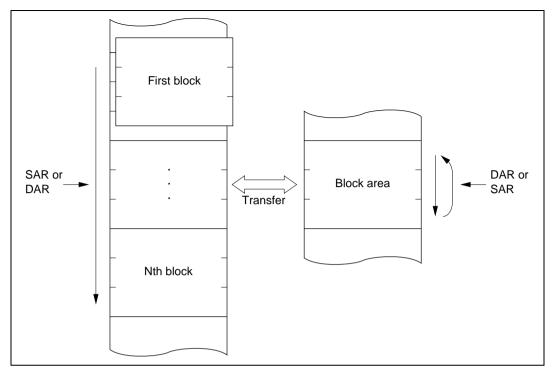


Figure 7.8 Memory Mapping in Block Transfer Mode

7.3.8 Chain Transfer

Setting the CHNE bit to 1 enables a number of data transfers to be performed consecutively in response to a single transfer request. SAR, DAR, CRA, CRB, MRA, and MRB, which define data transfers, can be set independently.

Figure 7.9 shows memory mapping for chain transfer.

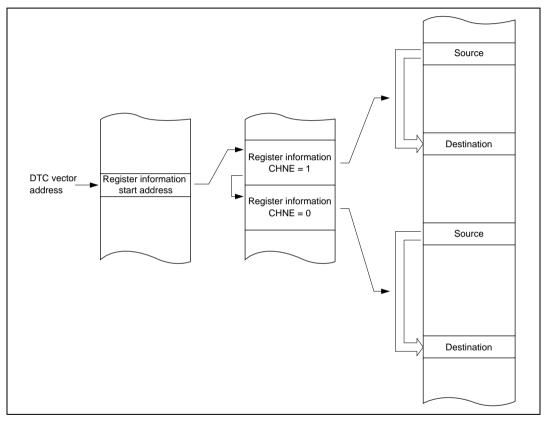


Figure 7.9 Memory Mapping in Chain Transfer

In the case of transfer with CHNE set to 1, an interrupt request to the CPU is not generated at the end of the specified number of transfers or by setting of the DISEL bit to 1, and the interrupt source flag for the activation source is not affected.

7.3.9 Operation Timing

Figures 7.10 to 7.12 show examples of DTC operation timing.

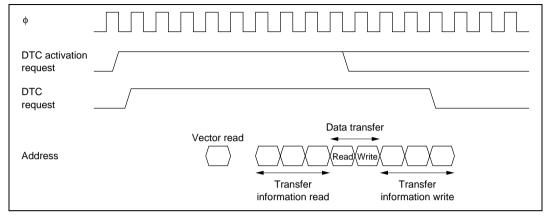


Figure 7.10 DTC Operation Timing (Normal Mode or Repeat Mode)

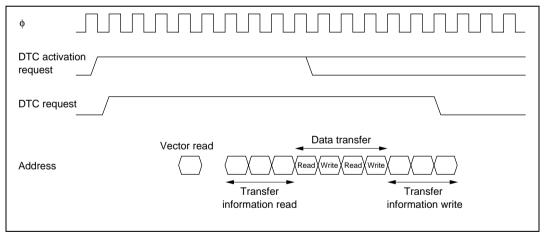


Figure 7.11 DTC Operation Timing (Block Transfer Mode, with Block Size of 2)

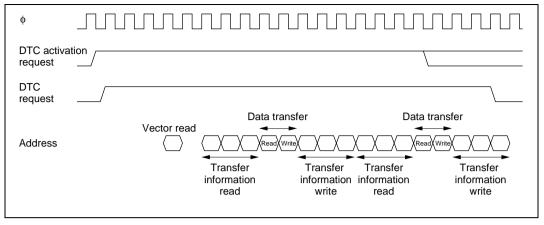


Figure 7.12 DTC Operation Timing (Chain Transfer)

7.3.10 Number of DTC Execution States

Table 7.8 lists execution phases for a single DTC data transfer, and table 7.9 shows the number of states required for each execution phase.

DTC Execution Phases Table 7.8

Mode	Vector Read I	Register Information Read/Write J	Data Read K	Data Write L	Internal Operation M
Normal	1	6	1	1	3
Repeat	1	6	1	1	3
Block transfer	1	6	N	N	3

Legend:

N: Block size (initial setting of CRAH and CRAL)

Table 7.9 Number of States Required for Each Execution Phase

Object of Access			On- Chip RAM	On- Chip ROM	Internal I/O Registers		External Devices			
Bus width			32	16	8	16	8	8	16	16
Access states		1	1	2	2	2	3	2	3	
Execution phase	Vector read	Sı	_	1	_	_	4	6+2m	2	3+m
	Register information read/write	S _J	1	_	_	_	_	_	_	_
	Byte data read	S _κ	1	1	2	2	2	3+m	2	3+m
	Word data read	S _K	1	1	4	2	4	6+2m	2	3+m
	Byte data write	S _L	1	1	2	2	2	3+m	2	3+m
	Word data write	S _L	1	1	4	2	4	6+2m	2	3+m
	Internal operation	S _M	1	1	1	1	1	1	1	1

The number of execution states is calculated from the formula below. Note that Σ means the sum of all transfers activated by one activation event (the number for which the CHNE bit is set to one, plus 1).

Number of execution states =
$$I \cdot S_1 + \Sigma (J \cdot S_1 + K \cdot S_K + L \cdot S_1) + M \cdot S_M$$

For example, when the DTC vector address table is located in on-chip ROM, normal mode is set, and data is transferred from the on-chip ROM to an internal I/O register, the time required for the DTC operation is 13 states. The time from activation to the end of the data write is 10 states.

7.3.11 Procedures for Using the DTC

Activation by Interrupt

The procedure for using the DTC with interrupt activation is as follows:

- 1. Set the MRA, MRB, SAR, DAR, CRA, and CRB register information in the on-chip RAM.
- 2. Set the start address of the register information in the DTC vector address.
- 3. Set the corresponding bit in DTCER to 1.
- 4. Set the enable bits for the interrupt sources to be used as the activation sources to 1. The DTC is activated when an interrupt used as an activation source is generated.
- 5. After the end of one data transfer, or after the specified number of data transfers have ended, the DTCE bit is cleared to 0 and a CPU interrupt is requested. If the DTC is to continue transferring data, set the DTCE bit to 1.

Activation by Software

The procedure for using the DTC with software activation is as follows:

- 1. Set the MRA, MRB, SAR, DAR, CRA, and CRB register information in the on-chip RAM.
- 2. Set the start address of the register information in the DTC vector address.
- 3. Check that the SWDTE bit is 0.
- 4. Write 1 in the SWDTE bit and the vector number to DTVECR.
- 5. Check the vector number written to DTVECR.
- 6. After the end of one data transfer, if the DISEL bit is 0 and a CPU interrupt is not requested, the SWDTE bit is cleared to 0. If the DTC is to continue transferring data, set the SWDTE bit to 1. When the DISEL bit is 1, or after the specified number of data transfers have ended, the SWDTE bit is held at 1 and a CPU interrupt is requested.



7.3.12 Examples of Use of the DTC

Normal Mode

An example is shown in which the DTC is used to receive 128 bytes of data via the SCI.

- 1. Set MRA to fixed source address (SM1 = SM0 = 0), incrementing destination address (DM1 = 1, DM0 = 0), normal mode (MD1 = MD0 = 0), and byte size (Sz = 0). The DTS bit can have any value. Set MRB for one data transfer by one interrupt (CHNE = 0, DISEL = 0). Set the SCI RDR address in SAR, the start address of the RAM area where the data will be received in DAR, and 128 (H'0080) in CRA. CRB can be set to any value.
- 2. Set the start address of the register information at the DTC vector address.
- 3. Set the corresponding bit in DTCER to 1.
- 4. Set the SCI to the appropriate receive mode. Set the RIE bit in SCR to 1 to enable the reception complete (RXI) interrupt. Since the generation of a receive error during the SCI reception operation will disable subsequent reception, the CPU should be enabled to accept receive error interrupts.
- 5. Each time reception of one byte of data ends on the SCI, the RDRF flag in SSR is set to 1, an RXI interrupt is generated, and the DTC is activated. The receive data is transferred from RDR to RAM by the DTC. DAR is incremented and CRA is decremented. The RDRF flag is automatically cleared to 0.
- 6. When CRA becomes 0 after the 128 data transfers have ended, the RDRF flag is held at 1, the DTCE bit is cleared to 0, and an RXI interrupt request is sent to the CPU. The interrupt handling routine should perform wrap-up processing.

Software Activation

An example is shown in which the DTC is used to transfer a block of 128 bytes of data by means of software activation. The transfer source address is H'1000 and the destination address is H'2000. The vector number is H'60, so the vector address is H'04C0.

- Set MRA to incrementing source address (SM1 = 1, SM0 = 0), incrementing destination address (DM1 = 1, DM0 = 0), block transfer mode (MD1 = 1, MD0 = 0), and byte size (Sz = 0). The DTS bit can have any value. Set MRB for one block transfer by one interrupt (CHNE = 0). Set the transfer source address (H'1000) in SAR, the destination address (H'2000) in DAR, and 128 (H'8080) in CRA. Set 1 (H'0001) in CRB.
- 2. Set the start address of the register information at the DTC vector address (H'04C0).
- 3. Check that the SWDTE bit in DTVECR is 0. Check that there is currently no transfer activated by software.
- 4. Write 1 to the SWDTE bit and the vector number (H'60) to DTVECR. The write data is H'E0.
- 5. Read DTVECR again and check that it is set to the vector number (H'60). If it is not, this indicates that the write failed. This is presumably because an interrupt occurred between steps 3 and 4 and led to a different software activation. To activate this transfer, go back to step 3.
- 6. If the write was successful, the DTC is activated and a block of 128 bytes of data is transferred.
- 7. After the transfer, an SWDTEND interrupt occurs. The interrupt handling routine should clear the SWDTE bit to 0 and perform other wrap-up processing.



7.4 Interrupts

An interrupt request is issued to the CPU when the DTC finishes the specified number of data transfers, or a data transfer for which the DISEL bit was set to 1. In the case of interrupt activation, the interrupt set as the activation source is generated. These interrupts to the CPU are subject to CPU mask level and interrupt controller priority level control.

In the case of activation by software, a software-activated data transfer end interrupt (SWDTEND) is generated.

When the DISEL bit is 1 and one data transfer has ended, or the specified number of transfers have ended, after data transfer ends, the SWDTE bit is held at 1 and an SWDTEND interrupt is generated. The interrupt handling routine should clear the SWDTE bit to 0.

When the DTC is activated by software, an SWDTEND interrupt is not generated during a data transfer wait or during data transfer even if the SWDTE bit is set to 1.

7.5 Usage Notes

Module Stop

When the MSTP14 bit in MSTPCR is set to 1, the DTC clock stops, and the DTC enters the module stop state. However, 1 cannot be written in the MSTP14 bit while the DTC is operating. When the DTC is placed in the module stop state, the DTCER registers must all be in the cleared state when the MSTP14 bit is set to 1.

On-Chip RAM

The MRA, MRB, SAR, DAR, CRA, and CRB registers are all located in on-chip RAM. When the DTC is used, the RAME bit in SYSCR must not be cleared to 0.

DTCE Bit Setting

For DTCE bit setting, read/write operations must be performed using bit-manipulation instructions such as BSET and BCLR. For the initial setting only, however, when multiple activation sources are set at one time, it is possible to disable interrupts and write after executing a dummy read on the relevant register.

Section 8 I/O Ports

8.1 Overview

This LSI have ten I/O ports (ports 1 to 6, 8, 9, A, and B), and one input-only port (port 7).

Tables 8.1 to 8.3 summarize the port functions. The pins of each port also have other functions.

Each port includes a data direction register (DDR) that controls input/output (not provided for the input-only port) and data registers (DR, ODR) that store output data.

Ports 1 to 3, 6, A, and B have a built-in MOS input pull-up function. For ports A and B, the on/off status of the MOS input pull-up is controlled by DDR and ODR. Ports 1 to 3 and 6 have a MOS input pull-up control register (PCR), in addition to DDR and DR, to control the on/off status of the MOS input pull-ups.

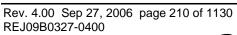
Ports 1 to 6, 8, 9, A, and B can drive a single TTL load and 30 pF capacitive load. All the I/O ports can drive a Darlington transistor when in output mode. Ports 1, 2, and 3 can drive an LED (10 mA sink current).

PA4 to PA7 in port A have bus buffer drive capability. In the H8S/2148 Group and H8S/2147N, P52 in port 5 and P97 in port 9 are NMOS push-pull outputs.

Note that the H8S/2144 Group and H8S/2147N have subset specifications that do not include some supporting modules. For differences in pin functions, see table 8.1, H8S/2148 Group Port Functions, table 8.2, H8S/2147N Port Functions, and table 8.3, H8S/2144 Group Port Functions.

Table 8.1 H8S/2148 Group Port Functions

			Expar	nded Modes	Single-Chip Mode
Port	Description	Pins	Mode 1	Mode 2, Mode 3 (EXPE = 1)	Mode 2, Mode 3 (EXPE = 0)
Port 1	8-bit I/O port Built-in MOS input pull-ups LED drive capability	P17 to P10/ A7 to A0/ PW7 to PW0	Lower address output (A7 to A0)	When DDR = 0 (after reset): input port When DDR = 1: lower address output (A7 to A0) or PWM timer output (PW7 to PW0)	I/O port also functioning as PWM timer output (PW7 to PW0)
Port 2	8-bit I/O port Built-in MOS input pull-ups LED drive capability	P27/A15/PW15/ CBLANK P26/A14/PW14 P25/A13/PW13 P24/A12/PW12 P23/A11/PW11 P22/A10/PW10 P21/A9/PW9 P20/A8/PW8	Upper address output (A15 to A8)	When DDR = 0 (after reset): input port or timer connection output (CBLANK) When DDR = 1: upper address output (A15 to A8), PWM timer output (PW15 to PW8), timer connection output (CBLANK), or output ports (P27 to P24)	I/O port also functioning as PWM timer output (PW15 to PW8) and timer connection output (CBLANK)
Port 3	8-bit I/O port Built-in MOS input pull-ups LED drive capability	P37 to P30/ HDB7 to HDB0/ D15 to D8	Data bus input/	output (D15 to D8)	I/O port also functioning as host interface data bus input/output (HDB7 to HDB0)





			Expar	nded Modes	Single-Chip Mode
Port	Description	Pins	Mode 1	Mode 2, Mode 3 (EXPE = 1)	Mode 2, Mode 3 (EXPE = 0)
Port 4	8-bit I/O port	P47/PWX1 P46/PWX0 P45/TMRI1/ HIRQ12/CSYNCI P44/TMO1/ HIRQ1/HSYNCO P43/TMCI1/ HIRQ11/HSYNCI P42/TMRI0/ SCK2/SDA1 P41/TMO0/ RxD2/IrRxD P40/TMCI0/ TxD2/IrTxD	I/O port also functioning as 14-bit PWM timer output (PWX1, PWX0), 8-bit timer 0 and 1 input/output (TMCI0, TMRI0, TMO0, TMCI1, TMRI1, TMO1), timer connection input/output (HSYNCO, CSYNCI, HSYNCI), SCI2 input/output (TxD2, RxD2, SCK2), IrDA interface input/output (IrTxD, IrRxD), and I ² C bus interface 1 (option) input/output (SDA1)		I/O port also functioning as 14-bit PWM timer output (PWX1, PWX0), 8-bit timer 0 and 1 input/output (TMCI0, TMRI0, TMO0, TMCI1, TMRI1, TMO1), timer connection input/output (HSYNCO, CSYNCI, HSYNCI), host interface host CPU interrupt request output (HIRQ12, HIRQ1, HIRQ11), SCI2 input/output (TxD2, RxD2, SCK2), IrDA interface input/output (IrTxD, IrRxD), and I²C bus interface 1 (option) input/output (SDA1)
Port 5	• 3-bit I/O port	P52/SCK0/SCL0 P51/RxD0 P50/TxD0	I/O port also functioning as SCI0 input/out SCK0) and I ² C bus interface 0 (option) input		ut/output (TxD0, RxD0,
Port 6	8-bit I/O port	P67/IRQ7/TMOX/ KIN7/CIN7 P66/IRQ6/FTOB/ KIN6/CIN6 P65/FTID/KIN5/ CIN5 P64/FTIC/KIN4/ CIN4/CLAMPO P63/FTIB/KIN3/ CIN3/VFBACKI P62/FTIA/TMIY/ KIN2/CIN2/ VSYNCI P61/FTOA/KIN1/ CIN1/VSYNCO P60/FTCI/TMIX/ KIN0/CIN0/ HFBACKI	FRT input/outp 8-bit timer X an connection inpu VSYNCO, HFB	ut (FTCI, FTOA, FTIA d Y input/output (TMC ut/output (CLAMPO, V	errupt input ($\overline{KIN7}$ to $\overline{KIN0}$),

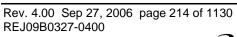
			Expar	nded Modes	Single-Chip Mode
Port	Description	Pins	Mode 1	Mode 2, Mode 3 (EXPE = 1)	Mode 2, Mode 3 (EXPE = 0)
Port 7	• 8-bit I/O port	P76/AN6/DA0 P75/AN5 P74/AN4 P73/AN3 P72/AN2 P71/AN1		functioning as A/D coconverter analog out	onverter analog input (AN7 to put (DA1, DA0)
Port 8	• 7-bit I/O port	P70/AN0 P86/IRQ5/SCK1/ SCL1 P85/IRQ4/RxD1 P84/IRQ3/TxD1 P83 P82/HIFSD P81/CS2/GA20 P80/HA0	I/O port also functioning as external interrupt input (IRQ5, IRQ4, IRQ3), SCI1 input/output (TxD1, RxD1, SCK1), and I²C bus interface 1 (option) input/output (SCL1)		I/O port also functioning as external interrupt input (IRQ5, IRQ4, IRQ3), SCI1 input/output (TxD1, RxD1, SCK1), host interface control input/output (CS2, GA20, HA0, HIFSD), and I ² C bus interface 1 (option) input/output (SCL1)
Port 9	• 8-bit I/O port	P97/WAIT/SDA0 P96/φ/EXCL	expanded data bus control input I ² C bus interface 0		, , ,
		P95/AS/IOS/CS1 P94/HWR/IOW P93/RD/IOR	expanded data (AS/IOS, HWR	bus control output RD)	I/O port also functioning as host interface control input (CS1, IOW, IOR)



			Expan	ded Modes	Single-Chip Mode
Port	Description	Pins	Mode 1	Mode 2, Mode 3 (EXPE = 1)	Mode 2, Mode 3 (EXPE = 0)
Port 9	8-bit I/O port	P92/IRQ0 P91/IRQ1	I/O port also fur	nctioning as external in	nterrupt input (IRQ0, IRQ1)
		P90/LWR/IRQ2/ ADTRG/ECS2	I/O port also functioning as expanded data bus control output (LWR), external interrupt input (IRQ2), and A/D converter external trigger input (ADTRG)		I/O port also functioning as external interrupt input (IRQ2), A/D converter external trigger input (ADTRG), and host interface control input (ECS2)
Port A	• 8-bit I/O port	CIN15/PS2CD PA6/A22/KIN14/ CIN14/PS2CC PA5/A21/KIN13/ CIN13/PS2BD PA4/A20/KIN12/	converter input	I/O port also functioning as address output (A23 to A16), key-sense interrupt input (KIN15 to KIN8), expansion A/D converter input	I/O port also functioning as key-sense interrupt input (KIN15 to KIN8), expansion A/D converter input (CIN15 to CIN8), and keyboard buffer controller input/output (PS2CD, PS2CC, PS2BD, PS2BC, PS2BC, PS2BD, PS2BC, PS2BC, PS2BD, PS2BC,
		CIN12/PS2BC (CIN15 to CIN8), an keyboard buffer controller input/outp. (PS2CD, PS2CD, PS2CC, PS2BD, PA0/A16/KIN8/ CIN8) PS2AD, CIN8 (PS2AD, PS2AD,	CIN8), and keyboard buffer controller input/output (PS2CD, PS2CC, PS2BD, PS2BC,	(CIN15 to CIN8), and keyboard buffer controller input/ output (PS2CD, PS2CC, PS2BD, PS2BC, PS2AD, PS2AC)	PS2AD, PS2AC)
Port B	• 8-bit I/O port	PB7/D7 PB6/D6 PB5/D5 PB4/D4 PB3/D3/CS4 PB2/D2/CS3 PB1/D1/HIRQ4 PB0/D0/HIRQ3	port	de (ABW = 1): I/O ode (ABW = 0): data t (D7 to D0)	I/O port also functioning as HIF control input/output pins (CS3, CS4, HIRQ3, HIRQ4)

Table 8.2 H8S/2147N Port Functions

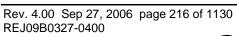
			Expar	ided Modes	Single-Chip Mode
Port	Description	Pins	Mode 1	Mode 2, Mode 3 (EXPE = 1)	Mode 2, Mode 3 (EXPE = 0)
Port 1	8-bit I/O port Built-in MOS input pull-ups LED drive capability	P17 to P10/ A7 to A0/ PW7 to PW0	Lower address output (A7 to A0)	When DDR = 0 (after reset): input port When DDR = 1: lower address output (A7 to A0) or PWM timer output (PW7 to PW0)	I/O port also functioning as PWM timer output (PW7 to PW0)
Port 2	8-bit I/O port Built-in MOS input pull-ups LED drive capability	P27/A15/PW15 P26/A14/PW14 P25/A13/PW13 P24/A12/PW12 P23/A11/PW11 P22/A10/PW10 P21/A9/PW9 P20/A8/PW8	Upper address output (A15 to A8)	When DDR = 0 (after reset): input port When DDR = 1: upper address output (A15 to A8), PWM timer output (PW15 to PW8), or output ports (P27 to P24)	I/O port also functioning as PWM timer output (PW15 to PW8)
Port 3	8-bit I/O port Built-in MOS input pull-ups LED drive capability	P37 to P30/ HDB7 to HDB0/ D15 to D8			I/O port also functioning as host interface data bus input/output (HDB7 to HDB0)
Port 4	• 8-bit I/O port	P47/PWX1 P46/PWX0 P45/TMRI1/ HIRQ12 P44/TMO1/ HIRQ1 P43/TMCI1/ HIRQ11 P42/TMRI0/ SCK2/SDA1 P41/TMO0/ RxD2/IrRxD P40/TMCI0/ TxD2/IrTxD	I/O port also functioning as 14-bit PWM timer output (PWX1, PWX0), 8-bit timer 0 and 1 input/output (TMCI0, TMRI0, TMO0, TMCI1, TMRI1, TMO1), SCI2 input/output (TxD2, RxD2, SCK2), IrDA interface input/output (IrTxD, IrRxD), and I ² C bus interface 1 (option) input/output (SDA1)		I/O port also functioning as 14-bit PWM timer output (PWX1, PWX0), 8-bit timer 0 and 1 input/output (TMCI0, TMRI0, TMO0, TMCI1, TMRI1, TMO1), host interface host CPU interrupt request output (HIRQ12, HIRQ1, HIRQ11), SCI2 input/output (TXD2, RXD2, SCK2), IrDA interface input/output (IrTxD, IrRxD), and I²C bus interface 1 (option) input/output (SDA1)





			Expanded Modes Single-Chi		Single-Chip Mode	
Port	Description	Pins	Mode 1	Mode 2, Mode 3 (EXPE = 1)	Mode 2, Mode 3 (EXPE = 0)	
Port 5	• 3-bit I/O port	P52/SCK0/SCL0	I/O port also functioning as SCI0 input/output (TxD0, RxD0, SCK0) and I ² C bus interface 0 (option) input/output (SCL0)			
		P51/RxD0				
		P50/TxD0				
Port 6	• 8-bit I/O port	P67/IRQ7/KIN7/ CIN7	I/O port also functioning as external interrupt input (IRQ7, IRC FRT input/output (FTCI, FTOA, FTIA, FTIB, FTIC, FTID, FTO			
		P66/IRQ6/FTOB/ KIN6/CIN6		out (TMIY), key-sense ansion A/D converter	interrupt input (KIN7 to input (CIN7 to CIN0)	
		P65/FTID/KIN5/ CIN5				
		P64/FTIC/KIN4/ CIN4				
		P63/FTIB/KIN3/ CIN3				
		P62/FTIA/TMIY/ KIN2/CIN2				
		P61/FTOA/KIN1/ CIN1				
		P60/FTCI/KINO/ CINO				
Port 7	• 8-bit I/O port	P77/AN7/DA1			nverter analog input (AN7 to	
		P76/AN6/DA0	AN0) and D/A	converter analog outp	ut (DA1, DA0)	
		P75/AN5				
		P74/AN4				
		P73/AN3				
		P72/AN2				
		P71/AN1				
		P70/AN0				
Port 8	• 7-bit I/O port	P86/IRQ5/SCK1/ SCL1	interrupt input	nctioning as external (IRQ5, IRQ4, IRQ3),	I/O port also functioning as external interrupt input	
		P85/IRQ4/RxD1		out (TxD1, RxD1,	(IRQ5, IRQ4, IRQ3), SCI1 input/output (TxD1, RxD1,	
		P84/IRQ3/TxD1	SCK1), and I ² C bus interface 1 input/output (TxD (option) input/output (SCL1) SCK1), host inter			
		P83			control input/output (CS2,	
		P82/HIFSD			GA20, HA0, HIFSD), and I ² C bus interface 1 (option)	
		P81/CS2/GA20			input/output (SCL1)	
		P80/HA0				

			Expar	ided Modes	Single-Chip Mode
Port	Description	Pins	Mode 1	Mode 2, Mode 3 (EXPE = 1)	Mode 2, Mode 3 (EXPE = 0)
Port 9	8-bit I/O port	P97/WAIT/SDA0	I/O port also functioning as expanded data bus control input (WAIT) and I ² C bus interface 0 (option) input/output (SDA0)		I/O port also functioning as I ² C bus interface 0 (option) input/output (SDA0)
		P96/ø/EXCL	When DDR = 0: input port or EXCL input When DDR = 1 (after reset): \$\phi\$ output	When DDR = 0 (after input When DDR = 1: ϕ ou	reset): input port or EXCL
		P95/AS/IOS/CS1 P94/HWR/IOW P93/RD/IOR	Expanded data (AS/IOS, HWR,	bus control output RD)	I/O port also functioning as host interface control input (CS1, IOW, IOR)
		P92/IRQ0 P91/IRQ1	I/O port also functioning as external in I/O port also functioning as expanded data bus control output (LWR), external interrupt input (IRQ2), and A/D converter external trigger input (ADTRG)		nterrupt input (IRQ0, IRQ1)
		P90/LWR/IRQ2/ ADTRG/ECS2			I/O port also functioning as external interrupt input (IRQ2), A/D converter external trigger input (ADTRG), and host interface control input (ECS2)
Port A	• 8-bit I/O port	PA7/A23/KIN15/ CIN15/PS2CD PA6/A22/KIN14/ CIN14/PS2CC PA5/A21/KIN13/ CIN13/PS2BD PA4/A20/KIN12/ CIN12/PS2BC PA3/A19/KIN11/ CIN11/PS2AD PA2/A18/KIN10/ CIN10/PS2AC PA1/A17/KIN9/ CIN9 PA0/A16/KIN8/ CIN8	I/O port also functioning as key-sense interrupt input (KIN15 to KIN8), expansion A/D converter input (CIN15 to CIN8), and keyboard buffer controller input/output (PS2CD, PS2CC, PS2BD, PS2BC, PS2AD, PS2AC)	I/O port also functioning as address output (A23 to A16), key-sense interrupt input (KIN15 to KIN8), expansion A/D converter input (CIN15 to CIN8), and keyboard buffer controller input/output (PS2CD, PS2CC, PS2BD, PS2BC, PS2AD, PS2AC)	I/O port also functioning as key-sense interrupt input (KIN15 to KIN8), expansion A/D converter input (CIN15 to CIN8), and keyboard buffer controller input/output (PS2CD, PS2CC, PS2BD, PS2BC, PS2AD, PS2AC)

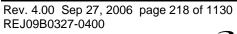




			Expar	nded Modes	Single-Chip Mode	
Port	ort Description Pins	Mode 1	Mode 2, Mode 3 (EXPE = 1)	Mode 2, Mode 3 (EXPE = 0)		
Port B	• 8-bit I/O port	PB7/D7	In 8-bit bus mode (ABW = 1): I/O port In 16-bit bus mode (ABW = 0): data bus input/output (D7 to D0)		I/O port also functioning as	
		PB6/D6			HIF control input/output pins (CS3, CS4, HIRQ3, HIRQ4)	
		PB5/D5				
		PB4/D4	bus inputoutpu	(07 10 00)		
		PB3/D3/CS4				
		PB2/D2/CS3				
		PB1/D1/HIRQ4				
		PB0/D0/HIRQ3				

Table 8.3 H8S/2144 Group Port Functions

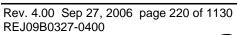
			Expar	nded Modes	Single-Chip Mode	
Port	Description	Pins	Mode 1 Mode 2, Mode 3 (EXPE = 1)		Mode 2, Mode 3 (EXPE = 0)	
Port 1	8-bit I/O port Built-in MOS input pull-	P17 to P10/ A7 to A0	Lower address output (A7 to A0)	When DDR = 0 (after reset): input port	I/O port	
	ups • LED drive capability			When DDR = 1: lower address output (A7 to A0)		
Port 2	8-bit I/O port Built-in MOS input pull-ups	P27 to P20/ A15 to A8	Upper address output (A15 to A8)	When DDR = 0 (after reset): input port When DDR = 1:	I/O port	
	LED drive capability			upper address output (A15 to A8) or output port (P27 to P24)		
Port 3	8-bit I/O port Built-in MOS input pull-ups	P37 to P30/ D15 to D8	Data bus input/output (D15 to D8)		I/O port	
	 LED drive capability 					
Port 4	8-bit I/O port	P47/PWX1 P46/PWX0	PWX0), 8-bit tir	ner 0 and 1 input/outp	/M timer output (PWX1, out (TMCI0, TMRI0, TMO0,	
		P45/TMRI1		TMO1), SCI2 input/o ace input/output (IrTxE	utput (TxD2, RxD2, SCK2),), IrRxD)	
		P44/TMO1			,	
		P43/TMCI1 P42/TMRI0/SCK2				
		P41/TMO0/RxD2/ IrRxD				
		P40/TMCI0/ TxD2/IrTxD				
Port 5	• 3-bit I/O port	P52/SCK0 P51/RxD0 P50/TxD0	I/O port also fur SCK0)	nctioning as SCI0 inpu	ut/output (TxD0, RxD0,	





			Expanded Modes Single-Chi		Single-Chip Mode	
Port	Description	Pins	Mode 1	Mode 2, Mode 3 (EXPE = 1)	Mode 2, Mode 3 (EXPE = 0)	
Port 6	8-bit I/O port	P67/IRQ7/KIN7/ CIN7	I/O port also functioning as external interrupt input (IRQ7 FRT input/output (FTCI, FTOA, FTIA, FTIB, FTIC, FTID, 8-bit timer Y input (TMIY), key-sense interrupt input (KIN KINO), and expansion A/D converter input (CIN7 to CINC			
		P66/IRQ6/FTOB/ KIN6/CIN6				
		P65/FTID/KIN5/ CIN5				
		P64/FTIC/KIN4/ CIN4				
		P63/FTIB/KIN3/ CIN3				
		P62/FTIA/TMIY/ KIN2/CIN2				
		P61/FTOA/KIN1/ CIN1				
		P60/FTCI/KINO/ CIN0				
Port 7	• 8-bit input	P77/AN7/DA1			overter analog input (AN7 to	
	port	P76/AN6/DA0	AN0) and D/A o	converter analog outpu	ut (DA1, DA0)	
		P75/AN5				
		P74/AN4				
		P73/AN3				
		P72/AN2				
		P71/AN1				
		P70/AN0				
Port 8	• 7-bit I/O port	P86/IRQ5/SCK1			nterrupt input (IRQ5, IRQ4,	
		P85/IRQ4/RxD1	IRQ3) and SCI1 input/output (TxD1, RxD1, SCK1)			
		P84/IRQ3/TxD1				
		P83				
		P82				
		P81				
		P80				

			Expar	nded Modes	Single-Chip Mode
Port	Description	Pins	Mode 1	Mode 2, Mode 3 (EXPE = 1)	Mode 2, Mode 3 (EXPE = 0)
Port 9	8-bit I/O port	P97/WAIT	I/O port also fur expanded data (WAIT)	nctioning as bus control input	I/O port
		P96/φ/EXCL	When DDR = 0 (after out or EXCL input When DDR = 1 (after reset):		reset): input port or EXCL
		P95/AS/IOS P94/HWR P93/RD	Expanded data output(AS/IOS,		I/O port
		P92/IRQ0 P91/IRQ1	I/O port also functioning as external in I/O port also functioning as expanded data bus control output (LWR), external interrupt input (IRQ2), and A/D converter external trigger input (ADTRG)		nterrupt input (IRQ0, IRQ1)
		P90/LWR/ IRQ2/ADTRG			I/O port also functioning as external interrupt input (IRQ2) and A/D converter external trigger input (ADTRG)
Port A	8-bit I/O port	PA7 to PA0/ A23 to A16/ KIN15 to KIN8/ CIN15 to CIN8	I/O port also functioning as key-sense interrupt input (KIN15 to KIN8), and expansion A/D converter input (CIN15 to CIN8)	I/O port also functioning as address output (A23 to A16), key-sense interrupt input (KIN15 to KIN8), and expansion A/D converter input (CIN15 to CIN8)	I/O port also functioning as key-sense interrupt input (KIN15 to KIN8) and expansion A/D converter input (CIN15 to CIN8)
Port B	8-bit I/O port	PB7 to PB0/ D7 to D0	In 8-bit bus mode (ABW = 1): I/O port In 16-bit bus mode (ABW = 0): data bus input/output (D7 to D0)		I/O port





8.2 Port 1

8.2.1 Overview

Port 1 is an 8-bit I/O port. Port 1 pins also function as address bus output pins (A7 to A0), and as 8-bit PWM output pins (PW7 to PW0) (H8S/2148 Group and H8S/2147N only). Port 1 functions change according to the operating mode. Port 1 has a built-in MOS input pull-up function that can be controlled by software.

Figure 8.1 shows the port 1 pin configuration.

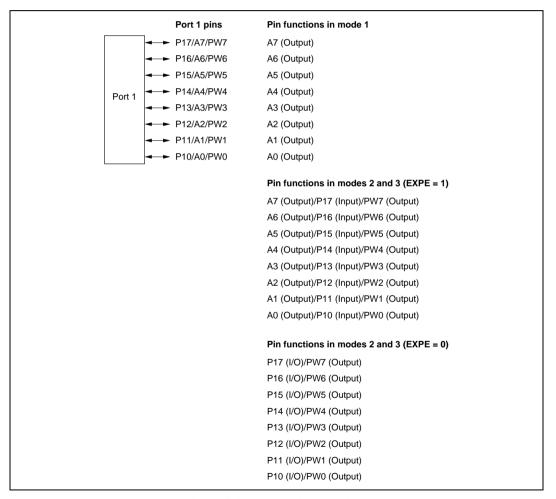


Figure 8.1 Port 1 Pin Functions

8.2.2 **Register Configuration**

Table 8.4 shows the port 1 register configuration.

Table 8.4 **Port 1 Registers**

Name	Abbreviation	R/W	Initial Value	Address*
Port 1 data direction register	P1DDR	W	H'00	H'FFB0
Port 1 data register	P1DR	R/W	H'00	H'FFB2
Port 1 MOS pull-up control register	P1PCR	R/W	H'00	H'FFAC

Lower 16 bits of the address. Note:

Port 1 Data Direction Register (P1DDR)

Bit	7	6	5	4	3	2	1	0
	P17DDR	P16DDR	P15DDR	P14DDR	P13DDR	P12DDR	P11DDR	P10DDR
Initial value	0	0	0	0	0	0	0	0
Read/Write	W	W	W	W	W	W	W	W

P1DDR is an 8-bit write-only register, the individual bits of which specify input or output for the pins of port 1. P1DDR cannot be read; if it is, an undefined value will be returned.

P1DDR is initialized to H'00 by a reset and in hardware standby mode. It retains its prior state in software standby mode. The address output pins maintain their output state in a transition to software standby mode.

Mode 1

The corresponding port 1 pins are address outputs, regardless of the P1DDR setting. In hardware standby mode, the address outputs go to the high-impedance state.

Modes 2 and 3 (EXPE = 1)

The corresponding port 1 pins are address outputs or PWM outputs when P1DDR bits are set to 1, and input ports when cleared to 0.

Modes 2 and 3 (EXPE = 0)

The corresponding port 1 pins are output ports or PWM outputs when P1DDR bits are set to 1, and input ports when cleared to 0.



Port 1 Data Register (P1DR)

Bit	7	6	5	4	3	2	1	0
	P17DR	P16DR	P15DR	P14DR	P13DR	P12DR	P11DR	P10DR
Initial value	0	0	0	0	0	0	0	0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

P1DR is an 8-bit readable/writable register that stores output data for the port 1 pins (P17 to P10). If a port 1 read is performed while P1DDR bits are set to 1, the P1DR values are read directly, regardless of the actual pin states. If a port 1 read is performed while P1DDR bits are cleared to 0, the pin states are read.

P1DR is initialized to H'00 by a reset and in hardware standby mode. It retains its prior state in software standby mode.

Port 1 MOS Pull-Up Control Register (P1PCR)

Bit	7	6	5	4	3	2	1	0
	P17PCR	P16PCR	P15PCR	P14PCR	P13PCR	P12PCR	P11PCR	P10PCR
Initial value	0	0	0	0	0	0	0	0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

P1PCR is an 8-bit readable/writable register that controls the port 1 built-in MOS input pull-ups on a bit-by-bit basis.

In modes 2 and 3, the MOS input pull-up is turned on when a P1PCR bit is set to 1 while the corresponding P1DDR bit is cleared to 0 (input port setting).

P1PCR is initialized to H'00 by a reset and in hardware standby mode. It retains its prior state in software standby mode.

8.2.3 Pin Functions in Each Mode

Mode 1

In mode 1, port 1 pins automatically function as address outputs. The port 1 pin functions are shown in figure 8.2.

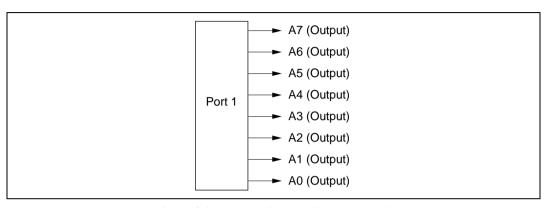


Figure 8.2 Port 1 Pin Functions (Mode 1)

Modes 2 and 3 (EXPE = 1)

In modes 2 and 3 (when EXPE = 1), port 1 pins function as address outputs, PWM outputs, or input ports, and input or output can be specified on a bit-by-bit basis. When a bit in P1DDR is set to 1, the corresponding pin functions as an address output or PWM output, and when cleared to 0, as an input port.

The port 1 pin functions are shown in figure 8.3.

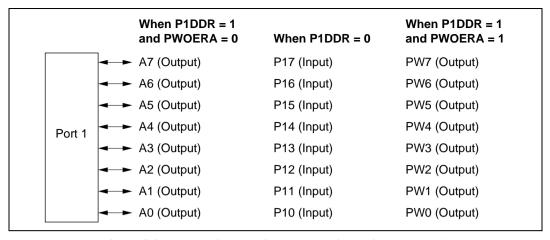


Figure 8.3 Port 1 Pin Functions (Modes 2 and 3 (EXPE = 1))

Modes 2 and 3 (EXPE = 0)

In modes 2 and 3 (when EXPE = 0), port 1 pins function as PWM outputs or I/O ports, and input or output can be specified on a bit-by-bit basis. When a bit in P1DDR is set to 1, the corresponding pin functions as a PWM output or output port, and when cleared to 0, as an input port.

The port 1 pin functions are shown in figure 8.4.

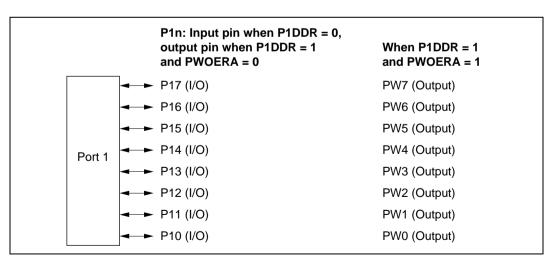


Figure 8.4 Port 1 Pin Functions (Modes 2 and 3 (EXPE = 0))

8.2.4 MOS Input Pull-Up Function

Port 1 has a built-in MOS input pull-up function that can be controlled by software. This MOS input pull-up function can be used in modes 2 and 3, and can be specified as on or off on a bit-by-bit basis

When a P1DDR bit is cleared to 0 in mode 2 or 3, setting the corresponding P1PCR bit to 1 turns on the MOS input pull-up for that pin.

The MOS input pull-up function is in the off state after a reset and in hardware standby mode. The prior state is retained in software standby mode.

Table 8.5 summarizes the MOS input pull-up states.

Table 8.5 MOS Input Pull-Up States (Port 1)

Mode	Reset	Hardware Standby Mode	Software Standby Mode	In Other Operations
1	Off	Off	Off	Off
2, 3	Off	Off	On/Off	On/Off

Legend:

Off: MOS input pull-up is always off.

On/Off: On when P1DDR = 0 and P1PCR = 1; otherwise off.

8.3 Port 2

8.3.1 Overview

Port 2 is an 8-bit I/O port. Port 2 pins also function as address bus output pins (A15 to A8), 8-bit PWM output pins (PW15 to PW8) (H8S/2148 Group and H8S/2147N only), and the timer connection output pin (CBLANK) (H8S/2148 Group only). Port 2 functions change according to the operating mode. Port 2 has a built-in MOS input pull-up function that can be controlled by software.

Figure 8.5 shows the port 2 pin configuration.

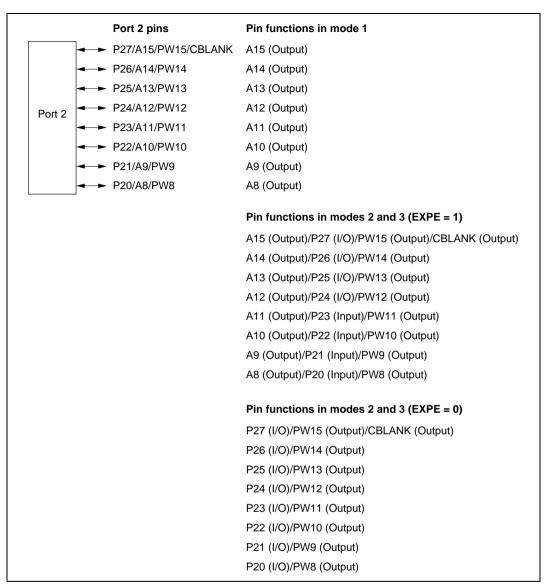


Figure 8.5 Port 2 Pin Functions

8.3.2 **Register Configuration**

Table 8.6 shows the port 2 register configuration.

Table 8.6 **Port 2 Registers**

Name	Abbreviation	R/W	Initial Value	Address*
Port 2 data direction register	P2DDR	W	H'00	H'FFB1
Port 2 data register	P2DR	R/W	H'00	H'FFB3
Port 2 MOS pull-up control register	P2PCR	R/W	H'00	H'FFAD

Note: Lower 16 bits of the address.

Port 2 Data Direction Register (P2DDR)

Bit	7	6	5	4	3	2	1	0
	P27DDR	P26DDR	P25DDR	P24DDR	P23DDR	P22DDR	P21DDR	P20DDR
Initial value	0	0	0	0	0	0	0	0
Read/Write	W	W	W	W	W	W	W	W

P2DDR is an 8-bit write-only register, the individual bits of which specify input or output for the pins of port 2. P2DDR cannot be read; if it is, an undefined value will be returned.

P2DDR is initialized to H'00 by a reset and in hardware standby mode. It retains its prior state in software standby mode. The address output pins maintain their output state in a transition to software standby mode.

Mode 1

The corresponding port 2 pins are address outputs, regardless of the P2DDR setting. In hardware standby mode, the address outputs go to the high-impedance state.

Modes 2 and 3 (EXPE = 1)

The corresponding port 2 pins are address outputs or PWM outputs when P2DDR bits are set to 1, and input ports when cleared to 0. P27 to P24 are switched from address outputs to output ports by setting the IOSE bit to 1.

P27 can be used as an on-chip supporting module output pin regardless of the P27DDR setting, but to ensure normal access to external space, P27 should not be set as an on-chip supporting module output pin when port 2 pins are used as address output pins.



• Modes 2 and 3 (EXPE = 0)

The corresponding port 2 pins are output ports or PWM outputs when P2DDR bits are set to 1, and input ports when cleared to 0.

P27 can be used as an on-chip supporting module output pin regardless of the P27DDR setting.

Port 2 Data Register (P2DR)

Bit	7	6	5	4	3	2	1	0
	P27DR	P26DR	P25DR	P24DR	P23DR	P22DR	P21DR	P20DR
Initial value	0	0	0	0	0	0	0	0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

P2DR is an 8-bit readable/writable register that stores output data for the port 2 pins (P27 to P20). If a port 2 read is performed while P2DDR bits are set to 1, the P2DR values are read directly, regardless of the actual pin states. If a port 2 read is performed while P2DDR bits are cleared to 0, the pin states are read.

P2DR is initialized to H'00 by a reset and in hardware standby mode. It retains its prior state in software standby mode.

Port 2 MOS Pull-Up Control Register (P2PCR)

Bit	7	6	5	4	3	2	1	0
	P27PCR	P26PCR	P25PCR	P24PCR	P23PCR	P22PCR	P21PCR	P20PCR
Initial value	0	0	0	0	0	0	0	0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

P2PCR is an 8-bit readable/writable register that controls the port 2 built-in MOS input pull-ups on a bit-by-bit basis.

In modes 2 and 3, the MOS input pull-up is turned on when a P2PCR bit is set to 1 while the corresponding P2DDR bit is cleared to 0 (input port setting).

P2PCR is initialized to H'00 by a reset and in hardware standby mode. It retains its prior state in software standby mode.

8.3.3 Pin Functions in Each Mode

Mode 1

In mode 1, port 2 pins automatically function as address outputs. The port 2 pin functions are shown in figure 8.6.

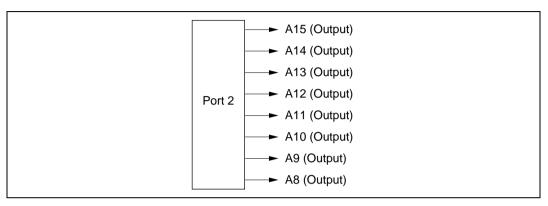


Figure 8.6 Port 2 Pin Functions (Mode 1)

Modes 2 and 3 (EXPE = 1)

In modes 2 and 3 (when EXPE = 1), port 2 pins function as address outputs, PWM outputs, or I/O ports, and input or output can be specified on a bit-by-bit basis. When a bit in P2DDR is set to 1, the corresponding pin functions as an address output or PWM output, and when cleared to 0, as an input port. P27 to P24 are switched from address outputs to output ports by setting the IOSE bit to 1. P27 can be used as an on-chip supporting module output pin regardless of the P27DDR setting, but to ensure normal access to external space, P27 should not be set as an on-chip supporting module output pin when port 2 pins are used as address output pins.

The port 2 pin functions are shown in figure 8.7.

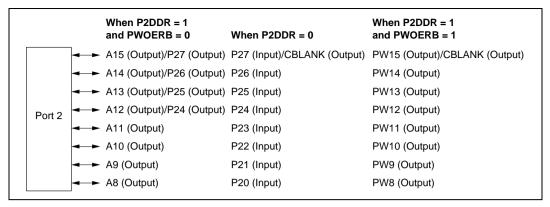


Figure 8.7 Port 2 Pin Functions (Modes 2 and 3 (EXPE = 1))

Modes 2 and 3 (EXPE = 0)

In modes 2 and 3 (when EXPE = 0), port 2 pins function as PWM outputs (P27 can also function as the timer connection output (CBLANK)) or I/O ports, and input or output can be specified on a bit-by-bit basis. When a bit in P2DDR is set to 1, the corresponding pin functions as a PWM output or output port, and when cleared to 0, as an input port. P27 can be used as an on-chip supporting module output pin regardless of the P27DDR setting.

The port 2 pin functions are shown in figure 8.8.

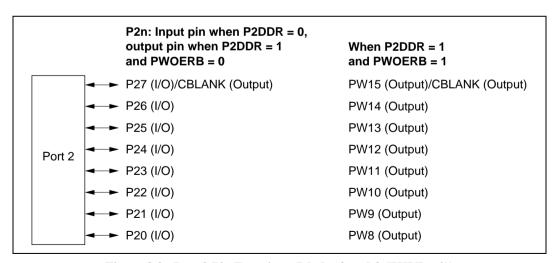


Figure 8.8 Port 2 Pin Functions (Modes 2 and 3 (EXPE = 0))

8.3.4 **MOS Input Pull-Up Function**

Port 2 has a built-in MOS input pull-up function that can be controlled by software. This MOS input pull-up function can be used in modes 2 and 3, and can be specified as on or off on a bit-bybit basis.

When a P2DDR bit is cleared to 0 in mode 2 or 3, setting the corresponding P2PCR bit to 1 turns on the MOS input pull-up for that pin.

The MOS input pull-up function is in the off state after a reset and in hardware standby mode. The prior state is retained in software standby mode.

Table 8.7 summarizes the MOS input pull-up states.

Table 8.7 MOS Input Pull-Up States (Port 2)

Mode	Reset	Hardware Standby Mode	Software Standby Mode	In Other Operations
1	Off	Off	Off	Off
2, 3	Off	Off	On/Off	On/Off

Legend:

Off: MOS input pull-up is always off.

On/Off: On when P2DDR = 0 and P2PCR = 1; otherwise off.



8.4 Port 3

8.4.1 Overview

Port 3 is an 8-bit I/O port. Port 3 pins also function as host data bus I/O pins (HDB7 to HDB0) (H8S/2148 Group and H8S/2147N only), and as data bus I/O pins. Port 3 functions change according to the operating mode. Port 3 has a built-in MOS input pull-up function that can be controlled by software.

Figure 8.9 shows the port 3 pin configuration.

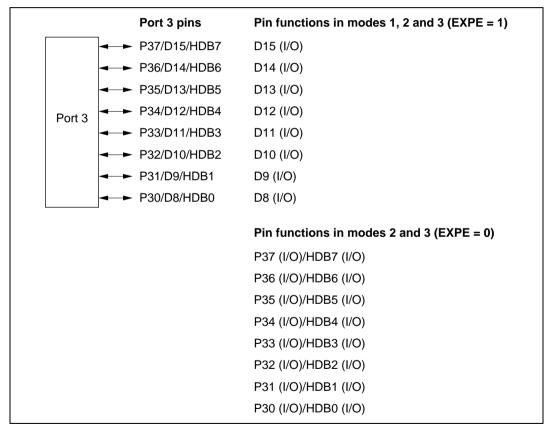


Figure 8.9 Port 3 Pin Functions

8.4.2 **Register Configuration**

Table 8.8 shows the port 3 register configuration.

Table 8.8 Port 3 Registers

Name	Abbreviation	R/W	Initial Value	Address*
Port 3 data direction register	P3DDR	W	H'00	H'FFB4
Port 3 data register	P3DR	R/W	H'00	H'FFB6
Port 3 MOS pull-up control register	P3PCR	R/W	H'00	H'FFAE

Lower 16 bits of the address. Note:

Port 3 Data Direction Register (P3DDR)

Bit	7	6	5	4	3	2	1	0
	P37DDR	P36DDR	P35DDR	P34DDR	P33DDR	P32DDR	P31DDR	P30DDR
Initial value	0	0	0	0	0	0	0	0
Read/Write	W	W	W	W	W	W	W	W

P3DDR is an 8-bit write-only register, the individual bits of which specify input or output for the pins of port 3. P3DDR cannot be read; if it is, an undefined value will be returned.

P3DDR is initialized to H'00 by a reset and in hardware standby mode. It retains its prior state in software standby mode.

Modes 1, 2, and 3 (EXPE = 1)

The input/output direction specified by P3DDR is ignored, and pins automatically function as data I/O pins.

After a reset, and in hardware standby mode or software standby mode, the data I/O pins go to the high-impedance state.

Modes 2 and 3 (EXPE = 0)

The corresponding port 3 pins are output ports when P3DDR bits are set to 1, and input ports when cleared to 0.



Port 3 Data Register (P3DR)

Bit	7	6	5	4	3	2	1	0
	P37DR	P36DR	P35DR	P34DR	P33DR	P32DR	P31DR	P30DR
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W							

P3DR is an 8-bit readable/writable register that stores output data for the port 3 pins (P37 to P30). If a port 3 read is performed while P3DDR bits are set to 1, the P3DR values are read directly, regardless of the actual pin states. If a port 3 read is performed while P3DDR bits are cleared to 0, the pin states are read.

P3DR is initialized to H'00 by a reset and in hardware standby mode. It retains its prior state in software standby mode.

Port 3 MOS Pull-Up Control Register (P3PCR)

Bit	7	6	5	4	3	2	1	0
	P37PCR	P36PCR	P35PCR	P34PCR	P33PCR	P32PCR	P31PCR	P30PCR
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W							

P3PCR is an 8-bit readable/writable register that controls the port 3 built-in MOS input pull-ups on a bit-by-bit basis.

In modes 2 and 3 (when EXPE = 0), the MOS input pull-up is turned on when a P3PCR bit is set to 1 while the corresponding P3DDR bit is cleared to 0 (input port setting).

P3PCR is initialized to H'00 by a reset and in hardware standby mode. It retains its prior state in software standby mode.

The MOS input pull-up function cannot be used in slave mode (when the host interface is enabled).

8.4.3 Pin Functions in Each Mode

Modes 1, 2, and 3 (EXPE = 1)

In modes 1, 2, and 3 (when EXPE = 1), port 3 pins automatically function as data I/O pins. The port 3 pin functions are shown in figure 8.10.

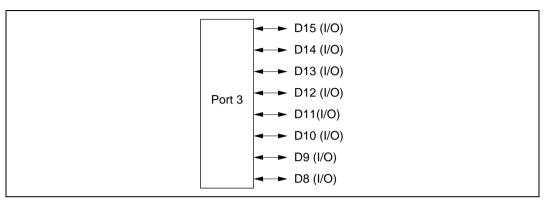


Figure 8.10 Port 3 Pin Functions (Modes 1, 2, and 3 (EXPE = 1))

Modes 2 and 3 (EXPE = 0)

In modes 2 and 3 (when EXPE = 0), port 3 functions as host interface data bus I/O pins (HDB7 to HDB0) or as I/O ports. When the HI12E bit is set to 1 in SYSCR2 and a transition is made to slave mode, port 3 functions as the host interface data bus. In slave mode, P3DR and P3DDR should be cleared to H'00. When the HI12E bit is cleared to 0, port 3 functions as an I/O port, and input or output can be specified on a bit-by-bit basis. When a bit in P3DDR is set to 1, the corresponding pin functions as an output port, and when cleared to 0, as an input port.

The port 3 pin functions are shown in figure 8.11.

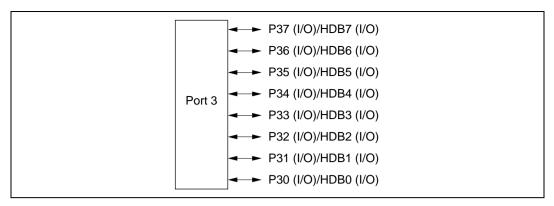


Figure 8.11 Port 3 Pin Functions (Modes 2 and 3 (EXPE = 0))

8.4.4 MOS Input Pull-Up Function

Port 3 has a built-in MOS input pull-up function that can be controlled by software. This MOS input pull-up function can be used in modes 2 and 3 (when EXPE = 0), and can be specified as on or off on a bit-by-bit basis.

When a P3DDR bit is cleared to 0 in mode 2 or 3 (when EXPE = 0), setting the corresponding P3PCR bit to 1 turns on the MOS input pull-up for that pin.

The MOS input pull-up function is in the off state after a reset and in hardware standby mode. The prior state is retained in software standby mode.

Table 8.9 summarizes the MOS input pull-up states.

Table 8.9 MOS Input Pull-Up States (Port 3)

Mode	Reset	Hardware Standby Mode	Software Standby Mode	In Other Operations
1, 2, 3 (EXPE = 1)	Off	Off	Off	Off
2, 3 (EXPE = 0)	Off	Off	On/Off	On/Off

Legend:

Off: MOS input pull-up is always off.

On/Off: On when P3DDR = 0 and P3PCR = 1; otherwise off.

8.5 Port 4

8.5.1 Overview

Port 4 is an 8-bit I/O port. Port 4 pins also function as 14-bit PWM output pins (PWX1, PWX0), 8-bit timer 0 and 1 (TMR0, TMR1) I/O pins (TMCI0, TMRI0, TMO0, TMCI1, TMRI1, TMO1), timer connection I/O pins (CSYNCI, HSYNCI, HSYNCO) (H8S/2148 Group only), SCI2 I/O pins (TxD2, RxD2, SCK2), IrDA interface I/O pins (IrTxD, IrRxD), host interface output pins (HIRQ12, HIRQ1, HIRQ11) (H8S/2148 Group and H8S/2147N only), and the IIC1 I/O pin (SDA1) (option in H8S/2148 Group and H8S/2147N only). Port 4 pin functions are the same in all operating modes.

Figure 8.12 shows the port 4 pin configuration.

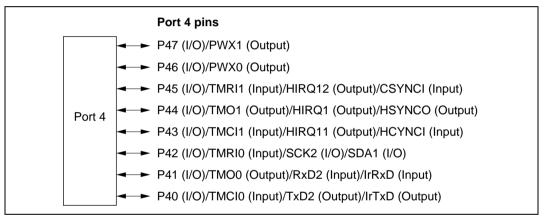


Figure 8.12 Port 4 Pin Functions

8.5.2 Register Configuration

Table 8.10 shows the port 4 register configuration.

Table 8.10 Port 4 Registers

Name	Abbreviation	R/W	Initial Value	Address*
Port 4 data direction register	P4DDR	W	H'00	H'FFB5
Port 4 data register	P4DR	R/W	H'00	H'FFB7

Note: * Lower 16 bits of the address.

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Port 4 Data Direction Register (P4DDR)

Bit	7	6	5	4	3	2	1	0
	P47DDR	P46DDR	P45DDR	P44DDR	P43DDR	P42DDR	P41DDR	P40DDR
Initial value	0	0	0	0	0	0	0	0
Read/Write	W	W	W	W	W	W	W	W

P4DDR is an 8-bit write-only register, the individual bits of which specify input or output for the pins of port 4. P4DDR cannot be read; if it is, an undefined value will be returned.

When a bit in P4DDR is set to 1, the corresponding pin functions as an output port, and when cleared to 0, as an input port.

P4DDR is initialized to H'00 by a reset and in hardware standby mode. It retains its prior state in software standby mode. As 14-bit PWM and SCI2 are initialized in software standby mode, the pin states are determined by the TMR0, TMR1, HIF, IIC1, P4DDR, and P4DR specifications.

Port 4 Data Register (P4DR)

Bit	7	6	5	4	3	2	1	0
	P47DR	P46DR	P45DR	P44DR	P43DR	P42DR	P41DR	P40DR
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W							

P4DR is an 8-bit readable/writable register that stores output data for the port 4 pins (P47 to P40). If a port 4 read is performed while P4DDR bits are set to 1, the P4DR values are read directly, regardless of the actual pin states. If a port 4 read is performed while P4DDR bits are cleared to 0, the pin states are read.

P4DR is initialized to H'00 by a reset and in hardware standby mode. It retains its prior state in software standby mode.

8.5.3 Pin Functions

Port 4 pins also function as 14-bit PWM output pins (PWX1, PWX0), 8-bit timer 0 and 1 (TMR0, TMR1) I/O pins (TMCI0, TMRI0, TMO0, TMCI1, TMRI1, TMO1), timer connection I/O pins (CSYNCI, HSYNCI, HSYNCO), SCI2 I/O pins (TxD2, RxD2, SCK2), IrDA interface I/O pins

Pin

P47/PWX1

(IrTxD, IrRxD), host interface output pins (HIRQ12, HIRQ1, HIRQ11), and the IIC1 I/O pin (SDA1). The port 4 pin functions are shown in table 8.11.

bit OEB in DACR of 14-bit PWM, and bit P47DDR.

The pin function is switched as shown below according to the combination of

Selection Method and Pin Functions

Table 8.11 Port 4 Pin Functions

	OEB		0					1
	P47DDR	0			1			_
	Pin function	P47 inpu	ıt pin	P	47 output pir	n	PWX1	output pin
P46/PWX0	•	ion is switched as shown below according to the cACR of 14-bit PWM, and bit P46DDR.					the com	bination of
	OEA		0					1
	P46DDR	0			1			_
	Pin function	P46 inpu	ıt pin	P	46 output pir	า	PWX0	output pin
P45/TMRI1/ HIRQ12/CSYNCI	The pin functi				elow accordi	ng to t	the com	bination of
	P45DDR	0 1		1	1			
	Operating mode	_		Not slave mode		е	Slave mode	
	Pin function	P45 input pin P45 output pin		n	HIRQ12	2 output pin		
			TMRI1	inpu	t pin, CSYN	CI inp	ut pin	
	When bits CC as the TMRI1							
P44/TMO1/ HIRQ1/HSYNCO	The pin functi the operating of the timer co	mode, bits O	S3 to OS	0 in ⁻	TCSR of TM			
	HOE			0				1
	OS3 to OS0		All 0			Not	t all 0	_
	P44DDR	0		1				_
	Operating mode	_	Not slav		Slave mode		_	_

Pin function

P44

output pin

HIRQ1

output pin

TMO1 output pin

HSYNCO

output pin

P44

input pin

Pin

Selection Method and Pin Functions

P43/TMCI1/ HIRQ11/HSYNCI

The pin function is switched as shown below according to the combination of the operating mode and bit P43DDR.

P43DDR	0		1		
Operating mode	_	Not slave mode	Slave mode		
Pin function	P43 input pin	P43 output pin	HIRQ11 output pin		
	TMCI1 input pin, HSYNCI input pin				

When an external clock is selected with bits CKS2 to CKS0 in TCR1 of TMR1, this pin is used as the TMCI1 input pin. It can also be used as the HSYNCI input pin.

P42/TMRI0/ SCK2/SDA1

The pin function is switched as shown below according to the combination of bit ICE in ICCR of IIC1, bits CKE1 and CKE0 in SCR of SCI2, bit C/\overline{A} in SMR of SCI2, and bit P42DDR.

ICE		0					
CKE1		(1	0			
C/A		0		1	_	0	
CKE0	()	1	_	_	0	
P42DDR	0	1	_	_	_	_	
Pin function	P42 P42 input pin output pin		SCK2 output pin	SCK2 output pin	SCK2 input pin	SDA1 I/O pin	
			TMRI0	input pin			

When this pin is used as the SDA1 I/O pin, bits CKE1 and CKE0 in SCR of SCI2 and bit C/\overline{A} in SMR of SCI2 must all be cleared to 0. SDA1 is an NMOS-only output, and has direct bus drive capability.

When bits CCLR1 and CCLR0 in TCR0 of TMR0 are set to 1, this pin is used as the TMR10 input pin.

Pin

Selection Method and Pin Functions

P41/TMO0/RxD2/ IrRxD

The pin function is switched as shown below according to the combination of bits OS3 to OS0 in TCSR of TMR0, bit RE in SCR of SCI2 and bit P41DDR.

OS3 to OS0		Not all 0		
RE	()	1	0
P41DDR	0	1	_	_
Pin function	P41 input pin	P41 output pin	RxD2/IrRxD input pin	TMO0 output pin

When this pin is used as the TMO0 output pin, bit RE in SCR of SCI2 must be cleared to 0.

IrTxD

P40/TMCI0/TxD2/ The pin function is switched as shown below according to the combination of bit TE in SCR of SCI2 and bit P40DDR.

TE	(0		
P40DDR	0	1	_	
Pin function	P40 input pin	P40 output pin	TxD2/IrTxD output pin	

When an external clock is selected with bits CKS2 to CKS0 in TCR0 of TMR0, this pin is used as the TMCI0 input pin.

8.6 Port 5

8.6.1 Overview

Port 5 is a 3-bit I/O port. Port 5 pins also function as SCI0 I/O pins (TxD0, RxD0, SCK0), and the IIC0 I/O pin (SCL0) (option in H8S/2148 Group and H8S/2147N only). In the H8S/2148 Group and H8S/2147N, P52 and SCK0 are NMOS push-pull outputs, and SCL0 is an NMOS open-drain output. Port 5 pin functions are the same in all operating modes.

Figure 8.13 shows the port 5 pin configuration.

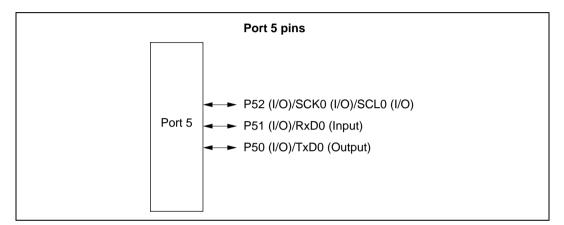


Figure 8.13 Port 5 Pin Functions

8.6.2 Register Configuration

Table 8.12 shows the port 5 register configuration.

Table 8.12 Port 5 Registers

Name	Abbreviation	R/W	Initial Value	Address*
Port 5 data direction register	P5DDR	W	H'F8	H'FFB8
Port 5 data register	P5DR	R/W	H'F8	H'FFBA

Note: * Lower 16 bits of the address.

Port 5 Data Direction Register (P5DDR)

Bit	7	6	5	4	3	2	1	0
	_	_	_	_	_	P52DDR	P51DDR	P50DDR
Initial value	1	1	1	1	1	0	0	0
Read/Write	_	_	_		_	W	W	W

P5DDR is an 8-bit write-only register, the individual bits of which specify input or output for the pins of port 5. P5DDR cannot be read; if it is, an undefined value will be returned. Bits 7 to 3 are reserved.

Setting a P5DDR bit to 1 makes the corresponding port 5 pin an output pin, while clearing the bit to 0 makes the pin an input pin.

P5DDR is initialized to HF8 by a reset and in hardware standby mode. It retains its prior state in software standby mode. As SCI0 is initialized, the pin states are determined by the IIC0 ICCR, P5DDR, and P5DR specifications.

Port 5 Data Register (P5DR)

Bit	7	6	5	4	3	2	1	0
	_	_	_	_	_	P52DR	P51DR	P50DR
Initial value	1	1	1	1	1	0	0	0
Read/Write	_	_	_	_	_	R/W	R/W	R/W

P5DR is an 8-bit readable/writable register that stores output data for the port 5 pins (P52 to P50). If a port 5 read is performed while P5DDR bits are set to 1, the P5DR values are read directly, regardless of the actual pin states. If a port 5 read is performed while P5DDR bits are cleared to 0, the pin states are read.

Bits 7 to 3 are reserved; they cannot be modified and are always read as 1.

P5DR is initialized to H'F8 by a reset and in hardware standby mode. It retains its prior state in software standby mode.



8.6.3 Pin Functions

Port 5 pins also function as SCI0 I/O pins (TxD0, RxD0, SCK0) and the IIC0 I/O pin (SCL0). The port 5 pin functions are shown in table 8.13.

Table 8.13 Port 5 Pin Functions

Pin Selection Method and Pin Functions

P52/SCK0/SCL0

The pin function is switched as shown below according to the combination of bits CKE1 and CKE0 in SCR of SCI0, bit C/\overline{A} in SMR of SCI0, bit ICE in ICCR of IIC0, and bit P52DDR.

ICE			1			
CKE1		()		1	0
C/A		0		1	_	0
CKE0	()	1	_	_	0
P52DDR	0	1	_	_	_	_
Pin function	P52 input pin	P52 output pin	SCK0 output pin	SCK0 output pin	SCK0 input pin	SCL0 I/O pin

When this pin is used as the SCL0 I/O pin, bits CKE1 and CKE0 in SCR of SCI0 and bit C/A in SMR of SCI0 must all be cleared to 0.

SCL0 is an NMOS open-drain output, and has direct bus drive capability. In the H8S/2148 Group and H8S/2147N, when set as the P52 output pin or SCK0 output pin, this pin is an NMOS push-pull output.

P51/RxD0

The pin function is switched as shown below according to the combination of bit RE in SCR of SCI0 and bit P51DDR.

RE	(0			
P51DDR	0	1	_		
Pin function	P51 input pin	P51 output pin	RxD0 input pin		

P50/TxD0

The pin function is switched as shown below according to the combination of bit TE in SCR of SCI0 and bit P50DDR.

TE	()	1
P50DDR	0	_	
Pin function	P50 input pin	P50 output pin	TxD0 output pin

8.7 Port 6

8.7.1 Overview

Port 6 is an 8-bit I/O port. Port 6 pins also function as the 16-bit free-running timer (FRT) I/O pins (FTOA, FTOB, FTIA to FTID, FTCI), timer X (TMRX) I/O pins (TMOX, TMIX) (H8S/2148 Group only), the timer Y (TMRY) input pin (TMIY), timer connection I/O pins (HFBACKI, VSYNCI, VSYNCO, VFBACKI, CLAMPO) (H8S/2148 Group only), key-sense interrupt input pins ($\overline{\text{KIN7}}$ to $\overline{\text{KIN0}}$), expansion A/D converter input pins (CIN7 to CIN0), and external interrupt input pins ($\overline{\text{IRQ7}}$, $\overline{\text{IRQ6}}$). In the H8S/2148 Group and H8S/2147N, the port 6 input level can be switched in four stages. Port 6 pin functions are the same in all operating modes.

Figure 8.14 shows the port 6 pin configuration.

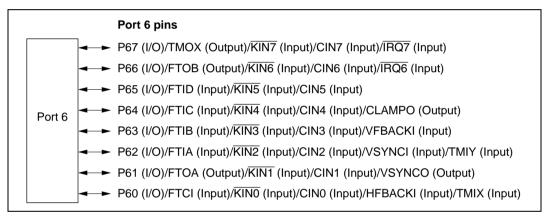


Figure 8.14 Port 6 Pin Functions

8.7.2 Register Configuration

Table 8.14 shows the port 6 register configuration.

Table 8.14 Port 6 Registers

Name	Abbreviation	R/W	Initial Value	Address*1
Port 6 data direction register	P6DDR	W	H'00	H'FFB9
Port 6 data register	P6DR	R/W	H'00	H'FFBB
Port 6 MOS pull-up control register	KMPCR	R/W	H'00	H'FFF2*2
System control register 2	SYSCR2	R/W	H'00	H'FF83

Notes: 1. Lower 16 bits of the address.

2. KMPCR has the same address as TICRR/TCORAY of TMRX/TMRY. To select KMPCR, set the HIE bit to 1 in SYSCR and set the MSTP2 bit to 0 in MSTPCRL.

Port 6 Data Direction Register (P6DDR)

Bit	7	6	5	4	3	2	1	0
	P67DDR	P66DDR	P65DDR	P64DDR	P63DDR	P62DDR	P61DDR	P60DDR
Initial value	0	0	0	0	0	0	0	0
Read/Write	W	W	W	W	W	W	W	W

P6DDR is an 8-bit write-only register, the individual bits of which specify input or output for the pins of port 6. P6DDR cannot be read; if it is, an undefined value will be returned.

Setting a P6DDR bit to 1 makes the corresponding port 6 pin an output pin, while clearing the bit to 0 makes the pin an input pin.

P6DDR is initialized to H'00 by a reset and in hardware standby mode. It retains its prior state in software standby mode.

Port 6 Data Register (P6DR)

Bit	7	6	5	4	3	2	1	0
	P67DR	P66DR	P65DR	P64DR	P63DR	P62DR	P61DR	P60DR
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W							

P6DR is an 8-bit readable/writable register that stores output data for the port 6 pins (P67 to P60). If a port 6 read is performed while P6DDR bits are set to 1, the P6DR values are read directly, regardless of the actual pin states. If a port 6 read is performed while P6DDR bits are cleared to 0, the pin states are read.

P6DR is initialized to H'00 by a reset and in hardware standby mode. It retains its prior state in software standby mode.

Port 6 MOS Pull-Up Control Register (KMPCR)

Bit	7	6	5	4	3	2	1	0
	KM7PCR	KM6PCR	KM5PCR	KM4PCR	KM3PCR	KM2PCR	KM1PCR	KM0PCR
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W							

KMPCR is an 8-bit readable/writable register that controls the port 6 built-in MOS input pull-ups on a bit-by-bit basis.

The MOS input pull-up is turned on when a KMPCR bit is set to 1 while the corresponding P6DDR bit is cleared to 0 (input port setting).

KMPCR is initialized to H'00 by a reset and in hardware standby mode. It retains its prior state in software standby mode.



System Control Register 2 (SYSCR2) (H8S/2148 Group and H8S/2147N Only)

Bit	7	6	5	4	3	2	1	0
	KWUL1	KWUL0	P6PUE		SDE	CS4E	CS3E	HI12E
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	_	R/W	R/W	R/W	R/W

SYSCR2 is an 8-bit readable/writable register that controls port 6 input level selection and the operation of host interface functions.

Only bits 7, 6, and 5 are described here. See section 18.2.2, System Control Register 2 (SYSCR2), for information on bits 4 to 0.

SYSCR2 is initialized to H'00 by a reset and in hardware standby mode.

Bits 7 and 6—Key Wakeup Level 1 and 0 (KWUL1, KWUL0): The port 6 input level setting can be changed by software, using these bits. The setting of these bits also changes the input level of the pin functions multiplexed with port 6.

Bit 7	Bit 6		
KWUL1	KWUL0	Description	
0	0	Standard input level is selected as port 6 input level	(Initial value)
	1	Input level 1 is selected as port 6 input level	
1	0	Input level 2 is selected as port 6 input level	
	1	Input level 3 is selected as port 6 input level	

Bit 5—Port 6 Input Pull-Up Extra (P6PUE): Controls and selects the current specification for the port 6 MOS input pull-up function connected by means of KMPCR settings.

Bit 5

P6PUE	Description
0	Standard current specification is selected for port 6 MOS input pull-up function (Initial value)
1	Current-limit specification is selected for port 6 MOS input pull-up function

8.7.3 Pin Functions

Port 6 pins also function as the 16-bit free-running timer (FRT) I/O pins (FTOA, FTOB, FTIA to FTID, FTCI), timer X (TMRX) I/O pins (TMOX, TMIX), the timer Y (TMRY) input pin (TMIY), timer connection I/O pins (HFBACKI, VSYNCI, VSYNCO, VFBACKI, CLAMPO), key-sense interrupt input pins ($\overline{\text{KIN7}}$ to $\overline{\text{KIN0}}$), comparator input pins (CIN7 to CIN0), and external interrupt input pins (IRO7, IRO6). In the H8S/2148 Group and H8S/2147N, the port 6 input level can be switched in four stages. The port 6 pin functions are shown in table 8.15.

Table 8.15 Port 6 Pin Functions

Pin Selection Method and Pin Functions

P67/TMOX/IBQ7/ KIN7/CIN7

The pin function is switched as shown below according to the combination of bits OS3 to OS0 in TCSR of TMRX and bit P67DDR.

OS3 to OS0	All	Not all 0	
P67DDR	0	1	_
Pin function	P67 input pin	P67 output pin	TMOX output pin
	ĪRQ7 input į	pin, KIN7 input pin, C	N7 input pin

This pin is used as the $\overline{IRQ7}$ input pin when bit IRQ7E is set to 1 in IER.

It can always be used as the $\overline{\text{KIN7}}$ or CIN7 input pin.

P66/FTOB/IRQ6/ KIN6/CIN6

The pin function is switched as shown below according to the combination of bit OEB in TOCR of the FRT and bit P66DDR.

OEB	(1	
P66DDR	0	1	_
Pin function	P66 input pin	P66 output pin	FTOB output pin
	IRQ6 input	pin, KIN6 input pin, CI	N6 input pin

This pin is used as the IRQ6 input pin when bit IRQ6E is set to 1 in IER.

It can always be used as the KIN6 or CIN6 input pin.

P65/FTID/KIN5/ CIN5

P65DDR	0	1					
Pin function	P65 input pin	P65 output pin					
	FTID input pin, KIN5 in	FTID input pin, KIN5 input pin, CIN5 input pin					

This pin can always be used as the FTID, KIN5, or CIN5 input pin.



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Selection Method and Pin Functions

P64/FTIC/KIN4/CIN4/CLAMPO

The pin function is switched as shown below according to the combination of bit CLOE in TCONRO of the timer connection function and bit P64DDR.

CLOE	(1			
P64DDR	0	0 1			
Pin function	P64 input pin	P64 output pin	CLAMPO output pin		
	FTIC input pin, KIN4 input pin, CIN4 input pin				

This pin can always be used as the FTIC, KIN4, or CIN4 input pin.

P63/FTIB/KIN3/ CIN3/VFBACKI

P63DDR	0	1				
Pin function	P63 input pin	P63 output pin				
	FTIB input pin, VFBACKI input pin, KIN3 input pin, CIN3 input pin					

This pin can always be used as the FTIB, KIN3, CIN3, or VFBACKI input pin.

P62/FTIA/TMIY/ KIN2/CIN2/ VSYNCI

P62DDR	0	1				
Pin function	P62 input pin	P62 output pin				
	FTIA input pin, VSYNCI input pin, TMIY input pin,					
	KIN2 input pin, CIN2 input pin					

This pin can always be used as the FTIA, TMIY, KIN2, CIN2, or VSYNCI input pin.

P61/FTOA/KIN1/ CIN1/VSYNCO

The pin function is switched as shown below according to the combination of bit OEA in TOCR of the FRT, bit VOE in TCONRO of the timer connection function, and bit P61DDR.

VOE		1				
OEA	(0	1	0		
P61DDR	0	1	_	_		
Pin function	P61 input pin	P61 output pin	FTOA output pin	VSYNCO output pin		
	KIN1 input pin, CIN1 input pin					

When this pin is used as the VSYNCO pin, bit OEA in TOCR of the FRT must be cleared to 0.

This pin can always be used as the KIN1 or CIN1 input pin.

Selection Method and Pin Functions

P60/FTCI/TMIX/ KIN0/CIN0/ HFBACKI

P60DDR	0	1				
Pin function	P60 input pin	P60 output pin				
	FTCI input pin, HFBACKI input pin, TMIX input pin, KIN0 input pin, CIN0 input pin					

This pin is used as the FTCI input pin when an external clock is selected with bits CKS1 and CKS0 in TCR of the FRT.

It can always be used as the TMIX, KINO, CINO, or HFBACKI input pin.

8.7.4 MOS Input Pull-Up Function

Port 6 has a built-in MOS input pull-up function that can be controlled by software. This MOS input pull-up function can be used in any operating mode, and can be specified as on or off on a bit-by-bit basis.

When a P6DDR bit is cleared to 0, setting the corresponding KMPCR bit to 1 turns on the MOS input pull-up for that pin. The MOS input pull-up current specification can be changed by means of the P6PUE bit. When a pin is designated as an on-chip supporting module output pin, the MOS input pull-up is always off.

The MOS input pull-up function is in the off state after a reset and in hardware standby mode. The prior state is retained in software standby mode.

Table 8.16 summarizes the MOS input pull-up states.

Table 8.16 MOS Input Pull-Up States (Port 6)

Mode	Reset	Hardware Standby Mode	Software Standby Mode	In Other Operations
1, 2, 3	Off	Off	On/Off	On/Off

Legend:

Off: MOS input pull-up is always off.

On/Off: On when P6DDR = 0 and KMPCR = 1; otherwise off.

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8.8 Port 7

8.8.1 Overview

Port 7 is an 8-bit input port. Port 7 pins also function as the A/D converter analog input pins (AN0 to AN7) and D/A converter analog output pins (DA0, DA1). Port 7 functions are the same in all operating modes.

Figure 8.15 shows the port 7 pin configuration.

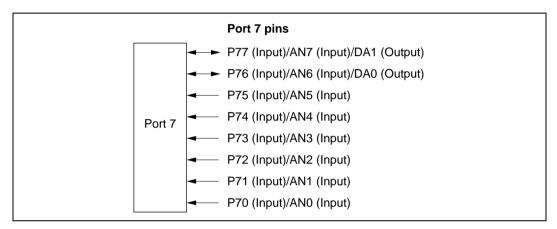


Figure 8.15 Port 7 Pin Functions

8.8.2 Register Configuration

Table 8.17 shows the port 7 register configuration. Port 7 is an input-only port, and does not have a data direction register or data register.

Table 8.17 Port 7 Registers

Name	Abbreviation	R/W	Initial Value	Address*1
Port 7 input data register	P7PIN	R	Undefined	H'FFBE*2

Notes: 1. Lower 16 bits of the address.

2. P7PIN has the same address as PBDDR.

Port 7 Input Data Register (P7PIN)

Bit	7	6	5	4	3	2	1	0
	P77PIN	P76PIN	P75PIN	P74PIN	P73PIN	P72PIN	P71PIN	P70PIN
Initial value	*	*	*	*	*	*	*	*
Read/Write	R	R	R	R	R	R	R	R

Note: * Determined by the state of pins P77 to P70.

When a P7PIN read is performed, the pin states are always read.

P7PIN has the same address as PBDDR; if a write is performed, data will be written into PBDDR and the port B setting will be changed.

8.8.3 **Pin Functions**

Port 7 pins also function as the A/D converter analog input pins (AN0 to AN7) and D/A converter analog output pins (DA0, DA1).



8.9 Port 8

8.9.1 Overview

Port 8 is an 8-bit I/O port. Port 8 pins also function as SCI1 I/O pins (TxD1, RxD1, SCK1), the IIC1 I/O pin (SCL1) (option in H8S/2148 Group and H8S/2147N only), HIF I/O pins ($\overline{\text{CS2}}$, GA20, HA0, HIFSD) (H8S/2148 Group and H8S/2147N only), and external interrupt input pins ($\overline{\text{IRQ5}}$ to $\overline{\text{IRQ3}}$). Port 8 pin functions are the same in all operating modes. Figure 8.16 shows the port 8 pin configuration.

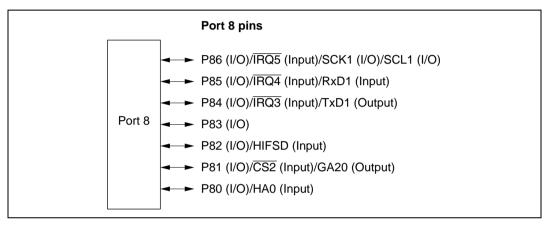


Figure 8.16 Port 8 Pin Functions

8.9.2 Register Configuration

Table 8.18 summarizes the port 8 registers.

Table 8.18 Port 8 Registers

Name	Abbreviation	R/W	Initial Value	Address*1
Port 8 data direction register	P8DDR	W	H'80	H'FFBD*2
Port 8 data register	P8DR	R/W	H'80	H'FFBF

Notes: 1. Lower 16 bits of the address.

2. P8DDR has the same address as PBPIN.

Port 8 Data Direction Register (P8DDR)

Bit	7	6	5	4	3	2	1	0
	_	P86DDR	P85DDR	P84DDR	P83DDR	P82DDR	P81DDR	P80DDR
Initial value	1	0	0	0	0	0	0	0
Read/Write	_	W	W	W	W	W	W	W

P8DDR is a 7-bit write-only register, the individual bits of which specify input or output for the pins of port 8. P8DDR has the same address as PBPIN, and if read, the port B state will be returned.

Setting a P8DDR bit to 1 makes the corresponding port 8 pin an output pin, while clearing the bit to 0 makes the pin an input pin.

P8DDR is initialized to H'80 by a reset and in hardware standby mode. It retains its prior state in software standby mode.

Port 8 Data Register (P8DR)

Bit	7	6	5	4	3	2	1	0
	_	P86DR	P85DR	P84DR	P83DR	P82DR	P81DR	P80DR
Initial value	1	0	0	0	0	0	0	0
Read/Write	_	R/W						

P8DR is a 7-bit readable/writable register that stores output data for the port 8 pins (P86 to P80). If a port 8 read is performed while P8DDR bits are set to 1, the P8DR values are read directly, regardless of the actual pin states. If a port 8 read is performed while P8DDR bits are cleared to 0, the pin states are read.

P8DR is initialized to H'80 by a reset and in hardware standby mode. It retains its prior state in software standby mode.

8.9.3 Pin Functions

Port 8 pins also function as SCI1 I/O pins (TxD1, RxD1, SCK1), the IIC1 I/O pin (SCL1), HIF I/O pins ($\overline{\text{CS2}}$, GA20, HA0, HIFSD), and external interrupt input pins ($\overline{\text{IRQ5}}$ to $\overline{\text{IRQ3}}$). The port 8 pin functions are shown in table 8.19.

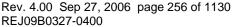




Table 8.19 Port 8 Pin Functions

Pin

Selection Method and Pin Functions

P86/IRQ5/SCK1/ SCL1

The pin function is switched as shown below according to the combination of bits CKE1 and CKE0 in SCR of SCI1, bit C/\overline{A} in SMR of SCI1, bit ICE in ICCR of IIC1, and bit P86DDR.

ICE		0 1						
CKE1		()		1	0		
C/A		0	_	0				
CKE0	()	1	_	_	0		
P86DDR	0	1	_	_	_	_		
Pin function	P86 input pin	P86 output pin	SCK1 output pin	SCK1 output pin	SCK1 input pin	SCL1 I/O pin		
		IRQ5 input pin						

When the IRQ5E bit in IER is set to 1, this pin is used as the $\overline{\text{IRQ5}}$ input pin. When this pin is used as the SCL1 I/O pin, bits CKE1 and CKE0 in SCR of SCI1 and bit C/ $\overline{\text{A}}$ in SMR of SCI1 must all be cleared to 0. SCL1 is an NMOS-only output, and has direct bus drive capability.

P85/IRQ4/RxD1

The pin function is switched as shown below according to the combination of bit RE in SCR of SCI1 and bit P85DDR.

RE	(1	
P85DDR	0	0 1	
Pin function	P85 input pin	P85 output pin	RxD1 input pin
		IRQ4 input pin	

When the IRQ4E bit in IER is set to 1, this pin is used as the IRQ4 input pin.

P84/IRQ3/TxD1

The pin function is switched as shown below according to the combination of bit TE in SCR of SCI1 and bit P84DDR.

TE	(1			
P84DDR	0	0 1			
Pin function	P84 input pin	P84 input pin P84 output pin			
		ĪRQ3 input pin			

When the IRQ3E bit in IER is set to 1, this pin is used as the IRQ3 input pin.

Pin Selection Method and Pin Functions

P83 The pin function is switched as shown below according to bit P83DDR.

P83DDR	0	1
Pin function	P83 input pin	P83 output pin

P82/HIFSD

The pin function is switched as shown below according to the combination of operating mode, bit SDE in SYSCR2, and bit P82DDR.

Operating mode	Not slav	e mode		Slave mode	
SDE	_	_	(1	
P82DDR	0	1	0	1	_
Pin function	P82 input pin	P82 output pin	P82 input pin	P82 output pin	HIFSD input pin

P81/GA20/CS2

The pin function is switched as shown below according to the combination of operating mode, bit CS2E in SYSCR, bit FGA20E in HICR of the HIF, and bit P81DDR.

Operating mode	Not slave mode		Slave mode				
FGA20E	_		0			1	
CS2E		_	0		1	_	
P81DDR	0	1	0	1	_	0	1
Pin function	P81 input pin	P81 output pin	P81 input pin	P81 output pin	CS2 input pin	P81 input pin	GA20 output pin

This pin should be used as the GA20 output pin or $\overline{\text{CS2}}$ input pin only in mode 2 or 3 (EXPE = 0).

P80/HA0

The pin function is switched as shown below according to the combination of operating mode and bit P80DDR.

Operating mode	Not slav	Slave mode			
P80DDR	0	0 1			
Pin function	P80 input pin	P80 output pin	HA0 input pin		



8.10 Port 9

8.10.1 Overview

Port 9 is an 8-bit I/O port. Port 9 pins also function as external interrupt input pins ($\overline{IRQ0}$ to $\overline{IRQ2}$), the A/D converter external trigger input pin (\overline{ADTRG}), host interface input pins ($\overline{ECS2}$, $\overline{CS1}$, \overline{IOW} , \overline{IOR}) (H8S/2148 Group and H8S/2147N only), the IICO I/O pin (SDA0) (option in H8S/2148 Group and H8S/2147N only), the subclock input pin (EXCL), bus control signal I/O pins ($\overline{AS/IOS}$, \overline{RD} , \overline{HWR} , \overline{LWR} , \overline{WAIT}), and the system clock (ϕ) output pin. In H8S/2148 Group and H8S/2147N, P97 is an NMOS push-pull output. SDA0 is an NMOS open-drain output, and has direct bus drive capability.

Figure 8.17 shows the port 9 pin configuration.

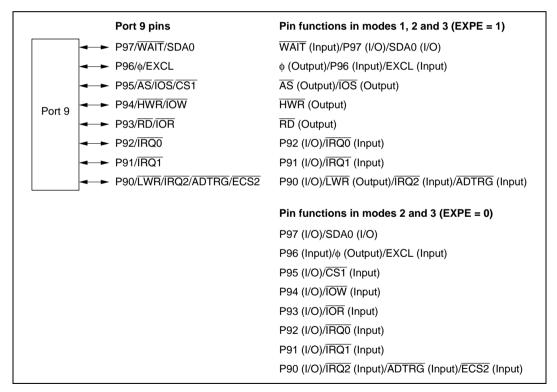


Figure 8.17 Port 9 Pin Functions

8.10.2 Register Configuration

Table 8.20 summarizes the port 9 registers.

Table 8.20 Port 9 Registers

Name	Abbreviation	R/W	Initial Value	Address*1
Port 9 data direction register	P9DDR	W	H'40/H'00*2	H'FFC0
Port 9 data register	P9DR	R/W	H'00	H'FFC1

Notes: 1. Lower 16 bits of the address.

2. Initial value depends on the mode.

Port 9 Data Direction Register (P9DDR)

Bit	7	6	5	4	3	2	1	0
	P97DDR	P96DDR	P95DDR	P94DDR	P93DDR	P92DDR	P91DDR	P90DDR
Mode 1	<u> </u>							
Initial value	0	1	0	0	0	0	0	0
Read/Write	W	W	W	W	W	W	W	W
Modes 2 and 3								
Initial value	0	0	0	0	0	0	0	0
Read/Write	W	W	W	W	W	W	W	W

P9DDR is an 8-bit write-only register, the individual bits of which specify input or output for the pins of port 9. P9DDR cannot be read; if it is, an undefined value will be returned.

P9DDR is initialized to H'40 (mode 1) or H'00 (modes 2 and 3) by a reset and in hardware standby mode. It retains its prior state in software standby mode.

• Modes 1, 2 and 3 (EXPE = 1)

Pin P97 functions as a bus control input (WAIT), the IIC0 I/O pin (SDA0), or an I/O port, according to the wait mode setting. When P97 functions as an I/O port, it becomes an output port when P97DDR is set to 1, and an input port when P97DDR is cleared to 0.

Pin P96 functions as the ϕ output pin when P96DDR is set to 1, and as the subclock input (EXCL) or an input port when P96DDR is cleared to 0.

Pins P95 to P93 automatically become bus control outputs ($\overline{AS/IOS}$, \overline{HWR} , \overline{RD}), regardless of the input/output direction indicated by P95DDR to P93DDR.

Pins P92 and P91 become output ports when P92DDR and P91DDR are set to 1, and input ports when P92DDR and P91DDR are cleared to 0.



When the ABW bit in WSCR is cleared to 0, pin P90 becomes a bus control output (\overline{LWR}) , regardless of the input/output direction indicated by P90DDR. When the ABW bit is 1, pin P90 becomes an output port if P90DDR is set to 1, and an input port if P90DDR is cleared to 0.

• Modes 2 and 3 (EXPE = 0)

When the corresponding P9DDR bits are set to 1, pin P96 functions as the ϕ output pin and pins P97 and P95 to P90 become output ports. When P9DDR bits are cleared to 0, the corresponding pins become input ports.

Port 9 Data Register (P9DR)

Bit	7	6	5	4	3	2	1	0
	P97DR	P96DR	P95DR	P94DR	P93DR	P92DR	P91DR	P90DR
Initial value	0	*	0	0	0	0	0	0
Read/Write	R/W	R	R/W	R/W	R/W	R/W	R/W	R/W

Note: * Determined by the state of pin P96.

P9DR is an 8-bit readable/writable register that stores output data for the port 9 pins (P97 to P90). With the exception of P96, if a port 9 read is performed while P9DDR bits are set to 1, the P9DR values are read directly, regardless of the actual pin states. If a port 9 read is performed while P9DDR bits are cleared to 0, the pin states are read.

P9DR is initialized to H'00 by a reset and in hardware standby mode. It retains its prior state in software standby mode.

8.10.3 Pin Functions

Port 9 pins also function as external interrupt input pins ($\overline{IRQ0}$ to $\overline{IRQ2}$), the A/D converter trigger input pin (\overline{ADTRG}), HIF input pins ($\overline{ECS2}$, $\overline{CS1}$, \overline{IOW} , \overline{IOR}), the IIC0 I/O pin ($\overline{SDA0}$), the subclock input pin (EXCL), bus control signal I/O pins ($\overline{AS/IOS}$, \overline{RD} , \overline{HWR} , \overline{LWR} , \overline{WAIT}), and the system clock (ϕ) output pin. The pin functions differ between the mode 1, 2, and 3 (EXPE = 1) expanded modes and the mode 2 and 3 (EXPE = 0) single-chip modes. The port 9 pin functions are shown in table 8.21.

Table 8.21 Port 9 Pin Functions

Pin Selection Method and Pin Functions

P97/WAIT/SDA0

The pin function is switched as shown below according to the combination of operating mode, bit WMS1 in WSCR, bit ICE in ICCR of IIC0, and bit P97DDR.

Operating mode	Mod	des 1, 2, 3	3 (EXPE :	Modes	2, 3 (EXI	PE = 0)	
WMS1	0 1					_	
ICE	()	1	_	0		1
P97DDR	0	1	_	_	0	1	_
Pin function	P97 input pin	P97 output pin	SDA0 I/O pin	WAIT input pin	P97 input pin	P97 output pin	SDA0 I/O pin

When this pin is set as the P97 output pin in the H8S/2148 Group and H8S/2147N, it is an NMOS push-pull output. SDA0 is an NMOS open-drain output, and has direct bus drive capability.

P96/ø/EXCL

The pin function is switched as shown below according to the combination of bit EXCLE in LPWRCR and bit P96DDR.

P96DDR	(0			
EXCLE	0	0 1			
Pin function	P96 input pin	EXCL input pin	φ output pin		

When this pin is used as the EXCL input pin, P96DDR should be cleared to 0.

P95/AS/IOS/CS1

The pin function is switched as shown below according to the combination of operating mode, bit IOSE in SYSCR, bit HI12E in SYSCR2, and bit P95DDR.

Operating mode		1, 2, 3 E = 1)	Modes 2, 3 (EXPE = 0)			
HI12E	_	_	()	1	
P95DDR	_	_	0	1	_	
IOSE	0	1	_	_		
Pin function	AS output pin	IOS output pin	P95 input pin	P95 output pin	CS1 input pin	

Pin	Selection Me	thod and Pin Fu	unctions				
P94/HWR/IOW		on is switched as de, bit HI12E in S			combination of		
	Operating mode	Modes 1, 2, 3 (EXPE = 1)	Мо	des 2, 3 (EXPE =	= 0)		
	HI12E	_	()	1		
	P94DDR	_	0	1	_		
	Pin function	HWR output pin	P94 input pin	P94 output pin	ĪOW input pin		
P93/RD/IOR		on is switched as de, bit HI12E in S			combination of		
	Operating mode	Modes 1, 2, 3 (EXPE = 0 (EXPE = 1)			= 0)		
	HI12E	_	()	1		
	P93DDR	_	0	1	_		
	Pin function	RD output pin	P93 input pin	P93 output pin	IOR input pin		
P92/ĪRQ0	P92DDR	(1		
	Pin function	P92 in	•		tput pin		
			IRQ0 ir	nput pin			
	When bit IRQ	0E in IER is set t	to 1, this pin is u	sed as the IRQ0	input pin.		
P91/IRQ1	P91DDR	()	1	1		
	Pin function	P91 in	put pin	P91 ou	tput pin		
			ĪRQ1 ir	input pin			
	When bit IRQ	1E in IER is set t	o 1, this pin is u	sed as the IRQ1	input pin.		

Selection Method and Pin Functions

P90/LWR/IRQ2/ ADTRG/ECS2

The pin function is switched as shown below according to the combination of operating mode, bit ABW in WSCR, bits HI12E and CS2E in SYSCR2, bit FGA20E in HICR, and bit P90DDR.

Operating mode	Modes	1, 2, 3 (EX	PE = 1)	Modes 2, 3 (EXPE = 0)			
ABW	0		1	_			
HI12E		_		Any	one 0	1	
FGA20E		_			1		
CS2E		_			1		
P90DDR	_	0	1	0	1	_	
Pin function	LWR	P90	P90	P90	P90	ECS2	
	output pin	input pin	output pin	input pin	output pin	input pin	
			IRQ2 input	pin, ADTF	RG input pir	1	

When the IRQ2E bit in IER is set to 1 in mode 1, 2, or 3 (EXPE = 1) with the ABW bit in WSCR set to 1, or in mode 2 and 3 (EXPE = 0), this pin is used as the $\overline{\text{IRQ2}}$ input pin.

When TRGS1 and TRGS0 in ADCR of the A/D converter are both set to 1, this pin is used as the \overline{ADTRG} input pin.

8.11 Port A

8.11.1 Overview

Port A is an 8-bit I/O port. Port A pins also function as keyboard buffer controller I/O pins (PS2AC, PS2AD, PS2BC, PS2BD, PS2CC, PS2CD) (H8S/2148 Group and H8S/2147N only), key-sense interrupt input pins (KIN15 to KIN8), expansion A/D converter input pins (CIN15 to CIN8), and address output pins (A23 to A16). Port A pin functions are the same in all operating modes. Figure 8.18 shows the port A pin configuration.

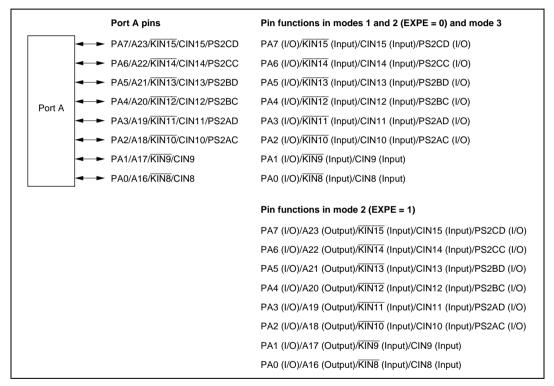


Figure 8.18 Port A Pin Functions

8.11.2 Register Configuration

Table 8.22 summarizes the port A registers.

Table 8.22 Port A Registers

Name	Abbreviation	R/W	Initial Value	Address*1
Port A data direction register	PADDR	W	H'00	H'FFAB*2
Port A output data register	PAODR	R/W	H'00	H'FFAA
Port A input data register	PAPIN	R	Undefined	H'FFAB*2

Notes: 1. Lower 16 bits of the address.

2. PADDR and PAPIN have the same address.

Port A Data Direction Register (PADDR)

Bit	7	6	5	4	3	2	1	0
	PA7DDR	PA6DDR	PA5DDR	PA4DDR	PA3DDR	PA2DDR	PA1DDR	PA0DDR
Initial value	0	0	0	0	0	0	0	0
Read/Write	W	W	W	W	W	W	W	W

PADDR is an 8-bit write-only register, the individual bits of which specify input or output for the pins of port A.

Setting a PADDR bit to 1 makes the corresponding port A pin an output pin, while clearing the bit to 0 makes the pin an input pin.

PADDR is initialized to H'00 by a reset and in hardware standby mode. It retains its prior state in software standby mode.

Port A Output Data Register (PAODR)

Bit	7	6	5	4	3	2	1	0
	PA7ODR	PA6ODR	PA5ODR	PA4ODR	PA3ODR	PA2ODR	PA10DR	PA0ODR
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W							

PAODR is an 8-bit readable/writable register that stores output data for the port A pins (PA7 to PA0). PAODR can always be read or written to, regardless of the contents of PADDR.

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PAODR is initialized to H'00 by a reset and in hardware standby mode. It retains its prior state in software standby mode.

Port A Input Data Register (PAPIN)

Bit	7	6	5	4	3	2	1	0
	PA7PIN	PA6PIN	PA5PIN	PA4PIN	PA3PIN	PA2PIN	PA1PIN	PA0PIN
Initial value	*	*	*	*	*	*	*	*
Read/Write	R	R	R	R	R	R	R	R

Note: * Determined by the state of pins PA7 to PA0.

Reading PAPIN always returns the pin states.

8.11.3 Pin Functions

Port A pins also function as keyboard buffer controller I/O pins (PS2AC, PS2AD, PS2BC, PS2BD, PS2CC, PS2CD), key-sense interrupt input pins (KIN15 to KIN8), expansion A/D converter input pins (CIN15 to CIN8), and address output pins (A23 to A16). The port A pin functions are shown in table 8.23.

Table 8.23 Port A Pin Functions

Pin Selection Method and Pin Functions

PA7/A23/PS2CD/ KIN15/CIN15 The pin function is switched as shown below according to the combination of operating mode, the KBIOE bit in KBCR2H of the keyboard buffer controller, the IOSE bit in SYSCR, and bit PA7DDR.

Operating mode	Modes 1	I, 2 (EXP	E = 0), 3	Mode 2 (EXPE = 1)			
KBIOE	0 1				1		
PA7DDR	0	1	_	0		1	
IOSE	_		_	_	0 1		_
Pin function	PA7 input pin	PA7 output pin IN15 inpu	PS2CD output pin It pin, CIN	A23 output pin pin, PS20	PA7 output pin CD input p	PS2CD output pin oin	

When the IICS bit in STCR is set to 1, this pin functions as a bus buffer. This pin can always be used as the PS2CD, $\overline{\text{KIN15}}$, or CIN15 input pin.

Selection Method and Pin Functions

PA6/A22/PS2CC/ KIN14/CIN14

The pin function is switched as shown below according to the combination of operating mode, the KBIOE bit in KBCR2H of the keyboard buffer controller, the IOSE bit in SYSCR, and bit PA6DDR.

Operating mode	Modes 1	, 2 (EXP	E = 0), 3	Mode 2 (EXPE = 1)						
KBIOE	0 1			0			1			
PA6DDR	0	1	_	0		_				
IOSE	_		_	_	0	1	_			
Pin function	PA6 input pin	PA6 PA6 PS2CC PA6 A22 PA6 PS2 input output output input output output output output								

When the IICS bit in STCR is set to 1, this pin functions as a bus buffer. This pin can always be used as the PS2CC, KIN14, or CIN14 input pin.

PA5/A21/PS2BD/ KIN13/CIN13

The pin function is switched as shown below according to the combination of operating mode, the KBIOE bit in KBCR1H of the keyboard buffer controller, the IOSE bit in SYSCR, and bit PA5DDR.

Operating mode	Modes 1	, 2 (EXP	E = 0), 3	Mode 2 (EXPE = 1)					
KBIOE	()	1			1			
PA5DDR	0	1	_	0	,	_			
IOSE	_	_	_	_	0 1		_		
Pin function	PA5 input pin	PA5 PA5 PS2BD PA5 A21 PA5 PS2 input output output input output output output output							

When the IICS bit in STCR is set to 1, this pin functions as a bus buffer. This pin can always be used as the PS2BD, KIN13, or CIN13 input pin.

Selection Method and Pin Functions

PA4/A20/PS2BC/ KIN12/CIN12

The pin function is switched as shown below according to the combination of operating mode, the KBIOE bit in KBCR1H of the keyboard buffer controller, the IOSE bit in SYSCR, and bit PA4DDR.

Operating mode	Modes 1	, 2 (EXP	E = 0), 3	Mode 2 (EXPE = 1)						
KBIOE	0 1				1					
PA4DDR	0	1	_	0	1		_			
IOSE	_		_	_	0 1		_			
Pin function	PA4 input pin	PA4 output pin	PS2BC output pin	PA4 input pin	A20 output pin	PA4 output pin	PS2BC output pin			
	K	KIN12 input pin, CIN12 input pin, PS2BC input pin								

When the IICS bit in STCR is set to 1, this pin functions as a bus buffer. This pin can always be used as the PS2BC, KIN12, or CIN12 input pin.

PA3/A19/PS2AD/ KIN11/CIN11

The pin function is switched as shown below according to the combination of operating mode, the KBIOE bit in KBCR0H of the keyboard buffer controller, the IOSE bit in SYSCR, and bit PA3DDR.

	·									
Operating mode	Modes 1	I, 2 (EXP	E = 0), 3	Mode 2 (EXPE = 1)						
KBIOE	()	1		1					
PA3DDR	0	1	_	0	1		_			
IOSE	_	_	_	_	0	1	_			
Pin function	PA3 input pin	input output output input output output								
	K	KIN11 input pin, CIN11 input pin, PS2AD input pin								

This pin can always be used as the PS2AD, KIN11, or CIN11 input pin.

Selection Method and Pin Functions

PA2/A18/PS2AC/ KIN10/CIN10

The pin function is switched as shown below according to the combination of operating mode, the KBIOE bit in KBCR0H of the keyboard buffer controller, the IOSE bit in SYSCR, and bit PA2DDR.

Operating mode	Modes 1, 2 (EXPE = 0), 3			Mode 2 (EXPE = 1)			
KBIOE	()	1		0	1	
PA2DDR	0	1	_	0	,	1	
IOSE	_			_	0 1		_
Pin function	PA2 input pin	PA2 output pin IN10 inpu	PS2AC output pin It pin, CIN	PA2 input pin 10 input	A18 output pin pin, PS2A	PA2 output pin C input p	PS2AC output pin

This pin can always be used as the PS2AC, KIN10, or CIN10 input pin.

PA1/A17/KIN9/ CIN9

The pin function is switched as shown below according to the combination of operating mode, the IOSE bit in SYSCR, and bit PA1DDR.

	3 3						
Operating mode	Modes 1, 2 (EXPE = 0), 3		Mode 2 (EXPE = 1)				
PA1DDR	0	1	0	1			
IOSE		_	_	0 1			
Pin function	PA1 input pin	PA1 output pin	PA1 A17 PA1 pin input pin output pin output pin				
		KIN9 input pin, CIN9 input pin					

This pin can always be used as the KIN9 or CIN9 input pin.

PA0/A16/KIN8/ CIN8

The pin function is switched as shown below according to the combination of operating mode, the IOSE bit in SYSCR, and bit PA0DDR.

Operating mode	Mode 2 (EXPE	es 1, E = 0), 3	Mode 2 (EXPE = 1)				
PA0DDR	0	1	0	1			
IOSE	_	_	_	0	0 1		
Pin function	PA0 input pin						
		KIN8 input pin, CIN8 input pin					

This pin can always be used as the $\overline{\text{KIN8}}$ or CIN8 input pin.



8.11.4 MOS Input Pull-Up Function

Port A has a built-in MOS input pull-up function that can be controlled by software. This MOS input pull-up function can be used in any operating mode, and can be specified as on or off on a bit-by-bit basis.

When a PADDR bit is cleared to 0, setting the corresponding PAODR bit to 1 turns on the MOS input pull-up for that pin. The MOS input pull-up for pins PA7 to PA4 is always off when IICS is set to 1. When the keyboard buffer control pin function is selected for pins PA7 to PA2, the MOS input pull-up is always off.

The MOS input pull-up function is in the off state after a reset and in hardware standby mode. The prior state is retained in software standby mode.

Table 8.24 summarizes the MOS input pull-up states.

Table 8.24 MOS Input Pull-Up States (Port A)

Mode	Reset	Hardware Standby Mode	Software Standby Mode	In Other Operations
1, 2, 3	Off	Off	On/Off	On/Off

Legend:

Off: MOS input pull-up is always off.

On/Off: On when PADDR = 0 and PAODR = 1; otherwise off.

8.12 Port B

8.12.1 Overview

Port B is an 8-bit I/O port. Port B pins also have HIF input/output pins (CS3, CS4, HIRQ3, HIRQ4) (H8S/2148 Group and H8S/2147N only), and a data bus input/output function (as D7 to D0). The pin functions depend on the operating mode.

Figure 8.19 shows the port B pin configuration.

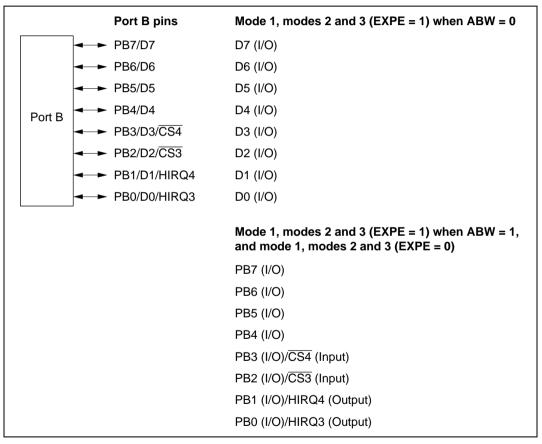


Figure 8.19 Port B Pin Functions

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8.12.2 Register Configuration

Table 8.25 summarizes the port B registers.

Table 8.25 Port B Registers

Name	Abbreviation	R/W	Initial Value	Address*1
Port B data direction register	PBDDR	W	H'00	H'FFBE*2
Port B output data register	PBODR	R/W	H'00	H'FFBC
Port B input data register	PBPIN	R	Undefined	H'FFBD*3

Notes: 1. Lower 16 bits of the address.

- 2. PBDDR has the same address as P7PIN.
- PBPIN has the same address as P8DDR.

Port B Data Direction Register (PBDDR)

Bit	7	6	5	4	3	2	1	0
	PB7DDR	PB6DDR	PB5DDR	PB4DDR	PB3DDR	PB2DDR	PB1DDR	PB0DDR
Initial value	0	0	0	0	0	0	0	0
Read/Write	W	W	W	W	W	W	W	W

PBDDR is an 8-bit write-only register, the individual bits of which specify input or output for the pins of port B. PBDDR has the same address as P7PIN, and if read, the port 7 pin states will be returned.

Setting a PBDDR bit to 1 makes the corresponding port B pin an output pin, while clearing the bit to 0 makes the pin an input pin.

PBDDR is initialized to H'00 by a reset and in hardware standby mode. It retains its prior state in software standby mode.

• Modes 1, 2 and 3 (EXPE = 1)

When the ABW bit in WSCR is cleared to 0, port B pins automatically become data I/O pins (D7 to D0), regardless of the input/output direction indicated by PBDDR. When the ABW bit is 1, a port B pin becomes an output port if the corresponding PBDDR bit is set to 1, and an input port if the bit is cleared to 0.

Data I/O pins go to the high-impedance state after a reset, and in hardware standby mode or software standby mode.

• Modes 2 and 3 (EXPE = 0)

A port B pin becomes an output port if the corresponding PBDDR bit is set to 1, and an input port if the bit is cleared to 0.

Port B Output Data Register (PBODR)

Bit	7	6	5	4	3	2	1	0
	PB7ODR	PB6ODR	PB5ODR	PB4ODR	PB3ODR	PB2ODR	PB1ODR	PB0ODR
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W							

PBODR is an 8-bit readable/writable register that stores output data for the port B pins (PB7 to PB0). PBODR can always be read or written to, regardless of the contents of PBDDR.

PBODR is initialized to H'00 by a reset and in hardware standby mode. It retains its prior state in software standby mode.

Port B Input Data Register (PBPIN)

Bit	7	6	5	4	3	2	1	0
	PB7PIN	PB6PIN	PB5PIN	PB4PIN	PB3PIN	PB2PIN	PB1PIN	PB0PIN
Initial value	*	*	*	*	*	*	*	*
Read/Write	R	R	R	R	R	R	R	R

Note: * Determined by the state of pins PB7 to PB0.

Reading PBPIN always returns the pin states.

PBPIN has the same address as P8DDR. If a write is performed, data will be written to P8DDR and the port 8 settings will change.

8.12.3 Pin Functions

Port B pins also function as HIF input pins ($\overline{CS3}$, $\overline{CS4}$, HIRQ3, HIRQ4) [H8S/2148 Group and H8S/2147N only] and data bus I/O pins (D7 to D0). The port B pin functions are shown in table 8.26.

Table 8.26 Port B Pin Functions

Selection Method and Pin Fu	nctions
	Selection Method and Pin Fui

PB7/D7

The pin function is switched as shown below according to the combination of the operating mode, bit PB7DDR, and bit ABW in WSCR.

Operating mode		Modes 1, 2, 3 (EXPE = 1)	Modes 2, 3 (EXPE = 0)		
ABW	0		1	_	
PB7DDR	_	0	1	0	1
Pin function	D7 I/O pin	PB7 input pin	PB7 output pin	PB7 input pin	PB7 output pin

PB6/D6

The pin function is switched as shown below according to the combination of the operating mode, bit PB6DDR, and bit ABW in WSCR.

the operating mead, but Bobbit, and but the work							
Operating mode	Modes 1, 2, 3 (EXPE = 1)				es 2, 3 E = 0)		
ABW	0		1	_	_		
PB6DDR	_	0	1	0	1		
Pin function	D6 I/O pin	PB6 input pin	PB6 output pin	PB6 input pin	PB6 output pin		

PB5/D5

The pin function is switched as shown below according to the combination of the operating mode, bit PB5DDR, and bit ABW in WSCR.

Operating mode		Modes 1, 2, 3 (EXPE = 1)		s 2, 3 E = 0)	
ABW	0		1	_	_
PB5DDR	_	0	1	0	1
Pin function	D5 I/O pin	PB5 input pin	PB5 output pin	PB5 input pin	PB5 output pin

Selection Method and Pin Functions

PB4/D4

The pin function is switched as shown below according to the combination of the operating mode, bit PB4DDR, and bit ABW in WSCR.

Operating mode	Modes 1, 2, 3 (EXPE = 1)			Modes 2, 3 (EXPE = 0)		
ABW	0	,	I	_		
PB4DDR	_	0 1		0	1	
Pin function	D4 I/O pin	PB4 PB4 input pin output pin		PB4 input pin	PB4 output pin	

PB3/D3/CS4

The pin function is switched as shown below according to the combination of the operating mode, bits HI12E and CS4E in SYSCR2, bit ABW in WSCR, and bit PB3DDR.

Operating mode	Modes 1, 2, 3 (EXPE = 1)			Modes 2, 3 (EXPE = 0)		
HI12E	_			Either cleared to 0		1
CS4E	_					1
ABW	0	1		_		_
PB3DDR	_	0 1		0	1	_
Pin function	D3 I/O pin	PB3 PB3 input pin output pin		PB3 input pin	PB3 output pin	CS4 input pin

PB2/D2/CS3

The pin function is switched as shown below according to the combination of the operating mode, bits HI12E and CS3E in SYSCR2, bit ABW in WSCR, and bit PB2DDR

Operating mode	Modes 1, 2, 3 (EXPE = 1)			Modes 2, 3 (EXPE = 0)		
HI12E	_			Either cleared to 0		1
CS3E	_					1
ABW	0	1		_		_
PB2DDR	_	0 1		0	1	_
Pin function	D2 I/O pin	PB2 PB2 output pin		PB2 input pin	PB2 output pin	CS3 input pin



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Selection Method and Pin Functions

PB1/D1/HIRQ4

The pin function is switched as shown below according to the combination of the operating mode, bits HI12E and CS4E in SYSCR2, bit ABW in WSCR, and bit PB1DDR.

Operating mode	Modes 1, 2, 3 (EXPE = 1)			Modes 2, 3 (EXPE = 0)		
HI12E	_			Either cleared to 0		1
CS4E	_					1
ABW	0	1		_	_	_
PB1DDR	_	0 1		0	1	_
Pin function	D1 I/O pin	PB1 PB1 input pin output pin		PB1	PB1	HIRQ4

PB0/D0/HIRQ3

The pin function is switched as shown below according to the combination of the operating mode, bits HI12E and CS3E in SYSCR2, bit ABW in WSCR, and bit PB0DDR.

Operating mode	Modes 1, 2, 3 (EXPE = 1)			Modes 2, 3 (EXPE = 0)		
HI12E	_			Either cleared to 0		1
CS3E	_					1
ABW	0	1		_		_
PB0DDR	_	0 1		0	1	_
Pin function	D0 I/O pin	PB0 PB0 output pin		PB0 input pin	PB0 output pin	HIRQ3 output pin

8.12.4 **MOS Input Pull-Up Function**

Port B has a built-in MOS input pull-up function that can be controlled by software. This MOS input pull-up function can be used in modes 1, 2 and 3 (EXPE = 1) with the ABW bit in WSCR set to 1, and in modes 2 and 3 (EXPE = 0), and can be specified as on or off on a bit-by-bit basis.

When a PBDDR bit is cleared to 0, setting the corresponding PBODR bit to 1 turns on the MOS input pull-up for that pin. When a pin is designated as an on-chip supporting module output pin, the MOS input pull-up is always off.

The MOS input pull-up function is in the off state after a reset and in hardware standby mode. The prior state is retained in software standby mode.

Table 8.27 summarizes the MOS input pull-up states.

Table 8.27 MOS Input Pull-Up States (Port B)

Mode	Reset	Hardware Standby Mode	Software Standby Mode	In Other Operations
1, 2, 3 (EXPE = 1) with ABW in WSCR = 0	Off	Off	Off	Off
1, 2, 3 (EXPE = 1) with ABW in WSCR = 1, and 2, 3 (EXPE = 0)	_		On/Off	On/Off

Legend:

Off: MOS input pull-up is always off.

On/Off: On when PBDDR = 0 and PBODR = 1; otherwise off.



Section 9 8-Bit PWM Timers

Provided in the H8S/2148 Group; not provided in the H8S/2144 Group and H8S/2147N.

9.1 Overview

The H8/2148 Group has an on-chip pulse width modulation (PWM) timer module with sixteen outputs. Sixteen output waveforms are generated from a common time base, enabling PWM output with a high carrier frequency to be produced using pulse division. The PWM timer module has sixteen 8-bit PWM data registers (PWDRs), and an output pulse with a duty cycle of 0 to 100% can be obtained as specified by PWDR and the port data register (P1DR or P2DR).

9.1.1 Features

The PWM timer module has the following features.

- Operable at a maximum carrier frequency of 1.25 MHz using pulse division (at 20-MHz operation)
- Duty cycles from 0 to 100% with 1/256 resolution (100% duty realized by port output)
- Direct or inverted PWM output, and PWM output enable/disable control

9.1.2 Block Diagram

Figure 9.1 shows a block diagram of the PWM timer module.

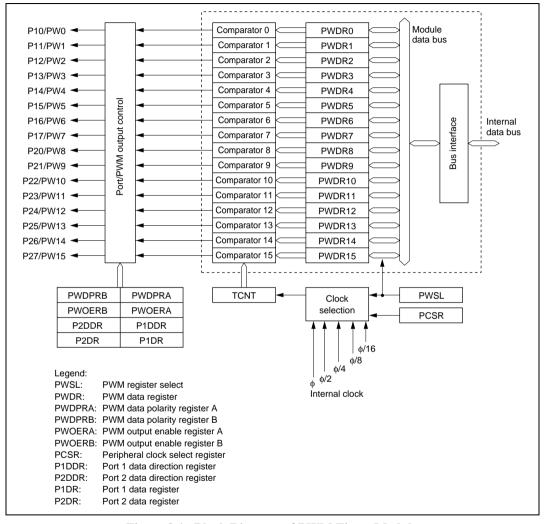


Figure 9.1 Block Diagram of PWM Timer Module

9.1.3 Pin Configuration

Table 9.1 shows the PWM output pin.

Table 9.1 Pin Configuration

Name	Abbreviation	I/O	Function
PWM output pin 0 to 15	PW0 to PW15	Output	PWM timer pulse output 0 to 15

9.1.4 Register Configuration

Table 9.2 lists the registers of the PWM timer module.

Table 9.2 PWM Timer Module Registers

Name	Abbreviation	R/W	Initial Value	Address*1
PWM register select	PWSL	R/W	H'20	H'FFD6
PWM data registers 0 to 15	PWDR0 to PWDR15	R/W	H'00	H'FFD7
PWM data polarity register A	PWDPRA	R/W	H'00	H'FFD5
PWM data polarity register B	PWDPRB	R/W	H'00	H'FFD4
PWM output enable register A	PWOERA	R/W	H'00	H'FFD3
PWM output enable register B	PWOERB	R/W	H'00	H'FFD2
Port 1 data direction register	P1DDR	W	H'00	H'FFB0
Port 2 data direction register	P2DDR	W	H'00	H'FFB1
Port 1 data register	P1DR	R/W	H'00	H'FFB2
Port 2 data register	P2DR	R/W	H'00	H'FFB3
Peripheral clock select register	PCSR	R/W	H'00	H'FF82*2
Module stop control register	MSTPCRH	R/W	H'3F	H'FF86
	MSTPCRL	R/W	H'FF	H'FF87

Note: 1. Lower 16 bits of the address.

2. Some registers in the 8-bit timer are assigned in the addresses as other registers. In this case, register selection is performed by the FLSHE bit in the serial timer control register (STCR).

9.2 Register Descriptions

9.2.1 PWM Register Select (PWSL)

Bit	7	6	5	4	3	2	1	0
	PWCKE	PWCKS	_	_	RS3	RS2	RS1	RS0
Initial value	0	0	1	0	0	0	0	0
Read/Write	R/W	R/W	_	_	R/W	R/W	R/W	R/W

PWSL is an 8-bit readable/writable register used to select the PWM timer input clock and the PWM data register.

PWSL is initialized to H'20 by a reset, and in the standby modes, watch mode, subactive mode, subsleep mode, and module stop mode.

Bits 7 and 6—PWM Clock Enable, PWM Clock Select (PWCKE, PWCKS): These bits, together with bits PWCKA and PWCKB in PCSR, select the internal clock input to TCNT in the PWM timer.

F	PWSL	F	PCSR					
Bit 7	Bit 6	Bit 2	Bit 1	_				
PWCKE PWCKS		PWCKB	PWCKA	 Description				
0	_	_	_	Clock input is disabled	(Initial value)			
1	0	_	_	φ (system clock) is selected				
	1	0	0					
			1	φ/4 is selected				
		1	0	φ/8 is selected				
			1	φ/16 is selected				

The PWM resolution, PWM conversion period, and carrier frequency depend on the selected internal clock, and can be found from the following equations.

Resolution (minimum pulse width) = 1/internal clock frequency PWM conversion period = resolution \times 256 Carrier frequency = 16/PWM conversion period

Thus, with a 20-MHz system clock (ϕ), the resolution, PWM conversion period, and carrier frequency are as shown below.

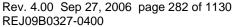




Table 9.3 Resolution, PWM Conversion Period, and Carrier Frequency when $\phi = 20 \text{ MHz}$

Internal Clock Frequency	Resolution	PWM Conversion Period	Carrier Frequency
ф	50 ns	12.8 μs	1250 kHz
φ/2	100 ns	25.6 μs	625 kHz
φ/4	200 ns	51.2 μs	312.5 kHz
ф/8	400 ns	102.4 μs	156.3 kHz
φ/16	800 ns	204.8 μs	78.1 kHz

Bit 5—Reserved: This bit is always read as 1 and cannot be modified.

Bit 4—Reserved: This bit is always read as 0 and cannot be modified.

Bits 3 to 0—Register Select (RS3 to RS0): These bits select the PWM data register.

Bit 3	Bit 2	Bit 1	Bit 0	
RS3	RS2	RS1	RS0	Register Selection
0	0	0	0	PWDR0 selected
			1	PWDR1 selected
		1	0	PWDR2 selected
			1	PWDR3 selected
	1	0	0	PWDR4 selected
			1	PWDR5 selected
		1	0	PWDR6 selected
			1	PWDR7 selected
1	0	0	0	PWDR8 selected
			1	PWDR9 selected
		1	0	PWDR10 selected
			1	PWDR11 selected
	1	0	0	PWDR12 selected
			1	PWDR13 selected
		1	0	PWDR14 selected
			1	PWDR15 selected

9.2.2 PWM Data Registers (PWDR0 to PWDR15)

Bit	7	6	5	4	3	2	1	0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W							

Each PWDR is an 8-bit readable/writable register that specifies the duty cycle of the basic pulse to be output, and the number of additional pulses. The value set in PWDR corresponds to a 0 or 1 ratio in the conversion period. The upper 4 bits specify the duty cycle of the basic pulse as 0/16 to 15/16 with a resolution of 1/16. The lower 4 bits specify how many extra pulses are to be added within the conversion period comprising 16 basic pulses. Thus, a specification of 0/256 to 255/256 is possible for 0/1 ratios within the conversion period. For 256/256 (100%) output, port output should be used.

PWDR is initialized to H'00 by a reset, and in the standby modes, watch mode, subactive mode, subsleep mode, and module stop mode.

9.2.3 PWM Data Polarity Registers A and B (PWDPRA and PWDPRB)

PWDPRA								
Bit	7	6	5	4	3	2	1	0
	OS7	OS6	OS5	OS4	OS3	OS2	OS1	OS0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
PWDPRB								
Bit	7	6	5	4	3	2	1	0
	OS15	OS14	OS13	OS12	OS11	OS10	OS9	OS8
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

Each PWDPR is an 8-bit readable/writable register that controls the polarity of the PWM output. Bits OS0 to OS15 correspond to outputs PW0 to PW15.

PWDPR is initialized to H'00 by a reset and in hardware standby mode.

os	Description	Description	
0	PWM direct output (PWDR value corresponds to high width of output) (Initial value)	PWM direct output (PWDR value corresponds to high width of output)	
1	PWM inverted output (PWDR value corresponds to low width of output)	PWM inverted output (PWDR value corresponds to low width of output)	_

9.2.4 PWM Output Enable Registers A and B (PWOERA and PWOERB)

PWOERA								
Bit	7	6	5	4	3	2	1	0
	OE7	OE6	OE5	OE4	OE3	OE2	OE1	OE0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
PWOERB								
Bit	7	6	5	4	3	2	1	0
	OE15	OE14	OE13	OE12	OE11	OE10	OE9	OE8
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

Each PWOER is an 8-bit readable/writable register that switches between PWM output and port output. Bits OE15 to OE0 correspond to outputs PW15 to PW0. To set a pin in the output state, a setting in the port direction register is also necessary. Bits P17DDR to P10DDR correspond to outputs PW7 to PW0, and bits P27DDR to P20DDR correspond to outputs PW15 to PW8.

PWOER is initialized to H'00 by a reset and in hardware standby mode.

DDR	OE	Description	
0	0	Port input	(Initial value)
	1	Port input	
1	0	Port output or PWM 256/256 output	
	1	PWM output (0 to 255/256 output)	

9.2.5 Peripheral Clock Select Register (PCSR)

Bit	7	6	5	4	3	2	1	0
	_	_	_	_	_	PWCKB	PWCKA	_
Initial value	0	0	0	0	0	0	0	0
Read/Write	_	_	_	_	_	R/W	R/W	_

PCSR is an 8-bit readable/writable register that selects the PWM timer input clock.

PCSR is initialized to H'00 by a reset, and in hardware standby mode.

Bits 7 to 3—Reserved: These bits cannot be modified and are always read as 0.

Bits 2 and 1—PWM Clock Select (PWCKB, PWCKA): Together with bits PWCKE and PWCKS in PWSL, these bits select the internal clock input to TCNT in the PWM timer. For details, see section 9.2.1, PWM Register Select (PWSL).

Bit 0—Reserved: Do not set this bit to 1.

Port 1 Data Direction Register (P1DDR) 9.2.6

Bit	7	6	5	4	3	2	1	0
	P17DDR	P16DDR	P15DDR	P14DDR	P13DDR	P12DDR	P11DDR	P10DDR
Initial value	0	0	0	0	0	0	0	0
Read/Write	W	W	W	W	W	W	W	W

P1DDR is an 8-bit write-only register that specifies the input/output direction and PWM output for each pin of port 1 on a bit-by-bit basis.

Port 1 pins are multiplexed with pins PW0 to PW7. The bit corresponding to a pin to be used for PWM output should be set to 1.

For details on P1DDR, see section 8.2, Port 1.



9.2.7 Port 2 Data Direction Register (P2DDR)

Bit	7	6	5	4	3	2	1	0
	P27DDR	P26DDR	P25DDR	P24DDR	P23DDR	P22DDR	P21DDR	P20DDR
Initial value	0	0	0	0	0	0	0	0
Read/Write	W	W	W	W	W	W	W	W

P2DDR is an 8-bit write-only register that specifies the input/output direction and PWM output for each pin of port J on a bit-by-bit basis.

Port 2 pins are multiplexed with pins PW8 to PW15. The bit corresponding to a pin to be used for PWM output should be set to 1.

For details on P2DDR, see section 8.3, Port 2.

9.2.8 Port 1 Data Register (P1DR)

Bit	7	6	5	4	3	2	1	0
	P17DR	P16DR	P15DR	15DR P14DR		P12DR	P11DR	P10DR
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

P1DR is an 8-bit readable/writable register used to fix PWM output at 1 (when OS = 0) or 0 (when OS = 1).

For details on P1DR, see section 8.2, Port 1.

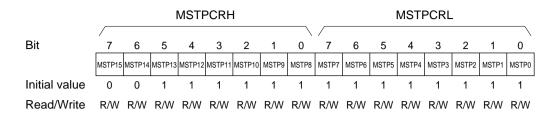
9.2.9 Port 2 Data Register (P2DR)

Bit	7	6	5	4	3	2	1	0	
	P27DR	P26DR	P25DR	P24DR	P23DR	P22DR	P21DR	P20DR	
Initial value	0	0	0	0	0	0	0	0	
Read/Write	R/W								

P2DR is an 8-bit readable/writable register used to fix PWM output at 1 (when OS = 0) or 0 (when OS = 1).

For details on P2DR, see section 8.3, Port 2.

9.2.10 Module Stop Control Register (MSTPCR)



MSTPCR comprises two 8-bit readable/writable registers, and is used to perform module stop mode control.

When the MSTP11 bit is set to 1, 8-bit PWM timer operation is halted and a transition is made to module stop mode. For details, see section 25.5, Module Stop Mode.

MSTPCR is initialized to H'3FFF by a reset and in hardware standby mode. It is not initialized in software standby mode.

MSTPCRH Bit 3—Module Stop (MSTP11): Specifies PWM module stop mode.

MSTPCRH Bit 3

MSTP11	Description	
0	PWM module stop mode is cleared	
1	PWM module stop mode is set	(Initial value)

9.3 Operation

9.3.1 Correspondence between PWM Data Register Contents and Output Waveform

The upper 4 bits of PWDR specify the duty cycle of the basic pulse as 0/16 to 15/16 with a resolution of 1/16, as shown in table 9.4.

Table 9.4 Duty Cycle of Basic Pulse

Upper 4 Bits	Basic Pulse Waveform (Internal)
0000	0 1 2 3 4 5 6 7 8 9 A B C D E F 0
0001	
0010	
0011	
0100	
0101	
0110	
0111	
:	
1000	
1001	
1010	
1011	
1100	
1101	
1110	
1111	

The lower 4 bits of PWDR specify the position of pulses added to the 16 basic pulses, as shown in table 9.5. An additional pulse consists of a high period (when OS = 0) with a width equal to the resolution, added before the rising edge of a basic pulse. When the upper 4 bits of PWDR are 0000, there is no rising edge of the basic pulse, but the timing for adding pulses is the same.

Table 9.5	Position	of Pulses	Added to	Basic Pulses

		Basic Pulse No.														
Lower 4 Bits	0	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15
0000																
0001																Yes
0010								Yes								Yes
0011								Yes				Yes				Yes
0100				Yes				Yes				Yes				Yes
0101				Yes				Yes				Yes		Yes		Yes
0110				Yes		Yes		Yes				Yes		Yes		Yes
0111				Yes												
1000		Yes		Yes		Yes		Yes		Yes		Yes		Yes		Yes
1001		Yes		Yes		Yes		Yes		Yes		Yes		Yes	Yes	Yes
1010		Yes		Yes		Yes	Yes	Yes		Yes		Yes		Yes	Yes	Yes
1011		Yes		Yes		Yes	Yes	Yes		Yes	Yes	Yes		Yes	Yes	Yes
1100		Yes	Yes	Yes		Yes	Yes	Yes		Yes	Yes	Yes		Yes	Yes	Yes
1101		Yes	Yes	Yes		Yes	Yes	Yes		Yes						
1110		Yes	Yes	Yes	Yes	Yes	Yes	Yes		Yes						
1111		Yes	Yes	Yes	Yes	Yes	Yes	Yes	Yes	Yes	Yes	Yes	Yes	Yes	Yes	Yes

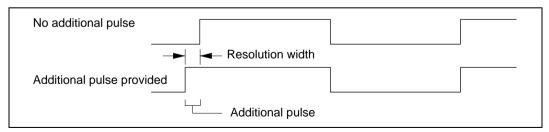


Figure 9.2 Example of Additional Pulse Timing (when Upper 4 Bits of PWDR = 1000)

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Section 10 14-Bit PWM Timer (PWMX)

10.1 Overview

This LSI have an on-chip 14-bit pulse-width modulator (PWM) with two output channels.

Each channel can be connected to an external low-pass filter to operate as a 14-bit D/A converter.

Both channels share the same counter (DACNT) and control register (DACR).

10.1.1 Features

The features of the 14-bit PWM (D/A) are listed below.

- The pulse is subdivided into multiple base cycles to reduce ripple.
- Two resolution settings and two base cycle settings are available
 The resolution can be set equal to one or two system clock cycles. The base cycle can be set equal to T × 64 or T × 256, where T is the resolution.
- Four operating rates
 The two resolution settings and two base cycle settings combine to give a selection of four operating rates.

10.1.2 Block Diagram

Figure 10.1 shows a block diagram of the PWM (D/A) module.

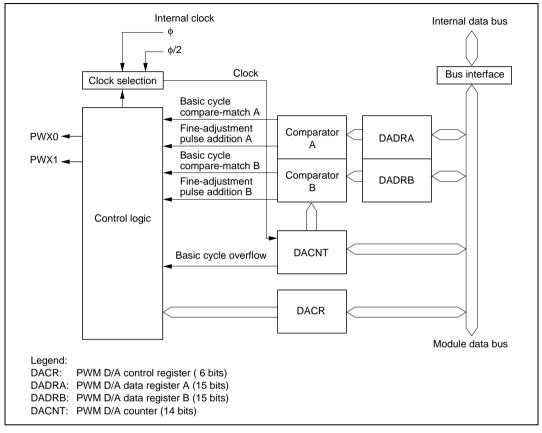


Figure 10.1 PWM D/A Block Diagram

10.1.3 Pin Configuration

Table 10.1 lists the pins used by the PWM (D/A) module.

Table 10.1 PWM Timer Input and Output Pins

Name	Abbr.	I/O	Function
PWM output pin 0	PWX0	Output	PWM output, channel A
PWM output pin 1	PWX1	Output	PWM output, channel B

10.1.4 Register Configuration

Table 10.2 lists the registers of the PWM (D/A) module.

Table 10.2 Register Configuration

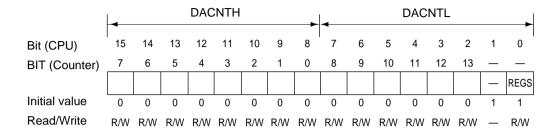
Name	Abbreviation	R/W	Initial value	Address*1
PWM D/A control register	DACR	R/W	H'30	H'FFA0*2
PWM D/A data register A high	DADRAH	R/W	H'FF	H'FFA0*2
PWM D/A data register A low	DADRAL	R/W	H'FF	H'FFA1*2
PWM D/A data register B high	DADRBH	R/W	H'FF	H'FFA6*2
PWM D/A data register B low	DADRBL	R/W	H'FF	H'FFA7*2
PWM D/A counter high	DACNTH	R/W	H'00	H'FFA6*2
PWM D/A counter low	DACNTL	R/W	H'03	H'FFA7*2
Module stop control register	MSTPCRH	R/W	H'3F	H'FF86
	MSTPCRL	R/W	H'FF	H'FF87

Notes: 1. Lower 16 bits of the address.

Registers in the 14-bit PWM timer are assigned to the same addresses as the other
registers. In this case, register selection is performed by the IICE bit in the serial timer
control register (STCR), and also the same addresses are shared by DADRAH and
DACR, and by DADRB and DACNT. Switching is performed by the REGS bit in DACNT
or DADRB.

10.2 Register Descriptions

10.2.1 PWM (D/A) Counter (DACNT)



DACNT is a 14-bit readable/writable up-counter that increments on an input clock pulse. The input clock is selected by the clock select bit (CKS) in DACR. The CPU can read and write the DACNT value, but since DACNT is a 16-bit register, data transfers between it and the CPU are performed using a temporary register (TEMP). See section 10.3, Bus Master Interface, for details.

DACNT functions as the time base for both PWM (D/A) channels. When a channel operates with 14-bit precision, it uses all DACNT bits. When a channel operates with 12-bit precision, it uses the lower 12 (counter) bits and ignores the upper two (counter) bits.

DACNT is initialized to H'0003 by a reset, in the standby modes, watch mode, subactive mode, subsleep mode, and module stop mode, and by the PWME bit.

Bit 1 of DACNTL (CPU) is not used, and is always read as 1.

DACNTL Bit 0—Register Select (REGS): DADRA and DACR, and DADRB and DACNT, are located at the same addresses. The REGS bit specifies which registers can be accessed. The REGS bit can be accessed regardless of whether DADRB or DACNT is selected.

Bit 0

REGS	Description	
0	DADRA and DADRB can be accessed	
1	DACR and DACNT can be accessed	(Initial value)

10.2.2 D/A Data Registers A and B (DADRA and DADRB)

	DADRH								DADRL							
Bit (CPU)	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Bit (Data)	13	12	11	10	9	8	7	6	5	4	3	2	1	0	_	
DADRA	DA13	DA12	DA11	DA10	DA9	DA8	DA7	DA6	DA5	DA4	DA3	DA2	DA1	DA0	CFS	
Initial value	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	_
DADRB	DΔ13	DA12	DA11	DA10	DAG	DAS	DA7	DAG	DAS	DA4	DA3	DA2	DA1	DAO	CES	REGS
Initial value		DAIZ	DATI	DATO	DAS							DAZ	DAI	DAU	4	KLG3
miliai value	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

There are two 16-bit readable/writable D/A data registers: DADRA and DADRB. DADRA corresponds to PWM (D/A) channel A, and DADRB to PWM (D/A) channel B. The CPU can read and write the PWM (D/A) data register values, but since DADRA and DADRB are 16-bit registers, data transfers between them and the CPU are performed using a temporary register (TEMP). See section 10.3, Bus Master Interface, for details.

The least significant (CPU) bit of DADRA is not used and is always read as 1.

DADR is initialized to H'FFFF by a reset, and in the standby modes, watch mode, subactive mode, subsleep mode, and module stop mode.

Bits 15 to 3—PWM D/A Data 13 to 0 (DA13 to DA0): The digital value to be converted to an analog value is set in the upper 14 bits of the PWM (D/A) data register.

In each base cycle, the DACNT value is continually compared with these upper 14 bits to determine the duty cycle of the output waveform, and to decide whether to output a fine-adjustment pulse equal in width to the resolution. To enable this operation, the data register must be set within a range that depends on the carrier frequency select bit (CFS). If the DADR value is outside this range, the PWM output is held constant.

A channel can be operated with 12-bit precision by keeping the two lowest data bits (DA0 and DA1) cleared to 0 and writing the data to be converted in the upper 12 bits. The two lowest data bits correspond to the two highest counter (DACNT) bits.

Bit 1—Carrier Frequency Select (CFS)

Bit 1

CFS	Description	
0	Base cycle = resolution (T) × 64 DADR range = H'0401 to H'FFFD	
1	Base cycle = resolution (T) × 256 DADR range = H'0103 to H'FFFF	(Initial value)

DADRA Bit 0—Reserved: This bit cannot be modified and is always read as 1.

DADRB Bit 0—Register Select (REGS): DADRA and DACR, and DADRB and DACNT, are located at the same addresses. The REGS bit specifies which registers can be accessed. The REGS bit can be accessed regardless of whether DADRB or DACNT is selected.

Bit 0

REGS	Description	
0	DADRA and DADRB can be accessed	_
1	DACR and DACNT can be accessed	(Initial value)

10.2.3 PWM D/A Control Register (DACR)

Bit	7	6	5	4	3	2	1	0
	TEST	PWME	_	_	OEB	OEA	os	CKS
Initial value	0	0	1	1	0	0	0	0
Read/Write	R/W	R/W	_	_	R/W	R/W	R/W	R/W

DACR is an 8-bit readable/writable register that selects test mode, enables the PWM outputs, and selects the output phase and operating speed.

DACR is initialized to H'30 by a reset, and in the standby modes, watch mode, subactive mode, subsleep mode, and module stop mode.

Bit 7—Test Mode (TEST): Selects test mode, which is used in testing the chip. Normally this bit should be cleared to 0



Bit 7

TEST	Description	
0	PWM (D/A) in user state: normal operation	(Initial value)
1	PWM (D/A) in test state: correct conversion results unobtainable	

Bit 6—PWM Enable (PWME): Starts or stops the PWM (D/A) counter (DACNT).

Bit 6

PWME	Description	
0	DACNT operates as a 14-bit up-counter	(Initial value)
1	DACNT halts at H'0003	

Bits 5 and 4—Reserved: These bits cannot be modified and are always read as 1.

Bit 3—Output Enable B (OEB): Enables or disables output on PWM (D/A) channel B.

Bit 3

OEB	Description	
0	PWM (D/A) channel B output (at the PWX1 pin) is disabled	(Initial value)
1	PWM (D/A) channel B output (at the PWX1 pin) is enabled	

Bit 2—Output Enable A (OEA): Enables or disables output on PWM (D/A) channel A.

Bit 2

OEA	Description	
0	PWM (D/A) channel A output (at the PWX0 pin) is disabled	(Initial value)
1	PWM (D/A) channel A output (at the PWX0 pin) is enabled	

Bit 1—Output Select (OS): Selects the phase of the PWM (D/A) output.

Bit 1

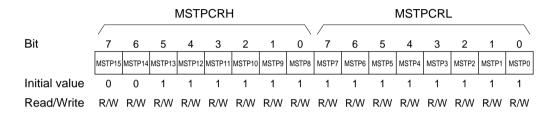
os	Description	
0	Direct PWM output	(Initial value)
1	Inverted PWM output	

Bit 0—Clock Select (CKS): Selects the PWM (D/A) resolution. If the system clock (φ) frequency is 10 MHz, resolutions of 100 ns and 200 ns can be selected.

D:4	\mathbf{a}
KIT	"

CKS	Description	
0	Operates at resolution (T) = system clock cycle time (t_{cyc})	(Initial value)
1	Operates at resolution (T) = system clock cycle time $(t_{cyc}) \times 2$	

10.2.4 Module Stop Control Register (MSTPCR)



MSTPCR comprises two 8-bit readable/writable registers, and is used to perform module stop mode control.

When the MSTP11 bit is set to 1, 14-bit PWM timer operation is halted and a transition is made to module stop mode. For details, see section 25.5, Module Stop Mode.

MSTPCR is initialized to H'3FFF by a reset and in hardware standby mode. It is not initialized in software standby mode.

MSTPCRH Bit 3—Module Stop (MSTP11): Specifies PWMX module stop mode.

MSTPCRH Bit 3

MSTP11	Description	
0	PWMX module stop mode is cleared	
1	PWMX module stop mode is set	(Initial value)

10.3 Bus Master Interface

DACNT, DADRA, and DADRB are 16-bit registers. The data bus linking the bus master and the on-chip supporting modules, however, is only 8 bits wide. When the bus master accesses these registers, it therefore uses an 8-bit temporary register (TEMP).

These registers are written and read as follows (taking the example of the CPU interface).

Write

When the upper byte is written, the upper-byte write data is stored in TEMP. Next, when the lower byte is written, the lower-byte write data and TEMP value are combined, and the combined 16-bit value is written in the register.

Read

When the upper byte is read, the upper-byte value is transferred to the CPU and the lower-byte value is transferred to TEMP. Next, when the lower byte is read, the lower-byte value in TEMP is transferred to the CPU.

These registers should always be accessed 16 bits at a time using an MOV instruction (by word access or two consecutive byte accesses), and the upper byte should always be accessed before the lower byte. Correct data will not be transferred if only the upper byte or only the lower byte is accessed. Also note that a bit-manipulation instruction cannot be used to access these registers.

Figure 10.2 shows the data flow for access to DACNT. The other registers are accessed similarly.

Example 1: Write to DACNT

MOV.W RO, @DACNT; Write RO contents to DACNT

Example 2: Read DADRA

MOV.W @DADRA, RO ; Transfer contents of DADRA to RO

Table 10.3 Read and Write Access Methods for 16-Bit Registers

	Read		Write		
Register Name	Word	Byte	Word	Byte	
DADRA and DADRB	Yes	Yes	Yes	×	
DACNT	Yes	×	Yes	×	

Legend:

Yes: Permitted type of access. Word access includes successive byte accesses to the upper byte (first) and lower byte (second).

x: This type of access may give incorrect results.

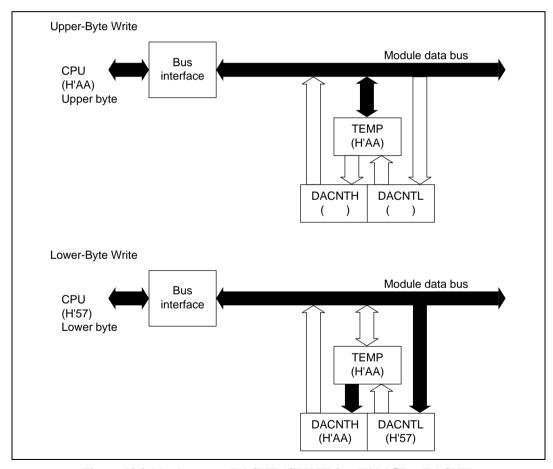


Figure 10.2 (a) Access to DACNT (CPU Writes H'AA57 to DACNT)

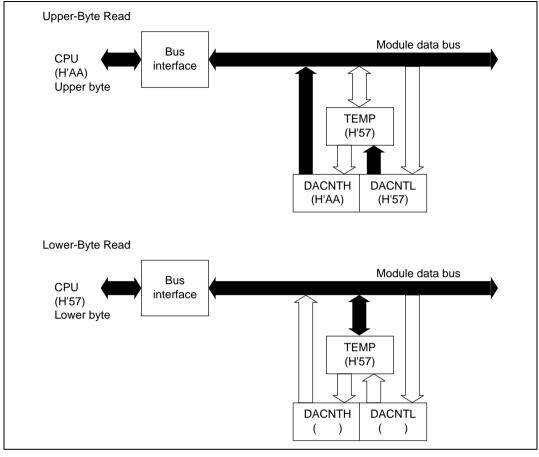


Figure 10.2 (b) Access to DACNT (CPU Reads H'AA57 from DACNT)

10.4 Operation

A PWM waveform like the one shown in figure 10.3 is output from the PWMX pin. When OS = 0, the value in DADR corresponds to the total width (T_L) of the low (0) pulses output in one conversion cycle (256 pulses when CFS = 0, 64 pulses when CFS = 1). When OS = 1, the output waveform is inverted and the DADR value corresponds to the total width (T_H) of the high (1) output pulses. Figure 10.4 shows the types of waveform output available.

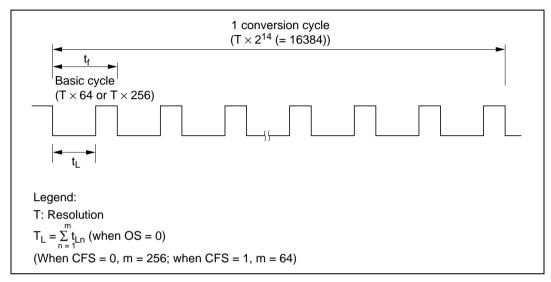


Figure 10.3 PWM D/A Operation

Table 10.4 summarizes the relationships of the CKS, CFS, and OS bit settings to the resolution, base cycle, and conversion cycle. The PWM output remains flat unless DADR contains at least a certain minimum value. Table 10.4 indicates the range of DADR settings that give an output waveform like the one in figure 10.3, and lists the conversion cycle length when low-order DADR bits are kept cleared to 0, reducing the conversion precision to 12 bits or 10 bits.

Table 10.4 Settings and Operation (Examples when $\phi = 10 \text{ MHz}$)

Resolution			Base	Conversion	sion T ((00 0)		Fixed DADR Bits						
скѕ	Т	CFS	Cycle	Cycle	T _L (if OS = 0) T _H (if OS = 1)	Precision	Е	3it I	Dat	а	Cycle*		
	(µs)		(µs)	(µs)	- H (4- 2-2 - 3)	(Bits)	3	2	1	0	(µs)		
0	0.1	0	6.4	1638.4	1. Always low (or high) level output (DADR = H'0001 to H'03FD)						1638.4		
					2. (Data value) × T (DADR = H'0401 to	12			0	0	409.6		
					H'FFFD)	10	0	0	0	0	102.4		
		1	25.6	1638.4	1. Always low (or high) level output (DADR = H'0003 to H'00FF)	14					1638.4		
					2. (Data value) × T	12			0	0	409.6		
					(DADR = H'0103 to H'FFFF)	10	0	0	0	0	102.4		
1	0.2	0	12.8	3276.8	1. Always low (or high) level output (DADR = H'0001 to H'03FD)	14					3276.8		
					2. (Data value) × T	12			0	0	819.2		
					(DADR = H'0401 to H'FFFD)	10	0	0	0	0	204.8		
		1	51.2	3276.8	1. Always low (or high) level output (DADR = H'0003 to H'00FF)	14					3276.8		
					2. (Data value) × T	12			0	0	819.2		
					(DADR = H'0103 to H'FFFF)	10	0	0	0	0	204.8		

Note: * This column indicates the conversion cycle when specific DADR bits are fixed.

1. OS = 0 (DADR corresponds to T_r)

a. CFS = 0 [base cycle = resolution $(T) \times 64$]

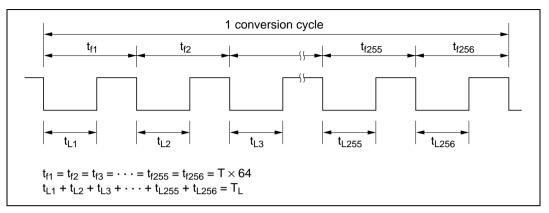


Figure 10.4 (1) Output Waveform

b. CFS = 1 [base cycle = resolution (T) \times 256]

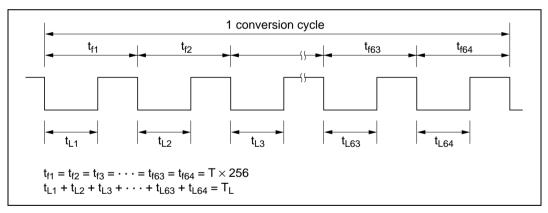


Figure 10.4 (2) Output Waveform

2. OS = 1 (DADR corresponds to T_{H})

a. CFS = 0 [base cycle = resolution (T) \times 64]

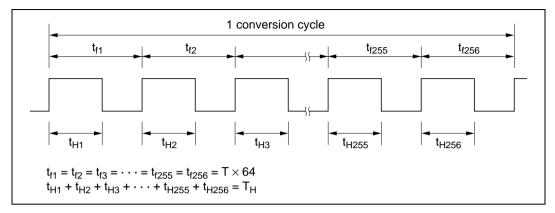


Figure 10.4 (3) Output Waveform

b. CFS = 1 [base cycle = resolution (T) \times 256]

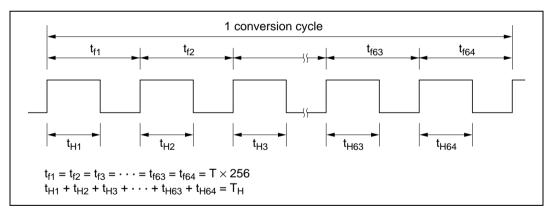


Figure 10.4 (4) Output Waveform

Section 11 16-Bit Free-Running Timer

11.1 Overview

This LSI have a single-channel on-chip 16-bit free-running timer (FRT) module that uses a 16-bit free-running counter as a time base. Applications of the FRT module include rectangular-wave output (up to two independent waveforms), input pulse width measurement, and measurement of external clock periods.

11.1.1 Features

The features of the free-running timer module are listed below.

- Selection of four clock sources
 - The free-running counter can be driven by an internal clock source ($\phi/2$, $\phi/8$, or $\phi/32$), or an external clock input (enabling use as an external event counter).
- Two independent comparators
 - Each comparator can generate an independent waveform.
- Four input capture channels
 - The current count can be captured on the rising or falling edge (selectable) of an input signal.
 - The four input capture registers can be used separately, or in a buffer mode.
- Counter can be cleared under program control
 - The free-running counters can be cleared on compare-match A.
- Seven independent interrupts
 - Two compare-match interrupts, four input capture interrupts, and one overflow interrupt can be requested independently.
- Special functions provided by automatic addition function
 - The contents of OCRAR and OCRAF can be added to the contents of OCRA automatically, enabling a periodic waveform to be generated without software intervention.
 - The contents of ICRD can be added automatically to the contents of OCRDM × 2, enabling input capture operations in this interval to be restricted.

11.1.2 Block Diagram

Figure 11.1 shows a block diagram of the free-running timer.

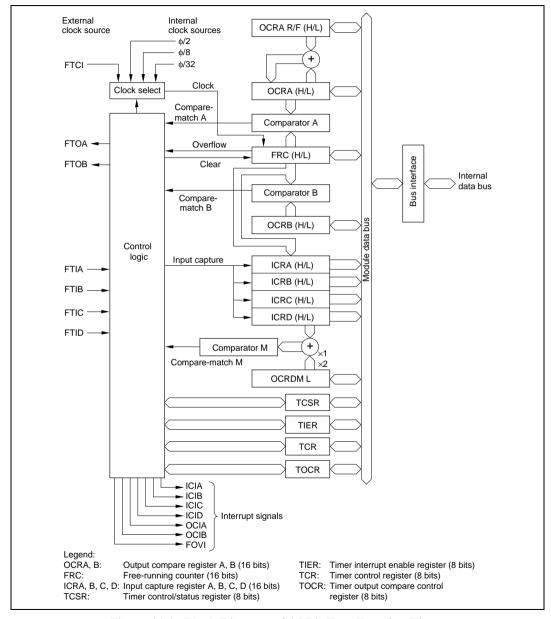


Figure 11.1 Block Diagram of 16-Bit Free-Running Timer

11.1.3 Input and Output Pins

Table 11.1 lists the input and output pins of the free-running timer module.

Table 11.1 Free-Running Timer Input and Output Pins

Name	Abbreviation	I/O	Function
Counter clock input	FTCI	Input	FRC counter clock input
Output compare A	FTOA	Output	Output compare A output
Output compare B	FTOB	Output	Output compare B output
Input capture A	FTIA	Input	Input capture A input
Input capture B	FTIB	Input	Input capture B input
Input capture C	FTIC	Input	Input capture C input
Input capture D	FTID	Input	Input capture D input

11.1.4 Register Configuration

Table 11.2 lists the registers of the free-running timer module.

Table 11.2 Register Configuration

Name	Abbreviation	R/W	Initial Value	Address*1
Timer interrupt enable register	TIER	R/W	H'01	H'FF90
Timer control/status register	TCSR	R/(W)*2	H'00	H'FF91
Free-running counter	FRC	R/W	H'0000	H'FF92
Output compare register A	OCRA	R/W	H'FFFF	H'FF94*3
Output compare register B	OCRB	R/W	H'FFFF	H'FF94*3
Timer control register	TCR	R/W	H'00	H'FF96
Timer output compare control register	TOCR	R/W	H'00	H'FF97
Input capture register A	ICRA	R	H'0000	H'FF98*4
Input capture register B	ICRB	R	H'0000	H'FF9A*4
Input capture register C	ICRC	R	H'0000	H'FF9C*4
Input capture register D	ICRD	R	H'0000	H'FF9E
Output compare register AR	OCRAR	R/W	H'FFFF	H'FF98*4
Output compare register AF	OCRAF	R/W	H'FFFF	H'FF9A*4
Output compare register DM	OCRDM	R/W	H'0000	H'FF9C*4
Module stop control register	MSTPCRH	R/W	H'3F	H'FF86
	MSTPCRL	R/W	H'FF	H'FF87

Notes: 1. Lower 16 bits of the address.

- 2. Bits 7 to 1 are read-only; only 0 can be written to clear the flags. Bit 0 is readable/writable.
- OCRA and OCRB share the same address. Access is controlled by the OCRS bit in TOCR.
- 4. ICRA, ICRB, and ICRC share the same addresses with OCRAR, OCRAF, and OCRDM. Access is controlled by the ICRS bit in TOCR.

11.2 Register Descriptions

11.2.1 Free-Running Counter (FRC)

Bit	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Initial value	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W					R/W								

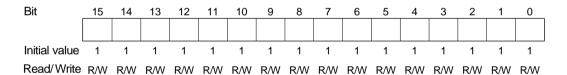
FRC is a 16-bit readable/writable up-counter that increments on an internal pulse generated from a clock source. The clock source is selected by bits CKS1 and CKS0 in TCR.

FRC can also be cleared by compare-match A.

When FRC overflows from H'FFFF to H'0000, the overflow flag (OVF) in TCSR is set to 1.

FRC is initialized to H'0000 by a reset and in hardware standby mode.

11.2.2 Output Compare Registers A and B (OCRA, OCRB)



OCRA and OCRB are 16-bit readable/writable registers, the contents of which are continually compared with the value in the FRC. When a match is detected, the corresponding output compare flags (OCFA or OCFB) is set in TCSR.

In addition, if the output enable bit (OEA or OEB) in TOCR is set to 1, when OCR and FRC values match, the logic level selected by the output level bit (OLVLA or OLVLB) in TOCR is output at the output compare pin (FTOA or FTOB). Following a reset, the FTOA and FTOB output levels are 0 until the first compare-match.

OCR is initialized to H'FFFF by a reset and in hardware standby mode.

11.2.3 Input Capture Registers A to D (ICRA to ICRD)

Bit	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Initial value	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
Read/Write	R	R	R	R	R	R	R	R	R	R	R	R	R	R	R	R

There are four input capture registers, A to D, each of which is a 16-bit read-only register.

When the rising or falling edge of the signal at an input capture input pin (FTIA to FTID) is detected, the current FRC value is copied to the corresponding input capture register (ICRA to ICRD). At the same time, the corresponding input capture flag (ICFA to ICFD) in TCSR is set to 1. The input capture edge is selected by the input edge select bits (IEDGA to IEDGD) in TCR.

ICRC and ICRD can be used as ICRA and ICRB buffer registers, respectively, and made to perform buffer operations, by means of buffer enable bits A and B (BUFEA, BUFEB) in TCR.

Figure 11.2 shows the connections when ICRC is specified as the ICRA buffer register (BUFEA = 1). When ICRC is used as the ICRA buffer, both rising and falling edges can be specified as transitions of the external input signal by setting IEDGA \neq IEDGC. When IEDGA = IEDGC, either the rising or falling edge is designated. See table 11.3.

Note: The FRC contents are transferred to the input capture register regardless of the value of the input capture flag (ICF).

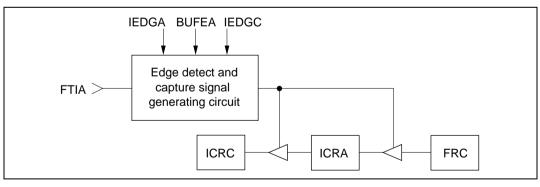


Figure 11.2 Input Capture Buffering (Example)

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Table 11.3 Buffered Input Capture Edge Selection (Example)

IEDGA	IEDGC	Description
0	0	Captured on falling edge of input capture A (FTIA) (Initial value)
	1	Captured on both rising and falling edges of input capture A (FTIA)
1	0	
	1	Captured on rising edge of input capture A (FTIA)

To ensure input capture, the width of the input capture pulse should be at least 1.5 system clock periods (1.5ϕ) . When triggering is enabled on both edges, the input capture pulse width should be at least 2.5 system clock periods (2.5ϕ) .

ICR is initialized to H'0000 by a reset and in hardware standby mode.

11.2.4 Output Compare Registers AR and AF (OCRAR, OCRAF)

Bit	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Initial value	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1
Read/Write	RΛΛ	RΛΛ	DΛΛ	DΛΛ	DΛΛ	DΛΛ	RΛΛ	DΛΛ								

OCRAR and OCRAF are 16-bit readable/writable registers.

When the OCRAMS bit in TOCR is set to 1, the operation of OCRA is changed to include the use of OCRAR and OCRAF. The contents of OCRAR and OCRAF are automatically added alternately to OCRA, and the result is written to OCRA. The write operation is performed on the occurrence of compare-match A. In the first compare-match A after the OCRAMS bit is set to 1, OCRAF is added.

The operation due to compare-match A varies according to whether the compare-match follows addition of OCRAR or OCRAF. The value of the OLVLA bit in TOCR is ignored, and 1 is output on a compare-match A following addition of OCRAF, while 0 is output on a compare-match A following addition of OCRAR.

When the OCRA automatically addition function is used, do not set internal clock $\phi/2$ as the FRC counter input clock together with an OCRAR (or OCRAF) value of H'0001 or less.

OCRAR and OCRAF are initialized to H'FFFF by a reset and in hardware standby mode.

11.2.5 Output Compare Register DM (OCRDM)

Bit _	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0	
Initial value	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	
Read/Write	R	R	R	R	R	R	R	R	R/W								

OCRDM is a 16-bit readable/writable register in which the upper 8 bits are fixed at H'00.

When the ICRDMS bit in TOCR is set to 1 and the contents of OCRDM are other than H'0000, the operation of ICRD is changed to include the use of OCRDM. The point at which input capture D occurs is taken as the start of a mask interval. Next, twice the contents of OCRDM is added to the contents of ICRD, and the result is compared with the FRC value. The point at which the values match is taken as the end of the mask interval. New input capture D events are disabled during the mask interval.

A mask interval is not generated when the ICRDMS bit is set to 1 and the contents of OCRDM are H'0000.

OCRDM is initialized to H'0000 by a reset and in hardware standby mode.

11.2.6 Timer Interrupt Enable Register (TIER)

Bit	7	6	5	4	3	2	1	0
	ICIAE	ICIBE	ICICE	ICIDE	OCIAE	OCIBE	OVIE	_
Initial value	0	0	0	0	0	0	0	1
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	

TIER is an 8-bit readable/writable register that enables and disables interrupts.

TIER is initialized to H'01 by a reset and in hardware standby mode.

Bit 7—Input Capture Interrupt A Enable (ICIAE): Selects whether to request input capture interrupt A (ICIA) when input capture flag A (ICFA) in TCSR is set to 1.



Bit 7

ICIAE	Description	
0	Input capture interrupt request A (ICIA) is disabled	(Initial value)
1	Input capture interrupt request A (ICIA) is enabled	

Bit 6—Input Capture Interrupt B Enable (ICIBE): Selects whether to request input capture interrupt B (ICIB) when input capture flag B (ICFB) in TCSR is set to 1.

Bit 6

ICIBE	Description	
0	Input capture interrupt request B (ICIB) is disabled	(Initial value)
1	Input capture interrupt request B (ICIB) is enabled	

Bit 5—Input Capture Interrupt C Enable (ICICE): Selects whether to request input capture interrupt C (ICIC) when input capture flag C (ICFC) in TCSR is set to 1.

Bit 5

ICICE	Description	
0	Input capture interrupt request C (ICIC) is disabled	(Initial value)
1	Input capture interrupt request C (ICIC) is enabled	

Bit 4—Input Capture Interrupt D Enable (ICIDE): Selects whether to request input capture interrupt D (ICID) when input capture flag D (ICFD) in TCSR is set to 1.

Bit 4

ICIDE	Description	
0	Input capture interrupt request D (ICID) is disabled	(Initial value)
1	Input capture interrupt request D (ICID) is enabled	

Bit 3—Output Compare Interrupt A Enable (OCIAE): Selects whether to request output compare interrupt A (OCIA) when output compare flag A (OCFA) in TCSR is set to 1.

Bit 3

OCIAE	Description	
0	Output compare interrupt request A (OCIA) is disabled	(Initial value)
1	Output compare interrupt request A (OCIA) is enabled	

Bit 2—Output Compare Interrupt B Enable (OCIBE): Selects whether to request output compare interrupt B (OCIB) when output compare flag B (OCFB) in TCSR is set to 1.

Bit 2

OCIBE	Description	
0	Output compare interrupt request B (OCIB) is disabled	(Initial value)
1	Output compare interrupt request B (OCIB) is enabled	

Bit 1—Timer Overflow Interrupt Enable (OVIE): Selects whether to request a free-running timer overflow interrupt (FOVI) when the timer overflow flag (OVF) in TCSR is set to 1.

Bit 1

OVIE	Description	
0	Timer overflow interrupt request (FOVI) is disabled	(Initial value)
1	Timer overflow interrupt request (FOVI) is enabled	

Bit 0—Reserved: This bit cannot be modified and is always read as 1.

11.2.7 Timer Control/Status Register (TCSR)

Bit	7	6	5	4	3	2	1	0
	ICFA	ICFB	ICFC	ICFD	OCFA	OCFB	OVF	CCLRA
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/(W)*	R/W						

Note: * Only 0 can be written in bits 7 to 1 to clear these flags.

TCSR is an 8-bit register used for counter clear selection and control of interrupt request signals.

TCSR is initialized to H'00 by a reset and in hardware standby mode.

Timing is described in section 11.3, Operation.

Bit 7—Input Capture Flag A (ICFA): This status flag indicates that the FRC value has been transferred to ICRA by means of an input capture signal. When BUFEA = 1, ICFA indicates that the old ICRA value has been moved into ICRC and the new FRC value has been transferred to ICRA.

ICFA must be cleared by software. It is set by hardware, however, and cannot be set by software.

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ICFA	Description	
0	[Clearing condition]	(Initial value)
	Read ICFA when ICFA = 1, then write 0 in ICFA	
1	[Setting condition]	
	When an input capture signal causes the FRC value to be transfelICRA	erred to

Bit 6—Input Capture Flag B (ICFB): This status flag indicates that the FRC value has been transferred to ICRB by means of an input capture signal. When BUFEB = 1, ICFB indicates that the old ICRB value has been moved into ICRD and the new FRC value has been transferred to ICRB.

ICFB must be cleared by software. It is set by hardware, however, and cannot be set by software.

Bit 6

ICFB	Description	
0	[Clearing condition]	(Initial value)
	Read ICFB when ICFB = 1, then write 0 in ICFB	
1	[Setting condition]	
	When an input capture signal causes the FRC value to be transferred to ICRB	

Bit 5—Input Capture Flag C (ICFC): This status flag indicates that the FRC value has been transferred to ICRC by means of an input capture signal. When BUFEA = 1, on occurrence of the signal transition in FTIC (input capture signal) specified by the IEDGC bit, ICFC is set but data is not transferred to ICRC. Therefore, in buffer operation, ICFC can be used as an external interrupt signal (by setting the ICICE bit to 1).

ICFC must be cleared by software. It is set by hardware, however, and cannot be set by software.

Bit 5

ICFC	Description	
0	[Clearing condition]	(Initial value)
	Read ICFC when ICFC = 1, then write 0 in ICFC	
1	[Setting condition]	
	When an input capture signal is received	

Bit 4—Input Capture Flag D (ICFD): This status flag indicates that the FRC value has been transferred to ICRD by means of an input capture signal. When BUFEB = 1, on occurrence of the signal transition in FTID (input capture signal) specified by the IEDGD bit, ICFD is set but data is not transferred to ICRD. Therefore, in buffer operation, ICFD can be used as an external interrupt by setting the ICIDE bit to 1.

ICFD must be cleared by software. It is set by hardware, however, and cannot be set by software.

Bit 4

ICFD	Description	
0	[Clearing condition]	(Initial value)
	Read ICFD when ICFD = 1, then write 0 in ICFD	
1	[Setting condition]	
	When an input capture signal is received	

Bit 3—Output Compare Flag A (OCFA): This status flag indicates that the FRC value matches the OCRA value. This flag must be cleared by software. It is set by hardware, however, and cannot be set by software.

Bit 3

OCFA	Description	
0	[Clearing condition]	(Initial value)
	Read OCFA when OCFA = 1, then write 0 in OCFA	
1	[Setting condition]	
	When FRC = OCRA	

Bit 2—Output Compare Flag B (OCFB): This status flag indicates that the FRC value matches the OCRB value. This flag must be cleared by software. It is set by hardware, however, and cannot be set by software.

Bit 2

OCFB	Description	
0	[Clearing condition]	(Initial value)
	Read OCFB when OCFB = 1, then write 0 in OCFB	
1	[Setting condition]	
	When FRC = OCRB	



Bit 1—Timer Overflow Flag (OVF): This status flag indicates that the FRC has overflowed (changed from H'FFFF to H'0000). This flag must be cleared by software. It is set by hardware, however, and cannot be set by software.

Bit 1

OVF	Description	
0	[Clearing condition]	(Initial value)
	Read OVF when OVF = 1, then write 0 in OVF	
1	[Setting condition]	
	When FRC changes from H'FFFF to H'0000	

Bit 0—Counter Clear A (CCLRA): This bit selects whether the FRC is to be cleared at comparematch A (when the FRC and OCRA values match).

Bit 0

CCLRA	Description	
0	FRC clearing is disabled	(Initial value)
1	FRC is cleared at compare-match A	

11.2.8 Timer Control Register (TCR)

Bit	7	6	5	4	3	2	1	0
	IEDGA	IEDGB	IEDGC	IEDGD	BUFEA	BUFEB	CKS1	CKS0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

TCR is an 8-bit readable/writable register that selects the rising or falling edge of the input capture signals, enables the input capture buffer mode, and selects the FRC clock source.

TCR is initialized to H'00 by a reset and in hardware standby mode

Bit 7—Input Edge Select A (IEDGA): Selects the rising or falling edge of the input capture A signal (FTIA).

Bit '	7
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IEDGA	 Description	
0	Capture on the falling edge of FTIA	(Initial value)
1	Capture on the rising edge of FTIA	

Bit 6—Input Edge Select B (IEDGB): Selects the rising or falling edge of the input capture B signal (FTIB).

Bit 6

IEDGB	Description	
0	Capture on the falling edge of FTIB	(Initial value)
1	Capture on the rising edge of FTIB	

Bit 5—Input Edge Select C (IEDGC): Selects the rising or falling edge of the input capture C signal (FTIC).

Bit 5

IEDGC	Description	
0	Capture on the falling edge of FTIC	(Initial value)
1	Capture on the rising edge of FTIC	

Bit 4—Input Edge Select D (IEDGD): Selects the rising or falling edge of the input capture D signal (FTID).

Bit 4

IEDGD	Description	
0	Capture on the falling edge of FTID	(Initial value)
1	Capture on the rising edge of FTID	

Bit 3—Buffer Enable A (BUFEA): Selects whether ICRC is to be used as a buffer register for ICRA.

Bit 3

BUFEA	Description	
0	ICRC is not used as a buffer register for input capture A	(Initial value)
1	ICRC is used as a buffer register for input capture A	

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Bit 2—Buffer Enable B (BUFEB): Selects whether ICRD is to be used as a buffer register for ICRB.

Bit 2

BUFEB	Description	
0	ICRD is not used as a buffer register for input capture B	(Initial value)
1	ICRD is used as a buffer register for input capture B	

Bits 1 and 0—Clock Select (CKS1, CKS0): Select external clock input or one of three internal clock sources for the FRC. External clock pulses are counted on the rising edge of signals input to the external clock input pin (FTCI).

Bit 1	Bit 0		
CKS1	CKS0	Description	
0	0	φ/2 internal clock source	(Initial value)
	1	φ/8 internal clock source	
1	0	φ/32 internal clock source	
	1	External clock source (rising edge)	

11.2.9 Timer Output Compare Control Register (TOCR)

Bit	7	6	5	4	3	2	1	0
	ICRDMS	OCRAMS	ICRS	OCRS	OEA	OEB	OLVLA	OLVLB
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

TOCR is an 8-bit readable/writable register that enables output from the output compare pins, selects the output levels, switches access between output compare registers A and B, controls the ICRD and OCRA operating mode, and switches access to input capture registers A, B, and C.

TOCR is initialized to H'00 by a reset and in hardware standby mode.

Bit 7—Input Capture D Mode Select (ICRDMS): Specifies whether ICRD is used in the normal operating mode or in the operating mode using OCRDM.

Bit 7

ICRDMS		
0	The normal operating mode is specified for ICRD	(Initial value)
1	The operating mode using OCRDM is specified for ICRD	

Bit 6—Output Compare A Mode Select (OCRAMS): Specifies whether OCRA is used in the normal operating mode or in the operating mode using OCRAR and OCRAF.

Bit 6

OCRAMS	Description	
0	The normal operating mode is specified for OCRA	(Initial value)
1	The operating mode using OCRAR and OCRAF is specified for OCRA	

Bit 5—Input Capture Register Select (ICRS): The same addresses are shared by ICRA and OCRAR, by ICRB and OCRAF, and by ICRC and OCRDM. The ICRS bit determines which registers are selected when the shared addresses are read or written to. The operation of ICRA, ICRB, and ICRC is not affected.

Bit 5

ICRS	Description	
0	The ICRA, ICRB, and ICRC registers are selected	(Initial value)
1	The OCRAR, OCRAF, and OCRDM registers are selected	

Bit 4—Output Compare Register Select (OCRS): OCRA and OCRB share the same address. When this address is accessed, the OCRS bit selects which register is accessed. This bit does not affect the operation of OCRA or OCRB.

Bit 4

OCRS	Description	
0	The OCRA register is selected	(Initial value)
1	The OCRB register is selected	

Bit 3—Output Enable A (OEA): Enables or disables output of the output compare A signal (FTOA).

Bit 3

OEA	Description	
0	Output compare A output is disabled	(Initial value)
1	Output compare A output is enabled	

Bit 2—Output Enable B (OEB): Enables or disables output of the output compare B signal (FTOB).

Bit 2

OEB	Description	
0	Output compare B output is disabled	(Initial value)
1	Output compare B output is enabled	

Bit 1—Output Level A (OLVLA): Selects the logic level to be output at the FTOA pin in response to compare-match A (signal indicating a match between the FRC and OCRA values). When the OCRAMS bit is 1, this bit is ignored.

Bit 1

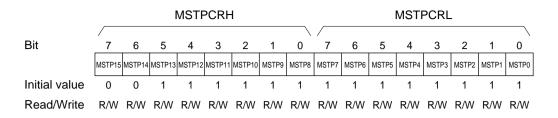
OLVLA	Description	
0	0 output at compare-match A	(Initial value)
1	1 output at compare-match A	

Bit 0—Output Level B (OLVLB): Selects the logic level to be output at the FTOB pin in response to compare-match B (signal indicating a match between the FRC and OCRB values).

Bit 0

OLVLB	Description	
0	0 output at compare-match B	(Initial value)
1	1 output at compare-match B	

11.2.10 Module Stop Control Register (MSTPCR)



MSTPCR, comprising two 8-bit readable/writable registers, performs module stop mode control.

When the MSTP13 bit is set to 1, FRT operation is stopped at the end of the bus cycle, and module stop mode is entered. For details, see section 25.5, Module Stop Mode.

MSTPCR is initialized to H'3FFF by a reset and in hardware standby mode. It is not initialized in software standby mode.

MSTPCRH Bit 5—Module Stop (MSTP13): Specifies the FRT module stop mode.

Bit 5

MSTPCRH	Description	
0	FRT module stop mode is cleared	
1	FRT module stop mode is set	(Initial value)

11.3 Operation

11.3.1 FRC Increment Timing

FRC increments on a pulse generated once for each period of the selected (internal or external) clock source.

Internal Clock

Any of three internal clocks ($\phi/2$, $\phi/8$, or $\phi/32$) created by division of the system clock (ϕ) can be selected by making the appropriate setting in bits CKS1 and CKS0 in TCR. Figure 11.3 shows the increment timing.

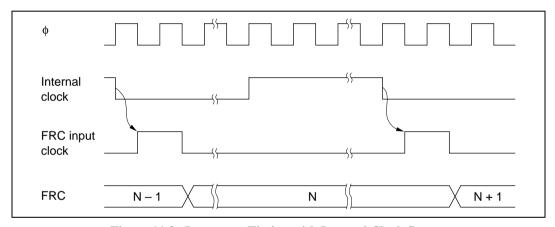


Figure 11.3 Increment Timing with Internal Clock Source

External Clock

If external clock input is selected by bits CKS1 and CKS0 in TCR, FRC increments on the rising edge of the external clock signal.

The pulse width of the external clock signal must be at least 1.5 system clock (ϕ) periods. The counter will not increment correctly if the pulse width is shorter than 1.5 system clock periods.

Figure 11.4 shows the increment timing.

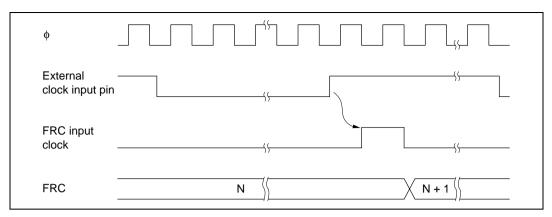


Figure 11.4 Increment Timing with External Clock Source

11.3.2 **Output Compare Output Timing**

When a compare-match occurs, the logic level selected by the output level bit (OLVLA or OLVLB) in TOCR is output at the output compare pin (FTOA or FTOB). Figure 11.5 shows the timing of this operation for compare-match A.

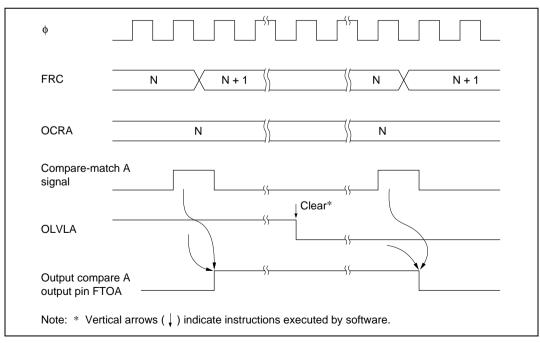


Figure 11.5 Timing of Output Compare A Output

11.3.3 FRC Clear Timing

FRC can be cleared when compare-match A occurs. Figure 11.6 shows the timing of this operation.

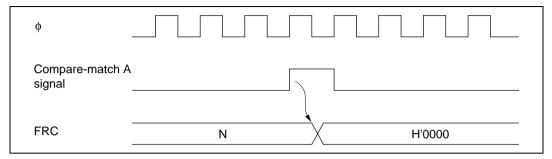


Figure 11.6 Clearing of FRC by Compare-Match A

11.3.4 Input Capture Input Timing

Input Capture Input Timing

An internal input capture signal is generated from the rising or falling edge of the signal at the input capture pin, as selected by the corresponding IEDGx (x = A to D) bit in TCR. Figure 11.7 shows the usual input capture timing when the rising edge is selected (IEDGx = 1).

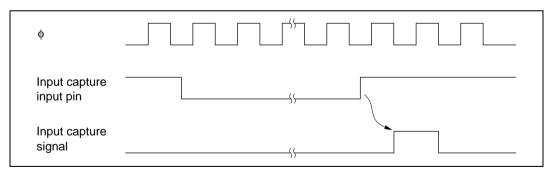


Figure 11.7 Input Capture Signal Timing (Usual Case)

If the upper byte of ICRA/B/C/D is being read when the corresponding input capture signal arrives, the internal input capture signal is delayed by one system clock (ϕ) period. Figure 11.8 shows the timing for this case.

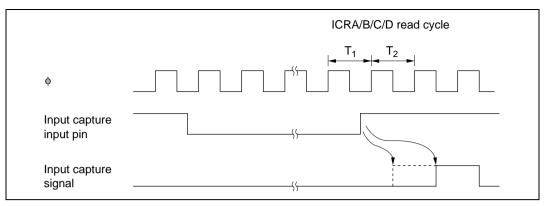


Figure 11.8 Input Capture Signal Timing (Input Capture Input when ICRA/B/C/D Is Read)

Buffered Input Capture Input Timing

ICRC and ICRD can operate as buffers for ICRA and ICRB.

Figure 11.9 shows how input capture operates when ICRA and ICRC are used in buffer mode and IEDGA and IEDGC are set to different values (IEDGA = 0 and IEDGC = 1, or IEDG A = 1 and IEDGC = 0), so that input capture is performed on both the rising and falling edges of FTIA.

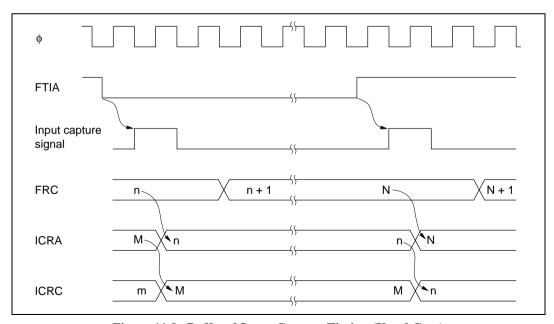


Figure 11.9 Buffered Input Capture Timing (Usual Case)

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When ICRC or ICRD is used as a buffer register, its input capture flag is set by the selected transition of its input capture signal. For example, if ICRC is used to buffer ICRA, when the edge transition selected by the IEDGC bit occurs on the FTIC input capture line, ICFC will be set, and if the ICIEC bit is set, an interrupt will be requested. The FRC value will not be transferred to ICRC, however.

In buffered input capture, if the upper byte of either of the two registers to which data will be transferred (ICRA and ICRC, or ICRB and ICRD) is being read when the input signal arrives, input capture is delayed by one system clock (ϕ) period. Figure 11.10 shows the timing when BUFEA = 1.

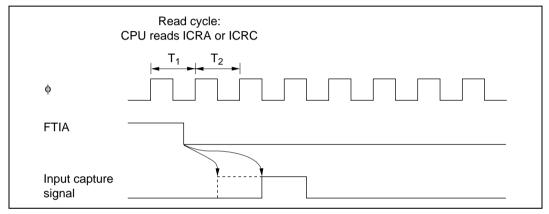


Figure 11.10 Buffered Input Capture Timing (Input Capture Input when ICRA or ICRC Is Read)

11.3.5 Timing of Input Capture Flag (ICF) Setting

The input capture flag ICFx (x = A, B, C, D) is set to 1 by the internal input capture signal. The FRC value is simultaneously transferred to the corresponding input capture register (ICRx). Figure 11.11 shows the timing of this operation.

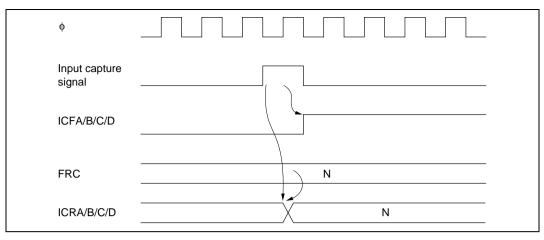


Figure 11.11 Setting of Input Capture Flag (ICFA/B/C/D)

11.3.6 Setting of Output Compare Flags A and B (OCFA, OCFB)

The output compare flags are set to 1 by an internal compare-match signal generated when the FRC value matches the OCRA or OCRB value. This compare-match signal is generated at the last state in which the two values match, just before FRC increments to a new value.

Accordingly, when the FRC and OCR values match, the compare-match signal is not generated until the next period of the clock source. Figure 11.12 shows the timing of the setting of OCFA and OCFB.

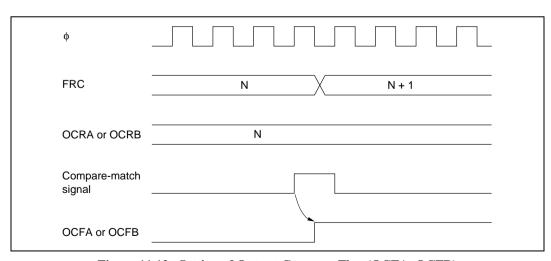


Figure 11.12 Setting of Output Compare Flag (OCFA, OCFB)

11.3.7 Setting of FRC Overflow Flag (OVF)

The FRC overflow flag (OVF) is set to 1 when FRC overflows (changes from H'FFFF to H'0000). Figure 11.13 shows the timing of this operation.

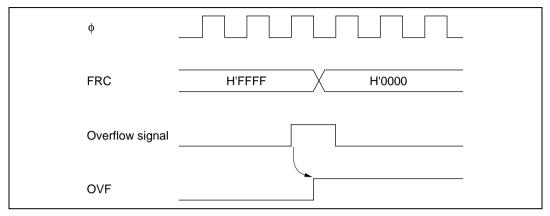


Figure 11.13 Setting of Overflow Flag (OVF)

11.3.8 Automatic Addition of OCRA and OCRAR/OCRAF

When the OCRAMS bit in TOCR is set to 1, the contents of OCRAR and OCRAF are automatically added to OCRA alternately, and when an OCRA compare-match occurs a write to OCRA is performed. The OCRA write timing is shown in figure 11.14.

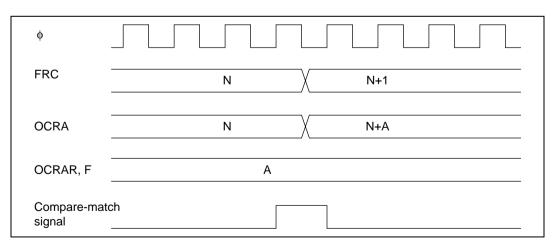


Figure 11.14 OCRA Automatic Addition Timing

11.3.9 ICRD and OCRDM Mask Signal Generation

When the ICRDMS bit in TOCR is set to 1 and the contents of OCRDM are other than H'0000, a signal that masks the ICRD input capture function is generated.

The mask signal is set by the input capture signal. The mask signal setting timing is shown in figure 11.15.

The mask signal is cleared by the sum of the ICRD contents and twice the OCRDM contents, and an FRC compare-match. The mask signal clearing timing is shown in figure 11.16.

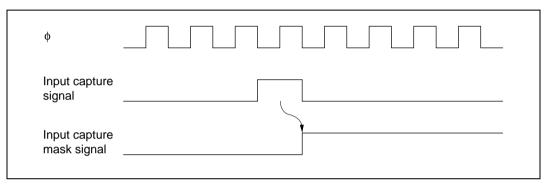


Figure 11.15 Input Capture Mask Signal Setting Timing

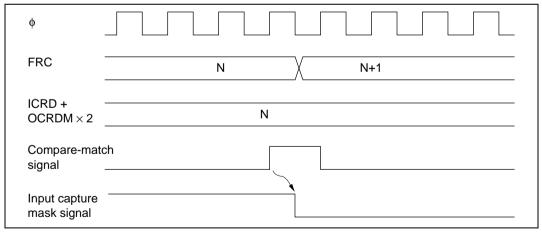


Figure 11.16 Input Capture Mask Signal Clearing Timing

11.4 Interrupts

The free-running timer can request seven interrupts (three types): input capture A to D (ICIA, ICIB, ICIC, ICID), output compare A and B (OCIA and OCIB), and overflow (FOVI). Each interrupt can be enabled or disabled by an enable bit in TIER. Independent signals are sent to the interrupt controller for each interrupt. Table 11.4 lists information about these interrupts.

Table 11.4 Free-Running Timer Interrupts

Interrupt	Description	DTC Activation	Priority
ICIA	Requested by ICFA	Possible	High
ICIB	Requested by ICFB	Possible	<u> </u>
ICIC	Requested by ICFC	Not possible	
ICID	Requested by ICFD	Not possible	
OCIA	Requested by OCFA	Possible	
OCIB	Requested by OCFB	Possible	
FOVI	Requested by OVF	Not possible	Low

11.5 Sample Application

In the example below, the free-running timer is used to generate pulse outputs with a 50% duty cycle and arbitrary phase relationship. The programming is as follows:

- The CCLRA bit in TCSR is set to 1.
- Each time a compare-match interrupt occurs, software inverts the corresponding output level bit in TOCR (OLVLA or OLVLB).

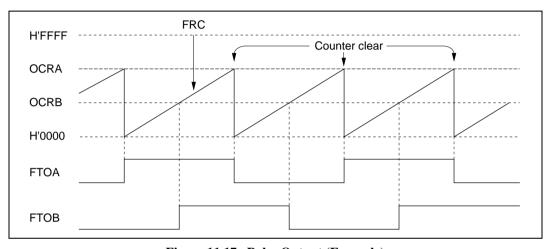


Figure 11.17 Pulse Output (Example)



11.6 Usage Notes

Application programmers should note that the following types of contention can occur in the freerunning timer.

Contention between FRC Write and Clear

If an internal counter clear signal is generated during the state after an FRC write cycle, the clear signal takes priority and the write is not performed.

Figure 11.18 shows this type of contention.

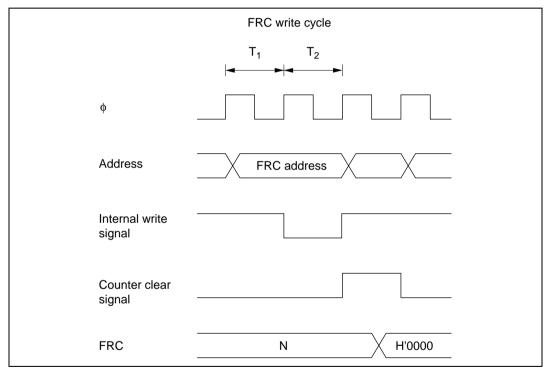


Figure 11.18 FRC Write-Clear Contention

Contention between FRC Write and Increment

If an FRC increment pulse is generated during the state after an FRC write cycle, the write takes priority and FRC is not incremented.

Figure 11.19 shows this type of contention.

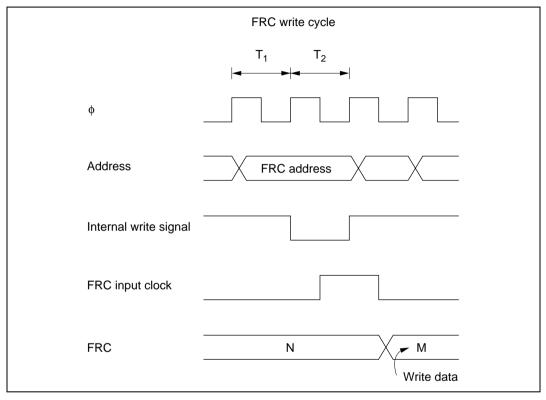


Figure 11.19 FRC Write-Increment Contention

Contention between OCR Write and Compare-Match

If a compare-match occurs during the state after an OCRA or OCRB write cycle, the write takes priority and the compare-match signal is inhibited.

Figure 11.20 shows this type of contention.

If automatic addition of OCRAR/OCRAF to OCRA is selected, and a compare-match occurs in the cycle following the OCRA, OCRAR and OCRAF write cycle, the OCRA, OCRAR and OCRAF write takes priority and the compare-match signal is inhibited. Consequently, the result of the automatic addition is not written to OCRA.

Figure 11.21 shows this type of contention.

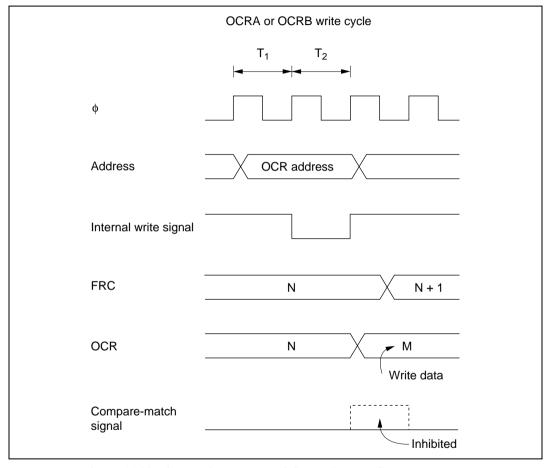


Figure 11.20 Contention between OCR Write and Compare-Match (When Automatic Addition Function Is Not Used)

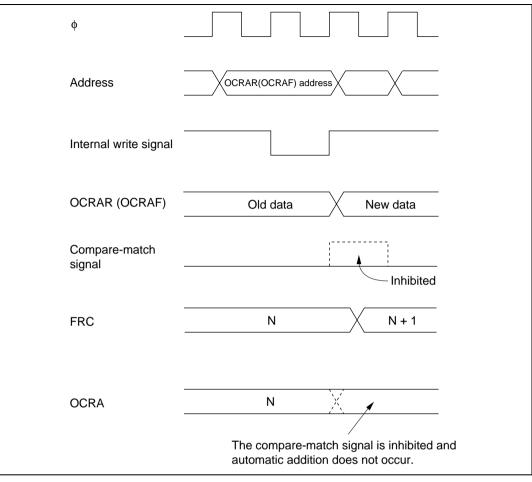


Figure 11.21 Contention between OCRAR/OCRAF Write and Compare-Match (When Automatic Addition Function Is Not Used)

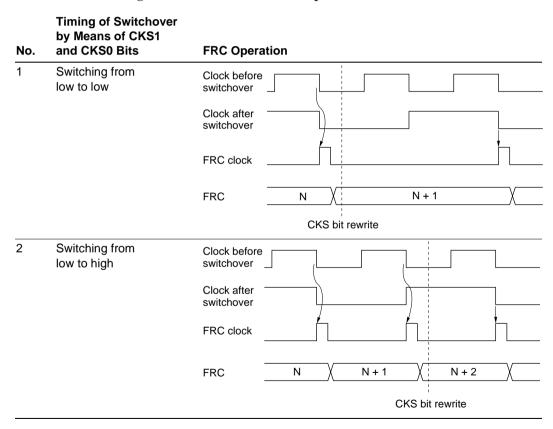
Switching of Internal Clock and FRC Operation

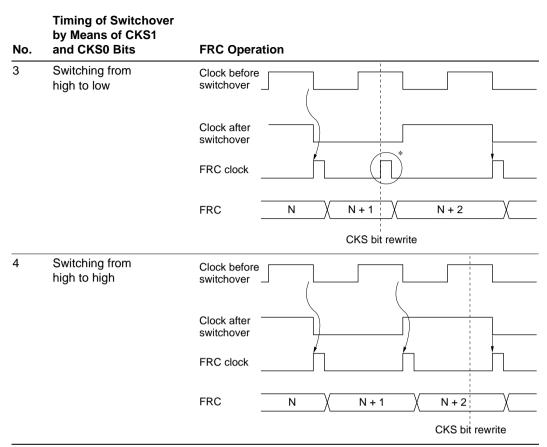
When the internal clock is changed, the changeover may cause FRC to increment. This depends on the time at which the clock select bits (CKS1 and CKS0) are rewritten, as shown in table 11.5.

When an internal clock is used, the FRC clock is generated on detection of the falling edge of the internal clock scaled from the system clock (ϕ). If the clock is changed when the old source is high and the new source is low, as in case no. 3 in table 11.5, the changeover is regarded as a falling edge that triggers the FRC increment clock pulse.

Switching between an internal and external clock can also cause FRC to increment.

Table 11.5 Switching of Internal Clock and FRC Operation





Generated on the assumption that the switchover is a falling edge; FRC is incremented. Note:



Section 12 8-Bit Timers

12.1 Overview

This LSI include an 8-bit timer module with two channels (TMR0 and TMR1). Each channel has an 8-bit counter (TCNT) and two time constant registers (TCORA and TCORB) that are constantly compared with the TCNT value to detect compare-matches. The 8-bit timer module can be used as a multifunction timer in a variety of applications, such as generation of a rectangular-wave output with an arbitrary duty cycle.

The H8S/2148 Group also has two similar 8-bit timer channels (TMRX and TMRY), and the H8S/2144 Group and H8S/2147N has one (TMRY). These channels can be used in a connected configuration using the timer connection function. TMRX and TMRY have greater input/output and interrupt function related restrictions than TMR0 and TMR1.

12.1.1 Features

- Selection of clock sources
 - TMR0, TMR1: The counter input clock can be selected from six internal clocks and an external clock (enabling use as an external event counter).
 - TMRX, TMRY: The counter input clock can be selected from three internal clocks and an external clock (enabling use as an external event counter).
- Selection of three ways to clear the counters
 - The counters can be cleared on compare-match A or B, or by an external reset signal.
- Timer output controlled by two compare-match signals
 - The timer output signal in each channel is controlled by two independent compare-match signals, enabling the timer to be used for various applications, such as the generation of pulse output or PWM output with an arbitrary duty cycle.

(Note: TMRY does not have a timer output pin.)

- Cascading of the two channels (TMR0, TMR1)
 - Operation as a 16-bit timer can be performed using channel 0 as the upper half and channel 1 as the lower half (16-bit count mode).
 - Channel 1 can be used to count channel 0 compare-match occurrences (compare-match count mode).
- Multiple interrupt sources for each channel
 - TMR0, TMR1, TMRY: Two compare-match interrupts and one overflow interrupt can be requested independently.
 - TMRX: One input capture source is available.

12.1.2 Block Diagram

Figure 12.1 shows a block diagram of the 8-bit timer module (TMR0 and TMR1).

TMRX and TMRY have a similar configuration, but cannot be cascaded. TMRX also has an input capture function. For details, see section 13, Timer Connection.

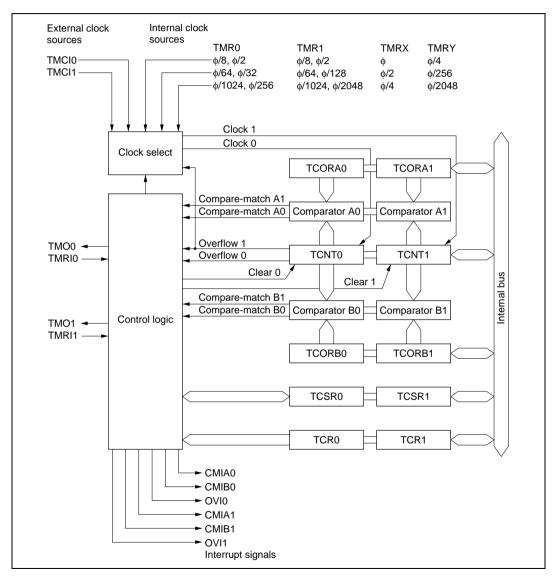


Figure 12.1 Block Diagram of 8-Bit Timer Module

12.1.3 Pin Configuration

Table 12.1 summarizes the input and output pins of the 8-bit timer module.

Table 12.1 8-Bit Timer Input and Output Pins

Channel	Name	Symbol*	I/O	Function
0	Timer output	TMO0	Output	Output controlled by compare-match
	Timer clock input	TMCI0	Input	External clock input for the counter
	Timer reset input	TMRI0	Input	External reset input for the counter
1	Timer output	TMO1	Output	Output controlled by compare-match
	Timer clock input	TMCI1	Input	External clock input for the counter
	Timer reset input	TMRI1	Input	External reset input for the counter
X	Timer output	TMOX	Output	Output controlled by compare-match
	Timer clock/ reset input	HFBACKI/TMIX (TMCIX/TMRIX)	Input	External clock/reset input for the counter
Y	Timer clock/reset input	VSYNCI/TMIY (TMCIY/TMRIY)	Input	External clock/reset input for the counter

Note: * The abbreviations TMO, TMCI, and TMRI are used in the text, omitting the channel number.

Channel X and Y I/O pins have the same internal configuration as channels 0 and 1, and therefore the same abbreviations are used.

12.1.4 **Register Configuration**

Table 12.2 summarizes the registers of the 8-bit timer module.

Table 12.2 8-Bit Timer Registers

Name	Abbreviation*3	R/W	Initial value	Address*1
Timer control register 0	TCR0	R/W	H'00	H'FFC8
Timer control/status register 0	TCSR0	R/(W)*2	H'00	H'FFCA
Time constant register A0	TCORA0	R/W	H'FF	H'FFCC
Time constant register B0	TCORB0	R/W	H'FF	H'FFCE
Time counter 0	TCNT0	R/W	H'00	H'FFD0
Timer control register 1	TCR1	R/W	H'00	H'FFC9
Timer control/status register 1	TCSR1	R/(W)*2	H'10	H'FFCB
Time constant register A1	TCORA1	R/W	H'FF	H'FFCD
Time constant register B1	TCORB1	R/W	H'FF	H'FFCF
Timer counter 1	TCNT1	R/W	H'00	H'FFD1
Serial/timer control register	STCR	R/W	H'00	H'FFC3
Module stop control register	MSTPCRH	R/W	H'3F	H'FF86
	MSTPCRL	R/W	H'FF	H'FF87
Timer connection register S	TCONRS	R/W	H'00	H'FFFE
Timer control register X	TCRX	R/W	H'00	H'FFF0
Timer control/status register X	TCSRX	R/(W)*2	H'00	H'FFF1
Time constant register AX	TCORAX	R/W	H'FF	H'FFF6
Time constant register BX	TCORBX	R/W	H'FF	H'FFF7
Timer counter X	TCNTX	R/W	H'00	H'FFF4
Time constant register C	TCORC	R/W	H'FF	H'FFF5
Input capture register R	TICRR	R	H'00	H'FFF2
Input capture register F	TICRF	R	H'00	H'FFF3
Timer control register Y	TCRY	R/W	H'00	H'FFF0
Timer control/status register Y	TCSRY	R/(W)*2	H'00	H'FFF1
Time constant register AY	TCORAY	R/W	H'FF	H'FFF2
Time constant register BY	TCORBY	R/W	H'FF	H'FFF3
Timer counter Y	TCNTY	R/W	H'00	H'FFF4
Timer input select register	TISR	R/W	H'FE	H'FFF5
	Timer control register 0 Timer constant register A0 Time constant register B0 Time constant register B0 Time counter 0 Timer control register 1 Timer control/status register 1 Timer constant register A1 Time constant register B1 Timer constant register B1 Timer counter 1 Serial/timer control register Module stop control register Timer connection register S Timer control/status register X Timer control/status register X Timer constant register AX Time constant register BX Timer constant register BX Timer counter X Time constant register C Input capture register C Input capture register F Timer control/status register Y Timer constant register AY Timer constant register AY Timer constant register AY Timer constant register BY Timer constant register BY Timer counter Y	Timer control register 0 TCR0 Timer control/status register 0 TCSR0 Time constant register A0 TCORA0 Time constant register B0 TCORB0 Time counter 0 TCNT0 Timer control register 1 TCR1 Timer control/status register 1 TCSR1 Time constant register A1 TCORA1 Time constant register B1 TCORB1 Timer counter 1 TCNT1 Serial/timer control register Module stop control register MSTPCRL Timer connection register S TCONRS Timer control register X TCRX Timer control/status register X TCRX Timer control/status register X TCSRX Timer control/status register X TCORAX Time constant register AX TCORAX Time constant register BX TCORBX Timer counter X TCNTX Time constant register C TCORC Input capture register F TICRR Input capture register Y TCRY Timer control/status register Y TCSRY Timer constant register AY TCORAY Time constant register BY TCORBY Timer constant register BY TCORBY Timer constant register BY TCORBY Timer constant register BY TCORBY	Timer control/status register 0 TCR0 R/W Timer control/status register 0 TCSR0 R/(W)*² Time constant register A0 TCORA0 R/W Time constant register B0 TCORB0 R/W Time counter 0 TCNT0 R/W Timer control register 1 TCR1 R/W Timer control/status register 1 TCSR1 R/W Time constant register A1 TCORA1 R/W Time constant register B1 TCORB1 R/W Timer counter 1 TCNT1 R/W Serial/timer control register STCR R/W Module stop control register STCR R/W MSTPCRH R/W MSTPCRH R/W Timer connection register S TCONRS R/W Timer control/status register X TCRX R/W Time constant register AX TCORAX R/W Time constant register BX TCORC R/W Time constant register C TCORC R/W Input capture register F TICRF R	Timer control register 0 TCR0 R/W H'00 Timer control/status register 0 TCSR0 R/(W)*2 H'00 Time constant register A0 TCORA0 R/W H'FF Time constant register B0 TCORB0 R/W H'FF Time constant register B0 TCNT0 R/W H'00 Time control register 1 TCR1 R/W H'00 Timer control register 1 TCSR1 R/W H'FF Timer control/status register A1 TCORA1 R/W H'FF Time constant register B1 TCORB1 R/W H'FF Timer counter 1 TCNT1 R/W H'00 Serial/timer control register STCR R/W H'00 Module stop control register MSTPCRH R/W H'3F MSTPCRL R/W H'3F MSTPCRL R/W H'00 Timer connection register S TCONRS R/W H'00 Timer control/status register X TCSRX R/W H'00 Time constant register BX

- Notes: 1. Lower 16 bits of the address.
 - 2. Only 0 can be written in bits 7 to 5, to clear these flags.
 - 3. The abbreviations TCR, TCSR, TCORA, TCORB, and TCNT are used in the text, omitting the channel designation (0, 1, X, or Y).

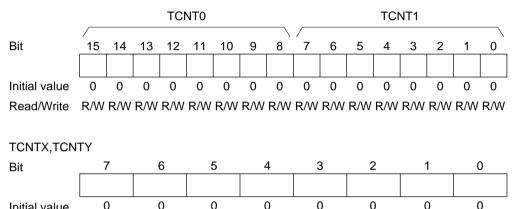


Each pair of registers for channel 0 and channel 1 comprises a 16-bit register with the upper 8 bits for channel 0 and the lower 8 bits for channel 1, so they can be accessed together by word access. (Access is not divided into two 8-bit accesses.)

In the H8S/2148 Group, certain of the channel X and channel Y registers are assigned to the same address. The TMRX/Y bit in TCONRS determines which register is accessed.

12.2 Register Descriptions

12.2.1 Timer Counter (TCNT)



 Initial value
 0
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Each TCNT is an 8-bit readable/writable up-counter.

TCNT0 and TCNT1 comprise a single 16-bit register, so they can be accessed together by word access.

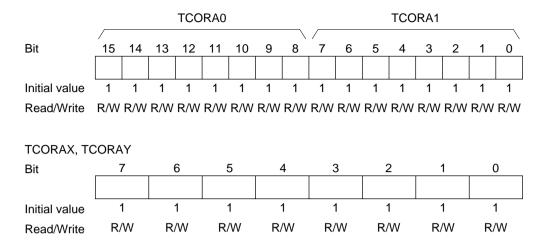
TCNT increments on pulses generated from an internal or external clock source. This clock source is selected by clock select bits CKS2 to CKS0 in TCR.

TCNT can be cleared by an external reset input signal or compare-match signal. Counter clear bits CCLR1 and CCLR0 in TCR select the method of clearing.

When TCNT overflows from H'FF to H'00, the overflow flag (OVF) in TCSR is set to 1.

The timer counters are initialized to H'00 by a reset and in hardware standby mode.

12.2.2 Time Constant Register A (TCORA)



TCORA is an 8-bit readable/writable register.

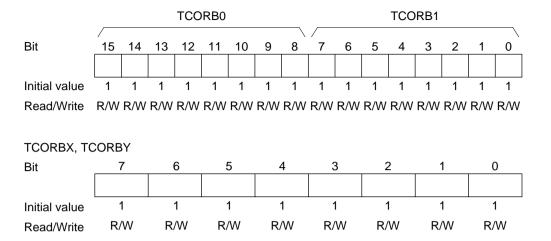
TCORA0 and TCORA1 comprise a single 16-bit register, so they can be accessed together by word access.

TCORA is continually compared with the value in TCNT. When a match is detected, the corresponding compare-match flag A (CMFA) in TCSR is set. Note, however, that comparison is disabled during the T2 state of a TCORA write cycle.

The timer output can be freely controlled by these compare-match signals and the settings of output select bits OS1 and OS0 in TCSR.

TCORA is initialized to H'FF by a reset and in hardware standby mode.

12.2.3 Time Constant Register B (TCORB)



TCORB is an 8-bit readable/writable register. TCORB0 and TCORB1 comprise a single 16-bit register, so they can be accessed together by word access.

TCORB is continually compared with the value in TCNT. When a match is detected, the corresponding compare-match flag B (CMFB) in TCSR is set. Note, however, that comparison is disabled during the T2 state of a TCORB write cycle.

The timer output can be freely controlled by these compare-match signals and the settings of output select bits OS3 and OS2 in TCSR.

TCORB is initialized to H'FF by a reset and in hardware standby mode.

12.2.4 Timer Control Register (TCR)

Bit	7	6	5	4	3	2	1	0
	CMIEB	CMIEA	OVIE	CCLR1	CCLR0	CKS2	CKS1	CKS0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

TCR is an 8-bit readable/writable register that selects the clock source and the time at which TCNT is cleared, and enables interrupts.

TCR is initialized to H'00 by a reset and in hardware standby mode.

For details of the timing, see section 12.3, Operation.

Bit 7—Compare-Match Interrupt Enable B (CMIEB): Selects whether the CMFB interrupt request (CMIB) is enabled or disabled when the CMFB flag in TCSR is set to 1.

Note that a CMIB interrupt is not requested by TMRX, regardless of the CMIEB value.

Bit 7

CMIEB	Description	
0	CMFB interrupt request (CMIB) is disabled	(Initial value)
1	CMFB interrupt request (CMIB) is enabled	

Bit 6—Compare-Match Interrupt Enable A (CMIEA): Selects whether the CMFA interrupt request (CMIA) is enabled or disabled when the CMFA flag in TCSR is set to 1.

Note that a CMIA interrupt is not requested by TMRX, regardless of the CMIEA value.

Bit 6

CMIEA	Description	
0	CMFA interrupt request (CMIA) is disabled	(Initial value)
1	CMFA interrupt request (CMIA) is enabled	



Bit 5—Timer Overflow Interrupt Enable (OVIE): Selects whether the OVF interrupt request (OVI) is enabled or disabled when the OVF flag in TCSR is set to 1.

Note that an OVI interrupt is not requested by TMRX, regardless of the OVIE value.

Bit 5

OVIE	Description	
0	OVF interrupt request (OVI) is disabled	(Initial value)
1	OVF interrupt request (OVI) is enabled	

Bits 4 and 3—Counter Clear 1 and 0 (CCLR1, CCLR0): These bits select the method by which the timer counter is cleared: by compare-match A or B, or by an external reset input.

Bit 4	Bit 3		
CCLR1	CCLR0	 Description	
0	0	Clearing is disabled	(Initial value)
	1	Cleared on compare-match A	
1	0	Cleared on compare-match B	
	1	Cleared on rising edge of external reset input	

Bits 2 to 0—Clock Select 2 to 0 (CKS2 to CKS0): These bits select whether the clock input to TCNT is an internal or external clock.

The input clock can be selected from either six or three clocks, all divided from the system clock (φ). The falling edge of the selected internal clock triggers the count.

When use of an external clock is selected, three types of count can be selected: at the rising edge, the falling edge, and both rising and falling edges.

Some functions differ between channel 0 and channel 1, because of the cascading function.

	TCR Bit 2 Bit 1 Bit 0			STCR Bit 1 Bit 0					
Channel	CKS2	CKS1	CKS0	ICKS1	ICKS0	Description			
0	0	0	0	_	_	Clock input disabled (Initial value)			
	0	0	1	_	0	φ/8 internal clock source, counted on the falling edge			
	0	0	1	_	1	φ/2 internal clock source, counted on the falling edge			
	0	1	0	_	0				
	0	1	0	_	1				
	0	1	1	_	0				
	0	1	1	_	1				
	1	0	0	_	_	Counted on TCNT1 overflow signal*			
1	0	0	0 0 —		_	Clock input disabled (Initial value)			
	0	0	1	0	_	φ/8 internal clock source, counted on the falling edge			
	0	0	1	1	_	φ/2 internal clock source, counted on the falling edge			
	0	1	0	0	_				
	0	1	0	1	_	φ/128 internal clock source, counted on the falling edge			
	0	1	1	0	_				
	0	1	1	1	_	φ/2048 internal clock source, counted on the falling edge			
	1	0	0	_	_	Counted on TCNT0 compare-match A*			

	TCR			STCR		
	Bit 2	Bit 1	Bit 0	Bit 1	Bit 0	•
Channel	CKS2	CKS1	CKS0	ICKS1	ICKS0	Description
X	0	0	0	_	_	Clock input disabled (Initial value)
	0	0	1	_	_	Counted on ϕ internal clock source
	0	1	0	_	_	φ/2 internal clock source, counted on the falling edge
	0	1	1	_	_	φ/4 internal clock source, counted on the falling edge
	1	0	0	_	_	Clock input disabled
Y	0	0	0	_	_	Clock input disabled (Initial value)
	0	0	1	_	_	φ/4 internal clock source, counted on the falling edge
	0	1	0	_	_	φ/256 internal clock source, counted on the falling edge
	0	1	1	_	_	φ/2048 internal clock source, counted on the falling edge
	1	0	0	_	_	Clock input disabled
Common	1	0	1	_	_	External clock source, counted at rising edge
	1	1	0	_	_	External clock source, counted at falling edge
	1	1	1	_	_	External clock source, counted at both rising and falling edges

Note: * If the count input of channel 0 is the TCNT1 overflow signal and that of channel 1 is the TCNT0 compare-match signal, no incrementing clock will be generated. Do not use this setting.

12.2.5 Timer Control/Status Register (TCSR)

TCSR0								
Bit	7	6	5	4	3	2	1	0
	CMFB	CMFA	OVF	ADTE	OS3	OS2	OS1	OS0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/(W)*	R/(W)*	R/(W)*	R/W	R/W	R/W	R/W	R/W
TCSR1								
Bit	7	6	5	4	3	2	1	0
	CMFB	CMFA	OVF	_	OS3	OS2	OS1	OS0
Initial value	0	0	0	1	0	0	0	0
Read/Write	R/(W)*	R/(W)*	R/(W)*	_	R/W	R/W	R/W	R/W
TCSRX								
Bit	7	6	5	4	3	2	1	0
	CMFB	CMFA	OVF	ICF	OS3	OS2	OS1	OS0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/(W)*	R/(W)*	R/(W)*	R/(W)*	R/W	R/W	R/W	R/W
TCSRY								
Bit	7	6	5	4	3	2	1	0
	CMFB	CMFA	OVF	ICIE	OS3	OS2	OS1	OS0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/(W)*	R/(W)*	R/(W)*	R/W	R/W	R/W	R/W	R/W

Note: * Only 0 can be written in bits 7 to 5, and in bit 4 in TCSRX, to clear these flags.

TCSR is an 8-bit register that indicates compare-match and overflow statuses (and input capture status in TMRX only), and controls compare-match output.

TCSR0, TCSRX, and TCSRY are initialized to H'00, and TCSR1 is initialized to H'10, by a reset and in hardware standby mode.

Bit 7—Compare-Match Flag B (CMFB): Status flag indicating whether the values of TCNT and TCORB match.

Bit 7

CMFB	 Description						
0	[Clearing conditions] (Initial						
	 Read CMFB when CMFB = 1, then write 0 in CMFB 						
	 When the DTC is activated by a CMIB interrupt 						
1	[Setting condition]						
	When TCNT = TCORB						

Bit 6—Compare-match Flag A (CMFA): Status flag indicating whether the values of TCNT and TCORA match.

Bit 6

CMFA	Description	
0	[Clearing conditions]	(Initial value)
	 Read CMFA when CMFA = 1, then write 0 in CMFA 	
	 When the DTC is activated by a CMIA interrupt 	
1	[Setting condition]	
	When TCNT = TCORA	

Bit 5—**Timer Overflow Flag (OVF):** Status flag indicating that TCNT has overflowed (changed from H'FF to H'00).

OVF	Description	
0	[Clearing condition]	(Initial value)
	Read OVF when OVF = 1, then write 0 in OVF	
1	[Setting condition]	
	When TCNT overflows from H'FF to H'00	

TCSR0

Bit 4—A/D Trigger Enable (ADTE): Enables or disables A/D converter start requests by compare-match A.

Bit 4

ADTE	Description	
0	A/D converter start requests by compare-match A are disabled	(Initial value)
1	A/D converter start requests by compare-match A are enabled	

TCSR1

Bit 4—Reserved: This bit cannot be modified and is always read as 1.

TCSRX

Bit 4—Input Capture Flag (ICF): Status flag that indicates detection of a rising edge followed by a falling edge in the external reset signal after the ICST bit in TCONRI has been set to 1.

Bit 4

ICF	Description	
0	[Clearing condition]	(Initial value)
	Read ICF when ICF = 1, then write 0 in ICF	
1	[Setting condition]	
	When a rising edge followed by a falling edge is detected in th after the ICST bit in TCONRI has been set to 1	e external reset signal

TCSRY

Bit 4—Input Capture Interrupt Enable (ICIE): Selects enabling or disabling of the interrupt request by ICF (ICIX) when the ICF bit in TCSRX is set to 1.

ICIE	Description	
0	Interrupt request by ICF (ICIX) is disabled	(Initial value)
1	Interrupt request by ICF (ICIX) is enabled	



Bits 3 to 0—Output Select 3 to 0 (OS3 to OS0): These bits specify how the timer output level is to be changed by a compare-match of TCOR and TCNT.

OS3 and OS2 select the effect of compare-match B on the output level, OS1 and OS0 select the effect of compare-match A on the output level, and both of them can be controlled independently.

Note, however, that priorities are set such that: trigger output > 1 output > 0 output. If comparematches occur simultaneously, the output changes according to the compare-match with the higher priority.

Timer output is disabled when bits OS3 to OS0 are all 0.

After a reset, the timer output is 0 until the first compare-match occurs.

Bit 3	Bit 2		
OS3	OS2	Description	
0	0	No change when compare-match B occurs	(Initial value)
	1	0 is output when compare-match B occurs	
1	0	1 is output when compare-match B occurs	
	1	Output is inverted when compare-match B occurs (toggle output)

Bit 1	Bit 0		
OS1	OS0	Description	
0	0	No change when compare-match A occurs	(Initial value)
	1	0 is output when compare-match A occurs	
1	0	1 is output when compare-match A occurs	
	1	Output is inverted when compare-match A occurs (toggle output)

12.2.6 Serial/Timer Control Register (STCR)

Bit	7	6	5	4	3	2	1	0
	IICS	IICX1	IICX0	IICE	FLSHE	_	ICKS1	ICKS0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

STCR is an 8-bit readable/writable register that controls register access, the IIC operating mode (when the on-chip IIC option is included), and on-chip flash memory (in F-ZTAT versions), and also selects the TCNT input clock.

For details on functions not related to the 8-bit timers, see section 3.2.4, Serial Timer Control Register (STCR), and the descriptions of the relevant modules. If a module controlled by STCR is not used, do not write 1 to the corresponding bit.

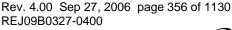
STCR is initialized to H'00 by a reset and in hardware standby mode.

Bits 7 to 4—I²C Control (IICS, IICX1, IICX0, IICE): These bits control the bus buffer function of port A and the operation of the I²C bus interface when the IIC option is included on-chip. See section 16.2.7, Serial/Timer Control Register (STCR), for details.

Bit 3—Flash Memory Control Register Enable (FLSHE): Controls CPU access to the flash memory control registers, the power-down mode control registers, and the supporting module control registers. See section 3.2.4, Serial Timer Control Register (STCR), for details..

Bit 2—Reserved: Do not write 1 to this bit.

Bits 1 and 0—Internal Clock Select 1 and 0 (ICKS1, ICKS0): These bits, together with bits CKS2 to CKS0 in TCR, select the clock to be input to TCNT. For details, see section 12.2.4, Timer Control Register (TCR).





12.2.7 System Control Register (SYSCR)

Bit	7	6	5	4	3	2	1	0
	CS2E	IOSE	INTM1	INTM0	XRST	NMIEG	HIE	RAME
Initial value	0	0	0	0	1	0	0	1
Read/Write	R/W	R/W	R	R/W	R	R/W	R/W	R/W

Only bit 1 is described here. For details on functions not related to the 8-bit timers, see sections 3.2.2 and 5.2.1, System Control Register (SYSCR), and the descriptions of the relevant modules.

Bit 1—Host Interface Enable (HIE): Controls CPU access to 8-bit timer (channel X and Y) data registers and control registers, and timer connection control registers.

Bit 1

HIE	Description
0	CPU access to 8-bit timer (channel X and Y) data registers and control registers, and timer connection control registers, is enabled (Initial value)
1	CPU access to 8-bit timer (channel X and Y) data registers and control registers, and timer connection control registers, is disabled

12.2.8 **Timer Connection Register S (TCONRS)**

Bit	7	6	5	4	3	2	1	0
	TMRX/Y	ISGENE	HOMOD1	HOMOD0	VOMOD1	VOMOD0	CLMOD1	CLMOD0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W							

TCONRS is an 8-bit readable/writable register that controls access to the TMRX and TMRY registers and timer connection operation.

TCONRS is initialized to H'00 by a reset and in hardware standby mode.

Bit 7—TMRX/TMRY Access Select (TMRX/Y): The TMRX and TMRY registers can only be accessed when the HIE bit in SYSCR is cleared to 0. In the H8S/2148 Group, some of the TMRX registers and the TMRY registers are assigned to the same memory space addresses (H'FFF0 to H'FFF5), and the TMRX/Y bit determines which registers are accessed. In the H8S/2144 Group and H8S/2147N, there is no control of TMRY register access by this bit.

Bit 7	Accessible Registers							
TMRX/Y	H'FFF0	H'FFF1	H'FFF2	H'FFF3	H'FFF4	H'FFF5	H'FFF6	H'FFF7
0	TMRX	TMRX	TMRX	TMRX	TMRX	TMRX	TMRX	TMRX
(Initial value)	TCRX	TCSRX	TICRR	TICRF	TCNTX	TCORC	TCORAX	TCORBX
1	TMRY	TMRY	TMRY	TMRY	TMRY	TMRY		
	TCRY	TCSRY	TCORAY	TCORBY	TCNTY	TISR		

12.2.9 Input Capture Register (TICR) [TMRX Additional Function]

Bit	7	6	5	4	3	2	1	0
Initial value	0	0	0	0	0	0	0	0
Read/Write	_		_	_	_	_	_	_

TICR is an 8-bit internal register to which the contents of TCNT are transferred on the falling edge of external reset input. The CPU cannot read or write to TICR directly.

The TICR function is used in timer connection. For details, see section 13, Timer Connection.



12.2.10 Time Constant Register C (TCORC) [TMRX Additional Function]

Bit	7	6	5	4	3	2	1	0
Initial value	1	1	1	1	1	1	1	1
Read/Write	R/W							

TCORC is an 8-bit readable/writable register. The sum of the contents of TCORC and TICR is continually compared with the value in TCNT. When a match is detected, a compare-match C signal is generated. Note, however, that comparison is disabled during the T2 state of a TCORC write cycle and a TICR input capture cycle.

TCORC is initialized to H'FF by a reset and in hardware standby mode.

The TCORC function is used in timer connection. For details, see section 13, Timer Connection.

12.2.11 Input Capture Registers R and F (TICRR, TICRF) [TMRX Additional Functions]

Bit	7	6	5	4	3	2	1	0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R	R	R	R	R	R	R	R

TICRR and TICRF are 8-bit read-only registers. When the ICST bit in TCONRI is set to 1, TICRR and TICRF capture the contents of TCNT successively on the rise and fall of the external reset input. When one capture operation ends, the ICST bit is cleared to 0.

TICRR and TICRF are each initialized to H'00 by a reset and in hardware standby mode.

The TICRR and TICRF functions are used in timer connection. For details, see section 12.3.6, Input Capture Operation, and section 13, Timer Connection.

Timer Input Select Register (TISR) [TMRY Additional Function] 12.2.12

Bit	7	6	5	4	3	2	1	0
	_	_	_	_	_	_	_	IS
Initial value	1	1	1	1	1	1	1	0
Read/Write	_	_	_	_	_	_	_	R/W

TISR is an 8-bit readable/writable register that selects the external clock/reset signal source for the counter.

TISR is initialized to H'FE by a reset and in hardware standby mode.

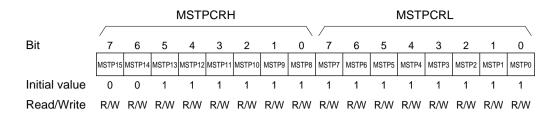
Bits 7 to 1—Reserved: Do not write 0 to these bits.

Bit 0—Input Select (IS): Selects the internal synchronization signal (IVG signal) or the timer clock/reset input pin (VSYNCI/TMIY (TMCIY/TMRIY)) as the external clock/reset signal source for the counter.

IS	Description	
0	IVG signal is selected (H8S/2148 Group) (Initial va	ılue)
	External clock/reset input is disabled (H8S/2144 Group and H8S/2147N)	
1	VSYNCI/TMIY (TMCIY/TMRIY) is selected	



12.2.13 Module Stop Control Register (MSTPCR)



MSTPCR comprises two 8-bit readable/writable registers, and is used to perform module stop mode control.

When the MSTP12 bit or MSTP8 bit is set to 1, 8-bit timer operation is halted on channels 0 and 1 or channels X and Y, respectively, and a transition is made to module stop mode. For details, see section 25.5, Module Stop Mode.

MSTPCR is initialized to H'3FFF by a reset and in hardware standby mode. It is not initialized in software standby mode.

MSTPCRH Bit 4—Module Stop (MSTP12): Specifies 8-bit timer (channel 0/1) module stop mode.

MSTPCRH Bit 4

MSTP12	Description	
0	8-bit timer (channel 0/1) module stop mode is cleared	
1	8-bit timer (channel 0/1) module stop mode is set	(Initial value)

MSTPCRH Bit 0—Module Stop (MSTP8): Specifies 8-bit timer (channel X/Y) and timer connection module stop mode.

MSTPCRH Bit 0

MSTP8	Description	
0	8-bit timer (channel X/Y) and timer connection module stop mode is	s cleared
1	8-bit timer (channel X/Y) and timer connection module stop mode is set	(Initial value)

12.3 Operation

12.3.1 TCNT Incrementation Timing

TCNT is incremented by input clock pulses (either internal or external).

Internal Clock

An internal clock created by dividing the system clock (ϕ) can be selected by setting bits CKS2 to CKS0 in TCR. Figure 12.2 shows the count timing.

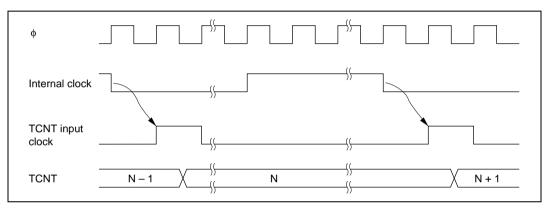


Figure 12.2 Count Timing for Internal Clock Input

External Clock

Three incrementation methods can be selected by setting bits CKS2 to CKS0 in TCR: at the rising edge, the falling edge, and both rising and falling edges.

Note that the external clock pulse width must be at least 1.5 states for incrementation at a single edge, and at least 2.5 states for incrementation at both edges. The counter will not increment correctly if the pulse width is less than these values.

Figure 12.3 shows the timing of incrementation at both edges of an external clock signal.

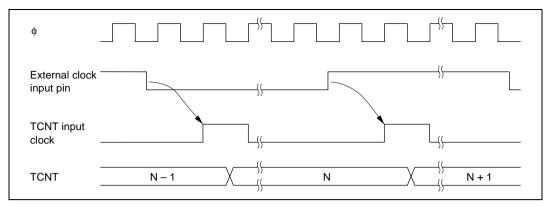


Figure 12.3 Count Timing for External Clock Input

12.3.2 Compare-Match Timing

Setting of Compare-Match Flags A and B (CMFA, CMFB)

The CMFA and CMFB flags in TCSR are set to 1 by a compare-match signal generated when the TCOR and TCNT values match. The compare-match signal is generated at the last state in which the match is true, just before the timer counter is updated.

Therefore, when TCOR and TCNT match, the compare-match signal is not generated until the next incrementation clock input. Figure 12.4 shows this timing.

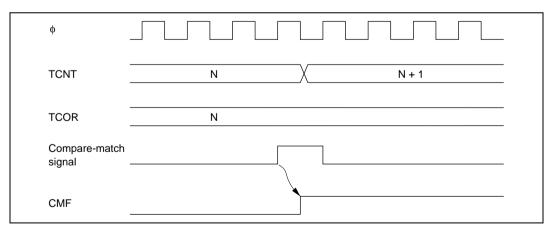


Figure 12.4 Timing of CMF Setting

Timer Output Timing

When compare-match A or B occurs, the timer output changes as specified by the output select bits (OS3 to OS0) in TCSR. Depending on these bits, the output can remain the same, be set to 0, be set to 1, or toggle.

Figure 12.5 shows the timing when the output is set to toggle at compare-match A.

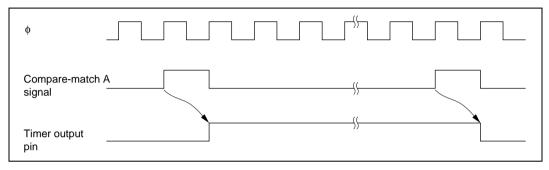


Figure 12.5 Timing of Timer Output

Timing of Compare-Match Clear

TCNT is cleared when compare-match A or B occurs, depending on the setting of the CCLR1 and CCLR0 bits in TCR. Figure 12.6 shows the timing of this operation.

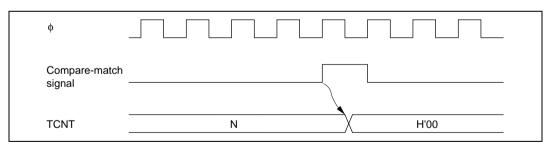


Figure 12.6 Timing of Compare-Match Clear

12.3.3 TCNT External Reset Timing

TCNT is cleared at the rising edge of an external reset input, depending on the settings of the CCLR1 and CCLR0 bits in TCR. The width of the clearing pulse must be at least 1.5 states. Figure 12.7 shows the timing of this operation.

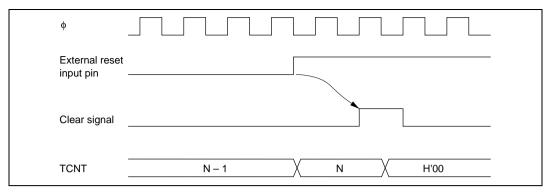


Figure 12.7 Timing of Clearing by External Reset Input

12.3.4 Timing of Overflow Flag (OVF) Setting

OVF in TCSR is set to 1 when the timer count overflows (changes from H'FF to H'00). Figure 12.8 shows the timing of this operation.

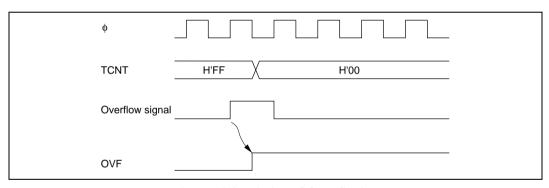


Figure 12.8 Timing of OVF Setting

12.3.5 Operation with Cascaded Connection

If bits CKS2 to CKS0 in either TCR0 or TCR1 are set to B'100, the 8-bit timers of the two channels are cascaded. With this configuration, a single 16-bit timer can be used (16-bit timer mode) or compare-matches of 8-bit channel 0 can be counted by the timer of channel 1 (compare-match count mode). In this case, the timer operates as described below.

16-Bit Count Mode

When bits CKS2 to CKS0 in TCR0 are set to B'100, the timer functions as a single 16-bit timer with channel 0 occupying the upper 8 bits and channel 1 occupying the lower 8 bits.

- Setting of compare-match flags
 - The CMF flag in TCSR0 is set to 1 when a 16-bit compare-match occurs.
 - The CMF flag in TCSR1 is set to 1 when a lower 8-bit compare-match occurs.
- Counter clear specification
 - If the CCLR1 and CCLR0 bits in TCR0 have been set for counter clear at compare-match, the 16-bit counter (TCNT0 and TCNT1 together) is cleared when a 16-bit compare-match occurs. The 16-bit counter (TCNT0 and TCNT1 together) is cleared even if counter clear by the TMRI0 pin has also been set.
 - The settings of the CCLR1 and CCLR0 bits in TCR1 are ignored. The lower 8 bits cannot be cleared independently.
- Pin output
 - Control of output from the TMO0 pin by bits OS3 to OS0 in TCSR0 is in accordance with the 16-bit compare-match conditions.
 - Control of output from the TMO1 pin by bits OS3 to OS0 in TCSR1 is in accordance with the lower 8-bit compare-match conditions.

Compare-Match Count Mode

When bits CKS2 to CKS0 in TCR1 are B'100, TCNT1 counts compare-match A's for channel 0.

Channels 0 and 1 are controlled independently. Conditions such as setting of the CMF flag, generation of interrupts, output from the TMO pin, and counter clearing are in accordance with the settings for each channel.

Usage Note

If the 16-bit count mode and compare-match count mode are set simultaneously, the input clock pulses for TCNT0 and TCNT1 are not generated and thus the counters will stop operating. Simultaneous setting of these two modes should therefore be avoided.



12.3.6 Input Capture Operation

TMRX has input capture registers of TICR, TICRR, and TICRF. Narrow pulse width can be measured with TICRR and TICRF, using one capture operation controlled by the ICST bit in the TCONRI register of the timer connection. When TMRIX detects a rising and falling edge successively after the ICST bit has been set to 1, the values of TCNT at that time are transferred to TICRR and TICRF and ICST bit is cleared to 0.

The TMRIX input signal can be selected by setting other bits in the TCONRI register.

(1) Input capture input timing

Figure 12.9 shows the timing of the input capture operation.

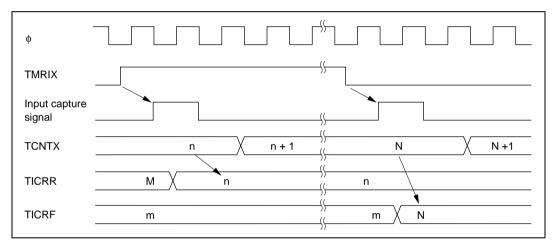


Figure 12.9 Timing of Input Capture Operation

If the input capture signal enters while TICRR and TICRF are being read, it is internally delayed one system clock (ϕ) period. Figure 12.10 shows the timing of this operation.

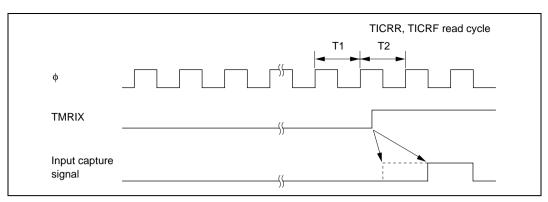


Figure 12.10 Timing of Input Capture Signal (When Input Capture Input Signal Enters while TICRR and TICRF Are Being Read)

(2) Input capture signal input selection

Input capture input signal (TMRIX) in TMRX is switched by setting bits in the TCONRI register.

Figure 12.11 and Table 12.3 show the input capture signal selections.

See section 13.2.1, Timer Connection Register I (TCONRI), for details.

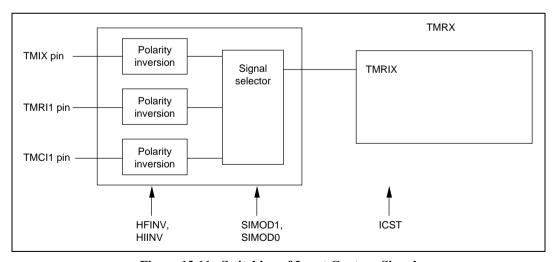


Figure 12.11 Switching of Input Capture Signal

Table 12.3 Input Capture Signal Selection

т	C	n	N	P	i
	u	u	14	\mathbf{r}	ı

Bit 4	Bit 7	Bit 6	Bit 3	Bit 1	
ICST	SIMOD1	SIMOD0	HFINV	HIINV	 Description
0	_	_	_	_	Input capture function not used
1	0	0	0	_	TMIX pin input signal
			1	_	Inverted signal of TMIX pin input
		1	_	0	TMRI1 pin input signal
			_	1	Inverted signal of TMRI1 pin input
	1	1	_	0	TMCI1 pin input signal
			_	1	Inverted signal of TMCI1 pin input

12.4 Interrupt Sources

The TMR0, TMR1, and TMRY 8-bit timers can generate three types of interrupt: compare-match A and B (CMIA and CMIB), and overflow (OVI). TMRX can generate only an ICIX interrupt. An interrupt is requested when the corresponding interrupt enable bit is set in TCR or TCSR. Independent signals are sent to the interrupt controller for each interrupt. It is also possible to activate the DTC by means of CMIA and CMIB interrupts from TMR0, TMR1 and TMRY.

An overview of 8-bit timer interrupt sources is given in tables 12.4 to 12.6.

Table 12.4 TMR0 and TMR1 8-Bit Timer Interrupt Sources

Interrupt source	Description	DTC Activation	Interrupt Priority
CMIA	Requested by CMFA	Possible	High
CMIB	Requested by CMFB	Possible	— ↑
OVI	Requested by OVF	Not possible	Low

Table 12.5 TMRX 8-Bit Timer Interrupt Source

Interrupt source	Description	DTC Activation
ICIX	Requested by ICF	Not possible

Table 12.6 TMRY 8-Bit Timer Interrupt Sources

Interrupt source	Description	DTC Activation	Interrupt Priority
CMIA	Requested by CMFA	Possible	High
CMIB	Requested by CMFB	Possible	↑
OVI	Requested by OVF	Not possible	Low

12.5 8-Bit Timer Application Example

In the example below, the 8-bit timer is used to generate a pulse output with a selected duty cycle, as shown in figure 12.12. The control bits are set as follows:

- In TCR, CCLR1 is cleared to 0 and CCLR0 is set to 1 so that the timer counter is cleared by a TCORA compare-match.
- In TCSR, bits OS3 to OS0 are set to B'0110, causing 1 output at a TCORA compare-match and 0 output at a TCORB compare-match.

With these settings, the 8-bit timer provides output of pulses at a rate determined by TCORA with a pulse width determined by TCORB. No software intervention is required.

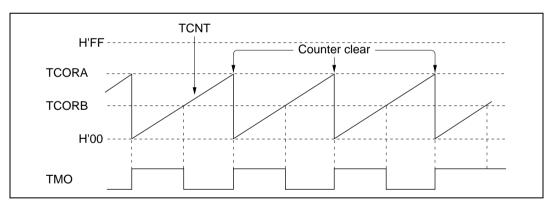


Figure 12.12 Pulse Output (Example)



12.6 Usage Notes

Application programmers should note that the following kinds of contention can occur in the 8-bit timer module.

12.6.1 Contention between TCNT Write and Clear

If a timer counter clock pulse is generated during the T2 state of a TCNT write cycle, the clear takes priority, so that the counter is cleared and the write is not performed. Figure 12.13 shows this operation.

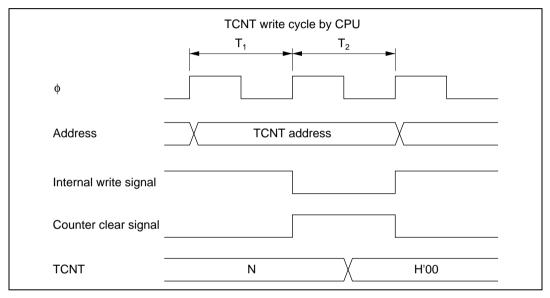


Figure 12.13 Contention between TCNT Write and Clear

12.6.2 Contention between TCNT Write and Increment

If a timer counter clock pulse is generated during the T2 state of a TCNT write cycle, the write takes priority and the counter is not incremented. Figure 12.14 shows this operation.

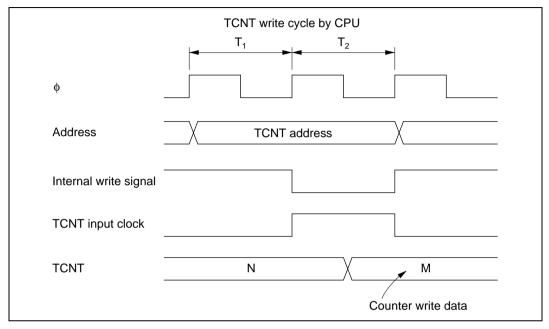


Figure 12.14 Contention between TCNT Write and Increment

12.6.3 Contention between TCOR Write and Compare-Match

During the T2 state of a TCOR write cycle, the TCOR write has priority even if a compare-match occurs and the compare-match signal is disabled. Figure 12.15 shows this operation.

With TMRX, an ICR input capture contends with a compare-match in the same way as with a write to TCORC. In this case, the input capture has priority and the compare-match signal is inhibited.

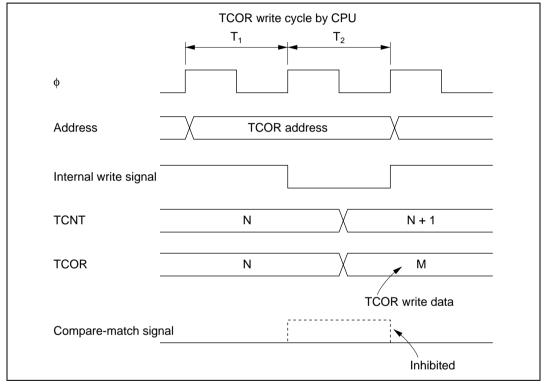


Figure 12.15 Contention between TCOR Write and Compare-Match

12.6.4 Contention between Compare-Matches A and B

If compare-matches A and B occur at the same time, the 8-bit timer operates in accordance with the priorities for the output states set for compare-match A and compare-match B, as shown in table 12.7.

Table 12.7 Timer Output Priorities

Output Setting	Priority
Toggle output	High
1 output	↑
0 output	
No change	Low

12.6.5 Switching of Internal Clocks and TCNT Operation

TCNT may increment erroneously when the internal clock is switched over. Table 12.8 shows the relationship between the timing at which the internal clock is switched (by writing to the CKS1 and CKS0 bits) and the TCNT operation

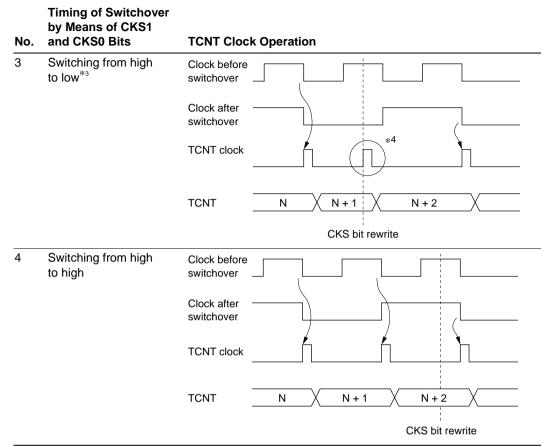
When the TCNT clock is generated from an internal clock, the falling edge of the internal clock pulse is detected. If clock switching causes a change from high to low level, as shown in no. 3 in table 12.8, a TCNT clock pulse is generated on the assumption that the switchover is a falling edge. This increments TCNT.

Erroneous incrementation can also happen when switching between internal and external clocks.



Table 12.8 Switching of Internal Clock and TCNT Operation

No.	Timing of Switchover by Means of CKS1 and CKS0 Bits	TCNT Clock Operation
1	Switching from low to low*1	Clock before switchover
		Clock after switchover
		TCNT clock
		TCNT N N+1
		CKS bit rewrite
2	Switching from low to high*2	Clock before switchover
		Clock after switchover
		TCNT clock
		TCNT N N+1 N+2
		CKS bit rewrite



Notes: 1. Includes switching from low to stop, and from stop to low.

- 2. Includes switching from stop to high.
- 3. Includes switching from high to stop.
- Generated on the assumption that the switchover is a falling edge; TCNT is incremented.

Section 13 Timer Connection

Provided in the H8S/2148 Group; not provided in the H8S/2144 Group and H8S/2147N.

13.1 Overview

The H8S/2148 Group allows interconnection between a combination of input signals, the input/output of the single free-running timer (FRT) channel and the three 8-bit timer channels (TMR1, TMRX, and TMRY). This capability can be used to implement complex functions such as PWM decoding and clamp waveform output. All the timers are initially set for independent operation.

13.1.1 Features

The features of the timer connection facility are as follows.

- Five input pins and four output pins, all of which can be designated for phase inversion.
 Positive logic is assumed for all signals used within the timer connection facility.
- An edge-detection circuit is connected to the input pins, simplifying signal input detection.
- TMRX can be used for PWM input signal decoding and clamp waveform generation.
- An external clock signal divided by TMR1 can be used as the FRT capture input signal.
- An internal synchronization signal can be generated using the FRT and TMRY.
- A signal generated/modified using an input signal and timer connection can be selected and output.

13.1.2 Block Diagram

Figure 13.1 shows a block diagram of the timer connection facility.

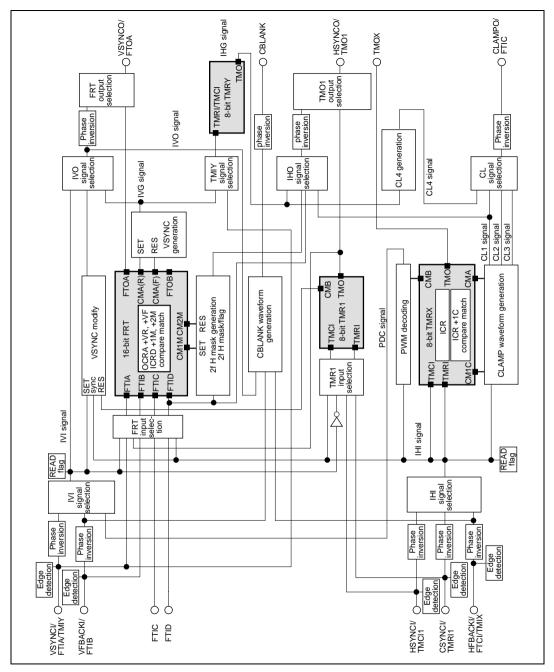


Figure 13.1 Block Diagram of Timer Connection Facility

13.1.3 Input and Output Pins

Table 13.1 lists the timer connection input and output pins.

Table 13.1 Timer Connection Input and Output Pins

Name	Abbreviation	Input/ Output	Function
Vertical synchronization signal input pin	VSYNCI	Input	Vertical synchronization signal input pin or FTIA input pin/TMIY input pin
Horizontal synchronization signal input pin	HSYNCI	Input	Horizontal synchronization signal input pin or TMCI1 input pin
Composite synchronization signal input pin	CSYNCI	Input	Composite synchronization signal input pin or TMRI1 input pin
Spare vertical synchronization signal input pin	VFBACKI	Input	Spare vertical synchronization signal input pin or FTIB input pin
Spare horizontal synchronization signal input pin	HFBACKI	Input	Spare horizontal synchronization signal input pin or FTCI input pin/TMIX input pin
Vertical synchronization signal output pin	VSYNCO	Output	Vertical synchronization signal output pin or FTOA output pin
Horizontal synchronization signal output pin	HSYNCO	Output	Horizontal synchronization signal output pin or TMO1 output pin
Clamp waveform output pin	CLAMPO	Output	Clamp waveform output pin or FTIC input pin
Blanking waveform output pin	CBLANK	Output	Blanking waveform output pin

13.1.4 Register Configuration

Table 13.2 lists the timer connection registers. Timer connection registers can only be accessed when the HIE bit in SYSCR is 0.

Table 13.2 Register Configuration

Name	Abbreviation	R/W	Initial Value	Address*1
Timer connection register I	TCONRI	R/W	H'00	H'FFFC
Timer connection register O	TCONRO	R/W	H'00	H'FFFD
Timer connection register S	TCONRS	R/W	H'00	H'FFFE
Edge sense register	SEDGR	R/(W)*2	H'00*3	H'FFFF
Module stop control register	MSTPRH	R/W	H'3F	H'FF86
	MSTPRL	R/W	H'FF	H'FF87

Notes: 1. Lower 16 bits of the address.

- 2. Bits 7 to 2: Only 0 can be written to clear the flags.
- 3. Bits 1 and 0: Undefined (reflect the pin states).

13.2 Register Descriptions

13.2.1 Timer Connection Register I (TCONRI)

Bit	7	6	5	4	3	2	1	0
	SIMOD1	SIMOD0	SCONE	ICST	HFINV	VFINV	HIINV	VIINV
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

TCONRI is an 8-bit readable/writable register that controls connection between timers, the signal source for synchronization signal input, phase inversion, etc.

TCONR1 is initialized to H'00 by a reset and in hardware standby mode.

Bits 7 and 6—Input Synchronization Mode Select 1 and 0 (SIMOD1, SIMOD0): These bits select the signal source of the IHI and IVI signals.

Bit 7	Bit 6	Description				
SIMOD1	SIMOD0	Mode		IHI Signal	IVI Signal	
0	0	No signal	(Initial value)	HFBACKI input	VFBACKI input	
	1	S-on-G mode	е	CSYNCI input	PDC input	
1	0	Composite n	node	HSYNCI input	PDC input	
	1	Separate mo	Separate mode		VSYNCI input	

Bit 5—Synchronization Signal Connection Enable (SCONE): Selects the signal source of the FRT FTI input and the TMR1 TMCI1/TMRI1 input.

Bit 5	Description							
SCONE	Mode	FTIA	FTIB	FTIC	FTID	TMCI1	TMRI1	
0	Normal connection (Initial value)	FTIA input	FTIB input	FTIC input	FTID input	TMCI1 input	TMRI1 input	
1	Synchronization signal connection mode	IVI signal	TMO1 signal	VFBACKI input	IHI signal	IHI signal	IVI inverse signal	

Bit 4—Input Capture Start Bit (ICST): The TMRX external reset input (TMRIX) is connected to the IHI signal. TMRX has input capture registers (TICR, TICRR, and TICRF). TICRR and TICRF can measure the width of a short pulse by means of a single capture operation under the control of the ICST bit. When a rising edge followed by a falling edge is detected on TMRIX after the ICST bit is set to 1, the contents of TCNT at those points are captured into TICRR and TICRF, respectively, and the ICST bit is cleared to 0.

Bit 4

ICST	Description			
0	The TICRR and TICRF input capture functions are stopped	(Initial value)		
	[Clearing condition]			
	When a rising edge followed by a falling edge is detected on TMRIX			
1	The TICRR and TICRF input capture functions are operating (Waiting for detection of a rising edge followed by a falling edge on TM	ИRIX)		
	[Setting condition]			
	When 1 is written in ICST after reading ICST = 0			

Bits 3 to 0—Input Synchronization Signal Inversion (HFINV, VFINV, HIINV, VIINV):

These bits select inversion of the input phase of the spare horizontal synchronization signal (HFBACKI), the spare vertical synchronization signal (VFBACKI), the horizontal synchronization signal and composite synchronization signal (HSYNCI, CSYNCI), and the vertical synchronization signal (VSYNCI).

Bit 3

HFINV	Description	
0	The HFBACKI pin state is used directly as the HFBACKI input	(Initial value)
1	The HFBACKI pin state is inverted before use as the HFBACKI input	

Bit 2

VFINV	Description	
0	The VFBACKI pin state is used directly as the VFBACKI input	(Initial value)
1	The VFBACKI pin state is inverted before use as the VFBACKI input	

Bit 1

HIINV	Description	
0	The HSYNCI and CSYNCI pin states are used directly as the HSYNCI and CSYNCI inputs	(Initial value)
1	The HSYNCI and CSYNCI pin states are inverted before use as the HS CSYNCI inputs	SYNCI and

VIINV	Description	
0	The VSYNCI pin state is used directly as the VSYNCI input	(Initial value)
1	The VSYNCI pin state is inverted before use as the VSYNCI input	

13.2.2 Timer Connection Register O (TCONRO)

Bit	7	6	5	4	3	2	1	0
	HOE	VOE	CLOE	CBOE	HOINV	VOINV	CLOINV	CBOINV
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

TCONRO is an 8-bit readable/writable register that controls output signal output, phase inversion, etc.

TCONRO is initialized to H'00 by a reset and in hardware standby mode.

Bits 7 and 4—Output Enable (HOE, VOE, CLOE, CBOE): These bits control enabling/disabling of horizontal synchronization signal (HSYNCO), vertical synchronization signal (VSYNCO), clamp waveform (CLAMPO), and blanking waveform (CBLANK) output. When output is disabled, the state of the relevant pin is determined by the port DR and DDR, FRT, TMR, and PWM settings.

Output enabling/disabling control does not affect the port, FRT, or TMR input functions, but some FRT and TMR input signal sources are determined by the SCONE bit in TCONRI.

Bit 7

HOE	Description	
0	The P44/TMO1/HIRQ1/HSYNCO pin functions as the P44/TMO1/HIRQ1 pin	(Initial value)
1	The P44/TMO1/HIRQ1/HSYNCO pin functions as the HSYNCO pin	

VOE	Description	
0	The P61/FTOA/KIN1/CIN1/VSYNCO pin functions as the P61/FTOA/KIN1/CIN1 pin	(Initial value)
1	The P61/FTOA/KIN1/CIN1/VSYNCO pin functions as the VSYNCO pin	

Bit 5

Dit		
CLOE	Description	
0	The P64/FTIC/KIN4/CIN4/CLAMPO pin functions as the P64/FTIC/KIN4/CIN4 pin	(Initial value)
1	The P64/FTIC/KIN4/CIN4/CLAMPO pin functions as the CLAMPO pin	

Bit 4

CBOE	Description	
0	The P27/A15/PW15/CBLANK pin functions as the P27/A15/PW15 pin	(Initial value)
1	In mode 1 (expanded mode with on-chip ROM disabled): The P27/A15/PW15/CBLANK pin functions as the A15 pin	
	In modes 2 and 3 (modes with on-chip ROM enabled): The P27/A15/PW15/CBLANK pin functions as the CBLANK pin	

Bits 3 to 0—Output Synchronization Signal Inversion (HOINV, VOINV, CLOINV,

CBOINV): These bits select inversion of the output phase of the horizontal synchronization signal (HSYNCO), the vertical synchronization signal (VSYNCO), the clamp waveform (CLAMPO), and the blank waveform (CBLANK).

Bit 3

HOINV	Description	
0	The IHO signal is used directly as the HSYNCO output	(Initial value)
1	The IHO signal is inverted before use as the HSYNCO output	

Bit 2

VOINV	Description	
0	The IVO signal is used directly as the VSYNCO output	(Initial value)
1	The IVO signal is inverted before use as the VSYNCO output	

CLOINV		
0	The CLO signal (CL1, CL2, CL3, or CL4 signal) is used directly as the CLAMPO output (Initial value)	ıe)
1	The CLO signal (CL1, CL2, CL3, or CL4 signal) is inverted before use as the CLAMPO output	

Bit 0		
CBOINV	Description	
0	The CBLANK signal is used directly as the CBLANK output	(Initial value)
1	The CBLANK signal is inverted before use as the CBLANK output	

13.2.3 **Timer Connection Register S (TCONRS)**

Bit	7	6	5	4	3	2	1	0
	TMRX/Y	ISGENE	HOMOD1	HOMOD0	VOMOD1	VOMOD0	CLMOD1	CLMOD0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W							

TCONRS is an 8-bit readable/writable register that selects 8-bit timer TMRX/TMRY access and the synchronization signal output signal source and generation method.

TCONRS is initialized to H'00 by a reset and in hardware standby mode.

Bit 7—TMRX/TMRY Access Select (TMRX/Y): The TMRX and TMRY registers can only be accessed when the HIE bit in SYSCR is cleared to 0. In the H8S/2148 Group, some of the TMRX registers and the TMRY registers are assigned to the same memory space addresses (H'FFF0 to H'FFF5), and the TMRX/Y bit determines which registers are accessed. In the H8S/2144 Group and H8S/2147N, there is no control of TMRY register access by this bit.

Bit 7

TMRX/Y	 Description	
0	The TMRX registers are accessed at addresses H'FFF0 to H'FFF5	(Initial value)
1	The TMRY registers are accessed at addresses H'FFF0 to H'FFF5	

Bit 6—Internal Synchronization Signal Select (ISGENE): Selects internal synchronization signals (IHG, IVG, and CL4 signals) as the signal sources for the IHO, IVO, and CLO signals.

Bits 5 and 4—Horizontal Synchronization Output Mode Select 1 and 0 (HOMOD1,

HOMOD0): These bits select the signal source and generation method for the IHO signal.

Bit 6	Bit 5	Bit 4	
ISGENE	HOMOD1	HOMOD0	Description
0	0	0	The IHI signal (without 2fH modification) is selected (Initial value)
		1	The IHI signal (with 2fH modification) is selected
	1	0	The CL1 signal is selected
		1	
1	0	0	The IHG signal is selected
		1	
	1	0	
		1	

Bits 3 and 2—Vertical Synchronization Output Mode Select 1 and 0 (VOMOD1, VOMOD0):

These bits select the signal source and generation method for the IVO signal.

Bit 6	Bit 3	Bit 2				
ISGENE	VOMOD1	VOMOD0	Description			
0	0	0	The IVI signal (without fall modification or IHI synchronization) is selected (Initial value)			
		1	The IVI signal (without fall modification, with IHI synchronization) is selected			
	1	0	The IVI signal (with fall modification, without IHI synchronization) is selected			
		1	The IVI signal (with fall modification and IHI synchronization) is selected			
1	0	0	The IVG signal is selected			
		1				
	1	0				
		1				



Bits 1 and 0—Clamp Waveform Mode Select 1 and 0 (CLMOD1, CLMOD0): These bits select the signal source for the CLO signal (clamp waveform).

Bit 6	Bit 1	Bit 0			
ISGENE	CLMOD1	CLMOD2	Description		
0	0	0	The CL1 signal is selected	(Initial value)	
		1	The CL2 signal is selected		
	1	0	The CL3 signal is selected		
		1	<u> </u>		
1	0	0	The CL4 signal is selected		
		1	<u> </u>		
	1	0	<u> </u>		
		1			

13.2.4 Edge Sense Register (SEDGR)

Bit	7	6	5	4	3	2	1	0
	VEDG	HEDG	CEDG	HFEDG	VFEDG	PREQF	IHI	IVI
Initial value	0	0	0	0	0	0	<u></u> *2	<u></u> *2
Read/Write	R/(W)*1	R/(W)*1	R/(W)*1	R/(W)*1	R/(W)*1	R/(W)*1	R	R

Notes: 1. Only 0 can be written, to clear the flags.

2. The initial value is undefined since it depends on the pin states.

SEDGR is an 8-bit readable/writable register used to detect a rising edge on the timer connection input pins and the occurrence of 2fH modification, and to determine the phase of the IVI and IHI signals.

The upper 6 bits of SEDGR are initialized to 0 by a reset and in hardware standby mode. The initial value of the lower 2 bits is undefined, since it depends on the pin states.

Bit 7—VSYNCI Edge (VEDG): Detects a rising edge on the VSYNCI pin.

Bit 7

VEDG	Description	
0	[Clearing condition]	(Initial value)
	When 0 is written in VEDG after reading VEDG = 1	
1	[Setting condition]	
	When a rising edge is detected on the VSYNCI pin	

Bit 6—HSYNCI Edge (HEDG): Detects a rising edge on the HSYNCI pin.

Bit 6

HEDG	Description	
0	[Clearing condition]	(Initial value)
	When 0 is written in HEDG after reading HEDG = 1	
1	[Setting condition]	
	When a rising edge is detected on the HSYNCI pin	

Bit 5—CSYNCI Edge (CEDG): Detects a rising edge on the CSYNCI pin.

Bit 5

CEDG	Description	
0	[Clearing condition]	(Initial value)
	When 0 is written in CEDG after reading CEDG = 1	
1	[Setting condition]	
	When a rising edge is detected on the CSYNCI pin	

Bit 4—HFBACKI Edge (HFEDG): Detects a rising edge on the HFBACKI pin.

HFEDG	 Description	
0	[Clearing condition]	(Initial value)
	When 0 is written in HFEDG after reading HFEDG = 1	
1	[Setting condition]	
	When a rising edge is detected on the HFBACKI pin	

Bit 3—VFBACKI Edge (VFEDG): Detects a rising edge on the VFBACKI pin.

Bit 3

VFEDG	Description				
0	[Clearing condition] (Initial value)				
	When 0 is written in VFEDG after reading VFEDG = 1				
1	[Setting condition]				
	When a rising edge is detected on the VFBACKI pin				

Bit 2—Pre-Equalization Flag (PREQF): Detects the occurrence of an IHI signal 2fH modification condition. The generation of a falling/rising edge in the IHI signal during a mask interval is expressed as the occurrence of a 2fH modification condition. For details, see section 13.3.4, IHI Signal 2fH Modification.

Bit 2

PREQF	Description	
0	[Clearing condition]	(Initial value)
	When 0 is written in PREQF after reading PREQF = 1	
1	[Setting condition]	
	When an IHI signal 2fH modification condition is detected	

Bit 1—IHI Signal Level (IHI): Indicates the current level of the IHI signal. Signal source and phase inversion selection for the IHI signal depends on the contents of TCONRI. Read this bit to determine whether the input signal is positive or negative, then maintain the IHI signal at positive phase by modifying TCONRI.

Bit 1

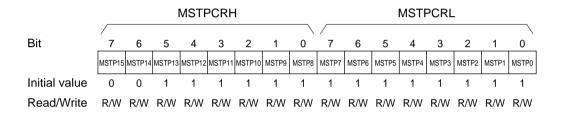
IHI	Description
0	The IHI signal is low
1	The IHI signal is high

Bit 0—IVI Signal Level (IVI): Indicates the current level of the IVI signal. Signal source and phase inversion selection for the IVI signal depends on the contents of TCONRI. Read this bit to determine whether the input signal is positive or negative, then maintain the IVI signal at positive phase by modifying TCONRI.

Bit 0

IVI	Description
0	The IVI signal is low
1	The IVI signal is high

13.2.5 Module Stop Control Register (MSTPCR)



MSTPCR, comprising two 8-bit readable/writable registers, performs module stop mode control. When the MSTP13, MSTP12, and MSTP8 bits are set to 1, the 16-bit free-running timer, 8-bit timer channels 0 and 1, and 8-bit timer channels X and Y and timer connection, respectively, halt and enter module stop mode. See section 25.5., Module Stop Mode, for details.

MSTPCR is initialized to H'3FFF by a reset and in hardware standby mode. It is not initialized in software standby mode.

MSTPCRH Bit 5—Module Stop (MSTP13): Specifies FRT module stop mode.

MSTPCRH Bit 5

MSTP13	Description	
0	FRT module stop mode is cleared	
1	FRT module stop mode is set	(Initial value)



MSTPCRH Bit 4—Module Stop (MSTP12): Specifies 8-bit timer channel 0 and 1 module stop mode.

MSTPCRH Bit 4

MSTP12		
0	8-bit timer channel 0 and 1 module stop mode is cleared	
1	8-bit timer channel 0 and 1 module stop mode is set	(Initial value)

MSTPCRH Bit 0—Module Stop (MSTP8): Specifies 8-bit timer channel X and Y and timer connection module stop mode.

MSTPCRH Bit 0

MSTP8	Description	
0	8-bit timer channel X and Y and timer connection module stop mode is cleared	
1	8-bit timer channel X and Y and timer connection module stop mode is (Initial value)	lue)

13.3 Operation

13.3.1 PWM Decoding (PDC Signal Generation)

The timer connection facility and TMRX can be used to decode a PWM signal in which 0 and 1 are represented by the pulse width. To do this, a signal in which a rising edge is generated at regular intervals must be selected as the IHI signal.

The timer counter (TCNT) in TMRX is set to count the internal clock pulses and to be cleared on the rising edge of the external reset signal (IHI signal). The value to be used as the threshold for deciding the pulse width is written in TCORB. The PWM decoder contains a delay latch which uses the IHI signal as data and compare-match signal B (CMB) as a clock, and the state of the IHI signal (the result of the pulse width decision) at the compare-match signal B timing after TCNT is reset by the rise of the IHI signal is output as the PDC signal. The pulse width setting using TICRR and TICRF of TMRX can be used to determine the pulse width decision threshold. Examples of TCR and TCORB settings are shown in tables 13.3 and 13.4, and the timing chart is shown in figure 13.2.

Table 13.3 Examples of TCR Settings

Bit(s)	Abbreviation	Contents	Description	
7	CMIEB	0	Interrupts due to compare-match and overflow	
6	CMIEA	0	disabled	
5	OVIE	0		
4 and 3	CCLR1, CCLR0	11	TCNT is cleared by the rising edge of the external reset signal (IHI signal)	
2 to 0	CKS2 to CKS0	001	Incremented on internal clock: φ	

Table 13.4 Examples of TCORB (Pulse Width Threshold) Settings

	φ:10 MHz	φ: 12 MHz	φ: 16 MHz	φ: 20 MHz
H'07	0.8 µs	0.67 μs	0.5 μs	0.4 μs
H'0F	1.6 µs	1.33 µs	1 µs	0.8 µs
H'1F	3.2 µs	2.67 µs	2 µs	1.6 µs
H'3F	6.4 µs	5.33 µs	4 μs	3.2 µs
H'7F	12.8 µs	10.67 µs	8 µs	6.4 µs

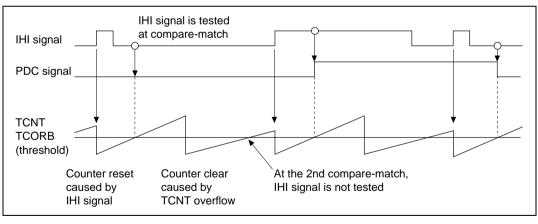


Figure 13.2 Timing Chart for PWM Decoding

13.3.2 Clamp Waveform Generation (CL1/CL2/CL3 Signal Generation)

The timer connection facility and TMRX can be used to generate signals with different duty cycles and rising/falling edges (clamp waveforms) in synchronization with the input signal (IHI signal). Three clamp waveforms can be generated: the CL1, CL2, and CL3 signals. In addition, the CL4 signal can be generated using TMRY.

The CL1 signal rises simultaneously with the rise of the IHI signal, and when the CL1 signal is high, the CL2 signal rises simultaneously with the fall of the IHI signal. The fall of both the CL1 and the CL2 signal can be specified by TCORA.

The rise of the CL3 signal can be specified as simultaneous with the sampling of the fall of the IHI signal using the system clock, and the fall of the CL3 signal can be specified by TCORC. The CL3 signal can also fall when the IHI signal rises.

TCNT in TMRX is set to count internal clock pulses and to be cleared on the rising edge of the external reset signal (IHI signal).

The value to be used as the CL1 signal pulse width is written in TCORA. Write a value of H'02 or more in TCORA when internal clock ϕ is selected as the TMRX counter clock, and a value or H'01 or more when ϕ /2 is selected. When internal clock ϕ is selected, the CL1 signal pulse width is (TCORA set value +3 ±0.5). When the CL2 signal is used, the setting must be made so that this pulse width is greater than the IHI signal pulse width.

The value to be used as the CL3 signal pulse width is written in TCORC. The TICR register in TMRX captures the value of TCNT at the inverse of the external reset signal edge (in this case, the falling edge of the IHI signal). The timing of the fall of the CL3 signal is determined by the sum of the contents of TICR and TCORC. Caution is required if the rising edge of the IHI signal precedes the fall timing set by the contents of TCORC, since the IHI signal will cause the CL3 signal to fall.

Examples of TMRX TCR settings are the same as those in table 13.3. The clamp waveform timing charts are shown in figures 13.3 and 13.4.

Since the rise of the CL1 and CL2 signals is synchronized with the edge of the IHI signal, and their fall is synchronized with the system clock, the pulse width variation is equivalent to the resolution of the system clock.

Both the rise and the fall of the CL3 signal are synchronized with the system clock and the pulse width is fixed, but there is a variation in the phase relationship with the IHI signal equivalent to the resolution of the system clock.

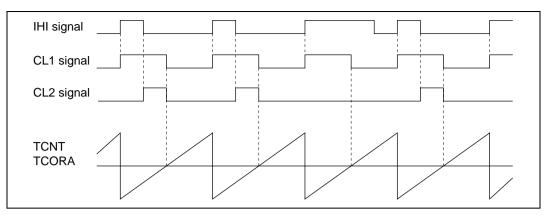


Figure 13.3 Timing Chart for Clamp Waveform Generation (CL1 and CL2 Signals)

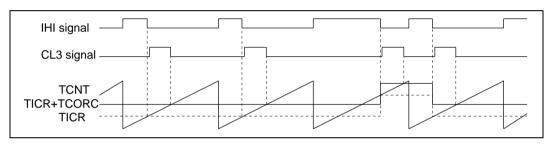


Figure 13.4 Timing Chart for Clamp Waveform Generation (CL3 Signal)

13.3.3 Measurement of 8-Bit Timer Divided Waveform Period

The timer connection facility, TMR1, and the free-running timer (FRT) can be used to measure the period of an IHI signal divided waveform. Since TMR1 can be cleared by a rising edge of inverted IVI signal, the rise and fall of the IHI signal divided waveform can be virtually synchronized with the IVI signal. This enables period measurement to be carried out efficiently.

To measure the period of an IHI signal divided waveform, TCNT in TMR1 is set to count the external clock (IHI signal) pulses and to be cleared on the rising edge of the external reset signal (inverted IVI signal). The value to be used as the division factor is written in TCORA, and the TMO output method is specified by the OS bits in TCSR. Examples of TCR and TCSR settings are shown in table 13.5, and the timing chart for measurement of the IVI signal and IHI signal divided waveform periods is shown in figure 13.5. The period of the IHI signal divided waveform is given by (ICRD(3) – ICRD(2)) × the resolution.

Table 13.5 Examples of TCR and TCSR Settings

Register	Bit(s)	Abbreviation	Contents	Description
TCR in TMR1	7	CMIEB	0	Interrupts due to compare-match
	6	CMIEA	0	and overflow are disabled
	5	OVIE	0	_
	4 and 3	CCLR1, CCLR0	11	TCNT is cleared by the rising edge of the external reset signal (inverted IVI signal)
	2 to 0	CKS2 to CKS0	101	TCNT is incremented on the rising edge of the external clock (IHI signal)
TCSR in TMR1	3 to 0	OS3 to OS0	0011	Not changed by compare-match B; output inverted by compare-match A (toggle output): division by 512
				or
			1001	when TCORB < TCORA, 1 output on compare-match B, and 0 output on compare-match A: division by 256
TCR in FRT	6	IEDGB	0/1	O: FRC value is transferred to ICRB on falling edge of input capture input B (IHI divided signal waveform)
				FRC value is transferred to ICRB on rising edge of input capture input B (IHI divided signal waveform)
	1 and 0	CKS1, CKS0	01	FRC is incremented on internal clock: \$\phi/8\$
TCSR in FRT	0	CCLRA	0	FRC clearing is disabled

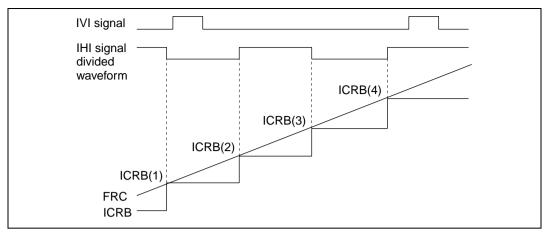


Figure 13.5 Timing Chart for Measurement of IVI Signal and IHI Signal Divided Waveform Periods

13.3.4 IHI Signal and 2fH Modification

By using the timer connection FRT, even if there is a part of the IHI signal with twice the frequency, this can be eliminated. In order for this function to operate properly, the duty cycle of the IHI signal must be approximately 30% or less, or approximately 70% or above.

The 8-bit OCRDM contents or twice the OCRDM contents can be added automatically to the data captured in ICRD in the FRT, and compare-matches generated at these points. The interval between the two compare-matches is called a mask interval. A value equivalent to approximately 1/3 the IHI signal period is written in OCRDM. ICRD is set so that capture is performed on the rise of the IHI signal.

Since the IHI signal supplied to the IHO signal selection circuit is normally set on the rise of the IHI signal and reset on the fall, its waveform is the same as that of the original IHI signal. When 2fH modification is selected, IHI signal edge detection is disabled during mask intervals. Capture is also disabled during these intervals.

Examples of FRT TCR settings are shown in table 13.6, and the 2fH modification timing chart is shown in figure 13.6.

Table 13.6 Examples of TCR, TCSR, TCOR, and OCRDM Settings

Register	Bit(s)	Abbreviation	Contents	Description
TCR in FRT	4	IEDGD	1	FRC value is transferred to ICRD on the rising edge of input capture input D (IHI signal)
	1 and 0	CKS1, CKS0	01	FRC is incremented on internal clock: $\phi/8$
TCSR in FRT	0	CCLRA	0	FRC clearing is disabled
TCOR in FRT	7	ICRDMS	1	ICRD is set to the operating mode in which OCRDM is used
OCRDM in FRT	7 to 0	OCRDM7 to 0	H'01 to H'FF	Specifies the period during which ICRD operation is masked

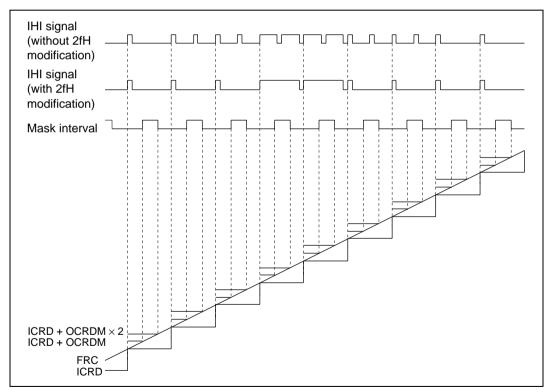


Figure 13.6 2fH Modification Timing Chart

13.3.5 IVI Signal Fall Modification and IHI Synchronization

By using the timer connection TMR1, the fall of the IVI signal can be shifted backward by the specified number of IHI signal waveforms. Also, the fall of the IVI signal can be synchronized with the rise of the IHI signal.

To perform 8-bit timer divided waveform period measurement, TCNT in TMR1 is set to count external clock (IHI signal) pulses, and to be cleared on the rising edge of the external reset signal (inverse of the IVI signal). The number of IHI signal pulses until the fall of the IVI signal is written in TCORB.

Since the IVI signal supplied to the IVO signal selection circuit is normally set on the rise of the IVI signal and reset on the fall, its waveform is the same as that of the original IVI signal. When fall modification is selected, a reset is performed on a TMR1 TCORB compare-match.

The fall of the waveform generated in this way can be synchronized with the rise of the IHI signal, regardless of whether or not fall modification is selected.

Examples of TMR1 TCORB, TCR, and TCSR settings are shown in table 13.7, and the fall modification/IHI synchronization timing chart is shown in figure 13.7.



Table 13.7 Examples of TCORB, TCR, and TCSR Settings

Register	Bit(s)	Abbreviation	Contents	Description
TCR in	7	CMIEB	0	Interrupts due to compare-match and
TMR1	6	CMIEA	0	overflow are disabled
	5	OVIE	0	_
	4 and 3	CCLR1, CCLR0	11	TCNT is cleared by the rising edge of the external reset signal (inverse of the IVI signal)
	2 to 0	CKS2 to CKS0	101	TCNT is incremented on the rising edge of the external clock (IHI signal)
TCSR in TMR1	3 to 0	OS3 to OS0	0011	Not changed by compare-match B; output inverted by compare-match A (toggle output)
				or
			1001	when TCORB \leq TCORA, 1 output on compare-match B, 0 output on comparematch A
TOCRB in TMR1		H'03 (example)	Compare-match on the 4th (example) rise of the IHI signal after the rise of the inverse of the IVI signal	

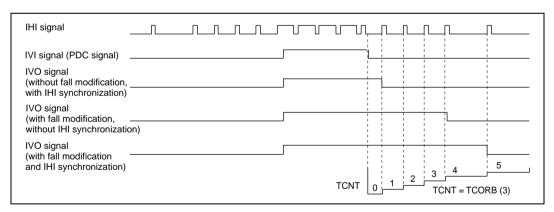


Figure 13.7 Fall Modification/IHI Synchronization Timing Chart

13.3.6 Internal Synchronization Signal Generation (IHG/IVG/CL4 Signal Generation)

By using the timer connection FRT and TMRY, it is possible to automatically generate internal signals (IHG and IVG signals) corresponding to the IHI and IVI signals. As the IHG signal is synchronized with the rise of the IVG signal, the IHG signal period must be made a divisor of the IVG signal period in order to keep it constant. In addition, the CL4 signal can be generated in synchronization with the IHG signal.

The contents of OCRA in the FRT are updated by the automatic addition of the contents of OCRAR or OCRAF, alternately, each time a compare-match occurs. A value corresponding to the 0 interval of the IVG signal is written in OCRAR, and a value corresponding to the 1 interval of the IVG signal is written in OCRAF. The IVG signal is set by a compare-match after an OCRAR addition, and reset by a compare-match after an OCRAF addition.

The IHG signal is the TMRY 8-bit timer output. TMRY is set to count internal clock pulses, and to be cleared on TCORA compare-match, to fix the period and set the timer output. TCORB is set so as to reset the timer output. The IVG signal is connected as the TMRY reset input (TMRI), and the rise of the IVG signal can be treated in the same way as a TCORA compare-match.

The CL4 signal is a waveform that rises within one system clock period after the fall of the IHG signal, and has a 1 interval of 6 system clock periods.

Examples of settings of TCORA, TCORB, TCR, and TCSR in TMRY, and OCRAR, OCRAF, and TCR in the FRT, are shown in table 13.8, and the IHG signal/IVG signal timing chart is shown in figure 13.8.



Table 13.8 Examples of OCRAR, OCRAF, TOCR, TCORA, TCORB, TCR, and TCSR Settings

Register	Bit(s)	Abbreviation	Contents	Description		
TCR in TMRY	7	CMIEB	0	Interrupts due to compare-match and overflow are disabled		
	6	CMIEA	0	_		
	5	OVIE	0	_		
	4 and 3	CCLR1, CCLR0	01	TCNT is cleared by	compare-match A	
	2 to 0	CKS2 to CKS0	001	TCNT is incremented on internal clock: $\phi/4$		
TCSR in TMRY	3 to 0	OS3 to OS0	0110	0 output on compare-match B 1 output on compare-match A		
TOCRA in TMRY			H'3F (example)	IHG signal period = $\phi \times 256$		
TOCRB in TMRY			H'03 (example)	IHG signal 1 interval = $\phi \times 16$		
TCR in FRT	1 and 0	CKS1, CKS0	01	FRC is incremented	l on internal clock: φ/8	
OCRAR in FRT			H'7FEF (example)	IVG signal 0 interval = φ × 262016	IVG signal period = $\phi \times 262144$ (1024 times IHG signal)	
OCRAF in FRT			H'000F (example)	IVG signal 1 interval = $\phi \times 128$		
TOCR in FRT	6	OCRAMS	1	OCRA is set to the operating mode in which OCRAR and OCRAF are used		

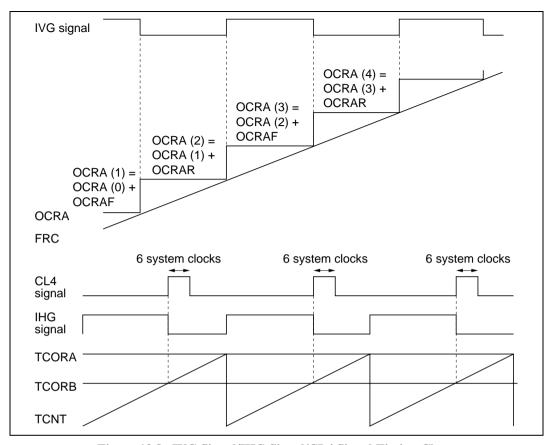


Figure 13.8 IVG Signal/IHG Signal/CL4 Signal Timing Chart

13.3.7 HSYNCO Output

With the HSYNCO output, the meaning of the signal source to be selected and use or non-use of modification varies according to the IHI signal source and the waveform required by external circuitry. The meaning of the HSYNCO output in each mode is shown in table 13.9.

Table 13.9 Meaning of HSYNCO Output in Each Mode

Mode	IHI Signal	IHO Signal	Meaning of IHO Signal
No signal	HFBACKI input	IHI signal (without 2fH modification)	HFBACKI input is output directly
		IHI signal (with 2fH modification)	Meaningless unless there is a double-frequency part in the HFBACKI input
		CL1 signal	HFBACKI input 1 interval is changed before output
		IHG signal	Internal synchronization signal is output
S-on-G mode	CSYNCI input	IHI signal (without 2fH modification)	CSYNCI input (composite synchronization signal) is output directly
		IHI signal (with 2fH modification)	Double-frequency part of CSYNCI input (composite synchronization signal) is eliminated before output
			CSYNCI input (composite synchronization signal) horizontal synchronization signal part is separated before output
		IHG signal	Internal synchronization signal is output
Composite mode	HSYNCI input	IHI signal (without 2fH modification)	HSYNCI input (composite synchronization signal) is output directly
		IHI signal (with 2fH modification)	Double-frequency part of HSYNCI input (composite synchronization signal) is eliminated before output
		CL1 signal	HSYNCI input (composite synchronization signal) horizontal synchronization signal part is separated before output
		IHG signal	Internal synchronization signal is output
Separate mode	HSYNCI input	IHI signal (without 2fH modification)	HSYNCI input (horizontal synchronization signal) is output directly
		IHI signal (with 2fH modification)	Meaningless unless there is a double-frequency part in the HSYNCI input (horizontal synchronization signal)
		CL1 signal	HSYNCI input (horizontal synchronization signal) 1 interval is changed before output
		IHG signal	Internal synchronization signal is output

13.3.8 VSYNCO Output

With the VSYNCO output, the meaning of the signal source to be selected and use or non-use of modification varies according to the IVI signal source and the waveform required by external circuitry. The meaning of the VSYNCO output in each mode is shown in table 13.10.

Table 13.10 Meaning of VSYNCO Output in Each Mode

Mode	IVI Signal	IVO Signal	Meaning of IVO Signal
No signal	VFBACKI input	IVI signal (without fall modification or IHI synchronization)	VFBACKI input is output directly
		IVI signal (without fall modification, with IHI synchronization)	Meaningless unless VFBACKI input is synchronized with HFBACKI input
		IVI signal (with fall modification, without IHI synchronization)	VFBACKI input fall is modified before output
		IVI signal (with fall modification and IHI synchronization)	VFBACKI input fall is modified and signal is synchronized with HFBACKI input before output
		IVG signal	Internal synchronization signal is output
S-on-G mode or composite mode	PDC signal	IVI signal (without fall modification or IHI synchronization)	CSYNCI/HSYNCI input (composite synchronization signal) vertical synchronization signal part is separated before output
		IVI signal (without fall modification, with IHI synchronization)	CSYNCI/HSYNCI input (composite synchronization signal) vertical synchronization signal part is separated, and signal is synchronized with CSYNCI/HSYNCI input before output
		IVI signal (with fall modification, without IHI synchronization)	CSYNCI/HSYNCI input (composite synchronization signal) vertical synchronization signal part is separated, and fall is modified before output
		IVI signal (with fall modification and IHI synchronization)	CSYNCI/HSYNCI input (composite synchronization signal) vertical synchronization signal part is separated, fall is modified, and signal is synchronized with CSYNCI/HSYNCI input before output
		IVG signal	Internal synchronization signal is output



Mode	IVI Signal	IVO Signal	Meaning of IVO Signal
Separate VSYNCI mode input		IVI signal (without fall modification or IHI synchronization)	VSYNCI input (vertical synchronization signal) is output directly
	IVI signal (without fall modification, with IHI synchronization)		Meaningless unless VSYNCI input (vertical synchronization signal) is synchronized with HSYNCI input (horizontal synchronization signal)
		IVI signal (with fall modification, without IHI synchronization)	VSYNCI input (vertical synchronization signal) fall is modified before output
		IVI signal (with fall modification and IHI synchronization)	VSYNCI input (vertical synchronization signal) fall is modified and signal is synchronized with HSYNCI input (horizontal synchronization signal) before output
		IVG signal	Internal synchronization signal is output

13.3.9 CBLANK Output

Using the signals generated/selected with timer connection, it is possible to generate a waveform based on the composite synchronization signal (blanking waveform).

One kind of blanking waveform is generated by combining HFBACKI and VFBACKI inputs, with the phase polarity made positive by means of bits HFINV and VFINV in TCONRI, with the IVO signal.

The composition logic is shown in figure 13.9.

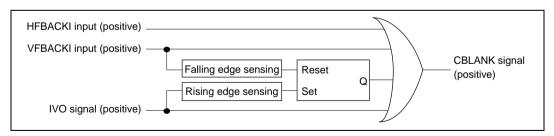


Figure 13.9 CBLANK Output Waveform Generation

Section 14 Watchdog Timer (WDT)

14.1 Overview

This LSI have an on-chip watchdog timer with two channels (WDT0, WDT1) for monitoring system operation. The WDT outputs an overflow signal (RESO) if a system crash prevents the CPU from writing to the timer counter, allowing it to overflow. At the same time, the WDT can also generate an internal reset signal or internal NMI interrupt signal.

When this watchdog function is not needed, the WDT can be used as an interval timer. In interval timer mode, an interval timer interrupt is generated each time the counter overflows.

14.1.1 Features

WDT features are listed below.

- Switchable between watchdog timer mode and interval timer mode
- Internal reset or internal interrupt generated when the timer counter overflows
 - WOVI interrupt generation in interval timer mode
 - Choice of internal reset or NMI interrupt generation in watchdog timer mode
- RESO output in watchdog timer mode
 - In watchdog timer mode, a low-level signal is output from the RESO pin when the counter overflows (when internal reset is selected)
- Choice of 8 (WDT0) or 16 (WDT1) counter input clocks
 - Maximum WDT interval: system clock period × 131072 × 256
 - Subclock can be selected for the WDT1 input counter
 Maximum interval when the subclock is selected: subclock period × 256 × 256

14.1.2 Block Diagram

Figures 14.1 (a) and (b) show block diagrams of WDT0 and WDT1.

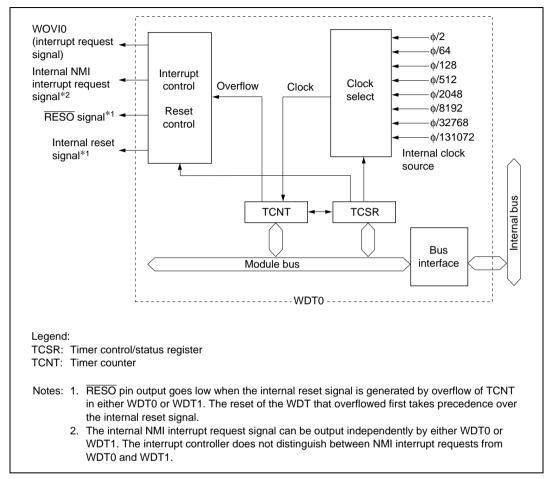


Figure 14.1 (a) Block Diagram of WDT0

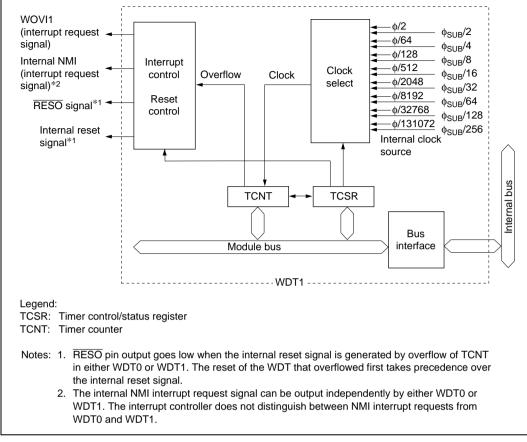


Figure 14.1 (b) Block Diagram of WDT1

14.1.3 Pin Configuration

Table 14.1 describes the WDT input pin.

Table 14.1 WDT Pin

Name	Symbol	I/O	Function
Reset output pin	RESO	Output	Watchdog timer mode counter overflow signal output
External subclock input pin	EXCL	Input	WDT1 prescaler counter input clock

14.1.4 **Register Configuration**

The WDT has four registers, as summarized in table 14.2. These registers control clock selection, WDT mode switching, the reset signal, etc.

Table 14.2 WDT Registers

					Add	ress*1
Channel	Name	Abbreviation	R/W	Initial Value	Write*2	Read
0	Timer control/status register 0	TCSR0	R/(W)*3	H'00	H'FFA8	H'FFA8
	Timer counter 0	TCNT0	R/W	H'00	H'FFA8	H'FFA9
1	Timer control/status register 1	TCSR1	R/(W)*3	H'00	H'FFEA	H'FFEA
	Timer counter 1	TCNT1	R/W	H'00	H'FFEA	H'FFEB
Common	System control register	SYSCR	R/W	H'09	H'FFC4	H'FFC4

Notes: 1. Lower 16 bits of the address.

- 2. For details of write operations, see section 14.2.4, Notes on Register Access.
- 3. Only 0 can be written in bit 7, to clear the flag.

14.2 **Register Descriptions**

14.2.1 **Timer Counter (TCNT)**

Bit	7	6	5	4	3	2	1	0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W							

TCNT is an 8-bit readable/writable* up-counter.

When the TME bit is set to 1 in TCSR, TCNT starts counting pulses generated from the internal clock source selected by bits CKS2 to CKS0 in TCSR. When the count overflows (changes from H'FF to H'00), the OVF flag in TCSR is set to 1. Watchdog timer overflow signal (RESO) output, an internal reset, NMI interrupt, interval timer interrupt (WOVI), etc., can be generated, depending on the mode selected by the WT/IT bit and RST/NMI bit.

TCNT is initialized to H'00 by a reset, in hardware standby mode, or when the TME bit is cleared to 0. It is not initialized in software standby mode.

Note: * TCNT is write-protected by a password to prevent accidental overwriting. For details see section 14.2.4, Notes on Register Access.

14.2.2 Timer Control/Status Register (TCSR)

TCSR0

Bit	7	6	5	4	3	2	1	0
	OVF	WT/ IT	TME	RSTS	RST/NMI	CKS2	CKS1	CKS0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/(W)*	R/W	R/W	R/W	R/W	R/W	R/W	R/W

Note: * Only 0 can be written, to clear the flag.

TCSR1

Bit	7	6	5	4	3	2	1	0
	OVF	WT/ IT	TME	PSS	RST/NMI	CKS2	CKS1	CKS0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/(W)*	R/W	R/W	R/W	R/W	R/W	R/W	R/W

Note: * Only 0 can be written, to clear the flag.

TCSR is an 8-bit readable/writable* register. Its functions include selecting the clock source to be input to TCNT, and the timer mode.

TCR is initialized to H'00 by a reset and in hardware standby mode. It is not initialized in software standby mode.

Note: * TCSR is write-protected by a password to prevent accidental overwriting. For details see section 14.2.4, Notes on Register Access.

Bit 7—Overflow Flag (OVF): A status flag that indicates that TCNT has overflowed from H'FF to H'00.

Bit 7

OVF	Description					
0	[Clearing condi	tions]				
	 Write 0 in th 	ie TME bit	(Initial value)			
	 Read TCSF 	R when OVF = 1*, then write 0 in OVF				
1	[Setting condition]					
	(When internal	rerflows (changes from H'FF to H'00) reset request generation is selected in watchoutically by the internal reset.)	dog timer mode, OVF is			
Note:	 When OVF is po at least twice. 	lled and the interval timer interrupt is disabled	d, OVF = 1 must be read			

Bit 6—Timer Mode Select (WT/IT): Selects whether the WDT is used as a watchdog timer or interval timer. If used as an interval timer, the WDT generates an interval timer interrupt request (WOVI) when TCNT overflows. If used as a watchdog timer, the WDT generates a reset or NMI interrupt when TCNT overflows. When internal reset is selected in watchdog timer mode, a lowlevel signal is output from the \overline{RESO} pin.

Bit 6

WT/IT	Description
0	Interval timer mode: Sends the CPU an interval timer interrupt request (WOVI) when TCNT overflows (Initial value)
1	Watchdog timer mode: Generates a reset or NMI interrupt when TCNT overflows
	At the same time, a low-level signal is output from the $\overline{\text{RESO}}$ pin (when internal reset is selected)

Bit 5—Timer Enable (TME): Selects whether TCNT runs or is halted.

Bit 5

TME	Description	
0	TCNT is initialized to H'00 and halted	(Initial value)
1	TCNT counts	



TCSR0 Bit 4—Reset Select (RSTS): Reserved. This bit should not be set to 1.

TCSR1 Bit 4—Prescaler Select (PSS): Selects the input clock source for TCNT in WDT1. For details, see the description of the CKS2 to CKS0 bits below.

WDT1 TCSR Bit 4

PSS	Description	
0	TCNT counts φ-based prescaler (PSM) divided clock pulses	(Initial value)
1	TCNT counts φSUB-based prescaler (PSS) divided clock pulses	

Bit 3—Reset or NMI (RST/NMI): Specifies whether an internal reset or NMI interrupt is requested on TCNT overflow in watchdog timer mode.

Bit 3

RST/NMI	Description	
0	An NMI interrupt is requested	(Initial value)
1	An internal reset is requested	

Bits 2 to 0—Clock Select 2 to 0 (CKS2 to CKS0): These bits select an internal clock source, obtained by dividing the system clock (ϕ) , or subclock (ϕSUB) for input to TCNT.

• WDT0 input clock selection

Bit 2	Bit 1	Bit 0	Description			
CKS2	CKS1	CKS0	Clock	Overflow Period* (when φ = 20 MHz)		
0	0	0	φ/2 (Initial value)	25.6 µs		
		1	ф/64	819.2 µs		
	1	0	ф/128	1.6 ms		
		1	φ/512	6.6 ms		
1	0	0	ф/2048	26.2 ms		
		1	ф/8192	104.9 ms		
	1	0	ф/32768	419.4 ms		
		1	φ/131072	1.68 s		

Note: * The overflow period is the time from when TCNT starts counting up from H'00 until overflow occurs.

WDT1 input clock selection

Bit 4	Bit 2	Bit 1	Bit 0	Description		
PSS	CKS2	CKS1	CKS0	Clock	Overflow Period* (when ϕ = 20 MHz and ϕ_{SUB} = 32.768 kHz)	
0	0	0	0	φ/2 (Initial value)	25.6 μs	
			1	φ/64	819.2 µs	
		1	0	φ/128	1.6 ms	
			1	φ/512	6.6 ms	
	1	0	0	φ/2048	26.2 ms	
			1	φ/8192	104.9 ms	
		1	0	ф/32768	419.4 ms	
			1	φ/131072	1.68 s	
1	0	0	0	φSUB/2	15.6 ms	
			1	φSUB/4	31.3 ms	
		1	0	φSUB/8	62.5 ms	
			1	φSUB/16	125 ms	
	1	0	0	φSUB/32	250 ms	
			1	φSUB/64	500 ms	
		1	0	φSUB/128	1 s	
			1	φSUB/256	2 s	

The overflow period is the time from when TCNT starts counting up from H'00 until Note: overflow occurs.

System Control Register (SYSCR) 14.2.3

Bit	7	6	5	4	3	2	1	0
	CS2E	IOSE	INTM1	INTM0	XRST	NMIEG	HIE	RAME
Initial value	0	0	0	0	1	0	0	1
Read/Write	R/W	R/W	R	R/W	R	R/W	R/W	R/W

Only bit 3 is described here. For details on functions not related to the watchdog timer, see sections 3.2.2 and 5.2.1, System Control Register (SYSCR), and the descriptions of the relevant modules.

Bit 3—External Reset (XRST): Indicates the reset source. When the watchdog timer is used, a reset can be generated by watchdog timer overflow in addition to external reset input. XRST is a read-only bit. It is set to 1 by an external reset, and when the RST/NMI bit is 1, is cleared to 0 by an internal reset due to watchdog timer overflow.

XRST	Description	
0	Reset is generated by watchdog timer overflow	
1	Reset is generated by external reset input	(Initial value)

14.2.4 Notes on Register Access

The watchdog timer's TCNT and TCSR registers differ from other registers in being more difficult to write to. The procedures for writing to and reading these registers are given below.

Writing to TCNT and TCSR (Example of WDT0)

These registers must be written to by a word transfer instruction. They cannot be written to with byte transfer instructions.

Figure 14.2 shows the format of data written to TCNT and TCSR. TCNT and TCSR both have the same write address. For a write to TCNT, the upper byte of the written word must contain H'5A and the lower byte must contain the write data. For a write to TCSR, the upper byte of the written word must contain H'A5 and the lower byte must contain the write data. This transfers the write data from the lower byte to TCNT or TCSR.

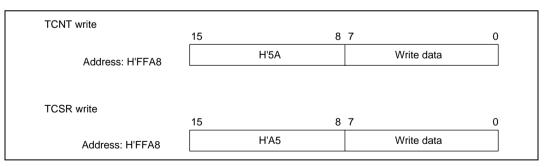


Figure 14.2 Format of Data Written to TCNT and TCSR (Example of WDT0)

Reading TCNT and TCSR (Example of WDT0)

These registers are read in the same way as other registers. The read addresses are H'FFA8 for TCSR, and H'FFA9 for TCNT.

14.3 Operation

14.3.1 Watchdog Timer Operation

To use the WDT as a watchdog timer, set the WT/IT and TME bits in TCSR to 1. Software must prevent TCNT overflows by rewriting the TCNT value (normally by writing H'00) before overflow occurs. This ensures that TCNT does not overflow while the system is operating normally. If TCNT overflows without being rewritten because of a system crash or other error, an internal reset or NMI interrupt request is generated.

When the RST/ $\overline{\text{NMI}}$ bit is set to 1, the chip is reset for 518 system clock periods (518 ϕ) by a counter overflow, and at the same time a low-level signal is output from the $\overline{\text{RESO}}$ pin for 132 states. This is illustrated in figure 14.3. The system can be reset using this $\overline{\text{RESO}}$ signal.

When the RST/ $\overline{\text{NMI}}$ bit cleared to 0, an NMI interrupt request is generated by a counter overflow. In this case, the $\overline{\text{RESO}}$ output signal remains high.

An internal reset request from the watchdog timer and reset input from the RES pin are handled via the same vector. The reset source can be identified from the value of the XRST bit in SYSCR.

If a reset caused by an input signal from the \overline{RES} pin and a reset caused by WDT overflow occur simultaneously, the \overline{RES} pin reset has priority, and the XRST bit in SYSCR is set to 1.

An NMI interrupt request from the watchdog timer and an interrupt request from the NMI pin are handled via the same vector. Simultaneous handling of a watchdog timer NMI interrupt request and an NMI pin interrupt request must therefore be avoided.



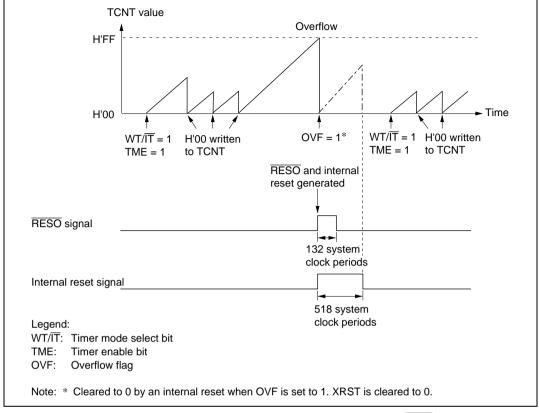


Figure 14.3 Operation in Watchdog Timer Mode (RST/ $\overline{NMI} = 1$)

14.3.2 Interval Timer Operation

To use the WDT as an interval timer, clear the WT/ $\overline{\text{IT}}$ bit in TCSR to 0 and set the TME bit to 1. An interval timer interrupt (WOVI) is generated each time TCNT overflows, provided that the WDT is operating as an interval timer, as shown in figure 14.4. This function can be used to generate interrupt requests at regular intervals.

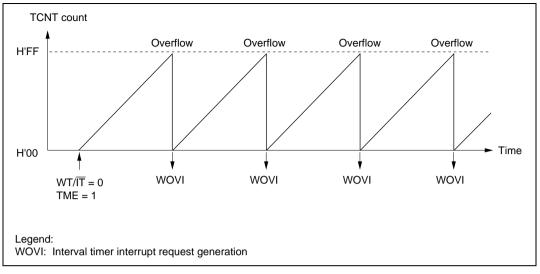


Figure 14.4 Operation in Interval Timer Mode

14.3.3 Timing of Setting of Overflow Flag (OVF)

The OVF flag is set to 1 if TCNT overflows during interval timer operation. At the same time, an interval timer interrupt (WOVI) is requested. This timing is shown in figure 14.5.

If NMI request generation is selected in watchdog timer mode, when TCNT overflows the OVF bit in TCSR is set to 1 and at the same time an NMI interrupt is requested.

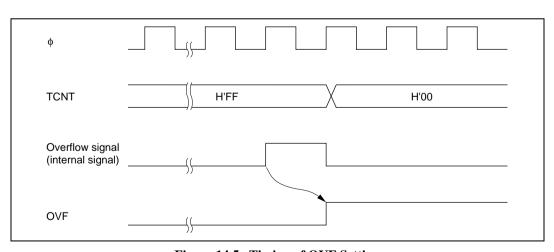


Figure 14.5 Timing of OVF Setting

RESO Signal Output Timing 14.3.4

When TCNT overflows in watchdog timer mode, the OVF bit is set to 1 in TCSR. If the RST/NMI bit is 1 at this time, an internal reset signal is generated for the entire chip, and at the same time a low-level signal is output from the \overline{RESO} pin. The timing is shown in figure 14.6.

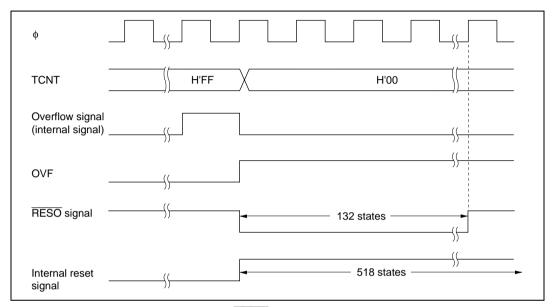


Figure 14.6 RESO Signal Output Timing

14.4 **Interrupts**

During interval timer mode operation, an overflow generates an interval timer interrupt (WOVI). The interval timer interrupt is requested whenever the OVF flag is set to 1 in TCSR. OVF must be cleared to 0 in the interrupt handling routine. When NMI interrupt request generation is selected in watchdog timer mode, an overflow generates an NMI interrupt request.

14.5 Usage Notes

14.5.1 Contention between Timer Counter (TCNT) Write and Increment

If a timer counter clock pulse is generated during the T_2 state of a TCNT write cycle, the write takes priority and the timer counter is not incremented. Figure 14.7 shows this operation.

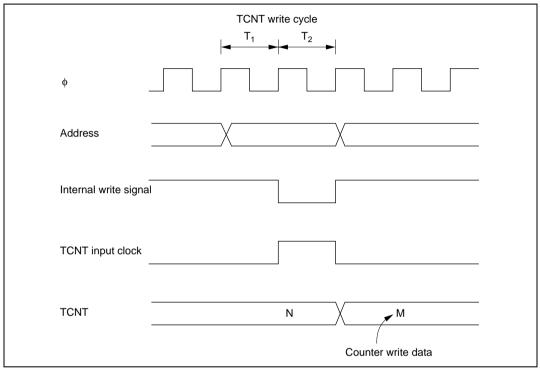


Figure 14.7 Contention between TCNT Write and Increment

14.5.2 Changing Value of CKS2 to CKS0

If bits CKS2 to CKS0 in TCSR are written to while the WDT is operating, errors could occur in the incrementation. Software must stop the watchdog timer (by clearing the TME bit to 0) before changing the value of bits CKS2 to CKS0.

14.5.3 Switching between Watchdog Timer Mode and Interval Timer Mode

If the mode is switched from watchdog timer to interval timer, or vice versa, while the WDT is operating, errors could occur in the incrementation. Software must stop the watchdog timer (by clearing the TME bit to 0) before switching the mode.

14.5.4 System Reset by $\overline{\text{RESO}}$ Signal

If the \overline{RESO} output signal is input to the chip's \overline{RES} pin, the chip will not be initialized correctly. Ensure that the \overline{RESO} signal is not logically input to the chip's \overline{RES} pin. When resetting the entire system with the \overline{RESO} signal, use a circuit such as that shown in figure 14.8.

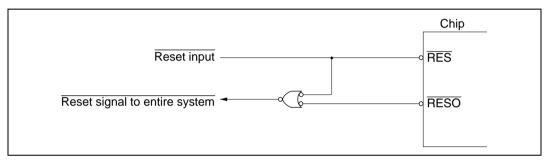


Figure 14.8 Sample Circuit for System Reset by RESO Signal

14.5.5 Counter Value in Transitions between High-Speed Mode, Subactive Mode, and Watch Mode

If the mode is switched between high-speed mode and subactive mode or between high-speed mode and watch mode when WDT1 is used as a realtime clock counter, an error will occur in the counter value when the internal clock is switched.

When the mode is switched from high-speed mode to subactive mode or watch mode, the increment timing is delayed by approximately 2 or 3 clock cycles when the WDT1 control clock is switched from the main clock to the subclock.

Also, since the main clock oscillator is halted during subclock operation, when the mode is switched from watch mode or subactive mode to high-speed mode, the clock is not supplied until internal oscillation stabilizes. As a result, after oscillation is started, counter incrementing is halted during the oscillation stabilization time set by bits STS2 to STS0 in SBYCR, and there is a corresponding discrepancy in the counter value.

Caution is therefore required when using WDT1 as the realtime clock counter.

No error occurs in the counter value while WDT1 is operating in the same mode.

14.5.6 OVF Flag Clear Condition

To clear OVF flag in WOVI handling routine, read TCSR when OVF = 1, then write with 0 to OVF, as stated above. When WOVI is masked and OVF flag is poling, if contention between OVF flag set and TCSR read is occurred, OVF = 1 is read but OVF can not be cleared by writing with 0 to OVF.

In this case, reading TCSR when OVF = 1 two times meet the requirements of OVF clear condition. Please read TCSR when OVF = 1 two times before writing with 0 to OVF.

```
LOOP BTST.B #7, @TCSR ; OVF flag read

BEQ LOOP ; if OVF=1, exit from loop

MOV.B @TCSR, ROL ; OVF=1 read again

MOV.W #H'A521, RO ; OVF flag clear

MOV.W RO.@TCSR ; :
```

Section 15 Serial Communication Interface (SCI, IrDA)

15.1 Overview

This LSI are equipped with a 3-channel serial communication interface (SCI). The SCI can handle both asynchronous and clocked synchronous serial communication. A function is also provided for serial communication between processors (multiprocessor communication function).

One of the three SCI channels can transmit and receive IrDA communication waveforms based on IrDA specification version 1.0.

15.1.1 Features

SCI features are listed below.

Choice of asynchronous or synchronous serial communication mode

Asynchronous mode:

- Serial data communication is executed using an asynchronous system in which synchronization is achieved character by character
 Serial data communication can be carried out with standard asynchronous communication chips such as a Universal Asynchronous Receiver/Transmitter (UART) or Asynchronous Communication Interface Adapter (ACIA)
- A multiprocessor communication function is provided that enables serial data communication with a number of processors
- Choice of 12 serial data transfer formats

Data length: 7 or 8 bits Stop bit length: 1 or 2 bits

Parity: Even, odd, or none

Multiprocessor bit: 1 or 0

— Receive error detection: Parity, overrun, and framing errors

— Break detection: Break can be detected by reading the RxD pin level

directly in case of a framing error

Synchronous mode:

- Serial data communication is synchronized with a clock
 - Serial data communication can be carried out with other chips that have a synchronous communication function
- One serial data transfer format

Data length: 8 bits

- Receive error detection: Overrun errors detected
- Full-duplex communication capability
 - The transmitter and receiver are mutually independent, enabling transmission and reception to be executed simultaneously
 - Double-buffering is used in both the transmitter and the receiver, enabling continuous transmission and continuous reception of serial data
- LSB-first or MSB-first transfer can be selected
 - This selection can be made regardless of the communication mode (with the exception of 7-bit data transfer in asynchronous mode)*

Note: * LSB-first transfer is used in the examples in this section.

- Built-in baud rate generator allows any bit rate to be selected
- Choice of serial clock source: internal clock from baud rate generator or external clock from SCK pin
- · Capability of transmit and receive clock output
 - The P86/SCK1 and P42/SCK2 pins are CMOS type outputs
 - The P52/SCK0 pin is an NMOS push-pull type output in the H8S/2148 Group and H8S/2147N, and is a CMOS output in the H8S/2144 Group (When the P52/SCK0 pin is used as an output in the H8S/2148 Group and H8S/2147N, external pull-up resistor must be connected in order to output high level)
- The transmit-data-empty interrupt and receive-data-full interrupt can activate the data transfer controller (DTC) to execute data transfer
- Four interrupt sources
 - Four interrupt sources (transmit-data-empty, transmit-end, receive-data-full, and receive error) that can issue requests independently
 - The transmit-data-empty interrupt and receive-data-full interrupt can activate the data transfer controller (DTC) to execute data transfer



15.1.2 Block Diagram

Figure 15.1 shows a block diagram of the SCI.

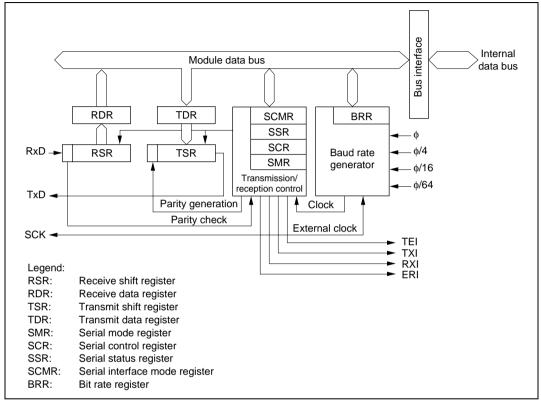


Figure 15.1 Block Diagram of SCI

15.1.3 **Pin Configuration**

Table 15.1 shows the serial pins used by the SCI.

Table 15.1 SCI Pins

Channel	Pin Name	Symbol*	I/O	Function
0	Serial clock pin 0	SCK0	I/O	SCI0 clock input/output
	Receive data pin 0	RxD0	Input	SCI0 receive data input
	Transmit data pin 0	TxD0	Output	SCI0 transmit data output
1	Serial clock pin 1	SCK1	I/O	SCI1 clock input/output
	Receive data pin 1	RxD1	Input	SCI1 receive data input
	Transmit data pin 1	TxD1	Output	SCI1 transmit data output
2	Serial clock pin 2	SCK2	I/O	SCI2 clock input/output
	Receive data pin 2	RxD2/IrRxD	Input	SCI2 receive data input (normal/IrDA)
	Transmit data pin 2	TxD2/IrTxD	Output	SCI2 transmit data output (normal/IrDA)

The abbreviations SCK, RxD, and TxD are used in the text, omitting the channel Note: number.

15.1.4 **Register Configuration**

The SCI has the internal registers shown in table 15.2. These registers are used to specify asynchronous mode or synchronous mode, the data format, and the bit rate, and to control the transmitter/receiver.



Table 15.2 SCI Registers

Channel	Name	Abbreviation	R/W	Initial Value	Address*1
0	Serial mode register 0	SMR0	R/W	H'00	H'FFD8*3
	Bit rate register 0	BRR0	R/W	H'FF	H'FFD9*3
	Serial control register 0	SCR0	R/W	H'00	H'FFDA
	Transmit data register 0	TDR0	R/W	H'FF	H'FFDB
	Serial status register 0	SSR0	R/(W)*2	H'84	H'FFDC
	Receive data register 0	RDR0	R	H'00	H'FFDD
	Serial interface mode register 0	SCMR0	R/W	H'F2	H'FFDE*3
1	Serial mode register 1	SMR1	R/W	H'00	H'FF88*3
	Bit rate register 1	BRR1	R/W	H'FF	H'FF89*3
	Serial control register 1	SCR1	R/W	H'00	H'FF8A
	Transmit data register 1	TDR1	R/W	H'FF	H'FF8B
	Serial status register 1	SSR1	R/(W)*2	H'84	H'FF8C
	Receive data register 1	RDR1	R	H'00	H'FF8D
	Serial interface mode register 1	SCMR1	R/W	H'F2	H'FF8E*3
2	Serial mode register 2	SMR2	R/W	H'00	H'FFA0*3
	Bit rate register 2	BRR2	R/W	H'FF	H'FFA1*3
	Serial control register 2	SCR2	R/W	H'00	H'FFA2
	Transmit data register 2	TDR2	R/W	H'FF	H'FFA3
	Serial status register 2	SSR2	R/(W)*2	H'84	H'FFA4
	Receive data register 2	RDR2	R	H'00	H'FFA5
	Serial interface mode register 2	SCMR2	R/W	H'F2	H'FFA6*3
	Keyboard comparator control register	KBCOMP	R/W	H'00	H'FEE4
Common	Module stop control register	MSTPCRH	R/W	H'3F	H'FF86
		MSTPCRL	R/W	H'FF	H'FF87

Notes: 1. Lower 16 bits of the address.

- 2. Only 0 can be written, to clear flags.
- 3. Some serial communication interface registers are assigned to the same addresses as other registers. In this case, register selection is performed by the IICE bit in the serial timer control register (STCR).

15.2 Register Descriptions

15.2.1 Receive Shift Register (RSR)

Bit	7	6	5	4	3	2	1	0
Read/Write	_	_	_	_	_	_	_	_

RSR is a register used to receive serial data.

The SCI sets serial data input from the RxD pin in RSR in the order received, starting with the LSB (bit 0), and converts it to parallel data. When one byte of data has been received, it is transferred to RDR automatically.

RSR cannot be directly read or written to by the CPU.

15.2.2 Receive Data Register (RDR)

Bit	7	6	5	4	3	2	1	0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R	R	R	R	R	R	R	R

RDR is a register that stores received serial data.

When the SCI has received one byte of serial data, it transfers the received serial data from RSR to RDR where it is stored, and completes the receive operation. After this, RSR is receive-enabled.

Since RSR and RDR function as a double buffer in this way, continuous receive operations can be performed.

RDR is a read-only register, and cannot be written to by the CPU.

RDR is initialized to H'00 by a reset, and in standby mode, watch mode, subactive mode, subsleep mode, and module stop mode.

15.2.3 Transmit Shift Register (TSR)



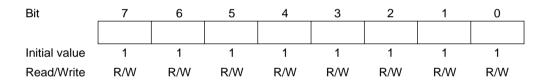
TSR is a register used to transmit serial data.

To perform serial data transmission, the SCI first transfers transmit data from TDR to TSR, then sends the data to the TxD pin starting with the LSB (bit 0).

When transmission of one byte is completed, the next transmit data is transferred from TDR to TSR, and transmission started, automatically. However, data transfer from TDR to TSR is not performed if the TDRE bit in SSR is set to 1.

TSR cannot be directly read or written to by the CPU.

15.2.4 Transmit Data Register (TDR)



TDR is an 8-bit register that stores data for serial transmission.

When the SCI detects that TSR is empty, it transfers the transmit data written in TDR to TSR and starts serial transmission. Continuous serial transmission can be carried out by writing the next transmit data to TDR during serial transmission of the data in TSR.

TDR can be read or written to by the CPU at all times.

TDR is initialized to H'FF by a reset, and in standby mode, watch mode, subactive mode, subsleep mode, and module stop mode.

15.2.5 Serial Mode Register (SMR)

Bit	7	6	5	4	3	2	1	0
	C/A	CHR	PE	O/E	STOP	MP	CKS1	CKS0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

SMR is an 8-bit register used to set the SCI's serial transfer format and select the baud rate generator clock source.

SMR can be read or written to by the CPU at all times.

SMR is initialized to H'00 by a reset, and in standby mode, watch mode, subactive mode, subsleep mode, and module stop mode.

Bit 7—Communication Mode ($\overline{C/A}$): Selects asynchronous mode or synchronous mode as the SCI operating mode.

Bit 7

C/Ā	Description	
0	Asynchronous mode	(Initial value)
1	Synchronous mode	

Bit 6—Character Length (CHR): Selects 7 or 8 bits as the data length in asynchronous mode. In synchronous mode, a fixed data length of 8 bits is used regardless of the CHR setting.

Bit 6

CHR	Description	
0	8-bit data	(Initial value)
1	7-bit data*	

When 7-bit data is selected, the MSB (bit 7) of TDR is not transmitted, and LSB-Note: first/MSB-first selection is not available.



Bit 5—Parity Enable (PE): In asynchronous mode, selects whether or not parity bit addition is performed in transmission, and parity bit checking in reception. In synchronous mode, or when a multiprocessor format is used, parity bit addition and checking is not performed, regardless of the PE bit setting.

Bit 5

PE		Description	
0		Parity bit addition and checking disabled	(Initial value)
1		Parity bit addition and checking enabled*	
Note:	*	When the PE bit is set to 1, the parity (even or odd) specified by transmit data before transmission. In reception, the parity bit is (even or odd) specified by the O/\overline{E} bit.	,

Bit 4—Parity Mode (O/\overline{E}) : Selects either even or odd parity for use in parity addition and checking.

The O/\overline{E} bit setting is only valid when the PE bit is set to 1, enabling parity bit addition and checking, in asynchronous mode. The O/\overline{E} bit setting is invalid in synchronous mode, when parity bit addition and checking is disabled in asynchronous mode, and when a multiprocessor format is used.

Bit 4

O/Ē	Description	
0	Even parity*1	(Initial value)
1	Odd parity*2	

- Notes: 1. When even parity is set, parity bit addition is performed in transmission so that the total number of 1 bits in the transmit character plus the parity bit is even.

 In reception, a check is performed to see if the total number of 1 bits in the receive character plus the parity bit is even.
 - 2. When odd parity is set, parity bit addition is performed in transmission so that the total number of 1 bits in the transmit character plus the parity bit is odd.
 In reception, a check is performed to see if the total number of 1 bits in the receive character plus the parity bit is odd.

Bit 3—Stop Bit Length (STOP): Selects 1 or 2 bits as the stop bit length in asynchronous mode. The STOP bit setting is only valid in asynchronous mode. If synchronous mode is set the STOP bit setting is invalid since stop bits are not added.

Bit 3

STOP	Description	
0	1 stop bit*1	(Initial value)
1	2 stop bits*2	

- Notes: 1. In transmission, a single 1 bit (stop bit) is added to the end of a transmit character before it is sent.
 - 2. In transmission, two 1 bits (stop bits) are added to the end of a transmit character before it is sent.

In reception, only the first stop bit is checked, regardless of the STOP bit setting. If the second stop bit is 1, it is treated as a stop bit; if it is 0, it is treated as the start bit of the next transmit character.

Bit 2—Multiprocessor Mode (MP): Selects multiprocessor format. When multiprocessor format is selected, the PE bit and O/E bit parity settings are invalid. The MP bit setting is only valid in asynchronous mode; it is invalid in synchronous mode.

For details of the multiprocessor communication function, see section 15.3.3, Multiprocessor Communication Function

Bit 2

MP	Description	
0	Multiprocessor function disabled	(Initial value)
1	Multiprocessor format selected	

Bits 1 and 0—Clock Select 1 and 0 (CKS1, CKS0): These bits select the clock source for the band rate generator. The clock source can be selected from ϕ , $\phi/4$, $\phi/16$, and $\phi/64$, according to the setting of bits CKS1 and CKS0.

For the relation between the clock source, the bit rate register setting, and the baud rate, see section 15.2.8, Bit Rate Register (BRR).

Bit 1	Bit 0		
CKS1	CKS0	Description	
0	0	φ clock	(Initial value)
	1	φ/4 clock	
1	0	φ/16 clock	
	1	φ/64 clock	

15.2.6 Serial Control Register (SCR)

Bit	7	6	5	4	3	2	1	0
	TIE	RIE	TE	RE	MPIE	TEIE	CKE1	CKE0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

SCR is a register that performs enabling or disabling of SCI transfer operations, serial clock output in asynchronous mode, and interrupt requests, and selection of the serial clock source.

SCR can be read or written to by the CPU at all times.

SCR is initialized to H'00 by a reset, and in standby mode, watch mode, subactive mode, subsleep mode, and module stop mode.

Bit 7—Transmit Interrupt Enable (TIE): Enables or disables transmit-data-empty interrupt (TXI) request generation when serial transmit data is transferred from TDR to TSR and the TDRE flag in SSR is set to 1.

Bit 7

TIE	Description	
0	Transmit-data-empty interrupt (TXI) request disabled*	(Initial value)
1	Transmit-data-empty interrupt (TXI) request enabled	
Note:	* TXI interrupt request cancellation can be performed by readin	g 1 from the TDRF flag

* TXI interrupt request cancellation can be performed by reading 1 from the TDRE flag, then clearing it to 0, or clearing the TIE bit to 0.

Bit 6—Receive Interrupt Enable (RIE): Enables or disables receive-data-full interrupt (RXI) request and receive-error interrupt (ERI) request generation when serial receive data is transferred from RSR to RDR and the RDRF flag in SSR is set to 1.

Bit 6

RIE		Description					
0		Receive-data-full interrupt (RXI) request and receive-error interrupt (ERI) request disabled* (Initial value)					
1		Receive-data-full interrupt (RXI) request and receive-error interrupt (ERI) request enabled					
Note:	*	RXI and ERI interrupt request cancellation can be performed by reading 1 from the RDRF, FER, PER, or ORER flag, then clearing the flag to 0, or clearing the RIE bit to 0					

Bit 5—Transmit Enable (TE): Enables or disables the start of serial transmission by the SCI.

Bit 5

TE	Description	
0	Transmission disabled*1	(Initial value)
1	Transmission enabled*2	

Notes: 1. The TDRE flag in SSR is fixed at 1.

2. In this state, serial transmission is started when transmit data is written to TDR and the TDRE flag in SSR is cleared to 0.

SMR setting must be performed to decide the transmission format before setting the TE bit to 1.

Bit 4—Receive Enable (RE): Enables or disables the start of serial reception by the SCI.

Bit 4

RE	Description	
0	Reception disabled*1	(Initial value)
1	Reception enabled*2	

Notes: 1. Clearing the RE bit to 0 does not affect the RDRF, FER, PER, and ORER flags, which retain their states.

2. Serial reception is started in this state when a start bit is detected in asynchronous mode or serial clock input is detected in synchronous mode.

SMR setting must be performed to decide the reception format before setting the RE bit to 1.



Bit 3—Multiprocessor Interrupt Enable (MPIE): Enables or disables multiprocessor interrupts. The MPIE bit setting is only valid in asynchronous mode when receiving with the MP bit in SMR set to 1.

The MPIE bit setting is invalid in synchronous mode or when the MP bit is cleared to 0.

Bit 3

MPIE		 Description				
0		Multiprocessor interrupts disabled (normal reception performed) (Initial value)				
		[Clearing conditions]				
		When the MPIE bit is cleared to 0				
		 When data with MPB = 1 is received 				
1		Multiprocessor interrupts enabled*				
		Receive interrupt (RXI) requests, receive-error interrupt (ERI) requests, and setting of the RDRF, FER, and ORER flags in SSR are disabled until data with the multiprocessor bit set to 1 is received.				
Note:	* When receive data including MPB = 0 is received, receive data transfer from RSR to RDR, receive error detection, and setting of the RDRF, FER, and ORER flags in SSR is not performed. When receive data with MPB = 1 is received, the MPB bit in SSR is set to 1, the MPIE bit is cleared to 0 automatically, and generation of RXI and ERI interrupts (when the TIE and RIE bits in SCR are set to 1) and FER and ORER flag setting is enabled.					

Bit 2—Transmit End Interrupt Enable (TEIE): Enables or disables transmit-end interrupt (TEI) request generation if there is no valid transmit data in TDR when the MSB is transmitted.

Bit 2

TEIE	Description	
0	Transmit-end interrupt (TEI) request disabled*	(Initial value)
1	Transmit-end interrupt (TEI) request enabled*	
Note: *	TEI cancellation can be performed by reading 1 from the TDRE flag clearing it to 0 and clearing the TEND flag to 0, or clearing the TEIE	,

Bits 1 and 0—Clock Enable 1 and 0 (CKE1, CKE0): These bits are used to select the SCI clock source and enable or disable clock output from the SCK pin. The combination of the CKE1 and CKE0 bits determines whether the SCK pin functions as an I/O port, the serial clock output pin, or the serial clock input pin.

The setting of the CKE0 bit, however, is only valid for internal clock operation (CKE1 = 0) in asynchronous mode. The CKE0 bit setting is invalid in synchronous mode, and in the case of

external clock operation (CKE1 = 1). The setting of bits CKE1 and CKE0 must be carried out before the SCI's operating mode is determined using SMR.

For details of clock source selection, see table 15.9 in section 15.3, Operation.

Bit 1	Bit 0		
CKE1	CKE0	Description	
0	0	Asynchronous mode	Internal clock/SCK pin functions as I/O port*1
		Synchronous mode	Internal clock/SCK pin functions as serial clock output*1
	1	Asynchronous mode	Internal clock/SCK pin functions as clock output*2
		Synchronous mode	Internal clock/SCK pin functions as serial clock output
1	0	Asynchronous mode	External clock/SCK pin functions as clock input*3
		Synchronous mode	External clock/SCK pin functions as serial clock input
	1	Asynchronous mode	External clock/SCK pin functions as clock input*3
		Synchronous mode	External clock/SCK pin functions as serial clock input

Notes: 1. Initial value

- 2. Outputs a clock of the same frequency as the bit rate.
- 3. Inputs a clock with a frequency 16 times the bit rate.

15.2.7 Serial Status Register (SSR)

Bit	7	6	5	4	3	2	1	0
	TDRE	RDRF	ORER	FER	PER	TEND	MPB	MPBT
Initial value	1	0	0	0	0	1	0	0
Read/Write	R/(W)*	R/(W)*	R/(W)*	R/(W)*	R/(W)*	R	R	R/W

Note: Only 0 can be written, to clear the flag.

SSR is an 8-bit register containing status flags that indicate the operating status of the SCI, and multiprocessor bits.

SSR can be read or written to by the CPU at all times. However, 1 cannot be written to flags TDRE, RDRF, ORER, PER, and FER. Also note that in order to clear these flags they must be read as 1 beforehand. The TEND flag and MPB flag are read-only flags and cannot be modified. SSR is initialized to H'84 by a reset, and in standby mode, watch mode, subactive mode, subsleep mode, and module stop mode.

Bit 7—Transmit Data Register Empty (TDRE): Indicates that data has been transferred from TDR to TSR and the next serial data can be written to TDR.

	_	
\mathbf{r}	:4	7

TDRE	 Description
0	[Clearing conditions]
	 When 0 is written in TDRE after reading TDRE = 1
	 When the DTC is activated by a TXI interrupt and writes data to TDR
1	[Setting conditions] (Initial value)
	When the TE bit in SCR is 0
	When data is transferred from TDR to TSR and data can be written to TDR

Bit 6—Receive Data Register Full (RDRF): Indicates that the received data is stored in RDR.

Bit 6

RDRF	Description
0	[Clearing conditions] (Initial value)
	 When 0 is written in RDRF after reading RDRF = 1
	 When the DTC is activated by an RXI interrupt and reads data from RDR
1	[Setting condition]
	When serial reception ends normally and receive data is transferred from RSR to RDR
Note:	RDR and the RDRF flag are not affected and retain their previous values when an error is

detected during reception or when the RE bit in SCR is cleared to 0.

If reception of the next data is completed while the RDRF flag is still set to 1, an overrun

If reception of the next data is completed while the RDRF flag is still set to 1, an overrur error will occur and the receive data will be lost.

Bit 5—Overrun Error (ORER): Indicates that an overrun error occurred during reception, causing abnormal termination.

Bit 5

ORER	Description	
0	[Clearing condition]	(Initial value)*1
	When 0 is written in ORER after reading ORER = 1	
1	[Setting condition]	
	When the next serial reception is completed while RDRF = 1*2	
Natas: 1	The ODED flow is not effected and notains its massicus state when the	DE 1:1: - 00D :-

Notes: 1. The ORER flag is not affected and retains its previous state when the RE bit in SCR is cleared to 0.

2. The receive data prior to the overrun error is retained in RDR, and the data received subsequently is lost. Also, subsequent serial reception cannot be continued while the ORER flag is set to 1. In synchronous mode, serial transmission cannot be continued, either.

Bit 4—Framing Error (FER): Indicates that a framing error occurred during reception in asynchronous mode, causing abnormal termination.

Bit 4

FER	Description	
0	[Clearing condition]	(Initial value)*1
	When 0 is written in FER after reading FER = 1	
1	[Setting condition]	
	When the SCI checks the stop bit at the end of the receive data vand the stop bit is 0^{*2}	when reception ends,
Notes: 1	1. The FER flag is not affected and retains its previous state when the	ne RE bit in SCR is

cleared to 0.

2. In 2-stop-bit mode, only the first stop bit is checked for a value of 0; the second stop bit is not checked. If a framing error occurs, the receive data is transferred to RDR but the RDRF flag is not set. Also, subsequent serial reception cannot be continued while the FER flag is set to 1. In synchronous mode, serial transmission cannot be continued, either.

Bit 3—Parity Error (PER): Indicates that a parity error occurred during reception using parity addition in asynchronous mode, causing abnormal termination.

Bit 3

PER	Description							
0	[Clearing condition]	(Initial value)*1						
	When 0 is written in PER after reading PER = 1							
1	[Setting condition]							
	When, in reception, the number of 1 bits in the receive data plus the parity bit does not match the parity setting (even or odd) specified by the O/\overline{E} bit in SMR*2							
Notes: 1	. The PER flag is not affected and retains its previous state when the cleared to 0.	e RE bit in SCR is						
2	. If a parity error occurs, the receive data is transferred to RDR but tl	he RDRF flag is not						

Bit 2—Transmit End (TEND): Indicates that there is no valid data in TDR when the last bit of the transmit character is sent, and transmission has been ended.

1. In synchronous mode, serial transmission cannot be continued, either.

set. Also, subsequent serial reception cannot be continued while the PER flag is set to

The TEND flag is read-only and cannot be modified.

Bit 2

TEND	Description									
0	[Clearing conditions]									
	 When 0 is written in TDRE after reading TDRE = 1 									
	When the DTC is activated by a TXI interrupt and writes data to TDR									
1	[Setting conditions] (I	Initial value)								
	When the TE bit in SCR is 0									
	When TDRE = 1 at transmission of the last bit of a 1-byte serial transmit	t character								

Bit 1—Multiprocessor Bit (MPB): When reception is performed using a multiprocessor format in asynchronous mode, MPB stores the multiprocessor bit in the receive data.

MPB is a read-only bit, and cannot be modified.

Bit 1

MPB		Description	
0		[Clearing condition]	(Initial value)*
		When data with a 0 multiprocessor bit is received	
1		[Setting condition]	
		When data with a 1 multiprocessor bit is received	
Note:	*	Retains its previous state when the RE bit in SCR is cleared to 0 w format.	vith multiprocessor

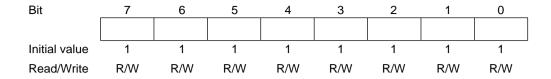
Bit 0—Multiprocessor Bit Transfer (MPBT): When transmission is performed using a multiprocessor format in asynchronous mode, MPBT stores the multiprocessor bit to be added to the transmit data.

The MPBT bit setting is invalid when a multiprocessor format is not used, when not transmitting, and in synchronous mode.

Bit 0

MPBT	Description	
0	Data with a 0 multiprocessor bit is transmitted	(Initial value)
1	Data with a 1 multiprocessor bit is transmitted	

15.2.8 Bit Rate Register (BRR)



BRR is an 8-bit register that sets the serial transfer bit rate in accordance with the baud rate generator operating clock selected by bits CKS1 and CKS0 in SMR.

BRR can be read or written to by the CPU at all times.

BRR is initialized to H'FF by a reset, and in standby mode, watch mode, subactive mode, subsleep mode, and module stop mode.

As baud rate generator control is performed independently for each channel, different values can be set for each channel.

Table 15.3 shows sample BRR settings in asynchronous mode, and table 15.4 shows sample BRR settings in synchronous mode.

Table 15.3 BRR Settings for Various Bit Rates (Asynchronous Mode)

Operating Frequency φ (MHz)

	φ = 2 MHz			ф	φ = 2.097152 MHz			$\phi = 2.4576 \text{ MHz}$			φ = 3 MHz		
Bit Rate (bits/s)	n	N	Error (%)	n	N	Error (%)	n	N	Error (%)	n	N	Error (%)	
110	1	141	0.03	1	148	-0.04	1	174	-0.26	1	212	0.03	
150	1	103	0.16	1	108	0.21	1	127	0.00	1	155	0.16	
300	0	207	0.16	0	217	0.21	0	255	0.00	1	77	0.16	
600	0	103	0.16	0	108	0.21	0	127	0.00	0	155	0.16	
1200	0	51	0.16	0	54	-0.70	0	63	0.00	0	77	0.16	
2400	0	25	0.16	0	26	1.14	0	31	0.00	0	38	0.16	
4800	0	12	0.16	0	13	-2.48	0	15	0.00	0	19	-2.34	
9600	_	_	_	0	6	-2.48	0	7	0.00	0	9	-2.34	
19200	_	_	_	_	_	_	0	3	0.00	0	4	-2.34	
31250	0	1	0.00	_	_	_	_	_	_	0	2	0.00	
38400	_	_	_	_	_	_	0	1	0.00	_	_	_	

Operating Frequency φ (MHz)

	φ = 3.6864 MHz				φ = 4 N	1Hz	(φ = 4.9152 MHz			φ = 5 MHz		
Bit Rate (bits/s)	n	N	Error (%)	n	N	Error (%)	n	N	Error (%)	n	N	Error (%)	
110	2	64	0.70	2	70	0.03	2	86	0.31	2	88	-0.25	
150	1	191	0.00	1	207	0.16	1	255	0.00	2	64	0.16	
300	1	95	0.00	1	103	0.16	1	127	0.00	1	129	0.16	
600	0	191	0.00	0	207	0.16	0	255	0.00	1	64	0.16	
1200	0	95	0.00	0	103	0.16	0	127	0.00	0	129	0.16	
2400	0	47	0.00	0	51	0.16	0	63	0.00	0	64	0.16	
4800	0	23	0.00	0	25	0.16	0	31	0.00	0	32	-1.36	
9600	0	11	0.00	0	12	0.16	0	15	0.00	0	15	1.73	
19200	0	5	0.00	_	_	_	0	7	0.00	0	7	1.73	
31250				0	3	0.00	0	4	-1.70	0	4	0.00	
38400	0	2	0.00	_			0	3	0.00	0	3	1.73	

Operating Frequency φ (MHz)

	φ = 6 MHz				$\phi = 6.144 \text{ MHz}$			$\phi = 7.3728 \text{ MHz}$			φ = 8 MHz		
Bit Rate (bits/s)	n	N	Error (%)	n	N	Error (%)	n	N	Error (%)	n	N	Error (%)	
110	2	106	-0.44	2	108	0.08	2	130	-0.07	2	141	0.03	
150	2	77	0.16	2	79	0.00	2	95	0.00	2	103	0.16	
300	1	155	0.16	1	159	0.00	1	191	0.00	1	207	0.16	
600	1	77	0.16	1	79	0.00	1	95	0.00	1	103	0.16	
1200	0	155	0.16	0	159	0.00	0	191	0.00	0	207	0.16	
2400	0	77	0.16	0	79	0.00	0	95	0.00	0	103	0.16	
4800	0	38	0.16	0	39	0.00	0	47	0.00	0	51	0.16	
9600	0	19	-2.34	0	19	0.00	0	23	0.00	0	25	0.16	
19200	0	9	-2.34	0	9	0.00	0	11	0.00	0	12	0.16	
31250	0	5	0.00	0	5	2.40	_	_	_	0	7	0.00	
38400	0	4	-2.34	0	4	0.00	0	5	0.00	_	_	_	

Operating Frequency φ (MHz)

	φ = 9.8304 MHz				φ = 10 MHz			φ = 12 MHz			φ = 12.288 MHz		
Bit Rate (bits/s)	n	N	Error (%)	n	N	Error (%)	n	N	Error (%)	n	N	Error (%)	
110	2	174	-0.26	2	177	-0.25	2	212	0.03	2	217	0.08	
150	2	127	0.00	2	129	0.16	2	155	0.16	2	159	0.00	
300	1	255	0.00	2	64	0.16	2	77	0.16	2	79	0.00	
600	1	127	0.00	1	129	0.16	1	155	0.16	1	159	0.00	
1200	0	255	0.00	1	64	0.16	1	77	0.16	1	79	0.00	
2400	0	127	0.00	0	129	0.16	0	155	0.16	0	159	0.00	
4800	0	63	0.00	0	64	0.16	0	77	0.16	0	79	0.00	
9600	0	31	0.00	0	32	-1.36	0	38	0.16	0	39	0.00	
19200	0	15	0.00	0	15	1.73	0	19	-2.34	0	19	0.00	
31250	0	9	-1.70	0	9	0.00	0	11	0.00	0	11	2.40	
38400	0	7	0.00	0	7	1.73	0	9	-2.34	0	9	0.00	

Operating	Frequency	ሐ (MHz)
Operannu	rieduelicv	W CIVITIZI

	φ = 14 MHz			φ = 14.7456 MHz			φ = 16 MHz			φ = 17.2032 MHz		
Bit Rate (bits/s)	n	N	Error (%)	n	N	Error (%)	n	N	Error (%)	n	N	Error (%)
110	2	248	-0.17	3	64	0.70	3	70	0.03	3	75	0.48
150	2	181	0.16	2	191	0.00	2	207	0.16	2	223	0.00
300	2	90	0.16	2	95	0.00	2	103	0.16	2	111	0.00
600	1	181	0.16	1	191	0.00	1	207	0.16	1	223	0.00
1200	1	90	0.16	1	95	0.00	1	103	0.16	1	111	0.00
2400	0	181	0.16	0	191	0.00	0	207	0.16	0	223	0.00
4800	0	90	0.16	0	95	0.00	0	103	0.16	0	111	0.00
9600	0	45	-0.93	0	47	0.00	0	51	0.16	0	55	0.00
19200	0	22	-0.93	0	23	0.00	0	25	0.16	0	27	0.00
31250	0	13	0.00	0	14	-1.70	0	15	0.00	0	16	1.20
38400				0	11	0.00	0	12	0.16	0	13	0.00

Operating Frequency φ (MHz)

		φ = 18 Ν	ИHz	¢	= 19.660	8 MHz		φ = 20 Ι	ИНz
Bit Rate (bits/s)	n	N	Error (%)	n	N	Error (%)	n	N	Error (%)
110	3	79	-0.12	3	86	0.31	3	88	-0.25
150	2	233	0.16	2	255	0.00	3	64	0.16
300	2	116	0.16	2	127	0.00	2	129	0.16
600	1	233	0.16	1	255	0.00	2	64	0.16
1200	1	116	0.16	1	127	0.00	1	129	0.16
2400	0	233	0.16	0	255	0.00	1	64	0.16
4800	0	116	0.16	0	127	0.00	0	129	0.16
9600	0	58	-0.69	0	63	0.00	0	64	0.16
19200	0	28	1.02	0	31	0.00	0	32	-1.36
31250	0	17	0.00	0	19	-1.70	0	19	0.00
38400	0	14	-2.34	0	15	0.00	0	15	1.73

Table 15.4 BRR Settings for Various Bit Rates (Synchronous Mode)

Operating Frequency ϕ (MHz)

Bit Rate	φ:	= 2 MHz	ф	= 4 MHz	ф	= 8 MHz	φ =	= 10 MHz	φ:	= 16 MHz	φ=	= 20 MHz
(bits/s)	n	N	n	N	n	N	n	N	n	N	n	N
110	3	70	_	_								
250	2	124	2	249	3	124	_	_	3	249		
500	1	249	2	124	2	249	_	_	3	124	_	_
1 k	1	124	1	249	2	124	_	_	2	249	_	_
2.5 k	0	199	1	99	1	199	1	249	2	99	2	124
5 k	0	99	0	199	1	99	1	124	1	199	1	249
10 k	0	49	0	99	0	199	0	249	1	99	1	124
25 k	0	19	0	39	0	79	0	99	0	159	0	199
50 k	0	9	0	19	0	39	0	49	0	79	0	99
100 k	0	4	0	9	0	19	0	24	0	39	0	49
250 k	0	1	0	3	0	7	0	9	0	15	0	19
500 k	0	0*	0	1	0	3	0	4	0	7	0	9
1 M			0	0*	0	1			0	3	0	4
2.5 M							0	0*			0	1
5 M											0	0*

Legend:

Blank: Cannot be set.

—: Can be set, but there will be a degree of error.

*: Continuous transfer is not possible.

Note: As far as possible, the setting should be made so that the error is no more than 1%.

The BRR setting is found from the following equations.

Asynchronous mode:

$$N = \frac{\phi}{64 \times 2^{2n-1} \times B} \times 10^6 - 1$$

Synchronous mode:

$$N = \frac{\phi}{8 \times 2^{2n-1} \times B} \times 10^6 - 1$$

Where B: Bit rate (bits/s)

N: BRR setting for band rate generator $(0 \le N \le 255)$

φ: Operating frequency (MHz)

n: Baud rate generator input clock (n = 0 to 3)
(See the table below for the relation between n and the clock.)

		SM	R Setting
n	Clock	CKS1	CKS0
0	ф	0	0
1	φ/4	0	1
2	ф/16	1	0
3	φ/64	1	1

The bit rate error in asynchronous mode is found from the following equation:

Error (%) =
$$\left\{ \frac{\phi \times 10^6}{(N+1) \times B \times 64 \times 2^{2n-1}} - 1 \right\} \times 100$$

Table 15.5 shows the maximum bit rate for each frequency in asynchronous mode. Tables 15.6 and 15.7 show the maximum bit rates with external clock input.

Table 15.5 Maximum Bit Rate for Each Frequency (Asynchronous Mode)

φ (MHz)	Maximum Bit Rate (bits/s)	n	N
2	62500	0	0
2.097152	65536	0	0
2.4576	76800	0	0
3	93750	0	0
3.6864	115200	0	0
4	125000	0	0
4.9152	153600	0	0
5	156250	0	0
6	187500	0	0
6.144	192000	0	0
7.3728	230400	0	0
8	250000	0	0
9.8304	307200	0	0
10	312500	0	0
12	375000	0	0
12.288	384000	0	0
14	437500	0	0
14.7456	460800	0	0
16	500000	0	0
17.2032	537600	0	0
18	562500	0	0
19.6608	614400	0	0
20	625000	0	0

Table 15.6 Maximum Bit Rate with External Clock Input (Asynchronous Mode)

φ (MHz)	External Input Clock (MHz)	Maximum Bit Rate (bits/s)
2	0.5000	31250
2.097152	0.5243	32768
2.4576	0.6144	38400
3	0.7500	46875
3.6864	0.9216	57600
4	1.0000	62500
4.9152	1.2288	76800
5	1.2500	78125
6	1.5000	93750
6.144	1.5360	96000
7.3728	1.8432	115200
8	2.0000	125000
9.8304	2.4576	153600
10	2.5000	156250
12	3.0000	187500
12.288	3.0720	192000
14	3.5000	218750
14.7456	3.6864	230400
16	4.0000	250000
17.2032	4.3008	268800
18	4.5000	281250
19.6608	4.9152	307200
20	5.0000	312500

Table 15.7 Maximum Bit Rate with External Clock Input (Synchronous Mode)

φ (MHz)	External Input Clock (MHz)	Maximum Bit Rate (bits/s)
2	0.3333	333333.3
4	0.6667	666666.7
6	1.0000	1000000.0
8	1.3333	1333333.3
10	1.6667	1666666.7
12	2.0000	2000000.0
14	2.3333	2333333.3
16	2.6667	2666666.7
18	3.0000	3000000.0
20	3.3333	3333333.3

15.2.9 Serial Interface Mode Register (SCMR)

Bit	7	6	5	4	3	2	1	0
	_	_	_	_	SDIR	SINV	_	SMIF
Initial value	1	1	1	1	0	0	1	0
Read/Write	_	_	_	_	R/W	R/W	_	R/W

SCMR is an 8-bit readable/writable register used to select SCI functions.

SCMR is initialized to H'F2 by a reset, and in standby mode, watch mode, subactive mode, subsleep mode, and module stop mode.

Bits 7 to 4—Reserved: These bits cannot be modified and are always read as 1.

Bit 3—Data Transfer Direction (SDIR): Selects the serial/parallel conversion format.

Bit 3

SDIR	Description	
0	TDR contents are transmitted LSB-first	(Initial value)
	Receive data is stored in RDR LSB-first	
1	TDR contents are transmitted MSB-first	
	Receive data is stored in RDR MSB-first	

Bit 2—Data Invert (SINV): Specifies inversion of the data logic level. The SINV bit does not affect the logic level of the parity bit(s): parity bit inversion requires inversion of the O/\overline{E} bit in SMR.

Bit 2

SINV	Description	
0	TDR contents are transmitted without modification	(Initial value)
	Receive data is stored in RDR without modification	
1	TDR contents are inverted before being transmitted	
	Receive data is stored in RDR in inverted form	

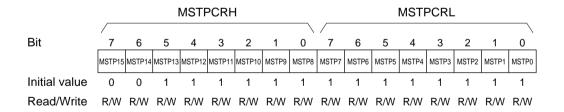
Bit 1—Reserved: This bit cannot be modified and is always read as 1.

Bit 0—Serial Communication Interface Mode Select (SMIF): Reserved bit. 1 should not be written in this bit.

Bit 0

SMIF	Description	
0	Normal SCI mode	(Initial value)
1	Reserved mode	

15.2.10 Module Stop Control Register (MSTPCR)



MSTPCR, comprising two 8-bit readable/writable registers, performs module stop mode control. When bit MSTP7, MSTP6, or MSTP5 is set to 1, SCI0, SCI1, or SCI2 operation, respectively, stops at the end of the bus cycle and a transition is made to module stop mode. For details, see section 25.5, Module Stop Mode.

MSTPCR is initialized to H'3FFF by a reset and in hardware standby mode. It is not initialized in software standby mode.

Bit 7—Module Stop (MSTP7): Specifies the SCI0 module stop mode.

MSTPCRL

Bit 7

MSTP7	Description	
0	SCI0 module stop mode is cleared	
1	SCI0 module stop mode is set	(Initial value)

Bit 6—Module Stop (MSTP6): Specifies the SCI1 module stop mode.

MSTPCRL

Bit 6

MSTP6	Description	
0	SCI1 module stop mode is cleared	
1	SCI1 module stop mode is set	(Initial value)

Bit 5—Module Stop (MSTP5): Specifies the SCI2 module stop mode.

MSTPCRL

Bit 5

MSTP5	Description	
0	SCI2 module stop mode is cleared	
1	SCI2 module stop mode is set	(Initial value)

15.2.11 Keyboard Comparator Control Register (KBCOMP)

Bit	7	6	5	4	3	2	1	0
	IrE	IrCKS2	IrCKS1	IrCKS0	KBADE	KBCH2	KBCH1	КВСН0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

KBCOMP is an 8-bit readable/writable register that selects the functions of SCI2 and the A/D converter.

KBCOMP is initialized to H'00 by a reset and in hardware standby mode.

Bit 7—IrDA Enable (IrE): Specifies normal SCI operation or IrDA operation for SCI2 input/output.

Bit 7

IrE	Description	
0	The TxD2/IrTxD and RxD2/IrRxD pins function as TxD2 and RxD2	(Initial value)
1	The TxD2/IrTxD and RxD2/IrRxD pins function as IrTxD and IrRxD	

Bits 6 to 4—IrDA Clock Select 2 to 0 (IrCKS2 to IrCKS0): These bits specify the high pulse width in IrTxD output pulse encoding when the IrDA function is enabled.

Bit 6	Bit 5	Bit 4						
IrCKS2	IrCKS1	IrCKS0	Description					
0	0	0	B × 3/16 (3/16 of the bit rate)	(Initial value)				
		1	φ/2					
	1	0	φ/4					
		1	φ/8					
1	0	0	φ/16					
		1	ф/32					
	1	0	φ/64					
		1	φ/128					

Bits 3 to 0—Keyboard Comparator Control: See the description in section 20, A/D converter.

15.3 Operation

15.3.1 Overview

The SCI can carry out serial communication in two modes: asynchronous mode in which synchronization is achieved character by character, and synchronous mode in which synchronization is achieved with clock pulses.

Selection of asynchronous or synchronous mode and the transmission format is made using SMR as shown in table 15.8. The SCI clock is determined by a combination of the C/\overline{A} bit in SMR and the CKE1 and CKE0 bits in SCR, as shown in table 15.9.

Asynchronous Mode:

- Data length: Choice of 7 or 8 bits
- Choice of parity addition, multiprocessor bit addition, and addition of 1 or 2 stop bits (the combination of these parameters determines the transfer format and character length)
- Detection of framing, parity, and overrun errors, and breaks, during reception
- Choice of internal or external clock as SCI clock source
 - When internal clock is selected:
 - The SCI operates on the baud rate generator clock and a clock with the same frequency as the bit rate can be output
 - When external clock is selected:
 - A clock with a frequency of 16 times the bit rate must be input (the built-in baud rate generator is not used)

Synchronous Mode:

- Transfer format: Fixed 8-bit data
- Detection of overrun errors during reception
- Choice of internal or external clock as SCI clock source
 - When internal clock is selected:
 - The SCI operates on the baud rate generator clock and a serial clock is output off-chip
 - When external clock is selected:
 - The built-in baud rate generator is not used, and the SCI operates on the input serial clock

Table 15.8 SMR Settings and Serial Transfer Format Selection

	S	MR Sett	ings		_		SCI Transfer Format					
	Bit 6	Bit 2	Bit 5	Bit 3	_ Mode	Data Length	Multi- processor Bit	Parity Bit	Stop Bit Length			
0	0	0	0	0	Asynchronous	8-bit data	No	No	1 bit			
				1	mode				2 bits			
			1	0	=			Yes	1 bit			
				1	_				2 bits			
	1	_	0	0	_	7-bit data	_	No	1 bit			
				1	_				2 bits			
			1	0	_			Yes	1 bit			
				1	_				2 bits			
	0	1	_	0	Asynchronous	8-bit data	Yes	No	1 bit			
			_	1	mode (multi- processor format)				2 bits			
	1	_	_	0	-processor format)	7-bit data	_		1 bit			
			_	1	_				2 bits			
1	_	_	_	_	Synchronous mode	8-bit data	No	_	None			

Table 15.9 SMR and SCR Settings and SCI Clock Source Selection

SMR	SCR	Setting		SCI Transfer Clock					
Bit 7	Bit 1	Bit 0	_	Clock					
	CKE1	CKE0	Mode	Source	SCK Pin Function				
0	0	0	Asynchronous	Internal	SCI does not use SCK pin				
		1	_mode		Outputs clock with same frequency as bit rate				
	1	0	_	External	Inputs clock with frequency of 16 times				
		1	_		the bit rate				
1	0	0	Synchronous	Internal	Outputs serial clock				
		1	mode						
	1	0	_	External	Inputs serial clock				
		1	_						

15.3.2 Operation in Asynchronous Mode

In asynchronous mode, characters are sent or received, each preceded by a start bit indicating the start of communication and followed by one or two stop bits indicating the end of communication. Serial communication is thus carried out with synchronization established on a character-by-character basis.

Inside the SCI, the transmitter and receiver are independent units, enabling full-duplex communication. Both the transmitter and the receiver also have a double-buffered structure, so that data can be read or written during transmission or reception, enabling continuous data transfer.

Figure 15.2 shows the general format for asynchronous serial communication.

In asynchronous serial communication, the transmission line is usually held in the mark state (high level). The SCI monitors the transmission line, and when it goes to the space state (low level), recognizes a start bit and starts serial communication.

One serial communication character consists of a start bit (low level), followed by data (in LSB-first order), a parity bit (high or low level), and finally one or two stop bits (high level).

In asynchronous mode, the SCI performs synchronization at the falling edge of the start bit in reception. The SCI samples the data on the 8th pulse of a clock with a frequency of 16 times the length of one bit, so that the transfer data is latched at the center of each bit.

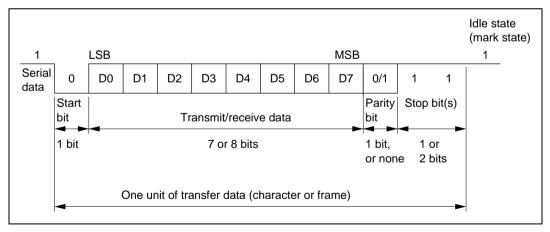


Figure 15.2 Data Format in Asynchronous Communication (Example with 8-Bit Data, Parity, Two Stop Bits)

Data Transfer Format

Table 15.10 shows the data transfer formats that can be used in asynchronous mode. Any of 12 transfer formats can be selected by settings in SMR.

Table 15.10 Serial Transfer Formats (Asynchronous Mode)

	SMR	Setting	gs	Serial Transfer Format and Frame Length													
CHR	PE	MP	STOP	1	2	3	4	5	6	7	8	9	10	11	12		
0	0	0	0	S					STOP								
0	0	0	1	S				8-bit	data				STOP	STOP STOP			
0	1	0	0	S				8-bit	data				Р	STOP			
0	1	0	1	S				8-bit	data				Р	STOP	STOP	-	
1	0	0	0	S	7-bit data							STOP)P				
1	0	0	1	S			7-	-bit da	ıta			STOP	DP STOP				
1	1	0	0	S			7-	-bit da	ıta			Р	STOP				
1	1	0	1	S			7-	-bit da	ıta			Р	STOP	STOP	=		
0	_	1	0	S	8-bit data								MPB STOP				
0	_	1	1	S	8-bit data								MPB STOP STOP			-	
1	_	1	0	S	7-bit data								B STOP				
1	_	1	1	S			7-	-bit da	ıta			МРВ	STOP	STOP	-		

Legend:

S: Start bit STOP: Stop bit P: Parity bit

MPB: Multiprocessor bit

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Clock

Either an internal clock generated by the built-in baud rate generator or an external clock input at the SCK pin can be selected as the SCI's serial clock, according to the setting of the C/\overline{A} bit in SMR and the CKE1 and CKE0 bits in SCR. For details of SCI clock source selection, see table 15.9.

When an external clock is input at the SCK pin, the clock frequency should be 16 times the bit rate used.

When the SCI is operated on an internal clock, the clock can be output from the SCK pin. The frequency of the clock output in this case is equal to the bit rate, and the phase is such that the rising edge of the clock is at the center of each transmit data bit, as shown in figure 15.3.

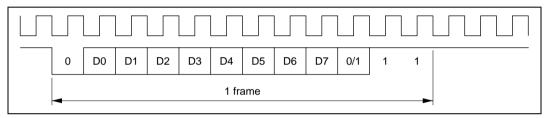


Figure 15.3 Relation between Output Clock and Transfer Data Phase (Asynchronous Mode)

Data Transfer Operations

SCI Initialization (**Asynchronous Mode**): Before transmitting and receiving data, first clear the TE and RE bits in SCR to 0, then initialize the SCI as described below.

When the operating mode, transfer format, etc., is changed, the TE and RE bits must be cleared to 0 before making the change using the following procedure. When the TE bit is cleared to 0, the TDRE flag is set to 1 and TSR is initialized. Note that clearing the RE bit to 0 does not change the contents of the RDRF, PER, FER, and ORER flags, or the contents of RDR.

When an external clock is used the clock should not be stopped during operation, including initialization, since operation is uncertain.

Figure 15.4 shows a sample SCI initialization flowchart.

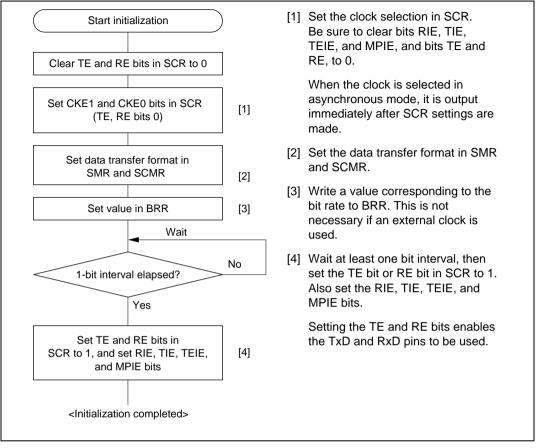


Figure 15.4 Sample SCI Initialization Flowchart

Serial Data Transmission (Asynchronous Mode): Figure 15.5 shows a sample flowchart for serial transmission.

The following procedure should be used for serial data transmission.

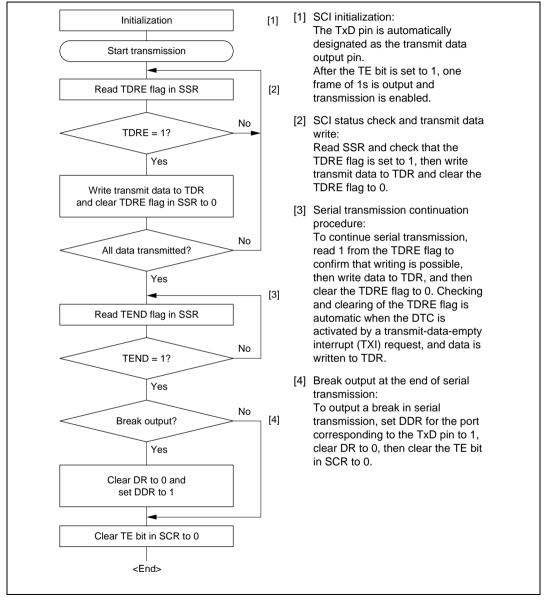


Figure 15.5 Sample Serial Transmission Flowchart

In serial transmission, the SCI operates as described below.

- 1. The SCI monitors the TDRE flag in SSR, and if it is 0, recognizes that data has been written to TDR, and transfers the data from TDR to TSR.
- 2. After transferring data from TDR to TSR, the SCI sets the TDRE flag to 1 and starts transmission.

If the TIE bit is set to 1 at this time, a transmit data empty interrupt (TXI) is generated.

The serial transmit data is sent from the TxD pin in the following order.

a. Start bit:

One 0-bit is output.

b. Transmit data:

8-bit or 7-bit data is output in LSB-first order.

c. Parity bit or multiprocessor bit:

One parity bit (even or odd parity), or one multiprocessor bit is output.

A format in which neither a parity bit nor a multiprocessor bit is output can also be selected.

d. Stop bit(s):

One or two 1-bits (stop bits) are output.

e. Mark state:

1 is output continuously until the start bit that starts the next transmission is sent.

3. The SCI checks the TDRE flag at the timing for sending the stop bit.

If the TDRE flag is cleared to 0, the data is transferred from TDR to TSR, the stop bit is sent, and then serial transmission of the next frame is started.

If the TDRE flag is set to 1, the TEND flag in SSR is set to 1, the stop bit is sent, and then the mark state is entered in which 1 is output continuously. If the TEIE bit in SCR is set to 1 at this time, a TEI interrupt request is generated.

Figure 15.6 shows an example of the operation for transmission in asynchronous mode.

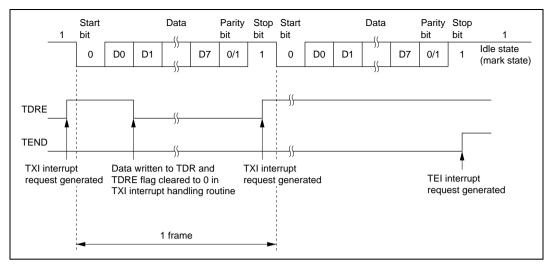


Figure 15.6 Example of Operation in Transmission in Asynchronous Mode (Example with 8-Bit Data, Parity, One Stop Bit)

Serial Data Reception (Asynchronous Mode): Figure 15.7 shows a sample flowchart for serial reception.

The following procedure should be used for serial data reception.

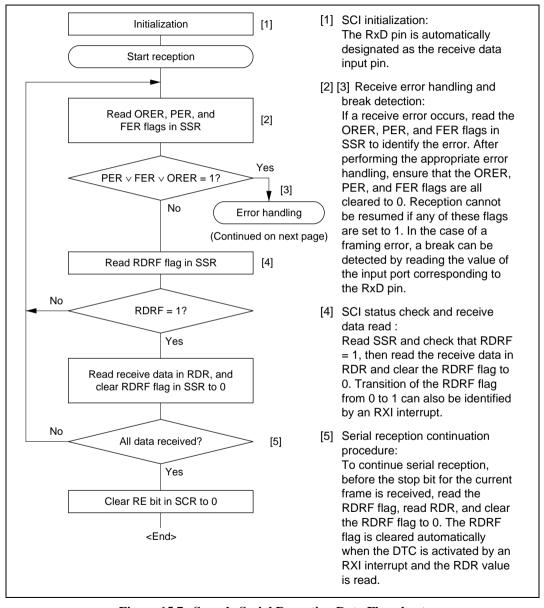


Figure 15.7 Sample Serial Reception Data Flowchart

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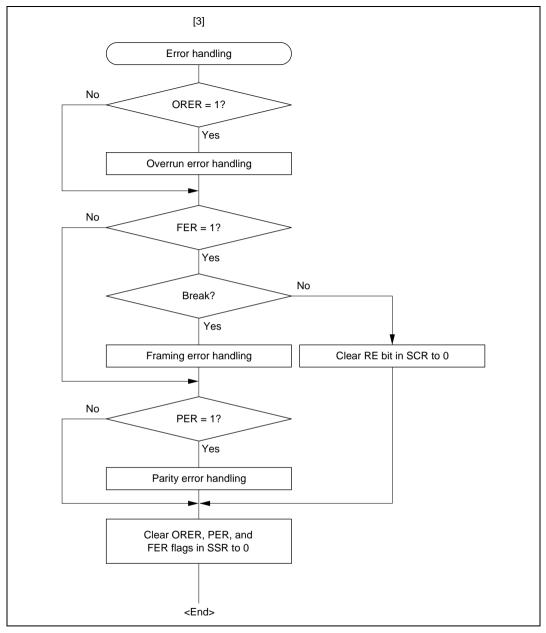


Figure 15.7 Sample Serial Reception Data Flowchart (cont)

In serial reception, the SCI operates as described below.

- 1. The SCI monitors the transmission line, and if a 0 stop bit is detected, performs internal synchronization and starts reception.
- 2. The received data is stored in RSR in LSB-to-MSB order.
- 3. The parity bit and stop bit are received.

After receiving these bits, the SCI carries out the following checks.

a. Parity check:

The SCI checks whether the number of 1 bits in the receive data agrees with the parity (even or odd) set in the O/\overline{E} bit in SMR.

b. Stop bit check:

The SCI checks whether the stop bit is 1.

If there are two stop bits, only the first is checked.

c. Status check:

The SCI checks whether the RDRF flag is 0, indicating that the receive data can be transferred from RSR to RDR.

If all the above checks are passed, the RDRF flag is set to 1, and the receive data is stored in RDR.

If a receive error* is detected in the error check, the operation is as shown in table 15.11.

Note: * Subsequent receive operations cannot be performed when a receive error has occurred.

Also note that the RDRF flag is not set to 1 in reception, and so the error flags must be cleared to 0.

4. If the RIE bit in SCR is set to 1 when the RDRF flag changes to 1, a receive-data-full interrupt (RXI) request is generated.

Also, if the RIE bit in SCR is set to 1 when the ORER, PER, or FER flag changes to 1, a receive-error interrupt (ERI) request is generated.



Table 15.11 Receive Errors and Conditions for Occurrence

Receive Error	Abbreviation	Occurrence Condition	Data Transfer
Overrun error	ORER	When the next data reception is completed while the RDRF flag in SSR is set to 1	Receive data is not transferred from RSR to RDR
Framing error	FER	When the stop bit is 0	Receive data is transferred from RSR to RDR
Parity error	PER	When the received data differs from the parity (even or odd) set in SMR	Receive data is transferred from RSR to RDR

Figure 15.8 shows an example of the operation for reception in asynchronous mode.

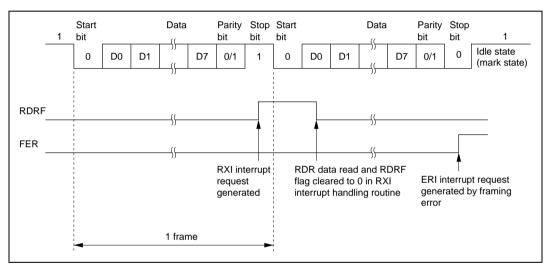


Figure 15.8 Example of SCI Operation in Reception (Example with 8-Bit Data, Parity, One Stop Bit)

15.3.3 Multiprocessor Communication Function

The multiprocessor communication function performs serial communication using a multiprocessor format, in which a multiprocessor bit is added to the transfer data, in asynchronous mode. Use of this function enables data transfer to be performed among a number of processors sharing transmission lines.

When multiprocessor communication is carried out, each receiving station is addressed by a unique ID code.

The serial communication cycle consists of two component cycles: an ID transmission cycle which specifies the receiving station, and a data transmission cycle. The multiprocessor bit is used to differentiate between the ID transmission cycle and the data transmission cycle.

The transmitting station first sends the ID of the receiving station with which it wants to perform serial communication as data with a 1 multiprocessor bit added. It then sends transmit data as data with a 0 multiprocessor bit added.

The receiving station skips the data until data with a 1 multiprocessor bit is sent.

When data with a 1 multiprocessor bit is received, the receiving station compares that data with its own ID. The station whose ID matches then receives the data sent next. Stations whose ID does not match continue to skip the data until data with a 1 multiprocessor bit is again received. In this way, data communication is carried out among a number of processors.

Figure 15.9 shows an example of inter-processor communication using a multiprocessor format.

Data Transfer Format

There are four data transfer formats.

When a multiprocessor format is specified, the parity bit specification is invalid.

For details, see table 15.10.

Clock

See the section on asynchronous mode.



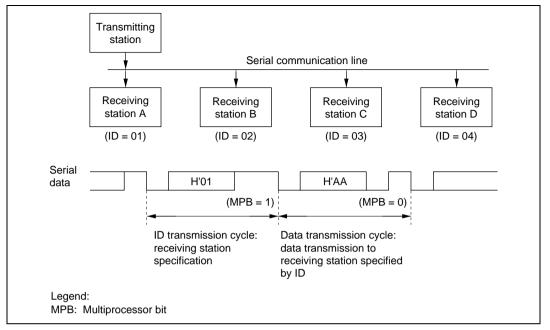


Figure 15.9 Example of Inter-Processor Communication Using Multiprocessor Format (Transmission of Data H'AA to Receiving Station A)

Data Transfer Operations

Multiprocessor Serial Data Transmission: Figure 15.10 shows a sample flowchart for multiprocessor serial data transmission.

The following procedure should be used for multiprocessor serial data transmission.

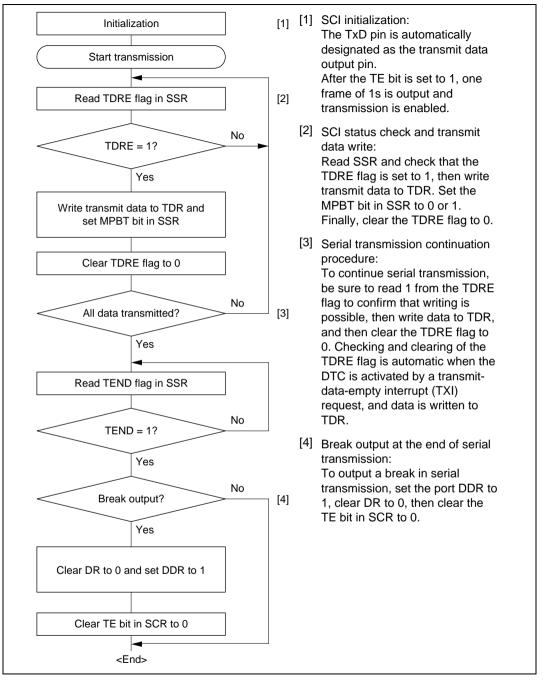


Figure 15.10 Sample Multiprocessor Serial Transmission Flowchart

In serial transmission, the SCI operates as described below.

- 1. The SCI monitors the TDRE flag in SSR, and if it is 0, recognizes that data has been written to TDR, and transfers the data from TDR to TSR.
- 2. After transferring data from TDR to TSR, the SCI sets the TDRE flag to 1 and starts transmission.

If the TIE bit is set to 1 at this time, a transmit-data-empty interrupt (TXI) is generated.

The serial transmit data is sent from the TxD pin in the following order.

a. Start bit:

One 0-bit is output.

b. Transmit data:

8-bit or 7-bit data is output in LSB-first order.

c. Multiprocessor bit

One multiprocessor bit (MPBT value) is output.

d. Stop bit(s):

One or two 1-bits (stop bits) are output.

e. Mark state:

1 is output continuously until the start bit that starts the next transmission is sent.

3. The SCI checks the TDRE flag at the timing for sending the stop bit.

If the TDRE flag is cleared to 0, data is transferred from TDR to TSR, the stop bit is sent, and then serial transmission of the next frame is started.

If the TDRE flag is set to 1, the TEND flag in SSR is set to 1, the stop bit is sent, and then the mark state is entered in which 1 is output continuously. If the TEIE bit in SCR is set to 1 at this time, a transmit-end interrupt (TEI) request is generated.

Figure 15.11 shows an example of SCI operation for transmission using a multiprocessor format.

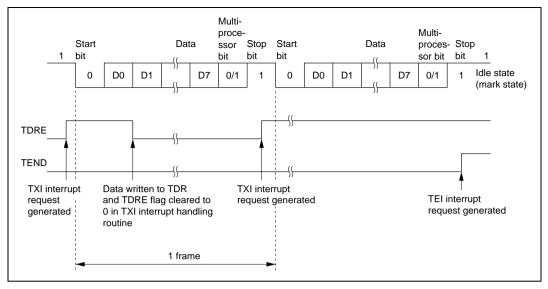


Figure 15.11 Example of SCI Operation in Transmission (Example with 8-Bit Data, Multiprocessor Bit, One Stop Bit)

Multiprocessor Serial Data Reception: Figure 15.12 shows a sample flowchart for multiprocessor serial reception.

The following procedure should be used for multiprocessor serial data reception.

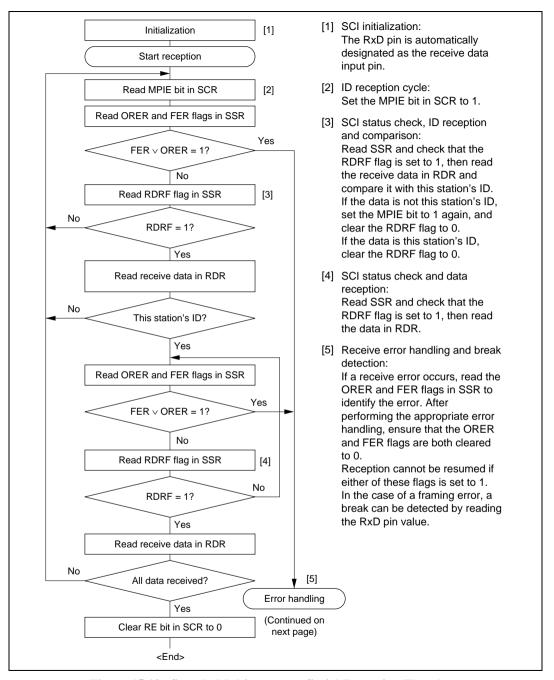


Figure 15.12 Sample Multiprocessor Serial Reception Flowchart

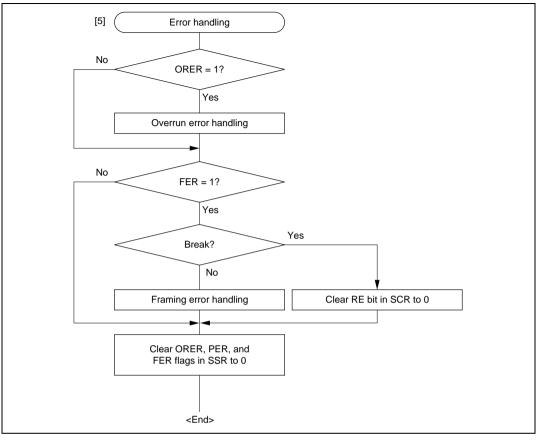


Figure 15.12 Sample Multiprocessor Serial Reception Flowchart (cont)

Figure 15.13 shows an example of SCI operation for multiprocessor format reception.

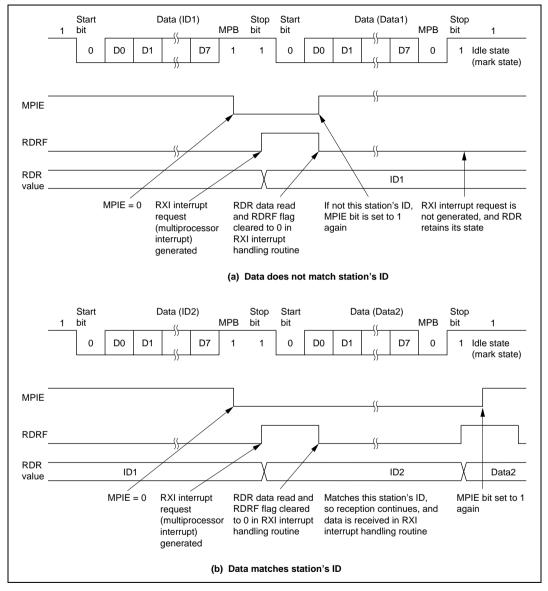


Figure 15.13 Example of SCI Operation in Reception (Example with 8-Bit Data, Multiprocessor Bit, One Stop Bit)

15.3.4 Operation in Synchronous Mode

In synchronous mode, data is transmitted or received in synchronization with clock pulses, making it suitable for high-speed serial communication.

Inside the SCI, the transmitter and receiver are independent units, enabling full-duplex communication by use of a common clock. Both the transmitter and the receiver also have a double-buffered structure, so that data can be read or written during transmission or reception, enabling continuous data transfer.

Figure 15.14 shows the general format for synchronous serial communication.

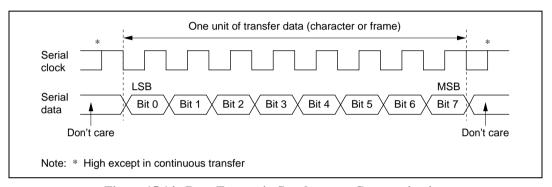


Figure 15.14 Data Format in Synchronous Communication

In synchronous serial communication, data on the transmission line is output from one falling edge of the serial clock to the next. Data is guaranteed valid at the rising edge of the serial clock.

In synchronous serial communication, one character consists of data output starting with the LSB and ending with the MSB. After the MSB is output, the transmission line holds the MSB state.

In synchronous mode, the SCI receives data in synchronization with the rising edge of the serial clock.

Data Transfer Format

A fixed 8-bit data format is used.

No parity or multiprocessor bits are added.



Clock

Either an internal clock generated by the built-in baud rate generator or an external serial clock input at the SCK pin can be selected, according to the setting of the C/\overline{A} bit in SMR and the CKE1 and CKE0 bits in SCR. For details on SCI clock source selection, see table 15.9.

When the SCI is operated on an internal clock, the serial clock is output from the SCK pin.

Eight serial clock pulses are output in the transfer of one character, and when no transfer is performed the clock is fixed high. When only receive operations are performed, however, the serial clock is output until an overrun error occurs or the RE bit is cleared to 0. To perform receive operations in units of one character, select an external clock as the clock source.

Data Transfer Operations

SCI Initialization (Synchronous Mode): Before transmitting and receiving data, first clear the TE and RE bits in SCR to 0, then initialize the SCI as described below.

When the operating mode, transfer format, etc., is changed, the TE and RE bits must be cleared to 0 before making the change using the following procedure. When the TE bit is cleared to 0, the TDRE flag is set to 1 and TSR is initialized. Note that clearing the RE bit to 0 does not change the settings of the RDRF, PER, FER, and ORER flags, or the contents of RDR.

Figure 15.15 shows a sample SCI initialization flowchart.

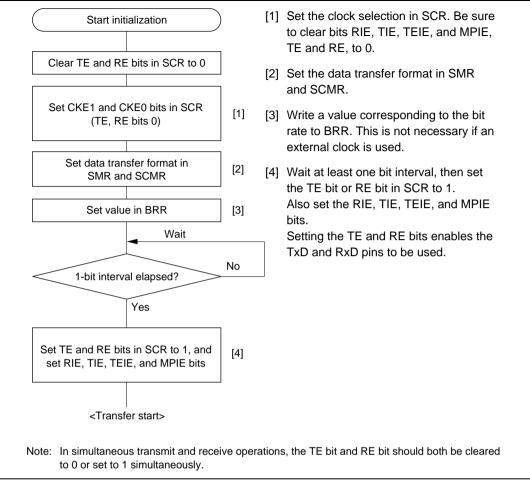


Figure 15.15 Sample SCI Initialization Flowchart

Serial Data Transmission (Synchronous Mode): Figure 15.16 shows a sample flowchart for serial transmission.

The following procedure should be used for serial data transmission.

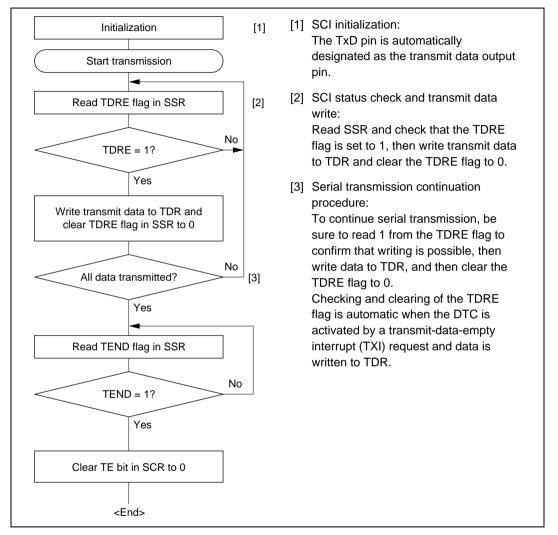


Figure 15.16 Sample Serial Transmission Flowchart

In serial transmission, the SCI operates as described below.

- 1. The SCI monitors the TDRE flag in SSR, and if it is 0, recognizes that data has been written to TDR, and transfers the data from TDR to TSR.
- 2. After transferring data from TDR to TSR, the SCI sets the TDRE flag to 1 and starts transmission. If the TIE bit is set to 1 at this time, a transmit-data-empty interrupt (TXI) is generated.

When clock output mode has been set, the SCI outputs 8 serial clock pulses. When use of an external clock has been specified, data is output synchronized with the input clock.

The serial transmit data is sent from the TxD pin starting with the LSB (bit 0) and ending with the MSB (bit 7).

- 3. The SCI checks the TDRE flag at the timing for sending the MSB (bit 7).
 - If the TDRE flag is cleared to 0, data is transferred from TDR to TSR, and serial transmission of the next frame is started.

If the TDRE flag is set to 1, the TEND flag in SSR is set to 1, the MSB (bit 7) is sent, and the TxD pin maintains its state.

If the TEIE bit in SCR is set to 1 at this time, a transmit-end interrupt (TEI) request is generated.

4. After completion of serial transmission, the SCK pin is held in a constant state.

Figure 15.17 shows an example of SCI operation in transmission.

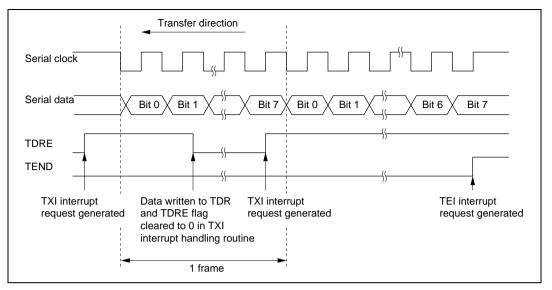


Figure 15.17 Example of SCI Operation in Transmission

Serial Data Reception (Synchronous Mode): Figure 15.18 shows a sample flowchart for serial reception.

The following procedure should be used for serial data reception.

When changing the operating mode from asynchronous to synchronous, be sure to check that the ORER, PER, and FER flags are all cleared to 0.

The RDRF flag will not be set if the FER or PER flag is set to 1, and neither transmit nor receive operations will be possible.

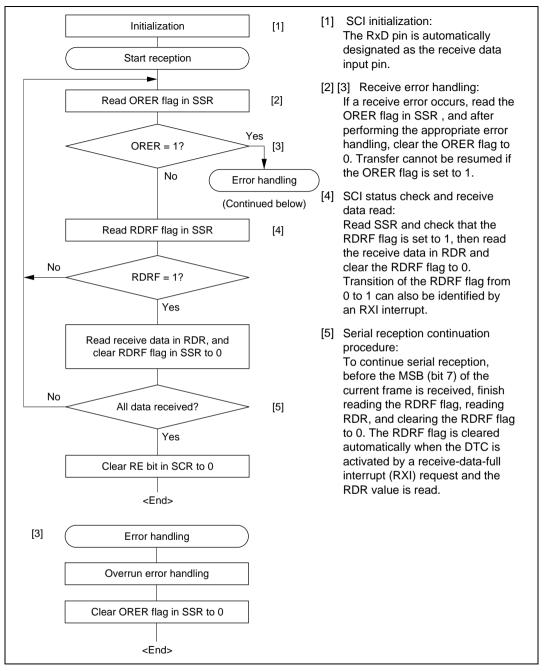


Figure 15.18 Sample Serial Reception Flowchart

In serial reception, the SCI operates as described below.

- 1. The SCI performs internal initialization in synchronization with serial clock input or output.
- 2. The received data is stored in RSR in LSB-to-MSB order.

After reception, the SCI checks whether the RDRF flag is 0 and the receive data can be transferred from RSR to RDR.

If this check is passed, the RDRF flag is set to 1, and the receive data is stored in RDR. If a receive error is detected in the error check, the operation is as shown in table 15.11.

Neither transmit nor receive operations can be performed subsequently when a receive error has been found in the error check.

3. If the RIE bit in SCR is set to 1 when the RDRF flag changes to 1, a receive-data-full interrupt (RXI) request is generated.

Also, if the RIE bit in SCR is set to 1 when the ORER flag changes to 1, a receive-error interrupt (ERI) request is generated.

Figure 15.19 shows an example of SCI operation in reception.

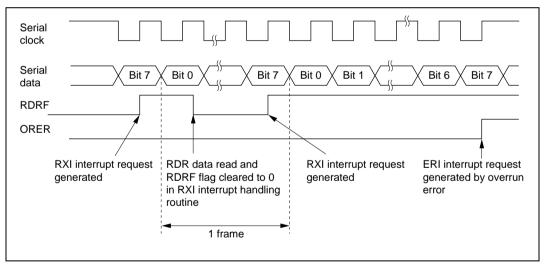


Figure 15.19 Example of SCI Operation in Reception

Simultaneous Serial Data Transmission and Reception (Synchronous Mode): Figure 15.20 shows a sample flowchart for simultaneous serial transmit and receive operations.

The following procedure should be used for simultaneous serial data transmit and receive operations.

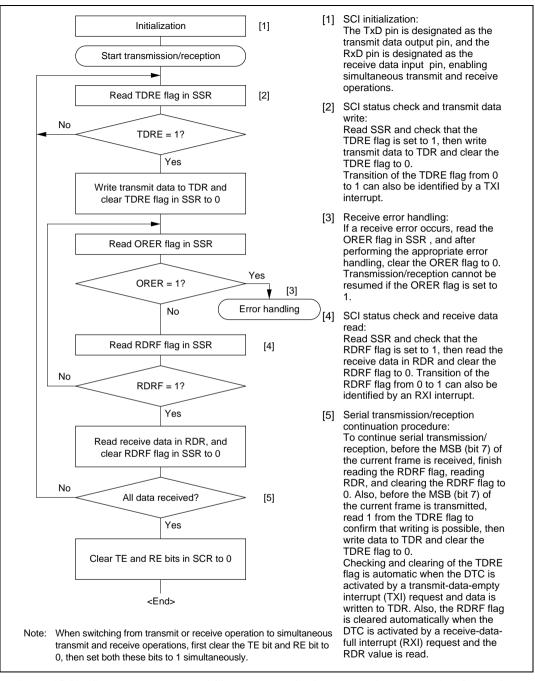


Figure 15.20 Sample Flowchart of Simultaneous Serial Transmit and Receive Operations

15.3.5 IrDA Operation

Figure 15.21 shows a block diagram of the IrDA function.

When the IrDA function is enabled with bit IrE in KBCOMP, the SCI channel 2 TxD2 and RxD2 signals are subjected to waveform encoding/decoding conforming to IrDA specification version 1.0 (IrTxD and IrRxD pins). By connecting these pins to an infrared transceiver/receiver, it is possible to implement infrared transmission/reception conforming to the IrDA specification version 1.0 system.

In the IrDA specification version 1.0 system, communication is started at a transfer rate of 9600 bps, and subsequently the transfer rate can be varied as necessary. As the IrDA interface in the this LSI does not include a function for varying the transfer rate automatically, the transfer rate setting must be changed by software.

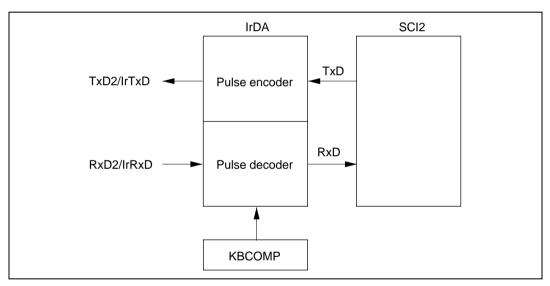


Figure 15.21 Block Diagram of IrDA Function

Transmission

In transmission, the output signal (UART frame) from the SCI is converted to an IR frame by the IrDA interface (see figure 15.22).

When the serial data is 0, a high-level pulses of 3/16 the bit rate (interval equivalent to the width of one bit) is output (initial value). The high-level pulse can be varied according to the setting of bits IrCKS2 to IrCKS0 in KBCOMP.

The high-level pulse width is fixed at a minimum of $1.41~\mu s$, and a maximum of $(3/16 + 2.5\%) \times$ bit rate or $(3/16 \times bit rate) + 1.08~\mu s$. When system clock ϕ is 20 MHz, 1.6 μ s can be set as the minimum high-level pulse width at 1.41 μ s or above.

When the serial data is 1, no pulse is output.

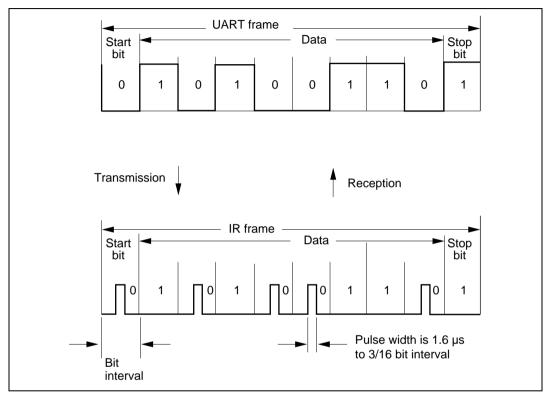


Figure 15.22 IrDA Transmit/Receive Operations

Reception

In reception, IR frame data is converted to a UART frame by the IrDA interface, and input to the SCI.

When a high-level pulse is detected, 0 data is output, and if there is no pulse during a one-bit interval, 1 data is output. Note that a pulse shorter than the minimum pulse width of 1.41 μ s will be identified as a 0 signal.

High-Level Pulse Width Selection

Operating

Table 15.12 shows possible settings for bits IrCKS2 to IrCKS0 (minimum pulse width), and this LSI operating frequencies and bit rates, for making the pulse width shorter than 3/16 times the bit rate in transmission.

Table 15.12 Bit IrCKS2 to IrCKS0 Settings

Bit Rate (bps) (Upper Row) / Bit Interval × 3/16 (μs) (Lower Row)

Operating Frequency	2400	9600	19200	38400	57600	115200
φ (MHz)	78.13	19.53	9.77	4.88	3.26	1.63
2	010	010	010	010	010	_
2.097152	010	010	010	010	010	_
2.4576	010	010	010	010	010	_
3	011	011	011	011	011	_
3.6864	011	011	011	011	011	011
4.9152	011	011	011	011	011	011
5	011	011	011	011	011	011
6	100	100	100	100	100	100
6.144	100	100	100	100	100	100
7.3728	100	100	100	100	100	100
8	100	100	100	100	100	100
9.8304	100	100	100	100	100	100
10	100	100	100	100	100	100
12	101	101	101	101	101	101
12.288	101	101	101	101	101	101
14	101	101	101	101	101	101
14.7456	101	101	101	101	101	101
16	101	101	101	101	101	101
16.9344	101	101	101	101	101	101
17.2032	101	101	101	101	101	101
18	101	101	101	101	101	101
19.6608	101	101	101	101	101	101
20	101	101	101	101	101	101

Legend:

^{—:} An SCI bit rate setting cannot be mode.

15.4 SCI Interrupts

The SCI has four interrupt sources: the transmit-end interrupt (TEI) request, receive-error interrupt (ERI) request, receive-data-full interrupt (RXI) request, and transmit-data-empty interrupt (TXI) request. Table 15.13 shows the interrupt sources and their relative priorities. Individual interrupt sources can be enabled or disabled with the TIE, RIE, and TEIE bits in SCR. Each kind of interrupt request is sent to the interrupt controller independently.

When the TDRE flag in SSR is set to 1, a TXI interrupt request is generated. When the TEND flag in SSR is set to 1, a TEI interrupt request is generated. A TXI interrupt can activate the DTC to perform data transfer. The TDRE flag is cleared to 0 automatically when data transfer is performed by the DTC. The DTC cannot be activated by a TEI interrupt request.

When the RDRF flag in SSR is set to 1, an RXI interrupt request is generated. When the ORER, PER, or FER flag in SSR is set to 1, an ERI interrupt request is generated. An RXI interrupt can activate the DTC to perform data transfer. The RDRF flag is cleared to 0 automatically when data transfer is performed by the DTC. The DTC cannot be activated by an ERI interrupt request.

Table 15.13 SCI Interrupt Sources

Channel	Interrupt Source	Description	DTC Activation	Priority*
0	ERI	Receive error (ORER, FER, or PER)	Not possible	High
	RXI	Receive data register full (RDRF)	Possible	_
	TXI	Transmit data register empty (TDRE)	Possible	_
	TEI	Transmit end (TEND)	Not possible	_
1	ERI	Receive error (ORER, PER, or PER)	Not possible	_
	RXI	Receive data register full (RDRF)	Possible	_
	TXI	Transmit data register empty (TDRE)	Possible	_
	TEI	Transmit end (TEND)	Not possible	_
2	ERI	Receive error (ORER, PER, or PER)	Not possible	_
	RXI	Receive data register full (RDRF)	Possible	_
	TXI	Transmit data register empty (TDRE)	Possible	_
	TEI	Transmit end (TEND)	Not possible	Low

Note: * The table shows the initial state immediately after a reset. Relative channel priorities can be changed by the interrupt controller.

The TEI interrupt is requested when the TEND flag is set to 1 while the TEIE bit is set to 1. The TEND flag is cleared at the same time as the TDRE flag. Consequently, if a TEI interrupt and a



TXI interrupt are requested simultaneously, the TXI interrupt will have priority for acceptance, and the TDRE flag and TEND flag may be cleared. Note that the TEI interrupt will not be accepted in this case.

15.5 Usage Notes

The following points should be noted when using the SCI.

Relation between Writes to TDR and the TDRE Flag

The TDRE flag in SSR is a status flag that indicates that transmit data has been transferred from TDR to TSR. When the SCI transfers data from TDR to TSR, the TDRE flag is set to 1.

Data can be written to TDR regardless of the state of the TDRE flag. However, if new data is written to TDR when the TDRE flag is cleared to 0, the data stored in TDR will be lost since it has not yet been transferred to TSR. It is therefore essential to check that the TDRE flag is set to 1 before writing transmit data to TDR.

Operation when Multiple Receive Errors Occur Simultaneously

If a number of receive errors occur at the same time, the state of the status flags in SSR is as shown in table 15.14. If there is an overrun error, data is not transferred from RSR to RDR, and the receive data is lost.

Table 15.14 State of SSR Status Flags and Transfer of Receive Data

	SSR St	atus Fla	gs	Receive Data Transfer	
RDRF	ORER	FER	PER	RSR to RDR	Receive Errors
1	1	0	0	Х	Overrun error
0	0	1	0	0	Framing error
0	0	0	1	0	Parity error
1	1	1	0	Х	Overrun error + framing error
1	1	0	1	Х	Overrun error + parity error
0	0	1	1	0	Framing error + parity error
1	1	1	1	Х	Overrun error + framing error + parity error

Notes: O: Receive data is transferred from RSR to RDR.

X: Receive data is not transferred from RSR to RDR.

Break Detection and Processing

When a framing error (FER) is detected, a break can be detected by reading the RxD pin value directly. In a break, the input from the RxD pin becomes all 0s, and so the FER flag is set, and the parity error flag (PER) may also be set.

Note that, since the SCI continues the receive operation after receiving a break, even if the FER flag is cleared to 0, it will be set to 1 again.

Sending a Break

The TxD pin has a dual function as an I/O port whose direction (input or output) is determined by DR and DDR. This feature can be used to send a break.

Between serial transmission initialization and setting of the TE bit to 1, the mark state is replaced by the value of DR (the pin does not function as the TxD pin until the TE bit is set to 1). Consequently, DDR and DR for the port corresponding to the TxD pin should first be set to 1.

To send a break during serial transmission, first clear DR to 0, then clear the TE bit to 0.

When the TE bit is cleared to 0, the transmitter is initialized regardless of the current transmission state, the TxD pin becomes an I/O port, and 0 is output from the TxD pin.

Receive Error Flags and Transmit Operations (Synchronous Mode Only)

Transmission cannot be started when a receive error flag (ORER, PER, or FER) is set to 1, even if the TDRE flag is cleared to 0. Be sure to clear the receive error flags to 0 before starting transmission.

Note also that receive error flags cannot be cleared to 0 even if the RE bit is cleared to 0.

Receive Data Sampling Timing and Reception Margin in Asynchronous Mode

In asynchronous mode, the SCI operates on a base clock with a frequency of 16 times the transfer rate.

In reception, the SCI samples the falling edge of the start bit using the base clock, and performs internal synchronization. Receive data is latched internally at the rising edge of the 8th pulse of the base clock. This is illustrated in figure 15.23.



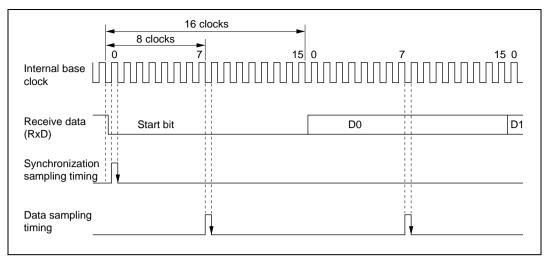


Figure 15.23 Receive Data Sampling Timing in Asynchronous Mode

Thus the receive margin in asynchronous mode is given by equation (1) below.

$$M = \left| \left(0.5 - \frac{1}{2N} \right) - (L - 0.5)F - \frac{|D - 0.5|}{N} (1 + F) \right| \times 100\%$$
(1)

Where M: Receive margin (%)

N: Ratio of bit rate to clock (N = 16)

D: Clock duty (D = 0 to 1.0)

L: Frame length (L = 9 to 12)

F: Absolute value of clock rate deviation

Assuming values of F = 0 and D = 0.5 in equation (1), a receive margin of 46.875% is given by equation (2) below.

When D = 0.5 and F = 0,

$$M = \left(0.5 - \frac{1}{2 \times 16}\right) \times 100\%$$

$$= 46.875\%$$
......(2)

However, this is only a theoretical value, and a margin of 20% to 30% should be allowed in system design.

Restrictions on Use of DTC

- When an external clock source is used as the serial clock, the transmit clock should not be input until at least 5 φ clock cycles after TDR is updated by the DTC. Misoperation may occur if the transmit clock is input within 4 clock cycles after TDR is updated. (Figure 15.24)
- When RDR is read by the DTC, be sure to set the activation source to the relevant SCI receivedata-full interrupt (RXI).

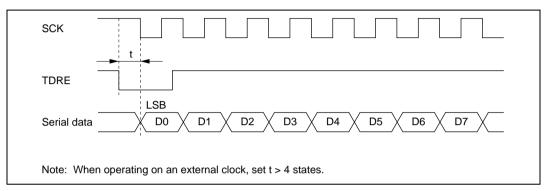


Figure 15.24 Example of Synchronous Transmission by DTC

Section 16 I²C Bus Interface [Option]

A two-channel I²C bus interface is available as an option in the H8S/2148 Group and H8S/2147N. The I²C bus interface is not available for the H8S/2144 Group. Observe the following notes when using this option.

1. For mask-ROM versions, a W is added to the part number in products in which this optional function is used.

Examples: HD6432147SWFA

2. The product number is identical for F-ZTAT versions. However, be sure to inform your Renesas sales representative if you will be using this option.

16.1 Overview

A two-channel I²C bus interface is available for the H8S/2148 Group and H8S/2147N as an option. The I²C bus interface conforms to and provides a subset of the Philips I²C bus (inter-IC bus) interface functions. The register configuration that controls the I²C bus differs partly from the Philips configuration, however.

Each I²C bus interface channel uses only one data line (SDA) and one clock line (SCL) to transfer data, saving board and connector space.

16.1.1 Features

- Selection of addressing format or non-addressing format
 - I²C bus format: addressing format with acknowledge bit, for master/slave operation
 - Serial format: non-addressing format without acknowledge bit, for master operation only
- Conforms to Philips I²C bus interface (I²C bus format)
- Two ways of setting slave address (I²C bus format)
- Start and stop conditions generated automatically in master mode (I²C bus format)
- Selection of acknowledge output levels when receiving (I²C bus format)
- Automatic loading of acknowledge bit when transmitting (I²C bus format)
- Wait function in master mode (I²C bus format)

A wait can be inserted by driving the SCL pin low after data transfer, excluding acknowledgement. The wait can be cleared by clearing the interrupt flag.

- Wait function in slave mode (I²C bus format)
 - A wait request can be generated by driving the SCL pin low after data transfer, excluding acknowledgement. The wait request is cleared when the next transfer becomes possible.
- Three interrupt sources
 - Data transfer end (including transmission mode transition with I²C bus format and address reception after loss of master arbitration)
 - Address match: when any slave address matches or the general call address is received in slave receive mode (I²C bus format)
 - Stop condition detection
- Selection of 16 internal clocks (in master mode)
- Direct bus drive (with SCL and SDA pins)
 - Two pins—P52/SCL0 and P97/SDA0—(normally NMOS push-pull outputs) function as NMOS open-drain outputs when the bus drive function is selected.
 - Two pins—P86/SCL1 and P42/SDA1—(normally CMOS pins) function as NMOS-only outputs when the bus drive function is selected.
- Automatic switching from formatless mode to I²C bus format (channel 0 only)
 - Formatless operation (no start/stop conditions, non-addressing mode) in slave mode
 - Operation using a common data pin (SDA) and independent clock pins (VSYNCI, SCL)
 - Automatic switching from formatless mode to I²C bus format on the fall of the SCL pin

16.1.2 Block Diagram

Figure 16.1 shows a block diagram of the I²C bus interface.

Figure 16.2 shows an example of I/O pin connections to external circuits. Channel 0 I/O pins and channel 1 I/O pins differ in structure, and have different specifications for permissible applied voltages. For details, see section 26, Electrical Characteristics.



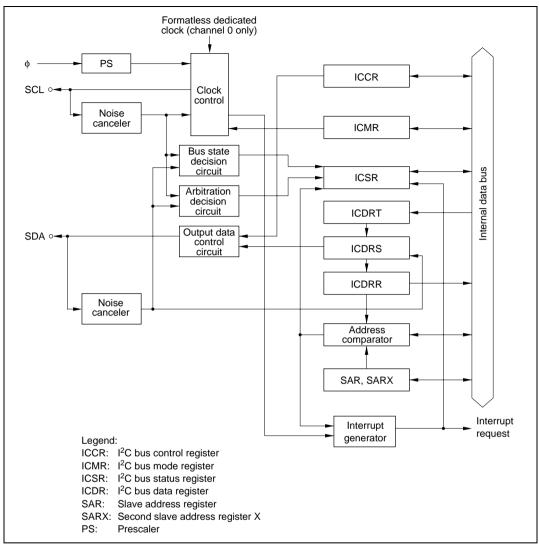


Figure 16.1 Block Diagram of I²C Bus Interface

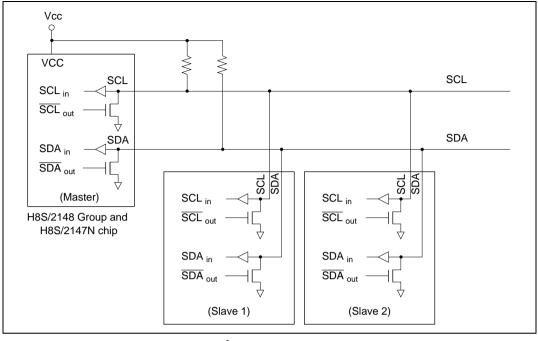


Figure 16.2 I²C Bus Interface Connections (Example: H8S/2148 Group and H8S/2147N Chip as Master)

16.1.3 Input/Output Pins

Table 16.1 summarizes the input/output pins used by the I²C bus interface.

Table 16.1 I²C Bus Interface Pins

Channel	Name	${\bf Abbreviation}^*$	I/O	Function
0	Serial clock	SCL0	I/O	IIC0 serial clock input/output
	Serial data	SDA0	I/O	IIC0 serial data input/output
	Formatless serial clock	VSYNCI	Input	IIC0 formatless serial clock input
1	Serial clock	SCL1	I/O	IIC1 serial clock input/output
	Serial data	SDA1	I/O	IIC1 serial data input/output

Note: $\ \ ^*$ In the text, the channel subscript is omitted, and only SCL and SDA are used.



16.1.4 Register Configuration

Table 16.2 summarizes the registers of the I²C bus interface.

Table 16.2 Register Configuration

Channel	Name	Abbreviation	R/W	Initial Value	Address*1
0	I ² C bus control register	ICCR0	R/W	H'01	H'FFD8
	I ² C bus status register	ICSR0	R/W	H'00	H'FFD9
	I ² C bus data register	ICDR0	R/W	_	H'FFDE*2
	I ² C bus mode register	ICMR0	R/W	H'00	H'FFDF*2
	Slave address register	SAR0	R/W	H'00	H'FFDF*2
	Second slave address register	SARX0	R/W	H'01	H'FFDE*2
1	I ² C bus control register	ICCR1	R/W	H'01	H'FF88
	I ² C bus status register	ICSR1	R/W	H'00	H'FF89
	I ² C bus data register	ICDR1	R/W	_	H'FF8E*2
	I ² C bus mode register	ICMR1	R/W	H'00	H'FF8F*2
	Slave address register	SAR1	R/W	H'00	H'FF8F*2
	Second slave address register	SARX1	R/W	H'01	H'FF8E*2
Common	Serial/timer control register	STCR	R/W	H'00	H'FFC3
	DDC switch register	DDCSWR	R/W	H'0F	H'FEE6
	Module stop control	MSTPCRH	R/W	H'3F	H'FF86
	register	MSTPCRL	R/W	H'FF	H'FF87

Notes: 1. Lower 16 bits of the address.

2. The register that can be written or read depends on the ICE bit in the I²C bus control register. The slave address register can be accessed when ICE = 0, and the I²C bus mode register can be accessed when ICE = 1.

The I²C bus interface registers are assigned to the same addresses as other registers. Register selection is performed by means of the IICE bit in the serial/timer control register (STCR).

16.2 Register Descriptions

16.2.1 I²C Bus Data Register (ICDR)

Bit	7	6	5	4	3	2	1	0
	ICDR7	ICDR6	ICDR5	ICDR4	ICDR3	ICDR2	ICDR1	ICDR0
Initial value	_	_	_	_	_	_	_	_
Read/Write	R/W							

ICDRR

Bit	7	6	5	4	3	2	1	0
	ICDRR7	ICDRR6	ICDRR5	ICDRR4	ICDRR3	ICDRR2	ICDRR1	ICDRR0
Initial value	_	_	_	_	_	_	_	
Read/Write	R	R	R	R	R	R	R	R

ICDRS

Bit	7	6	5	4	3	2	1	0
	ICDRS7	ICDRS6	ICDRR5	ICDRS4	ICDRS3	ICDRS2	ICDRS1	ICDRS0
Initial value	_	_	_	_	_	_	_	_
Read/Write	_	_	_	_	_	_	_	_

ICDRT

Bit	7	6	5	4	3	2	1	0
	ICDRT7	ICDRT6	ICDRT5	ICDRT4	ICDRT3	ICDRT2	ICDRT1	ICDRT0
Initial value	_	_	_	_	_	_	_	_
Read/Write	W	W	W	W	W	W	W	W

• TDRE, RDRF (internal flags)

Bit		
	TDRE	RDRF
Initial value	0	0
Read/Write	_	_

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ICDR is an 8-bit readable/writable register that is used as a transmit data register when transmitting and a receive data register when receiving. ICDR is divided internally into a shift register (ICDRS), receive buffer (ICDRR), and transmit buffer (ICDRT). ICDRS cannot be read or written by the CPU, ICDRR is read-only, and ICDRT is write-only. Data transfers among the three registers are performed automatically in coordination with changes in the bus state, and affect the status of internal flags such as TDRE and RDRF.

If IIC is in transmit mode and the next data is in ICDRT (the TDRE flag is 0) following transmission/reception of one frame of data using ICDRS, data is transferred automatically from ICDRT to ICDRS. If IIC is in receive mode and no previous data remains in ICDRR (the RDRF flag is 0) following transmission/reception of one frame of data using ICDRS, data is transferred automatically from ICDRS to ICDRR.

If the number of bits in a frame, excluding the acknowledge bit, is less than 8, transmit data and receive data are stored differently. Transmit data should be written justified toward the MSB side when MLS = 0, and toward the LSB side when MLS = 1. Receive data bits read from the LSB side should be treated as valid when MLS = 0, and bits read from the MSB side when MLS = 1.

ICDR is assigned to the same address as SARX, and can be written and read only when the ICE bit is set to 1 in ICCR.

The value of ICDR is undefined after a reset.

The TDRE and RDRF flags are set and cleared under the conditions shown below. Setting the TDRE and RDRF flags affects the status of the interrupt flags.

TDRE	Description								
0	The next transmit data is in ICDR (ICDRT), or transmission cannot be started (Initial value)								
	[Clearing conditions]								
	 When transmit data is written in ICDR (ICDRT) in transmit mode (TRS = 1) 								
	 When a stop condition is detected in the bus line state after a stop condition is issued with the I²C bus format or serial format selected 								
	 When a stop condition is detected with the I²C bus format selected 								
	 In receive mode (TRS = 0) 								
	(A 0 write to TRS during transfer is valid after reception of a frame containing an acknowledge bit)								
1	The next transmit data can be written in ICDR (ICDRT)								
	[Setting conditions]								
	 In transmit mode (TRS = 1), when a start condition is detected in the bus line state after a start condition is issued in master mode with the I²C bus format or serial format selected 								
	 At the first transmit mode setting (TRS = 1) (first transmit mode setting only) after the mode is switched from I²C bus mode to formatless mode 								
	When data is transferred from ICDRT to ICDRS								
	(Data transfer from ICDRT to ICDRS when TRS = 1 and TDRE = 0, and ICDRS is empty)								
	 When detecting a start condition and then switching from slave receive mode (TI = 0) state to transmit mode (TRS = 1) (first transmit mode switching only). 								

RDRF	Description	
0	The data in ICDR (ICDRR) is invalid	(Initial value)
	[Clearing condition]	
	When ICDR (ICDRR) receive data is read in receive mode	
1	The ICDR (ICDRR) receive data can be read	
	[Setting condition]	
	When data is transferred from ICDRS to ICDRR	
	(Data transfer from ICDRS to ICDRR in case of normal termination with RDRF = 0) $$	TRS = 0 and

16.2.2 Slave Address Register (SAR)

Bit	7	6	5	4	3	2	1	0
	SVA6	SVA5	SVA4	SVA3	SVA2	SVA1	SVA0	FS
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W						

SAR is an 8-bit readable/writable register that stores the slave address and selects the communication format. When the chip is in slave mode (and the addressing format is selected), if the upper 7 bits of SAR match the upper 7 bits of the first frame received after a start condition, the chip operates as the slave device specified by the master device. SAR is assigned to the same address as ICMR, and can be written and read only when the ICE bit is cleared to 0 in ICCR.

SAR is initialized to H'00 by a reset and in hardware standby mode.

Bits 7 to 1—Slave Address (SVA6 to SVA0): Set a unique address in bits SVA6 to SVA0, differing from the addresses of other slave devices connected to the I²C bus.

Bit 0—Format Select (FS): Used together with the FSX bit in SARX and the SW bit in DDCSWR to select the communication format.

- I²C bus format: addressing format with acknowledge bit
- Synchronous serial format: non-addressing format without acknowledge bit, for master mode only
- Formatless mode (channel 0 only): non-addressing format with or without acknowledge bit, slave mode only, start/stop conditions not detected

The FS bit also specifies whether or not SAR slave address recognition is performed in slave mode.

DDCSWR Bit 6	SAR Bit 0	SARX Bit 0	
sw	FS	FSX	Operating Mode
0	0	0	I ² C bus format
			 SAR and SARX slave addresses recognized
		1	I ² C bus format (Initial value)
			SAR slave address recognized
			SARX slave address ignored
	1	0	I ² C bus format
			SAR slave address ignored
			 SARX slave address recognized
		1	Synchronous serial format
			 SAR and SARX slave addresses ignored
1	0	0	Formatless mode (start/stop conditions not detected)
	0	1	Acknowledge bit used
	1	0	
	1	1	Formatless mode* (start/stop conditions not detected)
			No acknowledge bit

Note: * Do not set this mode when automatic switching to the I²C bus format is performed by means of the DDCSWR setting.

16.2.3 Second Slave Address Register (SARX)

Bit	7	6	5	4	3	2	1	0
	SVAX6	SVAX5	SVAX4	SVAX3	SVAX2	SVAX1	SVAX0	FSX
Initial value	0	0	0	0	0	0	0	1
Read/Write	R/W	R/W						

SARX is an 8-bit readable/writable register that stores the second slave address and selects the communication format. When the chip is in slave mode (and the addressing format is selected), if the upper 7 bits of SARX match the upper 7 bits of the first frame received after a start condition, the chip operates as the slave device specified by the master device. SARX is assigned to the same address as ICDR, and can be written and read only when the ICE bit is cleared to 0 in ICCR.

SARX is initialized to H'01 by a reset and in hardware standby mode.



Bits 7 to 1—Second Slave Address (SVAX6 to SVAX0): Set a unique address in bits SVAX6 to SVAX0, differing from the addresses of other slave devices connected to the I²C bus.

Bit 0—Format Select X (FSX): Used together with the FS bit in SAR and the SW bit in DDCSWR to select the communication format.

- I²C bus format: addressing format with acknowledge bit
- Synchronous serial format: non-addressing format without acknowledge bit, for master mode only
- Formatless mode: non-addressing format with or without acknowledge bit, slave mode only, start/stop conditions not detected

The FSX bit also specifies whether or not SARX slave address recognition is performed in slave mode. For details, see the description of the FS bit in SAR.

16.2.4 I²C Bus Mode Register (ICMR)

Bit	7	6	5	4	3	2	1	0
	MLS	WAIT	CKS2	CKS1	CKS0	BC2	BC1	BC0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

ICMR is an 8-bit readable/writable register that selects whether the MSB or LSB is transferred first, performs master mode wait control, and selects the master mode transfer clock frequency and the transfer bit count. ICMR is assigned to the same address as SAR. ICMR can be written and read only when the ICE bit is set to 1 in ICCR.

ICMR is initialized to H'00 by a reset and in hardware standby mode.

Bit 7—MSB-First/LSB-First Select (MLS): Selects whether data is transferred MSB-first or LSB-first.

If the number of bits in a frame, excluding the acknowledge bit, is less than 8, transmit data and receive data are stored differently. Transmit data should be written justified toward the MSB side when MLS = 0, and toward the LSB side when MLS = 1. Receive data bits read from the LSB side should be treated as valid when MLS = 0, and bits read from the MSB side when MLS = 1.

Do not set this bit to 1 when the I²C bus format is used.

Section 16 I2C Bus Interface [Option]

Bit 7

MLS	Description	
0	MSB-first	(Initial value)
1	LSB-first	

Bit 6—Wait Insertion Bit (WAIT): Selects whether to insert a wait between the transfer of data and the acknowledge bit, in master mode with the I²C bus format. When WAIT is set to 1, after the fall of the clock for the final data bit, the IRIC flag is set to 1 in ICCR, and a wait state begins (with SCL at the low level). When the IRIC flag is cleared to 0 in ICCR, the wait ends and the acknowledge bit is transferred. If WAIT is cleared to 0, data and acknowledge bits are transferred consecutively with no wait inserted.

The IRIC flag in ICCR is set to 1 on completion of the acknowledge bit transfer, regardless of the WAIT setting.

The setting of this bit is invalid in slave mode.

Bit 6

WAIT	Description	
0	Data and acknowledge bits transferred consecutively	(Initial value)
1	Wait inserted between data and acknowledge bits	

Bits 5 to 3—Serial Clock Select (CKS2 to CKS0): These bits, together with the IICX1 (channel 1) or IICX0 (channel 0) bit in the STCR register, select the serial clock frequency in master mode. They should be set according to the required transfer rate.

STCR
Bit 5 or 6 Bit 5 Bit 4 Bit 3

_						_			
	ro	n	- +	Δ	r	v	2	te	

IICX	CKS2	CKS1	CKS0	Clock	φ = 5 MHz	φ = 8 MHz	φ = 10 MHz	φ = 16 MHz	φ = 20 MHz
0	0	0	0	ф/28	179 kHz	286 kHz	357 kHz	571 kHz*	714 kHz*
			1	φ/40	125 kHz	200 kHz	250 kHz	400 kHz	500 kHz*
		1	0	ф/48	104 kHz	167 kHz	208 kHz	333 kHz	417 kHz*
			1	φ/64	78.1 kHz	125 kHz	156 kHz	250 kHz	313 kHz
	1	0	0	ф/80	62.5 kHz	100 kHz	125 kHz	200 kHz	250 kHz
			1	ф/100	50.0 kHz	80.0 kHz	100 kHz	160 kHz	200 kHz
		1	0	ф/112	44.6 kHz	71.4 kHz	89.3 kHz	143 kHz	179 kHz
			1	ф/128	39.1 kHz	62.5 kHz	78.1 kHz	125 kHz	156 kHz
1	0	0	0	φ/56	89.3 kHz	143 kHz	179 kHz	286 kHz	357 kHz
			1	φ/80	62.5 kHz	100 kHz	125 kHz	200 kHz	250 kHz
		1	0	ф/96	52.1 kHz	83.3 kHz	104 kHz	167 kHz	208 kHz
			1	ф/128	39.1 kHz	62.5 kHz	78.1 kHz	125 kHz	156 kHz
	1	0	0	ф/160	31.3 kHz	50.0 kHz	62.5 kHz	100 kHz	125 kHz
			1	ф/200	25.0 kHz	40.0 kHz	50.0 kHz	80.0 kHz	100 kHz
		1	0	φ/224	22.3 kHz	35.7 kHz	44.6 kHz	71.4 kHz	89.3 kHz
			1	φ/256	19.5 kHz	31.3 kHz	39.1 kHz	62.5 kHz	78.1 kHz

Note: * Outside the I²C bus interface specification range (normal mode: max. 100 kHz; high-speed mode: max. 400 kHz).

Bits 2 to 0—Bit Counter (BC2 to BC0): Bits BC2 to BC0 specify the number of bits to be transferred next. With the I²C bus format (when the FS bit in SAR or the FSX bit in SARX is 0), the data is transferred with one addition acknowledge bit. Bit BC2 to BC0 settings should be made during an interval between transfer frames. If bits BC2 to BC0 are set to a value other than 000, the setting should be made while the SCL line is low.

The bit counter is initialized to 000 by a reset and when a start condition is detected. The value returns to 000 at the end of a data transfer, including the acknowledge bit.

Bit 2	Bit 1	Bit 0	Bits	/Frame	
BC2 0	BC1	BC0	Synchronous Serial Format	I ² C Bus Format	
	0	0	8	9	(Initial value)
		1	1	2	
	1	0	2	3	
		1	3	4	
1	0	0	4	5	
		1	5	6	
	1	0	6	7	
		1	7	8	

16.2.5 I²C Bus Control Register (ICCR)

Bit	7	6	5	4	3	2	1	0
	ICE	IEIC	MST	TRS	ACKE	BBSY	IRIC	SCP
Initial value	0	0	0	0	0	0	0	1
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/(W)*	W

Note: * Only 0 can be written, to clear the flag.

ICCR is an 8-bit readable/writable register that enables or disables the I^2C bus interface, enables or disables interrupts, selects master or slave mode and transmission or reception, enables or disables acknowledgement, confirms the I^2C bus interface bus status, issues start/stop conditions, and performs interrupt flag confirmation.

ICCR is initialized to H'01 by a reset and in hardware standby mode.

Bit 7—I²**C Bus Interface Enable (ICE):** Selects whether or not the I²C bus interface is to be used. When ICE is set to 1, port pins function as SCL and SDA input/output pins and transfer operations are enabled. When ICE is cleared to 0, the I²C bus interface module is halted and its internal states are cleared.

The SAR and SARX registers can be accessed when ICE is 0. The ICMR and ICDR registers can be accessed when ICE is 1.



Bit 7	
ICE	Description
0	I ² C bus interface module disabled, with SCL and SDA signal pins set to port function (Initial value)
	I ² C bus interface module internal states initialized
	SAR and SARX can be accessed

I²C bus interface module enabled for transfer operations (pins SCL and SCA are

Bit 6—I²C **Bus Interface Interrupt Enable (IEIC):** Enables or disables interrupts from the I²C bus interface to the CPIJ

Bit 6

1

IEIC	 Description	
0	Interrupts disabled	(Initial value)
1	Interrupts enabled	

Bit 5—Master/Slave Select (MST)

Bit 4—Transmit/Receive Select (TRS)

driving the bus)

ICMR and ICDR can be accessed.

MST selects whether the I²C bus interface operates in master mode or slave mode.

TRS selects whether the I²C bus interface operates in transmit mode or receive mode.

In master mode with the I^2C bus format, when arbitration is lost, MST and TRS are both reset by hardware, causing a transition to slave receive mode. In slave receive mode with the addressing format (FS = 0 or FSX = 0), hardware automatically selects transmit or receive mode according to the R/\overline{W} bit in the first frame after a start condition.

Modification of the TRS bit during transfer is deferred until transfer of the frame containing the acknowledge bit is completed, and the changeover is made after completion of the transfer.

MST and TRS select the operating mode as follows.

Section 16 I²C Bus Interface [Option]

Bit 5	Bit 4		
MST	TRS	Operating Mode	
0	0	Slave receive mode	(Initial value)
	1	Slave transmit mode	
1	0	Master receive mode	
	1	Master transmit mode	

Bit 5

MST	Description				
0	Slave mode (Initial v	alue)			
	[Clearing conditions]				
	1. When 0 is written by software				
	 When bus arbitration is lost after transmission is started in I²C bus format master mode 				
1	Master mode				
	[Setting conditions]				
	1. When 1 is written by software (in cases other than clearing condition 2)				
	2. When 1 is written in MST after reading MST = 0 (in case of clearing condition	າ 2)			

Bit 4

TRS	Description					
0	Receive mode (Initial value)					
	[Clearing conditions]					
	1. When 0 is written by software (in cases other than setting condition 3)					
	2. When 0 is written in TRS after reading TRS = 1 (in case of clearing condition 3)					
	 When bus arbitration is lost after transmission is started in I²C bus format master mode 					
	4. When the SW bit in DDCSWR changes from 1 to 0					
1	Transmit mode					
	[Setting conditions]					
	1. When 1 is written by software (in cases other than clearing conditions 3 and 4)					
	 When 1 is written in TRS after reading TRS = 0 (in case of clearing conditions 3 and 4) 					
	3. When a 1 is received as the R/\overline{W} bit of the first frame in I^2C bus format slave mode					

Bit 3—Acknowledge Bit Judgement Selection (ACKE): Specifies whether the value of the acknowledge bit returned from the receiving device when using the I²C bus format is to be ignored and continuous transfer is performed, or transfer is to be aborted and error handling, etc., performed if the acknowledge bit is 1. When the ACKE bit is 0, the value of the received acknowledge bit is not indicated by the ACKB bit, which is always 0.

In the H8S/2148 Group and H8S/2147N, the DTC can be used to perform continuous transfer. The DTC is activated when the IRTR interrupt flag is set to 1 (IRTR is one of two interrupt flags, the other being IRIC). When the ACKE bit is 0, the TDRE, IRIC, and IRTR flags are set on completion of data transmission, regardless of the value of the acknowledge bit. When the ACKE bit is 1, the TDRE, IRIC, and IRTR flags are set on completion of data transmission when the acknowledge bit is 0, and the IRIC flag alone is set on completion of data transmission when the acknowledge bit is 1.

When the DTC is activated, the TDRE, IRIC, and IRTR flags are cleared to 0 after the specified number of data transfers have been executed. Consequently, interrupts are not generated during continuous data transfer, but if data transmission is completed with a 1 acknowledge bit when the ACKE bit is set to 1, the DTC is not activated and an interrupt is generated, if enabled.

Depending on the receiving device, the acknowledge bit may be significant, in indicating completion of processing of the received data, for instance, or may be fixed at 1 and have no significance.

Bit 3

ACKE	Description	
0	The value of the acknowledge bit is ignored, and continuous transfer is performed	(Initial value)
1	If the acknowledge bit is 1, continuous transfer is interrupted	

Bit 2—Bus Busy (BBSY): The BBSY flag can be read to check whether the I²C bus (SCL, SDA) is busy or free. In master mode, this bit is also used to issue start and stop conditions.

A high-to-low transition of SDA while SCL is high is recognized as a start condition, setting BBSY to 1. A low-to-high transition of SDA while SCL is high is recognized as a stop condition, clearing BBSY to 0.

To issue a start condition, use a MOV instruction to write 1 in BBSY and 0 in SCP. A retransmit start condition is issued in the same way. To issue a stop condition, use a MOV instruction to write 0 in BBSY and 0 in SCP. It is not possible to write to BBSY in slave mode; the I²C bus interface must be set to master transmit mode before issuing a start condition. MST and TRS should both be set to 1 before writing 1 in BBSY and 0 in SCP.

Bit 2

BBSY	Description	
0	Bus is free	(Initial value)
	[Clearing condition]	
	When a stop condition is detected	
1	Bus is busy	
	[Setting condition]	
	When a start condition is detected	

Bit 1—I²C Bus Interface Interrupt Request Flag (IRIC): Indicates that the I²C bus interface has issued an interrupt request to the CPU. IRIC is set to 1 at the end of a data transfer, when a slave address or general call address is detected in slave receive mode, when bus arbitration is lost in master transmit mode, and when a stop condition is detected. IRIC is set at different times depending on the FS bit in SAR and the WAIT bit in ICMR. See section 16.3.6, IRIC Setting Timing and SCL Control. The conditions under which IRIC is set also differ depending on the setting of the ACKE bit in ICCR.

IRIC is cleared by reading IRIC after it has been set to 1, then writing 0 in IRIC.

When the DTC is used, IRIC is cleared automatically and transfer can be performed continuously without CPU intervention.

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Bit 1	
IRIC	Description
0	Waiting for transfer, or transfer in progress (Initial value)
	[Clearing conditions]
	1. When 0 is written in IRIC after reading IRIC = 1
	2. When ICDR is written or read by the DTC
	(When the TDRE or RDRF flag is cleared to 0)
	(This is not always a clearing condition; see the description of DTC operation for details)
1	Interrupt requested
	[Cotting conditional

[Setting conditions]

- I²C bus format master mode
 - When a start condition is detected in the bus line state after a start condition is issued (when the TDRE flag is set to 1 because of first frame transmission)
 - 2. When a wait is inserted between the data and acknowledge bit when WAIT = 1
 - At the end of data transfer
 (at the rise of the 9th transmit/receive clock pulse when no wait is inserted, (WAIT=0) and, when a wait is inserted (WAIT=1), at the fall of the 8th transmit/receive clock pulse)
 - 4. When a slave address is received after bus arbitration is lost (when the AL flag is set to 1)
 - When 1 is received as the acknowledge bit when the ACKE bit is 1 (when the ACKB bit is set to 1)
- I²C bus format slave mode
 - When the slave address (SVA, SVAX) matches (when the AAS and AASX flags are set to 1) and at the end of data transfer up to the subsequent retransmission start condition or stop condition detection (when the TDRE or RDRF flag is set to 1)
 - When the general call address is detected (when FS = 0 and the ADZ flag is set to 1) and at the end of data transfer up to the subsequent retransmission start condition or stop condition detection (when the TDRE or RDRF flag is set to 1)
 - When 1 is received as the acknowledge bit when the ACKE bit is 1 (when the ACKB bit is set to 1)
 - 4. When a stop condition is detected (when the STOP or ESTP flag is set to 1)
- · Synchronous serial format, and formatless mode
 - At the end of data transfer (when the TDRE or RDRF flag is set to 1)
 - 2. When a start condition is detected with serial format selected
 - When the SW bit is set to 1 in DDCSWR

Except the above, when the conditions to set the TDRE or RDRF internal flag to 1 is generated

When, with the I²C bus format selected, IRIC is set to 1 and an interrupt is generated, other flags must be checked in order to identify the source that set IRIC to 1. Although each source has a corresponding flag, caution is needed at the end of a transfer.

When the TDRE or RDRF internal flag is set, the readable IRTR flag may or may not be set. The IRTR flag (the DTC start request flag) is not set at the end of a data transfer up to detection of a retransmission start condition or stop condition after a slave address (SVA) or general call address match in I²C bus format slave mode.

Even when the IRIC flag and IRTR flag are set, the TDRE or RDRF internal flag may not be set. The IRIC and IRTR flags are not cleared at the end of the specified number of transfers in continuous transfer using the DTC. The TDRE or RDRF flag is cleared, however, since the specified number of ICDR reads or writes have been completed.

Table 16.3 shows the relationship between the flags and the transfer states.

Table 16.3 Flags and Transfer States

MST	TRS	BBSY	ESTP	STOP	IRTR	AASX	AL	AAS	ADZ	ACKB	State
1/0	1/0	0	0	0	0	0	0	0	0	0	Idle state (flag clearing required)
1	1	0	0	0	0	0	0	0	0	0	Start condition issuance
1	1	1	0	0	1	0	0	0	0	0	Start condition established
1	1/0	1	0	0	0	0	0	0	0	0/1	Master mode wait
1	1/0	1	0	0	1	0	0	0	0	0/1	Master mode transmit/ receive end
0	0	1	0	0	0	1/0	1	1/0	1/0	0	Arbitration lost
0	0	1	0	0	0	0	0	1	0	0	SAR match by first frame in slave mode
0	0	1	0	0	0	0	0	1	1	0	General call address match
0	0	1	0	0	0	1	0	0	0	0	SARX match
0	1/0	1	0	0	0	0	0	0	0	0/1	Slave mode transmit/receive end (except after SARX match)
0	1/0	1	0	0	1	1	0	0	0	0	Slave mode
0	1	1	0	0	0	1	0	0	0	1	transmit/receive end (after SARX match)
0	1/0	0	1/0	1/0	0	0	0	0	0	0/1	Stop condition detected

Bit 0—Start Condition/Stop Condition Prohibit (SCP): Controls the issuing of start and stop conditions in master mode. To issue a start condition, write 1 in BBSY and 0 in SCP. A retransmit start condition is issued in the same way. To issue a stop condition, write 0 in BBSY and 0 in SCP. This bit is always read as 1. If 1 is written, the data is not stored.

Bit 0

SCP	Description	
0	Writing 0 issues a start or stop condition, in combination	on with the BBSY flag
1	Reading always returns a value of 1	(Initial value)
	Writing is ignored	

16.2.6 I²C Bus Status Register (ICSR)

Bit	7	6	5	4	3	2	1	0
	ESTP	STOP	IRTR	AASX	AL	AAS	ADZ	ACKB
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/(W)*	R/W						

Note: * Only 0 can be written, to clear the flags.

ICSR is an 8-bit readable/writable register that performs flag confirmation and acknowledge confirmation and control.

ICSR is initialized to H'00 by a reset and in hardware standby mode.

Bit 7—Error Stop Condition Detection Flag (ESTP): Indicates that a stop condition has been detected during frame transfer in I²C bus format slave mode.

Bit 7

ESTP	Description	
0	No error stop condition	(Initial value)
	[Clearing conditions]	
	1. When 0 is written in ESTP after reading ESTP = 1	
	2. When the IRIC flag is cleared to 0	
1	In I ² C bus format slave mode	
	Error stop condition detected	
	[Setting condition]	
	When a stop condition is detected during frame transfer	
	 In other modes 	
	No meaning	

Bit 6—Normal Stop Condition Detection Flag (STOP): Indicates that a stop condition has been detected after completion of frame transfer in I²C bus format slave mode.

Bit 6

Description
No normal stop condition (Initial value)
[Clearing conditions]
1. When 0 is written in STOP after reading STOP = 1
2. When the IRIC flag is cleared to 0
In I ² C bus format slave mode
Normal stop condition detected
[Setting condition]
When a stop condition is detected after completion of frame transfer
In other modes
No meaning

Bit 5—I²C Bus Interface Continuous Transmission/Reception Interrupt Request Flag (IRTR): Indicates that the I²C bus interface has issued an interrupt request to the CPU, and the source is completion of reception/transmission of one frame in continuous transmission/reception for which DTC activation is possible. When the IRTR flag is set to 1, the IRIC flag is also set to 1 at the same time.

IRTR flag setting is performed when the TDRE or RDRF flag is set to 1. IRTR is cleared by reading IRTR after it has been set to 1, then writing 0 in IRTR. IRTR is also cleared automatically when the IRIC flag is cleared to 0.

Bit 5

IRTR	Description					
0	Waiting for transfer, or transfer in progress	(Initial value)				
	[Clearing conditions]					
	1. When 0 is written in IRTR after reading IRTR = 1					
	2. When the IRIC flag is cleared to 0					
1	Continuous transfer state					
	[Setting conditions]					
	 In I²C bus interface slave mode 					
	When the TDRE or RDRF flag is set to 1 when AASX = 1					
	In other modes					
	When the TDRE or RDRF flag is set to 1					

Bit 4—Second Slave Address Recognition Flag (AASX): In I²C bus format slave receive mode, this flag is set to 1 if the first frame following a start condition matches bits SVAX6 to SVAX0 in SARX.

AASX is cleared by reading AASX after it has been set to 1, then writing 0 in AASX. AASX is also cleared automatically when a start condition is detected.

Bit 4

AASX	Description	
0	Second slave address not recognized (Ir	nitial value)
	[Clearing conditions]	
	1. When 0 is written in AASX after reading AASX = 1	
	2. When a start condition is detected	
	3. In master mode	
1	Second slave address recognized	
	[Setting condition]	
	When the second slave address is detected in slave receive mode while F	SX = 0

Bit 3—Arbitration Lost (AL): This flag indicates that arbitration was lost in master mode. The I^2C bus interface monitors the bus. When two or more master devices attempt to seize the bus at nearly the same time, if the I^2C bus interface detects data differing from the data it sent, it sets AL to 1 to indicate that the bus has been taken by another master.

AL is cleared by reading AL after it has been set to 1, then writing 0 in AL. In addition, AL is reset automatically by write access to ICDR in transmit mode, or read access to ICDR in receive mode.

Bit 3

AL	Description					
0	Bus arbitration won (Initial value)					
	[Clearing conditions]					
	1. When ICDR data is written (transmit mode) or read (receive mode)					
	2. When 0 is written in AL after reading AL = 1					
1	Arbitration lost					
	[Setting conditions]					
	 If the internal SDA and SDA pin disagree at the rise of SCL in master transmit mode 					
	2. If the internal SCL line is high at the fall of SCL in master transmit mode					

Bit 2—Slave Address Recognition Flag (AAS): In I²C bus format slave receive mode, this flag is set to 1 if the first frame following a start condition matches bits SVA6 to SVA0 in SAR, or if the general call address (H'00) is detected.

AAS is cleared by reading AAS after it has been set to 1, then writing 0 in AAS. In addition, AAS is reset automatically by write access to ICDR in transmit mode, or read access to ICDR in receive mode.

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AAS	 Description	
0	Slave address or general call address not recognized	(Initial value)
	[Clearing conditions]	
	1. When ICDR data is written (transmit mode) or read (receive mode)	
	2. When 0 is written in AAS after reading AAS = 1	
	3. In master mode	
1	Slave address or general call address recognized	
	[Setting condition]	
	When the slave address or general call address is detected in slave rece $FS = 0$	ive mode while

Bit 1—General Call Address Recognition Flag (ADZ): In I²C bus format slave receive mode, this flag is set to 1 if the first frame following a start condition is the general call address (H'00).

ADZ is cleared by reading ADZ after it has been set to 1, then writing 0 in ADZ. In addition, ADZ is reset automatically by write access to ICDR in transmit mode, or read access to ICDR in receive mode.

Bit 1

ADZ	Description	
0	General call address not recognized	(Initial value)
	[Clearing conditions]	
	1. When ICDR data is written (transmit mode) or read (receive mode)	
	2. When 0 is written in ADZ after reading ADZ = 1	
	3. In master mode	
1	General call address recognized	
	[Setting condition]	
	When the general call address is detected in slave receive mode while F	SX = 0 or FS = 0

Bit 0—Acknowledge Bit (ACKB): Stores acknowledge data. In transmit mode, after the receiving device receives data, it returns acknowledge data, and this data is loaded into ACKB. In receive mode, after data has been received, the acknowledge data set in this bit is sent to the transmitting device.

When this bit is read, in transmission (when TRS = 1), the value loaded from the bus line (returned by the receiving device) is read. In reception (when TRS = 0), the value set by internal software is read.

When this bit is written to, the acknowledge data transmitted at the receipt is rewritten regardless of the TRS value. The data loaded frm the receiving device is retained, therefore take care of using bit-manipulation instructions.

Bit 0

ACKB	Description	
0	Receive mode: 0 is output at acknowledge output timing	(Initial value)
	Transmit mode: Indicates that the receiving device has acknowledge is 0)	ed the data (signal
1	Receive mode: 1 is output at acknowledge output timing	
	Transmit mode: Indicates that the receiving device has not acknowled (signal is 1)	edged the data

16.2.7 Serial/Timer Control Register (STCR)

Bit	7	6	5	4	3	2	1	0
	IICS	IICX1	IICX0	IICE	FLSHE	_	ICKS1	ICKS0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

STCR is an 8-bit readable/writable register that controls register access, the I²C interface operating mode (when the on-chip IIC option is included), and on-chip flash memory (F-ZTAT versions), and selects the TCNT input clock source. For details of functions not related to the I²C bus interface, see section 3.2.4, Serial Timer Control Register (STCR), and the descriptions of the relevant modules. If a module controlled by STCR is not used, do not write 1 to the corresponding bit.

STCR is initialized to H'00 by a reset and in hardware standby mode.



Bit 7—I²C **Extra Buffer Select (IICS):** Designates bits 7 to 4 of port A as the same kind of output buffer as SCL and SDA. This bit is used when implementing the I²C interface by software only.

Bit 7

IICS	Description	
0	PA7 to PA4 are normal I/O pins	(Initial value)
1	PA7 to PA4 are I/O pins with bus driving capability	

Bits 6 and 5—I²C **Transfer Select 1 and 0 (IICX1 and 0):** This bit, together with bits CKS2 to CKS0 in ICMR, selects the transfer rate in master mode. For details, see section 16.2.4, I²C Bus Mode Register (ICMR).

Bit 4—I²C **Master Enable (IICE):** Controls CPU access to the I²C bus interface data and control registers (ICCR, ICSR, ICDR/SARX, ICMR/SAR).

Bit 4

IICE	Description	
0	CPU access to I ² C bus interface data and control registers is disabled	(Initial value)
1	CPU access to I ² C bus interface data and control registers is enabled	

Bit 3—Flash Memory Control Register Enable (FLSHE): Controls CPU access to the flash memory control registers, the power-down mode control registers, and the supporting module control registers. See section 3.2.4, Serial Timer Control Register (STCR), for details.

Bit 2—Reserved: Do not write 1 to this bit.

Bits 1 and 0—Internal Clock Source Select 1 and 0 (ICKS1, ICSK0): These bits, together with bits CKS2 to CKS0 in TCR, select the clock input to the timer counters (TCNT). For details, see section 12.2.4, Timer Control Register (TCR).

16.2.8 DDC Switch Register (DDCSWR)

Bit	7	6	5	4	3	2	1	0
	SWE	SW	IE	IF	CLR3	CLR2	CLR1	CLR0
Initial value	0	0	0	0	1	1	1	1
Read/Write	R/W	R/W	R/W	R/(W)*1	W^{*2}	W^{*2}	W*2	W^{*2}

Notes: 1. Only 0 can be written, to clear the flag.

2. Always read as 1.

DDCSWR is an 8-bit readable/writable register that is used to initialize IIC and controls IIC internal latch clearance.

DDCSWR is initialized to H'0F by a reset and in hardware standby mode.

Bit 7—DDC Mode Switch Enable (SWE): Selects the function for automatically switching IIC channel 0 from formatless mode to the I²C bus format.

Bit 7

SWE	Description	
0	Automatic switching of IIC channel 0 from formatless mode to I ² C bus format is disabled (Initial value)	
1	Automatic switching of IIC channel 0 from formatless mode to I ² C bus format is enabled	

Bit 6—DDC Mode Switch (SW): Selects either formatless mode or the I²C bus format for IIC channel 0.

Bit 6

SW	Description	
0	IIC channel 0 is used with the I ² C bus format	(Initial value)
	[Clearing conditions]	
	1. When 0 is written by software	
	2. When a falling edge is detected on the SCL pin when SWE = 1	
1	IIC channel 0 is used in formatless mode	
	[Setting condition]	
	When 1 is written in SW after reading SW = 0	

Bit 5—DDC Mode Switch Interrupt Enable Bit (IE): Enables or disables an interrupt request to the CPU when automatic format switching is executed for IIC channel 0.

Bit 5

IE	Description	
0	Interrupt when automatic format switching is executed is disabled	(Initial value)
1	Interrupt when automatic format switching is executed is enabled	

Bit 4—DDC Mode Switch Interrupt Flag (IF): Flag that indicates an interrupt request to the CPU when automatic format switching is executed for IIC channel 0.

Bit 4

IF	Description	
0	No interrupt is requested when automatic format switching is executed	(Initial value)
	[Clearing condition]	
	When 0 is written in IF after reading IF = 1	
1	An interrupt is requested when automatic format switching is executed	
	[setting condition]	
	When a falling edge is detected on the SCL pin when SWE = 1	

Bits 3 to 0—IIC Clear 3 to 0 (CLR3 to CLR0): These bits control initialization of the internal state of IIC0 and IIC1.

These bits can only be written to; if read they will always return a value of 1.

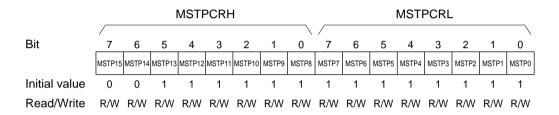
When a write operation is performed on these bits, a clear signal is generated for the internal latch circuit of the corresponding module(s), and the internal state of the IIC module(s) is initialized.

The write data for these bits is not retained. To perform IIC clearance, bits CLR3 to CLR0 must be written to simultaneously using an MOV instruction. Do not use a bit-manipulation instruction such as BCLR.

When clearing is required again, all the bits must be written to in accordance with the setting.

Bit 3	Bit 2	Bit 1	Bit 0	
CLR3	CLR2	CLR1	CLR0	 Description
0	0	_	_	Setting prohibited
	1	0	0	Setting prohibited
			1	IIC0 internal latch cleared
		1	0	IIC1 internal latch cleared
			1	IIC0 and IIC1 internal latches cleared
1	_	_	_	Invalid setting

16.2.9 Module Stop Control Register (MSTPCR)



MSTPCR comprises two 8-bit readable/writable registers, and is used to perform module stop mode control.

When the MSTP4 or MSTP3 bit is set to 1, operation of the corresponding IIC channel is halted at the end of the bus cycle, and a transition is made to module stop mode. For details, see section 25.5, Module Stop Mode.

MSTPCR is initialized to H'3FFF by a reset and in hardware standby mode. It is not initialized in software standby mode.

MSTPCRL Bit 4—Module Stop (MSTP4): Specifies IIC channel 0 module stop mode.

MSTPCRL Bit 4

MSTP4	Description	
0	IIC channel 0 module stop mode is cleared	
1	IIC channel 0 module stop mode is set	(Initial value)



MSTPCRL Bit 3—Module Stop (MSTP3): Specifies IIC channel 1 module stop mode.

MSTPCRL Bit 3

MSTP3	 Description	
0	IIC channel 1 module stop mode is cleared	
1	IIC channel 1 module stop mode is set	(Initial value)

16.3 Operation

16.3.1 I²C Bus Data Format

The I²C bus interface has serial and I²C bus formats.

The 1²C bus formats are addressing formats with an acknowledge bit. These are shown in figures 16.3 (a) and (b). The first frame following a start condition always consists of 8 bits.

IIC channel 0 only is capable of formatless operation, as shown in figure 16.4.

The serial format is a non-addressing format with no acknowledge bit. Although start and stop conditions must be issued, this format can be used as a synchronous serial format. This is shown in figure 16.5.

Figure 16.6 shows the I²C bus timing.

The symbols used in figures 16.3 to 16.6 are explained in table 16.4.

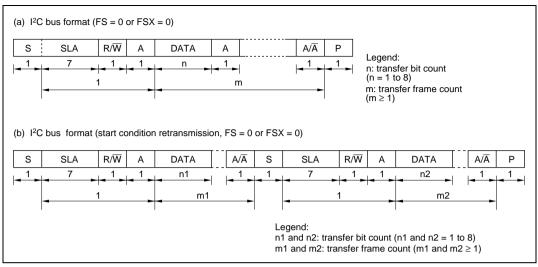


Figure 16.3 I²C Bus Data Formats (I²C Bus Formats)

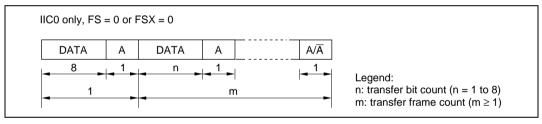


Figure 16.4 Formatless

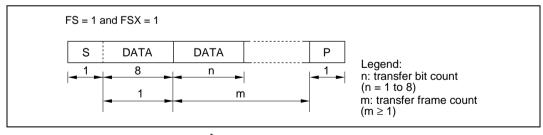


Figure 16.5 I²C Bus Data Format (Serial Format)

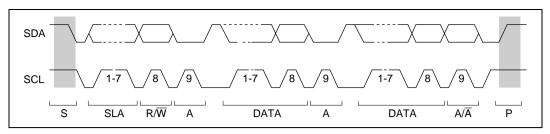


Figure 16.6 I²C Bus Timing

Table 16.4 I²C Bus Data Format Symbols

Legend	
S	Start condition. The master device drives SDA from high to low while SCL is high
SLA	Slave address, by which the master device selects a slave device
R/W	Indicates the direction of data transfer: from the slave device to the master device when R/\overline{W} is 1, or from the master device to the slave device when R/\overline{W} is 0
A	Acknowledge. The receiving device (the slave in master transmit mode, or the master in master receive mode) drives SDA low to acknowledge a transfer
DATA	Transferred data. The bit length is set by bits BC2 to BC0 in ICMR. The MSB-first or LSB-first format is selected by bit MLS in ICMR
Р	Stop condition. The master device drives SDA from low to high while SCL is high

16.3.2 Master Transmit Operation

In I²C bus format master transmit mode, the master device outputs the transmit clock and transmit data, and the slave device returns an acknowledge signal.

The transmission procedure and operations by which data is sequentially transmitted in synchronization with ICDR write operations, are described below.

- (1) Set the ICE bit in ICCR to l. Set bits MLS, WAIT, and CKS2 to CKS0 in ICMR, and bit IICX in STCR, according to the operation mode.
- (2) Read the BBSY flag to confirm that the bus is free.
- (3) Set the MST and TRS bits to 1 in ICCR to select master transmit mode.
- (4) Write 1 to BBSY and 0 to SCP. This switches SDA from high to low when SCL is high, and generates the start condition.
- (5) When the start condition is generated, the IRIC and IRTR flags are set to 1. If the IEIC bit in ICCR has been set to 1, an interrupt request is sent to the CPU.

(6) Write data to ICDR (slave address + R/\overline{W})

With the I²C bus format (when the FS bit in SAR or the FSX bit in SARX is 0), the first frame data following the start condition indicates the 7-bit slave address and transmit/receive direction.

Then clear the IRIC flag to indicate the end of transfer.

Writing to ICDR and clearing of the IRIC flag must be executed continuously, so that no interrupt is inserted.

If a period of time that is equal to transfer one byte has elapsed by the time the IRIC flag is cleared, the end of transfer cannot be identified.

The master device sequentially sends the transmit clock and the data written to ICDR with the timing shown in figure 16.7. The selected slave device (i.e., the slave device with the matching slave address) drives SDA low at the 9th transmit clock pulse and returns an acknowledge signal.

- (7) When one frame of data has been transmitted, the IRIC flag is set to 1 at the rise of the 9th transmit clock pulse. After one frame has been transmitted, SCL is automatically fixed low in synchronization with the internal clock until the next transmit data is written.
- (8) Read the ACKB bit to confirm that ACKB is 0. When the slave device has not returned an acknowledge signal and ACKB remains 1, execute the transmit end processing described in step (12) and perform transmit operation again.
- (9) Write the next data to be transmitted in ICDR. To indicate the end of data transfer, clear the IRIC flag to 0.
 - As described in step (6) above, writing to ICDR and clearing of the IRIC flag must be executed continuously so that no interrupt is inserted.

The next frame is transmitted in synchronization with the internal clock.

- (10) When one frame of data has been transmitted, the IRIC flag is set to 1 at the rise of the 9th transmit clock pulse. After one frame has been transmitted, SCL is automatically fixed low in synchronization with the internal clock until the next transmit data is written.
- (11) Read the ACKB bit of ICSR. Confirm that the slave device has returned an acknowledge signal and ACKB is 0. When more data is to be transmitted, return to step (9) to execute next transmit operation. If the slave device has not returned an acknowledge signal and ACKB is 1, execute the transmit end processing described in step (12).
- (12) Clear the IRIC flag to 0. Write BBSY and SCP of ICCR to 0. By doing so, SDA is changed from low to high while SCL is high and the transmit stop condition is generated.



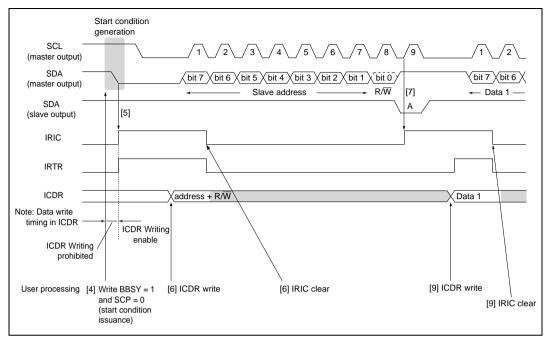


Figure 16.7 Example of Master Transmit Mode Operation Timing (MLS = WAIT = 0)

16.3.3 Master Receive Operation

In master receive mode, the master device outputs the receive clock, receives data, and returns an acknowledge signal. The slave device transmits data.

The receive procedure and operations by which data is sequentially received in synchronization with ICDR read operations, are described below.

- (1) Clear the TRS bit of ICCR to 0 and switch from transmit mode to receive mode. Set the WAIT bit to 1 and clear the ACKB bit of ICSR to 0 (acknowledge data setting).
- (2) When ICDR is read (dummy data read), reception is started and the receive clock is output, and data is received, in synchronization with the internal clock. To indicate the wait, clear the IRIC flag to 0.

Reading from ICDR and clearing of the IRIC flag must be executed continuously so that no interrupt is inserted.

If a period of time that is equal to transfer one byte has elapsed by the time the IRIC flag is cleared, the end of transfer cannot be identified.

- (3) The IRIC flag is set to 1 at the fall of the 8th clock of a one-frame reception clock. At this point, if the IEIC bit of ICCR is set to 1, an interrupt request is generated to the CPU. SCL is automatically fixed low in synchronization with the internal clock until the IRIC flag is cleared. If the first frame is the final reception frame, execute the end processing as described in (10).
- (4) Clear the IRIC flag to 0 to release from the wait state.

 The master device outputs the 9th receive clock pulse, sets SDA to low, and returns an
- acknowledge signal.
- (5) When one frame of data has been transmitted, the IRIC and IRTR flags are set to 1 at the rise of the 9th transmit clock pulse.
 - The master device continues to output the receive clock for the next receive data.
- (6) Read the ICDR receive data.
- (7) Clear the IRIC flag to indicate the next wait.
 - From clearing of the IRIC flag to completion of data transmission as described in steps (5), (6), and (7), must be performed within the time taken to transfer one byte, because releasing of the wait state as described in step (4) (or (9)).
- (8) The IRIC flag is set to 1 at the fall of the 8th one-frame reception clock pulse. SCL is automatically fixed low in synchronization with the internal clock until the IRIC flag is cleared. If this frame is the final reception frame, execute the end processing as described in (10).
- (9) Clear the IRIC flag to 0 to release from the wait state. The master device outputs the 9th reception clock pulse, sets SDA to low, and returns an acknowledge signal. By repeating steps (5) to (9) above, more data can be received.
- (10) Set the ACKB bit of ICSR to 1 and set the acknowledge data for the final reception. Set the TRS bit of ICCR to 1 to change receive mode to transmit mode.
- (11) Clear the IRIC flag to release from the wait state.
- (12) When one frame of data has been received, the IRIC flag is set to 1 at the rise of the 9th reception clock pulse.
- (13) Clear the WAIT bit of ICMR to 0 to cancel wait mode. Read the ICDR receive data and clear the IRIC flag to 0.
 - Clear the IRIC flag only when WAIT = 0.
 - (If the stop-condition generation command is executed after clearing the IRIC flag to 0 and then clearing the WAIT bit to 0, the SDA line is fixed low and the stop condition cannot be generated.)
- (14) Write 0 to BBSY and SCP. This changes SDA from low to high when SCL is high, and generates the stop condition.



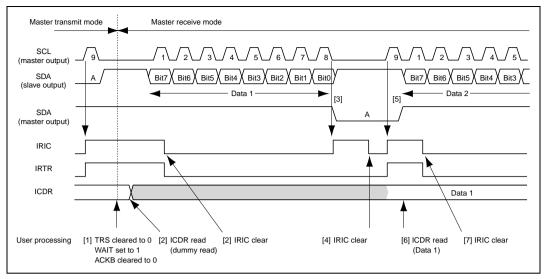


Figure 16.8 (a) Example of Master Receive Mode Operation Timing (MLS = ACKB = 0, WAIT = 1)

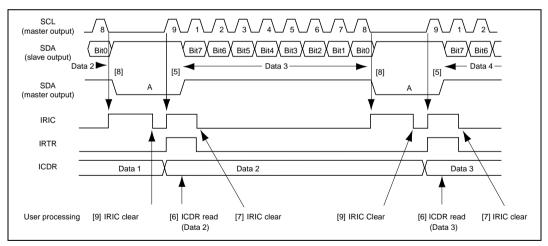


Figure 16.8 (b) Example of Master Receive Mode Operation Timing (MLS = ACKB = 0, WAIT = 1)

16.3.4 Slave Receive Operation

In slave receive mode, the master device outputs the transmit clock and transmit data, and the slave device returns an acknowledge signal. The reception procedure and operations in slave receive mode are described below.

- [1] Set the ICE bit in ICCR to 1. Set the MLS bit in ICMR and the MST and TRS bits in ICCR according to the operating mode.
- [2] When the start condition output by the master device is detected, the BBSY flag in ICCR is set to 1.
- [3] When the slave address matches in the first frame following the start condition, the device operates as the slave device specified by the master device. If the 8th data bit (R/W) is 0, the TRS bit in ICCR remains cleared to 0, and slave receive operation is performed.
- [4] At the 9th clock pulse of the receive frame, the slave device drives SDA low and returns an acknowledge signal. At the same time, the IRIC flag in ICCR is set to 1. If the IEIC bit in ICCR has been set to 1, an interrupt request is sent to the CPU. If the RDRF internal flag has been cleared to 0, it is set to 1, and the receive operation continues. If the RDRF internal flag has been set to 1, the slave device drives SCL low from the fall of the receive clock until data is read into ICDR.
- [5] Read ICDR and clear the IRIC flag in ICCR to 0. The RDRF flag is cleared to 0.

Receive operations can be performed continuously by repeating steps [4] and [5]. When SDA is changed from low to high when SCL is high, and the stop condition is detected, the BBSY flag in ICCR is cleared to 0.



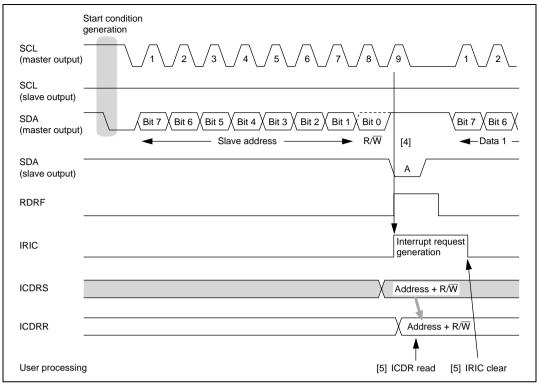


Figure 16.9 Example of Slave Receive Mode Operation Timing (1) (MLS = ACKB = 0)

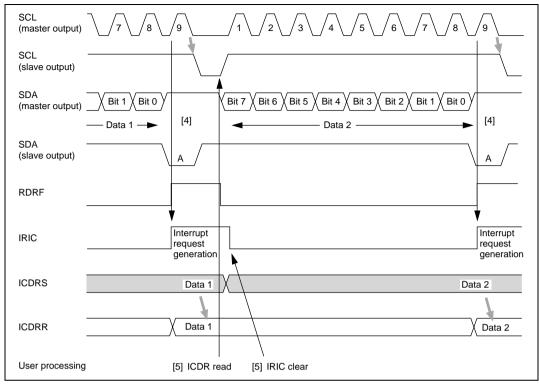


Figure 16.10 Example of Slave Receive Mode Operation Timing (2) (MLS = ACKB = 0)

16.3.5 Slave Transmit Operation

In slave transmit mode, the slave device outputs the transmit data, while the master device outputs the receive clock and returns an acknowledge signal. The transmission procedure and operations in slave transmit mode are described below.

- [1] Set the ICE bit in ICCR to 1. Set the MLS bit in ICMR and the MST and TRS bits in ICCR according to the operating mode.
- [2] When the slave address matches in the first frame following detection of the start condition, the slave device drives SDA low at the 9th clock pulse and returns an acknowledge signal. At the same time, the IRIC flag in ICCR is set to 1. If the IEIC bit in ICCR has been set to 1, an interrupt request is sent to the CPU. If the 8th data bit (R/W) is 1, the TRS bit in ICCR is set to 1, and the mode changes to slave transmit mode automatically. The TDRE internal flag is set to 1. The slave device drives SCL low from the fall of the transmit clock until ICDR data is written.
- [3] After clearing the IRIC flag to 0, write data to ICDR. The TDRE internal flag is cleared to 0. The written data is transferred to ICDRS, and the TDRE internal flag and the IRIC and IRTR flags are set to 1 again. After clearing the IRIC flag to 0, write the next data to ICDR. The slave device sequentially sends the data written into ICDR in accordance with the clock output by the master device at the timing shown in figure 16.11.
- [4] When one frame of data has been transmitted, the IRIC flag in ICCR is set to 1 at the rise of the 9th transmit clock pulse. If the TDRE internal flag has been set to 1, this slave device drives SCL low from the fall of the transmit clock until data is written to ICDR. The master device drives SDA low at the 9th clock pulse, and returns an acknowledge signal. As this acknowledge signal is stored in the ACKB bit in ICSR, this bit can be used to determine whether the transfer operation was performed normally. When the TDRE internal flag is 0, the data written into ICDR is transferred to ICDRS, transmission is started, and the TDRE internal flag and the IRIC and IRTR flags are set to 1 again.
- [5] To continue transmission, clear the IRIC flag to 0, then write the next data to be transmitted into ICDR. The TDRE internal flag is cleared to 0.

Transmit operations can be performed continuously by repeating steps [4] and [5]. To end transmission, write H'FF to ICDR to release SDA on the slave side. When SDA is changed from low to high when SCL is high, and the stop condition is detected, the BBSY flag in ICCR is cleared to 0.

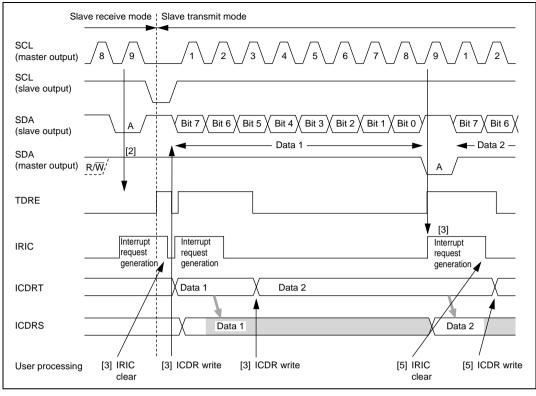


Figure 16.11 Example of Slave Transmit Mode Operation Timing (MLS = 0)

16.3.6 IRIC Setting Timing and SCL Control

The interrupt request flag (IRIC) is set at different times depending on the WAIT bit in ICMR, the FS bit in SAR, and the FSX bit in SARX. If the TDRE or RDRF internal flag is set to 1, SCL is automatically held low after one frame has been transferred; this timing is synchronized with the internal clock. Figure 16.12 shows the IRIC set timing and SCL control.

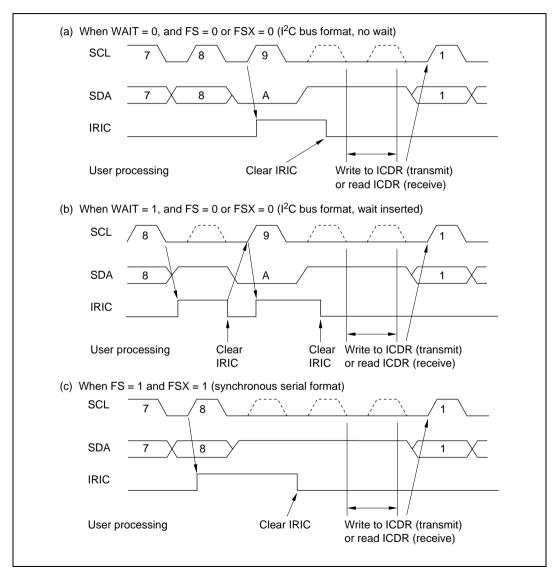


Figure 16.12 IRIC Setting Timing and SCL Control

16.3.7 Automatic Switching from Formatless Mode to I²C Bus Format

Setting the SW bit to 1 in DDCSWR enables formatless mode to be selected as the IIC0 operating mode. Switching from formatless mode to the I²C bus format (slave mode) is performed automatically when a falling edge is detected on the SCL pin.

The following four preconditions are necessary for this operation:

- A common data pin (SDA) for formatless and I²C bus format operation
- Separate clock pins for formatless operation (VSYNCI) and I²C bus format operation (SCL)
- A fixed 1 level for the SCL pin during formatless operation (the SCL pin does not output a low level)
- Settings of bits other than TRS in ICCR that allow I²C bus format operation

Automatic switching is performed from formatless mode to the I²C bus format when the SW bit in DDCSWR is automatically cleared to 0 on detection of a falling edge on the SCL pin. Switching from the I²C bus format to formatless mode is achieved by having software set the SW bit in DDCSWR to 1.

In formatless mode, bits (such as MSL and TRS) that control the I²C bus interface operating mode must not be modified. When switching from the I²C bus format to formatless mode, set the TRS bit to 1 or clear it to 0 according to the transmit data (transmission or reception) in formatless mode, then set the SW bit to 1. After automatic switching from formatless mode to the I²C bus format (slave mode), in order to wait for slave address reception, the TRS bit is automatically cleared to 0.

If a falling edge is detected on the SCL pin during formatless operation, the I²C bus interface operating mode is switched to the I²C bus format without waiting for a stop condition to be detected.



16.3.8 Operation Using the DTC

The I^2C bus format provides for selection of the slave device and transfer direction by means of the slave address and the R/\overline{W} bit, confirmation of reception with the acknowledge bit, indication of the last frame, and so on. Therefore, continuous data transfer using the DTC must be carried out in conjunction with CPU processing by means of interrupts.

Table 16.5 shows some examples of processing using the DTC. These examples assume that the number of transfer data bytes is known in slave mode.

Table 16.5 Examples of Operation Using the DTC

Item	Master Transmit Mode	Master Receive Mode	Slave Transmit Mode	Slave Receive Mode
Slave address + R/W bit transmission/ reception	Transmission by DTC (ICDR write)	Transmission by CPU (ICDR write)	Reception by CPU (ICDR read)	Reception by CPU (ICDR read)
Dummy data read	_	Processing by CPU (ICDR read)	_	_
Actual data transmission/ reception	Transmission by DTC (ICDR write)	Reception by DTC (ICDR read)	Transmission by DTC (ICDR write)	Reception by DTC (ICDR read)
Dummy data (H'FF) write	_	_	Processing by DTC (ICDR write)	_
Last frame processing	Not necessary	Reception by CPU (ICDR read)	Not necessary	Reception by CPU (ICDR read)
Transfer request processing after	1st time: Clearing by CPU	Not necessary	Automatic clearing on detection of end	Not necessary
last frame processing	2nd time: End condition issuance by CPU		condition during transmission of dummy data (H'FF)	
Setting of number of DTC transfer data frames	Transmission: Actual data count + 1 (+1 equivalent to slave address + R/W bits)	Reception: Actual data count	Transmission: Actual data count + 1 (+1 equivalent to dummy data (H'FF))	Reception: Actual data count

16.3.9 Noise Canceler

The logic levels at the SCL and SDA pins are routed through noise cancelers before being latched internally. Figure 16.13 shows a block diagram of the noise canceler circuit.

The noise canceler consists of two cascaded latches and a match detector. The SCL (or SDA) input signal is sampled on the system clock, but is not passed forward to the next circuit unless the outputs of both latches agree. If they do not agree, the previous value is held.

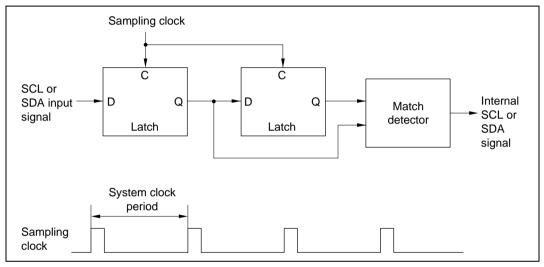


Figure 16.13 Block Diagram of Noise Canceler

16.3.10 Sample Flowcharts

Figures 16.14 to 16.17 show sample flowcharts for using the I²C bus interface in each mode.

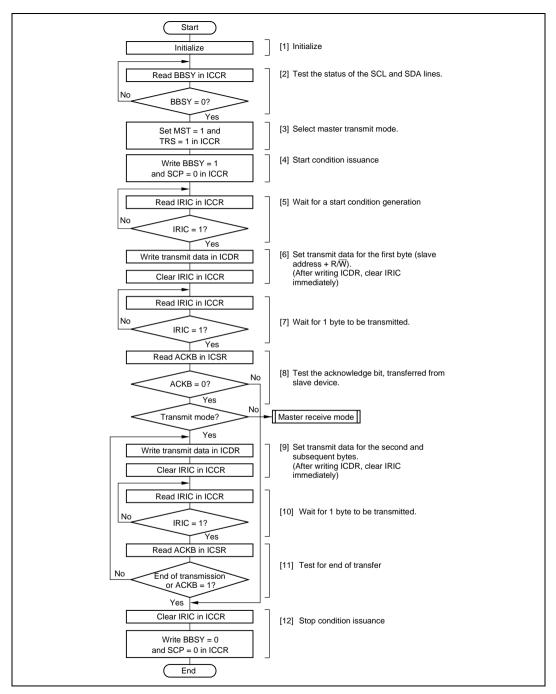


Figure 16.14 Flowchart for Master Transmit Mode (Example)

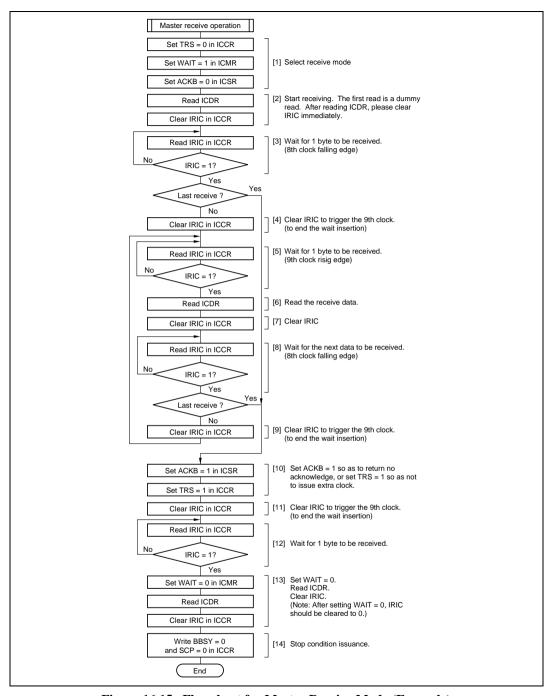


Figure 16.15 Flowchart for Master Receive Mode (Example)

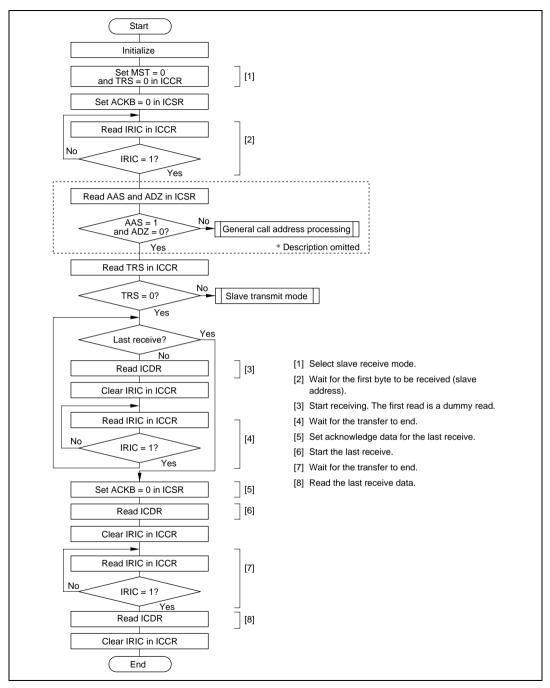


Figure 16.16 Flowchart for Slave Receive Mode (Example)

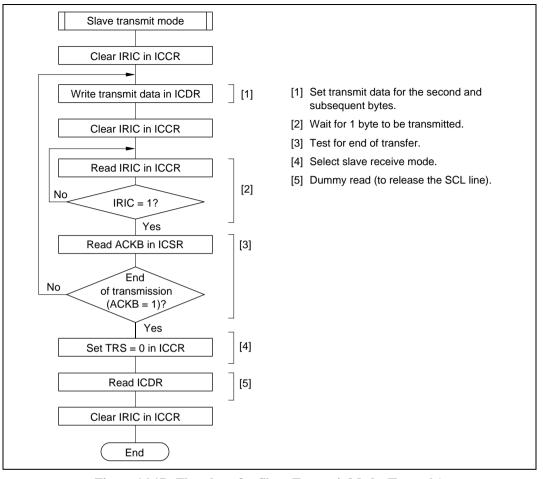


Figure 16.17 Flowchart for Slave Transmit Mode (Example)

16.3.11 Initialization of Internal State

The IIC has a function for forcible initialization of its internal state if a deadlock occurs during communication.

Initialization is executed by (1) setting bits CLR3 to CLR0 in the DDCSWR register or (2) clearing the ICE bit. For details of settings for bits CLR3 to CLR0, see section 16.2.8, DDC Switch Register (DDCSWR).

Scope of Initialization:

The initialization executed by this function covers the following items:

- TDRE and RDRF internal flags
- Transmit/receive sequencer and internal operating clock counter
- Internal latches for retaining the output state of the SCL and SDA pins (wait, clock, data output, etc.)

The following items are not initialized:

- Actual register values (ICDR, SAR, SARX, ICMR, ICCR, ICSR, DDCSWR, STCR)
- Internal latches used to retain register read information for setting/clearing flags in the ICMR, ICCR, ICSR, and DDCSWR registers
- The value of the ICMR register bit counter (BC2 to BC0)
- Generated interrupt sources (interrupt sources transferred to the interrupt controller)

Notes on Initialization:

- Interrupt flags and interrupt sources are not cleared, and so flag clearing measures must be taken as necessary.
- Basically, other register flags are not cleared either, and so flag clearing measures must be taken as necessary.
- When initialization is performed by means of the DDCSWR register, the write data for bits CLR3 to CLR0 is not retained. To perform IIC clearance, bits CLR3 to CLR0 must be written to simultaneously using an MOV instruction. Do not use a bit-manipulation instruction such as BCLR. Similarly, when clearing is required again, all the bits must be written to simultaneously in accordance with the setting.
- If a flag clearing setting is made during transmission/reception, the IIC module will stop transmitting/receiving at that point and the SCL and SDA pins will be released. When transmission/reception is started again, register initialization, etc., must be carried out as necessary to enable correct communication as a system.

The value of the BBSY bit cannot be modified directly by this module clear function, but since the stop condition pin waveform is generated according to the state and release timing of the SCL and SDA pins, the BBSY bit may be cleared as a result. Similarly, state switching of other bits and flags may also have an effect.

To prevent problems caused by these factors, the following procedure should be used when initializing the IIC state.

- 1. Execute initialization of the internal state by setting of bit CLR3 to CLR0 or by clearing ICE bit.
- 2. Execute a stop condition issuance instruction (write 0 to BBSY and SCP) to clear the BBSY bit to 0, and wait for two transfer rate clock cycles.
- 3. Re-execute initialization of the internal state by setting of bit CLR3 to CLR0 or by clearing ICE bit.
- 4. Initialize (re-set) the IIC registers.

16.4 Usage Notes

- In master mode, if an instruction to generate a start condition is immediately followed by an instruction to generate a stop condition, neither condition will be output correctly. To output consecutive start and stop conditions, after issuing the instruction that generates the start condition, read the relevant ports, check that SCL and SDA are both low, then issue the instruction that generates the stop condition. Note that SCL may not yet have gone low when BBSY is cleared to 0.
- Either of the following two conditions will start the next transfer. Pay attention to these conditions when reading or writing to ICDR.
 - Write access to ICDR when ICE = 1 and TRS = 1 (including automatic transfer from ICDRT to ICDRS)
 - Read access to ICDR when ICE = 1 and TRS = 0 (including automatic transfer from ICDRS to ICDRR)
- Table 16.6 shows the timing of SCL and SDA output in synchronization with the internal clock. Timings on the bus are determined by the rise and fall times of signals affected by the bus load capacitance, series resistance, and parallel resistance.



Table 16.6 I²C Bus Timing (SCL and SDA Output)

Item	Symbol	Output Timing	Unit	Notes
SCL output cycle time	t _{sclo}	28t _{cyc} to 256t _{cyc}	ns	Figure 26.28
SCL output high pulse width	t _{sclho}	0.5t _{sclo}	ns	(reference)
SCL output low pulse width	t _{scllo}	0.5t _{sclo}	ns	-
SDA output bus free time	t _{BUFO}	0.5t _{sclo} -1t _{cyc}	ns	-
Start condition output hold time	t _{STAHO}	0.5t _{sclo} -1t _{cyc}	ns	-
Retransmission start condition output setup time	t _{staso}	1t _{sclo}	ns	-
Stop condition output setup time	t _{stoso}	0.5t _{sclo} +2t _{cyc}	ns	-
Data output setup time (master)	t _{SDASO}	1t _{scllo} –3t _{cyc}	ns	-
Data output setup time (slave)		1t _{scll} -(6t _{cyc} or 12t _{cyc} *)		
Data output hold time	t _{SDAHO}	3t _{cyc}	ns	-

Note: * 6t_{cyc} when IICX is 0, 12t_{cyc} when 1.

- SCL and SDA input is sampled in synchronization with the internal clock. The AC timing therefore depends on the system clock cycle t_{cyc}, as shown in I²C Bus Timing in section 26, Electrical Characteristics, and as shown in table 26.10. Note that the I²C bus interface AC timing specifications will not be met with a system clock frequency of less than 5 MHz.
- The I²C bus interface specification for the SCL rise time t_{sr} is under 1000 ns (300 ns for high-speed mode). In master mode, the I²C bus interface monitors the SCL line and synchronizes one bit at a time during communication. If t_{sr} (the time for SCL to go from low to V_{IH}) exceeds the time determined by the input clock of the I²C bus interface, the high period of SCL is extended. The SCL rise time is determined by the pull-up resistance and load capacitance of the SCL line. To insure proper operation at the set transfer rate, adjust the pull-up resistance and load capacitance so that the SCL rise time does not exceed the values given in the table below.

Table 16.7 Permissible SCL Rise Time (t_{s.}) Values

		_
Tima	Indi	cation

IICX	t _{cyc} Indication		I ² C Bus Specification (Max.)	φ = 5 MHz	φ = 8 MHz	φ = 10 MHz	φ = 16 MHz	φ = 20 MHz
0	7.5t _{cyc}	Standard mode	1000 ns	1000 ns	937 ns	750 ns	468 ns	375 ns
		High-speed mode	300 ns	300 ns	300 ns	300 ns	300 ns	300 ns
1	17.5t _{cyc}	Standard mode	1000 ns	1000 ns	1000 ns	1000 ns	1000 ns	875 ns
		High-speed mode	300 ns	300 ns	300 ns	300 ns	300 ns	300 ns

• The I²C bus interface specifications for the SCL and SDA rise and fall times are under 1000 ns and 300 ns. The I²C bus interface SCL and SDA output timing is prescribed by t_{cyc}, as shown in table 16.6. However, because of the rise and fall times, the I²C bus interface specifications may not be satisfied at the maximum transfer rate. Table 16.8 shows output timing calculations for different operating frequencies, including the worst-case influence of rise and fall times.

 $t_{\mbox{\tiny BUFO}}$ fails to meet the I 2 C bus interface specifications at any frequency. The solution is either (a) to provide coding to secure the necessary interval (approximately 1 μ s) between issuance of a stop condition and issuance of a start condition, or (b) to select devices whose input timing permits this output timing for use as slave devices connected to the I 2 C bus.

 $t_{\scriptscriptstyle SCLLO}$ in high-speed mode and $t_{\scriptscriptstyle STASO}$ in standard mode fail to satisfy the I²C bus interface specifications for worst-case calculations of $t_{\scriptscriptstyle Sr}/t_{\scriptscriptstyle Sr}$. Possible solutions that should be investigated include (a) adjusting the rise and fall times by means of a pull-up resistor and capacitive load, (b) reducing the transfer rate to meet the specifications, or (c) selecting devices whose input timing permits this output timing for use as slave devices connected to the I²C bus.

Table 16.8 I²C Bus Timing (with Maximum Influence of t_{sr}/t_{sr})

Time Indication (at Maximum Transfer Rate) [ns]

			·····o ····aioaiio··· (at ····axiiiiaii · ··aiio·io· · ·tato) [··o]						
Item	t _{cyc} Indication		t _{s₁} /t _{s₁} Influence (Max.)	I ² C Bus Specifi- cation (Min.)	φ = 5 MHz	φ = 8 MHz	φ = 10 MHz	φ = 16 MHz	φ = 20 MHz
t _{sclho}	0.5t _{SCLO} (-t _{Sr})	Standard mode	-1000	4000	4000	4000	4000	4000	4000
		High-speed mode	-300	600	950	950	950	950	950
t _{scllo}	0.5t _{SCLO} (-t _{Sf})	Standard mode	-250	4700	4750	4750	4750	4750	4750
		High-speed mode	-250	1300	1000*1	1000*1	1000*1	1000*1	1000*1
t _{BUFO}	0.5t _{SCLO} -1t _{cyc} (-t _{Sr})	Standard mode	-1000	4700	3800*1	3875*1	3900*1	3938*1	3950 ^{*1}
		High-speed mode	-300	1300	750 ^{*1}	825*1	850*1	888*1	900*1
t _{STAHO}	0.5t _{SCLO} -1t _{cyc} (-t _{Sf})	Standard mode	-250	4000	4550	4625	4650	4688	4700
		High-speed mode	-250	600	800	875	900	938	950
t _{STASO}	$1t_{\text{SCLO}}(-t_{\text{Sr}})$	Standard mode	-1000	4700	9000	9000	9000	9000	9000
		High-speed mode	-300	600	2200	2200	2200	2200	2200
t _{stoso}	0.5t _{SCLO} +2t _{cyc} (-t _{Sr})	Standard mode	-1000	4000	4400	4250	4200	4125	4100
		High-speed mode	-300	600	1350	1200	1150	1075	1050
t _{SDASO} (master)	1t _{SCLLO} *3 -3t _{cyc} (-t _{Sr})	Standard mode	-1000	250	3100	3325	3400	3513	3550
		High-speed mode	-300	100	400	625	700	813	850
t _{sdaso} (slave)	1t _{SCLL} *3 -12t _{cyc} *2	Standard mode	-1000	250	1300	2200	2500	2950	3100
	(-t _{Sr})	High-speed mode	-300	100	-1400 ^{*1}	-500 ^{*1}	-200 ^{*1}	250	400

		Time Indication (at Maximum Tran					fer Rate)	[ns]	
Item	t _{cyc} Indicatior	1	t _{s/} /t _s Influence (Max.)	I ² C Bus Specifi- cation (Min.)	φ = 5 MHz	φ = 8 MHz	φ = 10 MHz	φ = 16 MHz	φ = 20 MHz
t _{SDAHO}	3t _{cyc}	Standard mode	0	0	600	375	300	188	150
		High-speed mode	0	0	600	375	300	188	150

- Notes: 1. Does not meet the I²C bus interface specification. Remedial action such as the following is necessary: (a) secure a start/stop condition issuance interval; (b) adjust the rise and fall times by means of a pull-up resistor and capacitive load; (c) reduce the transfer rate; (d) select slave devices whose input timing permits this output timing.
 - The values in the above table will vary depending on the settings of the IICX bit and bits CKS0 to CKS2. Depending on the frequency it may not be possible to achieve the maximum transfer rate; therefore whether or not the I²C bus interface specifications are met must be determined in accordance with the actual setting conditions.
 - 2. Value when the IICX bit is set to 1. When the IICX bit is cleared to 0, the value is $(t_{scl.} 6t_{cvc})$.
 - 3. Calculated using the I²C bus specification values (standard mode: 4700 ns min.; high-speed mode: 1300 ns min.).

• Note on ICDR Read at End of Master Reception

To halt reception at the end of a receive operation in master receive mode, set the TRS bit to 1 and write 0 to BBSY and SCP in ICCR. This changes SDA from low to high when SCL is high, and generates the stop condition. After this, receive data can be read by means of an ICDR read, but if data remains in the buffer the ICDRS receive data will not be transferred to ICDR, and so it will not be possible to read the second byte of data.

If it is necessary to read the second byte of data, issue the stop condition in master receive mode (i.e. with the TRS bit cleared to 0). When reading the receive data, first confirm that the BBSY bit in the ICCR register is cleared to 0, the stop condition has been generated, and the bus has been released, then read the ICDR register with TRS cleared to 0.

Note that if the receive data (ICDR data) is read in the interval between execution of the instruction for issuance of the stop condition (writing of 0 to BBSY and SCP in ICCR) and the actual generation of the stop condition, the clock may not be output correctly in subsequent master transmission.

Clearing of the MST bit after completion of master transmission/reception, or other modifications of IIC control bits to change the transmit/receive operating mode or settings, must be carried out during interval (a) in figure 16.18 (after confirming that the BBSY bit has been cleared to 0 in the ICCR register).

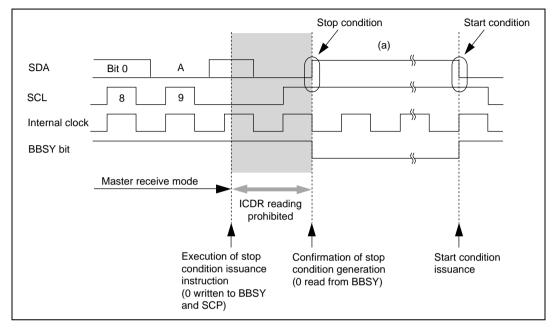


Figure 16.18 Points for Attention Concerning Reading of Master Receive Data

Notes on Start Condition Issuance for Retransmission Figure 16.19 shows the timing of start condition issuance for retransmission, and the timing for subsequently writing data to ICDR, together with the corresponding flowchart. After start condition issuance is done and determined the start condition, write the transmit data to ICDR, as shown below.

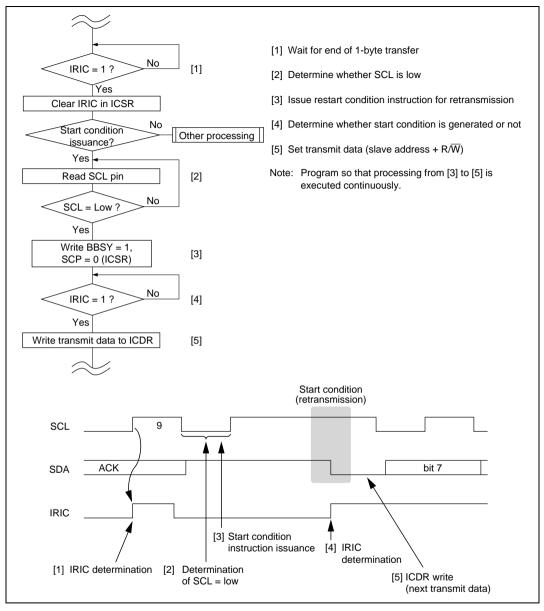


Figure 16.19 Flowchart and Timing of Start Condition Instruction Issuance for Retransmission

Notes on I²C Bus Interface Stop Condition Instruction Issuance

If the rise time of the 9th SCL clock exceeds the specification because the bus load capacitance is large, or if there is a slave device of the type that drives SCL low to effect a wait, after rising of the 9th SCL clock, issue the stop condition instruction after reading SCL and determining it to be low, as shown below.

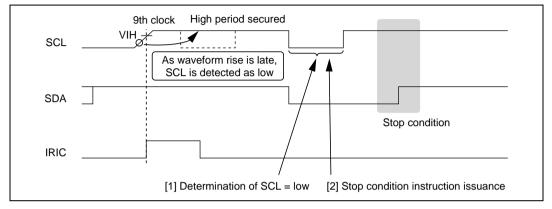


Figure 16.20 Timing of Stop Condition Issuance

Notes on WAIT Function

— Conditions to cause this phenomenon

When both of the following conditions are satisfied, the clock pulse of the 9th clock could be outputted continuously in master mode using the WAIT function due to the failure of the WAIT insertion after the 8th clock fall.

- (1) Setting the WAIT bit of the ICMR register to 1 and operating WAIT, in master mode
- (2) If the IRIC bit of interrupt flag is cleared from 1 to 0 between the fall of the 7th clock and the fall of the 8th clock.

Error phenomenon

Normally, WAIT State will be cancelled by clearing the IRIC flag bit from 1 to 0 after the fall of the 8th clock in WAIT State. In this case, if the IRIC flag bit is cleared between the 7th clock fall and the 8th clock fall, the IRIC flag clear- data will be retained internally. Therefore, the WAIT State will be cancelled right after WAIT insertion on 8th clock fall.

— Restrictions

Please clear the IRIC flag before the rise of the 7th clock (the counter value of BC2 through BC0 should be 2 or greater), after the IRIC flag is set to 1 on the rise of the 9th clock.

If the IRIC flag-clear is delayed due to the interrupt or other processes and the value of BC counter is turned to 1 or 0, please confirm the SCL pins are in L' state after the counter value of BC2 through BC0 is turned to 0, and clear the IRIC flag. (See figure 16.21.)

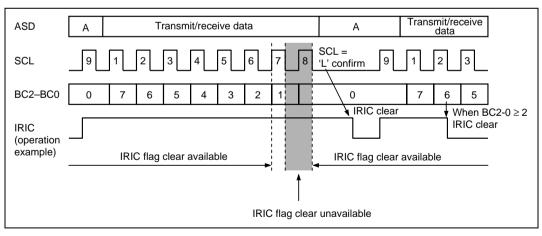


Figure 16.21 IRIC Flag Clear Timing on WAIT Operation

• Notes on ICDR Reads and ICCR Access in Slave Transmit Mode

In a transmit operation in the slave mode of the I²C bus interface, do not read the ICDR register or read or write to the ICCR register during the period indicated by the shaded portion in figure 16.22.

Normally, when interrupt processing is triggered in synchronization with the rising edge of the 9th clock cycle, the period in question has already elapsed when the transition to interrupt processing takes place, so there is no problem with reading the ICDR register or reading or writing to the ICCR register.

To ensure that the interrupt processing is performed properly, one of the following two conditions should be applied.

- (1) Make sure that reading received data from the ICDR register, or reading or writing to the ICCR register, is completed before the next slave address receive operation starts.
- (2) Monitor the BC2 to BC0 counter in the ICMR register and, when the value of BC2 to BC0 is 000 (8th or 9th clock cycle), allow a waiting time of at least 2 transfer clock cycles in order to involve the problem period in question before reading from the ICDR register, or reading or writing to the ICCR register.

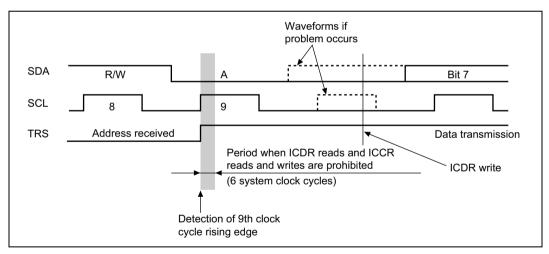


Figure 16.22 ICDR Read and ICCR Access Timing in Slave Transmit Mode

• Notes on TRS Bit Setting in Slave Mode

From the detection of the rising edge of the 9th clock cycle or of a stop condition to when the rising edge of the next SCL pin signal is detected (the period indicated as (a) in figure 16.23) in the slave mode of the I²C bus interface, the value set in the TRS bit in the ICCR register is effective immediately.

However, at other times (indicated as (b) in figure 16.23) the value set in the TRS bit is put on hold until the next rising edge of the 9th clock cycle or stop condition is detected, rather than taking effect immediately.

This results in the actual internal value of the TRS bit remaining 1 (transmit mode) and no acknowledge bit being sent at the 9th clock cycle address receive completion in the case of an address receive operation following a restart condition input with no stop condition intervening.

When receiving an address in the slave mode, clear the TRS bit to 0 during the period indicated as (a) in figure 16.23.

To cancel the holding of the SCL bit low by the wait function in the slave mode, clear the TRS bit to 0 and then perform a dummy read of the ICDR register.

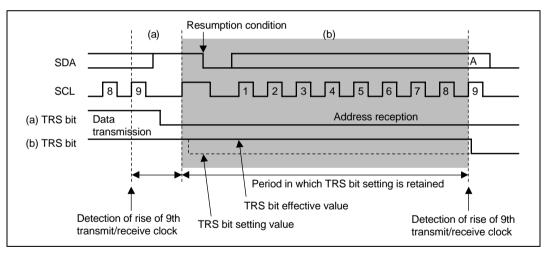


Figure 16.23 TRS Bit Setting Timing in Slave Mode

Notes on Arbitration Lost in Master Mode

The I²C bus interface recognizes the data in transmit/receive frame as an address when arbitration is lost in master mode and a transition to slave receive mode is automatically carried out.

When arbitration is lost not in the first frame but in the second frame or subsequent frame, transmit/receive data that is not an address is compared with the value set in the SAR or SARX register as an address. If the receive data matches with the address in the SAR or SARX register, the I²C bus interface erroneously recognizes that the address call has occurred. (See figure 16.24.)

In multi-master mode, a bus conflict could happen. When The I²C bus interface is operated in master mode, check the state of the AL bit in the ICSR register every time after one frame of data has been transmitted or received.

When arbitration is lost during transmitting the second frame or subsequent frame, take avoidance measures.

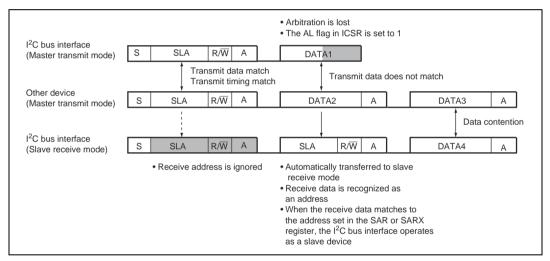


Figure 16.24 Diagram of Erroneous Operation when Arbitration is Lost

Though it is prohibited in the normal I²C protocol, the same problem may occur when the MST bit is erroneously set to 1 and a transition to master mode is occurred during data transmission or reception in slave mode. In multi-master mode, pay attention to the setting of the MST bit when a bus conflict may occur. In this case, the MST bit in the ICCR register should be set to 1 according to the order below.

- (a) Make sure that the BBSY flag in the ICCR register is 0 and the bus is free before setting the MST bit.
- (b) Set the MST bit to 1.

- (c) To confirm that the bus was not entered to the busy state while the MST bit is being set, check that the BBSY flag in the ICCR register is 0 immediately after the MST bit has been set.
- Notes on Interrupt Occurrence after ACKB Reception
 - Conditions to cause this failure

The IRIC flag is set to 1 when both of the following conditions are satisfied.

- 1 is received as the acknowledge bit for transmit data and the ACKB bit in ICSR is set to 1
- Rising edge of the 9th transmit/receive clock is input to the SCL pin

When the above two conditions are satisfied in slave receive mode, an unnecessary interrupt occurs.

Figure 16.25 shows the note on interrupt occurrence in slave mode after receiving 1 as the acknowledge bit (ACKB = 1).

(1) For the last transmit data in master transmit mode or slave transmit mode, 1 is received as the acknowledge bit.

If the ACKE bit in ICCR is set to 1 at this time, the ACKB bit in ICSR is set to 1.

- (2) After switching to slave receive mode, the start condition is input, and address reception is performed next.
- (3) Even if the received address does not match the address set in SAR or SARX, the IRIC flag is set to 1 at the rise of the 9th transmit/receive clock, thus causing an interrupt to occur.

Note that if the slave address matches, an interrupt is to be generated at the rise of the 9th transmit/receive clock as normal operation, so this is not erroneous operation.

— Restriction

In a transmit operation of the I²C bus interface module, carry out the following countermeasures.

- (1) After 1 is received as the acknowledge bit for transmit data, clear the ACKE bit in ICCR to 0 to clear the ACKB bit to 0.
- (2) To enable acknowledge bit reception afterwards, set the ACKE bit to 1 again.



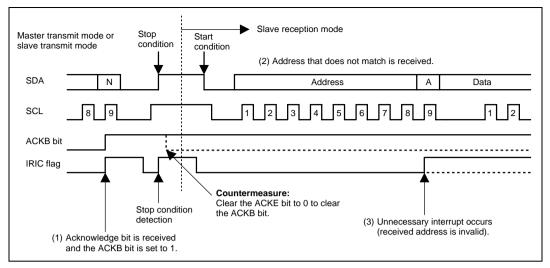


Figure 16.25 Note on Interrupt Occurrence in Slave Mode after ACKB = 1 Reception

Notes on TRS Bit Setting and ICDR Register Access

Conditions to cause this failure

Low-fixation of the SCL pins is cancelled incorrectly when the following conditions are satisfied

— Master mode

Figure 16.26 shows the notes on ICDR reading (TRS = 1) in master mode.

- (1) When previously received 2-bytes data remains in ICDR unread (ICDRS are full).
- (2) Reads ICDR register after switching to transmit mode (TRS = 1). (RDRF = 0 state)
- (3) Sets to receive mode (TRS = 0), after transmitting Rev.1 frame of issued start condition by master mode.

- Slave mode

Figure 16.27 shows the notes on ICDR writing (TRS = 0) in slave mode.

(1) Writes ICDR register in receive mode (TRS = 0), after entering the start condition by slave mode (TDRE = 0 state).

Address match with Rev.1 frame, receive 1 by R/W bit, and switches to transmit mode (TRS = 1).

When these conditions are satisfied, the low fixation of the SCL pins is cancelled without ICDR register access after Rev.1 frame is transferred.

- Restriction

Please carry out the following countermeasures when transmitting/receiving via the IIC bus interface module.

- (1) Please read the ICDR registers in receive mode, and write them in transmit mode.
- (2) In receiving operation with master mode, please issue the start condition after clearing the internal flag of the IIC bus interface module, using CLR3 to CLR0 bit of the DDCSWR register on bus-free state (BBSY = 0).



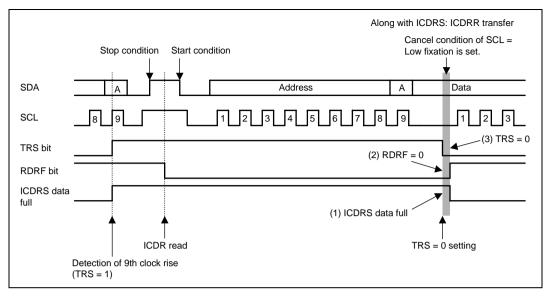


Figure 16.26 Notes on ICDR Reading with TRS = 1 Setting in Master Mode

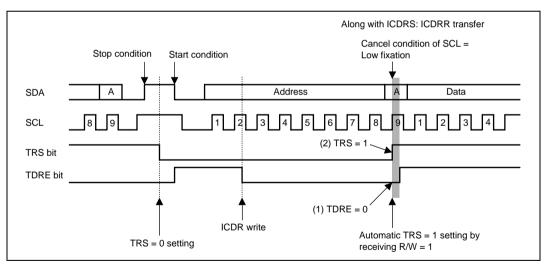


Figure 16.27 Notes on ICDR Writing with TRS = 0 Setting in Slave Mode

Section 17 Keyboard Buffer Controller

Provided in the H8S/2148 Group and H8S/2147N; not provided in the H8S/2144 Group.

17.1 Overview

The H8S/2148 Group and H8S/2147N have three on-chip keyboard buffer controller channels, designated 0, 1, and 2. The keyboard buffer controller is provided with functions conforming to the PS/2 interface specifications.

Data transfer using the keyboard buffer controller employs a data line (KD) and a clock line, providing economical use of connectors, board surface area, etc. Figure 17.1 shows how the keyboard buffer controller is connected.

17.1.1 Features

- Conforms to PS/2 interface specifications
- Direct bus drive (via the KCLK and KD pins)
- Interrupt sources: on completion of data reception and on detection of clock edge
- Error detection: parity error and stop bit monitoring

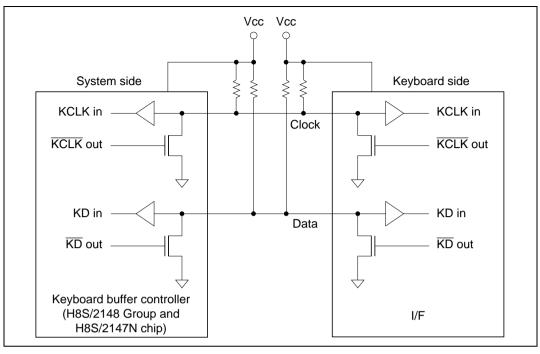


Figure 17.1 Keyboard Buffer Controller Connection

17.1.2 Block Diagram

Figure 17.2 shows a block diagram of the keyboard buffer controller.

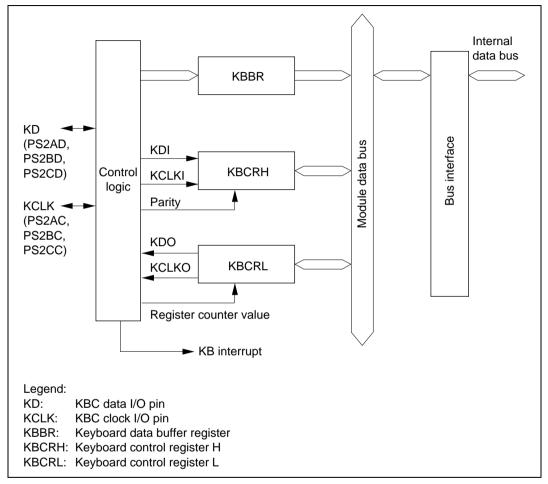


Figure 17.2 Block Diagram of Keyboard Buffer Controller

17.1.3 **Input/Output Pins**

Table 17.1 lists the input/output pins used by the keyboard buffer controller.

Table 17.1 Keyboard Buffer Controller Input/Output Pins

Channel	Name	${\sf Abbreviation}^*$	I/O	Function
0	KBC clock I/O pin (KCLK0)	PS2AC	I/O	KBC clock input/output
	KBC data I/O pin (KD0)	PS2AD	I/O	KBC data input/output
1	KBC clock I/O pin (KCLK1)	PS2BC	I/O	KBC clock input/output
	KBC data I/O pin (KD1)	PS2BD	I/O	KBC data input/output
2	KBC clock I/O pin (KCLK2)	PS2CC	I/O	KBC clock input/output
	KBC data I/O pin (KD2)	PS2CD	I/O	KBC data input/output

These are the external I/O pin names. In the text, clock I/O pins are referred to as KCLK Note: and data I/O pins as KD, omitting the channel designations.

Register Configuration 17.1.4

Table 17.2 lists the registers of the keyboard buffer controller.

Table 17.2 Keyboard Buffer Controller Registers

Channel	Name	Abbreviation	R/W	Initial Value	Address*1
0	Keyboard control register H	KBCRH0	R/(W)*2	H'70	H'FED8
	Keyboard control register L	KBCRL0	R/W	H'70	H'FED9
	Keyboard data buffer register	KBBR0	R	H'00	H'FEDA
1	Keyboard control register H	KBCRH1	R/(W)*2	H'70	H'FEDC
	Keyboard control register L	KBCRL1	R/W	H'70	H'FEDD
	Keyboard data buffer register	KBBR1	R	H'00	H'FEDE
2	Keyboard control register H	KBCRH2	R/(W)*2	H'70	H'FEE0
	Keyboard control register L	KBCRL2	R/W	H'70	H'FEE1
	Keyboard data buffer register	KBBR2	R	H'00	H'FEE2
Common	Module stop control register	MSTPCRH	R/W	H'3F	H'FF86
		MSTPCRL	R/W	H'FF	H'FF87

Notes: 1. Lower 16 bits of the address.

2. Only 0 can be written in bits 2 and 1, to clear the flags.



17.2 Register Descriptions

17.2.1 Keyboard Control Register H (KBCRH)

Bit	7	6	5	4	3	2	1	0
	KBIOE	KCLKI	KDI	KBFSEL	KBIE	KBF	PER	KBS
Initial value	0	1	1	1	0	0	0	0
Read/Write	R/W	R	R	R/W	R/W	R/(W)*	R/(W)*	R

Note: * Only 0 can be written, to clear the flags.

KBCRH is an 8-bit readable/writable register that indicates the operating status of the keyboard buffer controller.

KBCRH is initialized to H'70 by a reset, and in standby mode, watch mode, subactive mode, subsleep mode, and module stop mode. Bits 6, 5, and 2 to 0 are also initialized when KBIOE is cleared to 0.

Bit 7—Keyboard In/Out Enable (KBIOE): Selects whether or not the keyboard buffer controller is used. When KBIOE is set to 1, the keyboard buffer controller is enabled for transmission and reception and the port pins function as KCLK and KD I/O pins. When KBIOE is cleared to 0, the keyboard buffer controller stops functioning and the port pins go to the high-impedance state.

Bit 7

KBIOE	Description
0	The keyboard buffer controller is non-operational (KCLK and KD signal pins have port functions) (Initial value)
1	The keyboard buffer controller is enabled for transmission and reception (KCLK and KD signal pins are in the bus drive state)

Bit 6—Keyboard Clock In (KCLKI): Monitors the KCLK I/O pin. This bit cannot be modified.

Bit 6

KCLKI	Description	
0	KCLK I/O pin is low	
1	KCLK I/O pin is high	(Initial value)

Bit 5—Keyboard Data In (KDI): Monitors the KDI I/O pin. This bit cannot be modified.

Bit 5

KDI	 Description	
0	KD I/O pin is low	
1	KD I/O pin is high	(Initial value)

Bit 4—Keyboard Buffer Register Full Select (KBFSEL): Selects whether the KBF bit is used as the keyboard buffer register full flag or as the KCLK fall interrupt flag, When KBFSEL is cleared to 0, the KBE bit in the KBCRL register should be cleared to 0 to disable reception.

Bit 4

KBFSEL	Description	
0	KBF bit is used as KCLK fall interrupt flag	
1	KBF bit is used as keyboard buffer register full flag	(Initial value)

Bit 3—Keyboard Interrupt Enable (KBIE): Enables or disables interrupts from the keyboard buffer controller to the CPU.

Bit 3

KBIE	Description	
0	Interrupt requests are disabled	(Initial value)
1	Interrupt requests are enabled	

Bit 2—Keyboard Buffer Register Full (KBF): Indicates that data reception has been completed and the received data is in the keyboard data buffer register (KBBR).

Bit 2

KBF	Description
0	[Clearing condition] (Initial value)
	Read KBF when KBF =1, then write 0 in KBF
1	[Setting conditions]
	 When data has been received normally and has been transferred to KBBR (keyboard buffer register full flag)
	 When a KCLK falling edge is detected (while KBFSEL = 0) (KCLK interrupt flag)

Bit 1—Parity Error (PER): Indicates that an odd parity error has occurred.

Bit 1

PER	Description	
0	[Clearing condition]	(Initial value)
	Read PER when PER =1, then write 0 in PER	
1	[Setting condition]	
	When an odd parity error occurs	

Bit 0—Keyboard Stop (KBS): Indicates the receive data stop bit. Valid only when KBF = 1.

Bit 0

KBS	Description	
0	0 stop bit received	(Initial value)
1	1 stop bit received	

17.2.2 Keyboard Control Register L (KBCRL)

Bit	7	6	5	4	3	2	1	0
	KBE	KCLKO	KDO		RXCR3	RXCR2	RXCR1	RXCR0
Initial value	0	1	1	1	0	0	0	0
Read/Write	R/W	R/W	R/W	_	R	R	R	R

KBCRL is an 8-bit readable/writable register that enables the receive counter count and controls the keyboard buffer controller pin output.

KBCRL is initialized to H'70 by a reset, and in standby mode, watch mode, subactive mode, subsleep mode, and module stop mode.

Bit 7—Keyboard Enable (KBE): Enables or disables loading of receive data into the keyboard data buffer register (KBBR).

Bit 7

KBE	Description	
0	Loading of receive data into KBBR is disabled	(Initial value)
1	Loading of receive data into KBBR is enabled	

Bit 6—Keyboard Clock Out (KCLKO): Controls KBC clock I/O pin output.

Bit 6

KCLKO	Description	
0	Keyboard buffer controller clock I/O pin is low	
1	Keyboard buffer controller clock I/O pin is high	(Initial value)

Bit 5—Keyboard Data Out (KDO): Controls KBC data I/O pin output.

Bit 5

KDO	Description	
0	Keyboard buffer controller data I/O pin is low	
1	Keyboard buffer controller data I/O pin is high	(Initial value)

Bit 4—Reserved: This bit cannot be modified and is always read as 1.

Bits 3 to 0—Receive Counter (RXCR3 to RXCR0): These bits indicate the received data bit. Their value is incremented on the fall of KCLK. These bits cannot be modified.

The receive counter is initialized to 0000 by a reset and when 0 is written in KBE. Its value returns to 0000 after a stop bit is received.

Bit 3	Bit 2	Bit 1	Bit 0		
RXCR3	RXCR2	RXCR1	RXCR0	Receive Data Contents	
0	0	0	0	_	(Initial value)
			1	Start bit	
		1	0	KB0	
			1	KB1	
	1	0	0	KB2	
			1	KB3	
		1	0	KB4	
			1	KB5	
1	0	0	0	KB6	
			1	KB7	
		1	0	Parity bit	
			1	_	
	1	_	_	_	

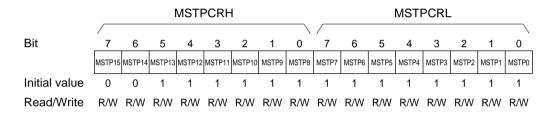
17.2.3 Keyboard Data Buffer Register (KBBR)

Bit	7	6	5	4	3	2	1	0
	KB7	KB6	KB5	KB4	KB3	KB2	KB1	KB0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R	R	R	R	R	R	R	R

KBBR is a read-only register that stores receive data. Its value is valid only when KBF = 1.

KBBR is initialized to H'00 by a reset, in standby mode, watch mode, subactive mode, subsleep mode, and module stop mode, and when KBIOE is cleared to 0.

17.2.4 Module Stop Control Register (MSTPCR)



MSTPCR, comprising two 8-bit readable/writable register, performs module stop mode control. When the MSTP2 bit is set to 1, the keyboard buffer controller halts and enters module stop mode. See section 25.5, Module Stop Mode, for details.

MSTPCR is initialized to H'3FFF by a reset and in hardware standby mode. It is not initialized in software standby mode.

MSTPCRL Bit 2—Module Stop (MSTP2): Specifies keyboard buffer controller module stop mode.

MSTPCRL Bit 2

MSTP2	Description	
0	Keyboard buffer controller module stop mode is cleared	
1	Keyboard buffer controller module stop mode is set	(Initial value)

17.3 Operation

17.3.1 Receive Operation

In a receive operation, both KCLK (clock) and KD (data) are outputs on the keyboard side and inputs on the H8S/2148 Group chip (system) side. KD receives a start bit, 8 data bits (LSB-first), an odd parity bit, and a stop bit, in that order. The KD value is valid when KCLK is low. A sample receive processing flowchart is shown in figure 17.3, and the receive timing in figure 17.4.

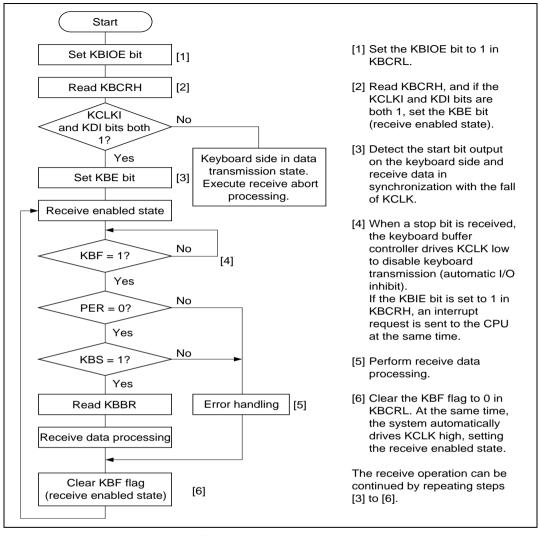


Figure 17.3 Sample Receive Processing Flowchart

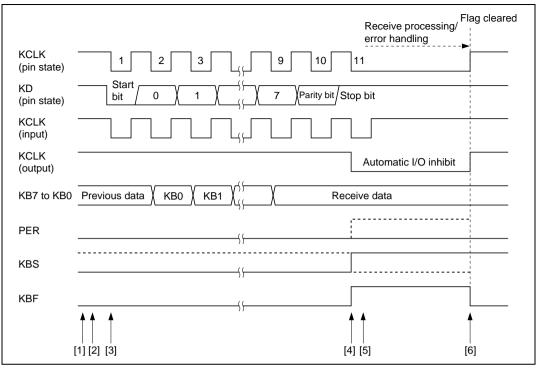


Figure 17.4 Receive Timing

17.3.2 Transmit Operation

In a transmit operation, KCLK (clock) is an output on the keyboard side, and KD (data) is an output on the H8S/2148 Group and H8S/2147N chip (system) side. KD outputs a start bit, 8 data bits (LSB-first), an odd parity bit, and a stop bit, in that order. The KD value is valid when KCLK is high. A sample transmit processing flowchart is shown in figure 17.5, and the transmit timing in figure 17.6.

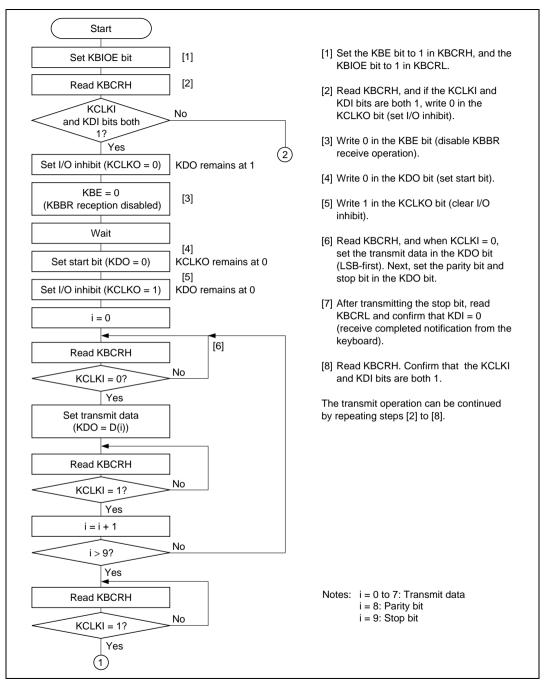


Figure 17.5 Sample Transmit Processing Flowchart

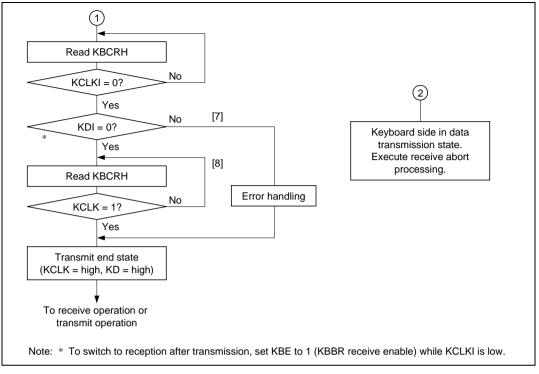


Figure 17.5 Sample Transmit Processing Flowchart (cont)

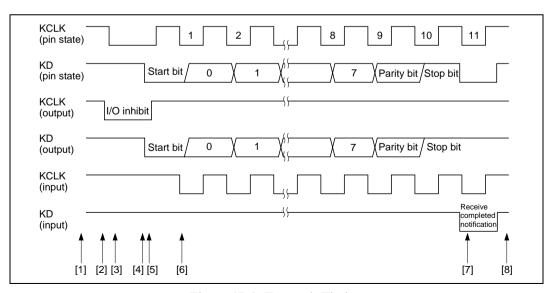


Figure 17.6 Transmit Timing

17.3.3 Receive Abort

The H8S/2148 Group and H8S/2147N device (system side) can forcibly abort transmission from the device connected to it (keyboard side) in the event of a protocol error, etc. In this case, the system holds the clock low. During reception, the keyboard also outputs a clock for synchronization, and the clock is monitored when the keyboard output clock is high. If the clock is low at this time, the keyboard judges that there is an abort request from the system, and data transmission from the keyboard is aborted. Thus the system can abort reception by holding the clock low for a certain period. A sample receive abort processing flowchart is shown in figure 17.7, and the receive abort timing in figure 17.8.

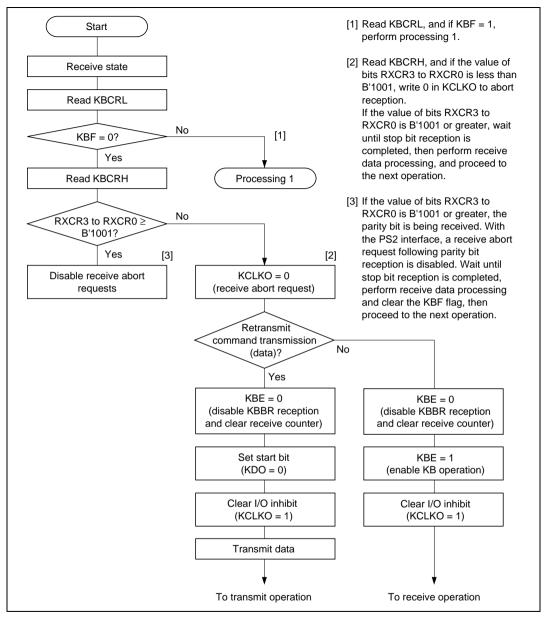


Figure 17.7 Sample Receive Abort Processing Flowchart

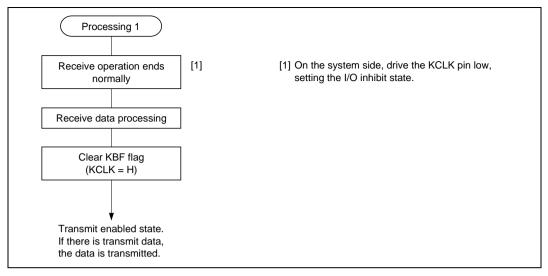


Figure 17.7 Sample Receive Abort Processing Flowchart (cont)

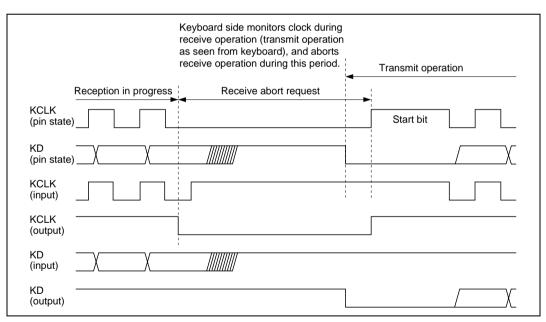


Figure 17.8 Receive Abort and Transmit Start (Transmission/Reception Switchover)
Timing

17.3.4 KCLKI and KDI Read Timing

Figure 17.9 shows the KCLKI and KDI read timing.

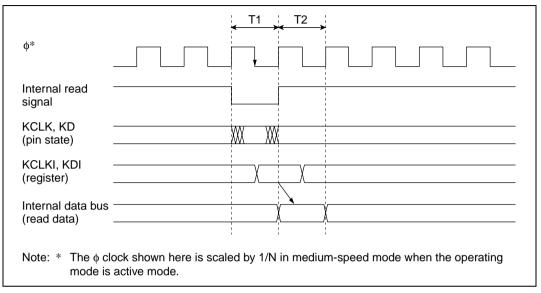


Figure 17.9 KCLKI and KDI Read Timing

17.3.5 KCLKO and KDO Write Timing

Figure 17.10 shows the KLCKO and KDO write timing and the KCLK and KD pin states.

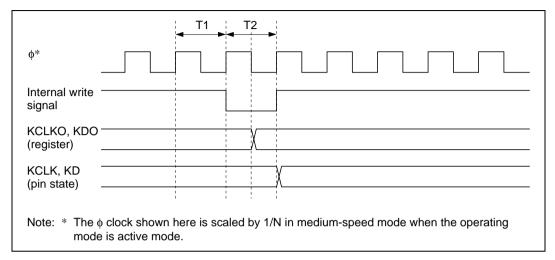


Figure 17.10 KCLKO and KDO Write Timing

17.3.6 KBF Setting Timing and KCLK Control

Figure 17.11 shows the KBF setting timing and the KCLK pin states.

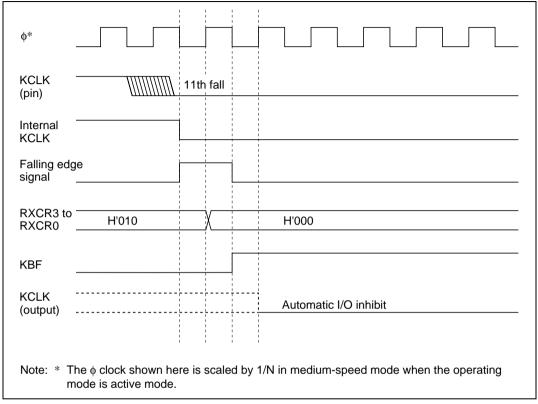


Figure 17.11 KBF Setting and KCLK Automatic I/O Inhibit Generation Timing

17.3.7 Receive Timing

Figure 17.12 shows the receive timing.

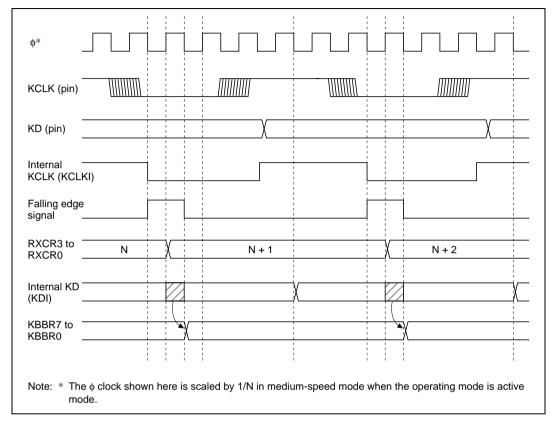


Figure 17.12 Receive Counter and KBBR Data Load Timing

17.3.8 KCLK Fall Interrupt Operation

In this device, clearing the KBFSEL bit to 0 in KBCRH enables the KBF bit in KBCRL to be used as a flag for the interrupt generated by the fall of KCLK input.

Figure 17.13 shows the setting method and an example of operation.

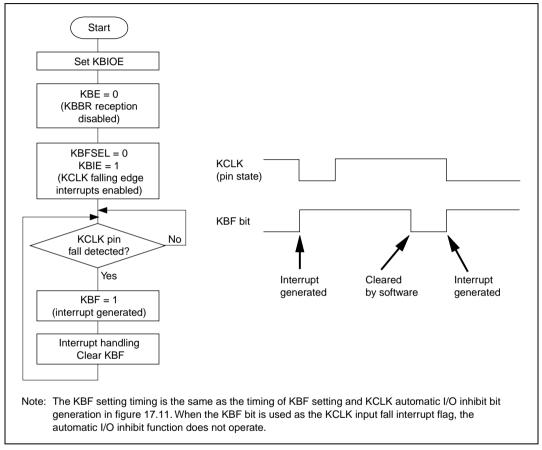


Figure 17.13 Example of KCLK Input Fall Interrupt Operation

17.3.9 Usage Note

When KBIOE is 0, the internal KCLK and internal KD settings are fixed at 1.

Therefore, if the KCLK pin is low when the KBIOE bit is set to 1, the edge detection circuit operates and the KCLK falling edge is detected.

If the KBFSEL bit and KBE bit are both 0 at this time, the KBF bit is set. Figure 17.14 shows the timing of KBIOE setting and KCLK falling edge detection.

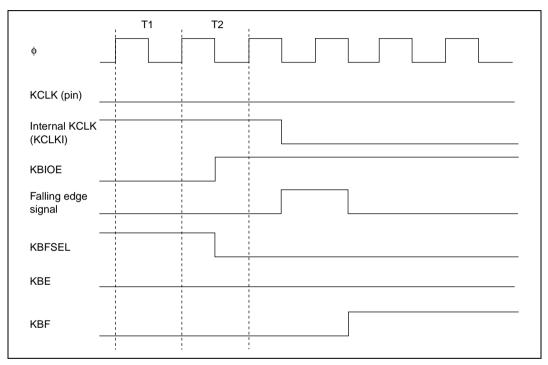


Figure 17.14 KBIOE Setting and KCLK Falling Edge Detection Timing

Section 18 Host Interface

Provided in the H8S/2148 Group and H8S/2147N; not provided in the H8S/2144 Group.

18.1 Overview

The H8S/2148 Group and H8S/2147N have an on-chip host interface (HIF) that enables connection to an ISA bus, widely used as the internal bus in personal computers. The host interface provides a four-channel parallel interface between the on-chip CPU and a host processor. The host interface is available only when the HI12E bit is set to 1 in SYSCR2. This mode is called slave mode, because it is designed for a master-slave communication system in which the H8S/2148 Group and H8S/2147N chip is slaved to a host processor.

18.1.1 Features

The features of the host interface are summarized below.

The host interface consists of 8-byte data registers, 4-byte status registers, a 2-byte control register, fast A20 gate logic, and a host interrupt request circuit. Communication is carried out via seven control signals from the host processor ($\overline{CS1}$, $\overline{CS2}$ or $\overline{ECS2}$, $\overline{CS3}$, $\overline{CS4}$, HA0, \overline{IOR} , and \overline{IOW}), six output signals to the host processor (GA20, HIRQ1, HIRQ11, HIRQ12, HIRQ3, and HIRQ4), and an 8-bit bidirectional command/data bus (HDB7 to HDB0). The $\overline{CS1}$, $\overline{CS2}$ (or $\overline{ECS2}$), $\overline{CS3}$, and $\overline{CS4}$ signals select one of the four interface channels.

18.1.2 **Block Diagram**

Figure 18.1 shows a block diagram of the host interface.

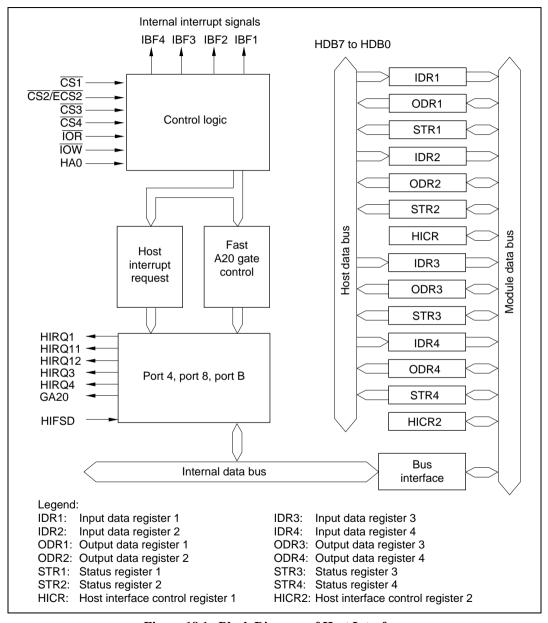


Figure 18.1 Block Diagram of Host Interface

18.1.3 Input and Output Pins

Table 18.1 lists the input and output pins of the host interface module.

Table 18.1 Host Interface Input/Output Pins

Name	Abbreviation	Port	I/O	Function
I/O read	ĪŌR	P93	Input	Host interface read signal
I/O write	ĪOW	P94	Input	Host interface write signal
Chip select 1	CS1	P95	Input	Host interface chip select signal for IDR1, ODR1, STR1
Chip select 2*	CS2	P81	Input	Host interface chip select signal for IDR2,
	ECS2	P90	_	ODR2, STR2
Chip select 3	CS3	PB2	Input	Host interface chip select signal for IDR3, ODR3, STR3
Chip select 4	CS4	PB3	Input	Host interface chip select signal for IDR4, ODR4, STR4
Command/data	HA0	P80	Input	Host interface address select signal.
				In host read access, this signal selects the status registers (STR1 to STR4) or data registers (ODR1 to ODR4). In host write access to the data registers (IDR1 to IDR3, and IDTR4), this signal indicates whether the host is writing a command or data.
Data bus	HDB7 to HDB0	P37 to P30	I/O	Host interface data bus
Host interrupt 1	HIRQ1	P44	Output	Interrupt output 1 to host
Host interrupt 11	HIRQ11	P43	Output	Interrupt output 11 to host
Host interrupt 12	HIRQ12	P45	Output	Interrupt output 12 to host
Host interrupt 3	HIRQ3	PB0	Output	Interrupt output 3 to host
Host interrupt 4	HIRQ4	PB1	Output	Interrupt output 4 to host
Gate A20	GA20	P81	Output	A20 gate control signal output
HIF shutdown	HIFSD	P82	Input	Host interface shutdown control signal

Note: * Selection of $\overline{\text{CS2}}$ or $\overline{\text{ECS2}}$ is by means of the CS2E bit in STCR and the FGA20E bit in HICR. Host interface channel 2 and the $\overline{\text{CS2}}$ pin can be used when CS2E = 1. When CS2E = 1, $\overline{\text{CS2}}$ is used when FGA20E =0, and $\overline{\text{ECS2}}$ is used when FGA20E = 1. In this manual, both are referred to as $\overline{\text{CS2}}$.

18.1.4 Register Configuration

Table 18.2 lists the host interface registers. Host interface registers HICR, IDR1, IDR2, ODR1, ODR2, STR1, and STR2 can only be accessed when the HIE bit is set to 1 in SYSCR.

Table 18.2 Host Interface Registers

	Abbrevia-	R/	W	_Initial	Slave		Mast	er Add	dress*4	.
Name	tion	Slave	Host	Value	Address*3	CS1	CS2	CS3	CS4	HA0
System control register	SYSCR	R/W*1	_	H'09	H'FFC4	_	_	_	_	_
System control register 2	SYSCR2	R/W	_	H'00	H'FF83	_	_	_	_	_
Host interface control register 1	HICR	R/W	_	H'F8	H'FFF0	_	_	_	_	_
Host interface control register 2	HICR2	R/W	_	H'F8	H'FE80	_	_	_	_	_
Input data register 1	IDR1	R	W	_	H'FFF4	0	1	1	1	0/1*5
Output data register 1	ODR1	R/W	R	_	H'FFF5	0	1	1	1	0
Status register 1	STR1	R/(W)*2	R	H'00	H'FFF6	0	1	1	1	1
Input data register 2	IDR2	R	W	_	H'FFFC	1	0	1	1	0/1*5
Output data register 2	ODR2	R/W	R	_	H'FFFD	1	0	1	1	0
Status register 2	STR2	R/(W)*2	R	H'00	H'FFFE	1	0	1	1	1
Input data register 3	IDR3	R	W	_	H'FE84	1	1	0	1	0/1*5
Output data register 3	ODR3	R/W	R	_	H'FE85	1	1	0	1	0
Status register 3	STR3	R/(W)*2	R	H'00	H'FE86	1	1	0	1	1
Input data register 4	IDR4	R	W	_	H'FE8C	1	1	1	0	0/1*5
Output data register 4	ODR4	R/W	R	_	H'FE8D	1	1	1	0	0
Status register 4	STR4	R/(W)*2	R	H'00	H'FE8E	1	1	1	0	1
Module stop control	MSTPCRH	R/W	_	H'3F	H'FF86	_	_	_	_	_
register	MSTPCRL	R/W	_	H'FF	H'FF87	_	_	_	_	_

Notes: 1. Bits 5 and 3 are read-only bits.

- 2. The user-defined bits (bits 7 to 4 and 2) are read/write accessible from the slave processor.
- 3. Address when accessed from the slave processor. The lower 16 bits of the address are shown.
- 4. Pin inputs used in access from the host processor.
- 5. The HA0 input discriminates between writing of commands and data.



18.2 Register Descriptions

18.2.1 System Control Register (SYSCR)

Bit	7	6	5	4	3	2	1	0
	CS2E	IOSE	INTM1	INTM0	XRST	NMIEG	HIE	RAME
Initial value	0	0	0	0	1	0	0	1
Read/Write	R/W	R/W	R	R/W	R	R/W	R/W	R/W

SYSCR is an 8-bit readable/writable register which controls H8S/2148 Group chip operations. Of the host interface registers, HICR, IDR1, ODR1, STR1, IDR2, ODR2, and STR2 can only be accessed when the HIE bit is set to 1. HICR2, IDR3, ODR3, STR3, IDR4, ODR4, and STR4 can be accessed regardless of the setting of the HIE bit. The host interface $\overline{CS2}$ and $\overline{ECS2}$ pins are controlled by the CS2E bit in SYSCR and the FGA20E bit in HICR. See section 3.2.2, System Control Register (SYSCR), and section 5.2.1, System Control Register (SYSCR), for information on other SYSCR bits. SYSCR is initialized to H'09 by a reset and in hardware standby mode.

Bit 7—CS2 Enable Bit (CS2E): Used together with the FGA20E bit in HICR to select the pin that performs the $\overline{\text{CS2}}$ function.

SYSCR Bit 7	Bit 0		
CS2E	FGA20E	Description	
0	0	CS2 pin function halted (CS2 fixed high internally)	(Initial value)
	1		
1	0	CS2 pin function selected for P81/CS2 pin	
	1	CS2 pin function selected for P90/ECS2 pin	

Bit 1—Host Interface Enable (HIE): Enables or disables CPU access to the host interface registers. When enabled, the host interface registers (HICR, IDR1, ODR1, STR1, IDR2, ODR2, and STR2) can be accessed.

Bit 1

HIE	Description	
0	Host interface register (HICR, IDR1, ODR1, STR1, IDR2, ODR2, STR2), CPU access is disabled (Initial value)	_
1	Host interface register (HICR, IDR1, ODR1, STR1, IDR2, ODR2, STR2), CPU access is enabled	

18.2.2 System Control Register 2 (SYSCR2)

Bit	7	6	5	4	3	2	1	0
	KWUL1	KWUL0	P6PUE	_	SDE	CS4E	CS3E	HI12E
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	_	R/W	R/W	R/W	R/W

SYSCR2 is an 8-bit readable/writable register which controls chip operations. Host interface functions are enabled or disabled by the HI12E bit in SYSCR2. The number of channels that can be used can be extended to a maximum of four by means of the CS3E bit and CS4E bit. SYSCR2 is initialized to H'00 by a reset and in hardware standby mode.

Bits 7 and 6—Key Wakeup Level 1 and 0 (KWUL1, KWUL0): The port 6 input level can be set and changed by software. For details see section 8, I/O Ports.

Bit 5—Port 6 Input Pull-Up Extra (P6PUE): Controls and selects the current specification for the port 6 MOS input pull-up function connected by means of KMPCR settings. For details see section 8, I/O Ports.

Bit 4—Reserved: Do not write 1 to this bit.

Bit 3—Shutdown Enable (SDE): Enables or disables the host interface pin shutdown function. When this function is enabled, host interface pin functions can be halted, and the pins placed in the high-impedance state, according to the state of the HIFSD pin.

Bit 3

SDE	Description	
0	Host interface pin shutdown function disabled	(Initial value)
1	Host interface pin shutdown function enabled	

Bit 2—CS4 Enable (CS4E): Enables or disables host interface channel 4 functions in slave mode. When these functions are enabled, channel 4 pins are enabled and processing can be performed for data transfer between the slave and the host.

Bit 2

CS4E	Description	
0	Host interface pin channel 4 functions disabled	(Initial value)
1	Host interface pin channel 4 functions enabled	

Bit 1—CS3 Enable (CS3E): Enables or disables host interface channel 3 functions in slave mode. When these functions are enabled, channel 3 pins are enabled and processing can be performed for data transfer between the slave and the host.

Bit 1

CS3E	Description	
0	Host interface pin channel 3 functions disabled	(Initial value)
1	Host interface pin channel 3 functions enabled	

Bit 0—Host Interface Enable Bit (HI12E): Enables or disables host interface functions in single-chip mode. When the host interface functions are enabled, slave mode is entered and processing is performed for data transfer between the slave and host.

Bit 0

HI12E	Description	
0	Host interface functions are disabled	(Initial value)
1	Host interface functions are enabled	

18.2.3 **Host Interface Control Register (HICR)**

HICR

Bit	7	6	5	4	3	2	1	0
	_	_	_	_	_	IBFIE2	IBFIE1	FGA20E
Initial value	1	1	1	1	1	0	0	0
Slave Read/Write	_	_	_	_	_	R/W	R/W	R/W
Host Read/Write	_	_	_	_	_	_	_	_

HICR2

Bit	7	6	5	4	3	2	1	0
	_	_	_	_	_	IBFIE4	IBFIE3	_
Initial value	1	1	1	1	1	0	0	0
Slave Read/Write	_	_	_	_	_	R/W	R/W	_
Host Read/Write	_	_	_	_	_	_	_	_

HICR is an 8-bit readable/writable register which controls host interface channel 1 and 2 interrupts and the fast A20 gate function. HICR2 is an 8-bit readable/writable register which controls host interface channel 3 and 4 interrupts. HICR and HICR2 are initialized to H'F8 by a reset and in hardware standby mode.

Bits 7 to 3—Reserved: These bits cannot be modified and are always read as 1.

HICR Bits 2 and 1—Input Data Register Full Interrupt Enable 2 and 1 (IBFIE2, IBFIE1) HICR2 Bits 2 and 1—Input Data Register Full Interrupt Enable 4 and 3 (IBFIE4, IBFIE3) These bits enable or disable the IBF1, IBF2, IBF3, and IBF4 interrupts to the internal CPU.



HICR2 Bit 2	HICR2 Bit 1	HICR Bit 2	HICR Bit 1	
IBFIE4	IBFIE3	IBFIE2	IBFIE1	Description
_	_	_	0	Input data register (IDR1) reception completed interrupt request disabled (Initial value)
_	_	_	1	Input data register (IDR1) reception completed interrupt request enabled
_	_	0	_	Input data register (IDR2) reception completed interrupt request disabled (Initial value)
_	_	1	_	Input data register (IDR2) reception completed interrupt request enabled
_	0	_	_	Input data register (IDR3) reception completed interrupt request disabled (Initial value)
_	1	_	_	Input data register (IDR3) reception completed interrupt request enabled
0	_	_	_	Input data register (IDR4) reception completed interrupt request disabled (Initial value)
1	_	_	_	Input data register (IDR4) reception completed interrupt request enabled

HICR Bit 0—Fast A20 Gate Function Enable (FGA20E): Enables or disables the fast A20 gate function. When the fast A20 gate is disabled, the normal A20 gate can be implemented byte firmware operation of the P81 output.

HICR Bit 0

FGA20E	Description	
0	Fast A20 gate function disabled	(Initial value)
1	Fast A20 gate function enabled	

HICR2 Bit 0—Reserved: Do not set this bit to 1.

18.2.4 **Input Data Register 1 (IDR1)**

Bit	7	6	5	4	3	2	1	0
	IDR7	IDR6	IDR5	IDR4	IDR3	IDR2	IDR1	IDR0
Initial value	_	_	_	_	_	_	_	_
Slave Read/Write	R	R	R	R	R	R	R	R
Host Read/Write	W	W	W	W	W	W	W	W

IDR1 is an 8-bit read-only register to the slave processor, and an 8-bit write-only register to the host processor. When \overline{CSn} (n = 1 to 4) is low, information on the host data bus is written into IDRn at the rising edge of \overline{IOW} . The HA0 state is also latched into the C/\overline{D} bit in STRn to indicate whether the written information is a command or data.

The initial values of IDR1 after a reset and in standby mode are undetermined.

18.2.5 **Output Data Register 1 (ODR)**

Bit	7	6	5	4	3	2	1	0
	ODR7	ODR6	ODR5	ODR4	ODR3	ODR2	ODR1	ODR0
Initial value	_	_	_	_	_	_	_	_
Slave Read/Write	R/W							
Host Read/Write	R	R	R	R	R	R	R	R

ODR1 is an 8-bit readable/writable register to the slave processor, and an 8-bit read-only register to the host processor. The ODRn contents are output on the host data bus when HA0 is low, $\overline{\text{CSn}}$ (n = 1 to 4) is low, and \overline{IOR} is low.

The initial values of ODR1 after a reset and in standby mode are undetermined.



18.2.6 Status Register (STR)

Bit	7	6	5	4	3	2	1	0
	DBU	DBU	DBU	DBU	C/D	DBU	IBF	OBF
Initial value	0	0	0	0	0	0	0	0
Slave Read/Write	R/W	R/W	R/W	R/W	R	R/W	R	R/(W)*
Host Read/Write	R	R	R	R	R	R	R	R

Note: * Only 0 can be written, to clear the flag.

STRn (n = 1 to 4) is an 8-bit register that indicates status information during host interface processing. Bits 3, 1, and 0 are read-only bits to both the host and slave processors.

STR is initialized to H'00 by a reset and in hardware standby mode.

Bits 7 to 4 and Bit 2—Defined by User (DBU): The user can use these bits as necessary.

Bit 3—Command/Data (C/\overline{D}) : Receives the HA0 input when the host processor writes to IDR1, and indicates whether IDR1 contains data or a command.

Bit 3

C/D	Description	
0	Contents of input data register (IDR1) are data	(Initial value)
1	Contents of input data register (IDR1) are a command	

Bit 1—Input Buffer Full (IBF): Set to 1 when the host processor writes to IDR1. This bit is an internal interrupt source to the slave processor. IBF is cleared to 0 when the slave processor reads IDR1.

The IBF flag setting and clearing conditions are different when the fast A20 gate is used. For details see table 18.7.

Bit 1

IBF	Description					
0	[Clearing condition]					
	When the slave processor reads IDR	(Initial value)				
1	[Setting condition]					
	When the host processor writes to IDR					

Bit 0—Output Buffer Full (OBF): Set to 1 when the slave processor writes to ODR1. Cleared to 0 when the host processor reads ODR.

Bit 0

OBF	Description
0	[Clearing condition]
	When the host processor reads ODR or the slave writes 0 in the OBF bit (Initial value)
1	[Setting condition]
	When the slave processor writes to ODR

Table 18.3 shows the conditions for setting and clearing the STR flags.

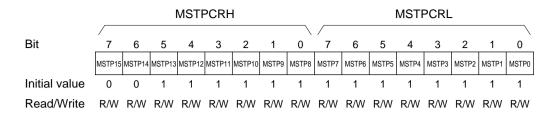
Table 18.3 Set/Clear Timing for STR Flags

Flag	Setting Condition	Clearing Condition
C/D	Rising edge of host's write signal (IOW) when HA0 is high	Rising edge of host's write signal (IOW) when HA0 is low
IBF*	Rising edge of host's write signal (IOW) when writing to IDR1	Falling edge of slave's internal read signal (RD) when reading IDR1
OBF	Falling edge of slave's internal write signal (WR) when writing to ODR1	Rising edge of host's read signal (IOR) when reading ODR1

The IBF flag setting and clearing conditions are different when the fast A20 gate is Note: used. For details see table 18.7.



18.2.7 Module Stop Control Register (MSTPCR)



MSTPCR, comprising two 8-bit readable/writable registers, performs module stop mode control. When the MSTP2 bit is set to 1, the host interface halts and enters module stop mode. See section 25.5, Module Stop Mode, for details.

MSTPCR is initialized to H'3FFF by a reset and in hardware standby mode. It is not initialized in software standby mode.

MSTPCRL Bit 2—Module Stop (MSTP2): Specifies host interface module stop mode.

MSTPCRL Bit 2

MSTP2	Description	
0	Host interface module stop mode is cleared	
1	Host interface module stop mode is set	(Initial value)

18.3 Operation

18.3.1 Host Interface Activation

The HIF (slave mode) is activated by setting the HI12E bit (bit 0) in SYSCR2 to 1 in single-chip mode. When the HIF (slave mode) is activated, all related I/O ports (data port 3, control ports 8 and 9, and host interrupt request port 4) become dedicated host interface ports. Setting the CS3E bit and CS4E bit to 1 enables the number of HIF channels to be extended to a four, and makes the channel 3 and 4 related I/O port (part of port B for control and host interrupt requests) a dedicated host interface port.

Table 18.4 shows HIF host interface channel selection and pin operation.

Table 18.4 Host Interface Channel Selection and Pin Operation

HI12E	CS2E	CS3E	CS4E	Operation		
0	_	_	_	Host interface functions halted		
1	0	0	0	Host interface channel 1 only operating		
				Operation of channels 2 to 4 halted		
				(No operation as $\overline{\text{CS2}}$ or $\overline{\text{ECS2}}$, $\overline{\text{CS3}}$, and $\overline{\text{CS4}}$ inputs. Pins P43, P81, P90, and PB0 to PB3 operate as I/O ports.)		
			1	Host interface channel 1 and 4 functions operating		
				Operation of channels 2 and 3 halted		
				(No operation as $\overline{\text{CS2}}$ or $\overline{\text{ECS2}}$ and $\overline{\text{CS3}}$ inputs. Pins P43, P81, P90, PB0, and PB2 operate as I/O ports.)		
		1	0	Host interface channel 1 and 3 functions operating		
				Operation of channels 2 and 4 halted		
				(No operation as $\overline{\text{CS2}}$ or $\overline{\text{ECS2}}$ and $\overline{\text{CS4}}$ inputs. Pins P43, P81, P90, PB1, and PB3 operate as I/O ports.)		
			1	Host interface channel 1, 3, and 4 functions operating		
				Operation of channel 2 halted		
				(No operation as $\overline{\text{CS2}}$ or $\overline{\text{ECS2}}$ input. Pins P43, P81, and P90 operate as I/O ports.)		
	1	0	0	Host interface channel 1 and 2 functions operating		
				Operation of channels 3 and 4 halted		
				(No operation as $\overline{\text{CS3}}$ and $\overline{\text{CS4}}$ inputs. Pins PB0 to PB3 operate as I/O ports.)		
			1	Host interface channel 1, 2, and 4 functions operating		
				Operation of channel 3 halted		
				(No operation as $\overline{\text{CS3}}$ input. Pins PB0 and PB2 operate as I/O ports.)		
		1	0	Host interface channel 1 to 3 functions operating		
				Operation of channel 4 halted		
				(No operation as $\overline{\text{CS4}}$ input. Pins PB1 and PB3 operate as I/O ports.)		
			1	Host interface channel 1 to 4 functions operating		

For host read/write timing, see section 26.7.5, Timing of On-Chip Supporting Modules.



18.3.2 Control States

Table 18.5 shows host interface operations from the HIF host, and slave operation.

Table 18.5 Host Interface Operations from HIF Host, and Slave Operation

Other than CSn	CSn	ĪŌR	ĪŌW	HA0	Operation
1	0	0	0	0	Setting prohibited
				1	Setting prohibited
			1	0	Data read from output data register n (ODRn)
				1	Status read from status register n (STRn)
		1	0	0	Data written to input data register n (IDRn)
				1	Command written to input data register n (IDRn)
			1	0	Idle state
				1	Idle state

Note: n = 1 to 4

18.3.3 A20 Gate

The A20 gate signal can mask address A20 to emulate an addressing mode used by personal computers with an 8086*-family CPU. In slave mode, a regular-speed A20 gate signal can be output under firmware control. Fast A20 gate output is enabled by setting the FGA20E bit (bit 0) to 1 in HICR (H'FFF0).

Note: * Intel microprocessor.

Regular A20 Gate Operation

Output of the A20 gate signal can be controlled by an H'D1 command followed by data. When the slave processor receives data, it normally uses an interrupt routine activated by the IBF1 interrupt to read IDR1. If the data follows an H'D1 command, software copies bit 1 of the data and outputs it at the gate A20 pin.

Fast A20 Gate Operation

When the FGA20E bit is set to 1, P81/GA20 is used for output of a fast A20 gate signal. Bit P81DDR must be set to 1 to assign this pin for output. The initial output from this pin will be a logic 1, which is the initial value. Afterward, the host processor can manipulate the output from

this pin by sending commands and data. This function is available only when register IDR1 is accessed using $\overline{CS1}$. Slave logic decodes the commands input from the host processor. When an H'D1 host command is detected, bit 1 of the data following the host command is output from the GA20 output pin. This operation does not depend on firmware or interrupts, and is faster than the regular processing using interrupts. Table 18.6 lists the conditions that set and clear GA20 (P81). Figure 18.2 shows the GA20 output in flowchart form. Table 18.7 indicates the GA20 output signal values.

Table 18.6 GA20 (P81) Set/Clear Timing

Pin Name	Setting Condition	Clearing Condition
GA20 (P81)	(IOW) when bit 1 of the written data is 1 and the data follows an H'D1 host command	Rising edge of the host's write signal (IOW) when bit 1 of the written data is 0 and the data follows an H'D1 host command
		Also, when bit FGA20E in HICR is cleared to 0

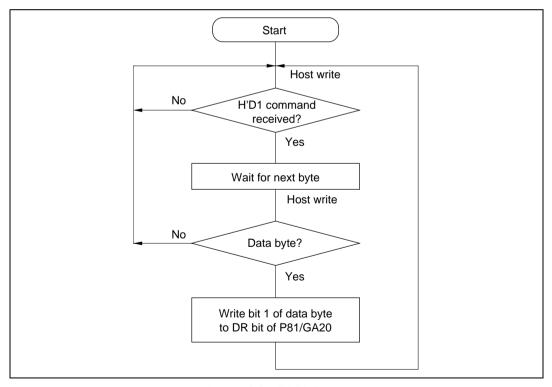


Figure 18.2 GA20 Output

Table 18.7 Fast A20 Gate Output Signal

HA0	Data/Command	Internal CPU Interrupt Flag	GA20 (P81)	Remarks
1	H'D1 command	0	Q	Turn-on sequence
0	1 data*1	0	1	
1	H'FF command	0	Q (1)	
1	H'D1 command	0	Q	Turn-off sequence
0	0 data ^{*2}	0	0	
1	H'FF command	0	Q (0)	
1	H'D1 command	0	Q	Turn-on sequence
0	1 data*1	0	1	(abbreviated form)
1/0	Command other than H'FF	1	Q (1)	
	and H'D1			
1	H'D1 command	0	Q	Turn-off sequence
0	0 data ^{*2}	0	0	(abbreviated form)
1/0	Command other than H'FF	1	Q (0)	
	and H'D1			
1	H'D1 command	0	Q	Cancelled sequence
1	Command other than H'D1	1	Q	
1	H'D1 command	0	Q	Retriggered sequence
1	H'D1 command	0	Q	
1	H'D1 command	0	Q	Consecutively executed
0	Any data	0	1/0	sequences
1	H'D1 command	0	Q(1/0)	

- Notes: 1. Arbitrary data with bit 1 set to 1.
 - 2. Arbitrary data with bit 1 cleared to 0.

18.3.4 **Host Interface Pin Shutdown Function**

Host interface output can be placed in the high-impedance state according to the state of the HIFSD pin. Setting the SDE bit to 1 in the SYSCR2 register enables the HIFSD pin is slave mode. The HIF constantly monitors the HIFSD pin, and when this pin goes low, places the host interface output pins (HIRQ1, HIRQ11, HIRQ12, HIRQ3, HIRQ4, and GA20) in the high-impedance state. At the same time, the host interface input pins ($\overline{CS1}$, $\overline{CS2}$ or $\overline{ECS2}$, $\overline{CS3}$, $\overline{CS4}$, \overline{IOW} , \overline{IOR} , and HA0) are disabled (fixed at the high input state internally) regardless of the pin states, and the signals of the multiplexed functions of these pins (input block) are similarly fixed internally. As a result, the host interface I/O pins (HDB7 to HDB0) also go to the high-impedance state.

This state is maintained while the HIFSD pin is low, and when the HIFSD pin returns to the highlevel state, the pins are restored to their normal operation as host interface pins.

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Table 18.8 shows the scope of HIF pin shutdown in slave mode.

Table 18.8 Scope of HIF Pin Shutdown in Slave Mode

Scope of

Abbreviation	Port	Scope of Shutdown in Slave Mode	I/O	Selection Conditions
ĪOR	P93	0	Input	Slave mode
IOW	P94	0	Input	Slave mode
CS1	P95	0	Input	Slave mode
CS2	P81	Δ	Input	Slave mode and CS2E = 1 and FGA20E = 0
ECS2	P90	Δ	Input	Slave mode and CS2E = 1 and FGA20E = 1
CS3	PB2	Δ	Input	Slave mode and CS3E = 1
CS4	PB3	Δ	Input	Slave mode and CS4E = 1
HA0	P80	0	Input	Slave mode
HDB7 to HDB0	P37 to P30	0	I/O	Slave mode
HIRQ11	P43	Δ	Output	Slave mode and CS2E = 1 and P43DDR = 1
HIRQ1	P44	Δ	Output	Slave mode and P44DDR = 1
HIRQ12	P45	Δ	Output	Slave mode and P45DDR = 1
HIRQ3	PB0	Δ	Output	Slave mode and CS3E = 1 and PB0DDR = 1
HIRQ4	PB1	Δ	Output	Slave mode and CS4E = 1 and PB1DDR = 1
GA20	P81	Δ	Output	Slave mode and FGA20E = 1
HIFSD	P82		Input	Slave mode and SDE = 1
				·

Legend:

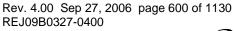
O: Pins shut down by shutdown function

The IRQ2/ADTRG input signal is also fixed in the case of P90 shutdown, the TMCI1/HSYNCI signal in the case of P43 shutdown, and the TMRI/CSYNCI in the case of P45 shutdown.

 Δ : Pins shut down only when the HIF function is selected by means of a register setting

-: Pin not shut down

Note: Slave mode: Single-chip mode and HI12E = 1





18.4 Interrupts

18.4.1 IBF1, IBF2, IBF3, IBF4

The host interface can issue two interrupt requests to the slave CPU: IBF1, IBF2, IBF3, and IBF4. They are input buffer full interrupts for input data registers IDR1, IDR2, IDR3, and IDR4 respectively. Each interrupt is enabled when the corresponding enable bit is set.

Table 18.9 Input Buffer Full Interrupts

Interrupt	Description
IBF1	Requested when IBFIE1 is set to 1 and IDR1 is full
IBF2	Requested when IBFIE2 is set to 1 and IDR2 is full
IBF3	Requested when IBFIE3 is set to 1 and IDR3 is full
IBF4	Requested when IBFIE4 is set to 1 and IDR4 is full

18.4.2 HIRQ11, HIRQ1, HIRQ12, HIRQ3, and HIRQ4

In slave mode (single-chip mode, with HI12E = 1 in SYSCR2), bits P45DR to P43DR in the port 4 data register (P4DR) and bits PB1ODR and PB0ODR in the port B data register (PBODR) can be used as host interrupt request latches

The corresponding bits in P4DR are cleared to 0 by the host processor's read signal (\overline{IOR}). If $\overline{CS1}$ and HA0 are low, when \overline{IOR} goes low and the host reads ODR1, HIRQ1 and HIRQ12 are cleared to 0. If $\overline{CS2}$ and HA0 are low, when \overline{IOR} goes low and the host reads ODR2, HIRQ11 is cleared to 0. The corresponding bit in PBODR is cleared to 0 by the host's read signal (\overline{IOR}). If $\overline{CS3}$ and HA0 are low, when \overline{IOR} goes low and the host reads ODR3, HIRQ3 is cleared to 0. If $\overline{CS4}$ and HA0 are low, when \overline{IOR} goes low and the host reads ODR4, HIRQ4 is cleared to 0. To generate a host interrupt request, normally on-chip firmware writes 1 in the corresponding bit. In processing the interrupt, the host's interrupt handling routine reads the output data register (ODR1, ODR2, ODR3, or ODR4) and this clears the host interrupt latch to 0.

Table 18.10 indicates how these bits are set and cleared. Figure 18.3 shows the processing in flowchart form.

Table 18.10 HIRQ Setting/Clearing Conditions

Host Interrupt Signal	Setting Condition	Clearing Condition
HIRQ11 (P43)	Internal CPU reads 0 from bit P43DR, then writes 1	Internal CPU writes 0 in bit P43DR, or host reads output data register 2
HIRQ1 (P44)	Internal CPU reads 0 from bit P44DR, then writes 1	Internal CPU writes 0 in bit P44DR, or host reads output data register 1
HIRQ12 (P45)	Internal CPU reads 0 from bit P45DR, then writes 1	Internal CPU writes 0 in bit P45DR, or host reads output data register 1
HIRQ3 (PB0)	Internal CPU reads 0 from bit PB0ODR, then writes 1	Internal CPU writes 0 in bit PB0ODR, or host reads output data register 3
HIRQ4 (PB1)	Internal CPU reads 0 from bit PB1ODR, then writes 1	Internal CPU writes 0 in bit PB1ODR, or host reads output data register 4

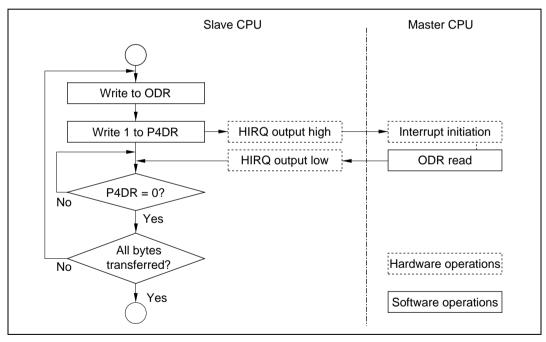


Figure 18.3 HIRQ Output Flowchart (Example of Channels 1 and 2)

HIRQ Setting/Clearing Contention

If there is contention between a P4DR or PBODR read/write by the CPU and P4DR (HIRQ11, HIRQ1, HIRQ12) or PBODR (HIRQ3, HIRQ4) clearing by the host, clearing by the host is held pending during the P4DR or PBODR read/write by the CPU. P4DR or PBODR clearing is executed after completion of the read/write.

18.5 Usage Note

The following points require attention when using the host interface.

(1) Host and slave transmission/reception procedures

The host interface provides buffering of asynchronous data from the host and slave processors, but an interface protocol must be followed to implement necessary functions and avoid data contention. For example, if the host and slave processors try to access the same input or output data register simultaneously, the data will be corrupted. Interrupts can be used to design a simple and effective protocol.

(2) Preventing data contention on the HDB

When the HIF function is used (HI12E = 1 in SYSCR2) and channel 3 or channel 4 has been set as deselected (CS3E = 0 or CS4E = 0 in SYSCR2), apply either of the following usage conditions.

- 1. Ensure that the $\overline{\text{CS}}$ pin for the deselected channel is fixed high.
- 2. Do not perform port B reads.

(3) Preventing through-current in pins $\overline{CS1}$ to $\overline{CS4}$

Also, if two or more of pins $\overline{CS1}$ to $\overline{CS4}$ are driven low simultaneously in attempting IDR or ODR access, signal contention will occur within the chip, and a through-current may result. This usage must therefore be avoided.

Section 19 D/A Converter

19.1 Overview

This LSI have an on-chip D/A converter module with two channels.

19.1.1 Features

Features of the D/A converter module are listed below.

- Eight-bit resolution
- Two-channel output
- Maximum conversion time: 10 µs (with 20-pF load capacitance)
- Output voltage: 0 V to AV_{ref}
- D/A output retention in software standby mode

19.1.2 Block Diagram

Figure 19.1 shows a block diagram of the D/A converter.

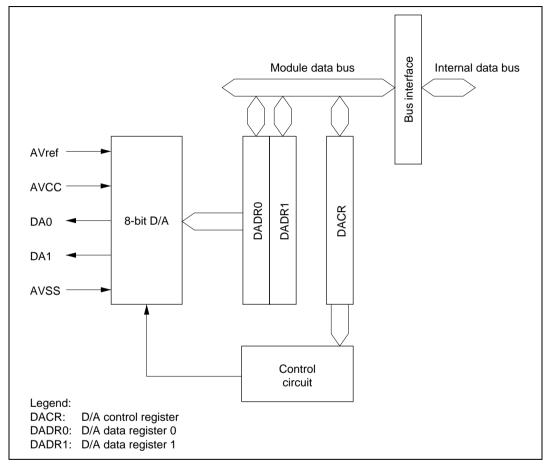


Figure 19.1 Block Diagram of D/A Converter

19.1.3 Input and Output Pins

Table 19.1 lists the input and output pins used by the D/A converter module.

Table 19.1 Input and Output Pins of D/A Converter Module

Name	Abbreviation	I/O	Function
Analog supply voltage	AVCC	Input	Power supply for analog circuits
Analog ground	AVSS	Input	Ground and reference voltage for analog circuits
Analog output 0	DA0	Output	Analog output channel 0
Analog output 1	DA1	Output	Analog output channel 1
Reference voltage pin	AVref	Input	Reference voltage for analog circuits

19.1.4 Register Configuration

Table 19.2 lists the registers of the D/A converter module.

Table 19.2 D/A Converter Registers

Name	Abbreviation	R/W	Initial Value	Address*
D/A data register 0	DADR0	R/W	H'00	H'FFF8
D/A data register 1	DADR1	R/W	H'00	H'FFF9
D/A control register	DACR	R/W	H'1F	H'FFFA
Module stop control	MSTPCRH	R/W	H'3F	H'FF86
register	MSTPCRL	R/W	H'FF	H'FF87

Note: * Lower 16 bits of the address.

19.2 **Register Descriptions**

19.2.1 D/A Data Registers 0 and 1 (DADR0, DADR1)

Bit	7	6	5	4	3	2	1	0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W							

D/A data registers 0 and 1 (DADR0 and DADR1) are 8-bit readable/writable registers that store data to be converted. When analog output is enabled, the value in the D/A data register is converted and output continuously at the analog output pin.

The D/A data registers are initialized to H'00 by a reset and in hardware standby mode.

D/A Control Register (DACR) 19.2.2

Bit	7	6	5	4	3	2	1	0	
	DAOE1	DAOE0	DAE	_	_	_	_	_	
Initial value	0	0	0	1	1	1	1	1	
Read/Write	R/W	R/W	R/W			_			

DACR is an 8-bit readable/writable register that controls the operation of the D/A converter module.

DACR is initialized to H'1F by a reset and in hardware standby mode.

Bit 7—D/A Output Enable 1 (DAOE1): Controls D/A conversion and analog output.

Bit 7

DAOE1	Description	
0	Analog output DA1 is disabled	(Initial value)
1	D/A conversion is enabled on channel 1. Analog output DA1 is enable	ed



Bit 6—D/A Output Enable 0 (DAOE0): Controls D/A conversion and analog output.

Bit 6

DAOE0	Description	
0	Analog output DA0 is disabled	(Initial value)
1	D/A conversion is enabled on channel 0. Analog output DA0 is ena	bled

Bit 5—D/A Enable (DAE): Controls D/A conversion, in combination with bits DAOE0 and DAOE1. D/A conversion is controlled independently on channels 0 and 1 when DAE = 0. Channels 0 and 1 are controlled together when DAE = 1.

Output of the converted results is always controlled independently by DAOE0 and DAOE1.

Bit 7	Bit 6	Bit 5	
DAOE1	DAOE0	DAE	D/A conversion
0	0	*	Disabled on channels 0 and 1
	1	0	Enabled on channel 0 Disabled on channel 1
		1	Enabled on channels 0 and 1
1	0	0	Disabled on channel 0 Enabled on channel 1
		1	Enabled on channels 0 and 1
	1	*	Enabled on channels 0 and 1

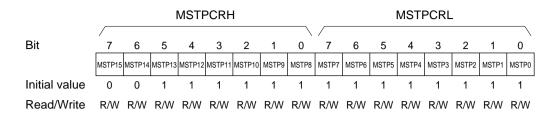
Legend:

If the chip enters software standby mode while D/A conversion is enabled, the D/A output is retained and the analog power supply current is the same as during D/A conversion. If it is necessary to reduce the analog power supply current in software standby mode, disable D/A output by clearing the DAOE0, DAOE1, and DAE bits to 0.

Bits 4 to 0—Reserved: These bits cannot be modified and are always read as 1.

^{*:} Don't care

19.2.3 Module Stop Control Register (MSTPCR)



MSTPCR, comprising two 8-bit readable/writable registers, performs module stop mode control. When the MSTP10 bit is set to 1, the D/A converter halts and enters module stop mode at the end of the bus cycle. See section 25.5, Module Stop Mode, for details.

MSTPCR is initialized to H'3FFF by a reset and in hardware standby mode. It is not initialized in software standby mode.

MSTPCRH Bit 2—Module Stop (MSTP10): Specifies D/A converter module stop mode.

MSTPCRH Bit 2

MSTP10	 Description	
0	D/A converter module stop mode is cleared	
1	D/A converter module stop mode is set	(Initial value)

19.3 Operation

The D/A converter module has two built-in D/A converter circuits that can operate independently.

D/A conversion is performed continuously whenever enabled by the D/A control register (DACR). When a new value is written in DADR0 or DADR1, conversion of the new value begins immediately. The converted result is output by setting the DAOE0 or DAOE1 bit to 1.

An example of conversion on channel 0 is given next. Figure 19.2 shows the timing.

- Software writes the data to be converted in DADR0.
- D/A conversion begins when the DAOE0 bit in DACR is set to 1. After the elapse of the
 conversion time, analog output appears at the DA0 pin. The output value is AVref × (DADR
 value)/256.

This output continues until a new value is written in DADR0 or the DAOE0 bit is cleared to 0.

- If a new value is written in DADR0, conversion begins immediately. Output of the converted result begins after the conversion time.
- When the DAOE0 bit is cleared to 0, DA0 becomes an input pin.

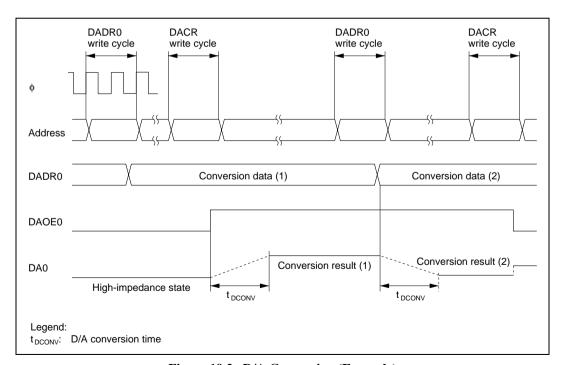


Figure 19.2 D/A Conversion (Example)

Section 20 A/D Converter

20.1 Overview

This LSI incorporate a 10-bit successive-approximations A/D converter that allows up to eight analog input channels to be selected.

In addition to the eight analog input channels, up to 16 channels of digital input can be selected for A/D conversion. Since the conversion precision falls when digital input is selected, digital input is ideal for use by a comparator identifying multi-valued inputs, for example.

20.1.1 Features

A/D converter features are listed below.

- 10-bit resolution
- Eight (analog) or 16 (digital) input channels
- Settable analog conversion voltage range
 - The analog conversion voltage range is set using the reference power supply voltage pin (AVref) as the analog reference voltage
- High-speed conversion
 - Minimum conversion time: 6.7 µs per channel (at 20-MHz operation)
- Choice of single mode or scan mode
 - Single mode: Single-channel A/D conversion
 - Scan mode: Continuous A/D conversion on 1 to 4 channels
- Four data registers
 - Conversion results are held in a 16-bit data register for each channel
- Sample and hold function
- Three kinds of conversion start
 - Choice of software or timer conversion start trigger (8-bit timer), or ADTRG pin
- A/D conversion end interrupt generation
 - An A/D conversion end interrupt (ADI) request can be generated at the end of A/D conversion

20.1.2 Block Diagram

Figure 20.1 shows a block diagram of the A/D converter.

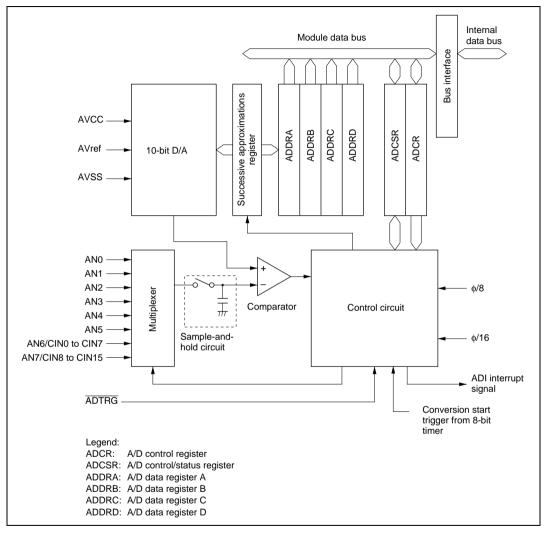


Figure 20.1 Block Diagram of A/D Converter

20.1.3 Pin Configuration

Table 20.1 summarizes the input pins used by the A/D converter.

The AVCC and AVSS pins are the power supply pins for the analog block in the A/D converter.

Table 20.1 A/D Converter Pins

Pin Name	Symbol	I/O	Function
Analog power supply pin	AVCC	Input	Analog block power supply
Analog ground pin	AVSS	Input	Analog block ground and A/D conversion reference voltage
Reference power supply pin	AVref	Input	A/D conversion reference voltage
Analog input pin 0	AN0	Input	Analog input channel 0
Analog input pin 1	AN1	Input	Analog input channel 1
Analog input pin 2	AN2	Input	Analog input channel 2
Analog input pin 3	AN3	Input	Analog input channel 3
Analog input pin 4	AN4	Input	Analog input channel 4
Analog input pin 5	AN5	Input	Analog input channel 5
Analog input pin 6	AN6	Input	Analog input channel 6
Analog input pin 7	AN7	Input	Analog input channel 7
A/D external trigger input pin	ADTRG	Input	External trigger input for starting A/D conversion
Expansion A/D input pins 0 to 15	CIN0 to CIN15	Input	Expansion A/D conversion input (digital input pin) channels 0 to 15

20.1.4 **Register Configuration**

Table 20.2 summarizes the registers of the A/D converter.

Table 20.2 A/D Converter Registers

Name	Abbreviation	R/W	Initial Value	Address*1
A/D data register AH	ADDRAH	R	H'00	H'FFE0
A/D data register AL	ADDRAL	R	H'00	H'FFE1
A/D data register BH	ADDRBH	R	H'00	H'FFE2
A/D data register BL	ADDRBL	R	H'00	H'FFE3
A/D data register CH	ADDRCH	R	H'00	H'FFE4
A/D data register CL	ADDRCL	R	H'00	H'FFE5
A/D data register DH	ADDRDH	R	H'00	H'FFE6
A/D data register DL	ADDRDL	R	H'00	H'FFE7
A/D control/status register	ADCSR	R/(W)*2	H'00	H'FFE8
A/D control register	ADCR	R/W	H'3F	H'FFE9
Module stop control register	MSTPCRH	R/W	H'3F	H'FF86
	MSTPCRL	R/W	H'FF	H'FF87
Keyboard comparator control register	KBCOMP	R/W	H'00	H'FEE4

Notes: 1. Lower 16 bits of the address.

2. Only 0 can be written in bit 7, to clear the flag.

20.2 **Register Descriptions**

A/D Data Registers A to D (ADDRA to ADDRD) 20.2.1

Bit	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
	AD9	AD8	AD7	AD6	AD5	AD4	AD3	AD2	AD1	AD0	_	_	_	_	_	_
Initial value	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
Read/Write	R	R	R	R	R	R	R	R	R	R	R	R	R	R	R	R

There are four 16-bit read-only ADDR registers, ADDRA to ADDRD, used to store the results of A/D conversion.



The 10-bit data resulting from A/D conversion is transferred to the ADDR register for the selected channel and stored there. The upper 8 bits of the converted data are transferred to the upper byte (bits 15 to 8) of ADDR, and the lower 2 bits are transferred to the lower byte (bits 7 and 6) and stored. Bits 5 to 0 are always read as 0.

The correspondence between the analog input channels and ADDR registers is shown in table 20.3.

The ADDR registers can always be read by the CPU. The upper byte can be read directly, but for the lower byte, data transfer is performed via a temporary register (TEMP). For details, see section 20.3. Interface to Bus Master.

The ADDR registers are initialized to H'0000 by a reset, and in standby mode, watch mode, subactive mode, subsleep mode, and module stop mode.

Table 20.3 Analog Input Channels and Corresponding ADDR Registers

A	nalog input Channel		
Group 0	Group 1	A/D Data Register	
AN0	AN4	ADDRA	
AN1	AN5	ADDRB	
AN2	AN6 or CIN0 to CIN7	ADDRC	
AN3	AN7 or CIN8 to CIN15	ADDRD	

20.2.2 A/D Control/Status Register (ADCSR)

Analag Innut Channal

Bit	7	6	5	4	3	2	1	0
	ADF	ADIE	ADST	SCAN	CKS	CH2	CH1	CH0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/(W)*	R/W	R/W	R/W	R/W	R/W	R/W	R/W

Note: * Only 0 can be written in bit 7, to clear the flag.

ADCSR is an 8-bit readable/writable register that controls A/D conversion operations.

ADCSR is initialized to H'00 by a reset, and in standby mode, watch mode, subactive mode, subsleep mode, and module stop mode.

Bit 7—A/D End Flag (ADF): Status flag that indicates the end of A/D conversion.

Bit 7

ADF	Description					
0	[Clearing conditions] (Initial value					
	 When 0 is written in the ADF flag after reading ADF = 1 					
	 When the DTC is activated by an ADI interrupt and ADDR is read 					
1	[Setting conditions]					
	Single mode: When A/D conversion ends					
	Scan mode: When A/D conversion ends on all specified channels					

Bit 6—A/D Interrupt Enable (ADIE): Selects enabling or disabling of interrupt (ADI) requests at the end of A/D conversion.

Bit 6

ADIE	Description	
0	A/D conversion end interrupt (ADI) request is disabled	(Initial value)
1	A/D conversion end interrupt (ADI) request is enabled	

Bit 5—A/D Start (ADST): Selects starting or stopping of A/D conversion. Holds a value of 1 during A/D conversion.

The ADST bit can be set to 1 by software, a timer conversion start trigger, or the A/D external trigger input pin (ADTRG).

Bit 5

ADST	Description	
0	A/D conversion	on stopped (Initial value
1	Single mode:	A/D conversion is started. Cleared to 0 automatically when conversion on the specified channel ends
	Scan mode:	A/D conversion is started. Conversion continues sequentially on the selected channels until ADST is cleared to 0 by software, a reset, or a transition to standby mode or module stop mode

Bit 4—Scan Mode (SCAN): Selects single mode or scan mode as the A/D conversion operating mode. See section 20.4, Operation, for single mode and scan mode operation. Only set the SCAN bit while conversion is stopped.

Bit 4

SCAN	Description	
0	Single mode	(Initial value)
1	Scan mode	

Bit 3—Clock Select (CKS): Sets the A/D conversion time. Only change the conversion time while ADST = 0.

Bit 3

CKS	Description	
0	Conversion time = 266 states (max.)	(Initial value)
1	Conversion time = 134 states (max.)	

Bits 2 to 0—Channel Select 2 to 0 (CH2 to CH0): Together with the SCAN bit, these bits select the analog input channel(s).

One analog input channel can be switched to digital input.

Only set the input channel while conversion is stopped.

Group Selection	Cha	nnel Selection	Description			
CH2	CH1	CH0	Single Mode	Scan Mode		
0	0	0	AN0 (Initial value)	AN0		
		1	AN1	AN0, AN1		
	1	0	AN2	AN0 to AN2		
		1	AN3	AN0 to AN3		
1	0	0	AN4	AN4		
		1	AN5	AN4, AN5		
	1	0	AN6 or CIN0 to CIN7	AN4, AN5, AN6 or CIN0 to CIN7		
		1	AN7 or CIN8 to CIN15	AN4, AN5, AN6 or CIN0 to CIN7		
				AN7 or CIN8 to CIN15		

20.2.3 A/D Control Register (ADCR)

Bit	7	6	5	4	3	2	1	0
	TRGS1	TRGS0	_	_	_	_		_
Initial value	0	0	1	1	1	1	1	1
Read/Write	R/W	R/W	_	_	_	_	_	_

ADCR is an 8-bit readable/writable register that enables or disables external triggering of A/D conversion operations.

ADCR is initialized to H'3F by a reset, and in standby mode, watch mode, subactive mode, subsleep mode, and module stop mode.

Bits 7 and 6—Timer Trigger Select 1 and 0 (TRGS1, TRGS0): These bits select enabling or disabling of the start of A/D conversion by a trigger signal. Only set bits TRGS1 and TRGS0 while conversion is stopped.

Bit 7	Bit 6	
TRGS1	TRGS0	Description
0	0	Start of A/D conversion by external trigger is disabled (Initial value)
	1	Start of A/D conversion by external trigger is disabled
1	0	Start of A/D conversion by external trigger (8-bit timer) is enabled
	1	Start of A/D conversion by external trigger pin is enabled

Bits 5 to 0—Reserved: Should always be written with 1.

Note: Some of these bits are readable/writable in products other than the HD64F2148, HD64F2147N, HD64F2144, HD64F2142R and HD6432142, however, when writing, be sure to write 1 here for software compatibility.

20.2.4	Kevboard	Comparator	Control R	Register	(KBCOMP)	١

Bit	7	6	5	4	3	2	1	0
	IrE	IrCKS2	IrCKS1	IrCKS0	KBADE	KBCH2	KBCH1	KBCH0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

KBCOMP is an 8-bit readable/writable register that controls the SCI2 IrDA function and selects the CIN input channels for A/D conversion.

KBCOMP is initialized to H'00 by a reset and in hardware standby mode.

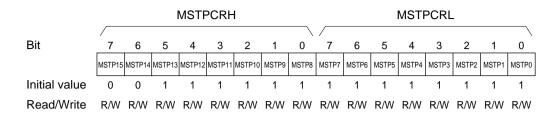
Bits 7 to 4—IrDA Control: See the description in section 15.2.11, Keyboard Comparator Control Register (KBCOMP).

Bit 3—Keyboard A/D Enable (KBADE): Selects either analog input pins (AN6, AN7) or digital input pins (CIN0 to CIN7, CIN8 to CIN15) for A/D converter channel 6 and channel 7 input.

Bits 2 to 0—Keyboard A/D Channel Select 2 to 0 (KBCH2 to KBCH0): These bits select the channels for A/D conversion from among the digital input pins. Only set the input channel while A/D conversion is stopped.

Bit 3	Bit 2	Bit 1	Bit 0	A/D Converter	A/D Converter
KBADE	KBCH2	KBCH1	KBCH0	Channel 6 Input	Channel 7 Input
0	_	_	_	AN6	AN7
1	0	0	0	CIN0	CIN8
			1	CIN1	CIN9
		1	0	CIN2	CIN10
			1	CIN3	CIN11
	1	0	0	CIN4	CIN12
			1	CIN5	CIN13
		1	0	CIN6	CIN14
			1	CIN7	CIN15

20.2.5 Module Stop Control Register (MSTPCR)



MSTPCR, comprising two 8-bit readable/writable registers, performs module stop mode control.

When the MSTP9 bit in MSTPCR is set to 1, A/D converter operation stops at the end of the bus cycle and a transition is made to module stop mode. Registers cannot be read or written to in module stop mode. For details, see section 25.5, Module Stop Mode.

MSTPCR is initialized to H'3FFF by a reset and in hardware standby mode. It is not initialized in software standby mode.

MSTPCRH Bit 1—Module Stop (MSTP9): Specifies the A/D converter module stop mode.

MSTPCRH Bit 1

MSTP9	Description	
0	A/D converter module stop mode is cleared	_
1	A/D converter module stop mode is set	(Initial value)

20.3 Interface to Bus Master

ADDRA to ADDRD are 16-bit registers, but the data bus to the bus master is only 8 bits wide. Therefore, in accesses by the bus master, the upper byte is accessed directly, but the lower byte is accessed via a temporary register (TEMP).

A data read from ADDR is performed as follows. When the upper byte is read, the upper byte value is transferred to the CPU and the lower byte value is transferred to TEMP. Next, when the lower byte is read, the TEMP contents are transferred to the CPU.

When reading ADDR, always read the upper byte before the lower byte. It is possible to read only the upper byte, but if only the lower byte is read, incorrect data may be obtained.

Figure 20.2 shows the data flow for ADDR access.

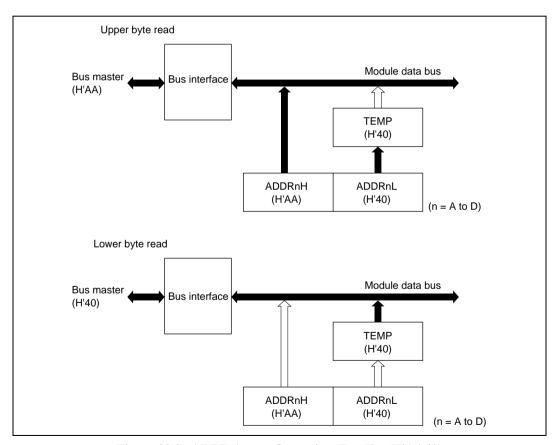


Figure 20.2 ADDR Access Operation (Reading H'AA40)

20.4 Operation

The A/D converter operates by successive approximations with 10-bit resolution. It has two operating modes: single mode and scan mode.

20.4.1 Single Mode (SCAN = 0)

Single mode is selected when A/D conversion is to be performed on a single channel only. A/D conversion is started when the ADST bit is set to 1 by software, or by external trigger input. The ADST bit remains set to 1 during A/D conversion, and is automatically cleared to 0 when conversion ends.

On completion of conversion, the ADF flag is set to 1. If the ADIE bit is set to 1 at this time, an ADI interrupt request is generated. The ADF flag is cleared by writing 0 after reading ADCSR.

When the operating mode or analog input channel must be changed during analog conversion, to prevent incorrect operation, first clear the ADST bit to 0 in ADCSR to halt A/D conversion. After making the necessary changes, set the ADST bit to 1 to start A/D conversion again. The ADST bit can be set at the same time as the operating mode or input channel is changed.

Typical operations when channel 1 (AN1) is selected in single mode are described next. Figure 20.3 shows a timing diagram for this example.

- 1. Single mode is selected (SCAN = 0), input channel AN1 is selected (CH1 = 0, CH0 = 1), the A/D interrupt is enabled (ADIE = 1), and A/D conversion is started (ADST = 1).
- 2. When A/D conversion is completed, the result is transferred to ADDRB. At the same time the ADF flag is set to 1, the ADST bit is cleared to 0, and the A/D converter becomes idle.
- 3. Since ADF = 1 and ADIE = 1, an ADI interrupt is requested.
- 4. The A/D interrupt handling routine starts.
- 5. The routine reads ADCSR, then writes 0 to the ADF flag.
- 6. The routine reads and processes the conversion result (ADDRB).
- 7. Execution of the A/D interrupt handling routine ends. After that, if the ADST bit is set to 1, A/D conversion starts again and steps 2 to 7 are repeated.



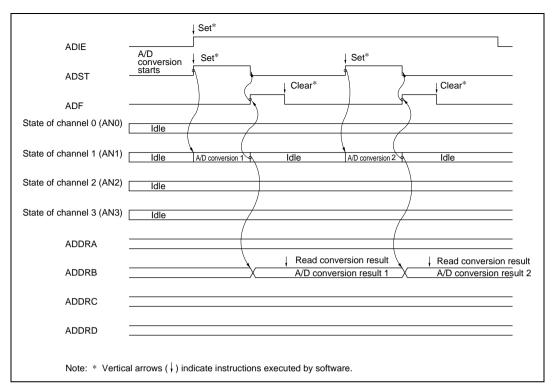


Figure 20.3 Example of A/D Converter Operation (Single Mode, Channel 1 Selected)

20.4.2 Scan Mode (SCAN = 1)

Scan mode is useful for monitoring analog inputs in a group of one or more channels. When the ADST bit is set to 1 by software, or by timer or external trigger input, A/D conversion starts on the first channel in the group (AN0 when CH2 = 0; AN4 when CH2 = 1). When two or more channels are selected, after conversion of the first channel ends, conversion of the second channel (AN1 or AN5) starts immediately. A/D conversion continues cyclically on the selected channels until the ADST bit is cleared to 0. The conversion results are transferred for storage into the ADDR registers corresponding to the channels.

When the operating mode or analog input channel must be changed during analog conversion, to prevent incorrect operation, first clear the ADST bit to 0 in ADCSR to halt A/D conversion. After making the necessary changes, set the ADST bit to 1 to start A/D conversion again. The ADST bit can be set at the same time as the operating mode or input channel is changed.

Typical operations when three channels (AN0 to AN2) are selected in scan mode are described next. Figure 20.4 shows a timing diagram for this example.

- 1. Scan mode is selected (SCAN = 1), scan group 0 is selected (CH2 = 0), analog input channels AN0 to AN2 are selected (CH1 = 1, CH0 = 0), and A/D conversion is started (ADST = 1)
- 2. When A/D conversion of the first channel (AN0) is completed, the result is transferred to ADDRA. Next, conversion of the second channel (AN1) starts automatically.
- 3. Conversion proceeds in the same way through the third channel (AN2).
- 4. When conversion of all the selected channels (AN0 to AN2) is completed, the ADF flag is set to 1 and conversion of the first channel (AN0) starts again. If the ADIE bit is set to 1 at this time, an ADI interrupt is requested after A/D conversion ends.
- 5. Steps 2 to 4 are repeated as long as the ADST bit remains set to 1. When the ADST bit is cleared to 0, A/D conversion stops. After that, if the ADST bit is set to 1, A/D conversion starts again from the first channel (AN0).



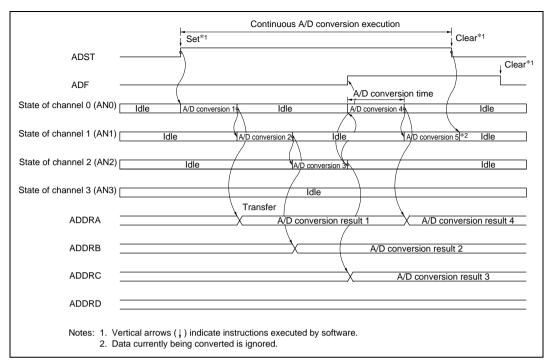


Figure 20.4 Example of A/D Converter Operation (Scan Mode, Channels AN0 to AN2 Selected)

20.4.3 Input Sampling and A/D Conversion Time

The A/D converter has a built-in sample-and-hold circuit. The A/D converter samples the analog input at a time t_D after the ADST bit is set to 1, then starts conversion. Figure 20.5 shows the A/D conversion timing. Table 20.4 indicates the A/D conversion time.

As indicated in figure 20.5, the A/D conversion time includes $t_{\scriptscriptstyle D}$ and the input sampling time. The length of $t_{\scriptscriptstyle D}$ varies depending on the timing of the write access to ADCSR. The total conversion time therefore varies within the ranges indicated in table 20.4.

In scan mode, the values given in table 20.4 apply to the first conversion time. In the second and subsequent conversions the conversion time is fixed at 256 states when CKS = 0 or 128 states when CKS = 1.

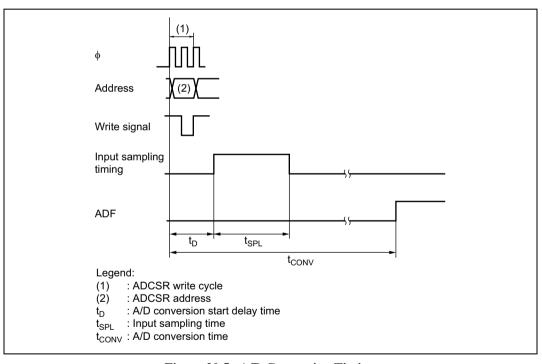


Figure 20.5 A/D Conversion Timing

Table 20.4 A/D Conversion Time (Single Mode)

			CKS =	· 0		CKS :	= 1
Item	Symbol	Min	Тур	Max	Min	Тур	Max
A/D conversion start delay	t _D	10	_	17	6	_	9
Input sampling time	t _{spl}	_	63	_	_	31	_
A/D conversion time	t _{conv}	259	_	266	131	_	134

Note: Values in the table are the number of states.

20.4.4 External Trigger Input Timing

A/D conversion can be externally triggered. When the TRGS1 and TRGS0 bits are set to 11 in ADCR, external trigger input is enabled at the ADTRG pin. A falling edge at the ADTRG pin sets the ADST bit to 1 in ADCSR, starting A/D conversion. Other operations, in both single and scan modes, are the same as when the ADST bit is set to 1 by software. Figure 20.6 shows the timing.

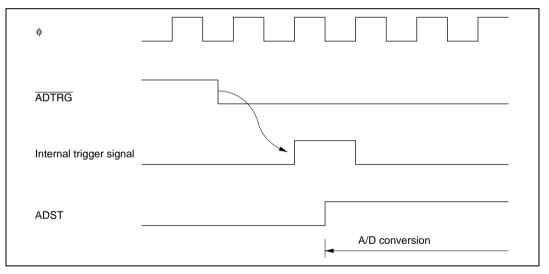


Figure 20.6 External Trigger Input Timing

20.5 Interrupts

The A/D converter generates an interrupt (ADI) at the end of A/D conversion. The ADI interrupt request can be enabled or disabled by the ADIE bit in ADCSR.

20.6 Usage Notes

The following points should be noted when using the A/D converter.

Setting Range of Analog Power Supply and Other Pins

- 1. Analog input voltage range
 - The voltage applied to the ANn analog input pins during A/D conversion should be in the range $AV_{ss} \le ANn \le AV_{ref}$ (n = 0 to 7).
- 2. Digital input voltage range

The voltage applied to the CINn digital input pins should be in the range $AV_{ss} \le CINn \le AV_{ref}$ and $V_{ss} \le CINn \le V_{cc}$ (n = 0 to 15).

- 3. Relation between AV_{cc} , AV_{ss} and V_{cc} , V_{ss} As the relationship between AV_{cc} , AV_{ss} and V_{cc} , V_{ss} , set $AV_{ss} = V_{ss}$. If the A/D converter is not used, the AVCC and AVSS pins must on no account be left open.
- 4. Setting Range of AVref Pin:

The reference voltage supplied via the AVref pin should be in the range $AV_{ref} \le AV_{CC}$.

If conditions 1 to 4 above are not met, the reliability of the device may be adversely affected.

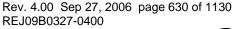
Notes on Board Design

In board design, digital circuitry and analog circuitry should be as mutually isolated as possible, and layout in which digital circuit signal lines and analog circuit signal lines cross or are in close proximity should be avoided as far as possible. Failure to do so may result in incorrect operation of the analog circuitry due to inductance, adversely affecting A/D conversion values.

Also, digital circuitry must be isolated from the analog input signals (AN0 to AN7), analog reference power supply (AVref), and analog power supply (AVCC) by the analog ground (AVSS). Also, the analog ground (AVSS) should be connected at one point to a stable digital ground (VSS) on the board.

Notes on Noise Countermeasures

A protection circuit connected to prevent damage due to an abnormal voltage such as an excessive surge at the analog input pins (AN0 to AN7) or analog reference power supply pin (AVref) should be connected between AVCC and AVSS as shown in figure 20.7.





Also, the bypass capacitors connected to AVCC, AVref and the filter capacitor connected to AN0 to AN7 must be connected to AVSS.

If a filter capacitor is connected as shown in figure 20.7, the input currents at the analog input pins (AN0 to AN7) are averaged, and so an error may arise. Also, when A/D conversion is performed frequently, as in scan mode, if the current charged and discharged by the capacitance of the sample-and-hold circuit in the A/D converter exceeds the current input via the input impedance (R_{in}) , an error will arise in the analog input pin voltage. Careful consideration is therefore required when deciding the circuit constants.

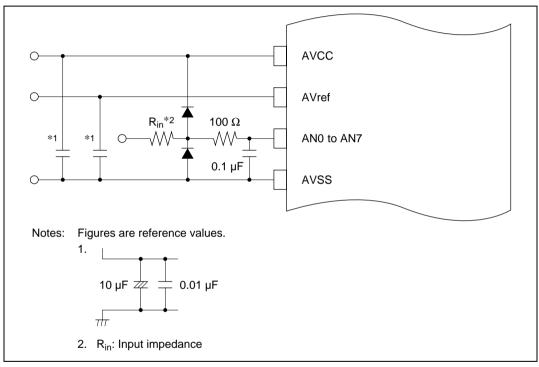


Figure 20.7 Example of Analog Input Protection Circuit

Table 20.5 Analog Pin Ratings

Item	Min	Max	Unit
Analog input capacitance	_	20	pF
Permissible signal source impedance	_	10*	kΩ

Note: * When V_{cc} = 4.0 to 5.5 V and $\phi \le$ 12 MHz.

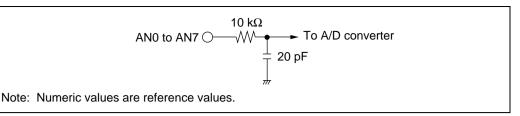


Figure 20.8 Analog Input Pin Equivalent Circuit

A/D Conversion Precision Definitions

The A/D conversion precision in this LSI is defined as follows.

Resolution

The number of A/D converter digital output codes

Offset error

The deviation of the analog input voltage value from the ideal A/D conversion characteristic when the digital output changes from the minimum voltage value B'0000000000 (H'000) to B'0000000001 (H'001) (see figure 20.10).

Full-scale error

The deviation of the analog input voltage value from the ideal A/D conversion characteristic when the digital output changes from B'1111111110 (H'3FE) to B'111111111 (H'3FF) (see figure 20.11).

Quantization error

The deviation inherent in the A/D converter, given by 1/2 LSB (see figure 20.9).

Nonlinearity error

The error with respect to the ideal A/D conversion characteristic between the zero voltage and the full-scale voltage. Does not include the offset error, full-scale error, or quantization error.

Absolute precision

The deviation between the digital value and the analog input value. Includes the offset error, full-scale error, quantization error, and nonlinearity error.



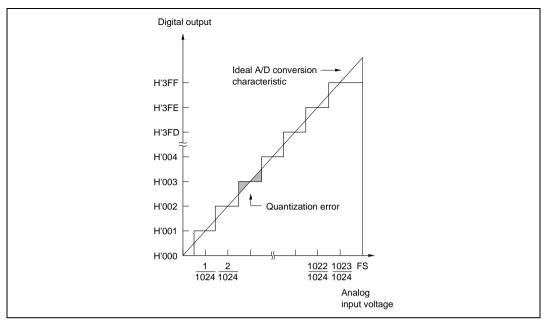


Figure 20.9 A/D Conversion Precision Definitions (1)

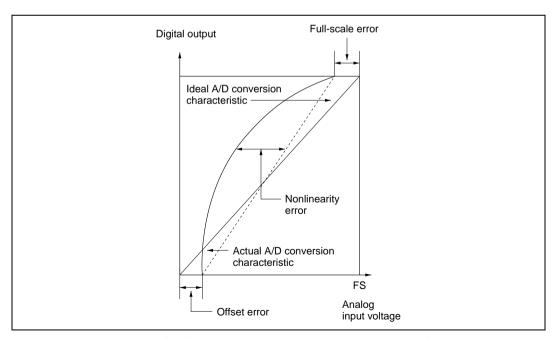


Figure 20.10 A/D Conversion Precision Definitions (2)

Permissible Signal Source Impedance

Analog input in this LSI is designed so that conversion precision is guaranteed for an input signal for which the signal source impedance is $10 \text{ k}\Omega$ (AVcc = 4.0 to 5.5 V, when $\phi \le 12 \text{ MHz}$) or less. This specification is provided to enable the A/D converter's sample-and-hold circuit input capacitance to be charged within the sampling time; if the sensor output impedance exceeds 10 k Ω (AVcc = 4.0 to 5.5 V, when $\phi \le 12$ MHz), charging may be insufficient and it may not be possible to guarantee the A/D conversion precision.

However, if a large capacitance is provided externally, the input load will essentially comprise only the internal input resistance of $10 \text{ k}\Omega$, and the signal source impedance is ignored.

But since a low-pass filter effect is obtained in this case, it may not be possible to follow an analog signal with a large differential coefficient (e.g., 5 mV/µsec or greater).

When converting a high-speed analog signal, a low-impedance buffer should be inserted.

Influences on Absolute Precision

Adding capacitance results in coupling with GND, and therefore noise in GND may adversely affect absolute precision. Be sure to make the connection to an electrically stable GND such as AV_{ss}.

Care is also required to insure that filter circuits do not communicate with digital signals on the mounting board, so acting as antennas.

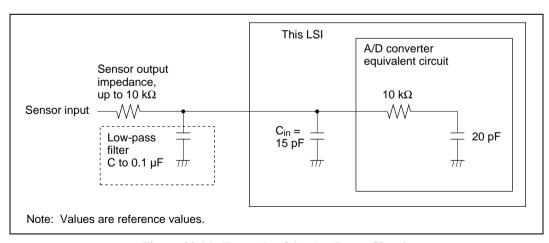


Figure 20.11 Example of Analog Input Circuit



Section 21 RAM

21.1 Overview

The H8S/2148, H8S/2144, and H8S/2143 have 4 kbytes of on-chip high-speed static RAM, and the H8S/2147, H8S/2147N, and H8S/2142 have 2 kbytes. The on-chip RAM is connected to the CPU by a 16-bit data bus, enabling both byte data and word data to be accessed in one state. This makes it possible to perform fast word data transfer.

The on-chip RAM can be enabled or disabled by means of the RAM enable bit (RAME) in the system control register (SYSCR).

21.1.1 Block Diagram

Figure 21.1 shows a block diagram of the on-chip RAM.

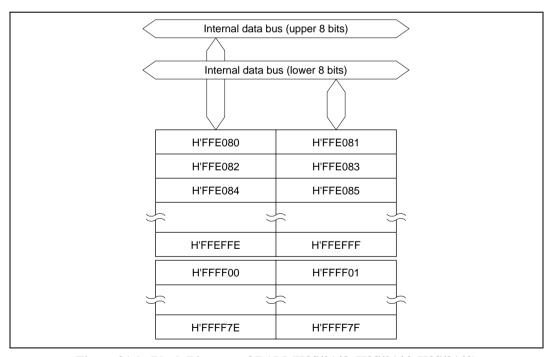


Figure 21.1 Block Diagram of RAM (H8S/2148, H8S/2144, H8S/2143)

Register Configuration 21.1.2

The on-chip RAM is controlled by SYSCR. Table 21.1 shows the register configuration.

Table 21.1 Register Configuration

Name	Abbreviation	R/W	Initial Value	Address*
System control register	SYSCR	R/W	H'09	H'FFC4

Note: Lower 16 bits of the address.

21.2 **System Control Register (SYSCR)**

Bit	7	6	5	4	3	2	1	0
	CS2E	IOSE	INTM1	INTM0	XRST	NMIEG	HIE	RAME
Initial value	0	0	0	0	1	0	0	1
Read/Write	R/W	R/W	R	R/W	R	R/W	R/W	R/W

The on-chip RAM is enabled or disabled by the RAME bit in SYSCR. For details of other bits in SYSCR, see section 3.2.2, System Control Register (SYSCR).

Bit 0—RAM Enable (RAME): Enables or disables the on-chip RAM. The RAME bit is initialized when the reset state is released. It is not initialized in software standby mode.

Bit 0

RAME	 Description	
0	On-chip RAM is disabled	
1	On-chip RAM is enabled	(Initial value)



21.3 Operation

21.3.1 Expanded Mode (Modes 1, 2, 3 (EXPE = 1))

When the RAME bit is set to 1, accesses to H8S/2148, H8S/2144, and H8S/2143 addresses H'(FF)E080 to H'(FF)EFFF and H'(FF)FF00 to H'(FF)FF7F, and H8S/2147, H8S/2147N, and H8S/2142 addresses H'(FF)E880 to H'(FF)EFFF and H'(FF)FF00 to H'(FF)FF7F, are directed to the on-chip RAM. When the RAME bit is cleared to 0, accesses to addresses H'(FF)E080 to H'(FF)EFFF and H'(FF)FF00 to H'(FF)FF7F, are directed to the off-chip address space.

Since the on-chip RAM is connected to the bus master by a 16-bit data bus, it can be written to and read in byte or word units. Each type of access is performed in one state.

Even addresses use the upper 8 bits, and odd addresses use the lower 8 bits. Word data must start at an even address.

21.3.2 Single-Chip Mode (Modes 2 and 3 (EXPE = 0))

When the RAME bit is set to 1, accesses to H8S/2148, H8S/2144, and H8S/2143 addresses H'(FF)E080 to H'(FF)EFFF and H'(FF)FF00 to H'(FF)FF7F, and H8S/2147, H8S/2147N, and H8S/2142 addresses H'(FF)E880 to H'(FF)EFFF and H'(FF)FF00 to H'(FF)FF7F, are directed to the on-chip RAM. When the RAME bit is cleared to 0, the on-chip RAM is not accessed. Undefined values are read from these bits, and writing is invalid.

Since the on-chip RAM is connected to the bus master by a 16-bit data bus, it can be written to and read in byte or word units. Each type of access is performed in one state.

Even addresses use the upper 8 bits, and odd addresses use the lower 8 bits. Word data must start at an even address.

Section 22 ROM (Mask ROM Version, H8S/2148 F-ZTAT, H8S/2147N F-ZTAT, H8S/2144 F-ZTAT, and H8S/2142 F-ZTAT)

22.1 Overview

The H8S/2148 and H8S/2144 have 128 kbytes of on-chip ROM (flash memory or mask ROM), the H8S/2143 has 96 kbytes, the H8S/2147, H8S/2147N, and H8S/2142 have 64 kbytes. The ROM is connected to the CPU by a 16-bit data bus. The CPU accesses both byte and word data in one state, enabling faster instruction fetches and higher processing speed.

The mode pins (MD1 and MD0) and the EXPE bit in MDCR can be set to enable or disable the on-chip ROM.

The flash memory versions of the H8S/2148, H8S/2147N, H8S/2144, and H8S/2142 can be erased and programmed on-board as well as with a general-purpose PROM programmer.

22.1.1 Block Diagram

Figure 22.1 shows a block diagram of the ROM.

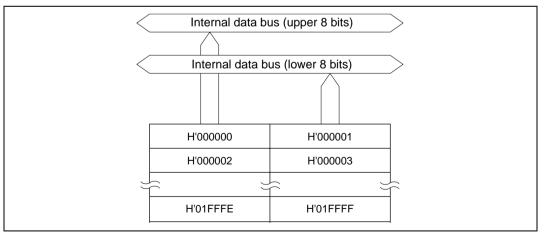


Figure 22.1 ROM Block Diagram (H8S/2148, H8S/2144)

22.1.2 **Register Configuration**

This group on-chip ROM is controlled by the operating mode and register MDCR. The register configuration is shown in table 22.1.

Table 22.1 ROM Register

Register Name	Abbreviation	R/W	Initial Value	Address*
Mode control register	MDCR	R/W	Undefined Depends on the operating mode	H'FFC5

Note: Lower 16 bits of the address.

22.2 **Register Descriptions**

22.2.1 **Mode Control Register (MDCR)**

Bit	7	6	5	4	3	2	1	0
	EXPE	_	_	_	_	_	MDS1	MDS0
Initial value	*	0	0	0	0	0	*	*
Read/Write	R/W*	_	_	_		_	R	R

Note: * Determined by the MD1 and MD0 pins.

MDCR is an 8-bit register used to set this group operating mode and monitor the current operating mode.

The EXPE bit is initialized in accordance with the mode pin states by a reset and in hardware standby mode.

Bit 7—Expanded Mode Enable (EXPE): Sets expanded mode. In mode 1, EXPE is fixed at 1 and cannot be modified. In modes 2 and 3, EXPE has an initial value of 0 and can be read or written.

Bit 7 EXPE	Description
0	Single-chip mode selected
1	Expanded mode selected

Bits 6 to 2—Reserved: These bits cannot be modified and are always read as 0.

Bits 1 and 0—Mode Select 1 and 0 (MDS1, MDS0): These bits indicate values that reflects the input levels of mode pins MD1 and MD0 (the current operating mode). Bits MDS1 and MDS0 correspond to pins MD1 and MD0, respectively. These are read-only bits, and cannot be modified. When MDCR is read, the input levels of mode pins MD1 and MD0 are latched in these bits.

22.3 **Operation**

The on-chip ROM is connected to the CPU by a 16-bit data bus, and both byte and word data is accessed in one state. Even addresses are connected to the upper 8 bits, and odd addresses to the lower 8 bits. Word data must start at an even address.

The mode pins (MD1 and MD0) and the EXPE bit in MDCR can be set to enable or disable the on-chip ROM, as shown in table 22.2.

In normal mode, the maximum amount of ROM that can be used is 56 kbytes.

Table 22.2 Operating Modes and ROM

Operating Mode

MCU Operating	CPU Operating		Mode Pins		MDCR	_	
Mode	Mode	Description	MD1	MD0	EXPE	On-Chip ROM	
Mode 1	Normal	Expanded mode with on-chip ROM disabled	0	1	1	Disabled	
Mode 2	Advanced	Single-chip mode	1	0	0	Enabled*	
	Advanced	Expanded mode with on-chip ROM enabled	-		1	_	
Mode 3	Normal	Normal Single-chip mode		1	0	Enabled	
	Normal	Expanded mode with on-chip ROM enabled			1	(max. 56 kbytes)	

Note: 128 kbytes in the H8S/2148 and H8S/2144. 96 kbytes in the H8S/2143. 64 kbytes in the H8S/2147, H8S/2147N, and H8S/2142.

22.4 Overview of Flash Memory

22.4.1 Features

The features of the flash memory are summarized below.

- Four flash memory operating modes
 - Program mode
 - Erase mode
 - Program-verify mode
 - Erase-verify mode
- Programming/erase methods

The flash memory is programmed 32 bytes at a time. Erasing is performed by block erase (in single-block units). When erasing multiple blocks, the individual blocks must be erased sequentially. Block erasing can be performed as required on 1-kbyte, 8-kbyte, 16-kbyte, 28-kbyte, and 32-kbyte blocks.

• Programming/erase times

The flash memory programming time is 10 ms (typ.) for simultaneous 32-byte programming, equivalent to 300 µs (typ.) per byte, and the erase time is 100 ms (typ.) per block.

• Reprogramming capability

The flash memory can be reprogrammed up to 100 times.

• On-board programming modes

There are two modes in which flash memory can be programmed/erased/verified on-board:

- Boot mode
- User program mode
- Automatic bit rate adjustment

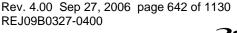
With data transfer in boot mode, the bit rate of the chip can be automatically adjusted to match the transfer bit rate of the host.

Protect modes

There are three protect modes, hardware, software, and error protect, which allow protected status to be designated for flash memory program/erase/verify operations.

Programmer mode

Flash memory can be programmed/erased in programmer mode, using a PROM programmer, as well as in on-board programming mode.





22.4.2 Block Diagram

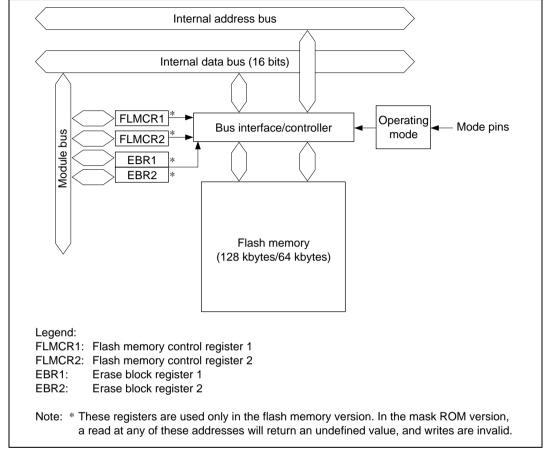


Figure 22.2 Block Diagram of Flash Memory

22.4.3 Flash Memory Operating Modes

Mode Transitions

When the mode pins are set in the reset state and a reset-start is executed, the MCU enters one of the operating modes shown in figure 22.3. In user mode, flash memory can be read but not programmed or erased.

Flash memory can be programmed and erased in boot mode, user program mode, and programmer mode.

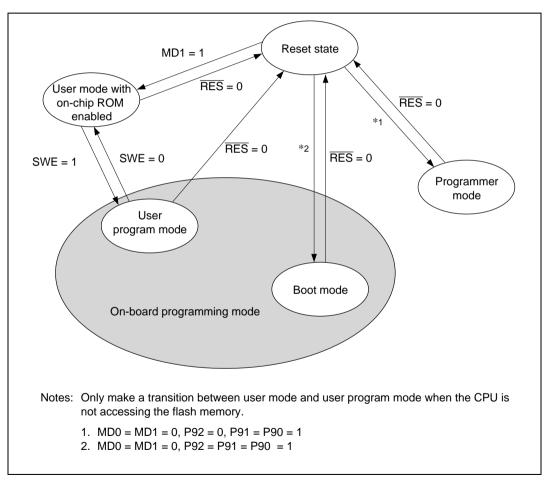


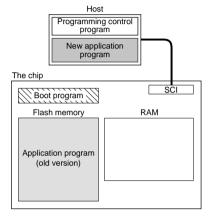
Figure 22.3 Flash Memory Mode Transitions

On-Board Programming Modes

Boot mode

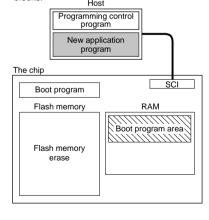
1. Initial state

The flash memory is in the erased state when the device is shipped. The description here applies to the case where the old program version or data is being rewritten. The user should prepare the programming control program and new application program beforehand in the host.



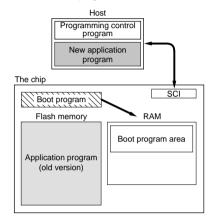
3. Flash memory initialization

The erase program in the boot program area (in RAM) is executed, and the flash memory is initialized (to H'FF). In boot mode, entire flash memory erasure is performed, without regard to blocks.



2. SCI communication check

When boot mode is entered, the boot program in the chip (originally incorporated in the chip) is started, an SCI communication check is carried out, and the boot program required for flash memory erasing is automatically transferred to the RAM boot program area.



4. Writing new application program

The programming control program transferred from the host to RAM by SCI communication is executed, and the new application program in the host is written into the flash memory.

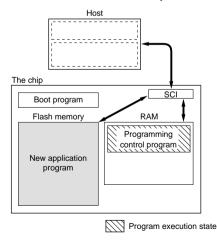
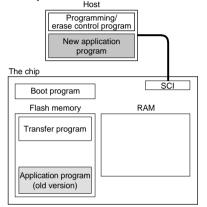


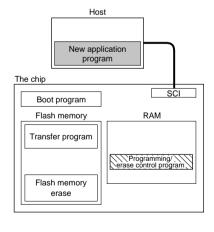
Figure 22.4 Boot Mode

User program mode

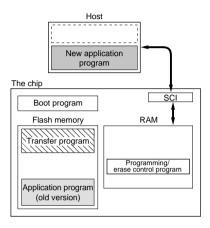
- 1. Initial state
 - (1) The program that will transfer the programming/ erase control program to on-chip RAM should be written into the flash memory by the user beforehand.
 - (2) The programming/erase control program should be prepared in the host or in the flash memory.



Flash memory initialization
 The programming/erase program in RAM is executed, and the flash memory is initialized (to H'FF). Erasing can be performed in block units, but not in byte units.



Programming/erase control program transfer Executes the transfer program in the flash memory, and transfers the programming/erase control program to RAM.



Writing new application program
 Next, the new application program in the host is written into the erased flash memory blocks. Do not write to unerased blocks.

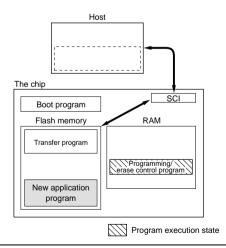


Figure 22.5 User Program Mode (Example)

Differences between Boot Mode and User Program Mode

	Boot Mode	User Program Mode
Entire memory erase	Yes	Yes
Block erase	No	Yes
Programming control program*	Program/program-verify	Program/program-verify
		Erase/erase-verify

Note: * To be provided by the user, in accordance with the recommended algorithm.

Block Configuration

The flash memory is divided into two 32-kbyte blocks (128-kbyte version only), two 8-kbyte blocks, one 16-kbyte block, one 28-kbyte block, and four 1-kbyte blocks.

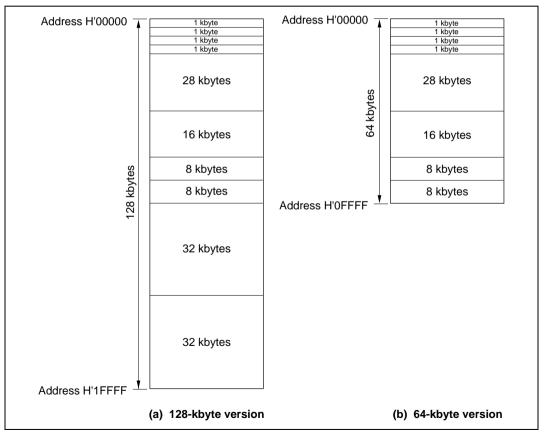


Figure 22.6 Flash Memory Block Configuration

22.4.4 Pin Configuration

The flash memory is controlled by means of the pins shown in table 22.3.

Table 22.3 Flash Memory Pins

Pin Name	Abbreviation	I/O	Function
Reset	RES	Input	Reset
Mode 1	MD1	Input	Sets MCU operating mode
Mode 0	MD0	Input	Sets MCU operating mode
Port 92	P92	Input	Sets MCU operating mode when MD1 = MD0 = 0
Port 91	P91	Input	Sets MCU operating mode when MD1 = MD0 = 0
Port 90	P90	Input	Sets MCU operating mode when MD1 = MD0 = 0
Transmit data	TxD1	Output	Serial transmit data output
Receive data	RxD1	Input	Serial receive data input

22.4.5 Register Configuration

The registers used to control the on-chip flash memory when enabled are shown in table 22.4. In order for these registers to be accessed, the FLSHE bit must be set to 1 in STCR.

Table 22.4 Flash Memory Registers

Register Name	Abbreviation	R/W	Initial Value	Address*1
Flash memory control register 1	FLMCR1*5	R/W*3	H'80	H'FF80*2
Flash memory control register 2	FLMCR2*5	R/W*3	H'00*4	H'FF81*2
Erase block register 1	EBR1*5	R/W*3	H'00*4	H'FF82*2
Erase block register 2	EBR2*5	R/W*3	H'00*4	H'FF83*2
Serial/timer control register	STCR	R/W	H'00	H'FFC3

Notes: 1. Lower 16 bits of the address.

- 2. Flash memory registers share addresses with other registers. Register selection is performed by the FLSHE bit in the serial/timer control register (STCR).
- 3. In modes in which the on-chip flash memory is disabled, a read will return H'00, and writes are invalid.
- 4. The SWE bit in FLMCR1 is not set, these registers are initialized to H'00.
- 5. FLMCR1, FLMCR2, EBR1, and EBR2 are 8-bit registers. Only byte accesses are valid for these registers, the access requiring 2 states. These registers are used only in the flash memory version. In the mask ROM version, a read at any of these addresses will return an undefined value, and writes are invalid.

22.5 Register Descriptions

22.5.1 Flash Memory Control Register 1 (FLMCR1)

Bit	7	6	5	4	3	2	1	0
	FWE	SWE	_	_	EV	PV	Е	Р
Initial value	1	0	0	0	0	0	0	0
Read/Write	R	R/W	_	_	R/W	R/W	R/W	R/W

FLMCR1 is an 8-bit register used for flash memory operating mode control. Program-verify mode or erase-verify mode is entered by setting SWE to 1. Program mode is entered by setting SWE to 1, then setting the PSU bit in FLMCR2, and finally setting the P bit. Erase mode is entered by setting SWE to 1, then setting the ESU bit in FLMCR2, and finally setting the E bit. FLMCR1 is initialized to H'80 by a reset, and in hardware standby mode, software standby mode, subactive mode, subsleep mode, and watch mode. When on-chip flash memory is disabled, a read will return H'00, and writes are invalid.

Writes to the EV and PV bits in FLMCR1 are enabled only when SWE=1; writes to the E bit only when SWE = 1, and ESU = 1; and writes to the P bit only when SWE = 1, and PSU = 1.

Bit 7—Flash Write Enable (FWE): Controls programming and erasing of the on-chip flash memory. This bit cannot be modified and is always read as 1.

Bit 6—Software Write Enable (SWE): Enables or disables flash memory programming. SWE should be set before setting bits ESU, PSU, EV, PV, E, P, and EB9 to EB0, and should not be cleared at the same time as these bits.

Bit 6

SWE	 Description	
0	Writes disabled	(Initial value)
1	Writes enabled	

Bits 5 and 4—Reserved: These bits cannot be modified and are always read as 0.

Bit 3—Erase-Verify (EV): Selects erase-verify mode transition or clearing. Do not set the SWE, ESU, PSU, PV, E, or P bit at the same time.

Bit 3

EV	Description	
0	Erase-verify mode cleared	(Initial value)
1	Transition to erase-verify mode	
	[Setting condition]	
	When SWE = 1	

Bit 2—Program-Verify (PV): Selects program-verify mode transition or clearing. Do not set the SWE, ESU, PSU, EV, E, or P bit at the same time.

Bit 2

PV	Description	
0	Program-verify mode cleared	(Initial value)
1	Transition to program-verify mode	
	[Setting condition]	
	When SWE = 1	

Bit 1—Erase (E): Selects erase mode transition or clearing. Do not set the SWE, ESU, PSU, EV, PV, or P bit at the same time.

Bit 1

E	Description	
0	Erase mode cleared	(Initial value)
1	Transition to erase mode	
	[Setting condition]	
	When SWE = 1, and ESU = 1	

Bit 0—Program (P): Selects program mode transition or clearing. Do not set the SWE, PSU, ESU, EV, PV, or E bit at the same time.

Bit 0

P	Description	
0	Program mode cleared	(Initial value)
1	Transition to program mode	
	[Setting condition]	
	When SWE = 1, and PSU = 1	

22.5.2 Flash Memory Control Register 2 (FLMCR2)

Bit	7	6	5	4	3	2	1	0
	FLER	_	_	_	_	_	ESU	PSU
Initial value	0	0	0	0	0	0	0	0
Read/Write	R	_	_	_	_	_	R/W	R/W

FLMCR2 is an 8-bit register that monitors the presence or absence of flash memory program/erase protection (error protection) and performs setup for flash memory program/erase mode. FLMCR2 is initialized to H'00 by a reset, and in hardware standby mode. The ESU and PSU bits are cleared to 0 in software standby mode, subactive mode, subsleep mode, and watch mode.

When on-chip flash memory is disabled, a read will return H'00, and writes are invalid.

Bit 7—Flash Memory Error (FLER): Indicates that an error has occurred during an operation on flash memory (programming or erasing). When FLER is set to 1, flash memory goes to the error-protection state.

Bit 7

DIL 1		
FLER	Description	
0	Flash memory is operating normally	(Initial value)
	Flash memory program/erase protection (error protection) is disabled	
	[Clearing condition]	
	Reset or hardware standby mode	
1	An error has occurred during flash memory programming/erasing	
	Flash memory program/erase protection (error protection) is enabled	
	[Setting condition]	
	See section 22.8.3, Error Protection	

Bits 6 to 2—Reserved: Always write 0 when writing to these bits.

Bit 1—Erase Setup (ESU): Prepares for a transition to erase mode. Set this bit to 1 before setting the E bit to 1 in FLMCR1. Do not set the SWE, PSU, EV, PV, E, or P bit at the same time.

Bit 1

ESU	Description	
0	Erase setup cleared	(Initial value)
1	Erase setup	
	[Setting condition]	
	When SWE = 1	

Bit 0—Program Setup (PSU): Prepares for a transition to program mode. Set this bit to 1 before setting the P bit to 1 in FLMCR1. Do not set the SWE, ESU, EV, PV, E, or P bit at the same time.

Bit 0

PSU	Description	
0	Program setup cleared	(Initial value)
1	Program setup	
	[Setting condition]	
	When SWE = 1	

22.5.3 Erase Block Registers 1 and 2 (EBR1, EBR2)

Bit	7	6	5	4	3	2	1	0
EBR1	_	_	_	_	_	_	EB9/—*2	EB8/—*2
Initial value	0	0	0	0	0	0	0	0
Read/Write	_	_	_	_	_	_	R/W*1 *2	R/W*1 *2
Bit	7	6	5	4	3	2	1	0
EBR2	EB7	EB6	EB5	EB4	EB3	EB2	EB1	EB0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W*1	R/W	R/W	R/W	R/W	R/W	R/W	R/W

Notes: 1. In normal mode, these bits cannot be modified and are always read as 0.

2. Bits EB8 and EB9 are not present in the 64-kbyte versions; they must not be set to 1.

EBR1 and EBR2 are registers that specify the flash memory erase area block by block; bits 1 and 2 in EBR1 (128 kB versions only) and bits 7 to 0 in EBR2 are readable/writable bits. EBR1 and EBR2 are each initialized to H'00 by a reset, in hardware standby mode, software standby mode, subactive mode, subsleep mode, and watch mode, and the SWE bit in FLMCR1 is not set. When a bit in EBR1 or EBR2 is set, the corresponding block can be erased. Other blocks are erase-protected. Set only one bit in EBR1 or EBR2 (more than one bit cannot be set). When on-chip flash memory is disabled, a read will return H'00, and writes are invalid.

The flash memory block configuration is shown in table 22.5.

Table 22.5 Flash Memory Erase Blocks

Block (Size)

128-kbyte Versions	64-kbyte Versions	Address
EB0 (1 kbyte)	EB0 (1 kbyte)	H'(00)0000 to H'(00)03FF
EB1 (1 kbyte)	EB1 (1 kbyte)	H'(00)0400 to H'(00)07FF
EB2 (1 kbyte)	EB2 (1 kbyte)	H'(00)0800 to H'(00)0BFF
EB3 (1 kbyte)	EB3 (1 kbytes)	H'(00)0C00 to H'(00)0FFF
EB4 (28 kbytes)	EB4 (28 kbytes)	H'(00)1000 to H'(00)7FFF
EB5 (16 kbytes)	EB5 (16 kbytes)	H'(00)8000 to H'(00)BFFF
EB6 (8 kbytes)	EB6 (8 kbytes)	H'(00)C000 to H'(00)DFFF
EB7 (8 kbytes)	EB7 (8 kbytes)	H'00E000 to H'00FFFF
EB8 (32 kbytes)	_	H'010000 to H'017FFF
EB9 (32 kbytes)	_	H'018000 to H'01FFFF

22.5.4 Serial/Timer Control Register (STCR)

Bit	7	6	5	4	3	2	1	0
	IICS	IICX1	IICX0	IICE	FLSHE	_	ICKS1	ICKS0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

STCR is an 8-bit readable/writable register that controls register access, the IIC operating mode (when the on-chip IIC option is included), and on-chip flash memory control (in F-ZTAT versions), and also selects the TCNT input clock. For details on functions not related to on-chip flash memory, see section 3.2.4, Serial Timer Control Register (STCR), and descriptions of individual modules. If a module controlled by STCR is not used, do not write 1 to the corresponding bit.

STCR is initialized to H'00 by a reset and in hardware standby mode.

Bits 7 to 4—I²C **Control (IICS, IICX1, IICX0, IICE):** When the on-chip IIC option is included, these bits control the operation of the I²C bus interface. For details, see section 16, I²C Bus Interface.

Bit 3—Flash Memory Control Register Enable (FLSHE): Setting the FLSHE bit to 1 enables read/write access to the flash memory control registers. If FLSHE is cleared to 0, the flash memory control registers are deselected. In this case, the flash memory control register contents are retained

Bit 3

FLSHE	Description	
0	Flash memory control registers deselected	(Initial value)
1	Flash memory control registers selected	

Bit 2—Reserved: Do not write 1 to this bit.

Bits 1 and 0—Internal Clock Select 1 and 0 (ICKS1, ICKS0): These bits control 8-bit timer operation. See section 12, 8-Bit Timers, for details.

22.6 On-Board Programming Modes

When pins are set to on-board programming mode, program/erase/verify operations can be performed on the on-chip flash memory. There are two on-board programming modes: boot mode and user program mode. The pin settings for transition to each of these modes are shown in table 22.6. For a diagram of the transitions to the various flash memory modes, see figure 22.3.

Only advanced mode setting is possible for boot mode.

In the case of user program mode, established in advanced mode or normal mode, depending on the setting of the MD0 pin. In normal mode, only programming of a 56-kbyte area of flash memory is possible.

Table 22.6 Setting On-Board Programming Modes

Ν	Λ	O	d	e

Mode Name	CPU Operating Mode	MD1	MD0	P92	P91	P90
Boot mode	Advanced mode	0	0	1*	1*	1*
User program mode	Advanced mode	1	0	_	_	_
	Normal mode		1	_	_	_

Note: * Can be used as I/O ports after boot mode is initiated.

22.6.1 Boot Mode

When boot mode is used, the flash memory programming control program must be prepared in the host beforehand. The channel 1 SCI to be used is set to asynchronous mode.

When a reset-start is executed after this group MCU's pins have been set to boot mode, the boot program built into the MCU is started and the programming control program prepared in the host is serially transmitted to the MCU via the SCI. In the MCU, the programming control program received via the SCI is written into the programming control program area in on-chip RAM. After the transfer is completed, control branches to the start address of the programming control program area and the programming control program execution state is entered (flash memory programming is performed).

The transferred programming control program must therefore include coding that follows the programming algorithm given later.

The system configuration in boot mode is shown in figure 22.7, and the boot program mode execution procedure in figure 22.8.

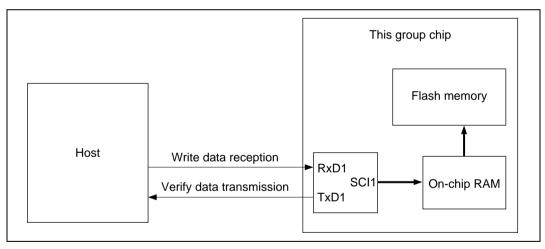


Figure 22.7 System Configuration in Boot Mode

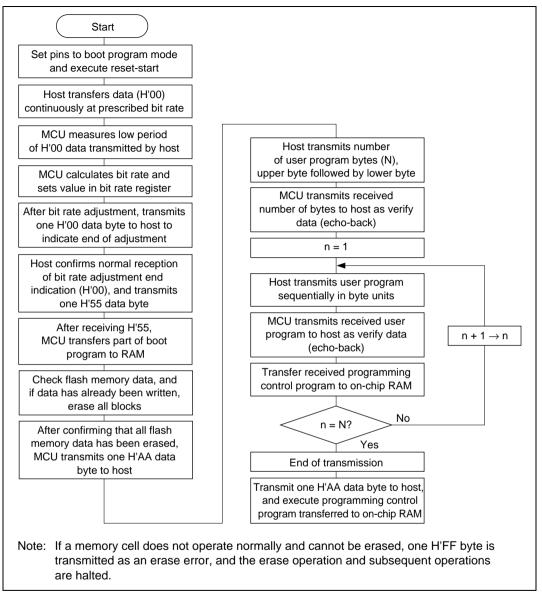


Figure 22.8 Boot Mode Execution Procedure

Automatic SCI Bit Rate Adjustment

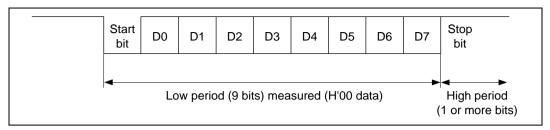


Figure 22.9 RxD1 Input Signal when Using Automatic SCI Bit Rate Adjustment

When boot mode is initiated, this group MCU measures the low period of the asynchronous SCI communication data (H'00) transmitted continuously from the host. The SCI transmit/receive format should be set as follows: 8-bit data, 1 stop bit, no parity. The MCU calculates the bit rate of the transmission from the host from the measured low period, and transmits one H'00 byte to the host to indicate the end of bit rate adjustment. The host should confirm that this adjustment end indication (H'00) has been received normally, and transmit one H'55 byte to the MCU. If reception cannot be performed normally, initiate boot mode again (reset), and repeat the above operations. Depending on the host's transmission bit rate and the MCU's system clock frequency, there will be a discrepancy between the bit rates of the host and the MCU. To ensure correct SCI operation, the host's transfer bit rate should be set to (2400, 4800, or 9600) bps.

Table 22.7 shows typical host transfer bit rates and system clock frequencies for which automatic adjustment of the MCU's bit rate is possible. The boot program should be executed within this system clock range.

Table 22.7 System Clock Frequencies for which Automatic Adjustment of This Group Bit Rate Is Possible

Host Bit Rate	System Clock Frequency for which Automatic Adjustment of This Group Bit Rate Is Possible
9600 bps	8 MHz to 20 MHz
4800 bps	4 MHz to 20 MHz
2400 bps	2 MHz to 18 MHz

On-Chip RAM Area Divisions in Boot Mode

In boot mode, the 128-byte area from H'(FF)FF00 to H'(FF)FF7F is reserved for use by the boot program, as shown in figure 22.10. The area to which the programming control program is transferred is H'(FF)E080 to H'(FF)EFFF (3968 bytes) in the 128-kbyte versions, or H'(FF)E880 to H'(FF)EFFF (1920 bytes) in the 64-kbyte versions. The boot program area can be used when the programming control program transferred into RAM enters the execution state. A stack area should be set up as required.

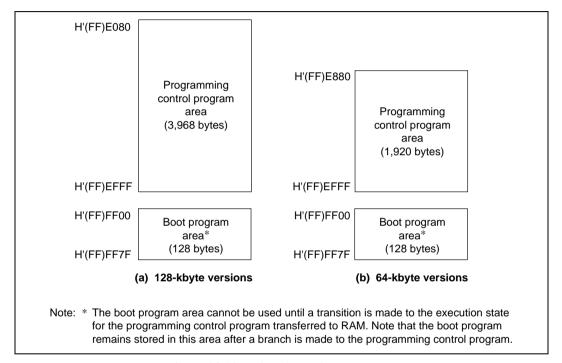


Figure 22.10 RAM Areas in Boot Mode

Notes on Use of User Mode

- When the chip comes out of reset in boot mode, it measures the low period of the input at the SCI's RxD1 pin. The reset should end with RxD1 high. After the reset ends, it takes about 100 states for the chip to get ready to measure the low period of the RxD1 input.
- In boot mode, if any data has been programmed into the flash memory (if all data is not 1), all
 flash memory blocks are erased. Boot mode is for use when user program mode is unavailable,
 such as the first time on-board programming is performed, or if the program activated in user
 program mode is accidentally erased.

- Interrupts cannot be used while the flash memory is being programmed or erased.
- The RxD1 and TxD1 pins should be pulled up on the board.
- Before branching to the programming control program (RAM area H'(FF)E080 (128-kbyte versions)) or H'(FF)E880 (64-kbyte versions)), the chip terminates transmit and receive operations by the on-chip SCI (channel 1) (by clearing the RE and TE bits in SCR to 0), but the adjusted bit rate value remains set in BRR. The transmit data output pin, TxD1, goes to the high-level output state (P84DDR = 1, P84DR = 1).

The contents of the CPU's internal general registers are undefined at this time, so these registers must be initialized immediately after branching to the programming control program. In particular, since the stack pointer (SP) is used implicitly in subroutine calls, etc., a stack area must be specified for use by the programming control program.

The initial values of other on-chip registers are not changed.

Boot mode can be entered by making the pin settings shown in table 22.6 and executing a
reset-start.

When the chip detects the boot mode setting at reset release*1, P92, P91, and P90 can be used as I/O ports.

Boot mode can be cleared by driving the reset pin low, waiting at least 20 states, then setting the mode pins, and executing reset release*1. Boot mode can also be cleared by a WDT overflow reset.

The mode pin input levels must not be changed in boot mode.

- If the mode pin input levels are changed (for example, from low to high) during a reset, the state of ports with multiplexed address functions and bus control output pins (AS, RD, WR) will change according to the change in the microcomputer's operating mode*2.
 - Therefore, care must be taken to make pin settings to prevent these pins from becoming output signal pins during a reset, or to prevent collision with signals outside the microcomputer.
- Notes: 1. Mode pin input must satisfy the mode programming setup time ($t_{MDS} = 4$ states) with respect to the reset release timing.
 - 2. Ports with multiplexed address functions will output a low level as the address signal if mode pin setting is for mode 1 is entered during a reset. In other modes, the port pins go to the high-impedance state. The bus control output signals will output a high level if mode pin setting is for mode 1 is entered during a reset. In other modes, the port pins go to the high-impedance state.



22.6.2 User Program Mode

When set to user program mode, the chip can program and erase its flash memory by executing a user program/erase control program. Therefore, on-board reprogramming of the on-chip flash memory can be carried out by providing on-board supply of programming data, and storing a program/erase control program in part of the program area as necessary.

To select user program mode, select a mode that enables the on-chip flash memory (mode 2 or 3). In this mode, on-chip supporting modules other than flash memory operate as they normally would in mode 2 and 3.

The flash memory itself cannot be read while the SWE bit is set to 1 to perform programming or erasing, so the control program that performs programming and erasing should be run in on-chip RAM or external memory.

Figure 22.11 shows the procedure for executing the program/erase control program when transferred to on-chip RAM.

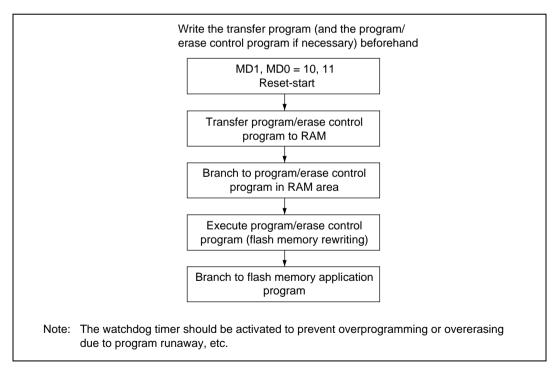


Figure 22.11 User Program Mode Execution Procedure

22.7 Programming/Erasing Flash Memory

In the on-board programming modes, flash memory programming and erasing is performed by software, using the CPU. There are four flash memory operating modes: program mode, erase mode, program-verify mode, and erase-verify mode. Transitions to these modes can be made by setting the PSU and ESU bits in FLMCR2, and the P, E, PV, and EV bits in FLMCR1.

The flash memory cannot be read while being programmed or erased. Therefore, the program that controls flash memory programming/erasing (the programming control program) should be located and executed in on-chip RAM or external memory.

- Notes: 1. Operation is not guaranteed if setting/resetting of the SWE, EV, PV, E, and P bits in FLMCR1, and the ESU and PSU bits in FLMCR2, is executed by a program in flash memory.
 - 2. Perform programming in the erased state. Do not perform additional programming on previously programmed addresses.

22.7.1 Program Mode

Follow the procedure shown in the program/program-verify flowchart in figure 21.12 to write data or programs to flash memory. Performing program operations according to this flowchart will enable data or programs to be written to flash memory without subjecting the device to voltage stress or sacrificing program data reliability. Programming should be carried out 32 bytes at a time.

For the wait times $(x, y, z, \alpha, \beta, \gamma, \epsilon, \eta)$ after setting/clearing individual bits in flash memory control registers 1 and 2 (FLMCR1, FLMCR2) and the maximum number of writes (N), see section 26.2.6, Flash Memory Characteristics.

Following the elapse of (x) µs or more after the SWE bit is set to 1 in flash memory control register 1 (FLMCR1), 32-byte program data is stored in the program data area and reprogram data area, and the 32-byte data in the reprogram data area written consecutively to the write addresses. The lower 8 bits of the first address written to must be H'00, H'20, H'40, H'60, H'80, H'A0, H'C0, or H'E0. Thirty-two consecutive byte data transfers are performed. The program address and program data are latched in the flash memory. A 32-byte data transfer must be performed even if writing fewer than 32 bytes; in this case, H'FF data must be written to the extra addresses.

Next, the watchdog timer is set to prevent overprogramming in the event of program runaway, etc. Set a value greater than $(y + z + \alpha + \beta)$ μ s as the WDT overflow period. After this, preparation for program mode (program setup) is carried out by setting the PSU bit in FLMCR2, and after the elapse of (y) μ s or more, the operating mode is switched to program mode by setting the P bit in



FLMCR1. The time during which the P bit is set is the flash memory programming time. Make a program setting so that the time for one programming operation is within the range of (z) μ s.

22.7.2 Program-Verify Mode

In program-verify mode, the data written in program mode is read to check whether it has been correctly written in the flash memory.

After the elapse of a given programming time, the programming mode is exited (the P bit in FLMCR1 is cleared, then the PSU bit in FLMCR2 is cleared at least (α) μ s later). The watchdog timer is cleared after the elapse of (β) μ s or more, and the operating mode is switched to programverify mode by setting the PV bit in FLMCR1. Before reading in program-verify mode, a dummy write of H'FF data should be made to the addresses to be read. The dummy write should be executed after the elapse of (γ) μ s or more. When the flash memory is read in this state (verify data is read in 16-bit units), the data at the latched address is read. Wait at least (ϵ) μ s after the dummy write before performing this read operation. Next, the originally written data is compared with the verify data, and reprogram data is computed (see figure 22.12) and transferred to the reprogram data area. After 32 bytes of data have been verified, exit program-verify mode, wait for at least (η) μ s, then clear the SWE bit in FLMCR1. If reprogramming is necessary, set program mode again, and repeat the program/program-verify sequence as before. However, ensure that the program/program-verify sequence is not repeated more than (N) times on the same bits.

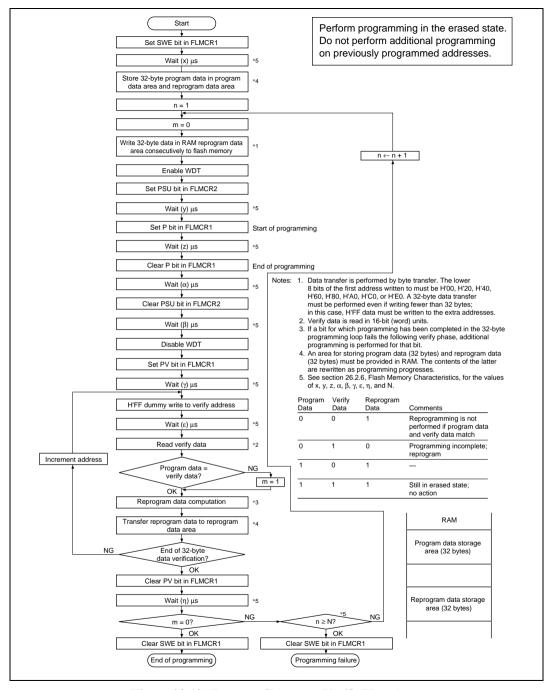


Figure 22.12 Program/Program-Verify Flowchart

22.7.3 Erase Mode

Flash memory erasing should be performed block by block following the procedure shown in the erase/erase-verify flowchart (single-block erase) shown in figure 22.13.

The wait times $(x, y, z, \alpha, \beta, \gamma, \epsilon, \eta)$ after setting/clearing individual bits in flash memory control registers 1 and 2 (FLMCR1, FLMCR2) and the maximum number of erases (N), see section 26.2.6, Flash Memory Characteristics.

To perform data or program erasure, make a 1 bit setting for the flash memory area to be erased in erase block register 1 or 2 (EBR1 or EBR2) at least (x) μ s after setting the SWE bit to 1 in flash memory control register 1 (FLMCR1). Next, the watchdog timer is set to prevent overerasing in the event of program runaway, etc. Set a value greater than $(y + z + \alpha + \beta)$ ms as the WDT overflow period. After this, preparation for erase mode (erase setup) is carried out by setting the ESU bit in FLMCR2, and after the elapse of (y) μ s or more, the operating mode is switched to erase mode by setting the E bit in FLMCR1. The time during which the E bit is set is the flash memory erase time. Ensure that the erase time does not exceed (z) ms.

Note: With flash memory erasing, preprogramming (setting all data in the memory to be erased to 0) is not necessary before starting the erase procedure.

22.7.4 Erase-Verify Mode

In erase-verify mode, data is read after memory has been erased to check whether it has been correctly erased.

After the elapse of the erase time, erase mode is exited (the E bit in FLMCR1 is cleared, then the ESU bit in FLMCR2 is cleared at least (α) μ s later), the watchdog timer is cleared after the elapse of (β) μ s or more, and the operating mode is switched to erase-verify mode by setting the EV bit in FLMCR1. Before reading in erase-verify mode, a dummy write of H'FF data should be made to the addresses to be read. The dummy write should be executed after the elapse of (γ) μ s or more. When the flash memory is read in this state (verify data is read in 16-bit units), the data at the latched address is read. Wait at least (ϵ) μ s after the dummy write before performing this read operation. If the read data has been erased (all 1), a dummy write is performed to the next address, and erase-verify is performed. If the read data has not been erased, set erase mode again, and repeat the erase/erase-verify sequence in the same way. However, ensure that the erase/erase-verify sequence is not repeated more than (N) times. When verification is completed, exit erase-verify mode, and wait for at least (η) μ s. If erasure has been completed on all the erase blocks, clear the SWE bit in FLMCR1. If there are any unerased blocks, make a 1 bit setting in EBR1 or EBR2 for the flash memory area to be erased, and repeat the erase/erase-verify sequence in the same way.

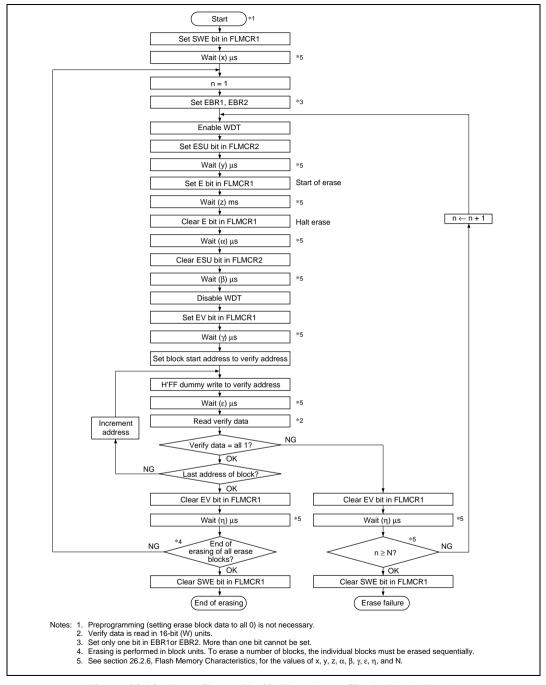


Figure 22.13 Erase/Erase-Verify Flowchart (Single-Block Erase)

22.8 **Flash Memory Protection**

There are three kinds of flash memory program/erase protection: hardware protection, software protection, and error protection.

22.8.1 Hardware Protection

Hardware protection refers to a state in which programming/erasing of flash memory is forcibly disabled or aborted. Hardware protection is reset by settings in flash memory control registers 1 and 2 (FLMCR1, FLMCR2) and erase block registers 1 and 2 (EBR1, EBR2). (See table 22.8.)

Table 22.8 Hardware Protection

		Fund	ctions
Item	Description	Program	Erase
Reset/standby protection	 In a reset (including a WDT overflow reset), software standby mode, subactive mode, subsleep mode, and watch mode, FLMCR1, FLMCR2, EBR1, and EBR2 are initialized, and the program/erase-protected state is entered. 	Yes	Yes
	 In a reset via the RES pin, the reset state is not entered unless the RES pin is held low until oscillation stabilizes after powering on. In the case of a reset during operation, hold the RES pin low for the RES pulse width specified in the AC Characteristics section. 		

22.8.2 **Software Protection**

Software protection can be implemented by setting the SWE bit in FLMCR1 and erase block registers 1 and 2 (EBR1, EBR2). When software protection is in effect, setting the P or E bit in flash memory control register 1 (FLMCR1) does not cause a transition to program mode or erase mode. (See table 22.9.)

Table 22.9 Software Protection

		Func	tions
Item	Description	Program	Erase
SWE bit protection	Clearing the SWE bit to 0 in FLMCR1 sets the program/erase-protected state for all blocks. (Execute in on-chip RAM or external memory.)	Yes	Yes
Block specification protection	 Erase protection can be set for individual blocks by settings in erase block registers 1 and 2 (EBR1, EBR2). 	_	Yes
	 Setting EBR1 and EBR2 to H'00 places all blocks in the erase-protected state. 		

22.8.3 Error Protection

In error protection, an error is detected when MCU runaway occurs during flash memory programming/erasing, or operation is not performed in accordance with the program/erase algorithm, and the program/erase operation is aborted. Aborting the program/erase operation prevents damage to the flash memory due to overprogramming or overerasing.

If the MCU malfunctions during flash memory programming/erasing, the FLER bit is set to 1 in FLMCR2 and the error protection state is entered. The FLMCR1, FLMCR2, EBR1, and EBR2 settings are retained, but program mode or erase mode is aborted at the point at which the error occurred. Program mode or erase mode cannot be re-entered by re-setting the P or E bit. However, PV and EV bit setting is enabled, and a transition can be made to verify mode.

FLER bit setting conditions are as follows:

- When flash memory is read during programming/erasing (including a vector read or instruction fetch)
- Immediately after exception handling (excluding a reset) during programming/erasing
- When a SLEEP instruction (transition to software standby, sleep, subactive, subsleep, or watch mode) is executed during programming/erasing
- When the bus is released during programming/erasing

Error protection is released only by a reset and in hardware standby mode.

Figure 22.14 shows the flash memory state transition diagram.



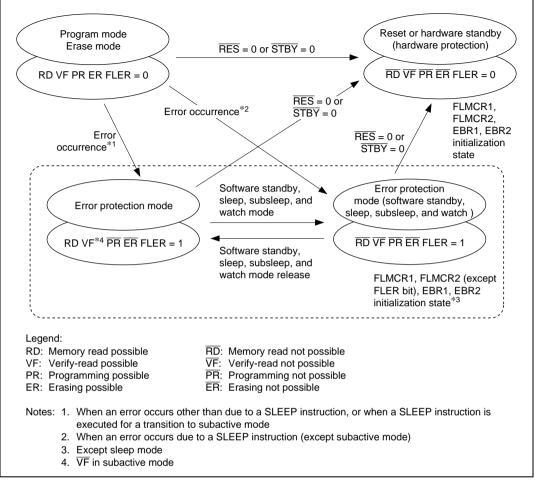


Figure 22.14 Flash Memory State Transitions

22.9 Interrupt Handling when Programming/Erasing Flash Memory

All interrupts, including NMI input is disabled when flash memory is being programmed or erased (when the P or E bit is set in FLMCR1), and while the boot program is executing in boot mode*1, to give priority to the program or erase operation. There are three reasons for this:

- 1. Interrupt during programming or erasing might cause a violation of the programming or erasing algorithm, with the result that normal operation could not be assured.
- 2. In the interrupt exception handling sequence during programming or erasing, the vector would not be read correctly*2, possibly resulting in MCU runaway.
- 3. If interrupt occurred during boot program execution, it would not be possible to execute the normal boot mode sequence.

For these reasons, in on-board programming mode alone there are conditions for disabling interrupt, as an exception to the general rule. However, this provision does not guarantee normal erasing and programming or MCU operation. All interrupt requests, including NMI input, must therefore be disabled inside and outside the MCU when programming or erasing flash memory. Interrupt is also disabled in the error-protection state while the P or E bit remains set in FLMCR1.

- Notes: 1. Interrupt requests must be disabled inside and outside the MCU until the programming control program has completed programming.
 - 2. The vector may not be read correctly in this case for the following two reasons:
 - If flash memory is read while being programmed or erased (while the P or E bit is set in FLMCR1), correct read data will not be obtained (undetermined values will be returned).
 - If the interrupt entry in the vector table has not been programmed yet, interrupt exception handling will not be executed correctly.



22.10 Flash Memory Programmer Mode

22.10.1 Programmer Mode Setting

Programs and data can be written and erased in programmer mode as well as in the on-board programming modes. In programmer mode, the on-chip ROM can be freely programmed using a PROM programmer that supports Renesas Technology microcomputer device types with 128-kbyte*** or 64-kbyte*** on-chip flash memory. Flash memory read mode, auto-program mode, auto-erase mode, and status read mode are supported with these device types. In auto-program mode, auto-erase mode, and status read mode, a status polling procedure is used, and in status read mode, detailed internal signals are output after execution of an auto-program or auto-erase operation.

Table 22.10 shows programmer mode pin settings.

Notes: 1. Applies to the H8S/2148 and H8S/2144.

- 2. Applies to the H8/2147N and H8S/2142.
- 3. Use products other than the A-mask version of the H8S/2148, H8S/2147N, H8/2144, and H8S/2142 (in either 5-V or 3-V version) with the writing voltage for the PROM programmer set to 5.0 V. Do not use the A-mask version with a 5.0-V PROM programmer setting.

Table 22.10 Programmer Mode Pin Settings

Pin Names	Setting/External Circuit Connection
Mode pins: MD1, MD0	Low-level input to MD1, MD0
STBY pin	High-level input (Hardware standby mode not set)
RES pin	Power-on reset circuit
XTAL and EXTAL pins	Oscillation circuit
Other setting pins: P97, P92, P91, P90, P67	Low-level input to P92, P67, high-level input to P97, P91, P90

22.10.2 Socket Adapters and Memory Map

In programmer mode, a socket adapter is mounted on the writer programmer to match the package concerned. Socket adapters are available for each writer manufacturer supporting Renesas Technology microcomputer device types with 128-kbyte or 64-kbyte on-chip flash memory.

Figure 22.15 shows the memory map in programmer mode. For pin names in programmer mode, see section 1.3.2, Pin Functions in Each Operating Mode.

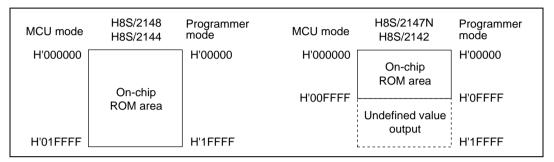


Figure 22.15 Memory Map in Programmer mode

22.10.3 Programmer Mode Operation

Table 22.11 shows how the different operating modes are set when using programmer mode, and table 22.12 lists the commands used in programmer mode. Details of each mode are given below.

- Memory Read Mode
 Memory read mode supports byte reads.
- Auto-Program Mode
 Auto-program mode supports programming of 128 bytes at a time. Status polling is used to confirm the end of auto-programming.
- Auto-Erase Mode
 Auto-erase mode supports automatic erasing of the entire flash memory. Status polling is used to confirm the end of auto-erasing.
- Status Read Mode
 Status polling is used for auto-programming and auto-erasing, and normal termination can be confirmed by reading the FO6 signal. In status read mode, error information is output if an error occurs.

Table 22.11 Settings for Each Operating Mode in Programmer Mode

Pin Names

Mode	CE	ŌĒ	WE	FO0 to FO7	FA0 to FA17
Read	L	L	Н	Data output	Ain
Output disable	L	Н	Н	Hi-Z	Х
Command write	L	Н	L	Data input	Ain ^{*2}
Chip disable*1	Н	Χ	Χ	Hi-Z	Χ

Notes: 1. Chip disable is not a standby state; internally, it is an operation state.

2. Ain indicates that there is also address input in auto-program mode.

Table 22.12 Programmer Mode Commands

Command Name	Number	1st Cycle			2nd Cycle		
	of Cycles	Mode	Address	Data	Mode	Address	Data
Memory read mode	1 + n	Write	Х	H'00	Read	RA	Dout
Auto-program mode	129	Write	Х	H'40	Write	WA	Din
Auto-erase mode	2	Write	Х	H'20	Write	Χ	H'20
Status read mode	2	Write	Χ	H'71	Write	Χ	H'71

Notes: 1. In auto-program mode, 129 cycles are required for command writing by a simultaneous 128-byte write.

2. In memory read mode, the number of cycles depends on the number of address write cycles (n).

22.10.4 Memory Read Mode

- After the end of an auto-program, auto-erase, or status read operation, the command wait state
 is entered. To read memory contents, a transition must be made to memory read mode by
 means of a command write before the read is executed.
- Command writes can be performed in memory read mode, just as in the command wait state.
- Once memory read mode has been entered, consecutive reads can be performed.
- After power-on, memory read mode is entered.

Table 22.13 AC Characteristics in Memory Read Mode

Conditions: $V_{CC} = 5.0 \text{ V} \pm 10\%$, $V_{SS} = 0 \text{ V}$, $T_a = 25^{\circ}\text{C} \pm 5^{\circ}\text{C}$

Item	Symbol	Min	Max	Unit	
Command write cycle	t _{nxtc}	20	_	μs	
CE hold time	t _{ceh}	0	_	ns	
CE setup time	t _{ces}	0	_	ns	
Data hold time	t _{dh}	50	_	ns	
Data setup time	t _{ds}	50	_	ns	
Write pulse width	t _{wep}	70	_	ns	
WE rise time	t _r	_	30	ns	
WE fall time	t _r	_	30	ns	

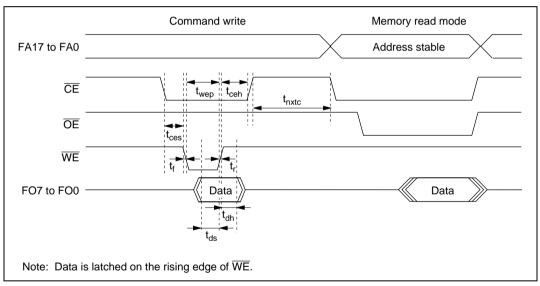


Figure 22.16 Memory Read Mode Timing Waveforms after Command Write

Table 22.14 AC Characteristics when Entering Another Mode from Memory Read Mode

Conditions: $V_{CC} = 5.0 \text{ V} \pm 10\%$, $V_{SS} = 0 \text{ V}$, $T_a = 25^{\circ}\text{C} \pm 5^{\circ}\text{C}$

Item	Symbol	Min	Max	Unit	
Command write cycle	t _{nxtc}	20	_	μs	
CE hold time	t _{ceh}	0	_	ns	
CE setup time	t _{ces}	0	_	ns	
Data hold time	t _{dh}	50	_	ns	
Data setup time	t _{ds}	50	_	ns	
Write pulse width	t _{wep}	70	_	ns	
WE rise time	t _r	_	30	ns	
WE fall time	t,	_	30	ns	

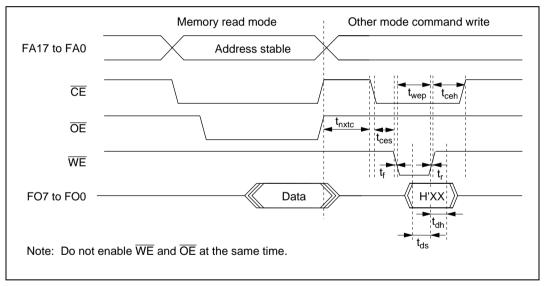


Figure 22.17 Timing Waveforms when Entering Another Mode from Memory Read Mode

Table 22.15 AC Characteristics in Memory Read Mode

Conditions: $V_{CC} = 5.0 \text{ V} \pm 10\%$, $V_{SS} = 0 \text{ V}$, $T_a = 25^{\circ}\text{C} \pm 5^{\circ}\text{C}$

Item	Symbol	Min	Max	Unit
Access time	t _{acc}	_	20	μs
CE output delay time	t _{ce}	_	150	ns
OE output delay time	t _{oe}	_	150	ns
Output disable delay time	t _{df}	_	100	ns
Data output hold time	t _{oh}	5	_	ns

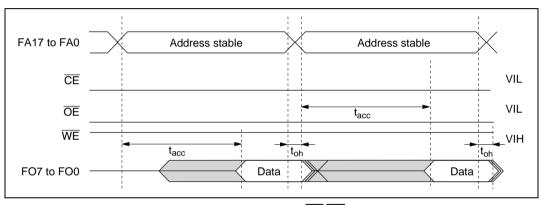


Figure 22.18 Timing Waveforms for $\overline{\text{CE}}/\overline{\text{OE}}$ Enable State Read

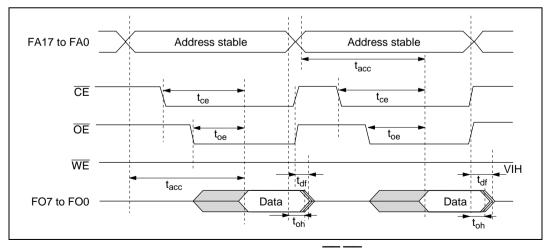


Figure 22.19 Timing Waveforms for $\overline{\text{CE}}/\overline{\text{OE}}$ Clocked Read

22.10.5 Auto-Program Mode

AC Characteristics

Table 22.16 AC Characteristics in Auto-Program Mode

Conditions: $V_{cc} = 5.0 \text{ V} \pm 10\%$, $V_{ss} = 0 \text{ V}$, $T_a = 25^{\circ}\text{C} \pm 5^{\circ}\text{C}$

Item	Symbol	Min	Max	Unit	
Command write cycle	t _{nxtc}	20	_	μs	
CE hold time	t _{ceh}	0	_	ns	
CE setup time	t _{ces}	0	_	ns	
Data hold time	t _{dh}	50	_	ns	
Data setup time	t _{ds}	50	_	ns	
Write pulse width	t _{wep}	70	_	ns	
Status polling start time	t _{wsts}	1	_	ms	
Status polling access time	t _{spa}	_	150	ns	
Address setup time	t _{as}	0	_	ns	
Address hold time	t _{ah}	60	_	ns	
Memory write time	t _{write}	1	3000	ms	
WE rise time	t _r	_	30	ns	
WE fall time	t,	_	30	ns	

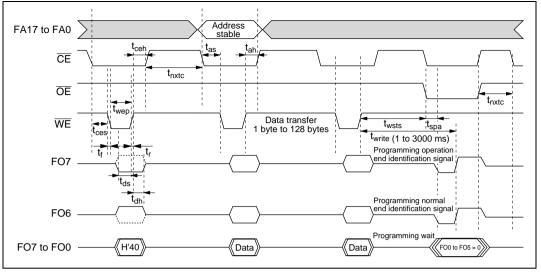


Figure 22.20 Auto-Program Mode Timing Waveforms

Notes on Use of Auto-Program Mode

- In auto-program mode, 128 bytes are programmed simultaneously. This should be carried out by executing 128 consecutive byte transfers.
- A 128-byte data transfer is necessary even when programming fewer than 128 bytes. In this case, H'FF data must be written to the extra addresses.
- The lower 8 bits of the transfer address must be H'00 or H'80. If a value other than an effective
 address is input, processing will switch to a memory write operation but a write error will be
 flagged.
- Memory address transfer is performed in the second cycle (figure 22.20). Do not perform transfer after the second cycle.
- Do not perform a command write during a programming operation.
- Perform one auto-programming operation for a 128-byte block for each address.
 Characteristics are not guaranteed for two or more programming operations.
- Confirm normal end of auto-programming by checking FO6. Alternatively, status read mode can also be used for this purpose (FO7 status polling uses the auto-program operation end identification pin).
- The status polling FO6 and FO7 pin information is retained until the next command write. Until the next command write is performed, reading is possible by enabling $\overline{\text{CE}}$ and $\overline{\text{OE}}$.



22.10.6 Auto-Erase Mode

AC Characteristics

Table 22.17 AC Characteristics in Auto-Erase Mode

Conditions: $V_{CC} = 5.0 \text{ V} \pm 10\%$, $V_{SS} = 0 \text{ V}$, $T_a = 25^{\circ}\text{C} \pm 5^{\circ}\text{C}$

Item	Symbol	Min	Max	Unit
Command write cycle	t _{nxtc}	20	_	μs
CE hold time	t _{ceh}	0	_	ns
CE setup time	t _{ces}	0	_	ns
Data hold time	t _{dh}	50	_	ns
Data setup time	t _{ds}	50	_	ns
Write pulse width	t_{wep}	70	_	ns
Status polling start time	t _{ests}	1	_	ms
Status polling access time	t _{spa}	_	150	ns
Memory erase time	t _{erase}	100	40000	ms
WE rise time	t _r	_	30	ns
WE fall time	t _f	_	30	ns

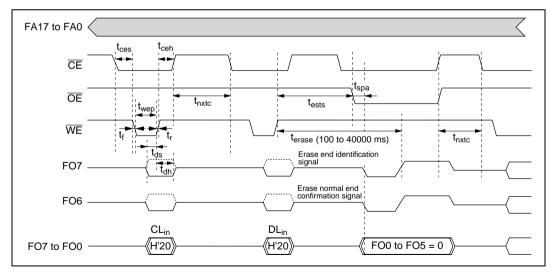


Figure 22.21 Auto-Erase Mode Timing Waveforms

Notes on Use of Erase-Program Mode

- Auto-erase mode supports only entire memory erasing.
- Do not perform a command write during auto-erasing.
- Confirm normal end of auto-erasing by checking FO6. Alternatively, status read mode can also be used for this purpose (FO7 status polling uses the auto-erase operation end identification pin).
- The status polling FO6 and FO7 pin information is retained until the next command write. Until the next command write is performed, reading is possible by enabling \overline{CE} and \overline{OE} .

22.10.7 Status Read Mode

- Status read mode is used to identify what type of abnormal end has occurred. Use this mode when an abnormal end occurs in auto-program mode or auto-erase mode.
- The return code is retained until a command write for other than status read mode is performed.

Table 22.18 AC Characteristics in Status Read Mode

Conditions: $V_{CC} = 5.0 \text{ V} \pm 10\%$, $V_{SS} = 0 \text{ V}$, $T_a = 25^{\circ}\text{C} \pm 5^{\circ}\text{C}$

Item	Symbol	Min	Max	Unit
Command write cycle	t _{nxtc}	20	_	μs
CE hold time	t _{ceh}	0	_	ns
CE setup time	t _{ces}	0	_	ns
Data hold time	\mathbf{t}_{dh}	50	_	ns
Data setup time	t _{ds}	50	_	ns
Write pulse width	t _{wep}	70	_	ns
OE output delay time	t _{oe}	_	150	ns
Disable delay time	t _{df}	_	100	ns
CE output delay time	\mathbf{t}_{ce}	_	150	ns
WE rise time	t,	_	30	ns
WE fall time	t,	_	30	ns

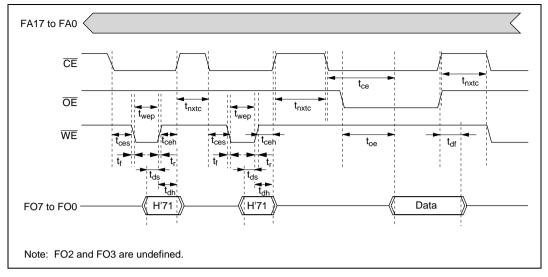


Figure 22.22 Status Read Mode Timing Waveforms

Table 22.19 Status Read Mode Return Commands

Pin Name	F07	FO6	FO5	FO4	FO3	FO2	F01	FO0
Attribute	Normal end identification	Command error	Program- ming error	Erase error	_	_	Program- ming or erase count exceeded	Effective address error
Initial value	0	0	0	0	0	0	0	0
Indications	Normal end: 0	Command error: 1	Program- ming	Erase error: 1	_	_	Count exceeded: 1	Effective address
	Abnormal end: 1	Otherwise: 0	error: 1 Otherwise: 0	Otherwise: ()		Otherwise: 0	error: 1 Otherwise: 0

Note: FO2 and FO3 are undefined.

22.10.8 Status Polling

- The FO7 status polling flag indicates the operating status in auto-program or auto-erase mode.
- The FO6 status polling flag indicates a normal or abnormal end in auto-program or auto-erase mode.

Table 22.20 Status Polling Output Truth Table

Pin Names	Internal Operation in Progress	Abnormal End	_	Normal End
FO7	0	1	0	1
FO6	0	0	1	1
FO0 to FO5	0	0	0	0

22.10.9 Programmer Mode Transition Time

Commands cannot be accepted during the oscillation stabilization period or the programmer mode setup period. After the programmer mode setup time, a transition is made to memory read mode.

Table 22.21 Command Wait State Transition Time Specifications

Item	Symbol	Min	Max	Unit
Standby release (oscillation stabilization time)	t _{osc1}	20	_	ms
Programmer mode setup time	t _{bmv}	10	_	ms
V _{cc} hold time	t _{dwn}	0	_	ms

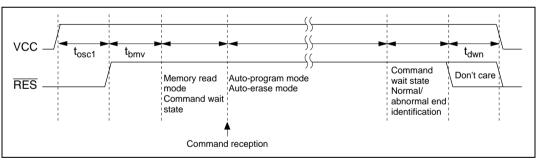


Figure 22.23 Oscillation Stabilization Time, Programmer Mode Setup Time, and Power Supply Fall Sequence

22.10.10 Notes on Memory Programming

- When programming addresses which have previously been programmed, carry out autoerasing before auto-programming.
- When performing programming using programmer mode on a chip that has been programmed/erased in an on-board programming mode, auto-erasing is recommended before carrying out auto-programming.
- Notes: 1. The flash memory is initially in the erased state when the device is shipped by Renesas. For other chips for which the erasure history is unknown, it is recommended that autoerasing be executed to check and supplement the initialization (erase) level.
 - 2. Auto-programming should be performed once only on the same address block.

22.11 Flash Memory Programming and Erasing Precautions

Precautions concerning the use of on-board programming mode and programmer mode are summarized below.

Use the specified voltages and timing for programming and erasing.

Applied voltages in excess of the rating can permanently damage the device. For a PROM programmer, use Renesas Technology microcomputer device types with 128-kbyte or 64-kbyte on-chip flash memory that support a 5.0-V programming voltage.

Do not select the HN28F101 or use a programming voltage of 3.3 V for the PROM programmer, and only use the specified socket adapter. Incorrect use will result in damaging the device.

Powering on and off

When applying or disconnecting $V_{\rm cc}$, fix the \overline{RES} pin low and place the flash memory in the hardware protection state.

The power-on and power-off timing requirements should also be satisfied in the event of a power failure and subsequent recovery.

Use the recommended algorithm when programming and erasing flash memory.

The recommended algorithm enables programming and erasing to be carried out without subjecting the device to voltage stress or sacrificing program data reliability. When setting the P or E bit in FLMCR1, the watchdog timer should be set beforehand as a precaution against program runaway, etc.

Do not set or clear the SWE bit during program execution in flash memory.

Clear the SWE bit before executing a program or reading data in flash memory. When the SWE bit is set, data in flash memory can be rewritten, but flash memory should only be accessed for verify operations (verification during programming/erasing).

Do not use interrupts while flash memory is being programmed or erased.

All interrupt requests, including NMI, should be disabled to give priority to program/erase operations.

Do not perform additional programming. Erase the memory before reprogramming.

In on-board programming, perform only one programming operation on a 32-byte programming unit block. In programmer mode, too, perform only one programming operation on a 128-byte programming unit block. Programming should be carried out with the entire programming unit block erased.

Before programming, check that the chip is correctly mounted in the PROM programmer.

Overcurrent damage to the device can result if the index marks on the PROM programmer socket, socket adapter, and chip are not correctly aligned.

Do not touch the socket adapter or chip during programming.

Touching either of these can cause contact faults and write errors.

22.12 Note on Switching from F-ZTAT Version to Mask ROM Version

The mask ROM version does not have the internal registers for flash memory control that are provided in the F-ZTAT version. Table 22.22 lists the registers that are present in the F-ZTAT version but not in the mask ROM version. If a register listed in table 22.22 is read in the mask ROM version, an undefined value will be returned, Therefore, if application software developed on the F-ZTAT version is switched to a mask ROM version product, it must be modified to ensure that the registers in table 22.22 have no effect.

Table 22.22 Registers Present in F-ZTAT Version but Absent in Mask ROM Version

Register	Abbreviation	Address
Flash memory control register 1	FLMCR1	H'FF80
Flash memory control register 2	FLMCR2	H'FF81
Erase block register 1	EBR1	H'FF82
Erase block register 2	EBR2	H'FF83



Section 23 ROM (H8S/2148 F-ZTAT A-Mask Version, H8S/2147 F-ZTAT A-Mask Version, H8S/2144 F-ZTAT A-Mask Version)

23.1 Overview

H8S/2148 F-ZTAT A-mask version and H8S/2144 F-ZTAT A-mask version have 128 kbytes, and H8S/2147 F-ZTAT A-mask version has 64 kbytes of on-chip flash memory. The flash memory is connected to the bus master by a 16-bit data bus. The bus master accesses both byte and word data in one state, enabling faster instruction fetches and higher processing speed.

The mode pins (MD1 and MD0) and the EXPE bit in MDCR can be set to enable or disable the on-chip ROM.

The flash memory versions of this group can be erased and programmed on-board as well as with a general-purpose PROM programmer.

23.1.1 Block Diagram

Figure 23.1 shows a block diagram of the ROM.

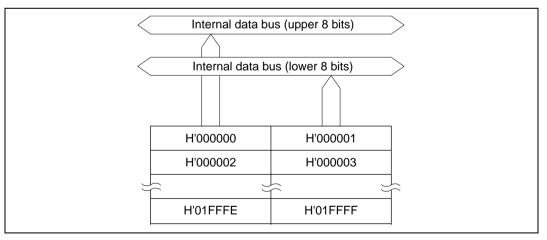


Figure 23.1 ROM Block Diagram (A-mask versions of the H8S/2148 F-ZTAT and H8S/2144 F-ZTAT)

23.1.2 Register Configuration

This group on-chip ROM is controlled by the operating mode and register MDCR. The register configuration is shown in table 23.1.

Table 23.1 ROM Register

Register Name	Abbreviation	R/W	Initial Value	Address*
Mode control register	MDCR	R/W	Undefined Depends on the operating mode	H'FFC5

Note: * Lower 16 bits of the address.

23.2 Register Descriptions

23.2.1 Mode Control Register (MDCR)

Bit	7	6	5	4	3	2	1	0
	EXPE	_	_	_	_	_	MDS1	MDS0
Initial value	*	0	0	0	0	0	*	*
Read/Write	R/W*	_	_	_	_	_	R	R

Note: * Determined by the MD1 and MD0 pins.

MDCR is an 8-bit read-only register used to set this group operating mode and monitor the current operating mode.

The EXPE bit is initialized in accordance with the mode pin states by a reset and in hardware standby mode.

Bit 7—Expanded Mode Enable (EXPE): Sets expanded mode. In mode 1, EXPE is fixed at 1 and cannot be modified. In modes 2 and 3, EXPE has an initial value of 0 and can be read or written.

Bit 7

EXPE	Description
0	Single-chip mode selected
1	Expanded mode selected

Bits 6 to 2—Reserved: These bits cannot be modified and are always read as 0.

Bits 1 and 0—Mode Select 1 and 0 (MDS1, MDS0): These bits indicate values that reflects the input levels of mode pins MD1 and MD0 (the current operating mode). Bits MDS1 and MDS0 correspond to pins MD1 and MD0, respectively. These are read-only bits, and cannot be modified. When MDCR is read, the input levels of mode pins MD1 and MD0 are latched in these bits.

23.3 **Operation**

The on-chip flash memory is connected to the CPU by a 16-bit data bus, and both byte and word data is accessed in one state. Even addresses are connected to the upper 8 bits, and odd addresses to the lower 8 bits. Word data must start at an even address.

The mode pins (MD1 and MD0) and the EXPE bit in MDCR can be set to enable or disable the on-chip ROM, as shown in table 23.2.

In normal mode, the maximum amount of ROM that can be used is 56 kbytes.

Table 23.2 Operating Modes and ROM

Operating Mode

MCU Operating	CPU Operating		Мо	de Pins	MDCR	
Mode	Mode	Description	MD1	MD0	EXPE	On-Chip ROM
Mode 1	Normal	Expanded mode with on-chip ROM disabled	0	1	1	Disabled
Mode 2	Advanced	Single-chip mode	1	0	0	Enabled*
	Advanced	Expanded mode with on-chip ROM enabled	=		1	_
Mode 3	Normal	Single-chip mode	-	1	0	Enabled
	Normal	Expanded mode with on-chip ROM enabled	=		1	(56 kbytes)

Note: H8S/2148 F-ZTAT A-mask version and H8S/2144 F-ZTAT A-mask version have 128 kbytes, and H8S/2147 F-ZTAT A-mask version has 64 kbytes of on-chip ROM.

RENESAS

23.4 Overview of Flash Memory

23.4.1 Features

The features of the flash memory are summarized below.

- Four flash memory operating modes
 - Program mode
 - Erase mode
 - Program-verify mode
 - Erase-verify mode
- Programming/erase methods

The flash memory is programmed 128 bytes at a time. Erasing is performed by block erase (in single-block units). When erasing multiple blocks, the individual blocks must be erased sequentially. Block erasing can be performed as required on 1-kbyte, 28-kbyte, 16-kbyte, 8-kbyte, and 32-kbyte blocks.

• Programming/erase times

The flash memory programming time is 10 ms (typ.) for simultaneous 128-byte programming, equivalent to approximately 80 µs (typ.) per byte, and the erase time is 100 ms (typ.) per block.

• Reprogramming capability

The flash memory can be reprogrammed up to 100 times.

• On-board programming modes

There are two modes in which flash memory can be programmed/erased/verified on-board:

- Boot mode
- User program mode
- Automatic bit rate adjustment

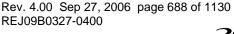
With data transfer in boot mode, the bit rate of the chip can be automatically adjusted to match the transfer bit rate of the host.

Protect modes

There are three protect modes, hardware, software, and error protect, which allow protected status to be designated for flash memory program/erase/verify operations.

Programmer mode

Flash memory can be programmed/erased in programmer mode, using a PROM programmer, as well as in on-board programming mode.





23.4.2 Block Diagram

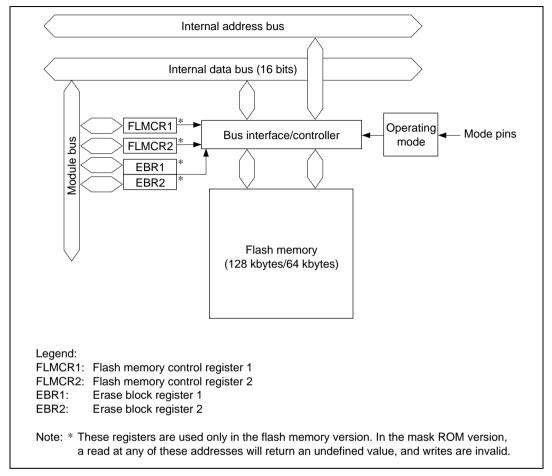


Figure 23.2 Block Diagram of Flash Memory

23.4.3 Flash Memory Operating Modes

Mode Transitions

When the mode pins are set in the reset state and a reset-start is executed, the MCU enters one of the operating modes shown in figure 23.3. In user mode, flash memory can be read but not programmed or erased.

Flash memory can be programmed and erased in boot mode, user program mode, and programmer mode.

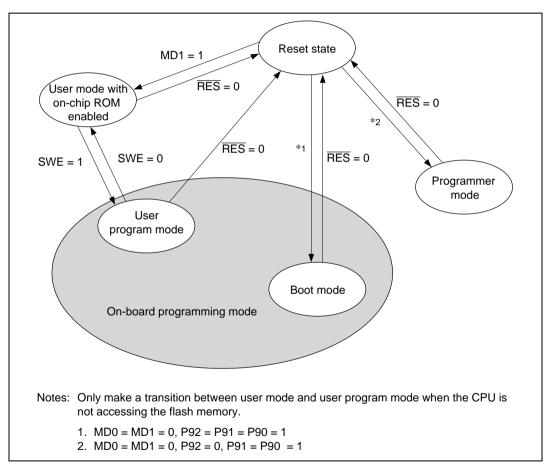


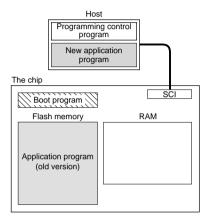
Figure 23.3 Flash Memory Mode Transitions

On-Board Programming Modes

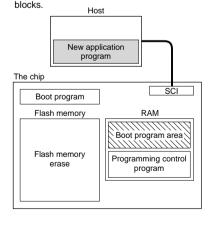
Boot mode

1. Initial state

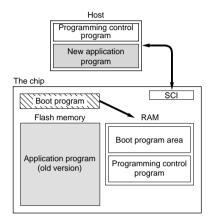
The flash memory is in the erased state when the device is shipped. The description here applies to the case where the old program version or data is being rewritten. The user should prepare the programming control program and new application program beforehand in the host.



Flash memory initialization
 The erase program in the boot program area (in RAM) is executed, and the flash memory is initialized (to H'FF). In boot mode, entire flash memory erasure is performed, without regard to



Programming control program transfer
 When boot mode is entered, the boot program in
 the chip (originally incorporated in the chip) is
 started, an SCI communication check is carried
 out, and the boot program required for flash
 memory erasing is automatically transferred to
 the RAM boot program area.



4. Writing new application program
The programming control program transferred
from the host to RAM by SCI communication is
executed, and the new application program in the
host is written into the flash memory.

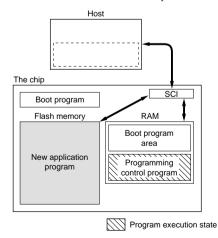


Figure 23.4 Boot Mode

User program mode

 Initial state

 (1) the program that will transfer the programming/ erase control program to on-chip RAM should be written into the flash memory by the user beforehand.
 (2) The programming/erase

control program should be prepared in the host

or in the flash memory.

Programming/
erase control program

New application
program

The chip

Boot program

Flash memory

RAM

Transfer program

Application program
(old version)

Flash memory initialization
 The programming/erase program in RAM is executed, and the flash memory is initialized (to H'FF). Erasing can be performed in block units, but not in byte units.

The chip

Boot program

Flash memory

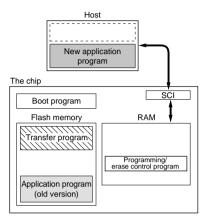
Flash memory

Flash memory

erase

Flash memory

 Programming/erase control program transfer The transfer program in the flash memory is executed, and the programming/erase control program is transferred to RAM.



Writing new application program
 Next, the new application program in the host is written into the erased flash memory blocks. Do not write to unerased blocks.

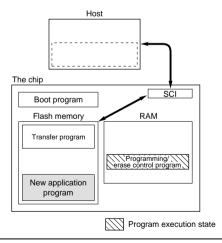


Figure 23.5 User Program Mode (Example)

Differences between Boot Mode and User Program Mode

	Boot Mode	User Program Mode
Entire memory erase	Yes	Yes
Block erase	No	Yes
Programming control program*	Program/program-verify	Program/program-verify
		Erase/erase-verify

Notes * To be provided by the user, in accordance with the recommended algorithm.

Block Configuration

The flash memory is divided into two 32-kbyte blocks (128-kbyte version only), two 8-kbyte blocks, one 16-kbyte block, one 28-kbyte block, and four 1-kbyte blocks.

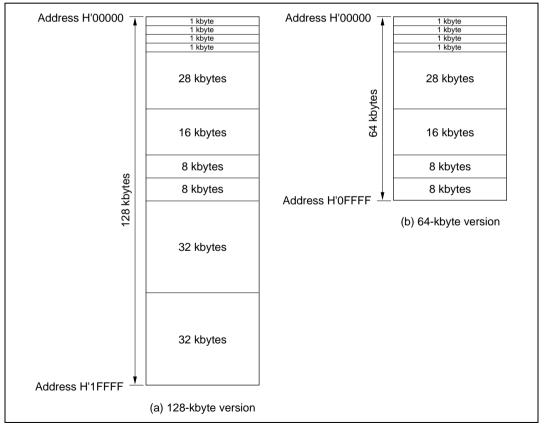


Figure 23.6 Flash Memory Block Configuration

23.4.4 Pin Configuration

The flash memory is controlled by means of the pins shown in table 23.3.

Table 23.3 Flash Memory Pins

Pin Name	Abbreviation	I/O	Function
Reset	RES	Input	Reset
Mode 1	MD1	Input	Sets MCU operating mode
Mode 0	MD0	Input	Sets MCU operating mode
Port 92	P92	Input	Sets MCU operating mode when MD1 = MD0 = 0
Port 91	P91	Input	Sets MCU operating mode when MD1 = MD0 = 0
Port 90	P90	Input	Sets MCU operating mode when MD1 = MD0 = 0
Transmit data	TxD1	Output	Serial transmit data output
Receive data	RxD1	Input	Serial receive data input

23.4.5 Register Configuration

The registers used to control the on-chip flash memory when enabled are shown in table 23.4. In order for these registers to be accessed, the FLSHE bit must be set to 1 in STCR.

Table 23.4 Flash Memory Registers

Register Name	Abbreviation	R/W	Initial Value	Address*1
Flash memory control register 1	FLMCR1*5	R/W*3	H'80	H'FF80*2
Flash memory control register 2	FLMCR2*5	R/W*3	H'00*4	H'FF81*2
Erase block register 1	EBR1*5	R/W*3	H'00*4	H'FF82*2
Erase block register 2	EBR2*5	R/W*3	H'00*4	H'FF83*2
Serial/timer control register	STCR	R/W	H'00	H'FFC3

Notes: 1. Lower 16 bits of the address.

- 2. Flash memory registers share addresses with other registers. Register selection is performed by the FLSHE bit in the serial/timer control register (STCR).
- 3. In modes in which the on-chip flash memory is disabled, a read will return H'00, and writes are invalid.
- 4. When the SWE bit in FLMCR1 is not set, these registers are initialized to H'00.
- 5. FLMCR1, FLMCR2, EBR1, and EBR2 are 8-bit registers. Only byte accesses are valid for these registers, the access requiring 2 states.

23.5 Register Descriptions

23.5.1 Flash Memory Control Register 1 (FLMCR1)

Bit	7	6	5	4	3	2	1	0
	FWE	SWE	_	_	EV	PV	Е	Р
Initial value	1	0	0	0	0	0	0	0
Read/Write	R	R/W	_	_	R/W	R/W	R/W	R/W

FLMCR1 is an 8-bit register used for flash memory operating mode control. Program-verify mode or erase-verify mode is entered by setting SWE to 1. Program mode is entered by setting SWE to 1, then setting the PSU bit in FLMCR2, and finally setting the P bit. Erase mode is entered by setting SWE to 1, then setting the ESU bit in FLMCR2, and finally setting the E bit. FLMCR1 is initialized to H'80 by a reset, and in hardware standby mode, software standby mode, subactive mode, subsleep mode, and watch mode. When on-chip flash memory is disabled, a read will return H'00, and writes are invalid.

Writes to the EV and PV bits in FLMCR1 are enabled only when SWE=1; writes to the E bit only when SWE = 1, and ESU = 1; and writes to the P bit only when SWE = 1, and PSU = 1.

Bit 7—Flash Write Enable (FWE): Sets hardware protection against flash memory programming/erasing. This bit cannot be modified and is always read as 1.

Bit 6—Software Write Enable (SWE): Enables or disables flash memory programming. SWE should be set before setting bits ESU, PSU, EV, PV, E, P, and EB9 to EB0, and should not be cleared at the same time as these bits.

Bit 6

SWE	Description	
0	Writes disabled	(Initial value)
1	Writes enabled	

Bit 5 and 4—Reserved: These bits cannot be modified and are always read as 0.

Bit 3—Erase-Verify (EV): Selects erase-verify mode transition or clearing. Do not set the SWE, ESU, PSU, PV, E, or P bit at the same time.

Bit 3

EV	Description	
0	Erase-verify mode cleared	(Initial value)
1	Transition to erase-verify mode	
	[Setting condition]	
	When SWE = 1	

Bit 2—Program-Verify (PV): Selects program-verify mode transition or clearing. Do not set the SWE, ESU, PSU, EV, E, or P bit at the same time.

Bit 2

PV	Description	
0	Program-verify mode cleared	(Initial value)
1	Transition to program-verify mode	
	[Setting condition]	
	When SWE = 1	

Bit 1—Erase (E): Selects erase mode transition or clearing. Do not set the SWE, ESU, PSU, EV, PV, or P bit at the same time.

Bit 1

E	Description	
0	Erase mode cleared	(Initial value)
1	Transition to erase mode	
	[Setting condition]	
	When SWE = 1, and ESU = 1	

Bit 0—Program (P): Selects program mode transition or clearing. Do not set the SWE, PSU, ESU, EV, PV, or E bit at the same time.

Bit 0

Р	Description	
0	Program mode cleared	(Initial value)
1	Transition to program mode	
	[Setting condition]	
	When SWE = 1, and PSU = 1	

23.5.2 Flash Memory Control Register 2 (FLMCR2)

Bit	7	6	5	4	3	2	1	0
	FLER	_	_	_	_	_	ESU	PSU
Initial value	0	0	0	0	0	0	0	0
Read/Write	R	_	_	_	_	_	R/W	R/W

FLMCR2 is an 8-bit register that monitors the presence or absence of flash memory program/erase protection (error protection) and performs setup for flash memory program/erase mode. FLMCR2 is initialized to H'00 by a reset, and in hardware standby mode. The ESU and PSU bits are cleared to 0 in software standby mode, subactive mode, subsleep mode, and watch mode.

When on-chip flash memory is disabled, a read will return H'00 and writes are invalid.

Bit 7—Flash Memory Error (FLER): Indicates that an error has occurred during an operation on flash memory (programming or erasing). When FLER is set to 1, flash memory goes to the error-protection state.

Bit 7

D					
FLER	Description				
0	Flash memory is operating normally	(Initial value)			
	Flash memory program/erase protection (error protection) is disabled				
	[Clearing condition]				
	Reset or hardware standby mode				
1	An error has occurred during flash memory programming/erasing				
	Flash memory program/erase protection (error protection) is enabled				
	[Setting condition]				
	See section 23.8.3, Error Protection				

Bits 6 to 2—Reserved: Always write 0 when writing to these bits.

Bit 1—Erase Setup (ESU): Prepares for a transition to erase mode. Set this bit to 1 before setting the E bit to 1 in FLMCR1. Do not set the SWE, PSU, EV, PV, E, or P bit at the same time.

Bit 1

ESU	 Description	
0	Erase setup cleared	(Initial value)
1	Erase setup	
	[Setting condition]	
	When SWE = 1	

Bit 0—Program Setup (PSU): Prepares for a transition to program mode. Set this bit to 1 before setting the P bit to 1 in FLMCR1. Do not set the SWE, ESU, EV, PV, E, or P bit at the same time.

Bit 0

PSU	Description	
0	Program setup cleared	(Initial value)
1	Program setup	
	[Setting condition]	
	When SWE = 1	

23.5.3 Erase Block Registers 1 and 2 (EBR1, EBR2)

Bit	7	6	5	4	3	2	1	0
EBR1	_	_	_	_	_	_	EB9/—*2	EB8/—*2
Initial value	0	0	0	0	0	0	0	0
Read/Write	_	_	_	_	_	_	R/W*1 *2	R/W*1 *2
Bit	7	6	5	4	3	2	1	0
EBR2	EB7	EB6	EB5	EB4	EB3	EB2	EB1	EB0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W*1	R/W	R/W	R/W	R/W	R/W	R/W	R/W

Notes: 1. In normal mode, these bits cannot be modified and are always read as 0.

2. Bits EB8 and EB9 are not present in the 64-kbyte versions; they must not be set to 1.

EBR1 and EBR2 are registers that specify the flash memory erase area block by block; bits 1 and 0 in EBR1 and bits 7 to 0 in EBR2 are readable/writable bits. EBR1 and EBR2 are each initialized to H'00 by a reset, in hardware standby mode, software standby mode, subactive mode, subsleep mode, and watch mode, and when the SWE bit in FLMCR1 is not set. When a bit in EBR1 and EBR2 is set, the corresponding block can be erased. Other blocks are erase-protected. Set only one bit in EBR1 and EBR2 (more than one bit cannot be set). When on-chip flash memory is disabled, a read will return H'00, and writes are invalid.

The flash memory block configuration is shown in table 23.5.

Table 23.5 Flash Memory Erase Blocks

Block (Size)

128-kbyte Version	64-kbyte Version	Address
EB0 (1 kbyte)	EB0 (1 kbyte)	H'(00)0000 to H'(00)03FF
EB1 (1 kbyte)	EB1 (1 kbyte)	H'(00)0400 to H'(00)07FF
EB2 (1 kbyte)	EB2 (1 kbyte)	H'(00)0800 to H'(00)0BFF
EB3 (1 kbytes)	EB3 (1 kbytes)	H'(00)0C00 to H'(00)0FFF
EB4 (28 kbytes)	EB4 (28 kbytes)	H'(00)1000 to H'(00)7FFF
EB5 (16 kbytes)	EB5 (16 kbytes)	H'(00)8000 to H'(00)BFFF
EB6 (8 kbytes)	EB6 (8 kbytes)	H'(00)C000 to H'(00)DFFF
EB7 (8 kbytes)	EB7 (8 kbytes)	H'00E000 to H'00FFFF
EB8 (32 kbytes)	_	H'010000 to H'017FFF
EB9 (32 kbytes)	_	H'018000 to H'01FFFF

23.5.4 Serial/Timer Control Register (STCR)

Bit	7	6	5	4	3	2	1	0
	IICS	IICX1	IICX0	IICE	FLSHE	_	ICKS1	ICKS0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

STCR is an 8-bit readable/writable register that controls register access, the IIC operating mode (when the on-chip IIC option is included), and on-chip flash memory, and also selects the TCNT input clock. For details on functions not related to on-chip flash memory, see section 3.2.4, Serial/Timer Control Register (STCR), and descriptions of individual modules. If a module controlled by STCR is not used, do not write 1 to the corresponding bit.

STCR is initialized to H'00 by a reset and in hardware standby mode.

Bits 7 to 4—I²C **Control** (**IICS, IICX1, IICX0, IICE**): These bits control the operation of the I²C bus interface for the I²C on-chip option. For details, see section 16, I²C Bus Interface.

Bit 3—Flash Memory Control Register Enable (FLSHE): Setting the FLSHE bit to 1 enables read/write access to the flash memory control registers. If FLSHE is cleared to 0, the flash memory control registers are deselected. In this case, the flash memory control register contents are retained

Bit 3

FLSHE		
0	Flash memory control registers deselected	(Initial value)
1	Flash memory control registers selected	

Bit 2—Reserved: Do not write 1 to this bit.

Bits 1 and 0—Internal Clock Select 1 and 0 (ICKS1, ICKS0): These bits control 8-bit timer operation. See section 12, 8-Bit Timers, for details.

23.6 On-Board Programming Modes

When pins are set to on-board programming mode, program/erase/verify operations can be performed on the on-chip flash memory. There are two on-board programming modes: boot mode and user program mode. The pin settings for transition to each of these modes are shown in table 23.6. For a diagram of the transitions to the various flash memory modes, see figure 23.3.

Only advanced mode setting is possible for boot mode.

In the case of user program mode, established in advanced mode or normal mode, depending on the setting of the MD0 pin. In normal mode, only programming of a 56-kbyte area of flash memory is possible.

Table 23.6 Setting On-Board Programming Modes

_	_		-	
	п.	_	_	_

Mode Name	lame CPU Operating Mode		MD0	P92	P91	P90
Boot mode	Advanced mode	0	0	1*	1*	1*
User program mode	r program mode Advanced mode		0	_	_	_
	Normal mode			_	_	_

Note: * Can be used as I/O ports after boot mode is initiated.

23.6.1 Boot Mode

When boot mode is used, the flash memory programming control program must be prepared in the host beforehand. The channel 1 SCI to be used is set to asynchronous mode.

When a reset-start is executed after the chip pins have been set to boot mode, the boot program built into the chip is started and the programming control program prepared in the host is serially transmitted to the chip via the SCI. In the chip, the programming control program received via the SCI is written into the programming control program area in on-chip RAM. After the transfer is completed, control branches to the start address of the programming control program area and the programming control program execution state is entered (flash memory programming is performed).

The transferred programming control program must therefore include coding that follows the programming algorithm given later.

The system configuration in boot mode is shown in figure 23.7, and the boot program mode execution procedure in figure 23.8.

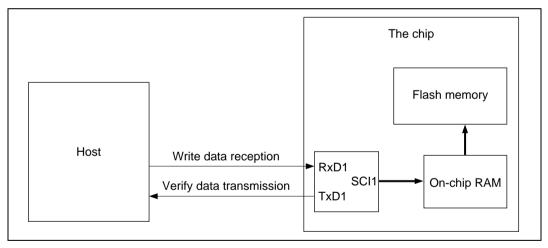


Figure 23.7 System Configuration in Boot Mode

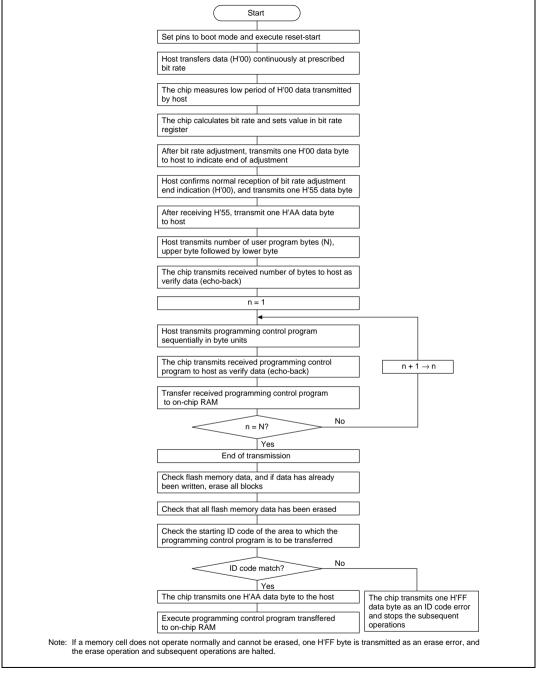


Figure 23.8 Boot Mode Execution Procedure

Automatic SCI Bit Rate Adjustment

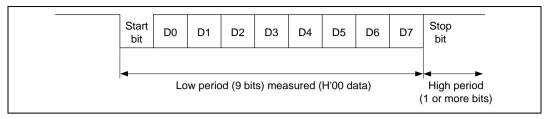


Figure 23.9 Automatic SCI Bit Rate Adjustment

When boot mode is initiated, the chip measures the low period of the asynchronous SCI communication data (H'00) transmitted continuously from the host. The SCI transmit/receive format should be set as follows: 8-bit data, 1 stop bit, no parity. The chip calculates the bit rate of the transmission from the host from the measured low period, and transmits one H'00 byte to the host to indicate the end of bit rate adjustment. The host should confirm that this adjustment end indication (H'00) has been received normally, and transmit one H'55 byte to the chip. If reception cannot be performed normally, initiate boot mode again (reset), and repeat the above operations. Depending on the host's transmission bit rate and the chips system clock frequency, there will be a discrepancy between the bit rates of the host and the chip. To ensure correct SCI operation, the host's transfer bit rate should be set to (4800, 9600, or 19200) bps.

Table 23.7 shows typical host transfer bit rates and system clock frequencies for which automatic adjustment of the chips bit rate is possible. The boot program should be executed within this system clock range.

Table 23.7 System Clock Frequencies for Which Automatic Adjustment of the Chips Bit Rate Is Possible

Host Bit Rate	System Clock Frequency for Which Automatic Adjustment of Bit Rate Is Possible
19200 bps	8 MHz to 20 MHz
9600 bps	4 MHz to 20 MHz
4800 bps	2 MHz to 18 MHz

On-Chip RAM Area Divisions in Boot Mode

In boot mode, the 1920-byte area from H'(FF)E880 to H'(FF) EFFF and the 128-byte area from H'(FF)FF00 to H'(FF)FF7F is reserved for use by the boot program, as shown in figure 23.10. The area to which the programming control program is transferred is H'(FF)E080 to H'(FF)E87F (2048 bytes). In the 64-kbyte version, this is a reserved area that is used only during the boot mode. However, the 8-byte area from H'(FF)E080 to H'(FF)E087 is reserved for ID codes as shown in

Figure 23.10. The boot program area can be used when the programming control program transferred into the RAM enters the execution state. A stack area should be set up as required.

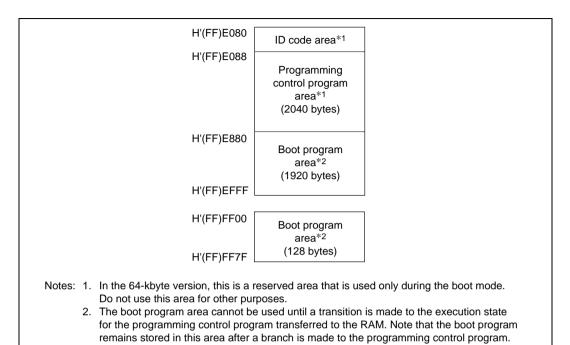


Figure 23.10 RAM Areas in Boot Mode

In the boot mode of this chip, the content in the 8-byte ID code area shown below is confirmed so that whether or not there is a programming control program that corresponds to the chip can be checked.

H'(FF)E080	40	FE	64	66	32	31	34	39
	↑ (produc							
H'(FF)E088 and above	Instructio	n code fo	r write co	ontrol prog	gram			

When a new programming control program for use in boot mode is created, add the 8-byte ID code described above to the head of the program.

Notes on Use of Boot Mode

 When the chip comes out of reset in boot mode, it measures the low period of the input at the SCI's RxD1 pin. The reset should end with RxD1 high. After the reset ends, it takes about 100 states for the chip to get ready to measure the low period of the RxD1 input.

- In boot mode, if any data has been programmed into the flash memory (if all data is not 1), all
 flash memory blocks are erased. Boot mode is for use when user program mode is unavailable,
 such as the first time on-board programming is performed, or if the program activated in user
 program mode is accidentally erased.
- Interrupts cannot be used while the flash memory is being programmed or erased.
- The RxD1 and TxD1 pins should be pulled up on the board.
- Before branching to the programming control program (RAM area H'(FF)E088), the chip terminates transmit and receive operations by the on-chip SCI (channel 1) (by clearing the RE and TE bits in SCR to 0), but the adjusted bit rate value remains set in BRR. The transmit data output pin, TxD1, goes to the high-level output state (P84DDR = 1, P84DR = 1).

The contents of the CPU's internal general registers are undefined at this time, so these registers must be initialized immediately after branching to the programming control program. In particular, since the stack pointer (SP) is used implicitly in subroutine calls, etc., a stack area must be specified for use by the programming control program.

The initial values of other on-chip registers are not changed.

- Boot mode can be entered by making the pin settings shown in table 23.6 and executing a
 reset-start.
 - When the chip detects the boot mode setting at reset release*1, P92, P91, and P90 can be used as I/O ports.
 - Boot mode can be cleared by driving the reset pin low, waiting at least 20 states, then setting the mode pins, and executing reset release*1. Boot mode can also be cleared by a WDT overflow reset. The mode pin input levels must not be changed in boot mode.
- If the mode pin input levels are changed (for example, from low to high) during a reset, the state of ports with multiplexed address functions and bus control output pins (AS, RD, HWR) will change according to the change in the microcomputer's operating mode*2.
 - Therefore, care must be taken to make pin settings to prevent these pins from becoming output signal pins during a reset, or to prevent collision with signals outside the microcomputer.
- Notes: 1. Mode pins input must satisfy the mode programming setup time ($t_{MDS} = 4$ states) with respect to the reset release timing.
 - 2. Ports with multiplexed address functions will output a low level as the address signal if mode pin setting is for mode 1 is entered during a reset. In other modes, the port pins go to the high-impedance state. The bus control output signals will output a high level if mode pin setting is for mode 1 is entered during a reset. In other modes, the port pins go to the high-impedance state.



23.6.2 User Program Mode

When set to user program mode, the chip can program and erase its flash memory by executing a user program/erase control program. Therefore, on-board reprogramming of the on-chip flash memory can be carried out by providing an on-board means of supplying programming data, and storing a program/erase control program in part of the program area as necessary.

To select user program mode, select a mode that enables the on-chip flash memory (mode 2 or 3). In this mode, on-chip supporting modules other than flash memory operate as they normally would in mode 2 and 3

The flash memory itself cannot be read while the SWE bit is set to 1 to perform programming or erasing, so the control program that performs programming and erasing should be run in on-chip RAM or external memory.

Figure 23.11 shows the procedure for executing the program/erase control program when transferred to on-chip RAM.

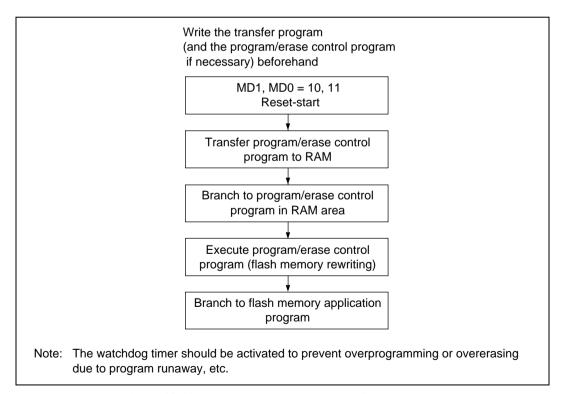


Figure 23.11 User Program Mode Execution Procedure

23.7 Programming/Erasing Flash Memory

In the on-board programming modes, flash memory programming and erasing is performed by software, using the CPU. There are four flash memory operating modes: program mode, erase mode, program-verify mode, and erase-verify mode. Transitions to these modes can be made by setting the PSU and ESU bits in FLMCR2, and the P, E, PV, and EV bits in FLMCR1.

The flash memory cannot be read while being programmed or erased. Therefore, the program that controls flash memory programming/erasing (the programming control program) should be located and executed in on-chip RAM or external memory.

- Notes: 1. Operation is not guaranteed if setting/resetting of the SWE, EV, PV, E, and P bits in FLMCR1, and the ESU and PSU bits in FLMCR2, is executed by a program in flash memory.
 - 2. Perform programming in the erased state. Do not perform additional programming on previously programmed addresses.

23.7.1 Program Mode

Follow the procedure shown in the program/program-verify flowchart in figure 23.12 to write data or programs to flash memory. Performing program operations according to this flowchart will enable data or programs to be written to flash memory without subjecting the device to voltage stress or sacrificing program data reliability. Programming should be carried out 128 bytes at a time.

The wait times $(x, y, z1, z2, z3, \alpha, \beta, \gamma, \epsilon, \eta, \theta)$ after setting/clearing individual bits in flash memory control registers 1 and 2 (FLMCR1, FLMCR2) and the maximum number of writes (N), see section 26.2.6, Flash Memory Characteristics.

Following the elapse of (x) µs or more after the SWE bit is set to 1 in flash memory control register 1 (FLMCR1), 128-byte program data is stored in the program data area and reprogram data area, and the 128-byte data in the reprogram data area written consecutively to the write addresses. The lower 8 bits of the first address written to must be H'00 or H'80. 128 consecutive byte data transfers are performed. The program address and program data are latched in the flash memory. A 128-byte data transfer must be performed even if writing fewer than 128 bytes; in this case, H'FF data must be written to the extra addresses.

Next, the watchdog timer is set to prevent overprogramming in the event of program runaway, etc. Set a value greater than $(y + z2 + \alpha + \beta)$ μ s as the WDT overflow period. After this, preparation for program mode (program setup) is carried out by setting the PSU bit in FLMCR2, and after the elapse of (y) μ s or more, the operating mode is switched to program mode by setting the P bit in FLMCR1. The time during which the P bit is set is the flash memory programming time. Make a



program setting so that the time for one programming operation is within the range of (z1), (z2) or (z3) μ s.

23.7.2 Program-Verify Mode

In program-verify mode, the data written in program mode is read to check whether it has been correctly written in the flash memory.

After the elapse of a given programming time, the programming mode is exited (the P bit in FLMCR1 is cleared, then the PSU bit in FLMCR2 is cleared at least (α) μs later). The watchdog timer is cleared after the elapse of (β) us or more, and the operating mode is switched to programverify mode by setting the PV bit in FLMCR1. Before reading in program-verify mode, a dummy write of H'FF data should be made to the addresses to be read. The dummy write should be executed after the elapse of (γ) us or more. When the flash memory is read in this state (verify data is read in 16-bit units), the data at the latched address is read. Wait at least (E) us after the dummy write before performing this read operation. Next, the originally written data is compared with the verify data, and reprogram data is computed (see figure 23.12) and transferred to the reprogram data area. After 128 bytes of data have been verified, exit program-verify mode, wait for at least (n) us. If the programming count is less than 6, the 128-byte data in the additional program data area should be written consecutively to the write addresses, and additional programming performed. Next clear the SWE bit in FLMCR1, and wait at least (θ) µs. If reprogramming is necessary, set program mode again, and repeat the program/program-verify sequence as before. However, ensure that the program/program-verify sequence is not repeated more than (N) times on the same bits.

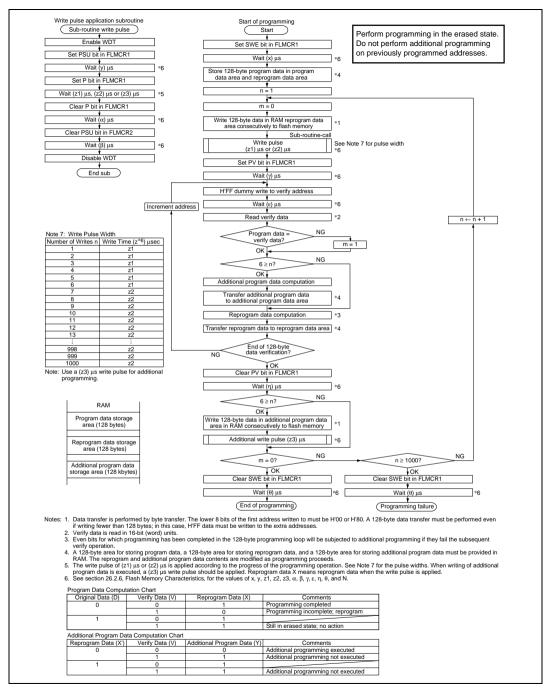


Figure 23.12 Program/Program-Verify Flowchart

23.7.3 Erase Mode

Flash memory erasing should be performed block by block following the procedure shown in the erase/erase-verify flowchart (single-block erase) shown in figure 23.13.

The wait times $(x, y, z, \alpha, \beta, \gamma, \epsilon, \eta, \theta)$ after setting/clearing individual bits in flash memory control registers 1 and 2 (FLMCR1, FLMCR2) and the maximum number of erase (N), see section 26.2.6, Flash Memory Characteristics.

To perform data or program erasure, make a 1 bit setting for the flash memory area to be erased in erase block register 1 or 2 (EBR1 or EBR2) at least (x) μ s after setting the SWE bit to 1 in flash memory control register 1 (FLMCR1). Next, the watchdog timer is set to prevent overerasing in the event of program runaway, etc. Set a value greater than $(y + z + \alpha + \beta)$ ms as the WDT overflow period. After this, preparation for erase mode (erase setup) is carried out by setting the ESU bit in FLMCR2, and after the elapse of (y) μ s or more, the operating mode is switched to erase mode by setting the E bit in FLMCR1. The time during which the E bit is set is the flash memory erase time. Ensure that the erase time does not exceed (z) ms.

Note: With flash memory erasing, preprogramming (setting all data in the memory to be erased to 0) is not necessary before starting the erase procedure.

23.7.4 Erase-Verify Mode

In erase-verify mode, data is read after memory has been erased to check whether it has been correctly erased.

After the elapse of the erase time, erase mode is exited (the E bit in FLMCR1 is cleared, then the ESU bit in FLMCR2 is cleared at least (α) μ s later), the watchdog timer is cleared after the elapse of (β) μ s or more, and the operating mode is switched to erase-verify mode by setting the EV bit in FLMCR1. Before reading in erase-verify mode, a dummy write of HTFF data should be made to the addresses to be read. The dummy write should be executed after the elapse of (γ) μ s or more. When the flash memory is read in this state (verify data is read in 16-bit units), the data at the latched address is read. Wait at least (ϵ) μ s after the dummy write before performing this read operation. If the read data has been erased (all 1), a dummy write is performed to the next address, and erase-verify is performed. If the read data has not been erased, set erase mode again, and repeat the erase/erase-verify sequence in the same way. However, ensure that the erase/erase-verify sequence is not repeated more than (N) times. When verification is completed, exit erase-verify mode, and wait for at least (η) μ s. If erasure has been completed on all the erase blocks, clear the SWE bit in FLMCR1, and wait (θ) μ s. If there are any unerased blocks, make a 1 bit setting in EBR1 or EBR2 for the flash memory area to be erased, and repeat the erase/erase-verify sequence in the same way.

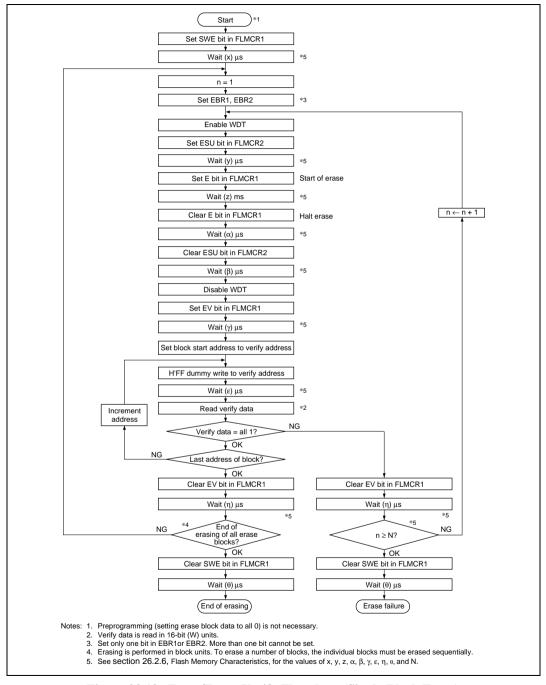


Figure 23.13 Erase/Erase-Verify Flowchart (Single-Block Erase)

23.8 **Flash Memory Protection**

There are three kinds of flash memory program/erase protection: hardware protection, software protection, and error protection.

23.8.1 Hardware Protection

Hardware protection refers to a state in which programming/erasing of flash memory is forcibly disabled or aborted. Hardware protection is reset by settings in flash memory control registers 1 and 2 (FLMCR1, FLMCR2) and erase block registers 1 and 2 (EBR1, EBR2). (See table 23.8.)

Table 23.8 Hardware Protection

		Fun	ctions
Item	Description	Program	Erase
Reset/standby protection	 In a reset (including a WDT overflow reset) and in hardware standby mode, software standby mode, subscleep mode, and watch mode, FLMCR1, FLMCR2, EBR1, and EBR2 are initialized, and the program/erase-protected state is entered. In a reset via the RES pin, the reset state is not entered unless the RES pin is held low until oscillation stabilizes after powering on. In the case of a reset during operation, hold the RES pin low for the RES pulse width specified in the AC Characteristics section. 	Yes	Yes

23.8.2 **Software Protection**

Software protection can be implemented by setting the SWE bit in FLMCR1 and erase block registers 1 and 2 (EBR1, EBR2). When software protection is in effect, setting the P or E bit in flash memory control register 1 (FLMCR1) does not cause a transition to program mode or erase mode. (See table 23.9.)

Table 23.9 Software Protection

	Functions			
Item	Description	Program	Erase	
SWE bit protection	Clearing the SWE bit to 0 in FLMCR1 sets the program/erase-protected state for all blocks. (Execute in on-chip RAM or external memory.)	Yes	Yes	
Block specification protection	 Erase protection can be set for individual blocks by settings in erase block registers 1 and 2 (EBR1, EBR2). 	_	Yes	
	 Setting EBR1 and EBR2 to H'00 places all blocks in the erase-protected state. 			

23.8.3 Error Protection

In error protection, an error is detected when MCU runaway occurs during flash memory programming/erasing, or operation is not performed in accordance with the program/erase algorithm, and the program/erase operation is aborted. Aborting the program/erase operation prevents damage to the flash memory due to overprogramming or overerasing.

If the MCU malfunctions during flash memory programming/erasing, the FLER bit is set to 1 in FLMCR2 and the error protection state is entered. The FLMCR1, FLMCR2, EBR1, and EBR2 settings are retained, but program mode or erase mode is aborted at the point at which the error occurred. Program mode or erase mode cannot be re-entered by re-setting the P or E bit. However, PV and EV bit setting is enabled, and a transition can be made to verify mode.

FLER bit setting conditions are as follows:

- When flash memory is read during programming/erasing (including a vector read or instruction fetch)
- Immediately after exception handling (excluding a reset) during programming/erasing
- When a SLEEP instruction (including software standby, sleep, subactive, subsleep and watch mode) is executed during programming/erasing
- When the bus is released during programming/erasing

Error protection is released only by a reset and in hardware standby mode.

Figure 23.14 shows the flash memory state transition diagram.



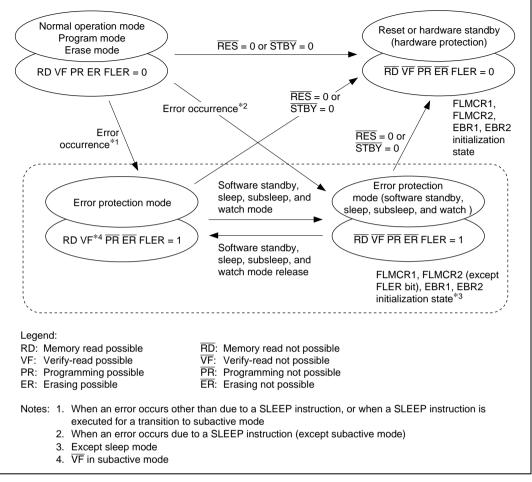


Figure 23.14 Flash Memory State Transitions

23.9 Interrupt Handling when Programming/Erasing Flash Memory

All interrupts, including NMI input is disabled when flash memory is being programmed or erased (when the P or E bit is set in FLMCR1), and while the boot program is executing in boot mode^{*1}, to give priority to the program or erase operation. There are three reasons for this:

- 1. Interrupt during programming or erasing might cause a violation of the programming or erasing algorithm, with the result that normal operation could not be assured.
- 2. In the interrupt exception handling sequence during programming or erasing, the vector would not be read correctly*2, possibly resulting in MCU runaway.
- 3. If interrupt occurred during boot program execution, it would not be possible to execute the normal boot mode sequence.

For these reasons, in on-board programming mode alone there are conditions for disabling interrupt, as an exception to the general rule. However, this provision does not guarantee normal erasing and programming or MCU operation. All interrupt requests, including NMI input, must therefore be disabled inside and outside the MCU when programming or erasing flash memory. Interrupt is also disabled in the error-protection state while the P or E bit remains set in FLMCR1.

- Notes: 1. Interrupt requests must be disabled inside and outside the MCU until the programming control program has completed programming.
 - 2. The vector may not be read correctly in this case for the following two reasons:
 - If flash memory is read while being programmed or erased (while the P or E bit is set in FLMCR1), correct read data will not be obtained (undetermined values will be returned).
 - If the interrupt entry in the vector table has not been programmed yet, interrupt exception handling will not be executed correctly.



23.10 Flash Memory Programmer Mode

23.10.1 Programmer Mode Setting

Programs and data can be written and erased in programmer mode as well as in the on-board programming modes. In programmer mode, the on-chip ROM can be freely programmed using a PROM programmer that supports Renesas Technology microcomputer device types with 128-kbyte or 64-kbyte on-chip flash memory*. See section 23.10.10, Notes on Memory Programming and section 23.11, Flash Memory Programming and Erasing Precautions, for notes on programmer mode. Flash memory read mode, auto-program mode, auto-erase mode, and status read mode are supported with these device types. In auto-program mode, auto-erase mode, and status read mode, a status polling procedure is used, and in status read mode, detailed internal signals are output after execution of an auto-program or auto-erase operation.

Table 23.10 shows programmer mode pin settings.

Note: * Use products of the H8S/2148 A-mask version, H8S/2147 A-mask version, and H8S/2144 A-mask version (in either 5-V or 3-V version) with the writing voltage for PROM programmer set to 3.3 V. Do not use products other than the A-mask version with 3.3V PROM programmer setting.

Table 23.10 Programmer Mode Pin Settings

Pin Names	Setting/External Circuit Connection
Mode pins: MD1, MD0	Low-level input to MD1, MD0
STBY pin	High-level input (Hardware standby mode not set)
RES pin	Power-on reset circuit
XTAL and EXTAL pins	Oscillation circuit
Other setting pins: P97, P92, P91, P90, P67	Low-level input to p92, p67, high-level input to P97, P91, P90

23.10.2 Socket Adapters and Memory Map

In programmer mode, a socket adapter is mounted on the writer programmer to match the package concerned. Socket adapters are available for each writer manufacturer supporting Renesas Technology microcomputer device types with 128-kbyte or 64-kbyte on-chip flash memory $(V_{pp} = 3.3 \text{ V})$.

Figure 23.15 shows the memory map in programmer mode. For pin names in programmer mode, see section 1.3.2, Pin Functions in Each Operating Mode.

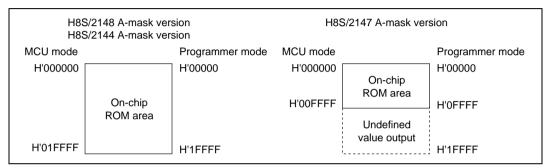


Figure 23.15 Memory Map in Programmer Mode

23.10.3 Programmer Mode Operation

Table 23.11 shows how the different operating modes are set when using programmer mode, and table 23.12 lists the commands used in programmer mode. Details of each mode are given below.

Memory Read Mode
 Memory read mode supports byte reads.

Auto-Erase Mode

- Auto-Program Mode
 Auto-program mode supports programming of 128 bytes at a time. Status polling is used to confirm the end of auto-programming.
- Auto-erase mode supports automatic erasing of the entire flash memory. Status polling is used to confirm the end of auto-erasing.
 - Status Read Mode

 Status polling is used for auto-programming and auto-erasing, and normal termination can be confirmed by reading the FO6 signal. In status read mode, error information is output if an error occurs.



Table 23.11 Settings for Each Operating Mode in Programmer Mode

Pin Names

Mode	CE	ŌĒ	WE	FO0 to FO7	FA0 to FA17
Read	L	L	Н	Data output	Ain
Output disable	L	Н	Н	Hi-Z	X
Command write	L	Н	L	Data input	Ain*2
Chip disable*1	Н	Χ	Х	Hi-Z	Χ

Notes: 1. Chip disable is not a standby state; internally, it is an operation state.

2. Ain indicates that there is also address input in auto-program mode.

Table 23.12 Programmer Mode Commands

	Number	1st Cycle			2nd Cycle		
Command Name	of Cycles	Mode	Address	Data	Mode	Address	Data
Memory read mode	1 + n	Write	Х	H'00	Read	RA	Dout
Auto-program mode	129	Write	Х	H'40	Write	WA	Din
Auto-erase mode	2	Write	Х	H'20	Write	Χ	H'20
Status read mode	2	Write	Χ	H'71	Write	Χ	H'71

Notes: 1. In auto-program mode. 129 cycles are required for command writing by a simultaneous 128-byte write.

2. In memory read mode, the number of cycles depends on the number of address write cycles (n).

23.10.4 Memory Read Mode

- After the end of an auto-program, auto-erase, or status read operation, the command wait state
 is entered. To read memory contents, a transition must be made to memory read mode by
 means of a command write before the read is executed.
- Command writes can be performed in memory read mode, just as in the command wait state.
- Once memory read mode has been entered, consecutive reads can be performed.
- After power-on, memory read mode is entered.

Table 23.13 AC Characteristics in Memory Read Mode

Item	Symbol	Min	Max	Unit
Command write cycle	t _{nxtc}	20	_	μs
CE hold time	t _{ceh}	0		ns
CE setup time	t _{ces}	0	_	ns
Data hold time	t _{dh}	50	_	ns
Data setup time	t _{ds}	50		ns
Write pulse width	t _{wep}	70	_	ns
WE rise time	t _r	_	30	ns
WE fall time	t _f	_	30	ns

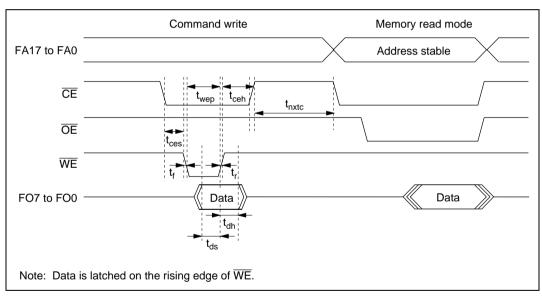


Figure 23.16 Memory Read Mode Timing Waveforms after Command Write

Table 23.14 AC Characteristics when Entering Another Mode from Memory Read Mode

Item	Symbol	Min	Max	Unit
Command write cycle	t _{nxtc}	20	_	μs
CE hold time	t _{ceh}	0	_	ns
CE setup time	t _{ces}	0	_	ns
Data hold time	t _{dh}	50	_	ns
Data setup time	t _{ds}	50	_	ns
Write pulse width	t _{wep}	70	_	ns
WE rise time	t,	_	30	ns
WE fall time	t _f	_	30	ns

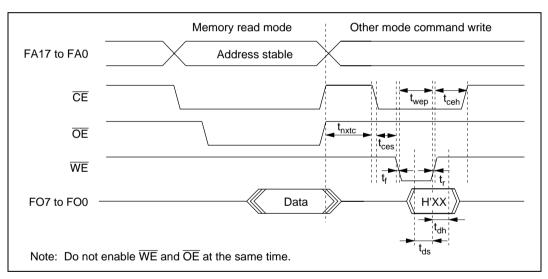


Figure 23.17 Timing Waveforms when Entering Another Mode from Memory Read Mode

Table 23.15 AC Characteristics in Memory Read Mode

Item	Symbol	Min	Max	Unit
Access time	t _{acc}	_	20	μs
CE output delay time	t _{ce}	_	150	ns
OE output delay time	t _{oe}	_	150	ns
Output disable delay time	t _{df}	_	100	ns
Data output hold time	t _{oh}	5	_	ns

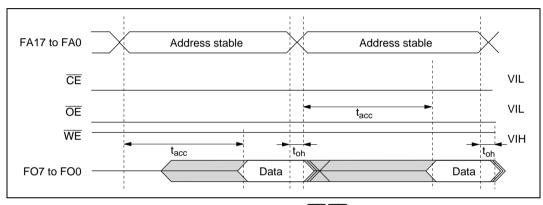


Figure 23.18 Timing Waveforms for $\overline{\text{CE}}/\overline{\text{OE}}$ Enable State Read

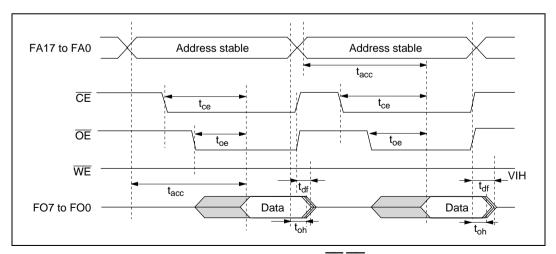


Figure 23.19 Timing Waveforms for $\overline{\text{CE}}/\overline{\text{OE}}$ Clocked Read

23.10.5 Auto-Program Mode

AC Characteristics

Table 23.16 AC Characteristics in Auto-Program Mode

Item	Symbol	Min	Max	Unit	
Command write cycle	t _{nxtc}	20	_	μs	
CE hold time	t _{ceh}	0	_	ns	
CE setup time	t _{ces}	0	_	ns	
Data hold time	t _{dh}	50	_	ns	
Data setup time	t _{ds}	50	_	ns	
Write pulse width	t _{wep}	70	_	ns	
Status polling start time	t _{wsts}	1	_	ms	
Status polling access time	t _{spa}	_	150	ns	
Address setup time	t _{as}	0	_	ns	
Address hold time	t _{ah}	60	_	ns	
Memory write time	t _{write}	1	3000	ms	
WE rise time	t _r	_	30	ns	
WE fall time	t _f	_	30	ns	

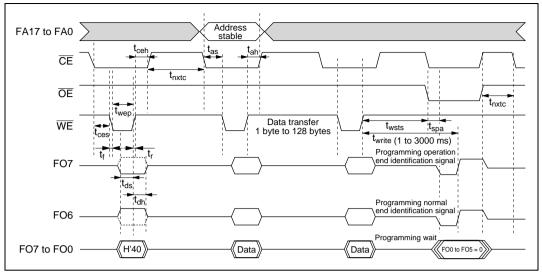


Figure 23.20 Auto-Program Mode Timing Waveforms

Notes on Use of Auto-Program Mode

- In auto-program mode, 128 bytes are programmed simultaneously. This should be carried out by executing 128 consecutive byte transfers.
- A 128-byte data transfer is necessary even when programming fewer than 128 bytes. In this case, H'FF data must be written to the extra addresses.
- The lower 8 bits of the transfer address must be H'00 or H'80. If a value other than an effective
 address is input, processing will switch to a memory write operation but a write error will be
 flagged.
- Memory address transfer is performed in the second cycle (figure 23.20). Do not perform transfer after the second cycle.
- Do not perform a command write during a programming operation.
- Perform one auto-programming operation for a 128-byte block for each address. Characteristics are not guaranteed for two or more programming operations.
- Confirm normal end of auto-programming by checking FO6. Alternatively, status read mode
 can also be used for this purpose (FO7 status polling uses the auto-program operation end
 identification pin).
- The status polling FO6 and FO7 pin information is retained until the next command write. Until the next command write is performed, reading is possible by enabling $\overline{\text{CE}}$ and $\overline{\text{OE}}$.



23.10.6 Auto-Erase Mode

AC Characteristics

Table 23.17 AC Characteristics in Auto-Erase Mode

Item	Symbol	Min	Max	Unit	
Command write cycle	t _{nxtc}	20	_	μs	
CE hold time	t _{ceh}	0	_	ns	
CE setup time	t _{ces}	0	_	ns	
Data hold time	t _{dh}	50	_	ns	
Data setup time	t _{ds}	50	_	ns	
Write pulse width	t _{wep}	70	_	ns	
Status polling start time	t _{ests}	1	_	ms	
Status polling access time	t _{spa}	_	150	ns	
Memory erase time	t _{erase}	100	40000	ms	
WE rise time	t _r	_	30	ns	
WE fall time	t _f	_	30	ns	

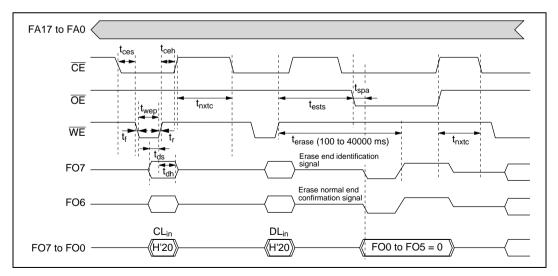


Figure 23.21 Auto-Erase Mode Timing Waveforms

Notes on Use of Auto-Erase-Program Mode

- Auto-erase mode supports only entire memory erasing.
- Do not perform a command write during auto-erasing.
- Confirm normal end of auto-erasing by checking FO6. Alternatively, status read mode can also be used for this purpose (FO7 status polling uses the auto-erase operation end identification pin).
- The status polling FO6 and FO7 pin information is retained until the next command write. Until the next command write is performed, reading is possible by enabling \overline{CE} and \overline{OE} .

23.10.7 Status Read Mode

- Status read mode is used to identify what type of abnormal end has occurred. Use this mode when an abnormal end occurs in auto-program mode or auto-erase mode.
- The return code is retained until a command write for other than status read mode is performed.

Table 23.18 AC Characteristics in Status Read Mode

Item	Symbol	Min	Max	Unit
Command write cycle	t _{nxtc}	20	_	μs
CE hold time	t _{ceh}	0	_	ns
CE setup time	t _{ces}	0	_	ns
Data hold time	t _{dh}	50	_	ns
Data setup time	t _{ds}	50	_	ns
Write pulse width	t _{wep}	70	_	ns
OE output delay time	t _{oe}	_	150	ns
Disable delay time	t _{df}	_	100	ns
CE output delay time	t _{ce}	_	150	ns
WE rise time	t _r	_	30	ns
WE fall time	t _f	_	30	ns

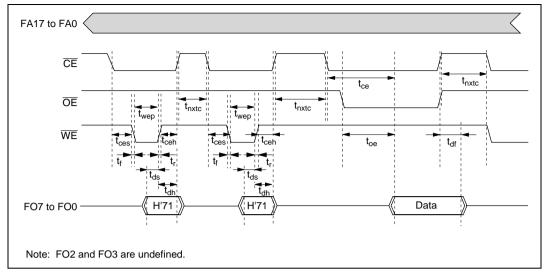


Figure 23.22 Status Read Mode Timing Waveforms

Table 23.19 Status Read Mode Return Commands

Pin Name	F07	FO6	FO5	FO4	FO3	FO2	FO1	FO0
Attribute	Normal end identification	Command error	Program- ming error	Erase error	_	_	Program- ming or erase count exceeded	Effective address error
Initial value	0	0	0	0	0	0	0	0
Indications	Normal end: 0 Abnormal	Command error: 1 Otherwise: 0	Program- ming error: 1	Erase error: 1 Otherwise: 0	_	_	Count exceeded: 1 Otherwise: 0	Effective address error: 1
	end: 1	Otherwise. 0	Otherwise: 0		,		Otherwise. 0	Otherwise: 0

Note: FO2 and FO3 are undefined.

23.10.8 Status Polling

- The FO7 status polling flag indicates the operating status in auto-program or auto-erase mode.
- The FO6 status polling flag indicates a normal or abnormal end in auto-program or auto-erase mode.

Table 23.20 Status Polling Output Truth Table

Pin Names	Internal Operation in Progress	Abnormal End	_	Normal End
FO7	0	1	0	1
FO6	0	0	1	1
FO0 to FO5	0	0	0	0

23.10.9 Programmer Mode Transition Time

Commands cannot be accepted during the oscillation stabilization period or the programmer mode setup period. After the programmer mode setup time, a transition is made to memory read mode.

Table 23.21 Command Wait State Transition Time Specifications

Item	Symbol	Min	Max	Unit	
Standby release (oscillation stabilization time)	t _{osc1}	20	_	ms	,
Programmer mode setup time	t _{bmv}	10	_	ms	
V _{cc} hold time	t _{dwn}	0	_	ms	

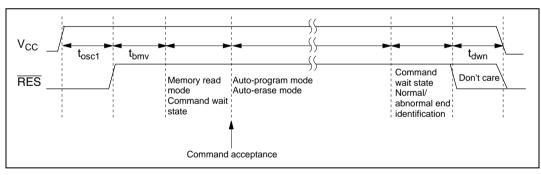


Figure 23.23 Oscillation Stabilization Time, Programmer Mode Setup Time, and Power Supply Fall Sequence

23.10.10 Notes on Memory Programming

- When programming addresses which have previously been programmed, carry out autoerasing before auto-programming.
- When performing programming using programmer mode on a chip that has been programmed/erased in an on-board programming mode, auto-erasing is recommended before carrying out auto-programming.
- Notes: 1. The flash memory is initially in the erased state when the device is shipped by Renesas. For other chips for which the erasure history is unknown, it is recommended that autoerasing be executed to check and supplement the initialization (erase) level.
 - 2. Auto-programming should be performed once only on the same address block.

23.11 Flash Memory Programming and Erasing Precautions

Precautions concerning the use of on-board programming mode and programmer mode are summarized below.

Use the specified voltages and timing for programming and erasing.

Applied voltages in excess of the rating can permanently damage the device. For a PROM programmer, use Renesas Technology microcomputer device types with 128-kbyte or 64-kbyte on-chip flash memory that support a 3.3-V programming voltage.

Do not select the HN28F101, or use a programming voltage of 5.0 V for the PROM programmer, and only use the specified socket adapter. Incorrect use will result in damaging the device.

Powering on and off

When applying or disconnecting $V_{\rm cc}$, fix the \overline{RES} pin low and place the flash memory in the hardware protection state.

The power-on and power-off timing requirements should also be satisfied in the event of a power failure and subsequent recovery.

Use the recommended algorithm when programming and erasing flash memory.

The recommended algorithm enables programming and erasing to be carried out without subjecting the device to voltage stress or sacrificing program data reliability. When setting the P or E bit in FLMCR1, the watchdog timer should be set beforehand as a precaution against program runaway, etc.

Do not set or clear the SWE bit during program execution in flash memory.

Wait for at least $100 \,\mu s$ after clearing the SWE bit before executing a program or reading data in flash memory. When the SWE bit is set, data in flash memory can be rewritten, but when SWE = 1 the flash memory can only be read in program-verify or erase-verify mode. Flash memory should only be accessed for verify operations (verification during programming/erasing). Do not clear the SWE bit during a program, erase, or verify operation.

Do not use interrupts while flash memory is being programmed or erased.

All interrupt requests, including NMI, should be disabled when programming and erasing flash memory to give priority to program/erase operations.

Do not perform additional programming. Erase the memory before reprogramming.

In on-board programming, perform only one programming operation on a 128-byte programming unit block. In programmer mode, too, perform only one programming operation on a 128-byte programming unit block. Programming should be carried out with the entire programming unit block erased

Before programming, check that the chip is correctly mounted in the PROM programmer.

Overcurrent damage to the device can result if the index marks on the PROM programmer socket, socket adapter, and chip are not correctly aligned.

Do not touch the socket adapter or chip during programming.

Touching either of these can cause contact faults and write errors.

23.12 Note on Switching from F-ZTAT Version to Mask ROM Version

The mask ROM version does not have the internal registers for flash memory control that are provided in the F-ZTAT version. Table 23.22 lists the registers that are present in the F-ZTAT version but not in the mask ROM version. If a register listed in table 23.22 is read in the mask ROM version, an undefined value will be returned. Therefore, if application software developed on the F-ZTAT version is switched to a mask ROM version product, it must be modified to ensure that the registers in table 23.22 have no effect

Table 23.22 Registers Present in F-ZTAT Version but Absent in Mask ROM Version

Register	Abbreviation	Address
Flash memory control register 1	FLMCR1	H'FF80
Flash memory control register 2	FLMCR2	H'FF81
Erase block register 1	EBR1	H'FF82
Erase block register 2	EBR2	H'FF83

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Section 24 Clock Pulse Generator

24.1 Overview

This LSI have a built-in clock pulse generator (CPG) that generates the system clock (ϕ) , the bus master clock, and internal clocks.

The clock pulse generator consists of an oscillator circuit, a duty adjustment circuit, clock selection circuit, medium-speed clock divider, bus master clock selection circuit, subclock input circuit, and waveform shaping circuit.

24.1.1 Block Diagram

Figure 24.1 shows a block diagram of the clock pulse generator.

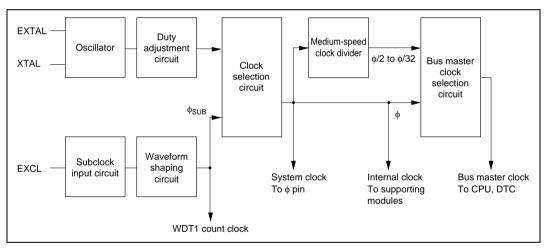


Figure 24.1 Block Diagram of Clock Pulse Generator

24.1.2 Register Configuration

The clock pulse generator is controlled by the standby control register (SBYCR) and low-power control register (LPWRCR). Table 24.1 shows the register configuration.

Table 24.1 CPG Registers

Name	Abbreviation	R/W	Initial Value	Address*
Standby control register	SBYCR	R/W	H'00	H'FF84
Low-power control register	LPWRCR	R/W	H'00	H'FF85

Note: * Lower 16 bits of the address.

24.2 Register Descriptions

24.2.1 Standby Control Register (SBYCR)

Bit	7	6	5	4	3	2	1	0
	SSBY	STS2	STS1	STS0	_	SCK2	SCK1	SCK0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	R/W		R/W	R/W	R/W

SBYCR is an 8-bit readable/writable register that performs power-down mode control.

Only bits 0 to 2 are described here. For a description of the other bits, see section 25.2.1, Standby Control Register (SBYCR).

SBYCR is initialized to H'00 by a reset and in hardware standby mode. It is not initialized in software standby mode.

Bits 2 to 0—System Clock Select 2 to 0 (SCK2 to SCK0): These bits select the bus master clock for high-speed mode and medium-speed mode.

When operating the device after a transition to subactive mode or watch mode bits SCK2 to SCK0 should all be cleared to 0.

Bit 2	Bit 1	Bit 0		
SCK2	SCK1	SCK0	Description	
0	0	0	Bus master is in high-speed mode	(Initial value)
		1	Medium-speed clock is φ/2	
	1	0	Medium-speed clock is φ/4	
		1	Medium-speed clock is φ/8	
1	0	0	Medium-speed clock is φ/16	
		1	Medium-speed clock is φ/32	
	1	_	_	

24.2.2 **Low-Power Control Register (LPWRCR)**

Bit	7	6	5	4	3	2	1	0	
	DTON	LSON	NESEL	EXCLE	_	_	_	_	
Initial value	0	0	0	0	0	0	0	0	_
Read/Write	R/W	R/W	R/W	R/W	_	_	_	_	

LPWRCR is an 8-bit readable/writable register that performs power-down mode control.

Only bit 4 is described here. For a description of the other bits, see section 25.2.2, Low-Power Control Register (LPWRCR).

LPWRCR is initialized to H'00 by a reset and in hardware standby mode. It is not initialized in software standby mode.

Bit 4—Subclock Input Enable (EXCLE): Controls subclock input from the EXCL pin.

Bit 4

EXCLE	Description	
0	Subclock input from EXCL pin is disabled	(Initial value)
1	Subclock input from EXCL pin is enabled	

24.3 Oscillator

Clock pulses can be supplied by connecting a crystal resonator, or by input of an external clock.

24.3.1 Connecting a Crystal Resonator

Circuit Configuration

A crystal resonator can be connected as shown in the example in figure 24.2. Select the damping resistance R_d according to table 24.2. An AT-cut parallel-resonance crystal should be used.

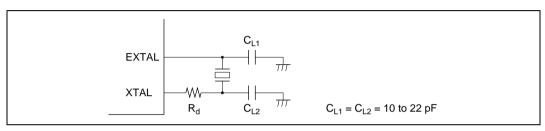


Figure 24.2 Connection of Crystal Resonator (Example)

Table 24.2 Damping Resistance Value

Frequency (MHz)	2	4	8	10	12	16	20	
$R_{d}(\Omega)$	1k	500	200	0	0	0	0	

Crystal resonator

Figure 24.3 shows the equivalent circuit of the crystal resonator. Use a crystal resonator that has the characteristics shown in table 24.3 and the same frequency as the system clock (ϕ) .

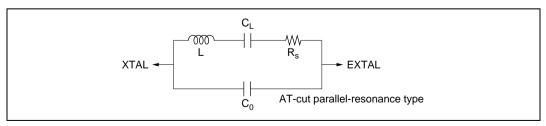


Figure 24.3 Crystal Resonator Equivalent Circuit

Table 24.3 Crystal Resonator Parameters

Frequency (MHz)	2	4	8	10	12	16	20	
R _s max (Ω)	500	120	80	70	60	50	40	
C₀ max (pF)	7	7	7	7	7	7	7	

Note on Board Design

When a crystal resonator is connected, the following points should be noted.

Other signal lines should be routed away from the oscillator circuit to prevent induction from interfering with correct oscillation. See figure 24.4.

When designing the board, place the crystal resonator and its load capacitors as close as possible to the XTAL and EXTAL pins.

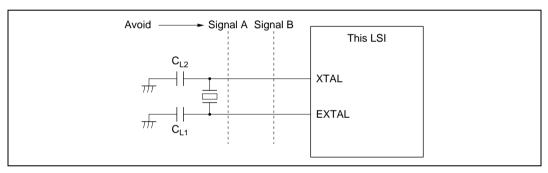


Figure 24.4 Example of Incorrect Board Design

24.3.2 External Clock Input

Circuit Configuration

An external clock signal can be input as shown in the examples in figure 24.5. If the XTAL pin is left open, make sure that stray capacitance is no more than 10 pF.

In example (b), make sure that the external clock is held high in standby mode, subactive mode, subsleep mode, and watch mode.

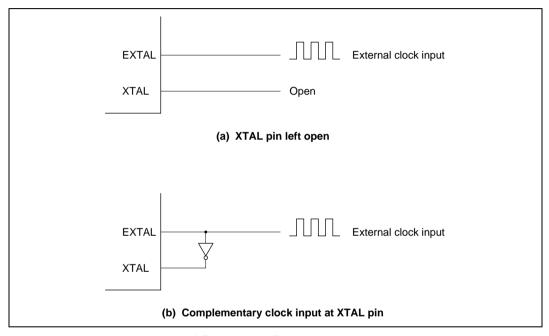


Figure 24.5 External Clock Input (Examples)

External Clock

The external clock signal should have the same frequency as the system clock (ϕ) .

Table 24.4 and figure 24.6 show the input conditions for the external clock.

Table 24.4 External Clock Input Conditions

		V _{cc} = 2.	7 to 5.5 V	$V_{cc} = 5$.	$V_{cc} = 5.0 \text{ V } \pm 10\%$			
Item	Symbol	Min	Max	Min	Max	Unit	Test Condi	tions
External clock input low pulse width	t _{EXL}	40	_	20	_	ns	Figure 24.6	
External clock input high pulse width	t _{EXH}	40	_	20	_	ns		
External clock rise time	t _{EXr}	_	10	_	5	ns	_	
External clock fall time	t _{EXf}	_	10	_	5	ns	_	
	t _{CL}	0.4	0.6	0.4	0.6	t _{cyc}	$\phi \ge 5 \text{ MHz}$	Figure 26.5
pulse width		80	_	80	_	ns	φ < 5 MHz	_
Clock high	t _{ch}	0.4	0.6	0.4	0.6	t _{cyc}	$\phi \ge 5 \text{ MHz}$	=
pulse width		80	_	80		ns	φ < 5 MHz	

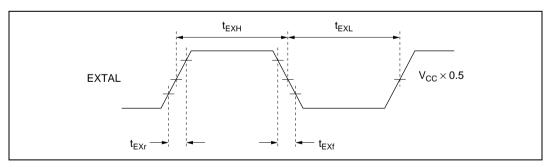


Figure 24.6 External Clock Input Timing

Table 24.5 shows the external clock output settling delay time, and figure 24.7 shows the external clock output settling delay timing. The oscillator and duty adjustment circuit have a function for adjusting the waveform of the external clock input at the EXTAL pin. When the prescribed clock signal is input at the EXTAL pin, internal clock signal output is fixed after the elapse of the

external clock output settling delay time (t_{DEXT}). As the clock signal output is not fixed during the t_{DEXT} period, the reset signal should be driven low to maintain the reset state.

Table 24.5 External Clock Output Settling Delay Time

Conditions: $V_{cc} = 2.7 \text{ V}$ to 5.5 V, $AV_{cc} = 2.7 \text{ V}$ to 5.5 V, $V_{ss} = AV_{ss} = 0 \text{ V}$

Item	Symbol	Min	Max	Unit	Notes
External clock output settling delay time	t *	500	_	μs	Figure 24.7

Note: * t_{DEXT} includes \overline{RES} pulse width (t_{RESW}) .

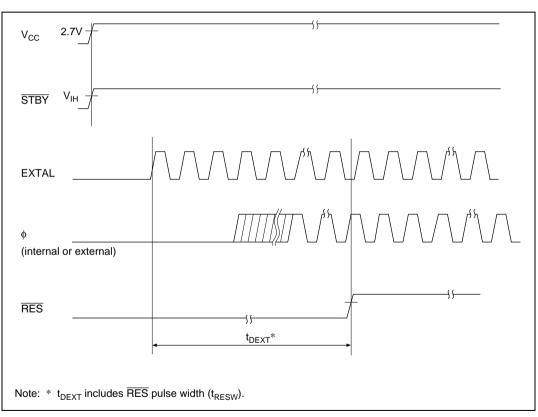


Figure 24.7 External Clock Output Settling Delay Timing

24.4 Duty Adjustment Circuit

When the oscillator frequency is 5 MHz or higher, the duty adjustment circuit adjusts the duty cycle of the clock signal from the oscillator to generate the system clock (ϕ).

24.5 Medium-Speed Clock Divider

The medium-speed clock divider divides the system clock to generate $\phi/2$, $\phi/4$, $\phi/8$, $\phi/16$, and $\phi/32$ clocks.

24.6 Bus Master Clock Selection Circuit

The bus master clock selection circuit selects the system clock (ϕ) or one of the medium-speed clocks (ϕ /2, ϕ /4, ϕ /8, ϕ /16, or ϕ /32) to be supplied to the bus master, according to the settings of bits SCK2 to SCK0 in SBYCR.

24.7 Subclock Input Circuit

The subclock input circuit controls the subclock input from the EXCL pin.

Inputting the Subclock

When a subclock is used, a 32.768 kHz external clock should be input from the EXCL pin. In this case, clear bit P96DDR to 0 in P9DDR and set bit EXCLE to 1 in LPWRCR.

The subclock input conditions are shown in table 24.6 and figure 24.8.

Table 24.6 Subclock Input Conditions

		'	$I_{\rm cc} = 2.7 \text{ to}$	5.5 V		
Item	Symbol	Min	Тур	Max	Unit	Test Conditions
Subclock input low pulse width	t _{EXCLL}	_	15.26	_	μs	Figure 24.8
Subclock input high pulse width	t _{EXCLH}	_	15.26	_	μs	
Subclock input rise time	t _{EXCLr}	_	_	10	ns	
Subclock input fall time	t _{EXCLf}	_	_	10	ns	

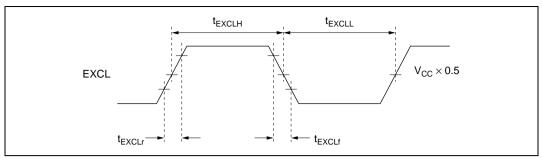


Figure 24.8 Subclock Input Timing

When Subclock Is Not Needed

Do not enable subclock input when the subclock is not needed

Note on Subclock Usage

In transiting to power-down mode, if at least two cycles of the 32-kHz clock are not input after the 32-kHz clock input is enabled (EXCLE = 1) until the SLEEP instruction is executed (power-down mode transition), the subclock input circuit is not initialized and an error may occur in the microcomputer.

Before power-down mode is entered with using the subclock, at least two cycle of the 32-kHz clock should be input after the 32-kHz clock input is enabled (EXCLE = 1).

As described in the hardware manual (clock pulse generator/subclock input circuit), when the subclock is not used, the subclock input should not be enabled (EXCLE = 0).

24.8 Subclock Waveform Shaping Circuit

To eliminate noise in the subclock input from the EXCL pin, this circuit samples the clock using a clock obtained by dividing the ϕ clock. The sampling frequency is set with the NESEL bit in LPWRCR. For details, see sections 24.2.2 and 25.2.2, Low-Power Control Register (LPWRCR). The clock is not sampled in subactive mode, subsleep mode, or watch mode.



24.9 Clock Selection Circuit

This circuit selects the system clock used in the MCU.

The clock signal generated in the EXTAL/XTAL pin oscillator is selected as the system clock when MCU is returned from high-speed mode, medium-speed mode, sleep mode, reset state, or standby mode.

In sub-active mode, sub-sleep mode, and watch mode, the sub-clock signal input from EXCL pin is selected as the system clock. In these modes, modules such as CPU, TMR0, TMR1, WDT0, WDT1, and I/O ports operate on the ϕ SUB clock. The count clock for each timer is a clock obtained by driving the ϕ SUB clock.

Section 25 Power-Down State

25.1 Overview

In addition to the normal program execution state, this LSI have a power-down state in which operation of the CPU and oscillator is halted and power dissipation is reduced. Low-power operation can be achieved by individually controlling the CPU, on-chip supporting modules, and so on.

This LSI has the following operating modes:

- 1. High-speed mode
- 2. Medium-speed mode
- 3. Subactive mode
- 4. Sleep mode
- 5. Subsleep mode
- Watch mode
- 7. Module stop mode
- 8. Software standby mode
- 9. Hardware standby mode

Of these, 2 to 9 are power-down modes. Sleep mode and subsleep mode are CPU modes, medium-speed mode is a CPU and bus master mode, subactive mode is a CPU, bus master, and on-chip supporting module mode, and module stop mode is an on-chip supporting module mode (including bus masters other than the CPU). Certain combinations of these modes can be set.

After a reset, the MCU is in high-speed mode and module stop mode (excluding the DTC).

Table 25.1 shows the internal chip states in each mode, and table 25.2 shows the conditions for transition to the various modes. Figure 25.1 shows a mode transition diagram.

Table 25.1 Internal States in Each Mode

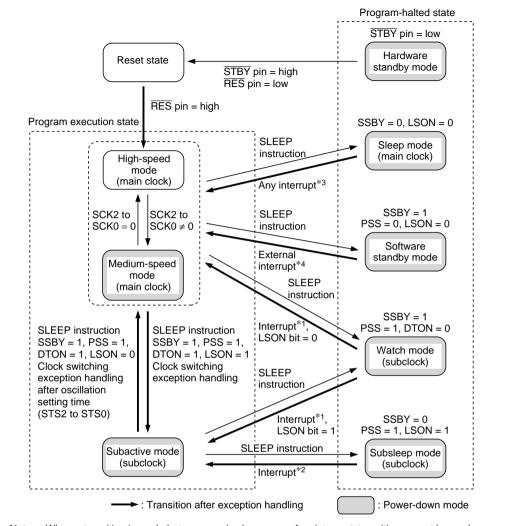
Func	tion	High- Speed	Medium- Speed	Sleep	Module Stop	Watch	Subactive	Subsleep	Software Standby	Hardware Standby
System clock oscillator		Function- ing	Function- ing	Function- ing	Function- ing	Halted	Halted	Halted	Halted	Halted
Subclock in	put	Function- ing	Function- ing	Function- ing	Function- ing	Function- ing	Function- ing	Function- ing	Halted	Halted
CPU operation	Instruc- tions	Function- ing	Medium- speed	Halted	Function- ing	Halted	Subclock operation	Halted	Halted	Halted
	Registers	Function- ing	Medium- speed	Retained	Function- ing	Retained	Subclock operation	Retained	Retained	Undefined
External interrupts	NMI IRQ0	Function- ing	Function- ing	Function- ing	Function- ing	Function- ing	Function- ing	Function- ing	Function- ing	Halted
	IRQ1									
	IRQ2									
On-chip supporting module	DTC	Function- ing	Medium- speed	Function- ing	Function- ing/halted (retained)	Halted (retained)	Halted (retained)	Halted (retained)	Halted (retained)	Halted (reset)
operation	WDT1	Function- ing	Function- ing	Function- ing	Function- ing	Subclock operation	Subclock operation	Subclock operation	Halted (retained)	Halted (reset)
	WDT0	Function- ing	Function- ing	Function- ing	Function- ing	Halted (retained)	Subclock operation	Subclock operation	Halted (retained)	Halted (reset)
	TMR0, 1	Function- ing	Function- ing	Function- ing	Function- ing/halted (retained)	Halted (retained)	Subclock operation	Subclock operation	Halted (retained)	Halted (reset)
	FRT	Function-ing Function-ing Function-				Halted	Halted	Halted	Halted	Halted
	TMRX, Y		ing	ing/halted (retained)	(retained)	(retained)	(retained)	(retained)	(reset)	
	Timer connec- tion									
	IIC0									
	IIC1									
	SCI0		Function-	Function-	Halted	Halted	Halted	Halted	Halted	
	SCI1	ing	ing	ing	ing/halted (reset)	(reset)	(reset)	(reset)	(reset)	(reset)
	SCI2									
	PWM									
	PWMX									
	HIF, PS2									
	D/A									
	A/D									
	RAM	Function- ing	Function- ing	Function- ing (DTC)	Function- ing	Retained	Function- ing	Retained	Retained	Retained
	I/O	Function- ing	Function- ing	Function- ing	Function- ing	Retained	Function- ing	Retained	Retained	High impedance

Notes: "Halted (retained)" means that internal register values are retained. The internal state is "operation suspended."

In module stop mode, only modules for which a stop setting has been made are halted (reset or retained).



[&]quot;Halted (reset)" means that internal register values and internal states are initialized.



Notes: When a transition is made between modes by means of an interrupt, transition cannot be made on interrupt source generation alone. Ensure that interrupt handling is performed after accepting the interrupt request.

From any state except hardware standby mode, a transition to the reset state occurs whenever $\overline{\text{RES}}$ goes low.

From any state, a transition to hardware standby mode occurs when STBY goes low. When a transition is made to watch mode or subactive mode, high-speed mode must be set.

- 1. NMI, IRQ0 to IRQ2, IRQ6, IRQ7, and WDT1 interrupts
- 2. NMI, IRQ0 to IRQ7, and WDT0 interrupts, WDT1 interrupt, TMR0 interrupt, TMR1 interrupt
- 3. All interrupts
- 4. NMI, IRQ0 to IRQ2, IRQ6, IRQ7

Figure 25.1 Mode Transitions

Table 25.2 Power-Down Mode Transition Conditions

State before	а		Bit State of Transit		State after Transition	State after Return	
Transition	SSBY	PSS	LSON	DTON	by SLEEP Instruction	by Interrupt	
High-speed/ medium-speed	0	*	0	*	Sleep	High-speed/ medium-speed	
	0	*	1	*	_	_	
	1	0	0	*	Software standby	High-speed/ medium-speed	
	1	0	1	*	_	_	
	1	1	0	0	Watch	High-speed	
	1	1	1	0	Watch	Subactive	
	1	1	0	1	_	_	
	1	1	1	1	Subactive	_	
Subactive	0	0	*	*	_	_	
	0	1	0	*	_	_	
	0	1	1	*	Subsleep	Subactive	
	1	0	*	*	_	_	
	1	1	0	0	Watch	High-speed	
	1	1	1	0	Watch	Subactive	
	1	1	0	1	High-speed	_	
	1	1	1	1	_	_	

Legend:

*: Don't care

-: Do not set.

25.1.1 Register Configuration

The power-down state is controlled by the SBYCR, LPWRCR, TCSR (WDT1), and MSTPCR registers. Table 25.3 summarizes these registers.

Table 25.3 Power-Down State Registers

Name	Abbreviation	R/W	Initial Value	Address*1
Standby control register	SBYCR	R/W	H'00	H'FF84*2
Low-power control register	LPWRCR	R/W	H'00	H'FF85*2
Timer control/status register (WDT1)	TCSR	R/W	H'00	H'FFEA
Module stop control register	MSTPCRH	R/W	H'3F	H'FF86*2
	MSTPCRL	R/W	H'FF	H'FF87*2

Notes: 1. Lower 16 bits of the address.

Some power down state registers are assigned to the same address as other registers. In this case, register selection is performed by the FLSHE bit in the serial timer control register (STCR).

25.2 Register Descriptions

25.2.1 Standby Control Register (SBYCR)

Bit	7	6	5	4	3	2	1	0
	SSBY	STS2	STS1	STS0	_	SCK2	SCK1	SCK0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	R/W	_	R/W	R/W	R/W

SBYCR is an 8-bit readable/writable register that performs power-down mode control.

SBYCR is initialized to H'00 by a reset and in hardware standby mode. It is not initialized in software standby mode.

Bit 7—Software Standby (SSBY): Determines the operating mode, in combination with other control bits, when a power-down mode transition is made by executing a SLEEP instruction. The SSBY setting is not changed by a mode transition due to an interrupt, etc.

Bit 7

SSBY	Description					
0	Transition to sleep mode after execution of SLEEP instruction in high-speed mode or medium-speed mode (Initial value)					
	Transition to subsleep mode after execution of SLEEP instruction in subactive mode					
1	Transition to software standby mode, subactive mode, or watch mode after execution of SLEEP instruction in high-speed mode or medium-speed mode					
	Transition to watch mode or high-speed mode after execution of SLEEP instruction in subactive mode					

Bits 6 to 4—Standby Timer Select 2 to 0 (STS2 to STS0): These bits select the time the MCU waits for the clock to stabilize when software standby mode, watch mode, or subactive mode is cleared and a transition is made to high-speed mode or medium-speed mode by means of a specific interrupt or instruction. With crystal oscillation, refer to table 25.4 and make a selection according to the operating frequency so that the standby time is at least 8 ms (the oscillation settling time). With an external clock, any selection can be made.

Bit 6	Bit 5	Bit 4		
STS2	STS1	STS0	Description	
0	0	0	Standby time = 8192 states	(Initial value)
		1	Standby time = 16384 states	
	1	0	Standby time = 32768 states	
		1	Standby time = 65536 states	
1	0	0	Standby time = 131072 states	
		1	Standby time = 262144 states	
	1	0	Reserved	
		1	Standby time = 16 states*	

Note: * This setting must not be used in the flash memory version.

Bit 3—Reserved: This bit cannot be modified and is always read as 0.

Bits 2 to 0—System Clock Select (SCK2 to SCK0): These bits select the clock for the bus master in high-speed mode and medium-speed mode. When operating the device after a transition to subactive mode or watch mode, bits SCK2 to SCK0 should all be cleared to 0.

Bit 2	Bit 1	Bit 0		
SCK2	SCK1	SCK0	Description	
0	0	0	Bus master is in high-speed mode	(Initial value)
		1	Medium-speed clock is φ/2	
	1	0	Medium-speed clock is φ/4	
		1	Medium-speed clock is φ/8	
1	0	0	Medium-speed clock is φ/16	
		1	Medium-speed clock is φ/32	
	1		_	

25.2.2 **Low-Power Control Register (LPWRCR)**

Bit	7	6	5	4	3	2	1	0	
	DTON	LSON	NESEL	EXCLE	_	_	_	_	
Initial value	0	0	0	0	0	0	0	0	_
Read/Write	R/W	R/W	R/W	R/W	_	_	_		

LPWRCR is an 8-bit readable/writable register that performs power-down mode control.

LPWRCR is initialized to H'00 by a reset and in hardware standby mode. It is not initialized in software standby mode.

Bit 7—Direct-Transfer On Flag (DTON): Specifies whether a direct transition is made between high-speed mode, medium-speed mode, and subactive mode when making a power-down transition by executing a SLEEP instruction. The operating mode to which the transition is made after SLEEP instruction execution is determined by a combination of other control bits.

be set.

Bit 7	
DTON	Description
0	When a SLEEP instruction is executed in high-speed mode or medium-speed mode, a transition is made to sleep mode, software standby mode, or watch mode*
	When a SLEEP instruction is executed in subactive mode, a transition is made to subsleep mode or watch mode (Initial value)
1	When a SLEEP instruction is executed in high-speed mode or medium-speed mode, a transition is made directly to subactive mode*, or a transition is made to sleep mode or software standby mode
	When a SLEEP instruction is executed in subactive mode, a transition is made directly to high-speed mode, or a transition is made to subsleep mode
Note:	* When a transition is made to watch mode or subactive mode, high-speed mode must

Bit 6—Low-Speed On Flag (LSON): Determines the operating mode in combination with other control bits when making a power-down transition by executing a SLEEP instruction. Also controls whether a transition is made to high-speed mode or to subactive mode when watch mode is cleared.

Bit 6								
LSON	Description							
0	When a SLEEP instruction is executed in high-speed mode or medium-speed mode a transition is made to sleep mode, software standby mode, or watch mode*							
	When a SLEEP instruction is executed in subactive mode, a transition is made to watch mode, or directly to high-speed mode							
	After watch mode is cleared, a transition is made to high-speed mode (Initial value)							
1	When a SLEEP instruction is executed in high-speed mode a transition is made to watch mode or subactive mode*							
	When a SLEEP instruction is executed in subactive mode, a transition is made to subsleep mode or watch mode							
	After watch mode is cleared, a transition is made to subactive mode							
Note: *	When a transition is made to watch mode or subactive mode, high-speed mode must be set.							

Bit 5—Noise Elimination Sampling Frequency Select (NESEL): Selects the frequency at which the subclock (ϕ SUB) input from the EXCL pin is sampled with the clock (ϕ) generated by the system clock oscillator. When $\phi = 5$ MHz or higher, clear this bit to 0.



Bit 5

NESEL	Description	
0	Sampling at φ divided by 32	(Initial value)
1	Sampling at φ divided by 4	

Bit 4—Subclock Input Enable (EXCLE): Controls subclock input from the EXCL pin.

Bit 4

EXCLE	Description	
0	Subclock input from EXCL pin is disabled	(Initial value)
1	Subclock input from EXCL pin is enabled	

Bits 3 to 0—Reserved: These bits cannot be modified and are always read as 0.

25.2.3 Timer Control/Status Register (TCSR)

TCSR1

Bit	7	6	5	4	3	2	1	0
	OVF	WT/IT	TME	PSS	RST/NMI	CKS2	CKS1	CKS0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/(W)*	R/W	R/W	R/W	R/W	R/W	R/W	R/W

Note: * Only 0 can be written in bit 7, to clear the flag.

TCSR1 is an 8-bit readable/writable register that performs selection of the WDT1 TCNT input clock, mode, etc.

Only bit 4 is described here. For details of the other bits, see section 14.2.2, Timer Control/Status Register (TCSR).

TCSR is initialized to H'00 by a reset and in hardware standby mode. It is not initialized in software standby mode.

Bit 4—Prescaler Select (PSS): Selects the WDT1 TCNT input clock.

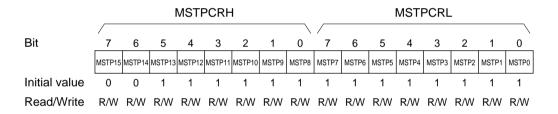
This bit also controls the operation in a power-down mode transition. The operating mode to which a transition is made after execution of a SLEEP instruction is determined in combination with other control bits.

For details, see the description of Clock Select 2 to 0 in section 14.2.2, Timer Control/Status Register (TCSR).

Bit 4

PSS	Description
0	TCNT counts φ-based prescaler (PSM) divided clock pulses
	When a SLEEP instruction is executed in high-speed mode or medium-speed mode, a transition is made to sleep mode or software standby mode (Initial value)
1	TCNT counts
	When a SLEEP instruction is executed in high-speed mode or medium-speed mode, a transition is made to sleep mode, watch mode*, or subactive mode*
	When a SLEEP instruction is executed in subactive mode, a transition is made to subsleep mode, watch mode, or high-speed mode
Note:	* When a transition is made to watch mode or subactive mode, high-speed mode must be set.

25.2.4 Module Stop Control Register (MSTPCR)



MSTPCR comprises two 8-bit readable/writable registers that perform module stop mode control.

MSTPCR is initialized to H'3FFF by a reset and in hardware standby mode. It is not initialized in software standby mode.

MSTRCRH and MSTPCRL Bits 7 to 0—Module Stop (MSTP 15 to MSTP 0): These bits specify module stop mode. See table 25.4 for the method of selecting on-chip supporting modules.

MSTPCRH, MSTPCRL Bits 7 to 0

MSTP15 to MSTP0	Description	
0	Module stop mode is cleared	(Initial value of MSTP15, MSTP14)
1	Module stop mode is set	(Initial value of MSTP13 to MSTP0)

25.3 Medium-Speed Mode

When the SCK2 to SCK0 bits in SBYCR are set to 1 in high-speed mode, the operating mode changes to medium-speed mode at the end of the bus cycle. In medium-speed mode, the CPU operates on the operating clock (ϕ /2, ϕ /4, ϕ /8, ϕ /16, or ϕ /32) specified by the SCK2 to SCK0 bits. The bus master other than the CPU (the DTC) also operates in medium-speed mode. On-chip supporting modules other than the bus masters always operate on the high-speed clock (ϕ).

In medium-speed mode, a bus access is executed in the specified number of states with respect to the bus master operating clock. For example, if $\phi/4$ is selected as the operating clock, on-chip memory is accessed in 4 states, and internal I/O registers in 8 states.

Medium-speed mode is cleared by clearing all of bits SCK2 to SCK0 to 0. A transition is made to high-speed mode and medium-speed mode is cleared at the end of the current bus cycle.

If a SLEEP instruction is executed when the SSBY bit in SBYCR and the LSON bit in LPWRCR are cleared to 0, a transition is made to sleep mode. When sleep mode is cleared by an interrupt, medium-speed mode is restored.

If a SLEEP instruction is executed when the SSBY bit in SBYCR is set to 1, and the LSON bit in LPWRCR and the PSS bit in TCSR (WDT1) are both cleared to 0, a transition is made to software standby mode. When software standby mode is cleared by an external interrupt, medium-speed mode is restored.

When the \overline{RES} pin is driven low, a transition is made to the reset state, and medium-speed mode is cleared. The same applies in the case of a reset caused by overflow of the watchdog timer.

When the \overline{STBY} pin is driven low, a transition is made to hardware standby mode.

Figure 25.2 shows the timing for transition to and clearance of medium-speed mode.

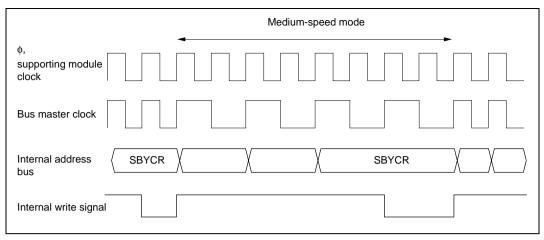


Figure 25.2 Medium-Speed Mode Transition and Clearance Timing

25.4 Sleep Mode

25.4.1 Sleep Mode

If a SLEEP instruction is executed when the SSBY bit in SBYCR and the LSON bit in LPWRCR are both cleared to 0, the CPU enters sleep mode. In sleep mode, CPU operation stops but the contents of the CPU's internal registers are retained. Other supporting modules do not stop.

25.4.2 **Clearing Sleep Mode**

Sleep mode is cleared by any interrupt, or with the \overline{RES} pin or \overline{STBY} pin.

Clearing with an Interrupt: When an interrupt request signal is input, sleep mode is cleared and interrupt exception handling is started. Sleep mode will not be cleared if interrupts are disabled, or if interrupts other than NMI have been masked by the CPU.

Clearing with the RES Pin: When the RES pin is driven low, the reset state is entered. When the RES pin is driven high after the prescribed reset input period, the CPU begins reset exception handling.

Clearing with the STBY Pin: When the STBY pin is driven low, a transition is made to hardware standby mode.

25.5 Module Stop Mode

25.5.1 Module Stop Mode

Module stop mode can be set for individual on-chip supporting modules.

When the corresponding MSTP bit in MSTPCR is set to 1, module operation stops at the end of the bus cycle and a transition is made to module stop mode. The CPU continues operating independently.

Table 25.4 shows MSTP bits and the corresponding on-chip supporting modules.

When the corresponding MSTP bit is cleared to 0, module stop mode is cleared and the module starts operating again at the end of the bus cycle. In module stop mode, the internal states of modules other than the SCI, A/D converter, 8-bit PWM module, and 14-bit PWM module, are retained.

After reset release, all modules other than the DTC are in module stop mode.

When an on-chip supporting module is in module stop mode, read/write access to its registers is disabled.

Table 25.4 MSTP Bits and Corresponding On-Chip Supporting Modules

Register	Bit	Module
MSTPCRH	MSTP15	_
	MSTP14*	Data transfer controller (DTC)
	MSTP13	16-bit free-running timer (FRT)
	MSTP12	8-bit timer (TMR0, TMR1)
	MSTP11	8-bit PWM timer (PWM), 14-bit PWM timer (PWMX)
	MSTP10	D/A converter
	MSTP9	A/D converter
	MSTP8	8-bit timers (TMRX, TMRY), timer connection
MSTPCRL	MSTP7	Serial communication interface 0 (SCI0)
	MSTP6	Serial communication interface 1 (SCI1)
	MSTP5	Serial communication interface 2 (SCI2)
	MSTP4*	I ² C bus interface (IIC) channel 0 (option)
	MSTP3*	I ² C bus interface (IIC) channel 1 (option)
	MSTP2	Host interface (HIF), keyboard matrix interrupt mask register (KMIMR, KMIMRA), port 6 MOS pull-up control register (KMPCR), keyboard buffer controller (PS2)
	MSTP1*	_
	MSTP0*	_

Notes: Do not set bit 15 to 1. Bits 1 and 0 can be read or written to, but do not affect operation.

25.5.2 **Usage Note**

If there is conflict between DTC module stop mode setting and a DTC bus request, the bus request has priority and the MSTP bit will not be set to 1.

Write 1 to the MSTP bit again after the DTC bus cycle.

When using the H8S/2144 Group and H8S/2147N, the MSTP bits for nonexistent modules must be set to 1.

Must be set to 1 in the H8S/2144 Group and H8S/2147N.

25.6 Software Standby Mode

25.6.1 Software Standby Mode

If a SLEEP instruction is executed when the SSBY bit in SBYCR is set to 1, the LSON bit in LPWRCR is cleared to 0, and the PSS bit in TCSR (WDT1) is cleared to 0, software standby mode is entered. In this mode, the CPU, on-chip supporting modules, and oscillator all stop. However, the contents of the CPU's internal registers, RAM data, and the states of on-chip supporting modules other than the SCI, PWM, and PWMX, and of the I/O ports, are retained.

In this mode the oscillator stops, and therefore power dissipation is significantly reduced.

25.6.2 Clearing Software Standby Mode

Software standby mode is cleared by an external interrupt (NMI pin, or pin $\overline{IRQ0}$, $\overline{IRQ1}$, $\overline{IRQ2}$, $\overline{IRQ6}$, or $\overline{IRQ7}$), or by means of the \overline{RES} pin or \overline{STBY} pin.

Clearing with an Interrupt: When an NMI, IRQ0, IRQ1, IRQ2, IRQ6, or IRQ7 interrupt request signal is input, clock oscillation starts, and after the elapse of the time set in bits STS2 to STS0 in SYSCR, stable clocks are supplied to the entire chip, software standby mode is cleared, and interrupt exception handling is started.

Software standby mode cannot be cleared with an IRQ0, IRQ1, IRQ2, IRQ6, or IRQ7 interrupt if the corresponding enable bit has been cleared to 0 or has been masked by the CPU.

Clearing with the \overline{RES} Pin: When the \overline{RES} pin is driven low, clock oscillation is started. At the same time as clock oscillation starts, clocks are supplied to the entire chip. Note that the \overline{RES} pin must be held low until clock oscillation stabilizes. When the \overline{RES} pin goes high, the CPU begins reset exception handling.

Clearing with the \overline{STBY} Pin: When the \overline{STBY} pin is driven low, a transition is made to hardware standby mode.

25.6.3 Setting Oscillation Settling Time after Clearing Software Standby Mode

Bits STS2 to STS0 in SBYCR should be set as described below.

Using a Crystal Oscillator

Set bits STS2 to STS0 so that the standby time is at least 8 ms (the oscillation settling time).

Table 25.5 shows the standby times for different operating frequencies and settings of bits STS2 to STS0.

Table 25.5 Oscillation Settling Time Settings

STS2	STS1	STS0	Standby Time	20 MHz	16 MHz	12 MHz	10 MHz	8 MHz	6 MHz	4 MHz	2 MHz	Unit
0	0	0	8192 states	0.41	0.51	0.65	8.0	1.0	1.3	2.0	4.1	ms
		1	16384 states	0.82	1.0	1.3	1.6	2.0	2.7	4.1	8.2	_
	1	0	32768 states	1.6	2.0	2.7	3.3	4.1	5.5	8.2	16.4	
		1	65536 states	3.3	4.1	5.5	6.6	8.2	10.9	16.4	32.8	
1	0	0	131072 states	6.6	8.2	10.9	13.1	16.4	21.8	32.8	65.5	_
		1	262144 states	13.1	16.4	21.8	26.2	32.8	43.6	65.6	131.2	
	1	0	Reserved	_	_	_	_	_	_	_	_	μs
		1	16 states*	8.0	1.0	1.3	1.6	2.0	2.7	4.0	8.0	_

: Recommended time setting

Note: * This setting must not be used in the flash memory version.

Using an External Clock

Any value can be set. Normally, use of the minimum time is recommended.

25.6.4 Software Standby Mode Application Example

Figure 25.3 shows an example in which a transition is made to software standby mode at the falling edge on the NMI pin, and software standby mode is cleared at the rising edge on the NMI pin.

In this example, an NMI interrupt is accepted with the NMIEG bit in SYSCR cleared to 0 (falling edge specification), then the NMIEG bit is set to 1 (rising edge specification), the SSBY bit is set to 1, and a SLEEP instruction is executed, causing a transition to software standby mode.

Software standby mode is then cleared at the rising edge on the NMI pin.

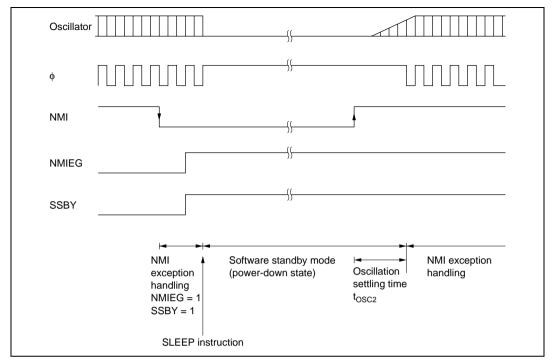


Figure 25.3 Software Standby Mode Application Example

25.6.5 Usage Note

In software standby mode, I/O port states are retained. Therefore, there is no reduction in current dissipation for the output current when a high-level signal is output.

Current dissipation increases while waiting for oscillation to settle.

25.7 Hardware Standby Mode

25.7.1 Hardware Standby Mode

When the STBY pin is driven low, a transition is made to hardware standby mode from any mode.

In hardware standby mode, all functions enter the reset state and stop operation, resulting in a significant reduction in power dissipation. As long as the prescribed voltage is supplied, on-chip RAM data is retained. I/O ports are set to the high-impedance state.

In order to retain on-chip RAM data, the RAME bit in SYSCR should be cleared to 0 before driving the \overline{STBY} pin low.

Do not change the state of the mode pins (MD1 and MD0) while the chip is in hardware standby mode.

Hardware standby mode is cleared by means of the \overline{STBY} pin and the \overline{RES} pin. When the \overline{STBY} pin is driven high while the \overline{RES} pin is low, the reset state is set and clock oscillation is started. Ensure that the \overline{RES} pin is held low until the clock oscillation settles (at least 8 ms—the oscillation settling time—when using a crystal oscillator). When the \overline{RES} pin is subsequently driven high, a transition is made to the program execution state via the reset exception handling state.

25.7.2 Hardware Standby Mode Timing

Figure 25.4 shows an example of hardware standby mode timing.

When the \overline{STBY} pin is driven low after the \overline{RES} pin has been driven low, a transition is made to hardware standby mode. Hardware standby mode is cleared by driving the \overline{STBY} pin high, waiting for the oscillation settling time, then changing the \overline{RES} pin from low to high.

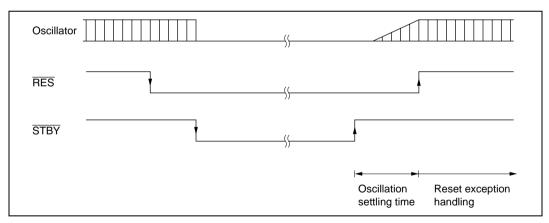


Figure 25.4 Hardware Standby Mode Timing

25.8 Watch Mode

25.8.1 Watch Mode

If a SLEEP instruction is executed in high-speed mode or subactive mode when the SSBY in SBYCR is set to 1, the DTON bit in LPWRCR is cleared to 0, and the PSS bit in TCSR (WDT1) is set to 1, the CPU makes a transition to watch mode.

In this mode, the CPU and all on-chip supporting modules except WDT1 stop. As long as the prescribed voltage is supplied, the contents of some of the CPU's internal registers and on-chip RAM are retained, and I/O ports retain their states prior to the transition.

25.8.2 Clearing Watch Mode

Watch mode is cleared by an interrupt (WOVI1 interrupt, NMI pin, or pin $\overline{IRQ0}$, $\overline{IRQ1}$, $\overline{IRQ2}$, $\overline{IRQ6}$, or $\overline{IRQ7}$), or by means of the \overline{RES} pin or \overline{STBY} pin.

Clearing with an Interrupt: When an interrupt request signal is input, watch mode is cleared and a transition is made to high-speed mode or medium-speed mode if the LSON bit in LPWRCR is cleared to 0, or to subactive mode if the LSON bit is set to 1. When making a transition to high-speed mode, after the elapse of the time set in bits STS2 to STS0 in SBYCR, stable clocks are supplied to the entire chip, and interrupt exception handling is started.

Watch mode cannot be cleared with an IRQ0, IRQ1, IRQ2, IRQ6, or IRQ7 interrupt if the corresponding enable bit has been cleared to 0, or with an on-chip supporting module interrupt if acceptance of the relevant interrupt has been disabled by the interrupt enable register or masked by the CPU.

See section 25.6.3, Setting Oscillation Settling Time after Clearing Software Standby Mode, for the oscillation settling time setting when making a transition from watch mode to high-speed mode.

Clearing with the \overline{RES} Pin: See "Clearing with the \overline{RES} Pin" in section 25.6.2, Clearing Software Standby Mode.

Clearing with the \overline{STBY} Pin: When the \overline{STBY} pin is driven low, a transition is made to hardware standby mode.

25.9 Subsleep Mode

25.9.1 Subsleep Mode

If a SLEEP instruction is executed in subactive mode when the SSBY in SBYCR is cleared to 0, the LSON bit in LPWRCR is set to 1, and the PSS bit in TCSR (WDT1) is set to 1, the CPU makes a transition to subsleep mode.

In this mode, the CPU and all on-chip supporting modules except TMR0, TMR1, WDT0, and WDT1 stop. As long as the prescribed voltage is supplied, the contents of some of the CPU's internal registers and on-chip RAM are retained, and I/O ports retain their states prior to the transition.

25.9.2 Clearing Subsleep Mode

Subsleep mode is cleared by an interrupt (on-chip supporting module interrupt, NMI pin, or pin $\overline{IRQ0}$ to $\overline{IRQ7}$), or by means of the \overline{RES} pin or \overline{STBY} pin.

Clearing with an Interrupt: When an interrupt request signal is input, subsleep mode is cleared and interrupt exception handling is started. Subsleep mode cannot be cleared with an IRQ0 to IRQ7 interrupt if the corresponding enable bit has been cleared to 0, or with an on-chip supporting module interrupt if acceptance of the relevant interrupt has been disabled by the interrupt enable register or masked by the CPU.

Clearing with the \overline{RES} Pin: See "Clearing with the \overline{RES} Pin" in section 25.6.2, Clearing Software Standby Mode.

Clearing with the \overline{STBY} Pin: When the \overline{STBY} pin is driven low, a transition is made to hardware standby mode



25.10 Subactive Mode

25.10.1 Subactive Mode

If a SLEEP instruction is executed in high-speed mode when the SSBY bit in SBYCR, the DTON bit in LPWRCR, and the PSS bit in TCSR (WDT1) are all set to 1, the CPU makes a transition to subactive mode. When an interrupt is generated in watch mode, if the LSON bit in LPWRCR is set to 1, a transition is made to subactive mode. When an interrupt is generated in subsleep mode, a transition is made to subactive mode.

In subactive mode, the CPU performs sequential program execution at low speed on the subclock. In this mode, all on-chip supporting modules except TMR0, TMR1, WDT0, and WDT1 stop.

When operating the device in subactive mode, bits SCK2 to SCK0 in SBYCR must all be cleared to 0.

25.10.2 Clearing Subactive Mode

Subsleep mode is cleared by a SLEEP instruction, or by means of the RES pin or STBY pin.

Clearing with a SLEEP Instruction: When a SLEEP instruction is executed while the SSBY bit in SBYCR is set to 1, the DTON bit in LPWRCR is cleared to 0, and the PSS bit in TCSR (WDT1) is set to 1, subactive mode is cleared and a transition is made to watch mode. When a SLEEP instruction is executed while the SSBY bit in SBYCR is cleared to 0, the LSON bit in LPWRCR is set to 1, and the PSS bit in TCSR (WDT1) is set to 1, a transition is made to subsleep mode. When a SLEEP instruction is executed while the SSBY bit in SBYCR is set to 1, the DTON bit is set to 1 and the LSON bit is cleared to 0 in LPWRCR, and the PSS bit in TCSR (WDT1) is set to 1, a transition is made directly to high-speed mode.

Fort details of direct transition, see section 25.11, Direct Transition.

Clearing with the \overline{RES} Pin: See "Clearing with the \overline{RES} Pin" in section 25.6.2, Clearing Software Standby Mode.

Clearing with the \overline{STBY} Pin: When the \overline{STBY} pin is driven low, a transition is made to hardware standby mode

25.11 Direct Transition

25.11.1 Overview of Direct Transition

There are three operating modes in which the CPU executes programs: high-speed mode, medium-speed mode, and subactive mode. A transition between high-speed mode and subactive mode without halting the program is called a direct transition. A direct transition can be carried out by setting the DTON bit in LPWRCR to 1 and executing a SLEEP instruction. After the transition, direct transition interrupt exception handling is started.

Direct Transition from High-Speed Mode to Subactive Mode: If a SLEEP instruction is executed in high-speed mode while the SSBY bit in SBYCR, the LSON bit and DTON bit in LPWRCR, and the PSS bit in TSCR (WDT1) are all set to 1, a transition is made to subactive mode.

Direct Transition from Subactive Mode to High-Speed Mode: If a SLEEP instruction is executed in subactive mode while the SSBY bit in SBYCR is set to 1, the LSON bit is cleared to 0 and the DTON bit is set to 1 in LPWRCR, and the PSS bit in TSCR (WDT1) is set to 1, after the elapse of the time set in bits STS2 to STS0 in SBYCR, a transition is made to directly to high-speed mode.

25.12 Usage Notes

- 1. When making a transition to subactive mode or watch mode, set the DTC to enter module stop mode (write 1 to the relevant bits in MSTPCR), and then read the relevant bits to confirm that they are set to 1 before mode transition. Do not clear module stop mode (write 0 to the relevant bits in MSTPCR) until a transition from subactive mode to high-speed mode or medium-speed mode has been performed.
 - If a DTC activation source occurs in sub-active mode, the DTC will be activated only after module stop mode has been cleared and high-speed mode or medium-speed mode has been entered.
- 2. The on-chip peripheral modules (DTC and TPU) which halt operation in subactive mode cannot clear an interrupt in subactive mode. Therefore, if a transition is made to sub-active mode while an interrupt is requested, the CPU interrupt source cannot be cleared. Disable the interrupts of each on-chip peripheral module before executing a SLEEP instruction to enter subactive mode or watch mode.



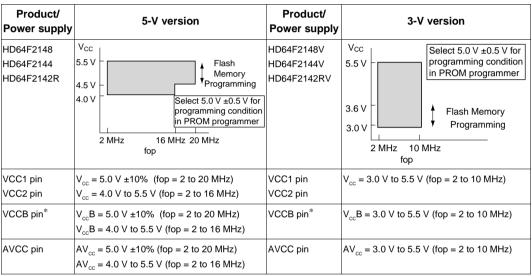
Section 26 Electrical Characteristics

26.1 Voltage of Power Supply and Operating Range

The power supply voltage and operating range (shaded part) for each product are shown in table 26.1.

Table 26.1 Power Supply Voltage and Operating Range (1)

(F-ZTAT Products)



Note: * Available only in the H8S/2148 Group.

Table 26.1 Power Supply Voltage and Operating Range (2)

(F-ZTAT Products)

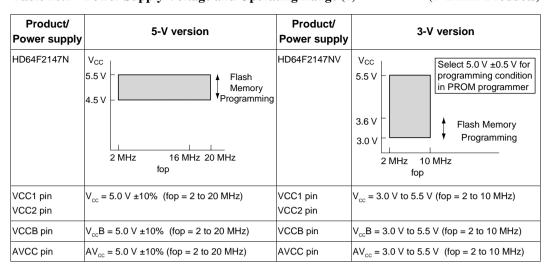


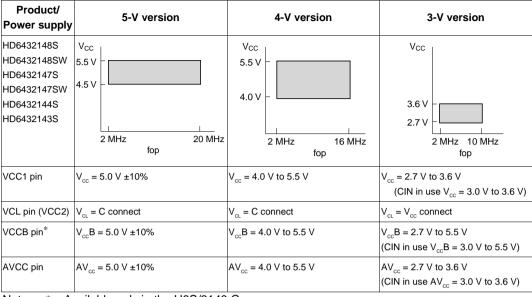
Table 26.1 Power Supply Voltage and Operating Range (3) (F-ZTAT A-Mask Products)

Product/ Power supply	5-V version	Product/ Power supply	3-V version
HD64F2148A HD64F2147A HD64F2144A	Vcc 5.5 V 4.5 V 4.0 V Select 3.3 V ±0.3 V for programming condition in PROM programmer 2 MHz 16 MHz 20 MHz fop	HD64F2148AV HD64F2147AV HD64F2144AV	Select 3.3 V ±0.3 V for programming condition in PROM programmer 3.6 V 3.0 V 2.7 V 4 Flash Memory Programming 2 MHz 10 MHz
VCC1 pin	$V_{cc} = 5.0 \text{ V} \pm 10\% \text{ (fop = 2 to 20 MHz)}$ $V_{cc} = 4.0 \text{ V to } 5.5 \text{ V (fop = 2 to 16 MHz)}$	VCC1 pin	$V_{cc} = 2.7 \text{ V to } 3.6 \text{ V (fop = 2 to 10 MHz)}$ (CIN in use $V_{cc} = 3.0 \text{ V to } 3.6 \text{ V)}$
VCL pin (VCC2)	V _{cL} = C connect	VCL pin (VCC2)	V _{cL} = V _{cc} connect
VCCB pin*	$V_{cc}B = 5.0 \text{ V} \pm 10\% \text{ (fop = 2 to 20 MHz)}$ $V_{cc}B = 4.0 \text{ V to 5.5 V (fop = 2 to 16 MHz)}$	VCCB pin*	$V_{cc}B = 2.7 \text{ V to } 5.5 \text{ V (fop = 2 to 10 MHz)}$ (CIN in use $V_{cc}B = 3.0 \text{ V to } 5.5 \text{ V)}$
AVCC pin	$AV_{cc} = 5.0 \text{ V} \pm 10\% \text{ (fop = 2 to 20 MHz)}$ $AV_{cc} = 4.0 \text{ V to 5.5 V (fop = 2 to 16 MHz)}$	AVCC pin	$AV_{cc} = 2.7 \text{ V to } 3.6 \text{ V (fop = 2 to 10 MHz)}$ (CIN in use $AV_{cc} = 3.0 \text{ V to } 3.6 \text{ V)}$

Note: * Available only in the H8S/2148 Group.

Table 26.1 Power Supply Voltage and Operating Range (4)

(Mask ROM Products)



Note: * Available only in the H8S/2148 Group.

Table 26.1 Power Supply Voltage and Operating Range (5)

(Mask ROM Products)

Product/ Power supply	5-V version	4-V version	3-V version		
	Vcc 5.5 V 4.5 V 2 MHz 20 MHz	Vcc 5.5 V 4.0 V 2 MHz 16 MHz fop	2.7 V 2 MHz 10 MHz fop		
VCC1 pin VCC2 pin	$V_{cc} = 5.0 \text{ V} \pm 10\%$	$V_{cc} = 4.0 \text{ V to } 5.5 \text{ V}$	$V_{cc} = 2.7 \text{ V to } 5.5 \text{ V}$		
AVCC pin	AV _{cc} = 5.0 V ±10%	AV _{cc} = 4.0 V to 5.5 V	AV _{cc} = 2.7 V to 5.5 V		

26.2 Electrical Characteristics of H8S/2148 F-ZTAT

26.2.1 Absolute Maximum Ratings

Table 26.2 lists the absolute maximum ratings.

Table 26.2 Absolute Maximum Ratings

Item	Symbol	Value	Unit
Power supply voltage*	V _{cc}	-0.3 to +7.0	V
Input/output buffer power supply (power supply for the port A)	V _{cc} B	-0.3 to +7.0	V
Input voltage (except ports 6, 7, and A)	V _{in}	-0.3 to V_{cc} +0.3	V
Input voltage (CIN input not selected for port 6)	V _{in}	-0.3 to V_{cc} +0.3	V
Input voltage (CIN input not selected for port A)	V _{in}	–0.3 to V _{cc} B +0.3	V
Input voltage (CIN input selected for port 6)	V _{in}	–0.3 V to lower of voltages $\rm V_{cc}$ +0.3 and $\rm AV_{cc}$ +0.3	V
Input voltage (CIN input selected for port A)	V _{in}	–0.3 V to lower of voltages $V_{\rm cc}B$ +0.3 and $AV_{\rm cc}$ +0.3	V
Input voltage (port 7)	V _{in}	-0.3 to AV _{cc} +0.3	V
Reference supply voltage	AV _{ref}	-0.3 to AV _{cc} +0.3	V
Analog power supply voltage	AV _{cc}	-0.3 to +7.0	V
Analog input voltage	V _{AN}	-0.3 to AV _{cc} +0.3	V
Operating temperature	T _{opr}	Regular specifications: -20 to +75	°C
		Wide-range specifications: -40 to +85	°C
Operating temperature (flash	T _{opr}	Regular specifications: 0 to +75	°C
memory programming/erasing)		Wide-range specifications: 0 to +85	°C
Storage temperature	T_{stg}	-55 to +125	°C

Caution: Permanent damage to the chip may result if absolute maximum ratings are exceeded.

Note: * Power supply voltage for VCC1 and VCC2 pins.



26.2.2 DC Characteristics

Table 26.3 lists the DC characteristics. Permitted output current values and bus drive characteristics are shown in tables 26.4 and 26.5, respectively.

Table 26.3 DC Characteristics (1)

$$\begin{array}{ll} \text{Conditions:} & V_{_{CC}} = 5.0 \text{ V} \pm 10\%, V_{_{CC}} B = 5.0 \text{ V} \pm 10\%, A V_{_{CC}}^{\quad \ *^1} = 5.0 \text{ V} \pm 10\%, \\ & A V_{_{ref}}^{\quad \ *^1} = 4.5 \text{ V} \text{ to } A V_{_{CC}}, V_{_{SS}} = A V_{_{SS}}^{\quad \ *^1} = 0 \text{ V}, T_{_a} = -20 \text{ to } +75^{\circ} \text{C}^{\ast 11} \text{ (regular specifications)}, \\ & T_{_a} = -40 \text{ to } +85^{\circ} \text{C}^{\ast 11} \text{ (wide-range specifications)} \end{array}$$

Item			Symbol	Min	Тур	Max	Unit	Test Conditions
• • •	P67 to P60(KWUL =	(1)	V _T -	1.0	_	_	V	
voltage	00)*2*6, KIN15 to KIN8*7*8, IRQ2 to IRQ0*3, IRQ5 to IRQ3		$V_{T}^{\; \star}$	_	_	$V_{cc} \times 0.7$ $V_{cc} B \times 0.7$		
			$V_T^+ - V_T^-$	0.4	_	_		
Schmitt trigger input voltage (in level switching)*6	P67 to P60		V _T	$V_{cc} \times 0.3$		_	V	
	(KWUL = 01)		V _T ⁺	_	_	$V_{cc} \times 0.7$	=	
			$V_T^+ - V_T^-$	$V_{\rm cc} \times 0.05$	_	_	=	
	P67 to P60 (KWUL = 10)		V _T	$V_{cc} \times 0.4$		_	=	
			V _T ⁺	_	_	$V_{cc} \times 0.8$	=	
			$V_T^+ - V_T^-$	$V_{\text{cc}} \times 0.03$	_	_	=	
	P67 to P60 (KWUL = 11)	_	V _T	$V_{cc} \times 0.45$	_	_	_	
			V _T ⁺	_	_	$V_{cc} \times 0.9$	_	
			$V_T^+ - V_T^-$	0.05	_	_	_	
Input high voltage	RES, STBY, NMI, MD1, MD0	(2)	V _{IH}	V _{cc} -0.7	_	V _{cc} +0.3	V	
	EXTAL			$V_{cc} \times 0.7$		V _{cc} +0.3	=	
	PA7 to PA0*7	=		$V_{cc}B \times 0.7$	_	V _{cc} B +0.3	=	
	Port 7	-		2.0	_	AV _{cc} +0.3	-	
	Input pins except (1) and (2) above		_	2.0	_	V _{cc} +0.3		

Item			Symbol	Min	Тур	Max	Unit	Test Conditions
Input low voltage	RES, STBY, MD1, MD0	(3)	V _{IL}	-0.3	_	0.5	V	
	PA7 to PA0	_		-0.3	_	1.0		
	NMI, EXTAL, input pins except (1) and (3) above		_	-0.3	_	0.8		
Output high	All output pins		V _{OH}	V _{cc} -0.5	_	_	V	$I_{OH} = -200 \mu A$
voltage	(except P97, and P52*4)*5 *8			V _{cc} B −0.5				
	P32)			3.5	_	_	V	$I_{OH} = -1 \text{ mA}$
	P97, P52*4		_	2.5	_	_	V	$I_{OH} = -1 \text{ mA}$
Output low voltage	All output pins (except RESO)*5		V _{oL}	_	_	0.4	V	I _{oL} = 1.6 mA
	Ports 1 to 3		_	_	_	1.0	V	I _{oL} = 10 mA
	RESO		_	_	_	0.4	V	I _{oL} = 2.6 mA
Input leakage	RES		l _{in}	_	_	10.0	μΑ	$V_{in} = 0.5 \text{ to} $ $V_{CC} = 0.5 \text{ V}$
current	STBY, NMI, MD1, MD0			_	_	1.0	μΑ	_
	Port 7			_	_	1.0	μΑ	$V_{in} = 0.5 \text{ to}$ AV _{CC} -0.5 V
Three-state leakage current (off state)	Ports 1 to 6, 8, 9, A*8, B		I _{TSI}	_	_	1.0	μA	$V_{in} = 0.5 \text{ to}$ $V_{cc} = -0.5 \text{ V},$ $V_{in} = 0.5 \text{ to}$ $V_{cc}B = -0.5 \text{ V}$
Input	Ports 1 to 3		$-I_P$	50	_	300	μΑ	$V_{in} = 0 V$
pull-up MOS current	Ports A*8, B, Port 6 (P6PUE = 0)		_	60	_	500	μA	_
	Port 6 (P6PUE = 1)			15	_	150	μΑ	

Input capacitance RES	Item		Symbol	Min	Тур	Max	Unit	Test Conditions	
Normal operation Hornard operation Horn	•	•	4) C _{in}	_	_	80	pF		
P52, P97, P42, P86 PA7 to PA2	capacitance	NMI		_	_	50	pF		
Current Normal operation I _{cc} — 85 120 mA f = 20 MHz		P42, P86		_	_	20	pF	– 1 _a – 23 O	
Sleep mode Standby mode*10 - 70 100 mA f = 20 MHz		· · · · · · · · · · · · · · · · · · ·		_	_	15	pF	_	
Standby mode*10 — 0.01 5.0 µA T _a ≤ 50°C — — 20.0 µA 50°C < T _a Analog power conversion Idle — 0.01 5.0 µA AV _{cc} = 2.0 V to 5.5 V Reference power supply current During A/D, D/A conversion Idle — 0.5 1.0 mA During A/D, D/A conversion Idle — 0.01 5.0 µA AV _{ref} = 2.0 V to AV _{cc} Analog power supply voltage*1 AV _{cc} 4.5 — 5.5 V Operating During a/D, D/D to AV _{cc} 4.5 — 5.5 V Idle/not used Analog power supply voltage AV _{cc} 4.5 — 5.5 V Idle/not used Idle/Idle/not used Idle/Idle/not used Idle/Idle/not used Idle/Idle/not used Idle/Idle/Idl			I _{cc}	_	85	120	mA	f = 20 MHz	
Analog During A/D, D/A Al _{cc} — 1.2 2.0 mA Exercise 2.0 V to 5.5 V	dissipation*9	Sleep mode		_	70	100	mA	f = 20 MHz	
Analog power supply current During A/D, D/A conversion AI _{cc} — 1.2 2.0 mA Reference power supply current During A/D conversion During A/D, D/A conversion AI _{ref} — 0.5 1.0 mA — 2.0 5.0 mA — 0.01 5.0 μA AV _{ref} = 2.0 V to AV _{cc} Analog power supply voltage*1 AV _{cc} 4.5 — 5.5 V Operating 2.0 — 5.5 V Idle/not used		Standby mode*10		_	0.01	5.0	μΑ	T _a ≤ 50°C	
				_	_	20.0	μΑ	50°C < T _a	
current Language Current <th colspan<="" td=""><td>J</td><td>•</td><td>Al_{cc}</td><td>_</td><td>1.2</td><td>2.0</td><td>mA</td><td></td></th>	<td>J</td> <td>•</td> <td>Al_{cc}</td> <td>_</td> <td>1.2</td> <td>2.0</td> <td>mA</td> <td></td>	J	•	Al _{cc}	_	1.2	2.0	mA	
		Idle		_	0.01	5.0	μΑ		
$\frac{\text{Supply current}}{\text{Idle}} = \frac{\frac{2.0 \text{ s.0}}{\text{max}}}{\frac{2.0 \text{ s.0}}{\text{max}}} = \frac{2.0 \text{ s.0}}{\text{max}}$ $\frac{\text{Conversion}}{\text{Idle}} = \frac{2.0 \text{ s.0}}{\text{max}} = \frac{2.0 \text{ v.0}}{\text{to AV}_{cc}}$ $\frac{\text{Analog power supply voltage*}^{*1}}{\text{AV}_{cc}} = \frac{\text{AV}_{cc}}{\text{4.5}} = \frac{4.5 \text{ s.5}}{\text{5.5}} = \frac{\text{V}}{\text{Operating}}$ $\frac{\text{Operating}}{2.0 \text{ s.5}} = \frac{\text{V}}{\text{Operating}}$	Reference	During A/D convers	sion Al _{ref}	_	0.5	1.0	mA		
$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$	supply	J ,		_	2.0	5.0	mA	_	
2.0 — 5.5 V Idle/not used		Idle		_	0.01	5.0	μΑ		
	Analog power	Analog power supply voltage*1		4.5	_	5.5	V	Operating	
RAM standby voltage V _{RAM} 2.0 — V				2.0	_	5.5	V	Idle/not used	
	RAM standb	y voltage	V_{RAM}	2.0	_	_	V		

Notes: 1. Do not leave the AVCC, AV_{ref}, and AVSS pins open even if the A/D converter and D/A converter are not used.

Even if the A/D converter and D/A converter are not used, apply a value in the range 2.0 V to 5.5 V to AVCC and AV_{ref} pins by connection to the power supply (V_{cc}), or some other method. Ensure that AV_{ref} \leq AV_{cc}.

- 2. P67 to P60 include supporting module inputs multiplexed on those pins.
- 3. IRQ2 includes the ADTRG signal multiplexed on that pin.
- In the H8S/2148 Group, P52/SCK0/SCL0 and P97/SDA0 are NMOS push-pull outputs. An external pull-up resistor is necessary to provide high-level output from SCL0 and SDA0 (ICE = 1).

In the H8S/2148 Group, P52/SCK0 and P97 (ICE = 0) high levels are driven by NMOS.

5. When IICS = 0, ICE = 0, and KBIOE = 0. Low-level output when the bus drive function is selected is determined separately.

- 6. The upper limit of the port 6 applied voltage is V_{cc} +0.3 V when CIN input is not selected, and the lower of V_{cc} +0.3 V and AV_{cc} +0.3 V when CIN input is selected. When a pin is in output mode, the output voltage is equivalent to the applied voltage.
- 7. The upper limit of the port A applied voltage is $V_{cc}B + 0.3 \text{ V}$ when CIN input is not selected, and the lower of $V_{cc}B + 0.3 \text{ V}$ and $AV_{cc} + 0.3 \text{ V}$ when CIN input is selected. When a pin is in output mode, the output voltage is equivalent to the applied voltage.
- 8. Port A characteristics depends on $V_{cc}B$ (or on V_{cc} when other pins are in output mode).
- 9. Current dissipation values are for V_{IH} min = V_{cc} –0.5 V, V_{cc} B –0.5 V, and V_{IL} max = 0.5 V with all output pins unloaded and the on-chip pull-up MOSs in the off state.
- 10. The values are for V_{RAM} \leq V_{CC} < 4.5V, V_{IH} min = V_{CC} \times 0.9, V_{CC}B \times 0.9, and V_{IL} max = 0.3 V.
- 11. For flash memory program/erase operations, the applicable range is $T_a = 0$ to +75°C (regular specifications) or $T_a = 0$ to +85°C (wide-range specifications).



Table 26.3 DC Characteristics (2)

Conditions: $V_{cc} = 4.0 \text{ V}$ to 5.5 V^{*11} , $V_{cc}B = 4.0 \text{ V}$ to 5.5 V, $AV_{cc}^{*1} = 4.0 \text{ V}$ to 5.5 V,

AV_{ref} *1 = 4.0 V to AV_{cc}, $V_{ss} = AV_{ss}^{*1} = 0$ V, $T_a = -20$ to $+75^{\circ}C^{*11}$ (regular specifications), $T_a = -40$ to $+85^{\circ}C^{*11}$ (wide-range specifications)

Test Symbol Item Min Unit **Conditions** Тур Max $V_{\scriptscriptstyle T}^{\scriptscriptstyle -}$ V $V_{cc} = 4.5 \text{ V to}$ Schmitt P67 to (1) 1.0 trigger input P60(KWUL = 5.5 V, V_{τ}^{\dagger} $V_{cc} \times 0.7$ 00)*2*6. $V_{cc}B = 4.5 \text{ V}$ voltage $V_{cc}B \times 0.7$ KIN15 to to 5.5 V $V_{T}^{+} - V_{T}^{-}$ 0.4 ٧ KIN8*7 *8 IRQ2 to IRQ0*3 8.0 ٧ $V_{cc} < 4.5 V$, IRQ5 to IRQ3 V_CB < 4.5 V $V_{cc} \times 0.7$ ٧ $V_{cc}B \times 0.7$ $V_{\tau}^{+} - V_{\tau}^{-}$ 0.3 V ٧ $V_{cc} = 4.0 \text{ V to}$ Schmitt P67 to P60 $V_{cc} \times 0.3$ trigger input (KWUL = 01) 5.5 V $V_{cc} \times 0.7$ voltage $V_{cc} \times 0.05$ (in level switching)*6 P67 to P60 $V_{cc} \times 0.4$ (KWUL = 10) $V_{cc} \times 0.8$ $V_{cc} \times 0.03$ $V_{cc} \times 0.45$ P67 to P60 (KWUL = 11) V_{τ}^{+} $V_{cc} \times 0.9$ $V_{\tau}^{+} - V_{\tau}^{-}$ 0.05 RES, STBY, V_{III} $V_{cc} - 0.7$ V_{cc} +0.3 V Input high (2)voltage NMI, MD1, MD0 $V_{cc} \times 0.7$ V_{cc} +0.3 **EXTAL** $V_{cc}B \times 0.7$ — $V_{cc}B + 0.3$ PA7 to PA0*7 AV_{cc} +0.3 Port 7 2.0 V_{cc} +0.3 Input pins 2.0 ٧ except (1) and (2) above

Item			Symbol	Min	Тур	Max	Unit	Test Conditions
Input low voltage	RES, STBY, MD1, MD0	(3)	V _{IL}	-0.3	_	0.5	V	
	PA7 to PA0	-		-0.3	_	1.0	V	$V_{cc}B = 4.5 \text{ V}$ to 5.5 V
				-0.3	_	0.8	V	V _{cc} B < 4.5 V
	NMI, EXTAL, input pins except (1) and (3) above		_	-0.3	_	0.8	V	
Output high voltage	(except P97, and P52*4)*5*8		V _{OH}	V _{cc} -0.5 V _{cc} B -0.5	_	_	V	$I_{OH} = -200 \mu A$
				3.5	_	_	V	$I_{OH} = -1 \text{ mA},$ $V_{CC} = 4.5 \text{ V to}$ $5.5 \text{ V},$ $V_{CC}B = 4.5 \text{ V}$ to 5.5 V
				3.0	_	_	V	$I_{OH} = -1 \text{ mA},$ $V_{CC} < 4.5 \text{ V},$ $V_{CC}B < 4.5 \text{ V}$
	P97, P52*4		_	2.0	_	_	V	I _{OH} = -1 mA
Output low voltage	All output pins (except RESO)*5		V _{oL}	_	_	0.4	V	I _{oL} = 1.6 mA
	Ports 1 to 3		_	_	_	1.0	V	I _{OL} = 10 mA
	RESO		_	_	_	0.4	V	I _{OL} = 2.6 mA
Input	RES		I _{in}	_	_	10.0	μΑ	V _{in} = 0.5 to
leakage current	STBY, NMI, MD1, MD0			_	_	1.0	μΑ	[−] V _{cc} −0.5 V
	Port 7		_	_	_	1.0	μΑ	$V_{in} = 0.5 \text{ to}$ AV _{CC} -0.5 V
Three-state leakage current (off state)	Ports 1 to 6 Ports 8, 9, A*8, B		I _{TSI}	_	_	1.0	μА	$V_{in} = 0.5 \text{ to}$ $V_{cc} - 0.5 \text{ V},$ $V_{in} = 0.5 \text{ to}$ $V_{cc}B - 0.5 \text{ V}$

Input pull-up pull-up Ports 1 to 3 Ports A**, B Port 6 Ports 1 to 3 Ports A**, B Port 6 Ports 1 to 3 Ports 1 to 3 Ports 1 to 3 Ports A**, B Port 6 Ports 1 to 3 Ports A**, B Port 6 Ports A**, B Ports A**	4.5 V to 4.5 V / / V, 5 V,
MOS current $(P6PUE = 0)$ $(P6PUE = 1)$ $Port 6$ $(P6PUE = 1)$ $Ports 1 to 3$ $Ports A^{*8}$, B $Port 6$ $Port$	4.5 V / V, 5 V,
Port 6 (P6PUE = 1) Ports 1 to 3 $0 - 200 \mu A V_{in} = 0.7$ Ports A*8, B Port 6 $0 - 400 \mu A V_{cc} \le 4 V_{cc} $	V, .5 V,
Ports A*8, B 40 — 400 μ A $V_{cc} \le 4$ V_{cc} B \le	.5 V,
Port 6 40 40 pA V _{cc} B ≤	
(1. 51. 52 - 5)	
Port 6 10 — 110 (P6PUE = 1)	
Input \overline{RES} (4) C_{in} — 80 pF $V_{in} = 0.7$	
capacitance ${NMI}$ ${T_a = 25}$	
P52, P97, P42, P86, PA7 to PA2	C
Input pins — — 15 pF except (4) above	
Current Normal operation I_{cc} — 70 100 mA $f = 16 M$	ЛHz
dissipation ^{*9} Sleep mode $-$ 60 85 mA f = 16 M	ЛHz
Standby mode ^{*10} $-$ 0.01 5.0 μ A $T_a \le 50$	$T_a \le 50^{\circ}C$
	T _a
Analog During A/D, D/A Al _{cc} — 1.2 2.0 mA power conversion	
supply Idle — 0.01 5.0 μ A AV _{cc} = to 5.5 \	
Reference During A/D conversion Al _{ref} — 0.5 1.0 mA	
power supply conversion — 2.0 5.0 mA — 2.0 conversion	
Idle	
Analog power supply voltage*1 AV _{cc} 4.0 — 5.5 V Operat	ng
2.0 — 5.5 V Idle/not	used
RAM standby voltage V _{RAM} 2.0 — V	

Notes: 1. Do not leave the AVCC, AV_{ref}, and AVSS pins open even if the A/D converter and D/A converter are not used.

Even if the A/D converter and D/A converter are not used, apply a value in the range 2.0 V to 5.5 V to AVCC and AV_{ref} pins by connection to the power supply (V_{cc}), or some other method. Ensure that AV_{ref} \leq AV_{cc}.

- 2. P67 to P60 include supporting module inputs multiplexed on those pins.
- 3. IRQ2 includes the ADTRG signal multiplexed on that pin.
- In the H8S/2148 Group, P52/SCK0/SCL0 and P97/SDA0 are NMOS push-pull outputs.
 An external pull-up resistor is necessary to provide high-level output from SCL0 and SDA0 (ICE = 1).

In the H8S/2148 Group, P52/SCK0 and P97 (ICE = 0) high levels are driven by NMOS.

- 5. When IICS = 0, ICE = 0, and KBIOE = 0. Low-level output when the bus drive function is selected is determined separately.
- 6. The upper limit of the port 6 applied voltage is V_{cc} +0.3 V when CIN input is not selected, and the lower of V_{cc} +0.3 V and AV_{cc} +0.3 V when CIN input is selected. When a pin is in output mode, the output voltage is equivalent to the applied voltage.
- 7. The upper limit of the port A applied voltage is $V_{cc}B + 0.3 \text{ V}$ when CIN input is not selected, and the lower of $V_{cc}B + 0.3 \text{ V}$ and $AV_{cc} + 0.3 \text{ V}$ when CIN input is selected. When a pin is in output mode, the output voltage is equivalent to the applied voltage.
- 8. The port A characteristics depend on $V_{cc}B$, and the other pins characteristics depend on V_{cc} .
- 9. Current dissipation values are for V_{H} min = V_{cc} -0.5 V, V_{cc} B -0.5 V, and V_{LL} max = 0.5 V with all output pins unloaded and the on-chip pull-up MOSs in the off state.
- 10. The values are for $V_{\text{RAM}} \le V_{\text{CC}} < 4.0 \text{ V}$, $V_{\text{IH}} \text{ min} = V_{\text{CC}} \times 0.9$, $V_{\text{CC}} B \times 0.9$, and $V_{\text{IL}} \text{ max} = 0.3 \text{ V}$.
- 11. For flash memory program/erase operations, the applicable ranges are V_{cc} = 4.5 V to 5.5 V and T_a = 0 to +75°C (regular specifications) or T_a = 0 to +85°C (wide-range specifications).



Table 26.3 DC Characteristics (3)

 $\begin{array}{ll} \text{Conditions:} & V_{_{CC}} = 3.0 \text{ V to } 5.5 \text{ V}^{*_{11}}, V_{_{CC}} B = 3.0 \text{ V to } 5.5 \text{ V}, \text{AV}_{_{CC}}^{\quad *_{1}} = 3.0 \text{ V to } 5.5 \text{ V}, \\ & AV_{_{ref}} = 3.0 \text{ V to } 5.5 \text{ V}, V_{_{SS}} = AV_{_{SS}}^{\quad *_{1}} = 0 \text{ V}, T_{_{a}} = -20 \text{ to } +75^{\circ}\text{C}^{*_{11}} \\ \end{array}$

Item			Symbol	Min	Тур	Max	Unit	Test Conditions
	P67 to P60(KWUL =	(1)	V _T -	$V_{cc} \times 0.2$ $V_{cc} B \times 0.2$	_	_	V	
voltage	00)*2*6, KIN15 to KIN8*7*8, IRQ2 to IRQ0*3, IRQ5 to IRQ3		V _T ⁺	_	_	$V_{cc} \times 0.7$ $V_{cc} B \times 0.7$	V	_
			$V_T^+ - V_T^-$	$V_{cc} \times 0.05$ $V_{cc} B \times$ 0.05	_	_	V	
Schmitt	P67 to P60		V _T	$V_{cc} \times 0.3$	_	_	V	
trigger input voltage (in level swiching)*6	(KWUL = 01)		V_{T}^{+}	_	_	$V_{cc} \times 0.7$		
		_	$V_{\scriptscriptstyle T}^{^+} - V_{\scriptscriptstyle T}^{^-}$	$V_{\text{cc}} \times 0.05$	_	_	_	
	P67 to P60 (KWUL = 10)		V _T	$V_{cc} \times 0.4$	_	_	=	
			V _T ⁺	_	_	$V_{cc} \times 0.8$	=	
		_	$V_T^+ - V_T^-$	$V_{cc} \times 0.03$	_	_	_	
	P67 to P60 (KWUL = 11)		V _T	$V_{cc} \times 0.45$	_	_	_	
			V _T ⁺	_	_	$V_{cc} \times 0.9$	=	
			$V_{T}^{+} - V_{T}^{-}$	0.05	_	_		
Input high voltage	RES, STBY, NMI, MD1, MD0	(2)	V_{IH}	$V_{cc} \times 0.9$	_	V _{cc} +0.3	V	
	EXTAL	-		$V_{cc} \times 0.7$	_	V _{cc} +0.3	V	_
	PA7 to PA0*7			$V_{cc}B \times 0.7$	_	V _{cc} B +0.3	V	_
	Port 7			$V_{cc} \times 0.7$	_	AV _{cc} +0.3	V	_
	Input pins except (1) and (2) above			V _{cc} × 0.7	_	V _{cc} +0.3	V	

Item			Symbol	Min	Тур	Max	Unit	Test Conditions
Input low voltage	RES, STBY, MD1, MD0	(3)	V _{IL}	-0.3	_	$V_{cc} \times 0.1$	V	
	PA7 to PA0	-		-0.3	_	$V_{cc}B \times 0.2$	V	V _{cc} B < 4.0 V
						0.8	V	$V_{cc}B = 4.0 \text{ V}$ to 5.5 V
	NMI, EXTAL,		_	-0.3	_	$V_{cc} \times 0.2$	V	V _{cc} < 4.0 V
	input pins except (1) and (3) above					0.8	V	$V_{cc} = 4.0 \text{ V to}$ 5.5 V
Output high voltage	(except P97, and P52*4)*5 *8		V _{OH}	V _{cc} -0.5 V _{cc} B -0.5	_	_	V	$I_{OH} = -200 \ \mu A$
				V _{cc} -1.0 V _{cc} B -1.0	_	_	V	$I_{OH} = -1 \text{ mA}$ $(V_{CC} < 4.0 \text{ V},$ $V_{CC}B < 4.0 \text{ V})$
	P97, P52*4		_	1.0	_	_	V	$I_{OH} = -1 \text{ mA}$
Output low voltage	All output pins (except RESO)*5		V _{oL}	_	_	0.4	V	I _{oL} = 1.6 mA
	Ports 1 to 3			_	_	1.0	V	$\begin{split} I_{_{OL}} &= 5 \text{ mA} \\ (V_{_{CC}} < 4.0 \text{ V}), \\ I_{_{OL}} &= 10 \text{ mA} \\ (4.0 \text{ V} \leq V_{_{CC}} \leq \\ 5.5 \text{ V}) \end{split}$
	RESO		_	_	_	0.4	V	I _{oL} = 1.6 mA
Input	RES		I _{in}	_	_	10.0	μΑ	$V_{in} = 0.5 \text{ to}$
leakage current	STBY, NMI, MD1, MD0		_	_	_	1.0	μΑ	[−] V _{cc} –0.5 V
	Port 7			_	_	1.0	μΑ	$V_{in} = 0.5 \text{ to}$ AV _{CC} -0.5 V
Three-state leakage current (off state)	Ports 1 to 6, 8, 9, A**, B		I _{TSI}	_	_	1.0	μA	$V_{in} = 0.5 \text{ to}$ $V_{cc} - 0.5 \text{ V},$ $V_{in} = 0.5 \text{ to}$ $V_{cc}B - 0.5 \text{ V}$

Item		Symbol	Min	Тур	Max	Unit	Test Conditions
Input pull-	Ports 1 to 3	-I _P	10	_	150	μΑ	$V_{in} = 0 V$,
up MOS current	Ports A*8, B, Port 6 (P6PUE = 0)	_	30	_	250	μΑ	$^{-}$ V _{cc} = 3.0 V to 3.6 V, V _{cc} B = 3.0 V - to 3.6 V
	Port 6 (P6PUE = 1)	_	3	_	70	μΑ	10 0.0 7
Input	RES (4)	C _{in}	_	_	80	pF	$V_{in} = 0 V$,
capacitance	NMI		_	_	50	pF	⁻f = 1 MHz, - T₂ = 25°C
	P52, P97, P42, P86, PA7 to PA2		_	_	20	pF	_
	Input pins except (4) above	_	_	_	15	pF	
Current	Normal operation	I _{cc}	_	50	70	mA	f = 10 MHz
dissipation*9	Sleep mode	_	_	40	60	mA	f = 10 MHz
	Standby mode*10	_	_	0.01	5.0	μΑ	T _a ≤ 50°C
			_	_	20.0	μΑ	50°C < T _a
Analog power	During A/D, D/A conversion	Al _{cc}	_	1.2	2.0	mA	
supply current	Idle	_	_	0.01	5.0	μΑ	$AV_{cc} = 2.0 \text{ V}$ to 5.5 V
Reference	During A/D conversion	AI_{ref}	_	0.5	1.0	mΑ	
power supply current	During A/D, D/A conversion	_	_	2.0	5.0	mA	_
Current	Idle	_	_	0.01	5.0	μΑ	$AV_{ref} = 2.0 \text{ V}$ to AV_{CC}
Analog power	er supply voltage*1	AV _{cc}	3.0	_	5.5	V	Operating
		-	2.0	_	5.5	V	Idle/not used
RAM standb	y voltage	V_{RAM}	2.0	_	_	V	

Notes: 1. Do not leave the AVCC, AV_{ref}, and AVSS pins open even if the A/D converter and D/A converter are not used.

Even if the A/D converter and D/A converter are not used, apply a value in the range 2.0 V to 5.5 V to AVCC and AV $_{\rm ref}$ pins by connection to the power supply (V $_{\rm cc}$), or some other method. Ensure that AV $_{\rm ref}$ \leq AV $_{\rm cc}$.

- 2. P67 to P60 include supporting module inputs multiplexed on those pins.
- 3. IRQ2 includes the ADTRG signal multiplexed on that pin.

- In the H8S/2148 Group, P52/SCK0/SCL0 and P97/SDA0 are NMOS push-pull outputs.
 An external pull-up resistor is necessary to provide high-level output from SCL0 and SDA0 (ICE = 1).
 - In the H8S/2148 Group, P52/SCK0 and P97 (ICE = 0) high levels are driven by NMOS.
- 5. When IICS = 0, ICE = 0, and KBIOE = 0. Low-level output when the bus drive function is selected is determined separately.
- 6. The upper limit of the port 6 applied voltage is V_{cc} +0.3 V when CIN input is not selected, and the lower of V_{cc} +0.3 V and AV_{cc} +0.3 V when CIN input is selected. When a pin is in output mode, the output voltage is equivalent to the applied voltage.
- 7. The upper limit of the port A applied voltage is $V_{cc}B + 0.3 \text{ V}$ when CIN input is not selected, and the lower of $V_{cc}B + 0.3 \text{ V}$ and $AV_{cc} + 0.3 \text{ V}$ when CIN input is selected. When a pin is in output mode, the output voltage is equivalent to the applied voltage.
- 8. The port A characteristics depend on $V_{\rm cc}B$, and the other pins characteristics depend on $V_{\rm cc}$.
- 9. Current dissipation values are for V_{IH} min = V_{cc} -0.5 V, V_{cc} B -0.5 V, and V_{IL} max = 0.5 V with all output pins unloaded and the on-chip pull-up MOSs in the off state.
- 10. The values are for V $_{\rm RAM} \le$ V $_{\rm CC}$ < 3.0 V, V $_{\rm IH}$ min = V $_{\rm CC} \times$ 0.9, V $_{\rm CC}$ B \times 0.9, and V $_{\rm IL}$ max = 0.3 V.
- 11. For flash memory program/erase operations, the applicable ranges are V_{cc} = 3.0 V to 3.6 V and Ta = 0 to +75°C.

Table 26.4 Permissible Output Currents

Conditions: $V_{cc} = 4.0 \text{ V}$ to 5.5 V, $V_{cc}B = 4.0 \text{ V}$ to 5.5 V, $V_{ss} = 0 \text{ V}$, $Ta = -20 \text{ to } +75^{\circ}\text{C}$ (regular specifications), $T_a = -40 \text{ to } +85^{\circ}\text{C}$ (wide-range specifications)

Item		Symbol	Min	Тур	Max	Unit
Permissible output low current (per pin)	SCL1, SCL0, SDA1, SDA0, PS2AC to PS2CC, PS2AD to PS2CD, PA7 to PA4 (bus drive function selected)	I _{OL}	_	_	20	mA
	Ports 1, 2, 3		_	_	10	mA
	RESO	_	_	_	3	mA
	Other output pins		_	_	2	mA
Permissible output	Total of ports 1, 2, and 3	\sum I _{OL}	_	_	80	mA
low current (total)	Total of all output pins, including the above		_	_	120	mA
Permissible output high current (per pin)	All output pins	-I _{OH}	_	_	2	mA
Permissible output high current (total)	Total of all output pins	$\Sigma - \mathbf{I}_{OH}$	_	_	40	mA

Conditions: $V_{cc} = 3.0 \text{ V}$ to 5.5 V, $V_{cc}B = 3.0 \text{ V}$ to 5.5 V, $V_{ss} = 0 \text{ V}$, $T_a = -20 \text{ to } +75^{\circ}\text{C}$

Item		Symbol	Min	Тур	Max	Unit
Permissible output low current (per pin)	SCL1, SCL0, SDA1, SDA0, PS2AC to PS2CC, PS2AD to PS2CD, PA7 to PA4 (bus drive function selected)	I _{OL}	_	_	10	mA
	Ports 1, 2, 3	_	_	_	2	mA
	RESO	_	_	_	1	mA
	Other output pins		_	_	1	mA
Permissible output	Total of ports 1, 2, and 3	\sum I _{OL}	_	_	40	mA
low current (total)	Total of all output pins, including the above		_	_	60	mA
Permissible output high current (per pin)	All output pins	-I _{OH}	_	_	2	mA
Permissible output high current (total)	Total of all output pins	$\Sigma - \mathbf{I}_{OH}$	_	_	30	mA

Notes: 1. To protect chip reliability, do not exceed the output current values in table 26.4.

2. When driving a Darlington pair or LED, always insert a current-limiting resistor in the output line, as show in figures 26.1 and 26.2.

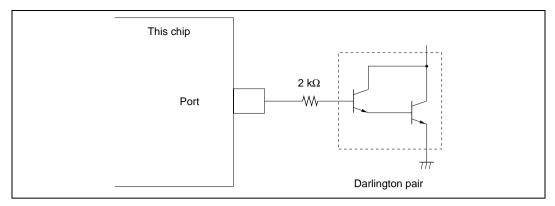


Figure 26.1 Darlington Pair Drive Circuit (Example)

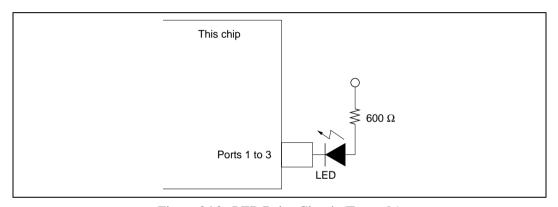


Figure 26.2 LED Drive Circuit (Example)

Table 26.5 Bus Drive Characteristics

Conditions: $V_{CC} = 3.0 \text{ V}$ to 5.5 V, $V_{ss} = 0 \text{ V}$

Applicable Pins: SCL1, SCL0, SDA1, SDA0 (bus drive function selected)

Item	Symbol	Min	Тур	Max	Unit	Test Conditions
Schmitt trigger	V _T -	$V_{cc} \times 0.3$	_	_	V	$V_{cc} = 3.0 \text{ V to } 5.5 \text{ V}$
input voltage	V _T ⁺	_	_	$V_{cc} \times 0.7$	_	$V_{cc} = 3.0 \text{ V to } 5.5 \text{ V}$
	$V_T^+ - V_T^-$	$V_{cc} \times 0.05$	_	_		V_{cc} = 3.0 V to 5.5 V
Input high voltage	V _{IH}	$V_{cc} \times 0.7$	_	V _{cc} +0.5	V	$V_{cc} = 3.0 \text{ V to } 5.5 \text{ V}$
Input low voltage	V _{IL}	-0.5	_	$V_{cc} \times 0.3$	_	$V_{cc} = 3.0 \text{ V to } 5.5 \text{ V}$
Output low voltage	V _{oL}	_	_	0.8	V	I _{OL} = 16 mA,
						$V_{cc} = 4.5 \text{ V to } 5.5 \text{ V}$
		_	_	0.5	_	$I_{OL} = 8 \text{ mA}$
		_	_	0.4		I _{oL} = 3 mA
Input capacitance	C _{in}	_	_	20	pF	$V_{in} = 0 \text{ V}, f = 1 \text{ MHz},$
						$T_a = 25^{\circ}C$
Three-state leakage	I _{TSI}	_	_	1.0	μΑ	$V_{in} = 0.5 \text{ to } V_{cc} - 0.5 \text{ V}$
current (off state)						
SCL, SDA output	t _{of}	20 + 0.1 Cb	_	250	ns	$V_{cc} = 3.0 \text{ V to } 5.5 \text{ V}$
fall time						

Conditions: $V_{cc} = 3.0 \text{ V}$ to 5.5 V, $V_{cc}B = 3.0 \text{ V}$ to 5.5 V, $V_{ss} = 0 \text{ V}$

Applicable Pins: PS2AC, PS2AD, PS2BC, PS2BD, PS2CC, PS2CD, PA7 to PA4 (bus drive

function selected)

Item	Symbol	Min	Тур	Max	Unit	Test Conditions
Output low voltage	V _{oL}	_	_	8.0	V	$I_{OL} = 16 \text{ mA},$ $V_{CC}B = 4.5 \text{ V to } 5.5 \text{ V}$
		_	_	0.5		I _{oL} = 8 mA
		_	_	0.4		I _{oL} = 3 mA

26.2.3 AC Characteristics

Clock timing, control signal timing, bus timing, and timing of on-chip supporting modules list the following.

Figure 26.4 shows the test conditions for the AC characteristics.



(1) Clock Timing

Table 26.6 shows the clock timing. The clock timing specified here covers clock (ϕ) output and clock pulse generator (crystal) and external clock input (EXTAL pin) oscillation settling times. For details of external clock input (EXTAL pin and EXCL pin) timing, see section 24, Clock Pulse Generator.

Table 26.6 Clock Timing

Condition A: $V_{cc} = 5.0 \text{ V} \pm 10\%$, $V_{cc}B = 5.0 \text{ V} \pm 10\%$, $V_{ss} = 0 \text{ V}$, $\phi = 2 \text{ MHz}$ to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$ (regular specifications), $T_a = -40 \text{ to } +85^{\circ}\text{C}$ (wide-range specifications)

Condition B: $V_{cc} = 4.0 \text{ V to } 5.5 \text{ V}, \ V_{cc}B = 4.0 \text{ V to } 5.5 \text{ V}, \ V_{ss} = 0 \text{ V}, \ \phi = 2 \text{ MHz to maximum}$ operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$ (regular specifications), $T_a = -40 \text{ to } +85^{\circ}\text{C}$ (wide-range specifications)

Condition C: $V_{cc} = 3.0 \text{ V}$ to 5.5 V, $V_{cc}B = 3.0 \text{ V}$ to 5.5 V, $V_{ss} = 0 \text{ V}$, $\phi = 2 \text{ MHz}$ to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$

		Cond	dition A	Cond	dition B	Con	dition C		
		20	MHz	16	MHz	10	MHz		Test
Item	Symbol	Min	Max	Min	Max	Min	Max	Unit	Conditions
Clock cycle time	t _{cyc}	50	500	62.5	500	100	500	ns	Figure 26.5
Clock high pulse width	t _{ch}	17	_	20	_	30	_	ns	Figure 26.5
Clock low pulse width	t _{cL}	17	_	20	_	30	_	ns	_
Clock rise time	t _{Cr}	_	8	_	10	_	20	ns	_
Clock fall time	t _{Cf}	_	8	_	10	_	20	ns	
Oscillation settling time at reset (crystal)	t _{osc1}	10	_	10	_	20	_	ms	Figure 26.6 Figure 26.7
Oscillation settling time in software standby (crystal)	t _{osc2}	8	_	8	_	8	_	ms	_
External clock output stabilization delay time	t _{DEXT}	500	_	500	_	500	_	μs	

(2) Control Signal Timing

Table 26.7 shows the control signal timing. The only external interrupts that can operate on the subclock ($\phi = 32.768 \text{ kHz}$) are NMI and IRO0, 1, 2, 6, and 7.

Table 26.7 Control Signal Timing

Condition A: $V_{cc} = 5.0 \text{ V} \pm 10\%$, $V_{cc}B = 5.0 \text{ V} \pm 10\%$, $V_{ss} = 0 \text{ V}$, $\phi = 32.768 \text{ kHz}$, 2 MHz to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$ (regular specifications), $T_a = -40 \text{ to } +85^{\circ}\text{C}$ (wide-range specifications)

Condition B: $V_{cc} = 4.0 \text{ V}$ to 5.5 V, $V_{cc}B = 4.0 \text{ V}$ to 5.5 V, $V_{ss} = 0 \text{ V}$, $\phi = 32.768 \text{ kHz}$, 2 MHz to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$ (regular specifications), $T_a = -40 \text{ to } +85^{\circ}\text{C}$ (wide-range specifications)

Condition C: $V_{cc} = 3.0 \text{ V}$ to 5.5 V, $V_{cc}B = 3.0 \text{ V}$ to 5.5 V, $V_{ss} = 0 \text{ V}$, $\phi = 32.768 \text{ kHz}$, 2 MHz to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$

		Cond	dition A	Cond	dition B	Cond	dition C		
		20	MHz	16	MHz	10	MHz		Test
Item	Symbol	Min	Max	Min	Max	Min	Max	Unit	Conditions
RES setup time	t _{RESS}	200	_	200	_	300	_	ns	Figure 26.8
RES pulse width	\mathbf{t}_{RESW}	20	_	20	_	20	_	$\mathbf{t}_{\mathrm{cyc}}$	
NMI setup time (NMI)	t _{NMIS}	150	_	150	_	250	_	ns	Figure 26.9
NMI hold time (NMI)	t _{nmih}	10	_	10	_	10	_	ns	_
NMI pulse width (NMI) (exiting software standby mode)	t _{NMIW}	200	_	200	_	200	_	ns	_
IRQ setup time (IRQ7 to IRQ0)	t _{IRQS}	150	_	150	_	250	_	ns	_
IRQ hold time (IRQ7 to IRQ0)	t _{IRQH}	10	_	10	_	10	_	ns	_
IRQ pulse width (IRQ7, IRQ6, IRQ2 to IRQ0) (exiting software standby mode)	t _{IRQW}	200	_	200	_	200	_	ns	_

(3) Bus Timing

Table 26.8 shows the bus timing. Operation in external expansion mode is not guaranteed when operating on the subclock ($\phi = 32.768 \text{ kHz}$).

Table 26.8 Bus Timing (1) (Nomal Mode)

- Condition A: $V_{cc} = 5.0 \text{ V} \pm 10\%$, $V_{cc}B = 5.0 \text{ V} \pm 10\%$, $V_{ss} = 0 \text{ V}$, $\phi = 2 \text{ MHz}$ to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$ (regular specifications), $T_a = -40 \text{ to } +85^{\circ}\text{C}$ (wide-range specifications)
- Condition B: $V_{cc} = 4.0 \text{ V}$ to 5.5 V, $V_{cc}B = 4.0 \text{ V}$ to 5.5 V, $V_{ss} = 0 \text{ V}$, $\phi = 2 \text{ MHz}$ to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$ (regular specifications), $T_a = -40 \text{ to } +85^{\circ}\text{C}$ (wide-range specifications)
- Condition C: $V_{cc} = 3.0 \text{ V}$ to 5.5 V, $V_{cc}B = 3.0 \text{ V}$ to 5.5 V, $V_{ss} = 0 \text{ V}$, $\phi = 2 \text{ MHz}$ to maximum operating frequency, $T_a = -20 \text{ to } +75 ^{\circ}\text{C}$

		Cond	ition A	Cond	ition B	Cond	ition C		
		20	MHz	16	MHz	10	MHz	•	Test
Item	Symbol	Min	Max	Min	Max	Min	Max	Unit	Conditions
Address delay time	t _{AD}	_	20	_	30	_	40	ns	Figure 26.10 to
Address setup time	t _{AS}	$0.5 \times t_{\text{cyc}} -15$	_	$0.5 \times t_{\text{cyc}}$ -20	_	$0.5 \times t_{cyc} -30$	_	ns	figure 26.14
Address hold time	t _{AH}	$0.5 \times t_{\rm cyc} -10$	_	$0.5 \times t_{\text{cyc}} -15$	_	$0.5 \times t_{\scriptscriptstyle ext{cyc}}$ -20	_	ns	
CS delay time (IOS)	t _{CSD}	_	20	_	30	_	40	ns	
AS delay time	t _{ASD}	_	30	_	45	_	60	ns	-
RD delay time 1	t _{RSD1}	_	30	_	45	_	60	ns	-
RD delay time 2	t _{RSD2}	_	30	_	45	_	60	ns	-
Read data setup time	t _{RDS}	15	_	20	_	35	_	ns	_
Read data hold time	t _{RDH}	0	_	0	_	0	_	ns	-

		Cond	lition A	Cond	ition B	Cond	ition C		
		20	MHz	16	MHz	10	MHz		Test
Item	Symbol	Min	Max	Min	Max	Min	Max	Unit	Conditions
Read data access time 1	t _{ACC1}	_	1.0 × t _{cyc} –30	_	1.0 × t _{cyc} –40	_	1.0 × t _{cyc} –60	ns	Figure 26.10 to figure 26.14
Read data access time 2	t _{ACC2}	_	1.5 × t _{cyc} –25	_	1.5 × t _{cyc} –35	_	1.5 × t _{cyc} –50	ns	_
Read data access time 3	t _{ACC3}	_	2.0 × t _{cyc} -30	_	2.0 × t _{cyc} -40	_	2.0 × t _{cyc} –60	ns	_
Read data access time 4	t _{ACC4}	_	2.5 × t _{cyc} –25	_	2.5 × t _{cyc} –35	_	2.5 × t _{cyc} –50	ns	_
Read data access time 5	t _{ACC5}	_	3.0 × t _{cyc} –30	_	3.0 × t _{cyc} –40	_	3.0 × t _{cyc} –60	ns	_
WR delay time 1	t _{WRD1}	_	30	_	45	_	60	ns	_
WR delay time 2	t _{WRD2}	_	30	_	45	_	60	ns	_
WR pulse width 1	t _{wsw1}	1.0 × t _{cyc} –20	_	1.0 × t _{cyc} –30	_	1.0 × t _{cyc} –40	_	ns	_
WR pulse width 2	t _{wsw2}	1.5 × t _{cyc} –20	_	1.5 × t _{cyc} –30	_	1.5 × t _{cyc} –40	_	ns	_
Write data delay time	t _{wdd}	_	30	_	45	_	60	ns	_
Write data setup time	t _{wds}	0	_	0	_	0	_	ns	_
Write data hold time	t _{wdh}	10	_	15	_	20	_	ns	_
WAIT setup time	t _{wrs}	30	_	45	_	60	_	ns	_
WAIT hold time	t _{wth}	5	_	5	_	10	_	ns	_

Table 26.8 Bus Timing (2) (Advanced mode)

Condition A: $V_{cc} = 5.0 \text{ V} \pm 10\%$, $V_{cc}B = 5.0 \text{ V} \pm 10\%$, $V_{ss} = 0 \text{ V}$, $\phi = 2 \text{ MHz}$ to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$ (regular specifications), $T_a = -40 \text{ to } +85^{\circ}\text{C}$ (wide-range specifications)

Condition B: $V_{cc} = 4.0 \text{ V}$ to 5.5 V, $V_{cc}B = 4.0 \text{ V}$ to 5.5 V, $V_{ss} = 0 \text{ V}$, $\phi = 2 \text{ MHz}$ to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$ (regular specifications), $T_a = -40 \text{ to } +85^{\circ}\text{C}$ (wide-range specifications)

Condition C: $V_{cc} = 3.0 \text{ V}$ to 5.5 V, $V_{cc}B = 3.0 \text{ V}$ to 5.5 V, $V_{ss} = 0 \text{ V}$, $\phi = 2 \text{ MHz}$ to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$

		Cond	ition A	Cond	ition B	Cond	ition C		
		20	MHz	16	MHz	10	MHz		Test
Item	Symbol	Min	Max	Min	Max	Min	Max	Unit	Conditions
Address delay time	t _{AD}	_	30	_	45	_	60	ns	Figure 26.10 to
Address setup time	t _{AS}	$0.5 \times \atop t_{\rm cyc} -25$	_	$0.5 \times t_{\rm cyc} -35$	_	$0.5 \times t_{\rm cyc} -50$	_	ns	figure 26.14
Address hold time	t _{AH}	$\begin{array}{l} 0.5 \times \\ t_{_{cyc}} - 10 \end{array}$	_	$0.5 \times t_{\text{cyc}} -15$	_	$0.5 \times t_{_{\mathrm{cyc}}}$ –20	_	ns	_
CS delay time (IOS)	t _{CSD}	_	30	_	45	_	60	ns	_
AS delay time	t _{ASD}	_	30	_	45	_	60	ns	
RD delay time 1	t _{RSD1}	_	30	_	45	_	60	ns	
RD delay time 2	t _{RSD2}	_	30	_	45	_	60	ns	
Read data setup time	t _{RDS}	15	_	20	_	35	_	ns	
Read data hold time	t _{rdh}	0	_	0	_	0	_	ns	
Read data access time 1	t _{ACC1}	_	$1.0 \times t_{\rm cyc}$ –40	_	$1.0 \times t_{\rm cyc}$ –55	-	$1.0 \times t_{\rm cyc}$ -80	ns	
Read data access time 2	t _{ACC2}	_	1.5 × t _{cyc} –25	_	1.5 × t _{cyc} –35	_	1.5 × t _{cyc} –50	ns	_

		Cond	ition A	Cond	ition B	Cond	ition C		
		20	MHz	16	MHz	10	MHz		Test
Item	Symbol	Min	Max	Min	Max	Min	Max	Unit	Conditions
Read data access time 3	t _{ACC3}	_	2.0 × t _{cyc} -40	_	$2.0 \times t_{\text{cyc}} -55$	_	2.0 × t _{cyc} -80	ns	Figure 26.10 to figure 26.14
Read data access time 4	t _{ACC4}	_	2.5 × t _{cyc} –25	_	$2.5 \times t_{\text{cyc}} -35$	_	$2.5 \times t_{\text{cyc}} -50$	ns	_
Read data access time 5	t _{ACC5}	_	3.0 × t _{cyc} –40	_	$3.0 \times t_{cyc} -55$	_	$3.0 \times t_{cyc} -80$	ns	_
WR delay time 1	t _{WRD1}	_	30	_	45	_	60	ns	_
WR delay time 2	t _{WRD2}	_	30	_	45	_	60	ns	_
WR pulse width 1	t _{wsw1}	1.0 × t _{cyc} –20	_	1.0 × t _{cyc} –30	_	1.0 × t _{cyc} –40	_	ns	_
WR pulse width 2	t _{wsw2}	1.5 × t _{cyc} –20	_	1.5 × t _{cyc} –30	_	1.5 × t _{cyc} –40	_	ns	_
Write data delay time	t _{wdd}	_	30	_	45	_	60	ns	_
Write data setup time	t _{wds}	0	_	0	_	0	_	ns	_
Write data hold time	t _{wdh}	10	_	15	_	20	_	ns	_
WAIT setup time	t _{wts}	30	_	45	_	60	_	ns	_
WAIT hold time	t _{wth}	5	_	5	_	10	_	ns	_

(4) Timing of On-Chip Supporting Modules

Tables 26.9 to 26.11 show the on-chip supporting module timing. The only on-chip supporting modules that can operate in subclock operation ($\phi = 32.768 \text{ kHz}$) are the I/O ports, external interrupts (NMI and IRQ0, 1, 2, 6, and 7), the watchdog timer, and the 8-bit timer (channels 0 and 1).

Table 26.9 Timing of On-Chip Supporting Modules (1)

Condition A: $V_{cc} = 5.0 \text{ V} \pm 10\%$, $V_{cc}B = 5.0 \text{ V} \pm 10\%$, $V_{ss} = 0 \text{ V}$, $\phi = 32.768 \text{ kHz}^*$, 2 MHz to maximum operating frequency, $T_a = -20$ to +75°C (regular specifications), $T_{\circ} = -40 \text{ to } +85^{\circ}\text{C}$ (wide-range specifications)

Condition B: $V_{cc} = 4.0 \text{ V}$ to 5.5 V, $V_{cc}B = 4.0 \text{ V}$ to 5.5 V, $V_{ss} = 0 \text{ V}$, $\phi = 32.768 \text{ kHz}^*$, 2 MHz to maximum operating frequency, $T_a = -20$ to +75°C (regular specifications), $T_a = -40 \text{ to } +85^{\circ}\text{C}$ (wide-range specifications)

Condition C: $V_{cc} = 3.0 \text{ V}$ to 5.5 V, $V_{cc}B = 3.0 \text{ V}$ to 5.5 V, $V_{ss} = 0 \text{ V}$, $\phi = 32.768 \text{ kHz}^*$, 2 MHz to maximum operating frequency, $T_a = -20$ to $+75^{\circ}$ C

				Cond	dition A	Cond	lition B	Conc	lition C		
				20	MHz	16	MHz	10	MHz		Test
Item			Symbol	Min	Max	Min	Max	Min	Max	Unit	Conditions
I/O ports	Output d	ata delay	t _{PWD}	_	50	_	50	_	100	ns	Figure 26.15
	Input dat time	a setup	t _{PRS}	30	_	30	_	50	_		
	Input dat time	a hold	t _{PRH}	30	_	30	_	50	_		
FRT	Timer ou time	tput delay	t _{FTOD}	_	50	_	50	_	100	ns	Figure 26.16
	Timer inp	out setup	t _{FTIS}	30	_	30	_	50	_		
	Timer clo	-	t _{FTCS}	30	_	30	_	50	_		Figure 26.17
	Timer clock	Single edge	t _{FTCWH}	1.5	_	1.5	_	1.5	_	t _{cyc}	-
	pulse width	Both edges	t _{FTCWL}	2.5	_	2.5	_	2.5	_		

				Cond	dition A	Conc	lition B	Conc	lition C		
				20	MHz	16	MHz	10	MHz		Test
Item			Symbol	Min	Max	Min	Max	Min	Max	Unit	Conditions
TMR	Timer of delay ti		\mathbf{t}_{TMOD}	_	50	_	50	_	100	ns	Figure 26.18
	Timer r setup ti	eset input me	\mathbf{t}_{TMRS}	30	_	30	_	50	_		Figure 26.20
	Timer of setup ti	clock input me	\mathbf{t}_{TMCS}	30	_	30	_	50	_		Figure 26.19
	Timer clock	Single edge	t _{TMCWH}	1.5	_	1.5	_	1.5	_	t _{cyc}	_
	pulse width	Both edges	t _{TMCWL}	2.5	_	2.5	_	2.5	_		
PWM, PWMX	Pulse delay ti		t _{PWOD}	_	50	_	50	_	100	ns	Figure 26.21
SCI	Input clock	Asynchro- nous	t _{Scyc}	4	_	4	_	4	_	t _{cyc}	Figure 26.22
	cycle	Synchro- nous	_	6	_	6	_	6	_		
	Input cl width	ock pulse	t _{sckw}	0.4	0.6	0.4	0.6	0.4	0.6	t _{Scyc}	_
	Input cl	ock rise	t _{SCKr}	_	1.5	_	1.5	_	1.5	t _{cyc}	=
	Input cl	ock fall	t _{SCKf}	_	1.5	_	1.5	_	1.5		
	Transm delay ti (synchi		t _{TXD}	_	50	_	50	_	100	ns	Figure 26.23
		e data setup ynchronous)	t _{RXS}	50	_	50	_	100	_	ns	_
		e data hold ynchronous)	t _{RXH}	50	_	50	_	100	_	ns	_
A/D conver- ter		input setup	t _{TRGS}	30	_	30	_	50	_	ns	Figure 26.24
WDT	RESO time	output delay	t _{RESD}	_	100	_	120	_	200	ns	Figure 26.25
	RESO width	output pulse	t _{RESOW}	132	_	132	_	132	_	t _{cyc}	_

Only supporting modules that can be used in subclock operation

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Table 26.9 Timing of On-Chip Supporting Modules (2)

Condition A: $V_{cc} = 5.0 \text{ V} \pm 10\%$, $V_{cc}B = 5.0 \text{ V} \pm 10\%$, $V_{ss} = 0 \text{ V}$, $\phi = 2 \text{ MHz}$ to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$ (regular specifications), $T_a = -40 \text{ to } +85^{\circ}\text{C}$ (wide-range specifications)

Condition B: $V_{cc} = 4.0 \text{ V}$ to 5.5 V, $V_{cc}B = 4.0 \text{ V}$ to 5.5 V, $V_{ss} = 0 \text{ V}$, $\phi = 2 \text{ MHz}$ to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$ (regular specifications), $T_a = -40 \text{ to } +85^{\circ}\text{C}$ (wide-range specifications)

Condition C: $V_{cc} = 3.0 \text{ V}$ to 5.5 V, $V_{cc}B = 3.0 \text{ V}$ to 5.5 V, $V_{ss} = 0 \text{ V}$, $\phi = 2 \text{ MHz}$ to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$

				Cond	dition A	Cond	dition B	Cond	dition C		
				20	MHz	16	MHz	10	MHz		Test
Item			Symbol	Min	Max	Min	Max	Min	Max	Unit	Conditions
HIF read cycle	CS/HA time	0 setup	t _{HAR}	10	_	10	_	10	_	ns	Figure 26.26
	CS/HA	0 hold time	t _{HRA}	10	_	10	_	10	_	ns	
	IOR pu	lse width	t _{HRPW}	120	_	120	_	220	_	ns	_
	HDB de	elay time	t _{HRD}	_	100	_	100	_	200	ns	_
	HDB h	old time	t _{HRF}	0	25	0	25	0	40	ns	_
	HIRQ o	delay time	t _{HIRQ}	_	120	_	120	_	200	ns	_
HIF write cycle	CS/HA time	0 setup	t _{HAW}	10	_	10	_	10	_	ns	_
	CS/HA	0 hold time	t _{HWA}	10	_	10	_	10	_	ns	
	IOW pu	ulse width	t _{HWPW}	60		60	_	100	_	ns	_
	HDB setup time	Fast A20 gate not used	t _{HDW}	30	_	30	_	50	_	ns	_
		Fast A20 gate used	_	45	_	55		85	_	ns	_
	HDB h	old time	t _{HWD}	15	_	15	_	25	_	ns	=
	GA20 d	delay time	t _{HGA}	_	90	_	90	_	180	ns	=

Table 26.10 Keyboard Buffer Controller Timing

Conditions: $V_{cc} = 3.0 \text{ V}$ to 5.5 V, $V_{cc}B = 3.0 \text{ V}$ to 5.5 V, $V_{ss} = 0 \text{ V}$, $\phi = 2 \text{ MHz}$ to maximum

operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$

		R	atings			Test	
Item	Symbol	Min	Тур	Max	Unit	Conditions	Notes
KCLK, KD output fall time	t _{KBF}	20 + 0.1 Cb	_	250	ns		Figure 26.27
KCLK, KD input data hold time	t _{KBIH}	150	_	_	ns		_
KCLK, KD input data setup time	t _{KBIS}	150	_	_	ns		_
KCLK, KD output delay time	t _{KBOD}	_	_	450	ns		_
KCLK, KD capacitive load	C _b	_	_	400	pF		_

Table 26.11 I²C Bus Timing

Conditions: $V_{cc} = 3.0 \text{ V}$ to 5.5 V, $V_{ss} = 0 \text{ V}$, $\phi = 5 \text{ MHz}$ to maximum operating frequency

			Ratings				
Item	Symbol	Min	Тур	Max	Unit	Test Conditions	Notes
SCL input cycle time	t _{SCL}	12	_	_	t _{cyc}		Figure 26.28
SCL input high pulse width	t _{SCLH}	3	_	_	t _{cyc}		_
SCL input low pulse width	t _{SCLL}	5	_	_	t _{cyc}		
SCL, SDA input rise time	t _{Sr}	_	_	7.5*	t _{cyc}		_
SCL, SDA input fall time	t _{sf}	_	_	300	ns		_
SCL, SDA input spike pulse elimination time	t _{sp}	_	_	1	t _{cyc}		_
SDA input bus free time	t _{BUF}	5	_	_	t _{cyc}		
Start condition input hold time	t _{STAH}	3	_	_	t _{cyc}		_
Retransmission start condition input setup time	t _{STAS}	3	_	_	t _{cyc}		_
Stop condition input setup time	t _{stos}	3	_	_	t _{cyc}		_
Data input setup time	t _{sdas}	0.5	_	_	t _{cyc}		_
Data input hold time	t _{sdah}	0	_	_	ns		_
SCL, SDA capacitive load	C _b	_	_	400	pF		_

Note: * 17.5t_{cyc} can be set according to the clock selected for use by the I²C module. For details, see section 16.4, Usage Notes.

26.2.4 A/D Conversion Characteristics

Tables 26.12 and 26.13 list the A/D conversion characteristics.

Table 26.12 A/D Conversion Characteristics

(AN7 to AN0 Input: 134/266-State Conversion)

Condition A: $V_{CC} = 5.0 \text{ V} \pm 10\%$, $AV_{CC} = 5.0 \text{ V} \pm 10\%$, $AV_{ref} = 4.5 \text{ V}$ to AV_{CC} ,

 $V_{ss} = AV_{ss} = 0 \text{ V}, \phi = 2 \text{ MHz}$ to maximum operating frequency,

 $T_{\circ} = -20 \text{ to } +75^{\circ}\text{C}$ (regular specifications),

 $T_a = -40 \text{ to } +85^{\circ}\text{C}$ (wide-range specifications)

Condition B: $V_{CC} = 4.0 \text{ V}$ to 5.5 V, $AV_{CC} = 4.0 \text{ V}$ to 5.5 V, $AV_{ref} = 4.0 \text{ V}$ to AV_{CC}

 $V_{ss} = AV_{ss} = 0 \text{ V}, \phi = 2 \text{ MHz}$ to maximum operating frequency,

 $T_{\circ} = -20 \text{ to } +75^{\circ}\text{C}$ (regular specifications),

 $T_a = -40 \text{ to } +85^{\circ}\text{C}$ (wide-range specifications)

Condition C: $V_{cc} = 3.0 \text{ V}$ to 5.5 V, $AV_{cc} = 3.0 \text{ V}$ to 5.5 V, $AV_{ref} = 3.0 \text{ V}$ to AV_{cc}

 $V_{ss} = AV_{ss} = 0 \text{ V}, \phi = 2 \text{ MHz}$ to maximum operating frequency,

 $T_{a} = -20 \text{ to } +75^{\circ}\text{C}$

		Condition A			Conditio	n B		on C		
		20 MF	łz		16 MF	lz		10 MF	lz	_
Item	Min	Тур	Max	Min	Тур	Max	Min	Тур	Max	Unit
Resolution	10	10	10	10	10	10	10	10	10	Bits
Conversion time*5	_	_	6.7	_	_	8.4	_	_	13.4	μs
Analog input capacitance	_	_	20	_	_	20	_	_	20	pF
Permissible signal- source impedance	_	_	10 ^{*3} 5 ^{*4}	-	_	10 ^{*3} 5 ^{*4}	_	_	10 ^{*1} 5 ^{*2}	kΩ
Nonlinearity error	_	_	±3.0	_	_	±3.0	_	_	±7.0	LSB
Offset error	_	_	±3.5	_	_	±3.5	_	_	±7.5	LSB
Full-scale error	_	_	±3.5	_	_	±3.5	_	_	±7.5	LSB
Quantization error	_	_	±0.5	_	_	±0.5	_	_	±0.5	LSB
Absolute accuracy	_	_	±4.0	_	_	±4.0	_	_	±8.0	LSB

Notes: 1. When $4.0 \text{ V} \le \text{AV}_{cc} \le 5.5 \text{ V}$

- 2. When $3.0 \text{ V} \le \text{AV}_{cc} < 4.0 \text{ V}$
- 3. When conversion time \geq 11. 17 µs (CKS = 1 and $\phi \leq$ 12 MHz, or CKS = 0)
- 4. When conversion time < 11. 17 μ s (CKS = 1 and ϕ > 12 MHz)
- 5. In single mode and ϕ = maximum operating frequency.



Table 26.13 A/D Conversion Characteristics (CIN15 to CIN0 Input: 134/266-State Conversion)

Condition A:
$$V_{cc} = 5.0 \text{ V} \pm 10\%$$
, $AV_{cc} = 5.0 \text{ V} \pm 10\%$, $AV_{ref} = 4.5 \text{ V}$ to AV_{cc} , $V_{ss} = AV_{ss} = 0 \text{ V}$, $\phi = 2 \text{ MHz}$ to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$ (regular specifications), $T_a = -40 \text{ to } +85^{\circ}\text{C}$ (wide-range specifications)

Condition B:
$$V_{cc} = 4.0 \text{ V to } 5.5 \text{ V}, \text{ AV}_{cc} = 4.0 \text{ V to } 5.5 \text{ V}, \text{ AV}_{ref} = 4.0 \text{ V to } \text{AV}_{cc},$$

$$V_{ss} = \text{AV}_{ss} = 0 \text{ V}, \phi = 2 \text{ MHz to maximum operating frequency},$$

$$T_a = -20 \text{ to } +75^{\circ}\text{C (regular specifications)},$$

$$T_a = -40 \text{ to } +85^{\circ}\text{C (wide-range specifications)}$$

Condition C:
$$V_{cc} = 3.0 \text{ V to } 5.5 \text{ V}, \text{ AV}_{cc} = 3.0 \text{ V to } 5.5 \text{ V}, \text{ AV}_{ref} = 3.0 \text{ V to AV}_{cc},$$

$$V_{ss} = \text{AV}_{ss} = 0 \text{ V}, \phi = 2 \text{ MHz to maximum operating frequency},$$

$$T_a = -20 \text{ to } +75^{\circ}\text{C}$$

		Conditio	n A		Conditio	n B		Conditio	on C	
		20 MF	lz	16 MHz				10 MF	lz	-
Item	Min	Тур	Max	Min	Тур	Max	Min	Тур	Max	Unit
Resolution	10	10	10	10	10	10	10	10	10	Bits
Conversion time*5	_	_	6.7	_	_	8.4	_	_	13.4	μs
Analog input capacitance	_	_	20	_	_	20	_	_	20	pF
Permissible signal- source impedance	_	_	10 ^{*3} 5 ^{*4}		_	10 ^{*3} 5 ^{*4}		_	10 ^{*1} 5 ^{*2}	kΩ
Nonlinearity error	_	_	±5.0	_	_	±5.0	_	_	±11.0	LSB
Offset error	_	_	±5.5	_	_	±5.5	_	_	±11.5	LSB
Full-scale error	_	_	±5.5	_	_	±5.5	_	_	±11.5	LSB
Quantization error	_	_	±0.5	_	_	±0.5	_	_	±0.5	LSB
Absolute accuracy	_	_	±6.0	_	_	±6.0	_	_	±12.0	LSB

Notes: 1. When $4.0 \text{ V} \le \text{AV}_{cc} \le 5.5 \text{ V}$

- 2. When $3.0 \text{ V} \le \text{AV}_{cc} < 4.0 \text{ V}$
- 3. When conversion time \geq 11. 17 μs (CKS = 1 and $\varphi \leq$ 12 MHz, or CKS = 0)
- 4. When conversion time < 11. 17 μ s (CKS = 1 and ϕ > 12 MHz)
- 5. In single mode and ϕ = maximum operating frequency.

26.2.5 D/A Conversion Characteristics

Table 26.14 lists the D/A conversion characteristics.

Table 26.14 D/A Conversion Characteristics

Condition A: $V_{cc} = 5.0 \text{ V} \pm 10\%$, $AV_{cc} = 5.0 \text{ V} \pm 10\%$, $AV_{ref} = 4.5 \text{ V}$ to AV_{cc} ,

 $V_{ss} = AV_{ss} = 0 \text{ V}, \phi = 2 \text{ MHz}$ to maximum operating frequency,

 $T_a = -20 \text{ to } +75^{\circ}\text{C}$ (regular specifications),

 $T_{\circ} = -40 \text{ to } +85^{\circ}\text{C}$ (wide-range specifications)

Condition B: $V_{CC} = 4.0 \text{ V}$ to 5.5 V, $AV_{CC} = 4.0 \text{ V}$ to 5.5 V, $AV_{pet} = 4.0 \text{ V}$ to AV_{CC}

 $V_{ss} = AV_{ss} = 0 \text{ V}, \phi = 2 \text{ MHz}$ to maximum operating frequency,

 $T_a = -20 \text{ to } +75^{\circ}\text{C}$ (regular specifications),

 $T_a = -40 \text{ to } +85^{\circ}\text{C}$ (wide-range specifications)

Condition C: $V_{CC} = 3.0 \text{ V}$ to 5.5 V, $AV_{CC} = 3.0 \text{ V}$ to 5.5 V, $AV_{ref} = 3.0 \text{ V}$ to AV_{CC}

 $V_{ss} = AV_{ss} = 0 \text{ V}, \phi = 2 \text{ MHz}$ to maximum operating frequency,

 $T_{a} = -20 \text{ to } +75^{\circ}\text{C}$

		С	onditio	on A	Condition B		on B	С	onditio	on C	
			20 MF	łz		16 MHz			10 MHz		
Item		Min	Тур	Max	Min	Тур	Max	Min	Тур	Max	Unit
Resolution		8	8	8	8	8	8	8	8	8	Bits
Conversion time	With 20-pF load capacitance	_	_	10	_	_	10	_	_	10	μs
Absolute accuracy	With 2-MΩ load resistance		±1.0	±1.5	_	±1.0	±1.5	_	±2.0	±3.0	LSB
	With 4-MΩ load resistance	_	_	±1.0	_	_	±1.0	_	_	±2.0	-

Toct

26.2.6 **Flash Memory Characteristics**

Table 26.15 shows the flash memory characteristics.

Table 26.15 Flash Memory Characteristics (Programming/erasing operating range)

Conditions (5 V version):
$$V_{cc} = 5.0 \text{ V} \pm 10\%$$
, $V_{ss} = 0 \text{ V}$, $T_a = 0 \text{ to } +75^{\circ}\text{C}$ (regular specifications), $T_a = 0 \text{ to } +85^{\circ}\text{C}$ (wide-range specifications) (3 V version): $V_{cc} = 3.0 \text{ V}$ to 3.6V, $V_{ss} = 0 \text{ V}$, $T_a = 0 \text{ to } +75^{\circ}\text{C}$

Item	Symbol	Min	Тур	Max	Unit	Test Condition	
Programming t	ime ^{*1 *2 *4}	tP	_	10	200	ms/ 32 bytes	
Erase time*1 *3	*6	tE	_	100	1200	ms/ block	
Reprogrammin	g count	N _{wec}	100*8	10000*9	'—	Times	
Data retention	time ^{*10}	t _{DRP}	10	_	_	Years	
Programming	Wait time after SWE-bit setting*1	х	10	_	_	μs	
	Wait time after PSU-bit setting*1	у	50	_	_	μs	
	Wait time after P-bit setting*1*4	Z	150	_	200	μs	
	Wait time after P-bit clear*1	α	10	_	_	μs	
	Wait time after PSU-bit clear*1	β	10	_	_	μs	
	Wait time after PV-bit setting*1	γ	4	_	_	μs	
	Wait time after dummy write*1	ε	2	_	_	μs	
	Wait time after PV-bit clear*1	η	4	_	_	μs	
	Maximum programming count*1*4*5	N	_	_	1000	Times	z = 200 μs
Erase	Wait time after SWE-bit setting*1	х	10	_	_	μs	
	Wait time after ESU-bit setting*1	у	200	_	_	μs	
	Wait time after E-bit setting*1 *6	z	5	_	10	ms	
	Wait time after E-bit clear*1	α	10	_	_	μs	
	Wait time after ESU-bit clear*1	β	10	_	_	μs	
	Wait time after EV-bit setting*1	γ	20	_	_	μs	
	Wait time after dummy write*1	ε	2	_	_	μs	
	Wait time after EV-bit clear*1		5	_	_	μs	
	Maximum erase count*1 *6 *7	N			120	Times	z = 10 ms

Notes: 1. Set the times according to the program/erase algorithms.

2. Programming time per 32 bytes (Shows the total period for which the P-bit in FLMCR1 is set. It does not include the programming verification time.)

- 3. Block erase time (Shows the total period for which the E-bit in FLMCR1 is set. It does not include the erase verification time.)
- 4. Maximum programming time (tP (max)) $tP(max) = wait time after P-bit setting (z) \times maximum programming count (N)$
- 5. Number of times when the wait time after P-bit setting (z) = $200 \mu s$. The number of writes should be set according to the actual set value of (z) to allow programming within the maximum programming time (tP(max)).
- Maximum erase time (tE (max)) tE(max) = wait time after E-bit setting (z) × maximum erase count (N)
- 7. Number of times when the wait time after E-bit setting (z) = 10 ms.

 The number of erases should be set according to the actual set value of (z) to allow erasing within the maximum erase time (tE(max)).
- 8. Minimum number of times for which all characteristics are guaranteed after rewriting (Guarantee range is 1 to minimum value).
- 9. Reference value for 25°C (as a guideline, rewriting should normally function up to this value).
- 10. Data retention characteristic when rewriting is performed within the specification range, including the minimum value.

26.2.7 Usage Note

- (1) The F-ZTAT and mask ROM versions have been confirmed as fully meeting the reference values for electrical characteristics shown in this manual. However, actual performance figures, operating margins, noise margins, and other properties may vary due to differences in the manufacturing process, on-chip ROM, layout patterns, etc.
 - When system evaluation testing is carried out using the F-ZTAT version, the same evaluation tests should also be conducted for the mask ROM version when changing over to that version.
- (2) On-chip power supply step-down circuit
 - The H8S/2148 F-ZTAT does not incorporate an internal power supply step-down circuit. When changing over to F-ZTAT versions or mask ROM versions incorporating an internal step-down circuit, the VCC2 pin has the same pin location as the VCL pin in a step-down circuit.

Therefore, note that the circuit patterns differ between these two types of products.



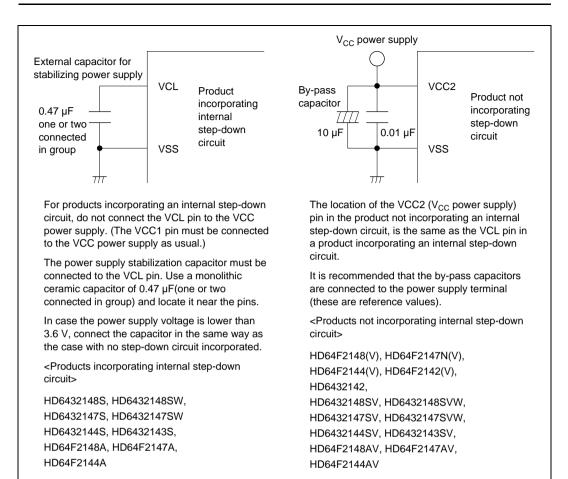


Figure 26.3 Connection of External Capacitor (Mask ROM Type Incorporating Step-Down Circuit and Product Not Incorporating Step-Down Circuit)

26.3 Electrical Characteristics of H8S/2148 F-ZTAT (A-mask version), H8S/2147 F-ZTAT (A-mask version), and Mask ROM Versions of H8S/2148 and H8S/2147

26.3.1 Absolute Maximum Ratings

Table 26.16 lists the absolute maximum ratings.

Table 26.16 Absolute Maximum Ratings

Item	Symbol	Value	Unit
Power supply voltage*1	V _{cc}	-0.3 to +7.0	V
Input/output buffer power supply (power supply for the port A)	V _{cc} B	-0.3 to +7.0	V
Power supply voltage*1 (3 V version)	V _{cc}	-0.3 to +4.3	V
Power supply voltage*2 (VCL pin)	V _{CL}	-0.3 to +4.3	V
Input voltage (except ports 6, 7, and A)	V _{in}	-0.3 to V_{cc} +0.3	V
Input voltage (CIN input not selected for port 6)	V _{in}	-0.3 to V_{cc} +0.3	V
Input voltage (CIN input not selected for port A)	V_{in}	-0.3 to V _{cc} B +0.3	V
Input voltage (CIN input selected for port 6)	V_{in}	–0.3 V to lower of voltages V $_{\rm cc}$ +0.3 and AV $_{\rm cc}$ +0.3	V
Input voltage (CIN input selected for port A)	V _{in}	-0.3 V to lower of voltages $V_{\rm cc}B$ +0.3 and $AV_{\rm cc}$ +0.3	V
Input voltage (port 7)	V _{in}	-0.3 to AV _{cc} +0.3	V
Reference supply voltage	AV_{ref}	-0.3 to AV _{cc} +0.3	V
Analog power supply voltage	AV _{cc}	-0.3 to +7.0	V
Analog power supply voltage (3 V version)	AV _{cc}	-0.3 to +4.3	V
Analog input voltage	$V_{\scriptscriptstyle AN}$	-0.3 to AV _{cc} +0.3	V
Operating temperature	T_{opr}	Regular specifications: -20 to +75	°C
		Wide-range specifications: -40 to +85	°C



Item	Symbol	Value	Unit
Operating temperature (flash	T _{opr}	Regular specifications: -20 to +75	°C
memory programming/erasing)		Wide-range specifications: -40 to +85	°C
Storage temperature	T _{stg}	-55 to +125	°C

Caution:

- Permanent damage to the chip may result if absolute maximum ratings are exceeded.
- 2. Never apply more than 7.0 V to any of the pins of the 5- or 4-V version or 4.3 V to any of the pins (except port A) of the 3-V version
- Notes: 1. Power supply voltage for VCC1 pin Never exceed the maximum rating of V_{cL} in the low-power version (3-V version) because both the VCC1 and VCL pins are connected to the V_{cc} power supply.
 - It is an operating power supply voltage pin on the chip.
 Never apply power supply voltage to the VCL pin in the 5- or 4-V version.
 Always connect an external capacitor between the VCL pin and ground for internal voltage stabilization.

26.3.2 DC Characteristics

Table 26.17 lists the DC characteristics. Permitted output current values and bus drive characteristics are shown in tables 26.18 and 26.19, respectively.

Table 26.17 DC Characteristics (1)

Conditions:
$$V_{cc} = 5.0 \text{ V} \pm 10\%$$
, $V_{cc}B = 5.0 \text{ V} \pm 10\%$, $AV_{cc}^{*1} = 5.0 \text{ V} \pm 10\%$, $AV_{ref}^{*1} = 4.5 \text{ V}$ to AV_{cc} , $V_{ss} = AV_{ss}^{*1} = 0 \text{ V}$, $T_{a} = -20 \text{ to } +75^{\circ}\text{C}$ (regular specifications), $T_{a} = -40 \text{ to } +85^{\circ}\text{C}$ (wide-range specifications)

Item			Symbol	Min	Тур	Max	Unit	Test Conditions
	P67 to P60(KWUL =	(1)	V _T -	1.0	_	_	V	
voltage	00)*2*6, KIN15 to KIN8*7*8,		V _T ⁺	_	_	$V_{cc} \times 0.7$ $V_{cc} B \times 0.7$	_	
	IRQ2 to IRQ0*3, IRQ5 to IRQ3		$V_T^+ - V_T^-$	0.4	_	_		
Schmitt	P67 to P60	_	V _T	$V_{cc} \times 0.3$	_	_	V	
trigger input voltage	(KWUL = 01)		V _T ⁺	_	_	$V_{cc} \times 0.7$	=	
(in level		_	$V_T^+ - V_T^-$	$\rm V_{cc} \times 0.05$	_	_		
switching)*6	P67 to P60		V _T	$V_{\text{cc}} \times 0.4$	_	_	_	
	(KWUL = 10)		V_{T}^{+}	_	_	$V_{cc} \times 0.8$	_	
		_	$V_T^+ - V_T^-$	$V_{\text{cc}} \times 0.03$	_	_	_	
	P67 to P60		V _T	$V_{cc} \times 0.45$	_	_		
	(KWUL = 11)		$V_{T}^{}^{T}}$	_	_	$V_{cc} \times 0.9$	_	
			$V_{T}^{+} - V_{T}^{-}$	0.05	_	_		
Input high voltage	RES, STBY, NMI, MD1, MD0	(2)	V _{IH}	V _{cc} -0.7	_	V _{cc} +0.3	V	
	EXTAL	_		$V_{cc} \times 0.7$	_	V _{cc} +0.3	=	
	PA7 to PA0*7	_		$V_{cc}B \times 0.7$	_	V _{cc} B +0.3	_	
	Port 7			2.0	_	AV _{cc} +0.3	_	
	Input pins except (1) and (2) above			2.0	_	V _{cc} +0.3		

Item			Symbol	Min	Тур	Max	Unit	Test Conditions
Input low voltage	RES, STBY, MD1, MD0	(3)	V _{IL}	-0.3	_	0.5	V	
	PA7 to PA0	-		-0.3	_	1.0		
	NMI, EXTAL, input pins except (1) and (3) above		_	-0.3	_	0.8		
Output high voltage	All output pins (except P97, and		V _{OH}	V _{cc} -0.5 V _{cc} B -0.5	_	_	V	$I_{OH} = -200 \mu A$
	P52*4)*5 *8			3.5	_	_	V	$I_{OH} = -1 \text{ mA}$
	P97, P52*4			2.0	_	_	V	$I_{OH} = -200 \mu A$
Output low voltage	All output pins (except RESO)*5		V _{oL}	_	_	0.4	V	I _{oL} = 1.6 mA
	Ports 1 to 3		_	_	_	1.0	V	I _{oL} = 10 mA
	RESO			_	_	0.4	V	I _{oL} = 2.6 mA
Input	RES		I _{in}	_	_	10.0	μΑ	$V_{in} = 0.5 \text{ to}$
leakage current	STBY, NMI, MD1, MD0			_	_	1.0	μΑ	[−] V _{cc} –0.5 V
	Port 7			_	_	1.0	μΑ	$V_{in} = 0.5 \text{ to}$ AV _{CC} -0.5 V
Three-state leakage current (off state)	Ports 1 to 6, 8, 9, A**, B		I _{TSI}	-	_	1.0	μА	$V_{in} = 0.5 \text{ to}$ $V_{cc} - 0.5 \text{ V},$ $V_{in} = 0.5 \text{ to}$ $V_{cc}B - 0.5 \text{ V}$
Input	Ports 1 to 3		-I _P	30	_	300	μA	$V_{in} = 0 V$
pull-up MOS	Ports A*8, B, Port 6 (P6PUE = 0)			60	_	600	μΑ	_
	Port 6 (P6PUE = 1)		_	15	_	200	μΑ	_

Item		Symbol	Min	Тур	Max	Unit	Test Conditions
Input	RES (4) C _{in}	_	_	80	pF	$V_{in} = 0 V$,
capacitance	NMI		_	_	50	pF	⁻f = 1 MHz, -T _a = 25°C
	P52, P97, P42, P86 PA7 to PA2		_	_	20	pF	– I _a – 23 O
	Input pins except (4) above		_	_	15	pF	_
Current	Normal operation	I _{cc}	_	55	70	mA	f = 20 MHz
dissipation*9	Sleep mode		_	36	55	mA	f = 20 MHz
	Standby mode*10		_	1.0	5.0	μΑ	T _a ≤ 50°C
			_	_	20.0	μΑ	50°C < T _a
Analog power	During A/D, D/A conversion	Al _{cc}	_	1.2	2.0	mA	
supply current	Idle		_	0.01	5.0	μΑ	AV _{cc} = 2.0 V to 5.5 V
Reference	During A/D convers	sion Al _{ref}	_	0.5	1.0	mA	
power supply current	During A/D, D/A conversion		_	2.0	5.0	mA	_
	Idle		_	0.01	5.0	μΑ	$AV_{ref} = 2.0 V$ to AV_{CC}
Analog power	Analog power supply voltage*1		4.5	_	5.5	V	Operating
			2.0	_	5.5	V	Idle/not used
RAM standb	RAM standby voltage		2.0	_	_	V	

Notes: 1. Do not leave the AVCC, AV_{ref}, and AVSS pins open even if the A/D converter and D/A converter are not used.

Even if the A/D converter and D/A converter are not used, apply a value in the range 2.0 V to 5.5 V to AVCC and AV_{ref} pins by connection to the power supply (V_{cc}), or some other method. Ensure that AV_{ref} \leq AV_{cc}.

- 2. P67 to P60 include supporting module inputs multiplexed on those pins.
- 3. $\overline{\text{IRQ2}}$ includes the $\overline{\text{ADTRG}}$ signal multiplexed on that pin.
- In the H8S/2148 Group, P52/SCK0/SCL0 and P97/SDA0 are NMOS push-pull outputs. An external pull-up resistor is necessary to provide high-level output from SCL0 and SDA0 (ICE = 1).

In the H8S/2148 Group, P52/SCK0 and P97 (ICE = 0) high levels are driven by NMOS.

5. When IICS = 0, ICE = 0, and KBIOE = 0. Low-level output when the bus drive function is selected is determined separately.



- 6. The upper limit of the port 6 applied voltage is V_{cc} +0.3 V when CIN input is not selected, and the lower of V_{cc} +0.3 V and AV_{cc} +0.3 V when CIN input is selected. When a pin is in output mode, the output voltage is equivalent to the applied voltage.
- 7. The upper limit of the port A applied voltage is $V_{cc}B$ +0.3 V when CIN input is not selected, and the lower of $V_{cc}B$ +0.3 V and AV_{cc} +0.3 V when CIN input is selected. When a pin is in output mode, the output voltage is equivalent to the applied voltage.
- 8. Port A characteristics depends on $V_{cc}B$ (or on V_{cc} when other pins are in output mode).
- 9. Current dissipation values are for V_{IH} min = V_{cc} -0.2 V, V_{cc} B -0.2 V, and V_{IL} max = 0.2 V with all output pins unloaded and the on-chip pull-up MOSs in the off state.
- 10. The values are for $V_{RAM} \le V_{CC} < 4.5 V$, $V_{IH} min = V_{CC} 0.2 V$, $V_{CC} B 0.2 V$, and $V_{IL} max = 0.2 V$.

Table 26.17 DC Characteristics (2)

Conditions: $V_{cc} = 4.0 \text{ V to } 5.5 \text{ V}, V_{cc}B = 4.0 \text{ V to } 5.5 \text{ V}, AV_{cc}^{*1} = 4.0 \text{ V to } 5.5 \text{ V}, AV_{ref}^{*1} = 4.0 \text{ V to } AV_{cc}, V_{ss} = AV_{ss}^{*1} = 0 \text{ V}, T_{a} = -20 \text{ to } +75^{\circ}\text{C} \text{ (regular specifications)}, T_{a} = -40 \text{ to } +85^{\circ}\text{C} \text{ (wide-range specifications)}$

Item			Symbol	Min	Тур	Max	Unit	Test Conditions		
Schmitt	P67 to	(1)	V _T -	1.0	_	_	V	$V_{cc} = 4.5 \text{ V to}$		
trigger input voltage	P60(KWUL = 00)*2 *6,		$V_{T}^{ +}$	_	_	$V_{cc} \times 0.7$	V	5.5 V, V _{cc} B = 4.5 V		
	KIN15 to					$V_{cc}B \times 0.7$		_to 5.5 V		
	KIN8*7 *8, IRQ2 to IRQ0*3,		$V_{T}^{+} - V_{T}^{-}$		_	_	V			
	IRQ5 to IRQ3		V _T	0.8	_	_	V	$V_{cc} = 4.0 \text{ V to}$		
			$V_{T}^{}^{T}}$	_	_	$V_{cc} \times 0.7$	V	4.5 V, V _{cc} B = 4.0 V		
						$V_{cc}B \times 0.7$		to 4.5 V		
			$V_{T}^{+} - V_{T}^{-}$	0.3	_	_	V			
Schmitt	P67 to P60 (KWUL = 01)	_	V _T	$V_{\text{cc}} \times 0.3$	_	_	V	$V_{cc} = 4.0 \text{ V to}$		
			$V_{T}^{}^{T}}$	_	_	$V_{cc} \times 0.7$	=	5.5 V		
voltage (in level			$V_T^+ - V_T^-$	$V_{cc} \times 0.05$	_	_	_			
	P67 to P60 (KWUL = 10)	-	V	$V_{cc} \times 0.4$	_	_	=			
			V _T ⁺	_	_	$V_{cc} \times 0.8$	=			
			$V_{T}^{+} - V_{T}^{-}$	$V_{cc} \times 0.03$		_	_			
	P67 to P60		$V_{\scriptscriptstyle T}^{\;-}$	$V_{\text{cc}} \times 0.45$	_	_				
	(KWUL = 11)		$V_{T}^{}^{T}}$	_	_	$V_{cc} \times 0.9$				
			$V_{T}^{+} - V_{T}^{-}$	0.05	_	_	_			
Input high voltage	RES, STBY, NMI, MD1, MD0	(2)	V _{IH}	V _{cc} -0.7	_	V _{cc} +0.3	V			
	EXTAL	$\frac{V_{cc} \times 0.7 - V_{cc} + 0.3 V}{V_{cc} \times 0.7 - V_{cc} + 0.3 V}$	V	_ ,						
	PA7 to PA0*7	=		$V_{cc}B \times 0.7$	_	V _{cc} B + 0.3	V	_ ,		
	Port 7	-				2.0	_	AV _{cc} +0.3	V	<u> </u>
	Input pins except (1) and (2) above			2.0	_	V _{cc} +0.3	V	_		

Item			Symbol	Min	Тур	Max	Unit	Test Conditions
Input low voltage	RES, STBY, MD1, MD0	(3)	V _{IL}	-0.3	_	0.5	V	
	PA7 to PA0	=		-0.3	_	1.0	V	$V_{cc}B = 4.5 \text{ V}$ to 5.5 V
				-0.3	_	0.8	V	$V_{cc}B = 4.0 \text{ V}$ to 4.5 V
	NMI, EXTAL, input pins except (1) and (3) above			-0.3	_	0.8	V	
Output high voltage	(except P97,		V _{OH}	V _{cc} -0.5 V _{cc} B -0.5	_	_	V	$I_{OH} = -200 \mu A$
	and P52*4)*5*8			3.5	_	_	V	$I_{OH} = -1 \text{ mA},$ $V_{CC} = 4.5 \text{ V to}$ 5.5 V, $V_{CC}B = 4.5 \text{ V}$ to 5.5 V
				3.0	_	_	V	$I_{OH} = -1 \text{ mA},$ $V_{CC} = 4.0 \text{ V to}$ 4.5 V, $V_{CC}B = 4.0 \text{ V}$ o 4.5 V
	P97, P52*4		_	1.5	_		V	$I_{OH} = -200 \mu A$
Output low voltage	All output pins (except RESO)*5		V _{oL}	_	_	0.4	V	I _{OL} = 1.6 mA
	Ports 1 to 3			_	_	1.0	V	$I_{OL} = 10 \text{ mA}$
	RESO		_	_	_	0.4	V	I _{OL} = 2.6 mA
Input	RES		I _{in}	_	_	10.0	μA	$V_{in} = 0.5 \text{ to}$
leakage current	STBY, NMI, MD1, MD0		_	_	_	1.0	μΑ	¯V _{cc} –0.5 V
	Port 7		_	_	_	1.0	μΑ	$V_{in} = 0.5 \text{ to}$ AV _{CC} -0.5 V
Three-state leakage current (off state)	Ports 1 to 6, 8, 9, A**, B		I _{TSI}	_	_	1.0	μА	$V_{in} = 0.5 \text{ to}$ $V_{cc} - 0.5 \text{ V},$ $V_{in} = 0.5 \text{ to}$ $V_{cc}B - 0.5 \text{ V}$

Item		Symbol	Min	Тур	Max	Unit	Test Conditions
Input	Ports 1 to 3	-I _P	30	_	300	μΑ	$V_{in} = 0 V$,
pull-up MOS current	Ports A*8, B Port 6 (P6PUE = 0)	_	60	_	600	μA	$^{-}$ V _{cc} = 4.5 V to 5.5 V, V _{cc} B = 4.5 V to 5.5 V
	Port 6 (P6PUE = 1)	_	15	_	200	_	10 0.3 V
	Ports 1 to 3	_	20	_	200	μΑ	$V_{in} = 0 V$,
	Ports A*8, B Port 6 (P6PUE = 0)	_	40	_	500	μA	$^{-}$ V _{cc} = 4.0 V to 4.5 V, V _{cc} B = 4.0 V to 4.5 V
	Port 6 (P6PUE = 1)	_	10	_	150	_	10 4.5 V
Input	RES (4)	C _{in}	_	_	80	pF	$V_{in} = 0 V$,
capacitance	NMI		_	_	50	pF	−f = 1 MHz, – T _a = 25°C
	P52, P97, P42, P86, PA7 to PA2	=	_	_	20	pF	- I _a = 25 C
	Input pins except (4) above	_	_	_	15	pF	_
Current	Normal operation	I _{cc}	_	45	58	mΑ	f = 16 MHz
dissipation*9	Olcop mode	_	_	30	46	mΑ	f = 16 MHz
	Standby mode*10	_	_	1.0	5.0	μΑ	$T_a \le 50$ °C
			_	_	20.0	_	50°C < T _a
Analog power	During A/D, D/A conversion	Al _{cc}	_	1.2	2.0	mA	
supply current	Idle	_	_	0.01	5.0	μΑ	$AV_{cc} = 2.0 \text{ V}$ to 5.5 V
Reference	During A/D conversion	Al _{ref}	_	0.5	1.0	mA	
power supply current	During A/D, D/A conversion		_	2.0	5.0	mA	_
Julion	Idle	_	_	0.01	5.0	μΑ	$AV_{ref} = 2.0 V$ to AV_{CC}
Analog power	er supply voltage*1	AV_{cc}	4.0	_	5.5	V	Operating
			2.0	_	5.5	V	Idle/not used
RAM standb	y voltage	$V_{\scriptscriptstyle{RAM}}$	2.0			V	

Notes: 1. Do not leave the AVCC, AV_{ref}, and AVSS pins open even if the A/D converter and D/A converter are not used.

Even if the A/D converter and D/A converter are not used, apply a value in the range 2.0 V to 5.5 V to AVCC and AV_{ref} pins by connection to the power supply (V_{cc}), or some other method. Ensure that AV_{ref} \leq AV_{cc}.

- 2. P67 to P60 include supporting module inputs multiplexed on those pins.
- 3. IRQ2 includes the ADTRG signal multiplexed on that pin.
- In the H8S/2148 Group, P52/SCK0/SCL0 and P97/SDA0 are NMOS push-pull outputs.
 An external pull-up resistor is necessary to provide high-level output from SCL0 and SDA0 (ICE = 1).

In the H8S/2148 Group, P52/SCK0 and P97 (ICE = 0) high levels are driven by NMOS.

- 5. When IICS = 0, ICE = 0, and KBIOE = 0. Low-level output when the bus drive function is selected is determined separately.
- 6. The upper limit of the port 6 applied voltage is V_{cc} +0.3 V when CIN input is not selected, and the lower of V_{cc} +0.3 V and AV_{cc} +0.3 V when CIN input is selected. When a pin is in output mode, the output voltage is equivalent to the applied voltage.
- 7. The upper limit of the port A applied voltage is $V_{cc}B + 0.3 \text{ V}$ when CIN input is not selected, and the lower of $V_{cc}B + 0.3 \text{ V}$ and $AV_{cc} + 0.3 \text{ V}$ when CIN input is selected. When a pin is in output mode, the output voltage is equivalent to the applied voltage.
- 8. The port A characteristics depend on $V_{cc}B$, and the other pins characteristics depend on V_{cc} .
- 9. Current dissipation values are for V_{IH} min = V_{CC} -0.2 V, V_{CC} B -0.2 V, and V_{IL} max = 0.2 V with all output pins unloaded and the on-chip pull-up MOSs in the off state.
- 10. The values are for $V_{RAM} \le V_{CC} < 4.0 \text{ V}$, $V_{IH} \text{ min} = V_{CC} 0.2 \text{ V}$, $V_{CC}B 0.2 \text{ V}$, and $V_{IL} \text{ max} = 0.2 \text{ V}$.

Table 26.17 DC Characteristics (3)

 $\begin{array}{ll} \text{Conditions:} & V_{_{CC}} = 2.7 \text{ V to } 3.6 \text{ V}^{*_{11}}, V_{_{CC}} B = 2.7 \text{ V to } 5.5 \text{ V}, \text{ AV}_{_{CC}}^{\quad *_{1}} = 2.7 \text{ V to } 3.6 \text{ V}, \\ & AV_{_{ref}} = 2.7 \text{ V to } 3.6 \text{ V}, V_{_{SS}} = AV_{_{SS}}^{\quad *_{1}} = 0 \text{ V}, T_{_{a}} = -20 \text{ to } +75^{\circ}\text{C} \\ \end{array}$

Item			Symbol	Min	Тур	Max	Unit	Test Conditions
Schmitt trigger input	P67 to P60(KWUL =	(1)	V _T -	$\begin{array}{c} V_{cc} \times 0.2 \\ V_{cc} B \times 0.2 \end{array}$	_	_	V	
voltage	00)*2*6, KIN15 to KIN8*7*8,		V _T ⁺	_	_	$V_{cc} \times 0.7$ $V_{cc} B \times 0.7$	V	_
	ĪRQ2 to ĪRQ0*3,		$V_T^+ - V_T^-$	$V_{cc} \times 0.05$	_	_	V	_
	IRQ5 to IRQ3			$V_{cc}B \times 0.05$				
Schmitt	P67 to P60	=	V _T -	$V_{cc} \times 0.3$	_	_	V	
trigger input voltage	(KWUL = 01)		$V_{T}^{}^{T}}$	_	_	$V_{cc} \times 0.7$	_	
(in level			$V_T^+ - V_T^-$	$V_{cc} \times 0.05$	_	_		
swiching)*6	P67 to P60 (KWUL = 10)	_	V _T	$V_{cc} \times 0.4$	_	_	_	
			$V_{T}^{}^{T}}$	_	_	$V_{\text{cc}} \times 0.8$	_	
		_	$V_{\scriptscriptstyle T}^{\;\scriptscriptstyle +} - V_{\scriptscriptstyle T}^{\;\scriptscriptstyle -}$	$\rm V_{cc} \times 0.03$	_	_	_	
	P67 to P60		V_{T}^{-}	$V_{cc} \times 0.45$	_	_	_	
	(KWUL = 11)		V_{T}^{+}	_	_	$V_{cc} \times 0.9$	_	
			$V_T^+ - V_T^-$	0.05	_	_		
Input high voltage	RES, STBY, NMI, MD1, MD0	(2)	V_{IH}	$V_{cc} \times 0.9$	_	V _{cc} +0.3	V	
	EXTAL	=		$V_{cc} \times 0.7$	_	V _{cc} +0.3	V	-
	PA7 to PA0*7	=		$V_{cc}B \times 0.7$	_	V _{cc} B +0.3	V	-
	Port 7	=		$V_{cc} \times 0.7$	_	AV _{cc} +0.3	V	-
	Input pins except (1) and (2) above			$V_{cc} \times 0.7$	_	V _{cc} +0.3	V	-

Item			Symbol	Min	Тур	Max	Unit	Test Conditions
Input low voltage	RES, STBY, MD1, MD0	(3)	V _{IL}	-0.3	_	$V_{cc} \times 0.1$	V	
	PA7 to PA0	-		-0.3	_	$V_{cc}B \times 0.2$	V	V _{cc} B = 2.7 V to 4.0 V
						0.8	V	V _{cc} B = 4.0 V to 5.5 V
	NMI, EXTAL, input pins except (1) and (3) above		_	-0.3	_	$V_{cc} \times 0.2$	V	V _{cc} = 2.7 V to 3.6 V
Output high voltage	(except P97,		V _{OH}	V _{cc} -0.5 V _{cc} B -0.5	_	_	V	$I_{OH} = -200 \mu A$
	and P52*4)*5*8			V _{cc} -1.0 V _{cc} B -1.0	_	_	V	$I_{OH} = -1 \text{ mA}$ $(V_{CC} = 2.7 \text{ V}$ to 3.6 V, $V_{CC}B = 2.7 \text{ V}$ to 4.0 V)
	P97, P52*4			0.5	_	_	V	$I_{OH} = -200 \mu A$
Output low voltage	All output pins (except RESO)*5	i	V _{oL}	_	_	0.4	V	I _{oL} = 1.6 mA
	Ports 1 to 3		_	_	_	1.0	V	I _{oL} = 5 mA
	RESO		_	_	_	0.4	V	I _{oL} = 1.6 mA
Input	RES		I _{in}	_	_	10.0	μΑ	$V_{in} = 0.5 \text{ to}$
leakage current	STBY, NMI, MD1, MD0			_	_	1.0	μΑ	[−] V _{cc} −0.5 V
	Port 7			_	_	1.0	μΑ	$V_{in} = 0.5 \text{ to}$ AV _{CC} -0.5 V
Three-state leakage current (off state)	Ports 1 to 6, 8, 9, A**, B		I _{TSI}	_	_	1.0	μΑ	$V_{in} = 0.5 \text{ to}$ $V_{cc} - 0.5 \text{ V},$ $V_{in} = 0.5 \text{ to}$ $V_{cc}B - 0.5 \text{ V}$

Item		Symbol	Min	Тур	Max	Unit	Test Conditions
Input pull-	Ports 1 to 3	-I _P	5	_	150	μA	$V_{in} = 0 V,$
up MOS current	Ports A*8, B Port 6 (P6PUE = 0)		30	_	300	μΑ	$V_{cc} = 2.7 \text{ V to}$ 3.6 V, $V_{cc}B = 2.7 \text{ V}$ -to 3.6 V
	Port 6 (P6PUE = 1)		3	_	100	μΑ	-10 3.0 V
Input capacitance	RES (4) C _{in}	_	_	80	pF	$V_{in} = 0 V,$ f = 1 MHz,
	NMI		_	_	50	pF	$T_a = 25^{\circ}C$
	P52, P97, P42, P86, PA7 to PA2		_	_	20	pF	_
	Input pins except (4) above		_	_	15	pF	_
Current	Normal operation	I _{cc}	_	30	40	mA	f = 10 MHz
dissipation*	Sleep mode		_	20	32	mA	f = 10 MHz
	Standby mode*10		_	1.0	5.0	μA	T _a ≤ 50°C
			_	_	20.0	μA	50°C < T _a
Analog power	During A/D, D/A conversion	Al _{cc}	_	1.2	2.0	mA	
supply current	Idle		_	0.01	5.0	μΑ	$AV_{cc} = 2.0 \text{ V}$ to 3.6 V
Reference	During A/D conversi	on Al _{ref}	_	0.5	1.0	mA	
power supply current	During A/D, D/A conversion		_	2.0	5.0	mA	_
	Idle		_	0.01	5.0	μΑ	$AV_{ref} = 2.0 \text{ V}$ to AV_{CC}
Analog power	er supply voltage*1	AV _{cc}	2.7		3.6	V	Operating
			2.0	_	3.6	V	Idle/not used
RAM standby voltage		$V_{\scriptscriptstyle{RAM}}$	2.0	_	_	V	

Notes: 1. Do not leave the AVCC, AV_{ref}, and AVSS pins open even if the A/D converter and D/A converter are not used.

Even if the A/D converter and D/A converter are not used, apply a value in the range 2.0 V to 3.6 V to AVCC and AV_{ref} pins by connection to the power supply (V_{cc}), or some other method. Ensure that AV_{ref} \leq AV_{cc}.

- 2. P67 to P60 include supporting module inputs multiplexed on those pins.
- 3. IRQ2 includes the ADTRG signal multiplexed on that pin.



- In the H8S/2148 Group, P52/SCK0/SCL0 and P97/SDA0 are NMOS push-pull outputs.
 An external pull-up resistor is necessary to provide high-level output from SCL0 and SDA0 (ICE = 1).
 - In the H8S/2148 Group, P52/SCK0 and P97 (ICE = 0) high levels are driven by NMOS.
- 5. When IICS = 0, ICE = 0, and KBIOE = 0. Low-level output when the bus drive function is selected is determined separately.
- 6. The upper limit of the port 6 applied voltage is V_{cc} +0.3 V when CIN input is not selected, and the lower of V_{cc} +0.3 V and AV_{cc} +0.3 V when CIN input is selected. When a pin is in output mode, the output voltage is equivalent to the applied voltage.
- 7. The upper limit of the port A applied voltage is $V_{cc}B + 0.3 \text{ V}$ when CIN input is not selected, and the lower of $V_{cc}B + 0.3 \text{ V}$ and $AV_{cc} + 0.3 \text{ V}$ when CIN input is selected. When a pin is in output mode, the output voltage is equivalent to the applied voltage.
- The port A characteristics depend on V_{cc}B, and the other pins characteristics depend on V_{cc}.
- 9. Current dissipation values are for V_{IH} min = V_{CC} -0.2 V, V_{CC} B -0.2 V, and V_{IL} max = 0.2 V with all output pins unloaded and the on-chip pull-up MOSs in the off state.
- 10. The values are for $V_{RAM} \le V_{CC} < 2.7 \text{ V}$, $V_{IH} \text{ min} = V_{CC} 0.2 \text{ V}$, $V_{CC}B 0.2 \text{ V}$, and $V_{IL} \text{ max} = 0.2 \text{ V}$.
- 11. For flash memory program/erase operations, the applicable ranges are $\rm V_{cc}$ = 3.0 V to 3.6 V.

Table 26.18 Permissible Output Currents

Conditions: $V_{cc} = 4.0 \text{ V}$ to 5.5 V, $V_{cc}B = 4.0 \text{ V}$ to 5.5 V, $V_{ss} = 0 \text{ V}$, $Ta = -20 \text{ to } +75 ^{\circ}\text{C}$ (regular)

specifications), $T_a = -40$ to +85°C (wide-range specifications)

Item		Symbol	Min	Тур	Max	Unit
Permissible output low current (per pin)	SCL1, SCL0, SDA1, SDA0, PS2AC to PS2CC, PS2AD to PS2CD, PA7 to PA4 (bus drive function selected)	I _{OL}	_	_	20	mA
	Ports 1, 2, 3		_	_	10	mA
	RESO	_	_	_	3	mA
	Other output pins	_	_	_	2	mA
Permissible output	Total of ports 1, 2, and 3	\sum I _{OL}	_	_	80	mA
low current (total)	Total of all output pins, including the above		_	_	120	mA
Permissible output high current (per pin)	All output pins	-I _{OH}	_	_	2	mA
Permissible output high current (total)	Total of all output pins	$\Sigma - \mathbf{I}_{OH}$	_	_	40	mA

Conditions: $V_{cc} = 2.7 \text{ V}$ to 3.6 V, $V_{cc}B = 2.7 \text{ V}$ to 5.5 V, $V_{ss} = 0 \text{ V}$, $T_a = -20 \text{ to } +75^{\circ}\text{C}$

Item		Symbol	Min	Тур	Max	Unit
Permissible output low current (per pin)	SCL1, SCL0, SDA1, SDA0, PS2AC to PS2CC, PS2AD to PS2CD, PA7 to PA4 (bus drive function selected)	I _{OL}	_	_	10	mA
	Ports 1, 2, 3		_	_	2	mA
	RESO		_	_	1	mA
	Other output pins	_	_	_	1	mA
Permissible output	Total of ports 1, 2, and 3	\sum I _{ol}	_	_	40	mA
low current (total)	Total of all output pins, including the above	_	_	_	60	mA
Permissible output high current (per pin)	All output pins	-I _{OH}	_	_	2	mA
Permissible output high current (total)	Total of all output pins	$\Sigma - \mathbf{I}_{OH}$	_	_	30	mA

Notes: 1. To protect chip reliability, do not exceed the output current values in table 26.18.

2. When driving a Darlington pair or LED, always insert a current-limiting resistor in the output line, as show in figures 26.1 and 26.2.

Table 26.19 Bus Drive Characteristics

Conditions: $V_{cc} = 4.0 \text{ V}$ to 5.5 V, $V_{cc} = 2.7 \text{ V}$ to 3.6 V (3 V version), $V_{ss} = 0 \text{ V}$

Applicable Pins: SCL1, SCL0, SDA1, SDA0 (bus drive function selected)

Item	Symbol	Min	Тур	Max	Unit	Test Conditions
Schmitt trigger	V _T -	$V_{cc} \times 0.3$	_	_	V	
input voltage	V _T ⁺	_	_	$V_{cc} \times 0.7$	_	
	$V_T^+ - V_T^-$	$V_{cc} \times 0.05$	_	_		
Input high voltage	V _{IH}	$V_{cc} \times 0.7$	_	V _{cc} +0.5	V	
Input low voltage	V _{IL}	-0.5	_	$V_{cc} \times 0.3$	_	
Output low voltage	V _{oL}	_	_	0.8	V	$I_{OL} = 16 \text{ mA},$ $V_{CC} = 4.5 \text{ V to } 5.5 \text{ V}$
		_	_	0.5		$I_{OL} = 8 \text{ mA}$
		_	_	0.4	_	I _{oL} = 3 mA
Input capacitance	C _{in}	_	_	20	pF	$V_{in} = 0 \text{ V, f} = 1 \text{ MHz,}$ $T_{a} = 25^{\circ}\text{C}$
Three-state leakage current (off state)	I _{TSI}	_	_	1.0	μΑ	$V_{in} = 0.5 \text{ to } V_{cc} - 0.5 \text{ V}$
SCL, SDA output fall time	t _{of}	20 + 0.1 Cb	_	250	ns	

Conditions: $V_{cc} = 4.0 \text{ V}$ to 5.5 V, $V_{cc} = 2.7 \text{ V}$ to 3.6 V (3 V version), $V_{cc}B = 2.7 \text{ V}$ to 5.5 V,

 $V_{ss} = 0 V$

Applicable Pins: PS2AC, PS2AD, PS2BC, PS2BD, PS2CC, PS2CD, PA7 to PA4 (bus drive

function selected)

Item	Symbol	Min	Тур	Max	Unit	Test Conditions
Output low voltage	V _{oL}	_	_	0.8	V	$I_{OL} = 16 \text{ mA},$ $V_{CC}B = 4.5 \text{ V to } 5.5 \text{ V}$
		_	_	0.5	_	$I_{OL} = 8 \text{ mA}$
		_	_	0.4	_	I _{oL} = 3 mA

26.3.3 AC Characteristics

Clock timing, control signal timing, bus timing, and timing of on-chip supporting modules list the following.

Figure 26.4 shows the test conditions for the AC characteristics.

(1) Clock Timing

Table 26.20 shows the clock timing. The clock timing specified here covers clock (ϕ) output and clock pulse generator (crystal) and external clock input (EXTAL pin) oscillation settling times. For details of external clock input (EXTAL pin and EXCL pin) timing, see section 24, Clock Pulse Generator.

Table 26.20 Clock Timing

Condition A: $V_{cc} = 5.0 \text{ V} \pm 10\%$, $V_{cc}B = 5.0 \text{ V} \pm 10\%$, $V_{ss} = 0 \text{ V}$, $\phi = 2 \text{ MHz}$ to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$ (regular specifications), $T_a = -40 \text{ to } +85^{\circ}\text{C}$ (wide-range specifications)

Condition B: $V_{cc} = 4.0 \text{ V to } 5.5 \text{ V}, \ V_{cc}B = 4.0 \text{ V to } 5.5 \text{ V}, \ V_{ss} = 0 \text{ V}, \ \phi = 2 \text{ MHz to maximum}$ operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$ (regular specifications), $T_a = -40 \text{ to } +85^{\circ}\text{C}$ (wide-range specifications)

Condition C: $V_{cc} = 2.7 \text{ V}$ to 3.6 V, $V_{cc}B = 2.7 \text{ V}$ to 5.5 V, $V_{ss} = 0 \text{ V}$, $\phi = 2 \text{ MHz}$ to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$

		Condition A		Condition B		Condition C				
		20 MHz		16 MHz		10 MHz			Test	
Item	Symbol	Min	Max	Min	Max	Min	Max	Unit	Conditions	
Clock cycle time	t _{cyc}	50	500	62.5	500	100	500	ns	Figure 26.5	
Clock high pulse width	t _{ch}	17	_	20	_	30	_	ns	Figure 26.5	
Clock low pulse width	t _{cL}	17	_	20	_	30	_	ns	_	
Clock rise time	t _{Cr}	_	8	_	10	_	20	ns	_	
Clock fall time	t _{Cf}	_	8	_	10	_	20	ns		
Oscillation settling time at reset (crystal)	t _{osc1}	10	_	10	_	20	_	ms	Figure 26.6	
Oscillation settling time in software standby (crystal)	t _{osc2}	8	_	8	_	8	_	ms	Figure 26.7	
External clock output stabilization delay time	t _{DEXT}	500	_	500	_	500	_	μs	_	

(2) Control Signal Timing

Table 26.21 shows the control signal timing. The only external interrupts that can operate on the subclock ($\phi = 32.768 \text{ kHz}$) are NMI and IRO0, 1, 2, 6, and 7.

Table 26.21 Control Signal Timing

Condition A: $V_{cc} = 5.0 \text{ V} \pm 10\%$, $V_{cc}B = 5.0 \text{ V} \pm 10\%$, $V_{ss} = 0 \text{ V}$, $\phi = 32.768 \text{ kHz}$, 2 MHz to maximum operating frequency, $T_a = -20$ to +75°C (regular specifications), $T_{a} = -40 \text{ to } +85^{\circ}\text{C}$ (wide-range specifications)

Condition B: $V_{cc} = 4.0 \text{ V}$ to 5.5 V, $V_{cc}B = 4.0 \text{ V}$ to 5.5 V, $V_{ss} = 0 \text{ V}$, $\phi = 32.768 \text{ kHz}$, 2 MHz to maximum operating frequency, $T_a = -20$ to +75°C (regular specifications), $T_a = -40 \text{ to } +85^{\circ}\text{C}$ (wide-range specifications)

Condition C: $V_{cc} = 2.7 \text{ V}$ to 3.6 V, $V_{cc}B = 2.7 \text{ V}$ to 5.5 V, $V_{ss} = 0 \text{ V}$, $\phi = 32.768 \text{ kHz}$, 2 MHz to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$

		Condition A		Cond	dition B	Condition C			
		20 MHz		16 MHz		10 MHz			Test
Item	Symbol	Min	Max	Min	Max	Min	Max	Unit	Conditions
RES setup time	t _{RESS}	200	_	200	_	300	_	ns	Figure 26.8
RES pulse width	t _{resw}	20	_	20	_	20	_	t _{cyc}	_
NMI setup time (NMI)	t _{NMIS}	150	_	150	_	250	_	ns	Figure 26.9
NMI hold time (NMI)	t _{nmih}	10	_	10	_	10	_		_
NMI pulse width (exiting software standby mode)	t _{nmiw}	200	_	200	_	200	_	ns	_
IRQ setup time (IRQ7 to IRQ0)	t _{IRQS}	150	_	150	_	250	_	ns	_
IRQ hold time (IRQ7 to IRQ0)	t _{IRQH}	10	_	10	_	10	_	ns	_
IRQ pulse width (IRQ7, IRQ6, IRQ2 to IRQ0) (exiting software standby mode)	t _{IRQW}	200	_	200	_	200	_	ns	_

(3) Bus Timing

Table 26.22 shows the bus timing. Operation in external expansion mode is not guaranteed when operating on the subclock ($\phi = 32.768 \text{ kHz}$).

Table 26.22 Bus Timing (1) (Nomal mode)

- Condition A: $V_{cc} = 5.0 \text{ V} \pm 10\%$, $V_{cc}B = 5.0 \text{ V} \pm 10\%$, $V_{ss} = 0 \text{ V}$, $\phi = 2 \text{ MHz}$ to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$ (regular specifications), $T_a = -40 \text{ to } +85^{\circ}\text{C}$ (wide-range specifications)
- Condition B: $V_{cc} = 4.0 \text{ V}$ to 5.5 V, $V_{cc}B = 4.0 \text{ V}$ to 5.5 V, $V_{ss} = 0 \text{ V}$, $\phi = 2 \text{ MHz}$ to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$ (regular specifications), $T_a = -40 \text{ to } +85^{\circ}\text{C}$ (wide-range specifications)
- Condition C: $V_{cc} = 2.7$ V to 3.6 V, $V_{cc}B = 2.7$ V to 5.5 V, $V_{ss} = 0$ V, $\phi = 2$ MHz to maximum operating frequency, $T_a = -20$ to +75°C

		Cond	ition A	Cond	ition B	Cond	ition C		
		20	MHz	16	MHz	10	MHz	•	Test
Item	Symbol	Min	Max	Min	Max	Min	Max	Unit	Conditions
Address delay time	t _{AD}	_	20	_	30	_	40	ns	Figure 26.10 to
Address setup time	t _{AS}	$0.5 \times t_{\text{cyc}} -15$	_	$0.5 \times t_{\text{cyc}}$ -20	_	$0.5 \times t_{\text{cyc}}$ –30	_	ns	figure 26.14
Address hold time	t _{AH}	$\begin{array}{l} 0.5 \times \\ t_{_{\text{cyc}}} - 10 \end{array}$	_	$0.5 \times t_{\rm cyc} -15$	_	$\begin{array}{c} 0.5 \times \\ t_{\mbox{\tiny cyc}} -20 \end{array}$	_	ns	_
CS delay time (IOS)	t _{CSD}	_	20	_	30	_	40	ns	
AS delay time	t _{ASD}	_	30	_	45	_	60	ns	-
RD delay time 1	t _{RSD1}	_	30	_	45	_	60	ns	-
RD delay time 2	t _{RSD2}	_	30	_	45	_	60	ns	-
Read data setup time	t _{RDS}	15	_	20	_	35	_	ns	_
Read data hold time	t _{RDH}	0	_	0	_	0	_	ns	-

		Cond	ition A	Cond	ition B	Cond	ition C		
		20	MHz	16	MHz	10	MHz		Test
Item	Symbol	Min	Max	Min	Max	Min	Max	Unit	Conditions
Read data access time 1	t _{ACC1}	_	1.0 × t _{cyc} –30	_	1.0 × t _{cyc} –40	_	1.0 × t _{cyc} –60	ns	Figure 26.10 to figure 26.14
Read data access time 2	t _{ACC2}	_	1.5 × t _{cyc} –25	_	1.5 × t _{cyc} –35	_	1.5 × t _{cyc} –50	ns	_
Read data access time 3	t _{ACC3}	_	2.0 × t _{cyc} –30	_	2.0 × t _{cyc} -40	_	2.0 × t _{cyc} –60	ns	_
Read data access time 4	t _{ACC4}	_	2.5 × t _{cyc} –25	_	2.5 × t _{cyc} –35	_	2.5 × t _{cyc} –50	ns	_
Read data access time 5	t _{ACC5}	_	3.0 × t _{cyc} –30	_	3.0 × t _{cyc} –40	_	3.0 × t _{cyc} –60	ns	_
WR delay time 1	t _{WRD1}	_	30	_	45	_	60	ns	_
WR delay time 2	t _{WRD2}	_	30	_	45	_	60	ns	_
WR pulse width 1	t _{wsw1}	1.0 × t _{cyc} –20	_	1.0 × t _{cyc} –30	_	1.0 × t _{cyc} –40	_	ns	_
WR pulse width 2	t _{wsw2}	$1.5 \times t_{\text{cyc}}$ –20	_	1.5 × t _{cyc} –30	_	1.5 × t _{cyc} –40	_	ns	_
Write data delay time	t _{wdd}	_	30	_	45	_	60	ns	_
Write data setup time	t _{wds}	0	_	0	_	0	_	ns	
Write data hold time	\mathbf{t}_{WDH}	10	_	15	_	20	_	ns	
WAIT setup time	t _{wts}	30	_	45	_	60	_	ns	_
WAIT hold time	t _{wth}	5	_	5	_	10	_	ns	_

Table 26.22 Bus Timing (2) (Advanced mode)

Condition A: $V_{cc} = 5.0 \text{ V} \pm 10\%$, $V_{cc}B = 5.0 \text{ V} \pm 10\%$, $V_{ss} = 0 \text{ V}$, $\phi = 2 \text{ MHz}$ to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$ (regular specifications), $T_a = -40 \text{ to } +85^{\circ}\text{C}$ (wide-range specifications)

Condition B: $V_{cc} = 4.0 \text{ V}$ to 5.5 V, $V_{cc}B = 4.0 \text{ V}$ to 5.5 V, $V_{ss} = 0 \text{ V}$, $\phi = 2 \text{ MHz}$ to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$ (regular specifications), $T_a = -40 \text{ to } +85^{\circ}\text{C}$ (wide-range specifications)

Condition C: $V_{cc} = 2.7 \text{ V}$ to 3.6 V, $V_{cc}B = 2.7 \text{ V}$ to 5.5 V, $V_{ss} = 0 \text{ V}$, $\phi = 2 \text{ MHz}$ to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$

		Cond	dition A Condition B		ition B				
		20	MHz	16	MHz	10	MHz		Test
Item	Symbol	Min	Max	Min	Max	Min	Max	Unit	Conditions
Address delay time	t _{AD}	_	30	_	45	_	60	ns	Figure 26.10 to
Address setup time	t _{AS}	$0.5 \times t_{\rm cyc} -25$	_	$0.5 \times t_{\text{cyc}} -35$	_	$0.5 \times t_{\text{cyc}}$ -50	_	ns	figure 26.14
Address hold time	t _{AH}	$\begin{array}{l} 0.5 \times \\ t_{_{cyc}} - 10 \end{array}$	_	$0.5 \times t_{_{\mathrm{cyc}}}$ -15	_	$\begin{array}{l} 0.5 \times \\ t_{_{\text{cyc}}} -20 \end{array}$	_	ns	
CS delay time (IOS)	t _{CSD}	_	30	_	45	_	60	ns	_
AS delay time	t _{ASD}	_	30	_	45	_	60	ns	
RD delay time 1	t _{RSD1}	_	30	_	45	_	60	ns	
RD delay time 2	t _{RSD2}	_	30	_	45	_	60	ns	_
Read data setup time	t _{RDS}	15	_	20	_	35	_	ns	_
Read data hold time	t _{rdh}	0	_	0	_	0	_	ns	_
Read data access time 1	t _{ACC1}	_	1.0 × t _{cyc} –40	-	$1.0 \times t_{\rm cyc}$ –55	_	$1.0 \times t_{\rm cyc} -80$	ns	
Read data access time 2	t _{ACC2}	_	1.5 × t _{cyc} –25	_	1.5 × t _{cyc} –35	_	1.5 × t _{cyc} –50	ns	

		Cond	ition A	Cond	ition B	Cond	ition C		
		20	MHz	16	MHz	10	MHz		Test
Item	Symbol	Min	Max	Min	Max	Min	Max	Unit	
Read data access time 3	t _{ACC3}	_	2.0 × t _{cyc} -40	_	2.0 × t _{cyc} –55	_	2.0 × t _{cyc} -80	ns	Figure 26.10 to figure 26.14
Read data access time 4	t _{ACC4}	_	2.5 × t _{cyc} –25	_	2.5 × t _{cyc} –35	_	$2.5 \times t_{\text{cyc}} -50$	ns	_
Read data access time 5	t _{ACC5}	_	3.0 × t _{cyc} –40	_	$3.0 \times t_{cyc} -55$	_	$3.0 \times t_{cyc} -80$	ns	_
WR delay time 1	t _{WRD1}	_	30	_	45	_	60	ns	_
WR delay time 2	t _{WRD2}	_	30	_	45	_	60	ns	_
WR pulse width 1	t _{wsw1}	1.0 × t _{cyc} –20	_	1.0 × t _{cyc} –30	_	1.0 × t _{cyc} –40	_	ns	_
WR pulse width 2	t _{wsw2}	1.5 × t _{cyc} –20	_	1.5 × t _{cyc} –30	_	1.5 × t _{cyc} –40	_	ns	_
Write data delay time	t _{wdd}	_	30	_	45	_	60	ns	=
Write data setup time	t _{wds}	0	_	0	_	0	_	ns	=
Write data hold time	t _{wdh}	10	_	15	_	20	_	ns	_
WAIT setup time	t _{wts}	30	_	45	_	60	_	ns	_
WAIT hold time	t _{wth}	5	_	5	_	10	_	ns	_

(4) Timing of On-Chip Supporting Modules

Tables 26.23 to 26.25 show the on-chip supporting module timing. The only on-chip supporting modules that can operate in subclock operation ($\phi = 32.768 \text{ kHz}$) are the I/O ports, external interrupts (NMI and IRQ0, 1, 2, 6, and 7), the watchdog timer, and the 8-bit timer (channels 0 and 1).

Table 26.23 Timing of On-Chip Supporting Modules (1)

Condition A: $V_{cc} = 5.0 \text{ V} \pm 10\%$, $V_{cc}B = 5.0 \text{ V} \pm 10\%$, $V_{ss} = 0 \text{ V}$, $\phi = 32.768 \text{ kHz}^*$, 2 MHz to maximum operating frequency, $T_a = -20 \text{ to} + 75^{\circ}\text{C}$ (regular specifications), $T_a = -40 \text{ to} + 85^{\circ}\text{C}$ (wide-range specifications)

Condition B: $V_{cc} = 4.0 \text{ V}$ to 5.5 V, $V_{cc}B = 4.0 \text{ V}$ to 5.5 V, $V_{ss} = 0 \text{ V}$, $\phi = 32.768 \text{ kHz}^*$, 2 MHz to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$ (regular specifications), $T_a = -40 \text{ to } +85^{\circ}\text{C}$ (wide-range specifications)

Condition C: $V_{cc} = 2.7 \text{ V}$ to 3.6 V, $V_{cc}B = 2.7 \text{ V}$ to 5.5 V, $V_{ss} = 0 \text{ V}$, $\phi = 32.768 \text{ kHz}^*$, 2 MHz to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$

				Cond	dition A	Cond	lition B	Conc	lition C		
				20	MHz	16	MHz	10	MHz		Test
Item			Symbol	Min	Max	Min	Max	Min	Max	Unit	Conditions
I/O ports	Output d	ata delay	t _{PWD}	_	50	_	50	_	100	ns	Figure 26.15
	Input dat time	a setup	t _{PRS}	30	_	30	_	50	_		
	Input dat time	a hold	t _{PRH}	30	_	30	_	50	_		
FRT	Timer ou time	tput delay	t _{FTOD}	_	50	_	50	_	100	ns	Figure 26.16
	Timer inp	out setup	t _{FTIS}	30	_	30	_	50	_		
	Timer clo	-	t _{FTCS}	30	_	30	_	50	_		Figure 26.17
	Timer clock	Single edge	t _{FTCWH}	1.5	_	1.5	_	1.5	_	t _{cyc}	-
	pulse width	Both edges	t _{FTCWL}	2.5	_	2.5	_	2.5	_		

				Cond	dition A	Cond	dition B	Conc	dition C		
				20	MHz	16	MHz	10	MHz		Test
Item			Symbol	Min	Max	Min	Max	Min	Max	Unit	Conditions
TMR	Timer of delay ti		\mathbf{t}_{TMOD}	_	50	_	50	_	100	ns	Figure 26.18
	Timer r	eset input me	\mathbf{t}_{TMRS}	30	_	30	_	50	_		Figure 26.20
	Timer of setup ti	clock input ime	\mathbf{t}_{TMCS}	30	_	30	_	50	_		Figure 26.19
	Timer clock	Single edge	t _{TMCWH}	1.5	_	1.5	_	1.5	_	t _{cyc}	_
	pulse width	Both edges	t _{TMCWL}	2.5	_	2.5	_	2.5	_		
PWM, PWMX	Pulse o		t _{PWOD}	_	50	_	50	_	100	ns	Figure 26.21
SCI	Input clock	Asynchro- nous	t _{Scyc}	4	_	4	_	4	_	t _{cyc}	Figure 26.22
	cycle	Synchro- nous	_	6	_	6	_	6	_		
	Input cl width	ock pulse	t _{sckw}	0.4	0.6	0.4	0.6	0.4	0.6	t _{Scyc}	_
	Input cl	ock rise	t _{SCKr}	_	1.5	_	1.5	_	1.5	t _{cyc}	=
	Input cl	ock fall	t _{sckf}	_	1.5	_	1.5	_	1.5		
	Transm delay ti (synchi		t _{TXD}	_	50	_	50	_	100	ns	Figure 26.23
		e data setup ynchronous)	t _{RXS}	50	_	50	_	100	_	ns	_
		e data hold ynchronous)	t _{RXH}	50	_	50	_	100	_	ns	_
A/D conver- ter		input setup	t _{TRGS}	30	_	30	_	50	_	ns	Figure 26.24
WDT	RESO time	output delay	t _{RESD}	_	100	_	120	_	200	ns	Figure 26.25
	RESO width	output pulse	t _{RESOW}	132	_	132	_	132	_	t _{cyc}	_

Note: * Only supporting modules that can be used in subclock operation

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Table 26.23 Timing of On-Chip Supporting Modules (2)

Condition A: $V_{cc} = 5.0 \text{ V} \pm 10\%$, $V_{cc}B = 5.0 \text{ V} \pm 10\%$, $V_{ss} = 0 \text{ V}$, $\phi = 2 \text{ MHz}$ to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$ (regular specifications), $T_a = -40 \text{ to } +85^{\circ}\text{C}$ (wide-range specifications)

Condition B: $V_{cc} = 4.0 \text{ V to } 5.5 \text{ V}, V_{cc}B = 4.0 \text{ V to } 5.5 \text{ V}, V_{ss} = 0 \text{ V}, \phi = 2 \text{ MHz to maximum operating frequency}, T_a = -20 \text{ to } +75^{\circ}\text{C} \text{ (regular specifications)}, T_a = -40 \text{ to } +85^{\circ}\text{C} \text{ (wide-range specifications)}$

Condition C: $V_{cc} = 2.7 \text{ V}$ to 3.6 V, $V_{cc}B = 2.7 \text{ V}$ to 5.5 V, $V_{ss} = 0 \text{ V}$, $\phi = 2 \text{ MHz}$ to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$

				Cond	dition A	Cond	dition B	Cond	lition C		
				20	MHz	16	MHz	10	MHz		Test
Item			Symbol	Min	Max	Min	Max	Min	Max	Unit	Conditions
HIF read cycle	CS/HA time	0 setup	t _{HAR}	10	_	10	_	10	_	ns	Figure 26.26
	CS/HA	0 hold time	t _{HRA}	10	_	10	_	10	_	ns	_
	IOR pu	lse width	t _{HRPW}	120	_	120	_	220	_	ns	_
	HDB de	elay time	t _{HRD}	_	100	_	100	_	200	ns	=
	HDB h	old time	t _{HRF}	0	25	0	25	0	40	ns	_
	HIRQ o	delay time	t _{HIRQ}		120	_	120	_	200	ns	=
HIF write cycle	CS/HA time	0 setup	t _{HAW}	10	_	10	_	10	_	ns	_
	CS/HA	0 hold time	t _{HWA}	10	_	10	_	10	_	ns	_
	IOW pu	ulse width	t _{HWPW}	60	_	60	_	100	_	ns	=
	HDB setup time	Fast A20 gate not used	t _{HDW}	30	_	30	_	50	_	ns	_
		Fast A20 gate used	_	45	_	55		85		ns	_
	HDB h	old time	t _{HWD}	15	_	15	_	25	_	ns	=
	GA20 d	delay time	t _{HGA}	_	90	_	90	_	180	ns	=

Table 26.24 Keyboard Buffer Controller Timing

Conditions: $V_{cc} = 4.0 \text{ V}$ to 5.5 V, $V_{cc} = 2.7 \text{ V}$ to 3.6 V (3 V version), $V_{cc}B = 2.7 \text{ V}$ to 5.5 V,

 $V_{ss} = 0 \text{ V}, \phi = 2 \text{ MHz}$ to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$

		R	atings			Test	
Item	Symbol	Min	Тур	Max	Unit	Conditions	Notes
KCLK, KD output fall time	t _{KBF}	20 + 0.1 Cb	_	250	ns		Figure 26.27
KCLK, KD input data hold time	t _{KBIH}	150	_	_	ns		_
KCLK, KD input data setup time	t _{KBIS}	150	_	_	ns		_
KCLK, KD output delay time	t _{KBOD}	_	_	450	ns		_
KCLK, KD capacitive load	Сь	_	_	400	pF		

Table 26.25 I²C Bus Timing

Conditions: $V_{cc} = 4.0 \text{ V}$ to 5.5 V, $V_{cc} = 2.7 \text{ V}$ to 3.6 V (3 V version), $V_{ss} = 0 \text{ V}$,

 $\phi = 5$ MHz to maximum operating frequency

Ratings				
Тур	Max	Unit	Test Conditions	Notes

Item	Symbol	Min	Тур	Max	Unit	Test Conditions	Notes
SCL input cycle time	t _{scL}	12	_	_	t _{cyc}		Figure 26.28
SCL input high pulse width	t _{sclh}	3	_	_	t _{cyc}		_
SCL input low pulse width	t _{scll}	5	_	_	t _{cyc}		_
SCL, SDA input rise time	t _{sr}	_	_	7.5*	t _{cyc}		_
SCL, SDA input fall time	t _{Sf}	_	_	300	ns		_
SCL, SDA output fall time	t _{of}	20 + 0.1 Cb	_	250	ns		_
SCL, SDA input spike pulse elimination time	t _{SP}	_	_	1	t _{cyc}		_
SDA input bus free time	t _{BUF}	5	_	_	t _{cyc}		_
Start condition input hold time	t _{STAH}	3	_	_	t _{cyc}		_
Retransmission start condition input setup time	t _{STAS}	3	_	_	t _{cyc}		_
Stop condition input setup time	t _{stos}	3	_	_	$t_{\rm cyc}$		
Data input setup time	t _{SDAS}	0.5	_	_	$t_{\rm cyc}$		
Data input hold time	t _{SDAH}	0			ns		_
SCL, SDA capacitive load	Сь	_	_	400	pF		

Note: * 17.5t_{cyc} can be set according to the clock selected for use by the I²C module. For details, see section 16.4, Usage Notes.

26.3.4 A/D Conversion Characteristics

Tables 26.26 and 26.27 list the A/D conversion characteristics.

Table 26.26 A/D Conversion Characteristics (AN7 to AN0 Input: 134/266-State Conversion)

O = 1'd = A = X = 5 0 X + 100/ AX = 5 0 X + 100/ AX = 4.5 X + 100/

Condition A: $V_{cc} = 5.0 \text{ V} \pm 10\%$, $AV_{cc} = 5.0 \text{ V} \pm 10\%$, $AV_{ref} = 4.5 \text{ V}$ to AV_{cc} ,

 $\boldsymbol{V}_{ss}=\boldsymbol{A}\boldsymbol{V}_{ss}=\boldsymbol{0}$ V, $\boldsymbol{\varphi}=\boldsymbol{2}$ MHz to maximum operating frequency,

 $T_a = -20 \text{ to } +75^{\circ}\text{C}$ (regular specifications),

 $T_a = -40 \text{ to } +85^{\circ}\text{C}$ (wide-range specifications)

Condition B: $V_{CC} = 4.0 \text{ V}$ to 5.5 V, $AV_{CC} = 4.0 \text{ V}$ to 5.5 V, $AV_{ref} = 4.0 \text{ V}$ to $AV_{CC} = 4.0 \text{ V}$

 $V_{ss} = AV_{ss} = 0 \text{ V}, \phi = 2 \text{ MHz}$ to maximum operating frequency,

 $T_a = -20 \text{ to } +75^{\circ}\text{C}$ (regular specifications),

 $T_a = -40 \text{ to } +85^{\circ}\text{C}$ (wide-range specifications)

Condition C: $V_{cc} = 2.7 \text{ V}$ to 3.6 V, $AV_{cc} = 2.7 \text{ V}$ to 3.6 V, $AV_{ref} = 2.7 \text{ V}$ to AV_{cc}

 $V_{ss} = AV_{ss} = 0 \text{ V}, \phi = 2 \text{ MHz}$ to maximum operating frequency,

 $T_a = -20 \text{ to } +75^{\circ}\text{C}$

		Conditio	n A		Conditio	n B		Conditio	on C	
		20 MF	lz		16 MF	lz		10 MF	lz	_
Item	Min	Тур	Max	Min	Тур	Max	Min	Тур	Max	Unit
Resolution	10	10	10	10	10	10	10	10	10	Bits
Conversion time*3	_	_	6.7	_	_	8.4	_	_	13.4	μs
Analog input capacitance	_	_	20	_	_	20	_	_	20	pF
Permissible signal- source impedance	_	_	10 ^{*1} 5 ^{*2}	_	_	10 ^{*1} 5 ^{*2}	_	_	5	kΩ
Nonlinearity error	_	_	±3.0	_	_	±3.0	_	_	±7.0	LSB
Offset error	_	_	±3.5	_	_	±3.5	_	_	±7.5	LSB
Full-scale error	_	_	±3.5	_	_	±3.5	_	_	±7.5	LSB
Quantization error	_	_	±0.5	_	_	±0.5	_	_	±0.5	LSB
Absolute accuracy	_	_	±4.0		_	±4.0	_	_	±8.0	LSB

Notes: 1. When conversion time \geq 11. 17 μs (CKS = 1 and $\phi \leq$ 12 MHz, or CKS = 0)

- 2. When conversion time < 11. 17 μs (CKS = 1 and ϕ > 12 MHz)
- 3. In single mode and ϕ = maximum operating frequency.

Table 26.27 A/D Conversion Characteristics (CIN15 to CIN0 Input: 134/266-State Conversion)

Condition A:
$$V_{cc} = 5.0 \text{ V} \pm 10\%$$
, $AV_{cc} = 5.0 \text{ V} \pm 10\%$, $AV_{ref} = 4.5 \text{ V}$ to AV_{cc} , $V_{ss} = AV_{ss} = 0 \text{ V}$, $\phi = 2 \text{ MHz}$ to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$ (regular specifications), $T_a = -40 \text{ to } +85^{\circ}\text{C}$ (wide-range specifications)

Condition B:
$$V_{cc} = 4.0 \text{ V to } 5.5 \text{ V}, \text{AV}_{cc} = 4.0 \text{ V to } 5.5 \text{ V}, \text{AV}_{ref} = 4.0 \text{ V to } \text{AV}_{cc},$$

$$V_{ss} = \text{AV}_{ss} = 0 \text{ V}, \phi = 2 \text{ MHz to maximum operating frequency},$$

$$T_a = -20 \text{ to } +75^{\circ}\text{C (regular specifications)},$$

$$T_a = -40 \text{ to } +85^{\circ}\text{C (wide-range specifications)}$$

Condition C: $V_{cc} = 3.0 \text{ V to } 3.6 \text{ V}^{*4}, \text{ AV}_{cc} = 3.0 \text{ V to } 3.6 \text{ V}^{*4}, \text{ AV}_{ref} = 3.0 \text{ V to } \text{AV}_{cc}^{*4}, \\ V_{cc}B = 3.0 \text{ V to } 5.5 \text{ V}^{*4}, V_{ss} = \text{AV}_{ss} = 0 \text{ V}, \\ \phi = 2 \text{ MHz to maximum operating frequency, } T_{a} = -20 \text{ to } +75^{\circ}\text{C}$

		Conditio	n A		Conditio	n B		Conditio	n C	
		20 MF	lz		16 MF	lz		10 MF	lz	•
Item	Min	Тур	Max	Min	Тур	Max	Min	Тур	Max	Unit
Resolution	10	10	10	10	10	10	10	10	10	Bits
Conversion time*3	_	_	6.7	_	_	8.4	_	_	13.4	μs
Analog input capacitance	_	_	20	_	_	20	_	_	20	pF
Permissible signal- source impedance	_	_	10 ^{*1} 5 ^{*2}		_	10 ^{*1} 5 ^{*2}		_	5	kΩ
Nonlinearity error		_	±5.0	_	_	±5.0	_	_	±11.0	LSB
Offset error	_	_	±5.5	_	_	±5.5	_	_	±11.5	LSB
Full-scale error	_	_	±5.5	_	_	±5.5	_	_	±11.5	LSB
Quantization error		_	±0.5	_	_	±0.5	_	_	±0.5	LSB
Absolute accuracy	_	_	±6.0	_	_	±6.0	_	_	±12.0	LSB

Notes: 1. When conversion time \geq 11. 17 μs (CKS = 1 and $\varphi \leq$ 12 MHz, or CKS = 0)

- 2. When conversion time < 11. 17 μs (CKS = 1 and φ > 12 MHz)
- 3. In single mode and ϕ = maximum operating frequency.
- 4. When using CIN, the applicable range is V_{cc} = 3.0 V to 3.6 V, AV_{cc} = 3.0 V to 3.6 V, AV_{cc} = 3.0 V to 3.6 V, and $V_{cc}B$ = 3.0 V to 5.5 V.

26.3.5 D/A Conversion Characteristics

Table 26.28 lists the D/A conversion characteristics.

Table 26.28 D/A Conversion Characteristics

Condition A: $V_{cc} = 5.0 \text{ V} \pm 10\%$, $AV_{cc} = 5.0 \text{ V} \pm 10\%$, $AV_{ref} = 4.5 \text{ V}$ to AV_{cc} ,

 $\boldsymbol{V}_{ss} = \boldsymbol{A}\boldsymbol{V}_{ss} = \boldsymbol{0}$ V, $\boldsymbol{\varphi} = \boldsymbol{2}$ MHz to maximum operating frequency,

 $T_a = -20 \text{ to } +75^{\circ}\text{C}$ (regular specifications),

 $T_a = -40 \text{ to } +85^{\circ}\text{C} \text{ (wide-range specifications)}$

Condition B: $V_{CC} = 4.0 \text{ V}$ to 5.5 V, $AV_{CC} = 4.0 \text{ V}$ to 5.5 V, $AV_{Ref} = 4.0 \text{ V}$ to AV_{CC}

 $V_{ss} = AV_{ss} = 0 \text{ V}, \phi = 2 \text{ MHz}$ to maximum operating frequency,

 $T_a = -20 \text{ to } +75^{\circ}\text{C}$ (regular specifications),

 $T_a = -40 \text{ to } +85^{\circ}\text{C}$ (wide-range specifications)

Condition C: $V_{CC} = 2.7 \text{ V}$ to 3.6 V, $AV_{CC} = 2.7 \text{ V}$ to 3.6 V, $AV_{ref} = 2.7 \text{ V}$ to AV_{CC}

 $V_{ss} = AV_{ss} = 0 \text{ V}, \phi = 2 \text{ MHz}$ to maximum operating frequency,

 $T_0 = -20 \text{ to } +75^{\circ}\text{C}$

		С	onditio	on A	Condition B		C				
			20 MF	łz	16 MHz		10 MHz			-	
Item		Min	Тур	Max	Min	Тур	Max	Min	Тур	Max	Unit
Resolution		8	8	8	8	8	8	8	8	8	Bits
Conversion time	With 20-pF load capacitance	_	_	10	_	_	10	_	_	10	μs
Absolute accuracy	With 2-MΩ load resistance	_	±1.0	±1.5	_	±1.0	±1.5	_	±2.0	±3.0	LSB
	With 4-MΩ load resistance	_	_	±1.0	_	_	±1.0	_	_	±2.0	-

26.3.6 Flash Memory Characteristics

Table 26.29 shows the flash memory characteristics.

Table 26.29 Flash Memory Characteristics (Programming/erasing operating range)

Conditions (5 V version): $V_{cc} = 4.0 \text{ V}$ to 5.5 V, $V_{ss} = 0 \text{ V}$, $T_a = -20 \text{ to } +75^{\circ}\text{C}$ (regular

specifications), $T_a = -40$ to $+85^{\circ}$ C (wide-range specifications)

(3 V version): $V_{CC} = 3.0 \text{ V}$ to 3.6 V, $V_{SS} = 0 \text{ V}$, $T_a = -20 \text{ to } 75^{\circ}\text{C}$

Item	ltem		Min	Тур	Max	Unit	Test Condition
Programming	time ^{*1 *2 *4}	tP	_	10	200	ms/ 128 bytes	
Erase time*1*	3 *6	tE	_	100	1200	ms/ block	
Reprogrammir	ng count	N _{WEC}	100*8	10000*9	_	Times	
Data retention	time ^{*10}	t _{DRP}	10	_	_	Years	
Programming	Wait time after SWE-bit setting*1	х	1	_	_	μs	
	Wait time after PSU-bit setting*1	у	50	_	_	μs	
	Wait time after P-bit setting*1*4	z1	28	30	32	μs	$1 \le n \le 6$
		z2	198	200	202	μs	$7 \le n \le 1000$
		z3	8	10	12	μs	Additional writing
	Wait time after P-bit clear*1	α	5	_	_	μs	
	Wait time after PSU-bit clear*1	β	5	_	_	μs	
	Wait time after PV-bit setting*1	γ	4	_	_	μs	
	Wait time after dummy write*1	ε	2	_	_	μs	
	Wait time after PV-bit clear*1	η	2	_	_	μs	
	Wait time after SWE-bit clear*1	θ	100	_	_	μs	
	Maximum programming count*1*4*5	N	_	_	1000	Times	
Erase	Wait time after SWE-bit setting*1	х	1	_	_	μs	
	Wait time after ESU-bit setting*1	у	100	_	_	μs	
	Wait time after E-bit setting*1 *6	Z	10	_	100	ms	
	Wait time after E-bit clear*1	α	10	_	_	μs	
	Wait time after ESU-bit clear*1	β	10	_	_	μs	
	Wait time after EV-bit setting*1	γ	20	_	_	μs	
	Wait time after H'FF dummy write*1	ε	2	_	_	μs	
	Wait time after EV-bit clear*1	η	4	_	_	μs	
	Wait time after SWE-bit clear*1	θ	100	_	_	μs	
	Maximum erase count*1 *6 *7	N	_	_	120	Times	

Notes: 1. Set the times according to the program/erase algorithms.

- 2. Programming time per 128 bytes (Shows the total period for which the P-bit in FLMCR1 is set. It does not include the programming verification time.)
- 3. Block erase time (Shows the total period for which the E-bit in FLMCR1 is set. It does not include the erase verification time.)
- 4. Maximum programming time (tP (max)) $tP(max) = (Wait time after P-bit setting (z1) + (z3)) \times 6 + Wait time after P-bit setting (z2) <math>\times ((N) 6)$
- 5. Maximum programming count (N) should be set according to the actual set value of (z1, z2, z3) to allow programming within the maximum programming time (tP(max)). The wait time after P-bit setting (z1, z2, z3) must be changed with the value of the number of writing times (n) as follows.

The number of times for writing n

$$1 \le n \le 6$$
 $z1 = 30 \,\mu s, z3 = 10 \,\mu s$

$$7 \le n \le 1000$$
 $z2 = 200 \text{ us}$

- Maximum erase time (tE (max)) tE(max) = Waiting time after E-bit setting (z) × Maximum erase count (N)
- 7. Maximum erase count (N) should be set according to the actual setting (z) to allow erase within the maximum erase time (tE(max)).
- 8. Minimum number of times for which all characteristics are guaranteed after rewriting (Guarantee range is 1 to minimum value).
- 9. Reference value for 25°C (as a guideline, rewriting should normally function up to this value).
- 10. Data retention characteristic when rewriting is performed within the specification range, including the minimum value.



26.3.7 Usage Note

(1) The F-ZTAT and mask ROM versions have been confirmed as fully meeting the reference values for electrical characteristics shown in this manual. However, actual performance figures, operating margins, noise margins, and other properties may vary due to differences in the manufacturing process, on-chip ROM, layout patterns, etc.

When system evaluation testing is carried out using the F-ZTAT version, the same evaluation tests should also be conducted for the mask ROM version when changing over to that version.

(2) On-chip power supply step-down circuit

The following products incorporate an internal power supply step-down circuit, which automatically drops down the internal power supply voltage to the optimum internal voltage level: the F-ZTAT A-mask versions of the H8S/2148, H8S/2147, and H8S/2144 (HD64F2148A, HD64F2147A, and HD64F2144A) and the mask ROM versions of the H8S/2148, H8S/2147, H8S/2144, and H8S/2143 (HD6432148S, HD6432148SW, HD6432147S, HD6432147SW, HD6432144S, and HD6432143S).

The voltage-stabilization capacitor (0.47 μ F one or two connected in group) must be connected between the VCL (internal power supply step-down) and VSS pins.

Figure 26.3 shows the connection of the external capacitors.

For the 5- or 4-V version whose power supply (V_{cc}) voltage exceeds 3.6 V, do not connect the VCL pin in a product incorporating an internal step-down circuit to the V_{cc} power supply. (Connect the VCC1 pin to the V_{cc} power supply as usual.)

For the 3-V version whose power supply (V_{CC}) voltage is 3.6 V or lower, connect both the VCL and VCC1 pins to the system power supply.

When changing from the F-ZTAT versions not incorporating an internal step-down circuit to the F-ZTAT A-mask versions or mask ROM versions incorporating an internal step-down circuit, the VCL pin has the same pin location as the VCC2 pin. Therefore, note that the circuit patterns differ between these two types of products.

26.4 Electrical Characteristics of H8S/2147N F-ZTAT

26.4.1 Absolute Maximum Ratings

Table 26.30 lists the absolute maximum ratings.

Table 26.30 Absolute Maximum Ratings

Symbol	Value	Unit
V _{cc}	-0.3 to +7.0	V
V _{cc} B	-0.3 to +7.0	V
V _{in}	-0.3 to V _{cc} +0.3	V
V_{in}	-0.3 to V _{cc} +0.3	V
V_{in}	–0.3 to V _{cc} B +0.3	V
V _{in}	–0.3 V to lower of voltages $V_{\rm cc}$ +0.3 and $AV_{\rm cc}$ +0.3	V
V_{in}	–0.3 V to lower of voltages $V_{\rm cc}B$ +0.3 and $AV_{\rm cc}$ +0.3	V
V _{in}	-0.3 to AV _{cc} +0.3	V
AV_{ref}	-0.3 to AV _{cc} +0.3	V
AV _{cc}	-0.3 to +7.0	V
V_{AN}	-0.3 to AV _{cc} +0.3	V
T _{opr}	Regular specifications: -20 to +75	°C
T_{opr}	Regular specifications: 0 to +75	°C
T_{stg}	-55 to +125	°C
	V _{CC} V _{CC} V _{In} T _{Opr}	$\begin{array}{cccccccccccccccccccccccccccccccccccc$

Caution: Permanent damage to the chip may result if absolute maximum ratings are exceeded.

Note: * Power supply voltage for VCC1 and VCC2 pins.



26.4.2 DC Characteristics

Table 26.31 lists the DC characteristics. Permitted output current values and bus drive characteristics are shown in tables 26.32 and 26.33, respectively.

Table 26.31 DC Characteristics (1)

$$\begin{array}{ll} \text{Conditions:} & V_{_{CC}} = 5.0 \text{ V} \pm 10\%, \, V_{_{CC}} B = 5.0 \text{ V} \pm 10\%, \, A V_{_{CC}}^{\quad \ \, *^{1}} = 5.0 \text{ V} \pm 10\%, \\ & A V_{_{ref}}^{\quad \ \, *^{1}} = 4.5 \text{ V} \text{ to } A V_{_{CC}}, \, V_{_{SS}} = A V_{_{SS}}^{\quad \ \, *^{1}} = 0 \text{ V}, \, T_{_{a}} = -20 \text{ to } +75^{\circ} \text{C}^{*^{11}} \\ \end{array}$$

11			0	N4:	T		11!1	Test
Item			Symbol	Min	Тур	Max	Unit	Conditions
Schmitt	P67 to	(1)	V _T	1.0	_	_	V	
trigger input voltage	P60(KWUL = $00)^{*2}$, KIN15 to		$V_{T}^{\; \star}$	_	_	$V_{cc} \times 0.7$ $V_{cc} B \times 0.7$		
	KIN8*7*8, IRQ2 to IRQ0*3, IRQ5 to IRQ3		$V_T^+ - V_T^-$	0.4	_	_		
Schmitt	P67 to P60	_	V _T	$V_{cc} \times 0.3$	_	_	V	
trigger input voltage	(KWUL = 01)		V _T ⁺	_	_	$V_{cc} \times 0.7$	=	
(in level			$V_{\scriptscriptstyle T}^{\;\scriptscriptstyle +} - V_{\scriptscriptstyle T}^{\;\scriptscriptstyle -}$	$V_{\text{cc}} \times 0.05$	_	_	_	
switching)*6	P67 to P60 (KWUL = 10)		$V_{\scriptscriptstyle T}^{\;-}$	$V_{\text{cc}} \times 0.4$	_	_		
			V _T ⁺	_	_	$V_{cc} \times 0.8$	_	
			$V_T^+ - V_T^-$	$V_{cc} \times 0.03$	_	_	=	
	P67 to P60 (KWUL = 11)	_	V _T	$V_{cc} \times 0.45$	_	_	=	
			V _T ⁺	_	_	$V_{cc} \times 0.9$	=	
			$V_T^+ - V_T^-$	0.05	_	_	=	
Input high voltage	RES, STBY, NMI, MD1, MD0	(2)	V _{IH}	V _{cc} -0.7	_	V _{cc} +0.3	V	
	EXTAL	-		$V_{cc} \times 0.7$	_	V _{cc} +0.3	-	
	PA7 to PA0*7	=		$V_{cc}B \times 0.7$	_	V _{cc} B +0.3	=	
	Port 7			2.0	_	AV _{cc} +0.3	-	
	Input pins except (1) and (2) above		_	2.0	_	V _{cc} +0.3	_	

Item			Symbol	Min	Тур	Max	Unit	Test Conditions
Input low voltage	RES, STBY, MD1, MD0	(3)	V _{IL}	-0.3	_	0.5	V	
	PA7 to PA0	=		-0.3	_	1.0		
	NMI, EXTAL, input pins except (1) and (3) above		_	-0.3	_	0.8		
Output high			V_{OH}	V _{cc} -0.5	_	_	V	$I_{OH} = -200 \ \mu A$
voltage	(except P97, and P52*4)*5 *8			$V_{cc}B$ -0.5				
	1 32)			3.5	_	_	V	$I_{OH} = -1 \text{ mA}$
	P97, P52*4			2.5	_	_	V	$I_{OH} = -1 \text{ mA}$
Output low voltage	All output pins (except RESO)*5		V _{oL}	_	_	0.4	V	I _{OL} = 1.6 mA
	Ports 1 to 3			_	_	1.0	V	I _{oL} = 10 mA
	RESO			_	_	0.4	V	I _{OL} = 2.6 mA
Input	RES		I _{in}	_	_	10.0	μΑ	V _{in} = 0.5 to
leakage current	STBY, NMI, MD1, MD0			_	_	1.0	μA	[−] V _{cc} –0.5 V
	Port 7			_	_	1.0	μA	$V_{in} = 0.5 \text{ to}$ AV _{cc} -0.5 V
Three-state leakage current (off state)	Ports 1 to 6, 8, 9, A**, B		I _{TSI}	_	_	1.0	μA	$V_{in} = 0.5 \text{ to}$ $V_{cc} - 0.5 \text{ V},$ $V_{in} = 0.5 \text{ to}$ $V_{cc}B - 0.5 \text{ V}$
Input	Ports 1 to 3		-I _P	50	_	300	μΑ	$V_{in} = 0 V$
pull-up MOS current	Ports A*8, B, Port 6 (P6PUE = 0)			60	_	500	μΑ	_
	Port 6 (P6PUE = 1)		_	15	_	150	μΑ	_

Item		Symb	ol Min	Тур	Max	Unit	Test Conditions
Input		(4) C _{in}	_	_	80	pF	$V_{in} = 0 V$,
capacitance	NMI		_	_	50	pF	[−] f = 1 MHz, – T _a = 25°C
	P52, P97, P42, P86 PA7 to PA2		_	_	20	pF	– I _a – 23 O
	Input pins except (4) above		_	_	15	pF	_
Current	Normal operation	I _{cc}	_	75	100	mA	f = 20 MHz
dissipation*9	Sleep mode		_	60	85	mA	f = 20 MHz
	Standby mode*10		_	0.01	5.0	μA	$T_a \le 50^{\circ}C$
			_	_	20.0	μA	50°C < T _a
Analog power	During A/D, D/A conversion	Al _{cc}	_	1.2	2.0	mA	
supply current	Idle		_	0.01	5.0	μΑ	AV _{cc} = 2.0 V to 5.5 V
Reference	During A/D conve	rsion Al _{ref}	_	0.5	1.0	mA	
power supply current	During A/D, D/A conversion		_	2.0	5.0	mA	_
	Idle		_	0.01	5.0	μΑ	$AV_{ref} = 2.0 V$ to AV_{CC}
Analog power	er supply voltage*1	AV_cc	4.5	_	5.5	V	Operating
			2.0	_	5.5	V	Idle/not used
RAM standb	RAM standby voltage		2.0	_	_	V	
Notes 1 F	On not loove the A	V _{RAM}	and AV/88 r	ino onon	oven if th	A/D	nyortor and

Notes: 1. Do not leave the AVCC, AV_{ref}, and AVSS pins open even if the A/D converter and D/A converter are not used.

Even if the A/D converter and D/A converter are not used, apply a value in the range 2.0 V to 5.5 V to AVCC and AV_{ref} pins by connection to the power supply (V_{cc}), or some other method. Ensure that AV_{ref} \leq AV_{cc}.

- 2. P67 to P60 include supporting module inputs multiplexed on those pins.
- 3. IRQ2 includes the ADTRG signal multiplexed on that pin.
- In the H8S/2147N, P52/SCK0/SCL0 and P97/SDA0 are NMOS push-pull outputs.
 An external pull-up resistor is necessary to provide high-level output from SCL0 and SDA0 (ICE = 1).

In the H8S/2147N, P52/SCK0 and P97 (ICE = 0) high levels are driven by NMOS.

5. When IICS = 0, ICE = 0, and KBIOE = 0. Low-level output when the bus drive function is selected is determined separately.

- 6. The upper limit of the port 6 applied voltage is V_{cc} +0.3 V when CIN input is not selected, and the lower of V_{cc} +0.3 V and AV_{cc} +0.3 V when CIN input is selected. When a pin is in output mode, the output voltage is equivalent to the applied voltage.
- 7. The upper limit of the port A applied voltage is $V_{cc}B + 0.3 \text{ V}$ when CIN input is not selected, and the lower of $V_{cc}B + 0.3 \text{ V}$ and $AV_{cc} + 0.3 \text{ V}$ when CIN input is selected. When a pin is in output mode, the output voltage is equivalent to the applied voltage.
- 8. Port A characteristics depends on $V_{cc}B$ (or on V_{cc} when other pins are in output mode).
- 9. Current dissipation values are for V_{IH} min = V_{CC} –0.5 V, $V_{CC}B$ –0.5 V, and V_{IL} max = 0.5 V with all output pins unloaded and the on-chip pull-up MOSs in the off state.
- 10. The values are for V_{RAM} \leq V_{CC} < 4.5V, V_{IH} min = V_{CC} \times 0.9, V_{CC}B \times 0.9, and V_{IL} max = 0.3 V.
- 11. For flash memory program/erase operations, the applicable range is T_a = 0 to +75°C (regular specifications).



Table 26.31 DC Characteristics (2)

 $\begin{array}{ll} \text{Conditions:} & V_{_{CC}} = 3.0 \text{ V to } 5.5 \text{ V}^{*_{11}}, V_{_{CC}} B = 3.0 \text{ V to } 5.5 \text{ V}, \text{AV}_{_{CC}}^{\quad *_{1}} = 3.0 \text{ V to } 5.5 \text{ V}, \\ & AV_{_{ref}} = 3.0 \text{ V to } 5.5 \text{ V}, V_{_{SS}} = AV_{_{SS}}^{\quad *_{1}} = 0 \text{ V}, T_{_{a}} = -20 \text{ to } +75^{\circ}\text{C}^{*_{11}} \\ \end{array}$

Item			Symbol	Min	Тур	Max	Unit	Test Conditions	
Schmitt trigger input	P67 to P60(KWUL =	(1)	V _T -	$V_{cc} \times 0.2$ $V_{cc} B \times 0.2$	_	_	V		
voltage	00)*2*6, KIN15 to KIN8*7*8.		V _T +	_	_	$V_{cc} \times 0.7$ $V_{cc} B \times 0.7$	V	_	
	IRQ2 to IRQ0*3, IRQ5 to IRQ3		$V_T^+ - V_T^-$	$V_{cc} \times 0.05$ $V_{cc} B \times$ 0.05	_	_	V		
Schmitt	P67 to P60	_	V _T	$V_{cc} \times 0.3$	_	_	V		
trigger input voltage (in level swiching)*6	(KWUL = 01)		$V_{\scriptscriptstyle T}^{^{\star}}$	_	_	$V_{cc} \times 0.7$			
		_	$V_{\scriptscriptstyle T}^{\;\scriptscriptstyle +} - V_{\scriptscriptstyle T}^{\;\scriptscriptstyle -}$	$V_{\text{cc}} \times 0.05$	_	_	_		
	P67 to P60 (KWUL = 10)		$V_{\scriptscriptstyle T}^{-}$	$V_{\text{cc}} \times 0.4$	_	_	_		
			V_{T}^{+}	_	_	$V_{cc} \times 0.8$	_		
	P67 to P60 (KWUL = 11)	_	_	$V_{T}^{+} - V_{T}^{-}$	$\rm V_{cc} \times 0.03$	_	_	_	
			$V_{\scriptscriptstyle T}^{\;-}$	$V_{\text{cc}} \times 0.45$	_	_	_		
			V_{T}^{+}	_	_	$V_{cc} \times 0.9$			
			$V_{\scriptscriptstyle T}^{^+} - V_{\scriptscriptstyle T}^{^-}$	0.05	_	_			
Input high voltage	RES, STBY, NMI, MD1, MD0	(2)	$V_{_{\mathrm{IH}}}$	$V_{cc} \times 0.9$	_	V _{cc} +0.3	V		
	EXTAL			$V_{cc} \times 0.7$	_	V _{cc} +0.3	V		
	PA7 to PA0*7	_		$V_{cc}B \times 0.7$	_	V _{cc} B +0.3	V	_	
	Port 7			$V_{cc} \times 0.7$	_	AV _{cc} +0.3	V		
	Input pins except (1) and (2) above			V _{cc} × 0.7	_	V _{cc} +0.3	V		

Item			Symbol	Min	Тур	Max	Unit	Test Conditions
Input low voltage	RES, STBY, MD1, MD0	(3)	V _{IL}	-0.3	_	$V_{cc} \times 0.1$	V	
	PA7 to PA0	=		-0.3	_	$V_{cc}B \times 0.2$	V	V _{cc} B = 3.0 V to 4.0 V
						0.8	V	$V_{cc}B = 4.0 \text{ V}$ to 5.5 V
	NMI, EXTAL, input pins except			-0.3	_	$V_{cc} \times 0.2$	V	$V_{cc} = 3.0 \text{ V to} $ 4.0 V
	(1) and (3) above					0.8	V	$V_{cc} = 4.0 \text{ V to} $ 5.5 V
Output high voltage	(except P97,		V_{OH}	V _{cc} -0.5 V _{cc} B -0.5	_	_	V	$I_{OH} = -200 \ \mu A$
	and P52*4)*5*8			V _{cc} -1.0 V _{cc} B -1.0	_	_	V	$I_{OH} = -1 \text{ mA}$ $(V_{CC} = 3.0 \text{ V}$ to 4.0 V, $V_{CC}B$ = 3.0 V to 4.0 V)
	P97, P52*4		_	1.0	_	_	V	$I_{OH} = -1 \text{ mA}$
Output low voltage	All output pins (except RESO)*5		V _{oL}	_	_	0.4	V	I _{oL} = 1.6 mA
	Ports 1 to 3		_	_	_	1.0	V	$I_{oL} = 5 \text{ mA}$ $(V_{cc} < 4.0 \text{ V}),$ $I_{oL} = 10 \text{ mA}$ $(4.0 \text{ V} \le V_{cc} \le 5.5 \text{ V})$
	RESO		_	_	_	0.4	V	I _{oL} = 1.6 mA
Input	RES		I _{in}	_	_	10.0	μΑ	$V_{in} = 0.5 \text{ to}$
leakage current	STBY, NMI, MD1, MD0		_	_	_	1.0	μΑ	[−] V _{cc} –0.5 V
	Port 7		_	_	_	1.0	μA	$V_{in} = 0.5 \text{ to}$ AV _{CC} -0.5 V
Three-state leakage current (off state)	Ports 1 to 6, 8, 9, A**, B		I _{TSI}	_	_	1.0	μA	$V_{in} = 0.5 \text{ to}$ $V_{CC} = -0.5 \text{ V},$ $V_{in} = 0.5 \text{ to}$ $V_{CC}B = -0.5 \text{ V}$

Item	Item		Min	Тур	Max	Unit	Test Conditions
Input pull-	Ports 1 to 3	-I _P	10	_	150	μA	$V_{in} = 0 V$,
up MOS current	Ports A*8, B Port 6 (P6PUE = 0)	_	30	_	250	μΑ	$V_{cc} = 3.0 \text{ V to}$ 3.6 V, $V_{cc}B = 3.0 \text{ V}$ - to 3.6 V
	Port 6 (P6PUE = 1)	_	3	_	70	μΑ	10 0.0 V
Input	RES (4)	C _{in}	_	_	80	pF	$V_{in} = 0 V$,
capacitance	NMI		_	_	50	pF	−f = 1 MHz, –T _a = 25°C
	P52, P97, P42, P86, PA7 to PA2		_	_	20	pF	– I _a = 25 C
	Input pins except (4) above	_	_	_	15	pF	_
Current	Normal operation	I _{cc}	_	45	60	mA	f = 10 MHz
dissipation*9	Sleep mode		_	35	50	mA	f = 10 MHz
	Standby mode*10	-	_	0.01	5.0	μA	T _a ≤ 50°C
			_	_	20.0	μΑ	50°C < T _a
Analog power	During A/D, D/A conversion	Al _{cc}	_	1.2	2.0	mA	
supply current	Idle	_	_	0.01	5.0	μΑ	$AV_{cc} = 2.0 \text{ V}$ to 5.5 V
Reference	During A/D conversion	Al _{ref}	_	0.5	1.0	mA	
power supply current	During A/D, D/A conversion	_	_	2.0	5.0	mA	_
	Idle	_	_	0.01	5.0	μΑ	$AV_{ref} = 2.0 \text{ V}$ to AV_{CC}
Analog power	er supply voltage*1	AV _{cc}	3.0	_	5.5	V	Operating
			2.0	_	5.5	V	Idle/not used
RAM standb	y voltage	V_{RAM}	2.0	_	_	V	
Notes 4 F	Do not looks the AVCC	AV 000	1 V/CC =	!	oven if th	A/D	

Notes: 1. Do not leave the AVCC, AV_{ref}, and AVSS pins open even if the A/D converter and D/A converter are not used.

Even if the A/D converter and D/A converter are not used, apply a value in the range 2.0 V to 5.5 V to AVCC and AV_{ref} pins by connection to the power supply (V_{cc}), or some other method. Ensure that AV_{ref} \leq AV_{cc}.

- 2. P67 to P60 include supporting module inputs multiplexed on those pins.
- 3. IRQ2 includes the ADTRG signal multiplexed on that pin.

- In the H8S/2147N, P52/SCK0/SCL0 and P97/SDA0 are NMOS push-pull outputs.
 An external pull-up resistor is necessary to provide high-level output from SCL0 and SDA0 (ICE = 1).
 - In the H8S/2147N, P52/SCK0 and P97 (ICE = 0) high levels are driven by NMOS.
- 5. When IICS = 0, ICE = 0, and KBIOE = 0. Low-level output when the bus drive function is selected is determined separately.
- 6. The upper limit of the port 6 applied voltage is V_{cc} +0.3 V when CIN input is not selected, and the lower of V_{cc} +0.3 V and AV_{cc} +0.3 V when CIN input is selected. When a pin is in output mode, the output voltage is equivalent to the applied voltage.
- 7. The upper limit of the port A applied voltage is $V_{cc}B + 0.3 \text{ V}$ when CIN input is not selected, and the lower of $V_{cc}B + 0.3 \text{ V}$ and $AV_{cc} + 0.3 \text{ V}$ when CIN input is selected. When a pin is in output mode, the output voltage is equivalent to the applied voltage.
- 8. The port A characteristics depend on $V_{cc}B$, and the other pins characteristics depend on V_{cc} .
- 9. Current dissipation values are for V_{IH} min = V_{cc} -0.5 V, V_{cc} B -0.5 V, and V_{IL} max = 0.5 V with all output pins unloaded and the on-chip pull-up MOSs in the off state.
- 10. The values are for $V_{\text{RAM}} \le V_{\text{CC}} < 3.0 \text{ V}$, $V_{\text{IH}} \min = V_{\text{CC}} \times 0.9$, $V_{\text{CC}} B \times 0.9$, and $V_{\text{IL}} \max = 0.3 \text{ V}$.
- 11. For flash memory program/erase operations, the applicable ranges are V_{cc} = 3.0 V to 3.6 V and Ta = 0 to +75°C.

Table 26.32 Permissible Output Currents

Conditions: $V_{cc} = 4.5 \text{ V}$ to 5.5 V, $V_{cc}B = 4.5 \text{ V}$ to 5.5 V, $V_{ss} = 0 \text{ V}$, $Ta = -20 \text{ to } +75^{\circ}\text{C}$

Item		Symbol	Min	Тур	Max	Unit
Permissible output low current (per pin)	SCL1, SCL0, SDA1, SDA0, PS2AC to PS2CC, PS2AD to PS2CD, PA7 to PA4 (bus drive function selected)	l _{oL}	_	_	20	mA
	Ports 1, 2, 3		_	_	10	mA
	RESO		_	_	3	mA
	Other output pins		_	_	2	mA
Permissible output	Total of ports 1, 2, and 3	\sum I _{ol}	_	_	80	mA
low current (total)	Total of all output pins, including the above	_	_	_	120	mA
Permissible output high current (per pin)	All output pins	–I _{OH}	_	_	2	mA
Permissible output high current (total)	Total of all output pins	$\Sigma - I_{OH}$	_	_	40	mA

Conditions: $V_{cc} = 3.0 \text{ V}$ to 5.5 V, $V_{cc}B = 3.0 \text{ V}$ to 5.5 V, $V_{ss} = 0 \text{ V}$, $T_a = -20 \text{ to } +75^{\circ}\text{C}$

Item		Symbol	Min	Тур	Max	Unit
Permissible output low current (per pin)	SCL1, SCL0, SDA1, SDA0, PS2AC to PS2CC, PS2AD to PS2CD, PA7 to PA4 (bus drive function selected)	I _{OL}	_	_	10	mA
	Ports 1, 2, 3		_	_	2	mA
	RESO	_	_	_	1	mA
	Other output pins	_	_	_	1	mA
Permissible output	Total of ports 1, 2, and 3	\sum I _{OL}	_	_	40	mA
low current (total)	Total of all output pins, including the above		_	_	60	mA
Permissible output high current (per pin)	All output pins	-I _{OH}	_	_	2	mA
Permissible output high current (total)	Total of all output pins	$\Sigma - \mathbf{I}_{OH}$	_	_	30	mA

Notes: 1. To protect chip reliability, do not exceed the output current values in table 26.32.

2. When driving a Darlington pair or LED, always insert a current-limiting resistor in the output line, as show in figures 26.1 and 26.2.



Table 26.33 Bus Drive Characteristics

Conditions: $V_{cc} = 3.0 \text{ V}$ to 5.5 V, $V_{ss} = 0 \text{ V}$

Applicable Pins: SCL1, SCL0, SDA1, SDA0 (bus drive function selected)

Item	Symbol	Min	Тур	Max	Unit	Test Conditions
Schmitt trigger	V _T	$V_{cc} \times 0.3$	_	_	V	V _{cc} = 3.0 V to 5.5 V
input voltage	V _T ⁺	_	_	$V_{cc} \times 0.7$	_	$V_{cc} = 3.0 \text{ V to } 5.5 \text{ V}$
	$V_T^+ - V_T^-$	$V_{cc} \times 0.05$	_	_		$V_{cc} = 3.0 \text{ V to } 5.5 \text{ V}$
Input high voltage	V _{IH}	$V_{cc} \times 0.7$	_	V _{cc} +0.5	V	V _{cc} = 3.0 V to 5.5 V
Input low voltage	V _{IL}	-0.5	_	$V_{cc} \times 0.3$		$V_{cc} = 3.0 \text{ V to } 5.5 \text{ V}$
Output low voltage	V _{oL}		_	0.8	V	I _{OL} = 16 mA,
						$V_{cc} = 4.5 \text{ V to } 5.5 \text{ V}$
		_	_	0.5	_	I _{OL} = 8 mA
		_	_	0.4		$I_{OL} = 3 \text{ mA}$
Input capacitance	C _{in}	_	_	20	pF	$V_{in} = 0 \text{ V}, f = 1 \text{ MHz},$
						$T_a = 25^{\circ}C$
Three-state leakage current (off state)	I _{TSI}	_	_	1.0	μΑ	$V_{in} = 0.5 \text{ to } V_{cc} - 0.5 \text{ V}$
SCL, SDA output fall time	t _{Of}	20 + 0.1 Cb	_	250	ns	$V_{cc} = 3.0 \text{ V to } 5.5 \text{ V}$

Conditions: $V_{cc} = 3.0 \text{ V}$ to 5.5 V, $V_{cc}B = 3.0 \text{ V}$ to 5.5 V, $V_{ss} = 0 \text{ V}$

Applicable Pins: PS2AC, PS2AD, PS2BC, PS2BD, PS2CC, PS2CD, PA7 to PA4 (bus drive

function selected)

Item	Symbol	Min	Тур	Max	Unit	Test Conditions
Output low voltage	V_{oL}	_	_	8.0	V	$I_{oL} = 16 \text{ mA},$ $V_{cc}B = 4.5 \text{ V to } 5.5 \text{ V}$
		_	_	0.5		$I_{OL} = 8 \text{ mA}$
		_	_	0.4		I _{oL} = 3 mA

26.4.3 AC Characteristics

Clock timing, control signal timing, bus timing, and timing of on-chip supporting modules list the following.

Figure 26.4 shows the test conditions for the AC characteristics.

(1) Clock Timing

Table 26.34 shows the clock timing. The clock timing specified here covers clock (φ) output and clock pulse generator (crystal) and external clock input (EXTAL pin) oscillation settling times. For details of external clock input (EXTAL pin and EXCL pin) timing, see section 24, Clock Pulse Generator.

Table 26.34 Clock Timing

Condition A: $V_{cc} = 5.0 \text{ V} \pm 10\%$, $V_{cc}B = 5.0 \text{ V} \pm 10\%$, $V_{ss} = 0 \text{ V}$, $\phi = 2 \text{ MHz}$ to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$

Condition B: $V_{cc} = 3.0 \text{ V}$ to 5.5 V, $V_{cc}B = 3.0 \text{ V}$ to 5.5 V, $V_{ss} = 0 \text{ V}$, $\phi = 2 \text{ MHz}$ to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$

		Condition A		Cor	ndition B			
		2	0 MHz	1	0 MHz	_	Test	
Item	Symbol	Min	Max	Min	Max	Unit	Conditions	
Clock cycle time	t _{cyc}	50	500	100	500	ns	Figure 26.5	
Clock high pulse width	t _{ch}	17	_	30	_	ns	Figure 26.5	
Clock low pulse width	t _{cL}	17	_	30	_	ns		
Clock rise time	t _{Cr}	_	8	_	20	ns		
Clock fall time	t _{Cf}	_	8	_	20	ns		
Oscillation settling time at reset (crystal)	t _{osc1}	10	_	20	_	ms	Figure 26.6 Figure 26.7	
Oscillation settling time in software standby (crystal)	t _{osc2}	8	_	8	_	ms		
External clock output stabilization delay time	t _{DEXT}	500	_	500	_	μs		



(2) Control Signal Timing

Table 26.35 shows the control signal timing. The only external interrupts that can operate on the subclock ($\phi = 32.768 \text{ kHz}$) are NMI and IRO0, 1, 2, 6, and 7.

Table 26.35 Control Signal Timing

Condition A: $V_{cc} = 5.0 \text{ V} \pm 10\%$, $V_{cc}B = 5.0 \text{ V} \pm 10\%$, $V_{ss} = 0 \text{ V}$, $\phi = 32.768 \text{ kHz}$, 2 MHz to maximum operating frequency, $T_a = -20 \text{ to} + 75^{\circ}\text{C}$

Condition B: $V_{cc} = 3.0 \text{ V}$ to 5.5 V, $V_{cc}B = 3.0 \text{ V}$ to 5.5 V, $V_{ss} = 0 \text{ V}$, $\phi = 32.768 \text{ kHz}$, 2 MHz to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$

		Con	dition A	Condition B				
		2	0 MHz	10 MHz		_	Test	
Item	Symbol	Min	Max	Min	Max	Unit	Conditions	
RES setup time	t _{RESS}	200	_	300	_	ns	Figure 26.8	
RES pulse width	t _{RESW}	20	_	20	_	t _{cyc}		
NMI setup time (NMI)	t _{nmis}	150	_	250	_	ns	Figure 26.9	
NMI hold time (NMI)	t _{nmih}	10	_	10	_	ns		
NMI pulse width (exiting software standby mode)	t _{nmiw}	200	_	200	_	ns		
IRQ setup time (IRQ7 to IRQ0)	t _{IRQS}	150	_	250	_	ns		
IRQ hold time (IRQ7 to IRQ0)	t _{IRQH}	10	_	10	_	ns		
IRQ pulse width (IRQ7, IRQ6, IRQ2 to IRQ0) (exiting software standby mode)	t _{IRQW}	200	_	200	_	ns	_	

(3) Bus Timing

Table 26.36 shows the bus timing. Operation in external expansion mode is not guaranteed when operating on the subclock ($\phi = 32.768 \text{ kHz}$).

Table 26.36 Bus Timing (1) (Nomal mode)

Condition A: $V_{cc} = 5.0 \text{ V} \pm 10\%$, $V_{cc}B = 5.0 \text{ V} \pm 10\%$, $V_{ss} = 0 \text{ V}$, $\phi = 2 \text{ MHz}$ to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$

Condition B: $V_{cc} = 3.0 \text{ V}$ to 5.5 V, $V_{cc}B = 3.0 \text{ V}$ to 5.5 V, $V_{ss} = 0 \text{ V}$, $\phi = 2 \text{ MHz}$ to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$

		Con	dition A	Condition B			
		20) MHz	10) MHz		Test
Item	Symbol	Min	Max	Min	Max	Unit	Conditions
Address delay time	t _{AD}	_	20	_	40	ns	Figure 26.10 to
Address setup time	t _{AS}	$0.5 \times t_{\text{cyc}} -15$	_	$\begin{array}{l} 0.5 \times \\ t_{_{cyc}} -30 \end{array}$	_	ns	figure 26.14
Address hold time	t _{AH}	$0.5 \times t_{\text{cyc}} -10$	_	$\begin{array}{l} 0.5 \times \\ t_{_{cyc}} -\!20 \end{array}$	_	ns	
CS delay time (IOS)	t _{CSD}	_	20	_	40	ns	
AS delay time	t _{ASD}	_	30	_	60	ns	
RD delay time 1	t _{RSD1}	_	30	_	60	ns	
RD delay time 2	t _{RSD2}	_	30	_	60	ns	
Read data setup time	t _{RDS}	15	_	35	_	ns	
Read data hold time	t _{RDH}	0	_	0	_	ns	

		Con	dition A	Con	dition B		
		20) MHz	10) MHz	_	Test
Item	Symbol	Min	Max	Min	Max	Unit	Conditions
Read data access time 1	t _{ACC1}	_	$1.0 \times t_{\rm cyc} -30$	_	1.0 × t _{cyc} –60	ns	Figure 26.10 to figure 26.14
Read data access time 2	t _{ACC2}	_	1.5 × t _{cyc} –25	_	1.5 × t _{cyc} –50	ns	
Read data access time 3	t _{ACC3}	_	2.0 × t _{cyc} –30	_	2.0 × t _{cyc} –60	ns	
Read data access time 4	t _{ACC4}	_	2.5 × t _{cyc} –25	_	2.5 × t _{cyc} –50	ns	
Read data access time 5	t _{ACC5}	_	$3.0 \times t_{\text{cyc}} -30$	_	3.0 × t _{cyc} –60	ns	
WR delay time 1	t _{WRD1}	_	30	_	60	ns	
WR delay time 2	t _{WRD2}	_	30	_	60	ns	
WR pulse width 1	t _{wsw1}	1.0 × t _{cyc} –20	_	1.0 × t _{cyc} –40	_	ns	
WR pulse width 2	t _{wsw2}	1.5 × t _{cyc} –20	_	1.5 × t _{cyc} –40	_	ns	
Write data delay time	t _{wdd}	_	30	_	60	ns	
Write data setup time	t _{wds}	0	_	0	_	ns	
Write data hold time	\mathbf{t}_{WDH}	10	_	20	_	ns	
WAIT setup time	t _{wrs}	30	_	60	_	ns	
WAIT hold time	t _{wth}	5	_	10	_	ns	

Table 26.36 Bus Timing (2) (Advanced mode)

Condition A: $V_{cc} = 5.0 \text{ V} \pm 10\%$, $V_{cc}B = 5.0 \text{ V} \pm 10\%$, $V_{ss} = 0 \text{ V}$, $\phi = 2 \text{ MHz}$ to maximum operating frequency, $T_a = -20 \text{ to} + 75^{\circ}\text{C}$

Condition B: $V_{cc} = 3.0 \text{ V}$ to 5.5 V, $V_{cc}B = 3.0 \text{ V}$ to 5.5 V, $V_{ss} = 0 \text{ V}$, $\phi = 2 \text{ MHz}$ to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$

		Con	dition A	Con	dition B			
		20) MHz	10) MHz	_	Test	
Item	Symbol	Min	Max	Min	Max	Unit	Conditions	
Address delay time	t _{AD}	_	30	_	60	ns	Figure 26.10 to	
Address setup time	t _{AS}	0.5 × t _{cyc} –25	_	$0.5 \times t_{\text{cyc}} -50$	_	ns	figure 26.14	
Address hold time	t _{AH}	0.5 × t _{cyc} –10	_	0.5 × t _{cyc} –20	_	ns		
CS delay time (IOS)	t _{CSD}	_	30	_	60	ns		
AS delay time	t _{ASD}	_	30	_	60	ns		
RD delay time 1	t _{RSD1}	_	30	_	60	ns		
RD delay time 2	t _{RSD2}	_	30		60	ns		
Read data setup time	t _{RDS}	15	_	35	_	ns		
Read data hold time	t _{RDH}	0	_	0	_	ns	<u> </u>	
Read data access time 1	t _{ACC1}	_	1.0 × t _{cyc} –40	_	1.0 × t _{cyc} –80	ns		
Read data access time 2	t _{ACC2}	_	$1.5 \times t_{\text{cyc}} -25$	_	$1.5 \times t_{\text{cyc}} -50$	ns		

		Con	dition A	Con	dition B			
	20 MHz 10 MHz) MHz	_	Test			
Item	Symbol	Min	Max	Min	Max	Unit	Conditions	
Read data access time 3	t _{ACC3}	_	$2.0 \times t_{\rm cyc}$ –40	_	$2.0 \times t_{cyc} -80$	ns	Figure 26.10 to figure 26.14	
Read data access time 4	t _{ACC4}	_	$2.5 \times t_{cyc}$ –25	_	$2.5 \times t_{\text{cyc}} -50$	ns		
Read data access time 5	t _{ACC5}	_	$3.0 \times t_{\text{cyc}} -40$	_	$3.0 \times t_{\rm cyc}$ -80	ns		
WR delay time 1	t _{WRD1}	_	30	_	60	ns		
WR delay time 2	t _{WRD2}	_	30	_	60	ns		
WR pulse width 1	t _{wsw1}	1.0 × t _{cyc} –20	_	1.0 × t _{cyc} –40		ns		
WR pulse width 2	t _{wsw2}	1.5 × t _{cyc} –20	_	1.5 × t _{cyc} –40	_	ns		
Write data delay time	t _{wdd}	_	30	_	60	ns		
Write data setup time	t _{wds}	0	_	0		ns		
Write data hold time	t _{wdh}	10	_	20		ns		
WAIT setup time	t _{wts}	30	_	60	_	ns	<u> </u>	
WAIT hold time	t _{wth}	5	_	10	_	ns		

(4) Timing of On-Chip Supporting Modules

Tables 26.37 to 26.39 show the on-chip supporting module timing. The only on-chip supporting modules that can operate in subclock operation ($\phi = 32.768 \text{ kHz}$) are the I/O ports, external interrupts (NMI and IRQ0, 1, 2, 6, and 7), the watchdog timer, and the 8-bit timer (channels 0 and 1).

Table 26.37 Timing of On-Chip Supporting Modules (1)

Condition A: $V_{cc} = 5.0 \text{ V} \pm 10\%$, $V_{cc}B = 5.0 \text{ V} \pm 10\%$, $V_{ss} = 0 \text{ V}$, $\phi = 32.768 \text{ kHz}^*$, 2 MHz to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$

Condition B: $V_{cc} = 3.0 \text{ V}$ to 5.5 V, $V_{cc}B = 3.0 \text{ V}$ to 5.5 V, $V_{ss} = 0 \text{ V}$, $\phi = 32.768 \text{ kHz}^*$, 2 MHz to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$

				Condition A Condit		ndition B			
				20 MHz		1	I0 MHz	_	Test
Item			Symbol	Min	Max	Min	Max	Unit	Conditions
I/O ports			t _{PWD}	_	50	_	100	ns	Figure 26.15
			t _{PRS}	30	_	50	_	_	
	Input dat time	Input data hold time		30	_	50	_	_	
FRT	Timer ou time	Timer output delay time		_	50	_	100	ns	Figure 26.16
	Timer inp	out setup	t _{FTIS}	30	_	50	_	_	
		Timer clock input setup time		30	_	50	_	_	Figure 26.17
	Timer clock	Single edge	t _{FTCWH}	1.5	_	1.5	_	t _{cyc}	_
	pulse width	Both edges	t _{FTCWL}	2.5	_	2.5	_	_	

				Co	ndition A	Coi	ndition B	_	
				2	20 MHz	1	0 MHz	_	Test
Item			Symbol	Min	Max	Min	Max	Unit	Conditions
TMR	Timer o		\mathbf{t}_{TMOD}	_	50	_	100	ns	Figure 26.18
	Timer r	eset input me	\mathbf{t}_{TMRS}	30	_	50	_		Figure 26.20
	Timer o	clock input me	t _{TMCS}	30	_	50	_		Figure 26.19
	Timer clock	Single edge	t _{TMCWH}	1.5	_	1.5	_	t _{cyc}	_
	pulse width	Both edges	t _{TMCWL}	2.5	_	2.5	_	_	
PWM, PWMX	Pulse output delay time		t _{PWOD}	_	50	_	100	ns	Figure 26.21
SCI	Input clock	Asynchro- nous	t _{Scyc}	4	_	4	_	t _{cyc}	Figure 26.22
	cycle	Synchro- nous	_	6	_	6	_	_	
	Input clock pulse width Input clock rise time		t _{sckw}	0.4	0.6	0.4	0.6	t _{Scyc}	
			t _{SCKr}	_	1.5	_	1.5	t _{cyc}	<u> </u>
	Input cl time	ock fall	t _{sckf}	_	1.5	_	1.5	_	
	Transmit data delay time (synchronous)		t _{TXD}	_	50	_	100	ns	Figure 26.23
		e data setup ynchronous)	t _{RXS}	50	_	100	_	ns	_
		e data hold ynchronous)	t _{RXH}	50	_	100	_	ns	_
A/D conver- ter		input setup	t _{TRGS}	30	_	50	_	ns	Figure 26.24
WDT	RESO time	output delay	t _{RESD}	_	100	_	200	ns	Figure 26.25
	RESO width	output pulse	t _{RESOW}	132	_	132	_	t _{cyc}	_

Note: * Only supporting modules that can be used in subclock operation

Table 26.37 Timing of On-Chip Supporting Modules (2)

Condition A: $V_{cc} = 5.0 \text{ V} \pm 10\%$, $V_{cc}B = 5.0 \text{ V} \pm 10\%$, $V_{ss} = 0 \text{ V}$, $\phi = 2 \text{ MHz}$ to maximum

operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$

Condition B: $V_{cc} = 3.0 \text{ V}$ to 5.5 V, $V_{cc}B = 3.0 \text{ V}$ to 5.5 V, $V_{ss} = 0 \text{ V}$, $\phi = 2 \text{ MHz}$ to maximum

operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$

				Condition A		Co	ndition B	l	
				2	0 MHz	1	0 MHz	_	Test
Item			Symbol	Min	Max	Min	Max	Unit	Conditions
HIF read cycle	CS/HA0 setup time		t _{HAR}	10	_	10	_	ns	Figure 26.26
	CS/HA	0 hold time	t _{HRA}	10	_	10	_	ns	_
	IOR pu	lse width	t _{HRPW}	120	_	220	_	ns	
	HDB de	elay time	t _{HRD}	_	100	_	200	ns	_
	HDB hold time		t _{HRF}	0	25	0	40	ns	
	HIRQ delay time		t _{HIRQ}	_	120	_	200	ns	_
HIF write cycle	CS/HA0 setup		t _{HAW}	10	_	10	_	ns	
	CS/HA	0 hold time	t _{HWA}	10	_	10	_	ns	_
	IOW pu	ulse width	t _{HWPW}	60	_	100	_	ns	_
	HDB setup time	Fast A20 gate not used	t _{HDW}	30	_	50	_	ns	_
		Fast A20 gate used	_	45	_	85	_	ns	
	HDB h	old time	t _{HWD}	15	_	25	_	ns	_
	GA20 (delay time	\mathbf{t}_{HGA}	_	90	_	180	ns	-

Table 26.38 Keyboard Buffer Controller Timing

Conditions: $V_{cc} = 3.0 \text{ V}$ to 5.5 V, $V_{cc}B = 3.0 \text{ V}$ to 5.5 V, $V_{ss} = 0 \text{ V}$, $\phi = 2 \text{ MHz}$ to maximum

operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$

		R	atings			Test	
Item	Symbol	Min	Тур	Max	Unit	Conditions	Notes
KCLK, KD output fall time	t _{KBF}	20 + 0.1 Cb	_	250	ns		Figure 26.27
KCLK, KD input data hold time	t _{KBIH}	150	_	_	ns		_
KCLK, KD input data setup time	t _{KBIS}	150	_	_	ns		_
KCLK, KD output delay time	t _{KBOD}	_	_	450	ns		_
KCLK, KD capacitive load	Сь	_	_	400	pF		_

Table 26.39 I²C Bus Timing

Conditions: $V_{cc} = 3.0 \text{ V}$ to 5.5 V, $V_{ss} = 0 \text{ V}$, $\phi = 5 \text{ MHz}$ to maximum operating frequency

			Ratings				
Item	Symbol	Min	Тур	Max	Unit	Test Conditions	Notes
SCL input cycle time	t _{SCL}	12	_	_	t _{cyc}		Figure 26.28
SCL input high pulse width	t _{SCLH}	3	_	_	t _{cyc}		_
SCL input low pulse width	t _{SCLL}	5	_	_	t _{cyc}		_
SCL, SDA input rise time	t _{Sr}	_	_	7.5*	t _{cyc}		_
SCL, SDA input fall time	t _{Sf}		_	300	ns		_
SCL, SDA input spike pulse elimination time	t _{sp}	_	_	1	t _{cyc}		_
SDA input bus free time	t _{BUF}	5	_	_	t _{cyc}		
Start condition input hold time	t _{STAH}	3	_	_	t _{cyc}		_
Retransmission start condition input setup time	t _{STAS}	3	_	_	t _{cyc}		
Stop condition input setup time	t _{stos}	3	_	_	t _{cyc}		
Data input setup time	t _{SDAS}	0.5	_	_	t _{cyc}		_
Data input hold time	t _{SDAH}	0	_	_	ns		_
SCL, SDA capacitive load	C _b	_	_	400	pF		

Note: * 17.5t_{cyc} can be set according to the clock selected for use by the I²C module. For details, see section 16.4, Usage Notes.

26.4.4 A/D Conversion Characteristics

Tables 26.40 and 26.41 list the A/D conversion characteristics.

Table 26.40 A/D Conversion Characteristics (AN7 to AN0 Input: 134/266-State Conversion)

Condition A:
$$V_{cc} = 5.0 \text{ V} \pm 10\%$$
, $AV_{cc} = 5.0 \text{ V} \pm 10\%$, $AV_{ref} = 4.5 \text{ V}$ to AV_{cc} , $V_{ss} = AV_{ss} = 0 \text{ V}$, $\phi = 2 \text{ MHz}$ to maximum operating frequency, $T_{c} = -20 \text{ to } +75^{\circ}\text{C}$

Condition B:
$$V_{cc}=3.0~V$$
 to 5.5 V, $AV_{cc}=3.0~V$ to 5.5 V, $AV_{ref}=3.0~V$ to AV_{cc} , $V_{ss}=AV_{ss}=0~V$, $\varphi=2~MHz$ to maximum operating frequency, $T_a=-20$ to $+75^{\circ}C$

		Condition	on A		Condition		
		20 MF	lz		10 MF	<u> </u>	
Item	Min	Тур	Max	Min	Тур	Max	Unit
Resolution	10	10	10	10	10	10	Bits
Conversion time*5	_	_	6.7	_	_	13.4	μs
Analog input capacitance	_	_	20	_	_	20	pF
Permissible signal- source impedance	_	_	10 ^{*3} 5 ^{*4}		_	10*1 5*2	kΩ
Nonlinearity error	_	_	±3.0	_	_	±7.0	LSB
Offset error	_	_	±3.5	_	_	±7.5	LSB
Full-scale error	_	_	±3.5	_	_	±7.5	LSB
Quantization error	_	_	±0.5	_	_	±0.5	LSB
Absolute accuracy	_	_	±4.0	_	_	±8.0	LSB

Notes: 1. When $4.0 \text{ V} \le \text{AV}_{cc} \le 5.5 \text{ V}$

- 2. When $3.0 \text{ V} \le \text{AV}_{cc} < 4.0 \text{ V}$
- 3. When conversion time \geq 11. 17 μs (CKS = 1 and $\varphi \leq$ 12 MHz, or CKS = 0)
- 4. When conversion time < 11. 17 μ s (CKS = 1 and ϕ > 12 MHz)
- 5. In single mode and ϕ = maximum operating frequency.

Table 26.41 A/D Conversion Characteristics (CIN15 to CIN0 Input: 134/266-State Conversion)

Condition A: $V_{cc} = 5.0 \text{ V} \pm 10\%$, $AV_{cc} = 5.0 \text{ V} \pm 10\%$, $AV_{ref} = 4.5 \text{ V}$ to AV_{cc} , $V_{ss} = AV_{ss} = 0 \text{ V}$, $\phi = 2 \text{ MHz}$ to maximum operating frequency, $T = -20 \text{ to } +75^{\circ}\text{C}$

Condition B: $V_{cc} = 3.0 \text{ V}$ to 5.5 V, $AV_{cc} = 3.0 \text{ V}$ to 5.5 V, $AV_{ref} = 3.0 \text{ V}$ to AV_{cc} , $V_{ss} = AV_{ss} = 0 \text{ V}$, $\phi = 2 \text{ MHz}$ to maximum operating frequency, $T_{c} = -20 \text{ to } +75^{\circ}\text{C}$

		Condition	on A		Condition	on B		
20 MHz				10 MF	łz	_		
Item	Min	Тур	Max	Min	Тур	Max	Unit	
Resolution	10	10	10	10	10	10	Bits	
Conversion time*5	_	_	6.7	_	_	13.4	μs	
Analog input capacitance	_	_	20	_	_	20	pF	
Permissible signal- source impedance	_	_	10 ^{*3} 5 ^{*4}		_	10*1 5*2	kΩ	
Nonlinearity error	_	_	±5.0	_	_	±11.0	LSB	
Offset error	_	_	±5.5	_	_	±11.5	LSB	
Full-scale error	_	_	±5.5	_	_	±11.5	LSB	
Quantization error	_	_	±0.5	_	_	±0.5	LSB	
Absolute accuracy	_	_	±6.0	_	_	±12.0	LSB	

Notes: 1. When 4.0 V \leq AV_{cc} \leq 5.5 V

- 2. When $3.0 \text{ V} \le \text{AV}_{cc} < 4.0 \text{ V}$
- 3. When conversion time \geq 11. 17 µs (CKS = 1 and $\phi \leq$ 12 MHz, or CKS = 0)
- 4. When conversion time < 11. 17 μ s (CKS = 1 and ϕ > 12 MHz)
- 5. In single mode and ϕ = maximum operating frequency.

26.4.5 D/A Conversion Characteristics

Table 26.42 lists the D/A conversion characteristics.

Table 26.42 D/A Conversion Characteristics

Condition A: $V_{cc} = 5.0 \text{ V} \pm 10\%$, $AV_{cc} = 5.0 \text{ V} \pm 10\%$, $AV_{ref} = 4.5 \text{ V}$ to AV_{cc} , $V_{ss} = AV_{ss} = 0 \text{ V}$, $\phi = 2 \text{ MHz}$ to maximum operating frequency,

 $T_a = -20$ to +75°C (regular specifications), $T_a = -40$ to +85°C (wide-range specifications)

Condition B: $V_{cc} = 3.0 \text{ V to } 5.5 \text{ V}, \text{AV}_{cc} = 3.0 \text{ V to } 5.5 \text{ V}, \text{AV}_{ref} = 3.0 \text{ V to AV}_{cc},$ $V_{ss} = \text{AV}_{ss} = 0 \text{ V}, \phi = 2 \text{ MHz to maximum operating frequency},$ $T_{s} = -20 \text{ to } +75^{\circ}\text{C}$

			Condition	on A		Condition	on B		
			20 MF	łz		10 MF	łz		
Item		Min	Тур	Max	Min	Тур	Max	Unit	
Resolution		8	8	8	8	8	8	Bits	
Conversion time	With 20-pF load capacitance	_	_	10	_	_	10	μs	
Absolute accuracy	With 2-MΩ load resistance	_	±1.0	±1.5	_	±2.0	±3.0	LSB	
	With 4-MΩ load resistance	_	_	±1.0	_	_	±2.0		

26.4.6 Flash Memory Characteristics

Table 26.43 shows the flash memory characteristics.

Table 26.43 Flash Memory Characteristics (Programming/erasing operating range)

Test

Conditions (5 V version): $V_{CC} = 5.0 \text{ V} \pm 10\%$, $V_{SS} = 0 \text{ V}$, $T_a = 0 \text{ to } +75^{\circ}\text{C}$ (3 V version): $V_{CC} = 3.0 \text{ V}$ to 3.6V, $V_{SS} = 0 \text{ V}$, $T_a = 0 \text{ to } +75^{\circ}\text{C}$

Item		Symbol	Min	Тур	Max	Unit	Test Condition
Programming	time ^{*1 *2 *4}	tP	_	10	200	ms/ 32 bytes	
Erase time*1*	3 *6	tE	_	100	1200	ms/ block	
Reprogrammir	ng count	N _{wec}	100*8	10000*9	_	Times	
Data retention	time ^{*10}	t _{DRP}	10	_	_	Years	
Programming	Wait time after SWE-bit setting*1	х	10	_	_	μs	
	Wait time after PSU-bit setting*1	у	50	_	_	μs	
	Wait time after P-bit setting*1*4	Z	150	_	200	μs	
	Wait time after P-bit clear*1	α	10	_	_	μs	
	Wait time after PSU-bit clear*1	β	10	_	_	μs	
	Wait time after PV-bit setting*1	γ	4	_	_	μs	
	Wait time after dummy write*1	ε	2	_	_	μs	
	Wait time after PV-bit clear*1	η	4	_	_	μs	
	Maximum programming count*1 *4 *5	N	_	_	1000	Times	z = 200 µs
Erase	Wait time after SWE-bit setting*1	х	10	_	_	μs	
	Wait time after ESU-bit setting*1	у	200	_	_	μs	
	Wait time after E-bit setting*1 *6	Z	5	_	10	ms	
	Wait time after E-bit clear*1	α	10	_	_	μs	
	Wait time after ESU-bit clear*1	β	10	_	_	μs	
	Wait time after EV-bit setting*1	γ	20	_	_	μs	
	Wait time after dummy write*1	ε	2	_	_	μs	
	Wait time after EV-bit clear*1	η	5	_	_	μs	
	Maximum erase count*1 *6 *7	N	_	_	120	Times	z = 10 ms

Notes: 1. Set the times according to the program/erase algorithms.

2. Programming time per 32 bytes (Shows the total period for which the P-bit in FLMCR1 is set. It does not include the programming verification time.)



- 3. Block erase time (Shows the total period for which the E-bit in FLMCR1 is set. It does not include the erase verification time.)
- 4. Maximum programming time (tP (max)) $tP(max) = wait time after P-bit setting (z) \times maximum programming count (N)$
- Number of times when the wait time after P-bit setting (z) = 200 μs.
 The number of writes should be set according to the actual set value of (z) to allow programming within the maximum programming time (tP(max)).
- Maximum erase time (tE (max)) tE(max) = wait time after E-bit setting (z) × maximum erase count (N)
- 7. Number of times when the wait time after E-bit setting (z) = 10 ms.

 The number of erases should be set according to the actual set value of (z) to allow erasing within the maximum erase time (tE(max)).
- 8. Minimum number of times for which all characteristics are guaranteed after rewriting (Guarantee range is 1 to minimum value).
- 9. Reference value for 25°C (as a guideline, rewriting should normally function up to this value).
- 10. Data retention characteristic when rewriting is performed within the specification range, including the minimum value.

26.4.7 Usage Note

- (1) The F-ZTAT and mask ROM versions have been confirmed as fully meeting the reference values for electrical characteristics shown in this manual. However, actual performance figures, operating margins, noise margins, and other properties may vary due to differences in the manufacturing process, on-chip ROM, layout patterns, etc.
 - When system evaluation testing is carried out using the F-ZTAT version, the same evaluation tests should also be conducted for the mask ROM version when changing over to that version.
- (2) On-chip power supply step-down circuit
 - The H8S/2147N F-ZTAT does not incorporate an internal power supply step-down circuit. When changing over to F-ZTAT versions or mask ROM versions incorporating an internal step-down circuit, the VCC2 pin has the same pin location as the VCL pin in a step-down circuit (See figure 26.3).
 - Therefore, note that the circuit patterns differ between these two types of products.

26.5 Electrical Characteristics of H8S/2144 F-ZTAT, H8S/2142 F-ZTAT, and Mask ROM Version of H8S/2142

26.5.1 Absolute Maximum Ratings

Table 26.44 lists the absolute maximum ratings.

Table 26.44 Absolute Maximum Ratings

Item	Symbol	Value	Unit
Power supply voltage*	V _{cc}	-0.3 to +7.0	V
Input voltage (except ports 6, 7, and A)	V _{in}	-0.3 to V _{cc} +0.3	V
Input voltage (CIN input not selected for port 6 and A)	V _{in}	-0.3 to V _{cc} +0.3	V
Input voltage (CIN input selected for port 6 and A)	V _{in}	–0.3 V to lower of voltages $\rm V_{cc}$ +0.3 and $\rm AV_{cc}$ +0.3	V
Input voltage (port 7)	V _{in}	-0.3 to AV _{cc} +0.3	V
Reference supply voltage	AV_{ref}	-0.3 to AV _{cc} +0.3	V
Analog power supply voltage	AV _{cc}	-0.3 to +7.0	V
Analog input voltage	V _{AN}	-0.3 to AV _{cc} +0.3	V
Operating temperature	T _{opr}	Regular specifications: -20 to +75	°C
		Wide-range specifications: -40 to +85	°C
Operating temperature (flash	T _{opr}	Regular specifications: 0 to +75	°C
memory programming/erasing)		Wide-range specifications: 0 to +85	°C
Storage temperature	T _{stg}	-55 to +125	°C

Caution: Permanent damage to the chip may result if absolute maximum ratings are exceeded.

Note: * Power supply voltage for VCC1 and VCC2 pins.

DC Characteristics 26.5.2

Table 26.45 lists the DC characteristics. Permitted output current values and bus drive characteristics are shown in tables 26.46 and 26.47, respectively.

Table 26.45 DC Characteristics (1)

$$\begin{array}{ll} \text{Conditions:} & \text{$V_{\text{CC}} = 5.0 \text{ V} \pm 10\%, \text{AV}_{\text{CC}}$}^{*1} = 5.0 \text{ V} \pm 10\%, \text{AV}_{\text{ref}}$}^{*1} = 4.5 \text{ V to AV}_{\text{CC}}$,} \\ & \text{$V_{\text{SS}} = \text{AV}_{\text{SS}}$}^{*1} = 0 \text{ V}, \text{$T_{\text{a}} = -20 \text{ to } +75^{\circ}\text{C}*8 (regular specifications),} \\ & \text{$T_{\text{a}} = -40 \text{ to } +85^{\circ}\text{C}*8 (wide-range specifications)} \\ \end{array}$$

Item			Symbol	Min	Тур	Max	Unit	Test Conditions
Schmitt	P67 to P60*2*5,	(1)	V _T -	1.0	_	_	V	
trigger input voltage	KIN15 to KIN8*5, IRQ2 to IRQ0*3,		$V_{T}^{ *}$	_	_	$V_{cc} \times 0.7$	V	_
	ĪRQ5 to ĪRQ3		$V_T^+ - V_T^-$	0.4	_	_	V	_
Input high voltage	RES, STBY, NMI, MD1, MD0	(2)	V _{IH}	V _{cc} -0.7	_	V _{cc} +0.3	V	
	EXTAL, PA7 to PA0*5	-		$V_{cc} \times 0.7$	_	V _{cc} +0.3	V	_
	Port 7	-		2.0	_	AV _{cc} +0.3	V	_
	Input pins except (1) and (2) above		_	2.0	_	V _{cc} +0.3	V	_
Input low voltage	RES, STBY, MD1, MD0	(3)	V _{IL}	-0.3	_	0.5	V	
	PA7 to PA0	-		-0.3	_	1.0	V	_
	NMI, EXTAL, input pins except (1) and (3) above		_	-0.3	_	0.8	V	_
	All output pins*4		V _{OH}	V _{cc} -0.5	_	_	V	$I_{OH} = -200 \mu A$
voltage				3.5	_	_	V	$I_{OH} = -1 \text{ mA}$
Output low voltage	All output pins (except RESO)*4		V_{oL}	_	_	0.4	V	I _{oL} = 1.6 mA
	Ports 1 to 3		=	_	_	1.0	V	I _{oL} = 10 mA
	RESO		_	_	_	0.4	V	$I_{OL} = 2.6 \text{ mA}$

Item			Symbol	Min	Тур	Max	Unit	Test Conditions
Input leakage	RES		I _{in}	_	_	10.0	μΑ	$V_{in} = 0.5 \text{ to} $ $V_{cc} - 0.5 \text{ V}$
current	STBY, NMI, MD1, MD0		-	_	_	1.0	μΑ	_
	Port 7		_	_	_	1.0	μΑ	$V_{in} = 0.5 \text{ to}$ AV _{CC} -0.5 V
Three-state leakage current (off state)	Ports 1 to 6, 8, 9, A, B		I _{TSI}	_	_	1.0	μΑ	$V_{in} = 0.5 \text{ to} $ $V_{cc} = -0.5 \text{ V}$
Input	Ports 1 to 3		-I _P	50	_	300	μΑ	$V_{in} = 0 V$
pull-up MOS current	Ports 6, A, B		_	60	_	500	μA	_
Input		(4)	C _{in}	_	_	80	pF	$V_{in} = 0 V$,
capacitance	NMI			_	_	50	pF	⁻f = 1 MHz, _Tॢ = 25°C
	P52, P97, P42, P86 PA7 to PA2			_	_	20	pF	- I _a - 25 0
	Input pins except (4) above		_	_	_	15	pF	_
Current	Normal operation		I _{cc}	_	75	100	mA	f = 20 MHz
dissipation*6	Sleep mode		=	_	60	85	mA	
	Standby mode*7		_	_	0.01	5.0	μΑ	T _a ≤ 50°C
				_	_	20.0	μΑ	$50^{\circ}\text{C} < \text{T}_{a}$
Analog power	During A/D, D/A conversion		Al_cc	_	1.2	2.0	mA	
supply current	Idle			_	0.01	5.0	μΑ	AV _{cc} = 2.0 V to 5.5 V
Reference power supply current	During A/D conver	sion	Al _{ref}	_	0.5	1.0	mA	
	During A/D, D/A conversion		rer	_	2.0	5.0	mA	_
	Idle		_	_	0.01	5.0	μΑ	AV _{ref} = 2.0 V to AV _{CC}

Item	Symbol	Min	Тур	Max	Unit	Test Conditions
Analog power supply voltage*1	AV_{cc}	4.5	_	5.5	V	Operating
		2.0	_	5.5	V	Idle/not used
RAM standby voltage	$V_{\scriptscriptstyle{RAM}}$	2.0	_	_	V	

Notes: 1. Do not leave the AVCC, AV_{ref}, and AVSS pins open even if the A/D converter and D/A converter are not used.

Even if the A/D converter and D/A converter are not used, apply a value in the range 2.0 V to 5.5 V to AVCC and AV_{ref} pins by connection to the power supply (V_{cc}), or some other method. Ensure that AV_{ref} \leq AV_{cc}.

- 2. P67 to P60 include supporting module inputs multiplexed on those pins.
- 3. IRQ2 includes the ADTRG signal multiplexed on that pin.
- 4. When IICS = 0. Low-level output when the bus drive function is selected is determined separately.
- 5. The upper limit of the applied voltage on Ports 6 and A is V_{cc} +0.3 V when CIN input is not selected, and the lower of V_{cc} +0.3 V and AV_{cc} +0.3 V when CIN input is selected. When a pin is in output mode, the output voltage is equivalent to the applied voltage.
- 6. Current dissipation values are for $V_{\rm IH}$ min = $V_{\rm cc}$ –0.5 V, and $V_{\rm IL}$ max = 0.5 V with all output pins unloaded and the on-chip pull-up MOSs in the off state.
- 7. The values are for $V_{RAM} \le V_{CC} < 4.5 \text{V}$, $V_{IH} \text{ min} = V_{CC} \times 0.9$, and $V_{IL} \text{ max} = 0.3 \text{ V}$.
- 8. For flash memory program/erase operations, the applicable range is $T_a = 0$ to +75°C (regular specifications) or $T_a = 0$ to +85°C (wide-range specifications).

Table 26.45 DC Characteristics (2)

Conditions: $V_{cc} = 4.0 \text{ V to } 5.5 \text{ V}^{*8}, \text{ AV}_{cc}^{*1} = 4.0 \text{ V to } 5.5 \text{ V}, \text{ AV}_{ref}^{*1} = 4.0 \text{ V to AV}_{cc},$ $V_{ss} = \text{AV}_{ss}^{*1} = 0 \text{ V}, \text{ T}_{a} = -20 \text{ to } +75^{\circ}\text{C}^{*8} \text{ (regular specifications)},$ $T_{a} = -40 \text{ to } +85^{\circ}\text{C}^{*8} \text{ (wide-range specifications)}$

Item			Symbol	Min	Тур	Max	Unit	Test Conditions
Schmitt	P67 to P60*2*5,	(1)	V _T	1.0	_	_	V	$V_{cc} = 4.5 \text{ V to}$
trigger input voltage	KIN15 to KIN8*5, IRQ2 to IRQ0*3,		V _T ⁺	_	_	$V_{cc} \times 0.7$	V	_5.5 V
voitage	IRQ5 to IRQ3		$V_T^+ - V_T^-$	0.4	_	_	V	_
			V _T	0.8	_	_	V	V_{cc} = 4.0 V to
			V _T ⁺	_	_	$V_{cc} \times 0.7$	V	_4.5 V
			$V_T^+ - V_T^-$	0.3	_	_	V	
Input high voltage	RES, STBY, NMI, MD1, MD0	(2)	V _{IH}	V _{cc} -0.7	_	V _{cc} +0.3	V	_
	EXTAL, PA7 to PA0*5	-		$V_{cc} \times 0.7$	_	V _{cc} +0.3	V	_
	Port 7	-		2.0	_	AV _{cc} +0.3	V	_
	Input pins except (1) and (2) above		_	2.0	_	V _{cc} +0.3	V	_
Input low voltage	RES, STBY, MD1, MD0	(3)	V _{IL}	-0.3	_	0.5	V	
	PA7 to PA0	-		-0.3	_	1.0	V	V _{cc} = 4.5 V to 5.5 V
				-0.3	_	0.8	V	$V_{cc} = 4.0 \text{ V to} $ 4.5 V
	NMI, EXTAL, input pins except (1) and (3) above		_	-0.3	_	0.8	V	
	All output pins*4		V _{OH}	V _{cc} -0.5	_	_	V	$I_{OH} = -200 \mu A$
voltage				3.5		_	V	$I_{OH} = -1 \text{ mA},$ $V_{CC} = 4.5 \text{ V to}$ 5.5 V
				3.0	_	_	V	$I_{OH} = -1 \text{ mA},$ $V_{CC} = 4.0 \text{ V to}$ 4.5 V

Item			Symbol	Min	Тур	Max	Unit	Test Conditions
Output low voltage	All output pins (except RESO)*4		V _{OL}	_	_	0.4	V	$I_{OL} = 1.6 \text{ mA}$
	Ports 1 to 3		_	_	_	1.0	V	I _{oL} = 10 mA
	RESO		_	_	_	0.4	V	I _{oL} = 2.6 mA
Input	RES	I _{in}		_	_	10.0	μA	$V_{in} = 0.5 \text{ to}$
leakage current	STBY, NMI, MD1, MD0			_	_	1.0	μΑ	[−] V _{cc} –0.5 V
	Port 7			_	_	1.0	μΑ	$V_{in} = 0.5 \text{ to}$ $AV_{cc} - 0.5 \text{ V}$
Three-state leakage current (off state)	Ports 1 to 6, 8, 9, A, B		I _{TSI}	_	_	1.0	μА	$V_{in} = 0.5 \text{ to}$ $V_{cc} = -0.5 \text{ V}$
Input	Ports 1 to 3		-I _P	50	_	300	μA	$V_{in} = 0 V$,
pull-up MOS	Ports 6, A, B			60	_	500	μΑ	$V_{cc} = 4.5 \text{ V to}$ 5.5 V
current	Ports 1 to 3		-	30	_	200	μΑ	$V_{in} = 0 V,$
	Ports 6, A, B			40	_	400	μΑ	$V_{cc} = 4.0 \text{ V to}$ 4.5 V
Input	RES	(4)	C_{in}	_	_	80	pF	$V_{in} = 0 V$,
capacitance	NMI			_	_	50	pF	$^{-}$ f = 1 MHz, $_{-}$ T $_{a}$ = 25°C
	P52, P97, P42, P86, PA7 to PA2			_	_	20	pF	_ 1 _a _ 2 5
	Input pins except (4) above		_	_	_	15	pF	_
Current	Normal operation		I _{cc}	_	65	85	mA	f = 16 MHz
dissipation*6	Sleep mode		_	_	50	70	mA	f = 16 MHz
	Standby mode*7		_	_	0.01	5.0	μΑ	$T_a \le 50^{\circ}C$
				_	_	20.0	μΑ	50°C < T _a
Analog power	During A/D, D/A conversion		Al _{cc}	_	1.2	2.0	mA	
supply current	Idle		_	_	0.01	5.0	μΑ	$AV_{cc} = 2.0 \text{ V}$ to 5.5 V

Item		Symbol	Min	Тур	Max	Unit	Test Conditions
Reference	During A/D conversion	Al _{ref}	_	0.5	1.0	mA	
power supply current	During A/D, D/A conversion	_	_	2.0	5.0	mA	_
	Idle	=	_	0.01	5.0	μΑ	$AV_{ref} = 2.0 V$ to AV_{cc}
Analog power supply voltage*1		AV_cc	4.0	_	5.5	V	Operating
			2.0	_	5.5	V	Idle/not used
RAM standby voltage		$V_{\scriptscriptstyle{RAM}}$	2.0	_	_	V	

Notes: 1. Do not leave the AVCC, AV_{ref}, and AVSS pins open even if the A/D converter and D/A converter are not used.

Even if the A/D converter and D/A converter are not used, apply a value in the range 2.0 V to 5.5 V to AVCC and AV_{ref} pins by connection to the power supply (V_{cc}), or some other method. Ensure that AV_{ref} \leq AV_{cc}.

- 2. P67 to P60 include supporting module inputs multiplexed on those pins.
- 3. IRQ2 includes the ADTRG signal multiplexed on that pin.
- 4. When IICS = 0. Low-level output when the bus drive function is selected is determined separately.
- 5. The upper limit of the applied voltage on Ports 6 and A is V_{cc} +0.3 V when CIN input is not selected, and the lower of V_{cc} +0.3 V and AV_{cc} +0.3 V when CIN input is selected. When a pin is in output mode, the output voltage is equivalent to the applied voltage.
- 6. Current dissipation values are for V_{IH} min = V_{CC} –0.5 V, and V_{IL} max = 0.5 V with all output pins unloaded and the on-chip pull-up MOSs in the off state.
- 7. The values are for $V_{RAM} \le V_{CC} < 4.0 \text{ V}$, $V_{IH} \text{ min} = V_{CC} \times 0.9$, and $V_{II} \text{ max} = 0.3 \text{ V}$.
- 8. For flash memory program/erase operations, the applicable ranges are V_{cc} = 4.5 V to 5.5 V and T_a = 0 to +75°C (regular specifications) or T_a = 0 to +85°C (wide-range specifications).



Table 26.45 DC Characteristics (3)

Conditions (Mask ROM version): $V_{cc} = 2.7 \text{ V}$ to 5.5 V, $AV_{cc}^{*1} = 2.7 \text{ V}$ to 5.5 V, $AV_{ref}^{*1} = 2.7 \text{ V}$ to 5.5 V, $AV_{ref}^{*1} = 2.7 \text{ V}$ to 5.5 V, $AV_{cc}^{*1} = 0$ V,

(Flash memory version): $V_{_{CC}} = 3.0 \text{ V to } 5.5 \text{ V}^{**}, \text{AV}_{_{CC}}^{**} = 3.0 \text{ V to } 5.5 \text{ V}, \text{AV}_{_{ref}} = 3.0 \text{ V}$ to $5.5 \text{ V}, \text{AV}_{_{ref}} = 3.0 \text{ V}$ to $5.5 \text{ V}, \text{V}_{_{SS}} = \text{AV}_{_{SS}}^{**} = 0 \text{ V}, \text{T}_{_{a}} = -20 \text{ to } +75^{\circ}\text{C}^{**}$

Item			Symbol	Min	Тур	Max	Unit	Test Conditions
Schmitt	P67 to P60*2*5,	(1)	V _T	$V_{cc} \times 0.2$	_	_	V	
trigger input voltage	KIN15 to KIN8*5, IRQ2 to IRQ0*3,		V _T ⁺	_	_	$V_{cc} \times 0.7$	V	_
voltage	IRQ5 to IRQ3		$V_T^+ - V_T^-$	$V_{cc} \times 0.05$	_	_	V	_
Input high voltage	RES, STBY, NMI, MD1, MD0	(2)	V _{IH}	$V_{cc} \times 0.9$	_	V _{cc} +0.3	V	
	EXTAL, PA7 to PA0*5	-		$V_{cc} \times 0.7$	_	V _{cc} +0.3	V	_
	Port 7	-		$V_{cc} \times 0.7$	_	AV _{cc} +0.3	V	
	Input pins except (1) and (2) above		_	$V_{cc} \times 0.7$	_	V _{cc} +0.3	V	_
Input low voltage	RES, STBY, MD1, MD0	(3)	V _{IL}	-0.3	_	$V_{cc} \times 0.1$	V	
	PA7 to PA0	-		-0.3	_	$V_{cc} \times 0.2$	V	V _{cc} < 4.0 V
						0.8	V	$V_{cc} = 4.0 \text{ V to}$ 5.5 V
	NMI, EXTAL,		=	-0.3	_	$V_{cc} \times 0.2$	V	V _{cc} < 4.0 V
	input pins except (1) and (3) above					0.8	V	V _{cc} = 4.0 V to 5.5 V
	All output pins*4		V _{OH}	V _{cc} -0.5	_	_	V	$I_{OH} = -200 \mu A$
voltage				V _{cc} -1.0	_	_	V	$I_{OH} = -1 \text{ mA}$ $(V_{CC} < 4.0 \text{ V})$

Item			Symbol	Min	Тур	Max	Unit	Test Conditions
Output low voltage	All output pins (except RESO)*4		V _{OL}	_	_	0.4	V	I _{oL} = 1.6 mA
	Ports 1 to 3		_	_	_	1.0	V	$\begin{split} I_{\text{OL}} &= 5 \text{ mA} \\ (V_{\text{CC}} < 4.0 \text{ V}), \\ I_{\text{OL}} &= 10 \text{ mA} \\ (4.0 \text{ V} \le V_{\text{CC}} \le 5.5 \text{ V}) \end{split}$
	RESO			_	_	0.4	V	$I_{OL} = 1.6 \text{ mA}$
Input	RES		I _{in}	_	_	10.0	μΑ	$V_{in} = 0.5 \text{ to}$
leakage current	STBY, NMI, MD1, MD0			_	_	1.0	μΑ	¯V _{cc} –0.5 V
	Port 7			_	_	1.0	μΑ	$V_{in} = 0.5 \text{ to}$ $AV_{cc} - 0.5 \text{ V}$
Three-state leakage current (off state)	Ports 1 to 6, 8, 9, A, B		I _{TSI}	_	_	1.0	μА	$V_{in} = 0.5 \text{ to}$ $V_{cc} - 0.5 \text{ V}$
Input pull-	Ports 1 to 3		-I _P	10	_	150	μΑ	V _{in} = 0 V,
up MOS current	Ports 6, A, B			30	_	250	μΑ	$V_{cc} = 2.7 \text{ V}^{*8}$ to 3.6 V
Input capacitance	RES	(4)	C_{in}	_	_	80	pF	$V_{in} = 0 V,$ f = 1 MHz,
	NMI			_	_	50	pF	$T_a = 25^{\circ}C$
	P52, P97, P42, P86, PA7 to PA2			_	_	20	pF	_
	Input pins except (4) above		_	_	_	15	pF	_
Current	Normal operation		I_{cc}	_	45	60	mA	f = 10 MHz
dissipation*6	Sleep mode			_	35	50	mA	_
	Standby mode*7			_	0.01	5.0	μA	T _a ≤ 50°C
				_	_	20.0	μΑ	50°C < T _a
Analog power	During A/D, D/A conversion		Al _{cc}	_	1.2	2.0	mA	
supply current	Idle			_	0.01	5.0	μΑ	AV _{cc} = 2.0 V o 5.5 V

Item		Symbol	Min	Тур	Max	Unit	Test Conditions
Reference	During A/D conversion	Al _{ref}	_	0.5	1.0	mA	
power supply current	During A/D, D/A conversion	_	_	2.0	5.0	mA	_
	Idle	-	_	0.01	5.0	μΑ	$AV_{ref} = 2.0 \text{ V}$ to AV_{cc}
Analog power supply voltage*1		AV _{cc}	2.7	_	5.5	V	Operating (mask ROM version)
			3.0	_	5.5	V	Operating (F-ZTAT version)
			2.0	_	5.5	V	Idle/not used
RAM stand	oy voltage	V_{RAM}	2.0	_	_	V	

Notes: 1. Do not leave the AVCC, AV_{ref}, and AVSS pins open even if the A/D converter and D/A converter are not used.

Even if the A/D converter and D/A converter are not used, apply a value in the range 2.0 V to 5.5 V to AVCC and AV_{ref} pins by connection to the power supply (V_{cc}), or some other method. Ensure that AV_{ref} \leq AV_{cc}.

- 2. P67 to P60 include supporting module inputs multiplexed on those pins.
- 3. IRQ2 includes the ADTRG signal multiplexed on that pin.
- 4. When IICS = 0. Low-level output when the bus drive function is selected is determined separately.
- 5. The upper limit of the applied voltage on Ports 6 and A is V_{cc} +0.3 V when CIN input is not selected, and the lower of V_{cc} +0.3 V and AV_{cc} +0.3 V when CIN input is selected. When a pin is in output mode, the output voltage is equivalent to the applied voltage.
- 6. Current dissipation values are for $V_{_{IH}}$ min = $V_{_{CC}}$ –0.5 V, and $V_{_{IL}}$ max = 0.5 V with all output pins unloaded and the on-chip pull-up MOSs in the off state.
- 7. The values are for V_{RAM} \leq V_{CC} < 2.7 V (mask ROM version), and V_{RAM} \leq V_{CC} < 3.0 V (F-ZTAT version), V_{IH} min = V_{CC} \times 0.9, and V_{IL} max = 0.3 V.
- 8. For flash memory program/erase operations, the applicable ranges are V_{cc} = 3.0 V to 3.6 V and Ta = 0 to +75°C. For the F-ZTAT versions, the test condition is V_{cc} = 3.0 V or higher.

Table 26.46 Permissible Output Currents

Conditions: $V_{cc} = 4.0 \text{ V}$ to 5.5 V, $V_{ss} = 0 \text{ V}$, $Ta = -20 \text{ to } +75^{\circ}\text{C}$ (regular specifications), $T_{a} = -40 \text{ to } +85^{\circ}\text{C}$ (wide-range specifications)

Item		Symbol	Min	Тур	Max	Unit
Permissible output low current (per pin)	PA7 to PA4 (bus drive function selected)	I _{OL}	_	_	20	mA
	Ports 1, 2, 3	_	_	_	10	mA
	RESO		_	_	3	mA
	Other output pins	_	_	_	2	mA
Permissible output	Total of ports 1, 2, and 3	\sum I _{OL}	_	_	80	mA
low current (total)	Total of all output pins, including the above		_	_	120	mA
Permissible output high current (per pin)	All output pins	-I _{OH}	_	_	2	mA
Permissible output high current (total)	Total of all output pins	$\Sigma - I_{OH}$	_	_	40	mA

Conditions: $V_{cc} = 2.7 \text{ V}$ to 5.5 V (mask ROM version), $V_{cc} = 3.0 \text{ V}$ to 5.5 V (F-ZTAT version), $V_{ss} = 0 \text{ V}, T_{s} = -20 \text{ to } +75^{\circ}\text{C}$

Item		Symbol	Min	Тур	Max	Unit
Permissible output low current (per pin)	PA7 to PA4 (bus drive function selected)	I _{OL}	_	_	10	mA
	Ports 1, 2, 3	-	_	_	2	mA
	RESO	_	_	_	1	mA
	Other output pins	_	_	_	1	mA
Permissible output	Total of ports 1, 2, and 3	\sum I _{OL}	_	_	40	mA
low current (total)	Total of all output pins, including the above		_	_	60	mA
Permissible output high current (per pin)	All output pins	-I _{OH}	_	_	2	mA
Permissible output high current (total)	Total of all output pins	$\Sigma - \mathbf{I}_{OH}$	_	_	30	mA

Notes: 1. To protect chip reliability, do not exceed the output current values in table 26.46.

2. When driving a Darlington pair or LED, always insert a current-limiting resistor in the output line, as show in figures 26.1 and 26.2.



Table 26.47 Bus Drive Characteristics

Conditions: $V_{CC} = 2.7 \text{ V}$ to 5.5 V (mask ROM version), $V_{CC} = 3.0 \text{ V}$ to 5.5 V (F-ZTAT version),

 $V_{ss} = 0 V$

Applicable Pins: PA7 to PA4 (bus drive function selected)

Item	Symbol	Min	Тур	Max	Unit	Test Conditions
Output low voltage	V _{oL}	_	_	0.8	V	$I_{oL} = 16 \text{ mA},$ $V_{cc} = 4.5 \text{ V to } 5.5 \text{ V}$
		_	_	0.5		$I_{OL} = 8 \text{ mA}$
		_	_	0.4		$I_{OL} = 3 \text{ mA}$

26.5.3 AC Characteristics

Clock timing, control signal timing, bus timing, and timing of on-chip supporting modules list the following.

Figure 26.4 shows the test conditions for the AC characteristics.

(1) Clock Timing

Table 26.48 shows the clock timing. The clock timing specified here covers clock (φ) output and clock pulse generator (crystal) and external clock input (EXTAL pin) oscillation settling times. For details of external clock input (EXTAL pin and EXCL pin) timing, see section 24, Clock Pulse Generator.

Table 26.48 Clock Timing

Condition A: $V_{CC} = 5.0 \text{ V} \pm 10\%$, $V_{SS} = 0 \text{ V}$, $\phi = 2 \text{ MHz}$ to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$ (regular specifications), $T_a = -40 \text{ to } +85^{\circ}\text{C}$ (wide-range specifications)

Condition B: $V_{cc} = 4.0 \text{ V}$ to 5.5 V, $V_{ss} = 0 \text{ V}$, $\phi = 2 \text{ MHz}$ to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$ (regular specifications), $T_a = -40 \text{ to } +85^{\circ}\text{C}$ (wide-range specifications)

Condition C: $V_{cc} = 2.7 \text{ V}$ to 5.5 V (mask ROM version), $V_{cc} = 3.0 \text{ V}$ to 5.5 V (F-ZTAT version) $V_{ss} = 0 \text{ V}$, $\phi = 2 \text{ MHz}$ to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$

		Condition A		Cond	Condition B		Condition C		
		20	MHz	16	MHz	10	MHz		Test
Item	Symbol	Min	Max	Min	Max	Min	Max	Unit	Conditions
Clock cycle time	t _{cyc}	50	500	62.5	500	100	500	ns	Figure 26.5
Clock high pulse width	t _{CH}	17	_	20	_	30	_	ns	Figure 26.5
Clock low pulse width	t _{CL}	17	_	20	_	30	_	ns	_
Clock rise time	t _{Cr}	_	8	_	10	_	20	ns	
Clock fall time	t _{Cf}	_	8	_	10		20	ns	
Oscillation settling time at reset (crystal)	t _{osc1}	10	_	10	_	20	_	ms	Figure 26.6 Figure 26.7
Oscillation settling time in software standby (crystal)	t _{osc2}	8	_	8	_	8	_	ms	_
External clock output stabilization delay time	t _{DEXT}	500	_	500	_	500	_	μs	_

(2) Control Signal Timing

Table 26.49 shows the control signal timing. The only external interrupts that can operate on the subclock ($\phi = 32.768 \text{ kHz}$) are NMI and IRO0, 1, 2, 6, and 7.

Table 26.49 Control Signal Timing

- Condition A: $V_{cc} = 5.0 \text{ V} \pm 10\%$, $V_{ss} = 0 \text{ V}$, $\phi = 32.768 \text{ kHz}$, 2 MHz to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$ (regular specifications), $T_a = -40 \text{ to } +85^{\circ}\text{C}$ (widerange specifications)
- Condition B: $V_{cc} = 4.0 \text{ V to } 5.5 \text{ V}$, $V_{ss} = 0 \text{ V}$, $\phi = 32.768 \text{ kHz}$, 2 MHz to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$ (regular specifications), $T_a = -40 \text{ to } +85^{\circ}\text{C}$ (widerange specifications)
- Condition C: $V_{cc} = 2.7 \text{ V}$ to 5.5 V (mask ROM version), $V_{cc} = 3.0 \text{ V}$ to 5.5 V (F-ZTAT version), $V_{ss} = 0 \text{ V}$, $\phi = 32.768 \text{ kHz}$, 2 MHz to maximum operating frequency, $T_{a} = -20 \text{ to } +75^{\circ}\text{C}$

		Cond	lition A	Cond	dition B	Cond	dition C		
		20	MHz	16	MHz	10	MHz		Test
Item	Symbol	Min	Max	Min	Max	Min	Max	Unit	Conditions
RES setup time	t _{ress}	200	_	200	_	300	_	ns	Figure 26.8
RES pulse width	t _{RESW}	20		20		20		t _{cyc}	_
NMI setup time (NMI)	t _{NMIS}	150	_	150	_	250	_	ns	Figure 26.9
NMI hold time (NMI)	t _{NMIH}	10	_	10	_	10	_	ns	_
NMI pulse width (exiting software standby mode)	t _{nmiw}	200	_	200	_	200	_	ns	_
IRQ setup time (IRQ7 to IRQ0)	t _{IRQS}	150	_	150	_	250	_	ns	_
IRQ hold time (IRQ7 to IRQ0)	t _{IRQH}	10	_	10	_	10	_	ns	_
IRQ pulse width (IRQ7, IRQ6, IRQ2 to IRQ0) (exiting software standby mode)	t _{IRQW}	200	_	200	_	200	_	ns	_

(3) Bus Timing

Table 26.50 shows the bus timing. Operation in external expansion mode is not guaranteed when operating on the subclock ($\phi = 32.768 \text{ kHz}$).

Table 26.50 Bus Timing

Condition A: $V_{cc} = 5.0 \text{ V} \pm 10\%$, $V_{ss} = 0 \text{ V}$, $\phi = 2 \text{ MHz}$ to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$ (regular specifications), $T_a = -40 \text{ to } +85^{\circ}\text{C}$ (wide-range specifications)

Condition B: $V_{CC} = 4.0 \text{ V}$ to 5.5 V, $V_{SS} = 0 \text{ V}$, $\phi = 2 \text{ MHz}$ to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$ (regular specifications), $T_a = -40 \text{ to } +85^{\circ}\text{C}$ (wide-range specifications)

Condition C: $V_{cc}=2.7~V$ to 5.5 V (mask ROM version), $V_{cc}=3.0~V$ to 5.5 V (F-ZTAT version), $V_{ss}=0~V$, $\phi=2~MHz$ to maximum operating frequency, $T_a=-20$ to $+75^{\circ}C$

		Cond	ition A	Cond	ition B	Cond	ition C		
		20	MHz	16	MHz	10	MHz		Test
Item	Symbol	Min	Max	Min	Max	Min	Max	Unit	Conditions
Address delay time	t _{AD}	_	20	_	30	_	40	ns	Figure 26.10 to
Address setup time	t _{AS}	$0.5 \times t_{\text{cyc}}$ -15	_	$0.5 \times t_{\text{cyc}}$ –20	_	0.5 × t _{cyc} –30	_	ns	figure 26.14
Address hold time	t _{AH}	$0.5 \times t_{\text{cyc}} -10$	_	$0.5 \times t_{\text{cyc}} -15$	_	0.5 × t _{cyc} –20	_	ns	_
CS delay time (IOS)	t _{CSD}	_	20	_	30	_	40	ns	
AS delay time	t _{ASD}	_	30	_	45	_	60	ns	_
RD delay time 1	t _{RSD1}	_	30	_	45	_	60	ns	_
RD delay time 2	t _{RSD2}	_	30	_	45	_	60	ns	_
Read data setup time	t _{RDS}	15	_	20	_	35	_	ns	_
Read data hold time	t _{RDH}	0	_	0	_	0	_	ns	_

		Cond	ition A	Cond	ition B	Cond	ition C		
		20	MHz	16	MHz	10	MHz		Test
Item	Symbol	Min	Max	Min	Max	Min	Max	Unit	Conditions
Read data access time 1	t _{ACC1}	_	$1.0 \times t_{\rm cyc}$ –30	-	$1.0 \times t_{\text{cyc}} -40$	-	$1.0 \times t_{\rm cyc}$ -60	ns	Figure 26.10 to figure 26.14
Read data access time 2	t _{ACC2}	_	1.5 × t _{cyc} –25	_	1.5 × t _{cyc} –35	_	1.5 × t _{cyc} –50	ns	_
Read data access time 3	t _{ACC3}	_	2.0 × t _{cyc} –30	_	2.0 × t _{cyc} –40	_	2.0 × t _{cyc} –60	ns	-
Read data access time 4	t _{ACC4}	_	2.5 × t _{cyc} –25	_	$2.5 \times t_{\text{cyc}} -35$	_	$2.5 \times t_{\text{cyc}}$ –50	ns	_
Read data access time 5	t _{ACC5}	_	3.0 × t _{cyc} –30	_	$3.0 \times t_{cyc}$ -40	_	3.0 × t _{cyc} –60	ns	_
WR delay time 1	t _{WRD1}	_	30	_	45	_	60	ns	_
WR delay time 2	t _{WRD2}	_	30	_	45	_	60	ns	_
WR pulse width 1	t _{wsw1}	1.0 × t _{cyc} –20	_	1.0 × t _{cyc} –30	_	1.0 × t _{cyc} -40	_	ns	_
WR pulse width 2	t _{wsw2}	1.5 × t _{cyc} –20	_	$1.5 \times t_{cyc}$ –30	_	$1.5 \times t_{\text{cyc}} -40$	_	ns	_
Write data delay time	t _{wdd}	_	30	_	45	_	60	ns	_
Write data setup time	t _{wds}	0	_	0	_	0	_	ns	_
Write data hold time	t _{wdh}	10	_	15	_	20	_	ns	_
WAIT setup time	t _{wrs}	30	_	45	_	60	_	ns	_
WAIT hold time	t _{wth}	5	_	5	_	10	_	ns	_

(4) Timing of On-Chip Supporting Modules

Tables 26.51 shows the on-chip supporting module timing. The only on-chip supporting modules that can operate in subclock operation ($\phi = 32.768 \text{ kHz}$) are the I/O ports, external interrupts (NMI and IRQ0, 1, 2, 6, and 7), the watchdog timer, and the 8-bit timer (channels 0 and 1).

Table 26.51 Timing of On-Chip Supporting Modules

Condition A: $V_{cc} = 5.0 \text{ V} \pm 10\%$, $V_{ss} = 0 \text{ V}$, $\phi = 32.768 \text{ kHz}^*$, 2 MHz to maximum operating frequency, $T_a = -20 \text{ to} +75^{\circ}\text{C}$ (regular specifications), $T_a = -40 \text{ to} +85^{\circ}\text{C}$ (widerange specifications)

Condition B: $V_{cc} = 4.0 \text{ V to } 5.5 \text{ V}, V_{ss} = 0 \text{ V}, \phi = 32.768 \text{ kHz}^*, 2 \text{ MHz to maximum operating}$ frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$ (regular specifications), $T_a = -40 \text{ to } +85^{\circ}\text{C}$ (widerange specifications)

Condition C: $V_{cc} = 2.7 \text{ V to } 5.5 \text{ V (mask ROM version)}, V_{cc} = 3.0 \text{ V to } 5.5 \text{ V (F-ZTAT version)}, V_{ss} = 0 \text{ V}, \phi = 32.768 \text{ kHz*}, 2 \text{ MHz to maximum operating frequency}, T_a = -20 \text{ to } +75^{\circ}\text{C}$

				Cond	dition A	Cond	dition B	Cond	dition C		
				20	MHz	16	MHz	10	MHz		Test
Item			Symbol	Min	Max	Min	Max	Min	Max	Unit	Conditions
I/O ports	Output d	lata delay	t _{PWD}	_	50	_	50	_	100	ns	Figure 26.15
	Input dat	ta setup	t _{PRS}	30	_	30	_	50	_		
	Input dat	ta hold	t _{PRH}	30	_	30	_	50	_		
FRT	Timer ou delay time	ıtput	t _{FTOD}	_	50	_	50	_	100	ns	Figure 26.16
	Timer in	put setup	t _{FTIS}	30	_	30	_	50	_		
	Timer clock input setup time		t _{FTCS}	30	_	30	_	50	_		Figure 26.17
	Timer clock	Single edge	t _{FTCWH}	1.5	_	1.5	_	1.5	_	t _{cyc}	_
	pulse width	Both edges	t _{FTCWL}	2.5	_	2.5	_	2.5	_		

				Conc	lition A	Cond	Condition B		lition C		
				20	MHz	16	MHz	10	MHz		Test
Item			Symbol	Min	Max	Min	Max	Min	Max	Unit	Conditions
TMR	Timer ou delay tim		t _{TMOD}	_	50	_	50	_	100	ns	Figure 26.18
	setup tin		\mathbf{t}_{TMRS}	30	_	30	_	50	_		Figure 26.20
	Timer closetup tin	ock input ne	t _{TMCS}	30	_	30	_	50	_		Figure 26.19
	Timer clock	Single edge	t _{TMCWH}	1.5	-	1.5	_	1.5	_	t _{cyc}	
	pulse width	Both edges	t _{TMCWL}	2.5	_	2.5	_	2.5	_		
PWMX	Pulse ou delay tim		t _{PWOD}	_	50	_	50	_	100	ns	Figure 26.21
SCI	Input clock	Asynchro- nous	· t _{Scyc}	4	_	4	_	4	_	t _{cyc}	Figure 26.22
	cycle	Synchro- nous	_	6	_	6	_	6	_		
	Input clo width	ck pulse	t _{sckw}	0.4	0.6	0.4	0.6	0.4	0.6	t _{Scyc}	-
	Input clo	ck rise	t _{SCKr}	_	1.5	_	1.5	_	1.5	t _{cyc}	=
	Input clo	ck fall	t _{SCKf}	_	1.5	_	1.5	_	1.5		
	Transmit delay tim (synchro	ne	t _{TXD}	_	50	_	50	_	100	ns	Figure 26.23
	Receive time (syr	data setup nchronous)	t _{RXS}	50	_	50	_	100	_	ns	-
		data hold nchronous)	t _{RXH}	50	_	50	_	100	_	ns	_
A/D conver- ter		nput setup	t _{TRGS}	30	_	30	_	50	_	ns	Figure 26.24
WDT	RESO o time	utput delay	t _{RESD}	_	100	_	120	_	200	ns	Figure 26.25
	RESO o width	utput pulse	t _{RESOW}	132	_	132	_	132	_	t _{cyc}	=

Note: * Only supporting modules that can be used in subclock operation

26.5.4 A/D Conversion Characteristics

Tables 26.52 and 26.53 list the A/D conversion characteristics.

Table 26.52 A/D Conversion Characteristics (AN7 to AN0 Input: 134/266-State Conversion)

- Condition A: $V_{cc} = 5.0 \text{ V} \pm 10\%$, $AV_{cc} = 5.0 \text{ V} \pm 10\%$, $AV_{ref} = 4.5 \text{ V}$ to AV_{cc} , $V_{ss} = AV_{ss} = 0 \text{ V}$, $\phi = 2$ MHz to maximum operating frequency, $T_a = -20$ to $+75^{\circ}$ C (regular specifications), $T_a = -40$ to $+85^{\circ}$ C (wide-range specifications)
- Condition B: $V_{CC} = 4.0 \text{ V}$ to 5.5 V, $AV_{CC} = 4.0 \text{ V}$ to 5.5 V, $AV_{ref} = 4.0 \text{ V}$ to AV_{CC} , $V_{ss} = AV_{ss} = 4.0 \text{ V}$ 0 V, $\phi = 2$ MHz to maximum operating frequency, $T_a = -20$ to +75°C (regular specifications), $T_a = -40$ to $+85^{\circ}$ C (wide-range specifications)
- Condition C (mask ROM version): $V_{CC} = 2.7 \text{ V}$ to 5.5 V, $AV_{CC} = 2.7 \text{ V}$ to 5.5 V, $AV_{ref} = 2.7 \text{ V}$ to AV_{CC} , $V_{SS} = AV_{SS} = 0$ V, $\phi = 2$ MHz to maximum operating frequency, $T_{.} = -20 \text{ to } +75^{\circ}\text{C}$
- Condition C (F-ZTAT version): $V_{CC} = 3.0 \text{ V}$ to 5.5 V, $AV_{CC} = 3.0 \text{ V}$ to 5.5 V, $AV_{ref} = 3.0 \text{ V}$ to AV_{CC} , $V_{SS} = AV_{SS} = 0$ V, $\phi = 2$ MHz to maximum operating frequency, $T_{a} = -20 \text{ to } +75^{\circ}\text{C}$

		Conditio	n A		Conditio	n B		Conditio	n C	
		20 MF	lz		16 MF	lz		10 MF	lz	_
Item	Min	Тур	Max	Min	Тур	Max	Min	Тур	Max	Unit
Resolution	10	10	10	10	10	10	10	10	10	Bits
Conversion time*5	_	_	6.7	_	_	8.4	_	_	13.4	μs
Analog input capacitance	_	_	20	_	_	20	_	_	20	pF
Permissible signal- source impedance		_	10 ^{*3} 5 ^{*4}		_	10 ^{*3} 5 ^{*4}	_	_	10 ^{*1} 5 ^{*2}	kΩ
Nonlinearity error	_	_	±3.0	_	_	±3.0	_	_	±7.0	LSB
Offset error	_	_	±3.5	_	_	±3.5	_	_	±7.5	LSB
Full-scale error	_	_	±3.5	_	_	±3.5	_	_	±7.5	LSB
Quantization error	_	_	±0.5	_	_	±0.5	_	_	±0.5	LSB
Absolute accuracy	_	_	±4.0	_	_	±4.0	_	_	±8.0	LSB

- Notes: 1. When $4.0 \text{ V} \le \text{AV}_{cc} \le 5.5 \text{ V}$
 - 2. When 2.7 V \leq AV_{cc} < 4.0 V (mask ROM version) or when 3.0 V \leq AV_{cc} < 4.0 V (F-ZTAT version)
 - 3. When conversion time \geq 11. 17 µs (CKS = 1 and $\phi \leq$ 12 MHz, or CKS = 0)
 - 4. When conversion time < 11. 17 μ s (CKS = 1 and ϕ > 12 MHz)
 - 5. In single mode and ϕ = maximum operating frequency.



Table 26.53 A/D Conversion Characteristics (CIN15 to CIN0 Input: 134/266-State Conversion)

- Condition A: $V_{cc} = 5.0 \text{ V} \pm 10\%$, $AV_{cc} = 5.0 \text{ V} \pm 10\%$, $AV_{ref} = 4.5 \text{ V}$ to AV_{cc} , $V_{ss} = AV_{ss} = 0 \text{ V}$, $\phi = 2 \text{ MHz}$ to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$ (regular specifications), $T_a = -40 \text{ to } +85^{\circ}\text{C}$ (wide-range specifications)
- Condition B: $V_{cc} = 4.0 \text{ V to } 5.5 \text{ V}, \text{AV}_{cc} = 4.0 \text{ V to } 5.5 \text{ V}, \text{AV}_{ref} = 4.0 \text{ V to } \text{AV}_{cc}, \text{V}_{ss} = \text{AV}_{ss} = 0 \text{ V}, \phi = 2 \text{ MHz to maximum operating frequency}, T_a = -20 \text{ to } +75^{\circ}\text{C} \text{ (regular specifications)}, T_a = -40 \text{ to } +85^{\circ}\text{C} \text{ (wide-range specifications)}$
- Condition C (mask ROM version): $V_{cc} = 2.7$ V to 5.5 V, $AV_{cc} = 2.7$ V to 5.5 V, $AV_{ref} = 2.7$ V to AV_{cc} , $AV_{cc} = 2.7$ V to AV_{cc} , $AV_{cc} = 2.7$ V to AV_{cc} , $AV_{cc} = 2.7$ V to $AV_{cc} = 2.$
- Condition C (F-ZTAT version): $V_{cc} = 3.0 \text{ V}$ to 5.5 V, $AV_{cc} = 3.0 \text{ V}$ to 5.5 V, $AV_{ref} = 3.0 \text{ V}$ to AV_{cc} , $V_{ss} = AV_{ss} = 0 \text{ V}$, $\phi = 2 \text{ MHz}$ to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$

		Conditio	n A		Conditio	n B		Conditio	on C	
		20 MF	lz		16 MF	lz		10 MF	łz	•
Item	Min	Тур	Max	Min	Тур	Max	Min	Тур	Max	Unit
Resolution	10	10	10	10	10	10	10	10	10	Bits
Conversion time*5	_	_	6.7	_	_	8.4	_	_	13.4	μs
Analog input capacitance	_	_	20	_	_	20	_	_	20	pF
Permissible signal- source impedance	_	_	10 ^{*3} 5 ^{*4}	_	_	10 ^{*3} 5 ^{*4}		_	10 ^{*1} 5 ^{*2}	kΩ
Nonlinearity error	_	_	±5.0	_	_	±5.0	_	_	±11.0	LSB
Offset error	_	_	±5.5	_	_	±5.5	_	_	±11.5	LSB
Full-scale error	_	_	±5.5	_	_	±5.5	_	_	±11.5	LSB
Quantization error	_	_	±0.5	_	_	±0.5	_	_	±0.5	LSB
Absolute accuracy	_	_	±6.0	_	_	±6.0	_	_	±12.0	LSB

- Notes: 1. When $4.0 \text{ V} \le \text{AV}_{cc} \le 5.5 \text{ V}$
 - 2. When 2.7 V \leq AV $_{\rm cc}$ < 4.0 V (mask ROM version) or when 3.0 V \leq AV $_{\rm cc}$ < 4.0 V (F-ZTAT version)
 - 3. When conversion time \geq 11. 17 μs (CKS = 1 and $\varphi \leq$ 12 MHz, or CKS = 0)
 - 4. When conversion time < 11. 17 μs (CKS = 1 and ϕ > 12 MHz)
 - 5. In single mode and ϕ = maximum operating frequency.

26.5.5 D/A Conversion Characteristics

Table 26.54 lists the D/A conversion characteristics.

Table 26.54 D/A Conversion Characteristics

- Condition A: $V_{CC} = 5.0 \text{ V} \pm 10\%$, $AV_{CC} = 5.0 \text{ V} \pm 10\%$, $AV_{ref} = 4.5 \text{ V}$ to AV_{CC} , $V_{SS} = AV_{SS} = 0 \text{ V}$, $\phi = 2$ MHz to maximum operating frequency, $T_a = -20$ to $+75^{\circ}$ C (regular specifications), $T_a = -40$ to $+85^{\circ}$ C (wide-range specifications)
- Condition B: $V_{CC} = 4.0 \text{ V}$ to 5.5 V, $AV_{CC} = 4.0 \text{ V}$ to 5.5 V, $AV_{ref} = 4.0 \text{ V}$ to AV_{CC} , $V_{ss} = AV_{ss} = 4.0 \text{ V}$ 0 V, $\phi = 2$ MHz to maximum operating frequency, $T_a = -20$ to +75°C (regular specifications), $T_a = -40$ to $+85^{\circ}$ C (wide-range specifications)
- Condition C (mask ROM version): $V_{CC} = 2.7 \text{ V}$ to 5.5 V, $AV_{CC} = 2.7 \text{ V}$ to 5.5 V, $AV_{ref} = 2.7 \text{ V}$ to AV_{cc} , $V_{ss} = AV_{ss} = 0$ V, $\phi = 2$ MHz to maximum operating frequency, $T_{a} = -20 \text{ to } +75^{\circ}\text{C}$
- Condition C (F-ZTAT version): $V_{CC} = 3.0 \text{ V}$ to 5.5 V, $AV_{CC} = 3.0 \text{ V}$ to 5.5 V, $AV_{ref} = 3.0 \text{ V}$ to AV_{CC} , $V_{SS} = AV_{SS} = 0$ V, $\phi = 2$ MHz to maximum operating frequency, $T_{a} = -20 \text{ to } +75^{\circ}\text{C}$

		С	onditio	on A	С	onditio	on B	С	onditio	on C		
			20 MHz			16 MHz			10 MHz			
Item		Min	Тур	Max	Min	Тур	Max	Min	Тур	Max	Unit	
Resolution		8	8	8	8	8	8	8	8	8	Bits	
Conversion time	With 20 pF load capacitance	_	_	10	_	_	10	_	_	10	μs	
Absolute accuracy	With 2 MΩ load resistance	_	±1.0	±1.5	_	±1.0	±1.5	_	±2.0	±3.0	LSB	
	With 4 MΩ load resistance	_	_	±1.0	_	_	±1.0	_	_	±2.0	-	

Tool

26.5.6 **Flash Memory Characteristics**

Table 26.55 shows the flash memory characteristics.

Table 26.55 Flash Memory Characteristics (Programming/erasing operating range)

Conditions (5 V version): $V_{cc} = 5.0 \text{ V} \pm 10\%$, $V_{ss} = 0 \text{ V}$, $T_a = 0 \text{ to } +75^{\circ}\text{C}$ (regular specifications),

 $T_a = 0$ to $+85^{\circ}$ C (wide-range specifications)

(3 V version): $V_{CC} = 3.0 \text{ V to } 3.6 \text{ V}, V_{SS} = 0 \text{ V}, T_a = 0 \text{ to } +75^{\circ}\text{C}$

Item		Symbol	Min	Тур	Max	Unit	Test Condition
Programming	time ^{*1 *2 *4}	tP	_	10	200	ms/ 32 bytes	
Erase time*1*3	*6	tE	_	100	1200	ms/ block	
Reprogrammir	ng count	N_{WEC}	100*8	10000*9	_	Times	
Data retention	time ^{*10}	t _{DRP}	10	_	_	Years	
Programming	Wait time after SWE-bit setting*1	х	10	_	_	μs	
	Wait time after PSU-bit setting*1	у	50	_	_	μs	
	Wait time after P-bit setting*1*4	Z	150	_	200	μs	
	Wait time after P-bit clear*1	α	10	_	_	μs	
	Wait time after PSU-bit clear*1	β	10	_	_	μs	
	Wait time after PV-bit setting*1	γ	4	_	_	μs	
	Wait time after dummy write*1	ε	2	_	_	μs	
	Wait time after PV-bit clear*1	η	4	_	_	μs	
	Maximum programming count**1 *4 *5	N	_	_	1000	Times	z = 200 µs
Erase	Wait time after SWE-bit setting*1	х	10	_	_	μs	
	Wait time after ESU-bit setting*1	у	200	_	_	μs	
	Wait time after E-bit setting*1 *6	Z	5	_	10	ms	
	Wait time after E-bit clear*1	α	10	_	_	μs	
	Wait time after ESU-bit clear*1	β	10	_	_	μs	
	Wait time after EV-bit setting*1	γ	20	_	_	μs	
	Wait time after dummy write*1	ε	2	_	_	μs	
	Wait time after EV-bit clear*1	η	5	_	_	μs	
	Maximum erase count*1 *6 *7	N	_		120	Times	z = 10 ms

Notes: 1. Set the times according to the program/erase algorithms.

2. Programming time per 32 bytes (Shows the total period for which the P-bit in FLMCR1 is set. It does not include the programming verification time.)

- 3. Block erase time (Shows the total period for which the E-bit in FLMCR1 is set. It does not include the erase verification time.)
- 4. Maximum programming time (tP (max)) tP (max) = wait time after P-bit setting (z) \times maximum programming count (N)
- 5. Number of times when the wait time after P-bit setting (z) = $200 \mu s$. The number of writes should be set according to the actual set value of (z) to allow programming within the maximum programming time (tP (max)).
- Maximum erase time (tE (max))
 tE (max) = wait time after E-bit setting (z) × maximum erase count (N))
- Number of times when the wait time after E-bit setting (z) = 10 ms.
 The number of erases should be set according to the actual set value of (z) to allow erasing within the maximum erase time (tE (max)).
- 8. Minimum number of times for which all characteristics are guaranteed after rewriting (Guarantee range is 1 to minimum value).
- 9. Reference value for 25°C (as a guideline, rewriting should normally function up to this value).
- 10. Data retention characteristic when rewriting is performed within the specification range, including the minimum value.

26.5.7 Usage Note

- (1) The F-ZTAT and mask ROM versions have been confirmed as fully meeting the reference values for electrical characteristics shown in this manual. However, actual performance figures, operating margins, noise margins, and other properties may vary due to differences in the manufacturing process, on-chip ROM, layout patterns, etc.
 - When system evaluation testing is carried out using the F-ZTAT version, the same evaluation tests should also be conducted for the mask ROM version when changing over to that version.
- (2) On-chip power supply step-down circuit
 - The H8S/2144F-ZTAT, H8S/2142F-ZTAT, and mask ROM version of H8S/2142 do not incorporate an internal power supply step-down circuit.
 - When changing over to F-ZTAT versions or mask ROM versions incorporating an internal step-down circuit, the VCC2 pin has the same pin location as the VCL pin in a step-down circuit (See figure 26.3).
 - Therefore, note that the circuit patterns differ between these two types of products.



26.6 Electrical Characteristics of H8S/2144 F-ZTAT (A-mask version) and Mask ROM Versions of H8S/2144 and H8S/2143

26.6.1 Absolute Maximum Ratings

Table 26.56 lists the absolute maximum ratings.

Table 26.56 Absolute Maximum Ratings

Item	Symbol	Value	Unit
Power supply voltage*1	V _{cc}	-0.3 to +7.0	V
Power supply voltage*1 (3-V version)	V _{cc}	-0.3 to +4.3	V
Power supply voltage*2 (VCL pin)	V _{CL}	-0.3 to +4.3	V
Input voltage (except ports 6, 7, and A)	V_{in}	-0.3 to V _{cc} +0.3	V
Input voltage (CIN input not selected for port 6 and A)	V_{in}	-0.3 to V_{cc} +0.3	V
Input voltage (CIN input selected for port 6 and A)	V _{in}	–0.3 V to lower of voltages $V_{\rm cc}$ +0.3 and $AV_{\rm cc}$ +0.3	V
Input voltage (port 7)	V_{in}	-0.3 to AV _{cc} +0.3	V
Reference supply voltage	AV_{ref}	-0.3 to AV _{cc} +0.3	V
Analog power supply voltage	AV _{cc}	-0.3 to +7.0	V
Analog power supply voltage (3 V version)	AV _{cc}	-0.3 to +4.3	V
Analog input voltage	$V_{_{AN}}$	-0.3 to AV _{cc} +0.3	V
Operating temperature	T _{opr}	Regular specifications: -20 to +75	°C
		Wide-range specifications: -40 to +85	°C
Operating temperature (flash memory	T_{opr}	Regular specifications: -20 to +75	°C
programming/erasing)		Wide-range specifications: -40 to +85	°C
Storage temperature	T_{stg}	-55 to +125	°C

Caution:

- Permanent damage to the chip may result if absolute maximum ratings are exceeded.
- 2. Never apply more than 7.0 V to any of the pins of the 5- or 4-V version or 4.3 V to any of the pins (except for port A) of the 3-V version.
- Notes: 1. Power supply voltage for VCC1 pin.

 Never exceed the maximum rating of V_{cL} in the low-power version (3-V version) because both the VCC1 and VCL pins are connected to the V_{cL} power supply.
 - It is an operating power supply voltage pin on the chip.
 Never apply power supply voltage to the VCL pin in the 5- or 4-V version.
 Always connect an external capacitor between the VCL pin and ground for internal voltage stabilization.

26.6.2 **DC** Characteristics

Table 26.57 lists the DC characteristics. Permitted output current values and bus drive characteristics are shown in tables 26.58 and 26.59, respectively.

Table 26.57 DC Characteristics (1)

Conditions:
$$V_{cc} = 5.0 \text{ V} \pm 10\%$$
, $AV_{cc}^{*1} = 5.0 \text{ V} \pm 10\%$, $AV_{ref}^{*1} = 4.5 \text{ V}$ to AV_{cc} , $V_{ss} = AV_{ss}^{*1} = 0 \text{ V}$, $T_a = -20 \text{ to } +75^{\circ}\text{C}$ (regular specifications), $T_a = -40 \text{ to } +85^{\circ}\text{C}$ (wide-range specifications)

Item			Symbol	Min	Тур	Max	Unit	Test Conditions
Schmitt	P67 to P60*2*5,	(1)	V _T -	1.0	_	_	V	
trigger input voltage	KIN15 to KIN8*5, IRQ2 to IRQ0*3,		V _T ⁺	_	_	$V_{cc} \times 0.7$	V	
voltage	IRQ5 to IRQ3		$V_T^+ - V_T^-$	0.4	_	_	V	
Input high voltage	RES, STBY, NMI, MD1, MD0	(2)	V _{IH}	V _{cc} -0.7	_	V _{cc} +0.3	V	
	EXTAL, PA7 to PA0*5			$V_{cc} \times 0.7$	_	V _{cc} +0.3	V	
	Port 7	-		2.0	_	AV _{cc} +0.3	V	
	Input pins except (1) and (2) above		_	2.0	_	V _{cc} +0.3	V	_
Input low voltage	RES, STBY, MD1, MD0	(3)	V _{IL}	-0.3	_	0.5	V	
	PA7 to PA0	-		-0.3	_	1.0	V	
	NMI, EXTAL, input pins except (1) and (3) above		_	-0.3	_	0.8	V	_
	All output pins*4		V _{OH}	V _{cc} -0.5	_	_	V	$I_{OH} = -200 \mu A$
voltage				3.5	_	_	V	$I_{OH} = -1 \text{ mA}$
Output low voltage	All output pins (except RESO)*4		V _{oL}	_	_	0.4	V	I _{OL} = 1.6 mA
	Ports 1 to 3		_	_	_	1.0	V	I _{oL} = 10 mA
	RESO		_	_	_	0.4	V	I _{oL} = 2.6 mA



Item			Symbol	Min	Тур	Max	Unit	Test Conditions
Input	RES		I _{in}	_	_	10.0	μΑ	$V_{in} = 0.5 \text{ to}$
leakage current	STBY, NMI, MD1, MD0		_	_	_	1.0	μΑ	[−] V _{cc} –0.5 V
	Port 7		_	_	_	1.0	μΑ	$V_{in} = 0.5 \text{ to}$ $AV_{cc} - 0.5 \text{ V}$
Three-state leakage current (off state)	Ports 1 to 6, 8, 9, A, B		I _{TSI}	_	_	1.0	μA	$V_{in} = 0.5 \text{ to}$ $V_{CC} = -0.5 \text{ V}$
Input	Ports 1 to 3		-I _P	30	_	300	μΑ	$V_{in} = 0 V$
pull-up MOS current	Ports 6, A, B		_	60	_	600	μA	_
Input	RES	(4)	C _{in}	_	_	80	pF	$V_{in} = 0 V$,
capacitance	NMI			_	_	50	pF	⁻f = 1 MHz, _T₂ = 25°C
	P52, P97, P42, P86 PA7 to PA2			_	_	20	pF	_ I _a = 25 C
	Input pins except (4) above		_	_	_	15	pF	_
Current	Normal operation		I _{cc}	_	55	70	mA	f = 20 MHz
dissipation*6	Sleep mode		_	_	36	55	mA	
	Standby mode*7		_	_	1.0	5.0	μΑ	T _a ≤ 50°C
				_	_	20.0	μΑ	50°C < T _a
Analog power	During A/D, D/A conversion		Al _{cc}	_	1.2	2.0	mA	
supply current	Idle			_	0.01	5.0	μΑ	AV _{cc} = 2.0 V to 5.5 V
Reference	During A/D conve	rsion	Al _{ref}	_	0.5	1.0	mA	
power supply current	During A/D, D/A conversion		_	_	2.0	5.0	mA	_
	Idle		_	_	0.01	5.0	μA	$AV_{ref} = 2.0 V$ to AV_{CC}

Item	Symbol	Min	Тур	Max	Unit	Test Conditions
Analog power supply voltage*1	AV _{cc}	4.5	_	5.5	V	Operating
		2.0	_	5.5	V	Idle/not used
RAM standby voltage	V_{RAM}	2.0	_	_	V	

Notes: 1. Do not leave the AVCC, AV_{ref}, and AVSS pins open even if the A/D converter and D/A converter are not used.

Even if the A/D converter and D/A converter are not used, apply a value in the range 2.0 V to 5.5 V to AVCC and AV_{ref} pins by connection to the power supply (V_{cc}), or some other method. Ensure that AV_{ref} \leq AV_{cc}.

- 2. P67 to P60 include supporting module inputs multiplexed on those pins.
- 3. IRQ2 includes the ADTRG signal multiplexed on that pin.
- 4. When IICS = 0. Low-level output when the bus drive function is selected is determined separately.
- 5. The upper limit of the applied voltage on Ports 6 and A is V_{cc} +0.3 V when CIN input is not selected, and the lower of V_{cc} +0.3 V and AV_{cc} +0.3 V when CIN input is selected. When a pin is in output mode, the output voltage is equivalent to the applied voltage.
- 6. Current dissipation values are for V_{IH} min = V_{CC} –0.2 V, and V_{IL} max = 0.2 V with all output pins unloaded and the on-chip pull-up MOSs in the off state.
- 7. The values are for $V_{RAM} \le V_{CC} < 4.5 \text{V}$, $V_{IH} \text{ min} = V_{CC} 0.2 \text{ V}$, and $V_{IL} \text{ max} = 0.2 \text{ V}$.

Table 26.57 DC Characteristics (2)

Conditions: $V_{cc} = 4.0 \text{ V}$ to 5.5 V, $AV_{cc}^{*1} = 4.0 \text{ V}$ to 5.5 V, $AV_{ref}^{*1} = 4.0 \text{ V}$ to AV_{cc}

 $V_{ss} = AV_{ss}^{*1} = 0 \text{ V}, T_a = -20 \text{ to } +75^{\circ}\text{C} \text{ (regular specifications)},$

 $T_a = -40 \text{ to } +85^{\circ}\text{C}$ (wide-range specifications)

Item			Symbol	Min	Тур	Max	Unit	Test Conditions
Schmitt	P67 to P60*2*5,	(1)	V _T	1.0	_	_	V	V_{cc} = 4.5 V to
trigger input voltage	$\overline{\text{KIN15}}$ to $\overline{\text{KIN8}}^{*5}$, $\overline{\text{IRQ2}}$ to $\overline{\text{IRQ0}}^{*3}$,		V _T ⁺	_		$V_{cc} \times 0.7$	V	_5.5 V
voltage	IRQ5 to IRQ3		$V_T^+ - V_T^-$	0.4		_	V	_
			V _T	0.8	_	_	V	V _{cc} < 4.5 V
			V _T ⁺	_	_	$V_{cc} \times 0.7$	V	_
			$V_T^+ - V_T^-$	0.3	_	_	V	_
Input high voltage	RES, STBY, NMI, MD1, MD0	(2)	V _{IH}	V _{cc} -0.7	_	V _{cc} +0.3	V	
	EXTAL, PA7 to PA0*5			$V_{cc} \times 0.7$	_	V _{cc} +0.3	V	_
	Port 7			2.0	_	AV _{cc} +0.3	V	_
	Input pins except (1) and (2) above		_	2.0	_	V _{cc} +0.3	V	
Input low voltage	RES, STBY, MD1, MD0	(3)	V_{IL}	-0.3	_	0.5	V	
	PA7 to PA0			-0.3	_	1.0	V	$V_{cc} = 4.5 \text{ V to}$ 5.5 V
				-0.3	_	0.8	V	V _{cc} < 4.5 V
	NMI, EXTAL, input pins except (1) and (3) above		_	-0.3	_	0.8	V	
	All output pins*4		V_{OH}	V _{cc} -0.5	_	_	V	$I_{OH} = -200 \mu A$
voltage				3.5	_	_	V	$I_{OH} = -1 \text{ mA},$ $V_{CC} = 4.5 \text{ V to}$ 5.5 V
				3.0	_	_	V	$I_{OH} = -1 \text{ mA},$ $V_{CC} < 4.5 \text{ V}$

Item			Symbol	Min	Тур	Max	Unit	Test Conditions
Output low voltage	All output pins (except RESO)*4		V _{OL}	_	_	0.4	V	$I_{OL} = 1.6 \text{ mA}$
	Ports 1 to 3		_	_	_	1.0	V	I _{oL} = 10 mA
	RESO		_	_	_	0.4	V	I _{oL} = 2.6 mA
Input	RES		I _{in}	_	_	10.0	μA	$V_{in} = 0.5 \text{ to}$
leakage current	STBY, NMI, MD1, MD0			_	_	1.0	μΑ	[−] V _{cc} –0.5 V
	Port 7			_	_	1.0	μΑ	$V_{in} = 0.5 \text{ to}$ AV _{cc} -0.5 V
Three-state leakage current (off state)	Ports 1 to 6, 8, 9, A, B		I _{TSI}	_	_	1.0	μА	$V_{in} = 0.5 \text{ to}$ $V_{CC} = -0.5 \text{ V}$
Input	Ports 1 to 3		-I _P	30	_	300	μΑ	$V_{in} = 0 V$,
pull-up MOS	Ports 6, A, B			60	_	600	μΑ	$V_{cc} = 4.5 \text{ V to}$ 5.5 V
current	Ports 1 to 3		_	20		200	μΑ	$V_{in} = 0 V$,
	Ports 6, A, B			40	_	500	μΑ	V _{cc} < 4.5 V
Input	RES	(4)	$C_{_{in}}$	_	_	80	pF	$V_{in} = 0 V$,
capacitance	NMI			_	_	50	pF	⁻f = 1 MHz, - T₂ = 25°C
	P52, P97, P42, P86, PA7 to PA2			_	_	20	pF	- 1 _a - 25 O
	Input pins except (4) above			_	_	15	pF	_
Current	Normal operation		I _{cc}	_	45	58	mA	f = 16 MHz
dissipation*6	Sleep mode		_	_	30	46	mA	f = 16 MHz
	Standby mode*7		_	_	1.0	5.0	μΑ	$T_a \le 50$ °C
				_	_	20.0	μΑ	50°C < T _a
Analog power	During A/D, D/A conversion		Al _{cc}	_	1.2	2.0	mA	
supply current	Idle		_	_	0.01	5.0	μΑ	$AV_{cc} = 2.0 \text{ V}$ to 5.5 V

Item		Symbol	Min	Тур	Max	Unit	Test Conditions
Reference power supply current	During A/D conversion	Al _{ref}	_	0.5	1.0	mA	
	During A/D, D/A conversion	_	_	2.0	5.0	mA	_
	Idle	_	_	0.01	5.0	μΑ	$AV_{ref} = 2.0 V$ to AV_{CC}
Analog power supply voltage*1		AV _{cc}	4.0	_	5.5	V	Operating
			2.0	_	5.5	V	Idle/not used
RAM standb	oy voltage	V _{RAM}	2.0	_	_	V	

Notes: 1. Do not leave the AVCC, AV_{ref}, and AVSS pins open even if the A/D converter and D/A converter are not used.

Even if the A/D converter and D/A converter are not used, apply a value in the range 2.0 V to 5.5 V to AVCC and AV_{ref} pins by connection to the power supply (V_{cc}), or some other method. Ensure that AV_{ref} \leq AV_{CC}.

- 2. P67 to P60 include supporting module inputs multiplexed on those pins.
- 3. IRQ2 includes the ADTRG signal multiplexed on that pin.
- 4. When IICS = 0. Low-level output when the bus drive function is selected is determined separately.
- 5. The upper limit of the applied voltage on Ports 6 and A is V_{cc} +0.3 V when CIN input is not selected, and the lower of V_{cc} +0.3 V and AV_{cc} +0.3 V when CIN input is selected. When a pin is in output mode, the output voltage is equivalent to the applied voltage.
- 6. Current dissipation values are for $V_{\rm IH}$ min = $V_{\rm cc}$ –0.2 V, and $V_{\rm IL}$ max = 0.2 V with all output pins unloaded and the on-chip pull-up MOSs in the off state.
- 7. The values are for $V_{RAM} \le V_{CC} < 4.0 \text{ V}$, $V_{HH} \min = V_{CC} 0.2 \text{ V}$, and $V_{H} \max = 0.2 \text{ V}$.

Table 26.57 DC Characteristics (3)

Conditions: $V_{cc} = 2.7 \text{ V to } 3.6 \text{ V}^{**}, \text{AV}_{cc}^{**} = 2.7 \text{ V to } 3.6 \text{ V}, \text{AV}_{ref}^{**} = 2.7 \text{ V to } 3.6 \text{ V},$ $V_{ss} = \text{AV}_{ss}^{**} = 0 \text{ V}, T_{a} = -20 \text{ to } +75 ^{\circ}\text{C}$

Item			Symbol	Min	Тур	Max	Unit	Test Conditions
Schmitt	P67 to P60*2*5,	(1)	V _T -	$V_{cc} \times 0.2$	_	_	V	
trigger input voltage	KIN15 to KIN8*5, IRQ2 to IRQ0*3,		V _T ⁺	_	_	$V_{cc} \times 0.7$	V	_
voitage	IRQ5 to IRQ3		$V_T^+ - V_T^-$	$V_{\rm cc} \times 0.05$	_	_	V	_
Input high voltage	RES, STBY, NMI, MD1, MD0	(2)	V _{IH}	$V_{cc} \times 0.9$	_	V _{cc} +0.3	V	
	EXTAL, PA7 to PA0*5	-		$V_{cc} \times 0.7$	_	V _{cc} +0.3	V	_
	Port 7	-		$V_{cc} \times 0.7$	_	AV _{cc} +0.3	V	
	Input pins except (1) and (2) above		_	$V_{cc} \times 0.7$	_	V _{cc} +0.3	V	_
Input low voltage	RES, STBY, MD1, MD0	(3)	$V_{_{\rm IL}}$	-0.3	_	$V_{cc} \times 0.1$	V	
-	PA7 to PA0	-		-0.3	_	$V_{cc} \times 0.2$	V	
	NMI, EXTAL, input pins except (1) and (3) above		_	-0.3	_	$V_{cc} \times 0.2$	V	
	All output pins*4		V _{OH}	V _{cc} -0.5	_	_	V	$I_{OH} = -200 \mu A$
voltage				V _{cc} -1.0	_	_	V	$I_{OH} = -1 \text{ mA}$ ($V_{CC} = 2.7 \text{ V}$ to 3.6 V
Output low voltage	All output pins (except RESO)*4		V _{oL}	_	_	0.4	V	$I_{OL} = 1.6 \text{ mA}$
	Ports 1 to 3		_	_	_	1.0	V	$I_{OL} = 5 \text{ mA}$
	RESO		_	_	_	0.4	V	I _{oL} = 1.6 mA
Input	RES		I _{in}	_	_	10.0	μΑ	$V_{in} = 0.5 \text{ to}$
eakage	STBY, NMI, MD1, MD0	1 in 1	_	_	1.0	μΑ	¯V _{cc} –0.5 V	
	Port 7				_	1.0	μA	$V_{in} = 0.5 \text{ to}$ AV _{CC} -0.5 V

Item		Symbol	Min	Тур	Max	Unit	Test Conditions
Three-state leakage current (off state)	Ports 1 to 6, 8, 9, A, B	I _{TSI}	_	_	1.0	μА	$V_{in} = 0.5 \text{ to}$ $V_{cc} = -0.5 \text{ V}$
Input pull-	Ports 1 to 3	-I _P	5	_	150	μΑ	$V_{in} = 0 V,$
up MOS current	Ports 6, A, B		30	_	300	μA	$V_{cc} = 2.7 \text{ V to}$ 3.6 V
Input	RES (4)	C_{in}	_	_	80	pF	$V_{in} = 0 V$
capacitance	NMI		_	_	50	pF	⁻f = 1 MHz, - T _a = 25°C
P52, P97, P42, P86, PA7 to PA2			_	_	20	pF	- 1 _a - 25 C
	Input pins except (4) above		_	_	15	pF	_
Current	Normal operation	I _{cc}	_	30	40	mA	f = 10 MHz
dissipation*6	Sleep mode		_	20	32	mA	_
	Standby mode*7		_	1.0	5.0	μΑ	T _a ≤ 50°C
			_	_	20.0	μΑ	50°C < T _a
Analog power	During A/D, D/A conversion	Al _{cc}	_	1.2	2.0	mA	
supply current	Idle	_	_	0.01	5.0	μA	$AV_{cc} = 2.0 \text{ V}$ to 3.6 V
Reference	During A/D conversion	Al _{ref}	_	0.5	1.0	mA	
power supply current	During A/D, D/A conversion	_	_	2.0	5.0	mA	_
	Idle	_	_	0.01	5.0	μΑ	$AV_{ref} = 2.0 V$ to AV_{cc}
Analog power supply voltage*1		AV _{cc}	2.7	_	3.6	V	Operating
			2.0	_	3.6	V	Idle/not used
RAM standb	RAM standby voltage		2.0	_	_	V	

Notes: 1. Do not leave the AVCC, AV_{ref}, and AVSS pins open even if the A/D converter and D/A converter are not used.

Even if the A/D converter and D/A converter are not used, apply a value in the range 2.0 V to 3.6 V to AVCC and AV_{ref} pins by connection to the power supply (V_{cc}), or some other method. Ensure that AV_{ref} \leq AV_{cc}.

- 2. P67 to P60 include supporting module inputs multiplexed on those pins.
- 3. IRQ2 includes the ADTRG signal multiplexed on that pin.

- 4. When IICS = 0. Low-level output when the bus drive function is selected is determined separately.
- 5. The upper limit of the applied voltage on Ports 6 and A is V_{cc} +0.3 V when CIN input is not selected, and the lower of V_{cc} +0.3 V and AV_{cc} +0.3 V when CIN input is selected. When a pin is in output mode, the output voltage is equivalent to the applied voltage.
- 6. Current dissipation values are for $V_{\rm H}$ min = $V_{\rm cc}$ –0.2 V, and $V_{\rm L}$ max = 0.2 V with all output pins unloaded and the on-chip pull-up MOSs in the off state.
- 7. The values are for $V_{RAM} \le V_{CC} < 2.7 \text{ V}$, $V_{IH} \min = V_{CC} = -0.2 \text{ V}$, and $V_{II} \max = 0.2 \text{ V}$.
- 8. For flash memory program/erase operations, the applicable range is $\rm V_{cc}$ = 3.0 V to 3.6 V.



Table 26.58 Permissible Output Currents

Conditions: $V_{CC} = 4.0 \text{ V}$ to 5.5 V, $V_{SS} = 0 \text{ V}$, $Ta = -20 \text{ to } +75^{\circ}\text{C}$ (regular specifications), $T_a = -40 \text{ to } +85^{\circ}\text{C}$ (wide-range specifications)

Item		Symbol	Min	Тур	Max	Unit
Permissible output low current (per pin)	PA7 to PA4 (bus drive function selected)	I _{OL}	_	_	20	mA
	Ports 1, 2, 3		_		10	mA
	RESO		_	_	3	mA
	Other output pins		_	_	2	mA
Permissible output	Total of ports 1, 2, and 3	\sum I _{OL}	_	_	80	mA
low current (total)	Total of all output pins, including the above		_	_	120	mA
Permissible output high current (per pin)	All output pins	-I _{OH}	_	_	2	mA
Permissible output high current (total)	Total of all output pins	$\Sigma - I_{OH}$	_	_	40	mA

Conditions: $V_{CC} = 2.7 \text{ V to } 3.6 \text{ V}, V_{SS} = 0 \text{ V}, T_{a} = -20 \text{ to } +75^{\circ}\text{C}$

Item		Symbol	Min	Тур	Max	Unit
Permissible output low current (per pin)	PA7 to PA4 (bus drive function selected)	I _{OL}	_	_	10	mA
	Ports 1, 2, 3	-	_	_	2	mA
	RESO	-	_	_	1	mA
	Other output pins		_	_	1	mA
Permissible output	Total of ports 1, 2, and 3	\sum I _{OL}	_	_	40	mA
low current (total)	Total of all output pins, including the above	_	_	_	60	mA
Permissible output high current (per pin)	All output pins	–I _{OH}	_	_	2	mA
Permissible output high current (total)	Total of all output pins	$\Sigma - \mathbf{I}_{OH}$	_	_	30	mA

Notes: 1. To protect chip reliability, do not exceed the output current values in table 26.58.

2. When driving a Darlington pair or LED, always insert a current-limiting resistor in the output line, as show in figures 26.1 and 26.2.

Table 26.59 Bus Drive Characteristics

Conditions: $V_{CC} = 4.0 \text{ V}$ to 5.5 V, $V_{CC} = 2.7 \text{ V}$ to 3.6 V (3 V version), $V_{SS} = 0 \text{ V}$

Applicable Pins: PA7 to PA4 (bus drive function selected)

Item	Symbol	Min	Тур	Max	Unit	Test Conditions
Output low voltage	V _{oL}	_	_	8.0	V	$I_{oL} = 16 \text{ mA},$ $V_{cc} = 4.5 \text{ V to } 5.5 \text{ V}$
		_	_	0.5	_	$I_{OL} = 8 \text{ mA}$
		_	_	0.4	-	I _{oL} = 3 mA

26.6.3 **AC Characteristics**

Clock timing, control signal timing, bus timing, and timing of on-chip supporting modules list the following.

Figure 26.4 shows the test conditions for the AC characteristics.



(1) Clock Timing

Table 26.60 shows the clock timing. The clock timing specified here covers clock (ϕ) output and clock pulse generator (crystal) and external clock input (EXTAL pin) oscillation settling times. For details of external clock input (EXTAL pin and EXCL pin) timing, see section 24, Clock Pulse Generator.

Table 26.60 Clock Timing

Condition A: $V_{cc} = 5.0 \text{ V} \pm 10\%$, $V_{ss} = 0 \text{ V}$, $\phi = 2 \text{ MHz}$ to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$ (regular specifications), $T_a = -40 \text{ to } +85^{\circ}\text{C}$ (wide-range specifications)

Condition B: $V_{cc} = 4.0 \text{ V}$ to 5.5 V, $V_{ss} = 0 \text{ V}$, $\phi = 2 \text{ MHz}$ to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$ (regular specifications), $T_a = -40 \text{ to } +85^{\circ}\text{C}$ (wide-range specifications)

Condition C: $V_{cc} = 2.7$ V to 3.6 V, $V_{ss} = 0$ V, $\phi = 2$ MHz to maximum operating frequency, $T_a = -20$ to $+75^{\circ}$ C

		Condition A		Cond	dition B	Con	dition C		
		20	MHz	16	MHz	10	MHz		Test
Item	Symbol	Min	Max	Min	Max	Min	Max	Unit	Conditions
Clock cycle time	t _{cyc}	50	500	62.5	500	100	500	ns	Figure 26.5
Clock high pulse width	t _{ch}	17	_	20	_	30	_	ns	Figure 26.5
Clock low pulse width	t _{cL}	17	_	20	_	30	_	ns	_
Clock rise time	t _{Cr}	_	8	_	10	_	20	ns	_
Clock fall time	t _{Cf}	_	8	_	10	_	20	ns	
Oscillation settling time at reset (crystal)	t _{osc1}	10	_	10	_	20	_	ms	Figure 26.6 Figure 26.7
Oscillation settling time in software standby (crystal)	t _{osc2}	8	_	8	_	8	_	ms	_
External clock output stabilization delay time	t _{DEXT}	500	_	500	_	500	_	μs	

(2) Control Signal Timing

Table 26.61 shows the control signal timing. The only external interrupts that can operate on the subclock ($\phi = 32.768 \text{ kHz}$) are NMI and IRO0, 1, 2, 6, and 7.

Table 26.61 Control Signal Timing

Condition A: $V_{cc} = 5.0 \text{ V} \pm 10\%$, $V_{ss} = 0 \text{ V}$, $\phi = 32.768 \text{ kHz}$, 2 MHz to maximum operating frequency, $T_a = -20 \text{ to} +75^{\circ}\text{C}$ (regular specifications), $T_a = -40 \text{ to} +85^{\circ}\text{C}$ (widerange specifications)

Condition B: $V_{cc} = 4.0 \text{ V to } 5.5 \text{ V}$, $V_{ss} = 0 \text{ V}$, $\phi = 32.768 \text{ kHz}$, 2 MHz to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$ (regular specifications), $T_a = -40 \text{ to } +85^{\circ}\text{C}$ (widerange specifications)

Condition C: $V_{cc} = 2.7$ V to 3.6 V, $V_{ss} = 0$ V, $\phi = 32.768$ kHz, 2 MHz to maximum operating frequency, $T_a = -20$ to $+75^{\circ}$ C

		Condition A		Cond	dition B	Cond	dition C		
		20	MHz	16	MHz	10	MHz		Test
Item	Symbol	Min	Max	Min	Max	Min	Max	Unit	Conditions
RES setup time	t _{RESS}	200	_	200	_	300	_	ns	Figure 26.8
RES pulse width	t _{resw}	20	_	20	_	20	_	t _{cyc}	_
NMI setup time (NMI)	t _{NMIS}	150	_	150	_	250	_	ns	Figure 26.9
NMI hold time (NMI)	t _{nmih}	10	_	10	_	10	_	ns	_
NMI pulse width (exiting software standby mode)	t _{nmiw}	200	_	200	_	200	_	ns	_
IRQ setup time (IRQ7 to IRQ0)	t _{IRQS}	150	_	150	_	250	_	ns	_
IRQ hold time (IRQ7 to IRQ0)	t _{IRQH}	10	_	10	_	10	_	ns	_
IRQ pulse width (IRQ7, IRQ6, IRQ2 to IRQ0) (exiting software standby mode)	t _{IRQW}	200	_	200	_	200	_	ns	_

(3) Bus Timing

Table 26.62 shows the bus timing. Operation in external expansion mode is not guaranteed when operating on the subclock ($\phi = 32.768 \text{ kHz}$).

Table 26.62 Bus Timing

Condition A: $V_{cc} = 5.0 \text{ V} \pm 10\%$, $V_{ss} = 0 \text{ V}$, $\phi = 2 \text{ MHz}$ to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$ (regular specifications), $T_a = -40 \text{ to } +85^{\circ}\text{C}$ (wide-range specifications)

Condition B: $V_{cc} = 4.0 \text{ V to } 5.5 \text{ V}, V_{ss} = 0 \text{ V}, \phi = 2 \text{ MHz to maximum operating frequency},$ $T_a = -20 \text{ to } +75^{\circ}\text{C}$ (regular specifications), $T_a = -40 \text{ to } +85^{\circ}\text{C}$ (wide-range specifications)

Condition C: $V_{cc}=2.7$ V to 3.6 V, $V_{ss}=0$ V, $\phi=2$ MHz to maximum operating frequency, $T_a=-20$ to $+75^{\circ}C$

		Cond	lition A	Condition B		Cond	ition C		
		20	MHz	16 MHz		10	MHz	•	Test
Item	Symbol	Min	Max	Min	Max	Min	Max	Unit	Conditions
Address delay time	t _{AD}	_	20	_	30	_	40	ns	Figure 26.10 to
Address setup time	t _{AS}	$0.5 \times t_{\text{cyc}} -15$	_	0.5 × t _{cyc} –20	_	0.5 × t _{cyc} –30	_	ns	figure 26.14
Address hold time	t _{AH}	$0.5 \times t_{\text{cyc}} -10$	_	$0.5 \times t_{\text{cyc}} -15$	_	$0.5 \times t_{\text{cyc}}$ –20	_	ns	
CS delay time (IOS)	t _{CSD}	_	20	_	30	_	40	ns	
AS delay time	t _{ASD}	_	30	_	45	_	60	ns	-
RD delay time 1	t _{RSD1}	_	30	_	45	_	60	ns	-
RD delay time 2	t _{RSD2}	_	30	_	45	_	60	ns	_
Read data setup time	t _{RDS}	15	_	20	_	35	_	ns	_
Read data hold time	t _{RDH}	0	_	0	_	0	_	ns	

		Cond	lition A	Condition B		Cond	ition C		
		20	MHz	16	MHz	10	MHz		Test
Item	Symbol	Min	Max	Min	Max	Min	Max	Unit	
Read data access time 1	t _{ACC1}	_	1.0 × t _{cyc} -30	_	1.0 × t _{cyc} -40	_	1.0 × t _{cyc} –60	ns	Figure 26.10 to figure 26.14
Read data access time 2	t _{ACC2}	_	1.5 × t _{cyc} –25	_	1.5 × t _{cyc} –35	_	1.5 × t _{cyc} –50	ns	_
Read data access time 3	t _{ACC3}	_	$2.0 \times t_{\rm cyc} -30$	-	$2.0 \times t_{\rm cyc}$ -40	_	$2.0 \times t_{\text{cyc}}$ -60	ns	
Read data access time 4	t _{ACC4}	_	$2.5 \times t_{\text{cyc}} -25$	_	$2.5 \times t_{cyc} -35$	_	$2.5 \times t_{\text{cyc}} -50$	ns	_
Read data access time 5	t _{ACC5}	_	3.0 × t _{cyc} –30	_	3.0 × t _{cyc} –40	_	3.0 × t _{cyc} –60	ns	_
WR delay time 1	t _{wrd1}	_	30	_	45	_	60	ns	_
WR delay time 2	t _{WRD2}	_	30	_	45	_	60	ns	_
WR pulse width 1	t _{wsw1}	1.0 × t _{cyc} –20	_	1.0 × t _{cyc} –30	_	1.0 × t _{cyc} –40	_	ns	_
WR pulse width 2	t _{wsw2}	$1.5 \times t_{\text{cyc}}$ –20	_	$1.5 \times t_{\text{cyc}} -30$	_	1.5 × t _{cyc} –40	_	ns	_
Write data delay time	t _{wdd}	_	30	_	45	_	60	ns	_
Write data setup time	t _{wds}	0	_	0	_	0	_	ns	_
Write data hold time	t _{wdh}	10	_	15	_	20	_	ns	_
WAIT setup time	t _{wts}	30	_	45	_	60	_	ns	
WAIT hold time	t _{wth}	5	_	5	_	10	_	ns	_

(4) Timing of On-Chip Supporting Modules

Tables 26.63 shows the on-chip supporting module timing. The only on-chip supporting modules that can operate in subclock operation ($\phi = 32.768 \text{ kHz}$) are the I/O ports, external interrupts (NMI and IRQ0, 1, 2, 6, and 7), the watchdog timer, and the 8-bit timer (channels 0 and 1).

Table 26.63 Timing of On-Chip Supporting Modules

Condition A: $V_{cc} = 5.0 \text{ V} \pm 10\%$, $V_{ss} = 0 \text{ V}$, $\phi = 32.768 \text{ kHz}^*$, 2 MHz to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$ (regular specifications), $T_a = -40 \text{ to } +85^{\circ}\text{C}$ (widerange specifications)

Condition B: $V_{cc} = 4.0 \text{ V to } 5.5 \text{ V}, V_{ss} = 0 \text{ V}, \phi = 32.768 \text{ kHz}^*, 2 \text{ MHz to maximum operating}$ frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$ (regular specifications), $T_a = -40 \text{ to } +85^{\circ}\text{C}$ (widerange specifications)

Condition C: $V_{cc} = 2.7 \text{ V}$ to 3.6 V, $V_{ss} = 0 \text{ V}$, $\phi = 32.768 \text{ kHz}^*$, 2 MHz to maximum operating frequency, $T_a = -20 \text{ to } +75 ^{\circ}\text{C}$

				Condition A		Cond	dition B	Cond	lition C			
				20	MHz	16	MHz	10 MHz			Test	
Item			Symbol	Min	Max	Min	Max	Min	Max	Unit	Conditions	
I/O ports	Output data delay time		t _{PWD}	_	50	_	50	_	100	ns	Figure 26.15	
	Input dat	a setup	t _{PRS}	30	_	30	_	50	_			
	Input dat	a hold time	t _{PRH}	30	_	30	_	50	_			
FRT	Timer ou time	tput delay	t _{FTOD}	_	50	_	50	_	100	ns	Figure 26.16	
	Timer inp	out setup	t _{FTIS}	30	_	30	_	50	_			
	Timer clo	-	t _{FTCS}	30	_	30	_	50	_		Figure 26.17	
	Timer clock	Single edge	t _{FTCWH}	1.5	_	1.5	_	1.5	_	t _{cyc}	_	
	pulse width	Both edges	t _{FTCWL}	2.5	_	2.5	_	2.5	_			

				Cond	dition A	Cond	dition B	Cond	lition C		
				20	MHz	16	MHz	10	MHz		Test
Item			Symbol	Min	Max	Min	Max	Min	Max	Unit	Conditions
TMR	Timer of delay ti		\mathbf{t}_{TMOD}	_	50	_	50	_	100	ns	Figure 26.18
	Timer r	eset input me	t _{TMRS}	30	_	30	_	50	_		Figure 26.20
	Timer of setup ti	clock input me	t _{TMCS}	30	_	30	_	50	_		Figure 26.19
	Timer clock	Single edge	t _{TMCWH}	1.5	_	1.5	_	1.5	_	t _{cyc}	_
	pulse width	Both edges	t _{TMCWL}	2.5	_	2.5	_	2.5	_		
PWMX	Pulse o	•	t _{PWOD}	_	50	_	50	_	100	ns	Figure 26.21
SCI	Input clock	Asynchro- nous	t _{Scyc}	4	_	4	_	4	_	t _{cyc}	Figure 26.22
	cycle	Synchro- nous	_	6	_	6	_	6	_		
	Input cl width	ock pulse	t _{sckw}	0.4	0.6	0.4	0.6	0.4	0.6	t _{Scyc}	_
	Input cl	ock rise	t _{SCKr}	_	1.5	_	1.5	_	1.5	t _{cyc}	-
	Input cl	ock fall	t _{SCKf}	_	1.5	_	1.5	_	1.5		
	Transm delay ti (synchi	me	t _{TXD}	_	50	_	50	_	100	ns	Figure 26.23
	Receive time (sy	e data setup ynchronous)	t _{RXS}	50	_	50	_	100	_	ns	_
		e data hold ynchronous)	t _{RXH}	50	_	50	_	100	_	ns	_
A/D conver- ter	Trigger input setup time		t _{TRGS}	30	_	30	_	50	_	ns	Figure 26.24
WDT	RESO time	output delay	t _{RESD}	_	100	_	120	_	200	ns	Figure 26.25
	RESO width	output pulse	t _{RESOW}	132	_	132	_	132	_	t _{cyc}	=

* Only supporting modules that can be used in subclock operation

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26.6.4 A/D Conversion Characteristics

Tables 26.64 and 26.65 list the A/D conversion characteristics.

Table 26.64 A/D Conversion Characteristics (AN7 to AN0 Input: 134/266-State Conversion)

Condition A:
$$V_{cc} = 5.0 \text{ V} \pm 10\%$$
, $AV_{cc} = 5.0 \text{ V} \pm 10\%$, $AV_{ref} = 4.5 \text{ V}$ to AV_{cc} , $V_{ss} = AV_{ss} = 0 \text{ V}$, $\phi = 2 \text{ MHz}$ to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$ (regular specifications), $T_a = -40 \text{ to } +85^{\circ}\text{C}$ (wide-range specifications)

Condition B:
$$V_{cc} = 4.0 \text{ V to } 5.5 \text{ V}, \text{AV}_{cc} = 4.0 \text{ V to } 5.5 \text{ V}, \text{AV}_{ref} = 4.0 \text{ V to } \text{AV}_{cc},$$

$$V_{ss} = \text{AV}_{ss} = 0 \text{ V}, \phi = 2 \text{ MHz to maximum operating frequency},$$

$$T_a = -20 \text{ to } +75^{\circ}\text{C (regular specifications)},$$

$$T_a = -40 \text{ to } +85^{\circ}\text{C (wide-range specifications)}$$

Condition C:
$$V_{cc} = 2.7$$
 V to 3.6 V, $AV_{cc} = 2.7$ V to 3.6 V, $AV_{ref} = 2.7$ V to AV_{cc} , $V_{ss} = AV_{ss} = 0$ V, $\phi = 2$ MHz to maximum operating frequency, $T_a = -20$ to $+75^{\circ}$ C

		Conditio	n A		Conditio	n B		Condition C								
		20 MF	lz		16 MF	lz		10 MF	lz	_						
Item	Min	Тур	Max	Min	Тур	Max	Min	Тур	Max	Unit						
Resolution	10	10	10	10	10	10	10	10	10	Bits						
Conversion time*3	_	_	6.7	_	_	8.4	_	_	13.4	μs						
Analog input capacitance	_	_	20	_	_	20	_	_	20	pF						
Permissible signal- source impedance	_	_	10 ^{*1} 5 ^{*2}	_	_	10 ^{*1} 5 ^{*2}	_	_	5	kΩ						
Nonlinearity error	_	_	±3.0	_	_	±3.0	_	_	±7.0	LSB						
Offset error	_	_	±3.5	_	_	±3.5	_	_	±7.5	LSB						
Full-scale error	_	_	±3.5	_	_	±3.5	_	_	±7.5	LSB						
Quantization error	_	_	±0.5	_	_	±0.5	_	_	±0.5	LSB						
Absolute accuracy	_	_	±4.0	_	_	±4.0	_	_	±8.0	LSB						

Notes: 1. When conversion time \geq 11. 17 μs (CKS = 1 and $\varphi \leq$ 12 MHz, or CKS = 0)

- 2. When conversion time < 11. 17 μs (CKS = 1 and ϕ > 12 MHz)
- 3. In single mode and ϕ = maximum operating frequency.

Table 26.65 A/D Conversion Characteristics

(CIN15 to CIN0 Input: 134/266-State Conversion)

Condition A: $V_{cc} = 5.0 \text{ V} \pm 10\%$, $AV_{cc} = 5.0 \text{ V} \pm 10\%$, $AV_{ref} = 4.5 \text{ V}$ to AV_{cc} ,

 $V_{ss} = AV_{ss} = 0 \text{ V}, \phi = 2 \text{ MHz}$ to maximum operating frequency,

 $T_a = -20 \text{ to } +75 ^{\circ}\text{C}$ (regular specifications), $T_a = -40 \text{ to } +85 ^{\circ}\text{C}$ (wide-range specifications)

Condition B: $V_{CC} = 4.0 \text{ V}$ to 5.5 V, $AV_{CC} = 4.0 \text{ V}$ to 5.5 V, $AV_{RE} = 4.0 \text{ V}$ to AV_{CC}

 $V_{ss} = AV_{ss} = 0 \text{ V}, \phi = 2 \text{ MHz}$ to maximum operating frequency,

 $T_a = -20 \text{ to } +75^{\circ}\text{C}$ (regular specifications),

 $T_a = -40 \text{ to } +85^{\circ}\text{C}$ (wide-range specifications)

Condition C: $V_{CC} = 3.0 \text{ V}$ to 3.6 V^{*4} , $AV_{CC} = 3.0 \text{ V}$ to 3.6 V^{*4} , $AV_{ref} = 3.0 \text{ V}$ to AV_{CC}^{*4} ,

 $V_{ss} = AV_{ss} = 0 \text{ V}, \phi = 2 \text{ MHz}$ to maximum operating frequency,

 $T_a = -20 \text{ to } +75^{\circ}\text{C}$

		Conditio	on A		Conditio	n B		Condition C							
	_	20 MF	łz		16 MF	łz		10 MF	lz	-					
Item	Min	Тур	Max	Min	Тур	Max	Min	Тур	Max	Unit					
Resolution	10	10	10	10	10	10	10	10	10	Bits					
Conversion time*3	_	_	6.7	_	_	8.4	_	_	13.4	μs					
Analog input capacitance	_	_	20	_	_	20	_	_	20	pF					
Permissible signal- source impedance	_	_	10 ^{*1} 5 ^{*2}	_	_	10 ^{*1} 5 ^{*2}		_	5	kΩ					
Nonlinearity error	_	_	±5.0	_	_	±5.0	_	_	±11.0	LSB					
Offset error	_	_	±5.5	_	_	±5.5	_	_	±11.5	LSB					
Full-scale error	_	_	±5.5	_	_	±5.5	_	_	±11.5	LSB					
Quantization error	_	_	±0.5	_	_	±0.5	_	_	±0.5	LSB					
Absolute accuracy	_	_	±6.0	_	_	±6.0	_	_	±12.0	LSB					

Notes: 1. When conversion time \geq 11. 17 µs (CKS = 1 and $\phi \leq$ 12 MHz, or CKS = 0)

- 2. When conversion time < 11. 17 μ s (CKS = 1 and ϕ > 12 MHz)
- 3. In single mode and ϕ = maximum operating frequency.
- 4. When using CIN, the applicable range is V_{cc} = 3.0 V to 3.6 V, AV_{cc} = 3.0 V to 3.6 V, and AVref = 3.0 V to 3.6 V.

26.6.5 D/A Conversion Characteristics

Table 26.66 lists the D/A conversion characteristics.

Table 26.66 D/A Conversion Characteristics

Condition A:
$$V_{cc} = 5.0 \text{ V} \pm 10\%$$
, $AV_{cc} = 5.0 \text{ V} \pm 10\%$, $AV_{ref} = 4.5 \text{ V}$ to AV_{cc} , $V_{ss} = AV_{ss} = 0 \text{ V}$, $\phi = 2 \text{ MHz}$ to maximum operating frequency, $T_a = -20 \text{ to } +75^{\circ}\text{C}$ (regular specifications),

$$T_a = -20$$
 to +/5°C (regular specifications),
 $T_a = -40$ to +85°C (wide-range specifications)

Condition B:
$$V_{cc} = 4.0 \text{ V to } 5.5 \text{ V}, \text{AV}_{cc} = 4.0 \text{ V to } 5.5 \text{ V}, \text{AV}_{ref} = 4.0 \text{ V to } \text{AV}_{cc},$$

$$V_{ss} = \text{AV}_{ss} = 0 \text{ V}, \phi = 2 \text{ MHz to maximum operating frequency},$$

$$T_a = -20 \text{ to } +75^{\circ}\text{C (regular specifications)},$$

$$T_a = -40 \text{ to } +85^{\circ}\text{C (wide-range specifications)}$$

Condition C:
$$V_{cc}=2.7~V$$
 to 3.6 V, $AV_{cc}=2.7~V$ to 3.6 V, $AV_{ref}=2.7~V$ to AV_{cc} , $V_{ss}=AV_{ss}=0~V$, $\varphi=2~MHz$ to maximum operating frequency, $T_a=-20$ to $+75^{\circ}C$

	С	onditio	on A	C	onditio	on B	С	Condition C						
			20 MF	łz		16 MF	łz		-					
Item		Min	Тур	Max	Min	Тур	Max	Min	Тур	Max	Unit			
Resolution		8	8	8	8	8	8	8	8	8	Bits			
Conversion time	With 20-pF load capacitance	_	_	10	_	_	10	_	_	10	μs			
Absolute accuracy	With 2-MΩ load resistance	_	±1.0	±1.5	_	±1.0	±1.5	_	±2.0	±3.0	LSB			
	With 4-MΩ load resistance	_	_	±1.0	_	_	±1.0	_	_	±2.0	-			

26.6.6 Flash Memory Characteristics

Table 26.67 shows the flash memory characteristics.

Table 26.67 Flash Memory Characteristics (Programming/erasing operating range)

Conditions (5 V version): $V_{cc} = 4.0 \text{ V}$ to 5.5 $V_{cs} = 0 \text{ V}$, $T_{a} = -20 \text{ to } +75 ^{\circ}\text{C}$ (regular

specifications), $T_a = -40$ to $+85^{\circ}$ C (wide-range specifications)

(3 V version): $V_{cc} = 3.0 \text{ V}$ to 3.6V, $V_{ss} = 0 \text{ V}$, $T_a = -20 \text{ to } 75^{\circ}\text{C}$

Item		Symbol	Min	Тур	Max	ms/ 128 bytes ms/ block Times Years μ s Addition	Test Condition
Programming	time ^{*1*2*4}	tP	_	10	200		
Erase time*1*	*6	tE	_	100	1200		
Reprogrammir	ng count	N _{wec}	100*8	10000*9	_	Times	
Data retention	time ^{*10}	t _{DRP}	10	_	_	Years	
Programming	Wait time after SWE-bit setting*1	х	1	_	_	μs	
Programming times Erase time******* Reprogramming condition times Programming Wan	Wait time after PSU-bit setting*1	у	50	_	_	μs	
	Wait time after P-bit setting*1*4	z1	28	30	32	μs	$1 \le n \le 6$
		z2	198	200	202	μs	$7 \le n \le 1000$
		z3	8	10	12	μs	Additional writing
	Wait time after P-bit clear*1	α	10	_	_	μs	
	Wait time after PSU-bit clear*1	β	10	_	_	μs	
	Wait time after PV-bit setting*1	γ	4	_	_	μs	
	Wait time after dummy write*1	ε	2	_	_	μs	
	Wait time after PV-bit clear*1	η	4	_	_	μs	
	Wait time after SWE-bit clear*1	θ	100	_	_	μs	
	Maximum programming count*1 *4 *5	N	_	_	1000	Times	
Erase	Wait time after SWE-bit setting*1	х	1	_	_	μs	
	Wait time after ESU-bit setting*1	у	100	_	_	μs	
	Wait time after E-bit setting*1 *6	Z	10	_	100	ms	
	Wait time after E-bit clear*1	α	10	_	_	μs	
	Wait time after ESU-bit clear*1	β	10	_	_	μs	
	Wait time after EV-bit setting*1	γ	20	_	_	μs	
	Wait time after H'FF dummy write*1	ε	2	_	_	μs	
	Wait time after EV-bit clear*1	η	4	_	_	μs	
	Wait time after SWE-bit clear*1	θ	100	_	_	μs	
	Maximum erase count*1 *6 *7	N	_	_	120	Times	

Notes: 1. Set the times according to the program/erase algorithms.

- 2. Programming time per 128 bytes (Shows the total period for which the P-bit in FLMCR1 is set. It does not include the programming verification time.)
- 3. Block erase time (Shows the total period for which the E-bit in FLMCR1 is set. It does not include the erase verification time.)
- Maximum programming time (tP (max))
 tP (max) = (Wait time after P-bit setting (z1) + (z3)) × 6 + Wait time after P-bit setting (z2) × ((N) 6)
- Maximum programming count (N) should be set according to the actual set value of (z1, z2, z3) to allow programming within the maximum programming time (tP (max)). The wait time after P-bit setting (z1, z2, z3) must be changed with the value of the number of writing times (n) as follows.

The number of times for writing n

$$1 \le n \le 6$$
 $z1 = 30 \ \mu s, \ z3 = 10 \ \mu s$ $7 \le n \le 1000$ $z2 = 200 \ \mu s$

6. Maximum erase time (tE (max))

 $tE(max) = Waiting time after E-bit setting (z) \times Maximum erase count (N)$

- 7. Maximum erase count (N) should be set according to the actual setting (z) to allow erase within the maximum erase time (tE (max)).
- 8. Minimum number of times for which all characteristics are guaranteed after rewriting (Guarantee range is 1 to minimum value).
- 9. Reference value for 25°C (as a guideline, rewriting should normally function up to this value).
- 10. Data retention characteristic when rewriting is performed within the specification range, including the minimum value.

26.6.7 Usage Note

(1) The F-ZTAT and mask ROM versions have been confirmed as fully meeting the reference values for electrical characteristics shown in this manual. However, actual performance figures, operating margins, noise margins, and other properties may vary due to differences in the manufacturing process, on-chip ROM, layout patterns, etc.

When system evaluation testing is carried out using the F-ZTAT version, the same evaluation tests should also be conducted for the mask ROM version when changing over to that version.

(2) On-chip power supply step-down circuit

The following products incorporate an internal power supply step-down circuit, which automatically drops down the internal power supply voltage to the optimum internal voltage level: the F-ZTAT A-mask versions of the H8S/2144 (HD64F2144A) and the mask ROM versions of the H8S/2144, and H8S/2143 (HD6432144S and HD6432143S).

The voltage-stabilization capacitor (0.47 μ F one or two connected in series) must be connected between the VCL (internal power supply step-down) and VSS pins.

Figure 26.3 shows the connection of the external capacitors.

For the 5- or 4-V version, do not connect the VCL pin in a product incorporating an internal step-down circuit to the $V_{\rm cc}$ power supply. (Connect the VCC1 pin to the $V_{\rm cc}$ power supply as usual.)

For the 3-V version, connect both the VCL and VCC1 pins to the system power supply.

When changing from the F-ZTAT versions not incorporating an internal step-down circuit to the mask ROM versions incorporating an internal step-down circuit, the VCL pin has the same pin location as the VCC2 pin.

Therefore, note that the circuit patterns differ between these two types of products.



26.7 Operational Timing

26.7.1 Test Conditions for the AC Characteristics

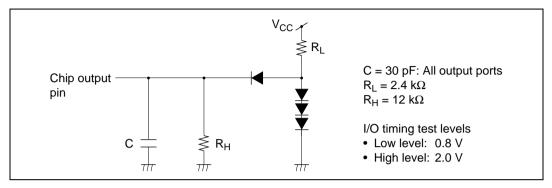


Figure 26.4 Output Load Circuit

26.7.2 Clock Timing

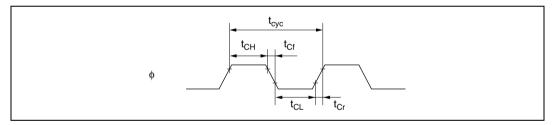


Figure 26.5 System Clock Timing

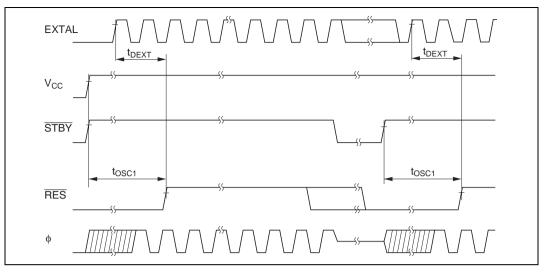


Figure 26.6 Oscillation Settling Timing

26.7.3 Control Signal Timing

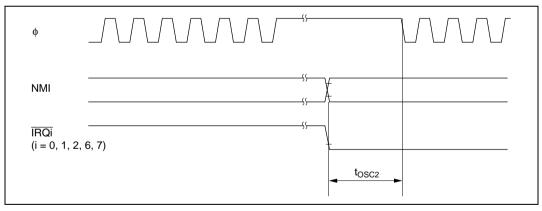


Figure 26.7 Oscillation Setting Timing (Exiting Software Standby Mode)

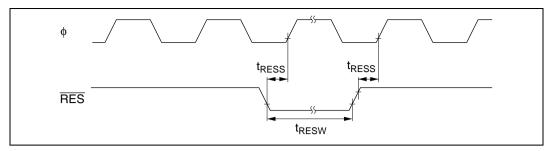


Figure 26.8 Reset Input Timing

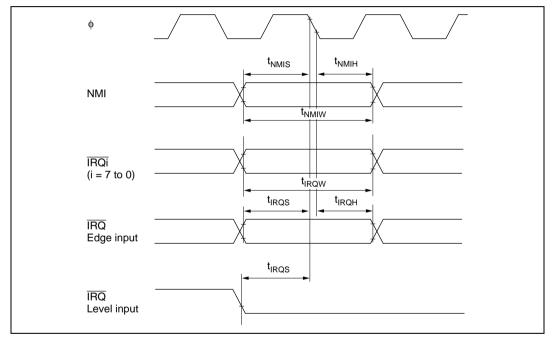


Figure 26.9 Interrupt Input Timing

26.7.4 Bus Timing

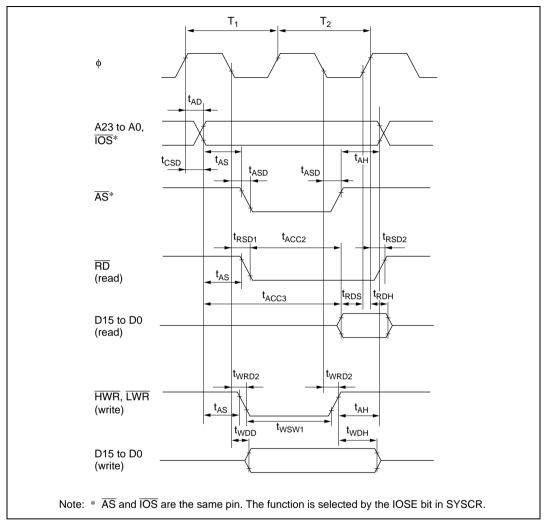


Figure 26.10 Basic Bus Timing (Two-State Access)

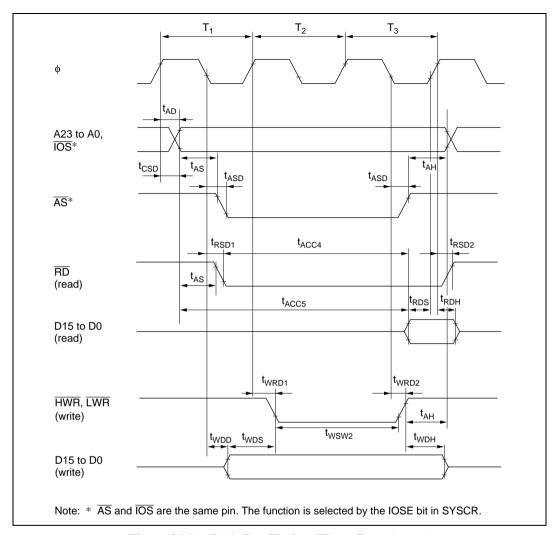


Figure 26.11 Basic Bus Timing (Three-State Access)

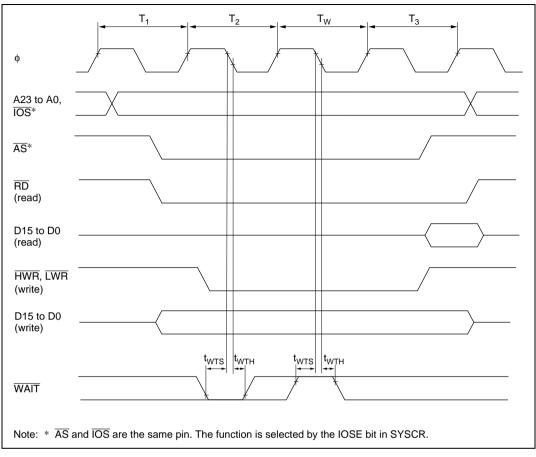


Figure 26.12 Basic Bus Timing (Three-State Access with One Wait State)

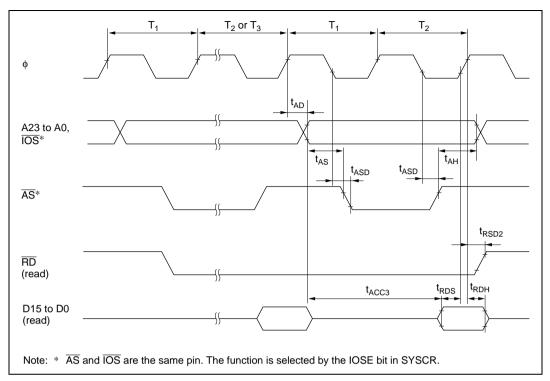


Figure 26.13 Burst ROM Access Timing (Two-State Access)

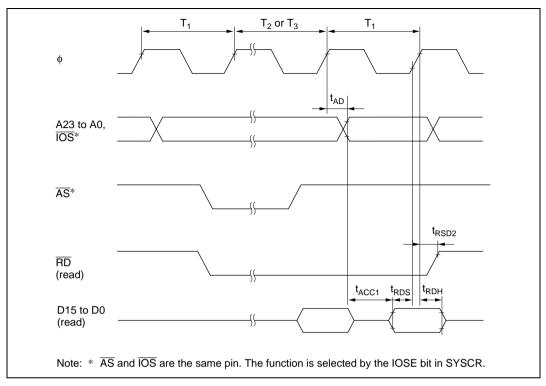


Figure 26.14 Burst ROM Access Timing (One-State Access)

26.7.5 Timing of On-Chip Supporting Modules

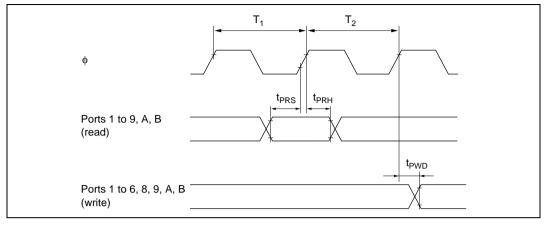


Figure 26.15 I/O Port Input/Output Timing

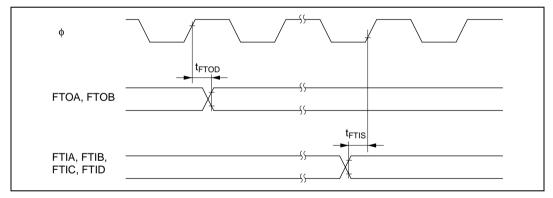


Figure 26.16 FRT Input/Output Timing

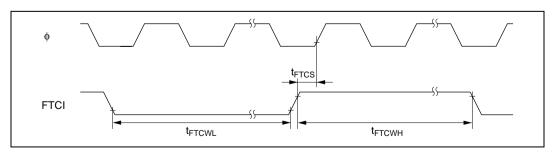


Figure 26.17 FRT Clock Input Timing

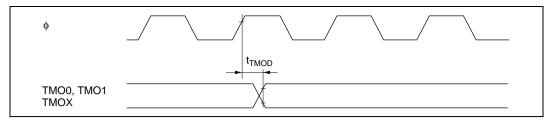


Figure 26.18 8-Bit Timer Output Timing

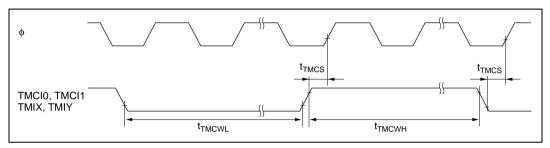


Figure 26.19 8-Bit Timer Clock Input Timing

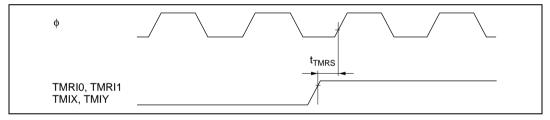


Figure 26.20 8-Bit Timer Reset Input Timing

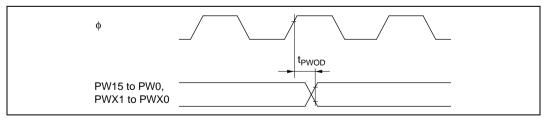


Figure 26.21 PWM, PWMX Output Timing

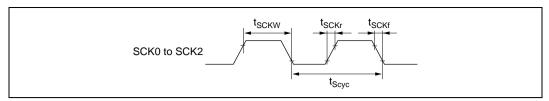


Figure 26.22 SCK Clock Input Timing

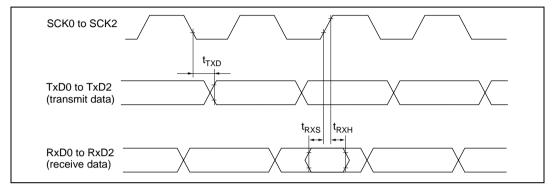


Figure 26.23 SCI Input/Output Timing (Synchronous Mode)

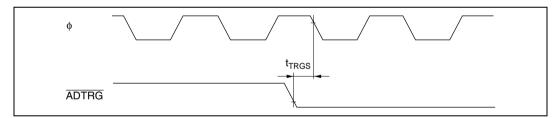


Figure 26.24 A/D Converter External Trigger Input Timing

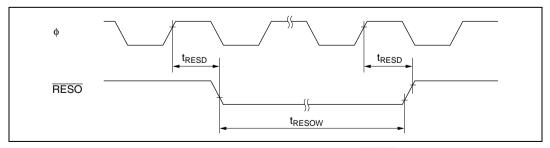


Figure 26.25 WDT Output Timing (RESO)

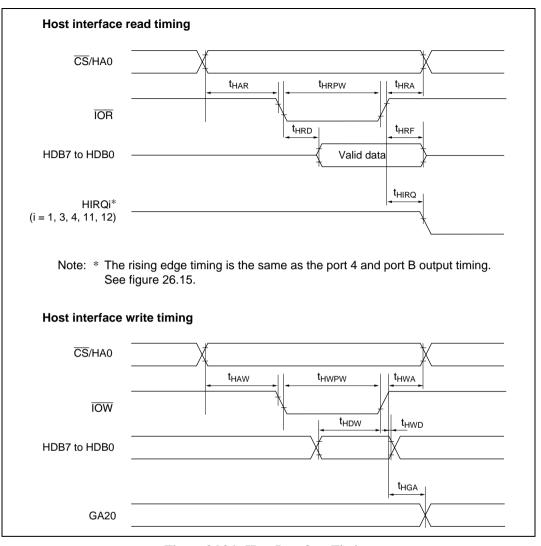


Figure 26.26 Host Interface Timing

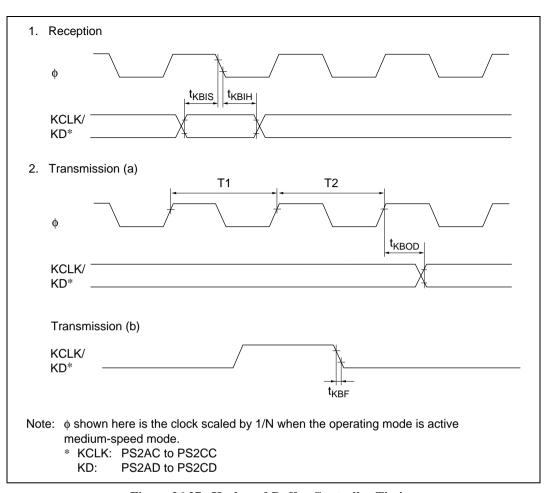


Figure 26.27 Keyboard Buffer Controller Timing

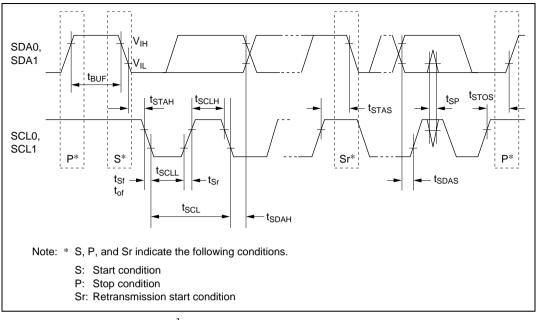


Figure 26.28 I²C Bus Interface Input/Output Timing (Option)

Appendix A Instruction Set

A.1 Instruction

Operation Notation

Rd	General register (destination)*1
Rs	General register (source)*1
Rn	General register*1
ERn	General register (32-bit register)
MAC	Multiply-and-accumulate register (32-bit register)*2
(EAd)	Destination operand
(EAs)	Source operand
EXR	Extend register
CCR	Condition code register
N	N (negative) flag in CCR
Z	Z (zero) flag in CCR
V	V (overflow) flag in CCR
С	C (carry) flag in CCR
PC	Program counter
SP	Stack pointer
#IMM	Immediate data
disp	Displacement
+	Addition
_	Subtraction
×	Multiplication
÷	Division
^	Logical AND
<u> </u>	Logical OR
\oplus	Exclusive logical OR
\rightarrow	Transfer from left-hand operand to right-hand operand, or transition from left-hand state to right-hand state
7	NOT (logical complement)
() < >	Operand contents
:8/:16/:24/:32	8-, 16-, 24-, or 32-bit length

Notes: 1. General registers include 8-bit registers (R0H to R7H, R0L to R7L), 16-bit registers (R0 to R7, E0 to E7), and 32-bit registers (ER0 to ER7).

2. MAC instructions cannot be used in this LSI.

Condition Code Notation

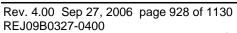
Symbol	Meaning
‡	Changes according operation result.
*	Indeterminate (value not guaranteed).
0	Always cleared to 0.
1	Always set to 1.
_	Not affected by operation result.

Table A.1 Instruction Set

1. Data Transfer Instructions

				Addressing Mode and Instruction Length (Bytes))			Con	No. of States*1					
	Mnemonic		xx#	Rn	@ERn	@ (d,ERn)	@-ERn/@ERn+	@ aa	@ (q,PC)	@ @aa	1	Operation	ı	н	N	z	v	С	Normal	Advanced
MOV	MOV.B #xx:8,Rd	В	2									#xx:8→Rd8	-	-	\$	‡	0	_	1	
	MOV.B Rs,Rd	В		2								Rs8→Rd8	-	-	\$	‡	0	_	1	ı
	MOV.B @ERs,Rd	В			2							@ERs→Rd8	<u> </u>	-	\$	‡	0	_	2	2
	MOV.B @(d:16,ERs),Rd	В				4						@(d:16,ERs)→Rd8	<u> </u>	-	‡	‡	0	_	3	3
	MOV.B @(d:32,ERs),Rd	В				8						@(d:32,ERs)→Rd8	-	<u> </u>	\$	‡	0	_	5	5
	MOV.B @ERs+,Rd	В					2					@ERs→Rd8,ERs32+1→ERs32	-	<u> </u>	\$	‡	0	_	3	3
	MOV.B @aa:8,Rd	В						2				@aa:8→Rd8	_	<u> </u>	1	‡	0	_	2	2
	MOV.B @aa:16,Rd	В						4				@aa:16→Rd8	-	-	‡	‡	0	_	3	3
	MOV.B @aa:32,Rd	В						6				@aa:32→Rd8	-	-	‡	‡	0	_	4	ŀ
	MOV.B Rs,@ERd	В			2							Rs8→@ERd	F		1	1	0		- 2	2
	MOV.B Rs,@(d:16,ERd)	В				4						Rs8→@(d:16,ERd)	<u> </u>	<u> </u>	\$	1	0	_	3	3
	MOV.B Rs,@(d:32,ERd)	В				8						Rs8→@(d:32,ERd)	<u> </u>	<u> </u>	\$	1	0	_	5	 j
	MOV.B Rs,@-ERd	В					2					ERd32-1→ERd32,Rs8→@ERd	-	-	‡	‡	0	_	3	3
	MOV.B Rs,@aa:8	В						2				Rs8→@aa:8	-	-	‡	‡	0	_	2	2
	MOV.B Rs,@aa:16	В						4				Rs8→@aa:16	<u> </u>	<u> </u>	\$	1	0	_	3	3
	MOV.B Rs,@aa:32	В						6				Rs8→@aa:32	-	-	‡	‡	0	_	4	ŀ
	MOV.W #xx:16,Rd	W	4									#xx:16→Rd16	<u> </u>	<u> </u>	\$	1	0	_	2	2
	MOV.W Rs,Rd	W		2								Rs16→Rd16	-	-	‡	‡	0	_	1	
	MOV.W @ERs,Rd	W			2							@ERs→Rd16	-	-	‡	‡	0	_	2	?
	MOV.W @(d:16,ERs),Rd	W				4						@(d:16,ERs)→Rd16	-	-	‡	‡	0	_	3	3
	MOV.W @(d:32,ERs),Rd	W				8						@(d:32,ERs)→Rd16	-	-	‡	‡	0	_	5	
	MOV.W @ERs+,Rd	W					2					@ERs→Rd16,ERs32+2→ERs32	-	-	‡	‡	0	_	3	3
	MOV.W @aa:16,Rd	W						4				@aa:16→Rd16	<u> </u>	<u> </u>	\$	1	0	_	3	3
	MOV.W @aa:32,Rd	W						6				@aa:32→Rd16	1-	-	‡	1	0	_	4	ļ
	MOV.W Rs,@ERd	W			2							Rs16→@ERd	1-	-	\$	1	0	_	2	?
	MOV.W Rs,@(d:16,ERd)	W				4						Rs16→@(d:16,ERd)	1-	-	‡	‡	0	_	3	3
	MOV.W Rs,@(d:32,ERd)	W				8						Rs16→@(d:32,ERd)	1-	-	\$	1	0	_	5	;
	MOV.W Rs,@-ERd	W					2					ERd32-2→ERd32,Rs16→@ERd	1-	-	‡	1	0	_	3	3
	MOV.W Rs,@aa:16	w						4				Rs16→@aa:16	1-	<u> </u>	1	‡	0	_	3	3
	MOV.W Rs,@aa:32	w						6				Rs16→@aa:32	1-	1-	1	‡	0	_	4	ļ

			Ir			ssir on l	•)		(Con	е	No. of States*1				
	Mnemonic	Size	#xx	Rn	@ERn	@ (d,ERn)	@-ERn/@ERn+	@ aa	@(d,PC)	@ @aa	1	Operation	ı	н	N	z	v	С	Normal	Advanced
MOV	MOV.L #xx:32,ERd	L	6									#xx:32→ERd32	_	_	1	\$	0	_	3	3
	MOV.L ERs,ERd	L		2								ERs32→ERd32	-	_	\$	1	0	_	,	1
	MOV.L @ERs,ERd	L			4							@ERs→ERd32	-	_	\$	1	0	_	4	4
	MOV.L @(d:16,ERs),ERd	L				6						@(d:16,ERs)→ERd32	-	_	\$	1	0	_		5
	MOV.L @(d:32,ERs),ERd	L				10						@(d:32,ERs)→ERd32	-	_	\$	1	0	_	7	7
	MOV.L @ERs+,ERd	L					4					@ERs→ERd32,ERs32+4→ERs32	-	_	1	1	0	_		5
	MOV.L @aa:16,ERd	L						6				@aa:16→ERd32	-	_	1	1	0	_		5
	MOV.L @aa:32,ERd	L						8				@aa:32→ERd32	-	_	1	1	0	_	6	3
	MOV.L ERs,@ERd	L			4							ERs32→@ERd	<u> </u>	_	1	1	0	_	4	4
	MOV.L ERs,@(d:16,ERd)	L				6						ERs32→@(d:16,ERd)	<u> </u>	_	1	1	0	_		5
	MOV.L ERs,@(d:32,ERd)	L				10						ERs32→@(d:32,ERd)	-	_	1	1	0	_	7	7
	MOV.L ERs,@-ERd	L					4					ERd32-4→ERd32,ERs32→@ERd	-	_	1	1	0	_		5
	MOV.L ERs,@aa:16	L						6				ERs32→@aa:16	<u> </u>	_	1	1	0	_		5
	MOV.L ERs,@aa:32	L						8				ERs32→@aa:32	<u> </u>	_	1	1	0	_	6	3
POP	POP.W Rn	W									2	@SP→Rn16,SP+2→SP	<u> </u>	_	1	1	0	_	:	3
	POP.L ERn	L									4	@SP→ERn32,SP+4→SP	-	_	1	1	0	_		5
PUSH	PUSH.W Rn	W									2	SP-2→SP,Rn16→@SP	<u> </u>	_	\$	1	0	_	3	3
	PUSH.L ERn	L									4	SP-4→SP,ERn32→@SP	-	_	1	1	0	_	Ę	5
LDM* ⁴	LDM @SP+,(ERm-ERn)	L									4	(@SP→ERn32,SP+4→SP) Repeated for each restored register.	-	-	-	_	_	_	7/9/1	1 [1]
STM* ⁴	STM (ERm-ERn),@-SP	L									4	(SP-4→SP,ERn32→@SP) Repeated for each saved register.	-	-	-	-	_	-	7/9/1	1 [1]
MOVFPE	MOVFPE @aa:16,Rd	Car	nnot	be	use	d wi	th th	nis L	SI.			•	_			•		•	[2	2]
MOVTPE	MOVTPE Rs,@aa:16																		[2	2]

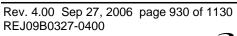




2. Arithmetic Instructions

			In	Ad			-		e an (By)			Con	diti	on (Cod	е	No. Stat	of tes ^{*1}
	Mnemonic	Size	*xx	Rn	@ERn	@(d,ERn)	@-ERn/@ERn+	@aa	@(d,PC)	@ @aa	1	Operation	1	н	N	z	v	С	Normal	Advanced
ADD	ADD.B #xx:8,Rd	В	2									Rd8+#xx:8→Rd8	-	1	1	‡	1	‡	1	1
	ADD.B Rs,Rd	В		2								Rd8+Rs8→Rd8	_	\$	\$	‡	‡	‡	1	1
	ADD.W #xx:16,Rd	W	4									Rd16+#xx:16→Rd16	_	[3]	\$	‡	‡	‡	2	2
	ADD.W Rs,Rd	W		2								Rd16+Rs16→Rd16	_	[3]	\$	‡	‡	‡	1	1
	ADD.L #xx:32,ERd	L	6									ERd32+#xx:32→ERd32	-	[4]	1	‡	1	‡	3	3
	ADD.L ERs,ERd	L		2								ERd32+ERs32→ERd32	-	[4]	1	‡	1	‡	1	1
ADDX	ADDX #xx:8,Rd	В	2									Rd8+#xx:8+C→Rd8	-	‡	1	[5]	1	‡	1	1
	ADDX Rs,Rd	В		2								Rd8+Rs8+C→Rd8	-	‡	\$	[5]	1	‡	1	1
ADDS	ADDS #1,ERd	L		2								ERd32+1→ERd32	-	-	_	_	_	_	1	1
	ADDS #2,ERd	L		2								ERd32+2→ERd32	-	_	_	_	_	_	1	1
	ADDS #4,ERd	L		2								ERd32+4→ERd32	-	-	-	_	_	_	1	1
INC	INC.B Rd	В		2								Rd8+1→Rd8	-	-	\$	‡	‡	_	1	1
	INC.W #1,Rd	w		2								Rd16+1→Rd16	-	-	\$	‡	‡	_	1	1
	INC.W #2,Rd	w		2								Rd16+2→Rd16	-	-	\$	‡	‡	_	1	1
	INC.L #1,ERd	L		2								ERd32+1→ERd32	-	_	1	‡	1	_	1	1
	INC.L #2,ERd	L		2								ERd32+2→ERd32	-	-	\$	‡	‡	_	1	1
DAA	DAA Rd	В		2								Rd8 decimal adjust →Rd8	-	*	\$	‡	*	‡	1	1
SUB	SUB.B Rs,Rd	В		2								Rd8-Rs8→Rd8	-	‡	\$	‡	‡	‡	1	1
	SUB.W #xx:16,Rd	w	4									Rd16-#xx:16→Rd16	-	[3]	\$	‡	‡	‡	2	2
	SUB.W Rs,Rd	w		2								Rd16-Rs16→Rd16	1-	[3]	\$	‡	‡	‡	1	1
	SUB.L #xx:32,ERd	L	6									ERd32-#xx:32→ERd32	-	[4]	\$	‡	‡	‡	3	3
	SUB.L ERs,ERd	L		2								ERd32-ERs32→ERd32	-	[4]	\$	‡	‡	‡	1	1
SUBX	SUBX #xx:8,Rd	В	2									Rd8-#xx:8-C→Rd8	-	‡	\$	[5]	‡	‡	1	1
	SUBX Rs,Rd	В		2								Rd8-Rs8-C→Rd8	-	‡	\$	[5]	‡	‡	1	1
SUBS	SUBS #1,ERd	L		2								ERd32-1→ERd32	-	_	-	_	_	_	1	1
	SUBS #2,ERd	L		2								ERd32-2→ERd32	1-	_	-	_	_	_	1	1
	SUBS #4,ERd	L		2								ERd32-4→ERd32	-	-	-	_	_	_	1	1
DEC	DEC.B Rd	В		2								Rd8-1→Rd8	-	-	\$	‡	1	-	1	1
	DEC.W #1,Rd	w		2								Rd16-1→Rd16	1-	_	\$	‡	1	_	1	1
	DEC.W #2,Rd	w		2								Rd16-2→Rd16	-	-	1	‡	‡	_	1	1
	DEC.L #1,ERd	L		2								ERd32-1→ERd32	1-	_	\$	‡	1	_	1	1
	DEC.L #2,ERd	L		2								ERd32-2→ERd32	-	-	1	‡	‡	_	1	1

			In			ssin on L	_)			Con	ditio	on C	Code	9	No. Stat	-
	Mnemonic	Size	#xx	Rn	@ERn	@(d,ERn)	@-ERn/@ERn+	@aa	@(q,PC)	@ @ aa	ı	Operation	ı	н	N	z	v	С	Normal	Advanced
DAS	DAS Rd	В		2								Rd8 decimal adjust →Rd8	-	*	\$	\$	*	_	1	ı
MULXU	MULXU.B Rs,Rd	В		2								Rd8×Rs8→Rd16 (unsigned multiplication)	-	_	_	_	_	_	1	2
	MULXU.W Rs,ERd	W		2								Rd16×Rs16→ERd32 (unsigned multiplication)	-	-	_	_	_	_	2	0
MULXS	MULXS.B Rs,Rd	В		4								Rd8×Rs8→Rd16 (signed multiplication)	_	_	\$	\$	_	_	1	3
	MULXS.W Rs,ERd	W		4								Rd16×Rs16→ERd32 (signed multiplication)	-	_	\$	\$	_	_	2	1
DIVXU	DIVXU.B Rs,Rd	В		2								Rd16÷Rs8→Rd16 (RdH: remainder, RdL: quotient) (unsigned division)	_	_	[6]	[7]	_	-	1	2
	DIVXU.W Rs,ERd	W		2								ERd32÷Rs16→ERd32 (Ed: remainder, Rd: quotient) (unsigned division)	-	_	[6]	[7]	_	_	2	0
DIVXS	DIVXS.B Rs,Rd	В		4								Rd16÷Rs8→Rd16 (RdH: remainder, RdL: quotient) (signed division)	-	_	[8]	[7]	_	_	1	3
	DIVXS.W Rs,ERd	W		4								ERd32÷Rs16→ERd32 (Ed: remainder, Rd: quotient) (signed division)	-	_	[8]	[7]	_	_	2	1
CMP	CMP.B #xx:8,Rd	В	2									Rd8-#xx:8	1-	1	\$	\$	‡	‡	1	
	CMP.B Rs,Rd	В		2								Rd8-Rs8	-	1	\$	\$	‡	‡	1	
	CMP.W #xx:16,Rd	W	4									Rd16-#xx:16	-	[3]	\$	‡	‡	‡	2	2
	CMP.W Rs,Rd	w		2								Rd16-Rs16	1-	[3]	\$	‡	‡	‡	1	ı
	CMP.L #xx:32,ERd	L	6									ERd32-#xx:32	1-	[4]	‡	\$	‡	‡	3	3
	CMP.L ERs,ERd	L		2								ERd32-ERs32	-	[4]	‡	\$	‡	‡	1	I
NEG	NEG.B Rd	В		2								0-Rd8→Rd8	1-	1	‡	\$	‡	‡	1	l
	NEG.W Rd	W		2								0-Rd16→Rd16	1-	1	‡	\$	‡	‡	1	l
	NEG.L ERd	L		2								0-ERd32→ERd32	1-	1	‡	\$	‡	‡	1	I
EXTU	EXTU.W Rd	W		2								0 → (<bits 15="" 8="" to=""> of Rd16)</bits>	1-	_	0	\$	0	_	1	I
	EXTU.L ERd	L		2								0 → (<bits 16="" 31="" to=""> of ERd32)</bits>	-	_	0	\$	0	_	1	
EXTS	EXTS.W Rd	W		2								(<bit 7=""> of Rd16) → (<bits 15="" 8="" to=""> of Rd16)</bits></bit>	_		\$	\$	0	_	1	
	EXTS.L ERd	L		2								(<bit 15=""> of ERd32) → (<bits 16="" 31="" to=""> of ERd32)</bits></bit>		_	\$	\$	0		1	1
TAS	TAS @ERd*3	В			4							@ERd-0 → CCR set, (1) → (<bit 7=""> of @ERd)</bit>		_	\$	\$	0		4	+





			lr		dre: ucti)			Cor	diti	on (Cod	е		. of tes ^{*1}
	Mnemonic	Size	xx#	Rn	@ERn	@(d,ERn)	@-ERn/@ERn+	@aa	@(d,PC)	@ @ aa	ı	Operation	ı	н	N	z	v	С	Normal	Advanced
MAC	MAC @ERn+,@ERm+	Car	nnot	be	use	d wi	th th	nis L	SI.				ı						[:	2]
CLRMAC	CLRMAC	1																		
LDMAC	LDMAC ERs,MACH																			
	LDMAC ERs,MACL																			
STMAC	STMAC MACH,ERd																			
	STMAC MACL,ERd																			

3. Logic Instructions

			lr			ssir on l)		(Con	ditio	on C	Code	е	No. Stat	of es*1
	Mnemonic	Size	xx#	Rn	@ ERn	@(d,ERn)	@-ERn/@ERn+	@ aa	@(d,PC)	@ @aa	ı	Operation	ı	н	N	z	v	С	Normal	Advanced
AND	AND.B #xx:8,Rd	В	2									Rd8∧#xx:8→Rd8	_	-	‡	‡	0	_		1
	AND.B Rs,Rd	В		2								Rd8∧Rs8→Rd8	<u> </u>	_	‡	‡	0	_	•	1
	AND.W #xx:16,Rd	w	4									Rd16∧#xx:16→Rd16	<u> </u>	_	‡	‡	0	_	2	2
	AND.W Rs,Rd	w		2								Rd16∧Rs16→Rd16	<u> </u>	_	\$	‡	0	_	•	1
	AND.L #xx:32,ERd	L	6									ERd32∧#xx:32→ERd32	<u> </u>	_	‡	‡	0	_	3	3
	AND.L ERs,ERd	L		4								ERd32∧ERs32→ERd32	<u> </u>	_	\$	‡	0	_	2	2
OR	OR.B #xx:8,Rd	В	2									Rd8∨#xx:8→Rd8	<u> </u>	_	‡	‡	0	_		1
	OR.B Rs,Rd	В		2								Rd8∨Rs8→Rd8	<u> </u>	_	\$	‡	0	_		1
	OR.W #xx:16,Rd	w	4									Rd16∨#xx:16→Rd16	<u> </u>	_	\$	‡	0	_	2	2
	OR.W Rs,Rd	w		2								Rd16∨Rs16→Rd16	<u> </u>	_	‡	‡	0	_		1
	OR.L #xx:32,ERd	L	6									ERd32√#xx:32→ERd32	<u> </u>	_	‡	‡	0	_	3	3
	OR.L ERs,ERd	L		4								ERd32√ERs32→ERd32	<u> </u>	_	‡	‡	0	_	2	2
XOR	XOR.B #xx:8,Rd	В	2									Rd8⊕#xx:8→Rd8	<u> </u>	_	‡	‡	0	_	•	1
	XOR.B Rs,Rd	В		2								Rd8⊕Rs8→Rd8	<u> </u>	_	‡	‡	0	_	•	1
	XOR.W #xx:16,Rd	w	4									Rd16⊕#xx:16→Rd16	<u> </u>	_	\$	‡	0	_	2	2
	XOR.W Rs,Rd	w		2								Rd16⊕Rs16→Rd16	<u> </u>	_	‡	‡	0	_		1
	XOR.L #xx:32,ERd	L	6									ERd32⊕#xx:32→ERd32	_	_	‡	‡	0	_	3	3
	XOR.L ERs,ERd	L		4								ERd32⊕ERs32→ERd32	-	_	1	‡	0	_	2	2
NOT	NOT.B Rd	В		2								¬ Rd8→Rd8	_	_	‡	‡	0	_	•	1
	NOT.W Rd	W		2								¬ Rd16→Rd16	_	_	‡	‡	0	_	•	1
	NOT.L ERd	L		2								¬ ERd32→ERd32	_	_	1	‡	0	_		1



4. Shift Instructions

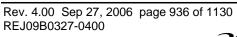
			In	Ad	dre: ucti						5)								•	Con	ditio	on (Cod	e		o. of tes*1
	Mnemonic	Size	xx#	Rn	@ERn	@(d,ERn)	@-ERn/@ERn+	@aa	@(d,PC)	@ @ aa	1			Oţ	peratio	on			ı	н	N	z	v	С	Normal	Advanced
SHAL	SHAL.B Rd	В		2															_	_	‡	\$	‡	‡		1
	SHAL.B #2,Rd	В		2															_	_	‡	\$	‡	‡		1
	SHAL.W Rd	w		2								J∏⊶						- −0	_	_	‡	\$	‡	‡		1
	SHAL.W #2,Rd	W		2								С	MSB	-			LSB		_	_	‡	\$	‡	‡		1
	SHAL.L ERd	L		2															_	_	‡	‡	‡	‡		1
	SHAL.L #2,ERd	L		2															_	_	\$	\$	‡	\$		1
SHAR	SHAR.B Rd	В		2															_	_	‡	\$	0	\$		1
	SHAR.B #2,Rd	В		2								1	_						_	_	1	‡	0	‡		1
	SHAR.W Rd	w		2								۱,	_					-	_	_	1	‡	0	‡		1
	SHAR.W #2,Rd	w		2									MSB	_		-	LSB	С	_	_	1	‡	0	‡		1
	SHAR.L ERd	L		2								1							_	_	1	‡	0	‡		1
	SHAR.L #2,ERd	L		2								1							_	_	\$	‡	0	‡		1
SHLL	SHLL.B Rd	В		2															-	_	\$	‡	0	‡		1
	SHLL.B #2,Rd	В		2								1							_	_	\$	‡	0	‡		1
	SHLL.W Rd	w		2								Ī₽∙						- −0	_	_	\$	‡	0	‡		1
	SHLL.W #2,Rd	w		2								С	MSB	-			LSB		_	_	\$	‡	0	‡		1
	SHLL.L ERd	L		2															_	_	\$	‡	0	‡		1
	SHLL.L #2,ERd	L		2															_	_	\$	‡	0	‡		1
SHLR	SHLR.B Rd	В		2															-	_	0	‡	0	‡		1
	SHLR.B #2,Rd	В		2															_	_	0	‡	0	‡		1
	SHLR.W Rd	w		2								0→						→	_	_	0	‡	0	‡		1
	SHLR.W #2,Rd	w		2									MSB	_		-	LSB	С	_	_	0	‡	0	‡		1
	SHLR.L ERd	L		2								1							_	_	0	‡	0	‡		1
	SHLR.L #2,ERd	L		2															_	_	0	‡	0	‡		1
ROTXL	ROTXL.B Rd	В		2															-	-	\$	‡	0	‡		1
	ROTXL.B #2,Rd	В		2															_	_	\$	\$	0	\$		1
	ROTXL.W Rd	w		2								1 [-	-					_	_	_	\$	\$	0	\$		1
	ROTXL.W #2,Rd	w		2									ı ∟ M:	SB	_		— L	 SB	_	_	\$	\$	0	\$		1
	ROTXL.L ERd	L		2								1					_,		_	_	\$	\$	0	\$		1
	ROTXL.L #2,ERd	L		2															_	_	\$	‡	0	\$		1

			Ir			ssin on l)		(Con	ditio	on C	ode	е	No. State	
	Mnemonic	Size	xx#	Rn	@ERn	@(d,ERn)	@-ERn/@ERn+	@aa	@(d,PC)	@ @ aa	ı	Operation	ı	Н	N	z	v	ပ	Normal	Advanced
ROTXR	ROTXR.B Rd	В		2									_	_	‡	1	0	‡	1	
	ROTXR.B #2,Rd	В		2									_	_	‡	‡	0	\leftrightarrow	1	
	ROTXR.W Rd	W		2									_	_	‡	‡	0	‡	1	
	ROTXR.W #2,Rd	W		2								MSB → LSB C	_	_	‡	\$	0	‡	1	
	ROTXR.L ERd	L		2								WISB - LSB C	_	_	‡	\$	0	‡	1	
	ROTXR.L #2,ERd	L		2									_	_	‡	‡	0	‡	1	
ROTL	ROTL.B Rd	В		2									_	_	‡	\$	0	‡	1	
	ROTL.B #2,Rd	В		2									_	_	‡	\$	0	‡	1	
	ROTL.W Rd	W		2									_	_	‡	\$	0	‡	1	
	ROTL.W #2,Rd	W		2								C MSB ← LSB	_	_	‡	‡	0	‡	1	
	ROTL.L ERd	L		2								C MISB - LOB	_	_	‡	‡	0	‡	1	
	ROTL.L #2,ERd	L		2									_	_	‡	‡	0	‡	1	
ROTR	ROTR.B Rd	В		2									_	_	‡	‡	0	‡	1	
	ROTR.B #2,Rd	В		2									_	_	‡	1	0	‡	1	
	ROTR.W Rd	w		2									_	_	‡	1	0	‡	1	
	ROTR.W #2,Rd	W		2								MSB → LSB C	_	_	‡	1	0	‡	1	
	ROTR.L ERd	L		2								MSB ──► LSB C	_	_	‡	1	0	‡	1	
	ROTR.L #2,ERd	L		2									_	_	‡	1	0	‡	1	

5. Bit-Manipulation Instructions

			In		dre:		_)		(Con	ditio	on C	Code	9	No. Stat	of es*1
	Mnemonic	Size	xx#	Rn	@ERn	@ (d,ERn)	@-ERn/@ERn+	@ aa	@(q,PC)	@ @aa		Operation	ı	н	N	z	v	С	Normal	Advanced
BSET	BSET #xx:3,Rd	В		2								(#xx:3 of Rd8)←1	_	_	_	_	_	_	1	I
	BSET #xx:3,@ERd	В			4							(#xx:3 of @ERd)←1	_	_	_	_	_	_	4	ļ
	BSET #xx:3,@aa:8	В						4				(#xx:3 of @aa:8)←1	_	_	_	_	_	_	4	1
	BSET #xx:3,@aa:16	В						6				(#xx:3 of @aa:16)←1	_	_	_	_	_	_	5	5
	BSET #xx:3,@aa:32	В						8				(#xx:3 of @aa:32)←1	_	_	_	_	_	_	6	;
	BSET Rn,Rd	В		2								(Rn8 of Rd8)←1	_	_	-	_	-	_	1	I
	BSET Rn,@ERd	В			4							(Rn8 of @ERd)←1	_	_	_	_	_	_	4	ļ
	BSET Rn,@aa:8	В						4				(Rn8 of @aa:8)←1	_	_	_	_	_	_	4	1
	BSET Rn,@aa:16	В						6				(Rn8 of @aa:16)←1	_	_	_	_	_	_	5	5
	BSET Rn,@aa:32	В						8				(Rn8 of @aa:32)←1	_	_	_	_	_	_	6	;
BCLR	BCLR #xx:3,Rd	В		2								(#xx:3 of Rd8)←0	_	_	_	_	_	_	1	ı
	BCLR #xx:3,@ERd	В			4							(#xx:3 of @ERd)←0	_	_	_	_	_	_	4	ŀ
	BCLR #xx:3,@aa:8	В						4				(#xx:3 of @aa:8)←0	_	_	_	_	_	_	4	1
	BCLR #xx:3,@aa:16	В						6				(#xx:3 of @aa:16)←0	_	_	_	_	_	_	5	
	BCLR #xx:3,@aa:32	В						8				(#xx:3 of @aa:32)←0	_	_	_	_	_	_	6	;
	BCLR Rn,Rd	В		2								(Rn8 of Rd8)←0	_	_	_	_	_	_	1	
	BCLR Rn,@ERd	В			4							(Rn8 of @ERd)←0	_	_	_	_	_	_	4	ŀ
	BCLR Rn,@aa:8	В						4				(Rn8 of @aa:8)←0	_	_	_	_	_	_	4	1
	BCLR Rn,@aa:16	В						6				(Rn8 of @aa:16)←0	_	_	_	_	_	_	5	
	BCLR Rn,@aa:32	В						8				(Rn8 of @aa:32)←0	_	_	_	_	_	_	ε	6
BNOT	BNOT #xx:3,Rd	В		2								(#xx:3 of Rd8)← [¬ (#xx:3 of Rd8)]	_	_	_	_	_	_	1	
	BNOT #xx:3,@ERd	В			4							(#xx:3 of @ERd)← [¬ (#xx:3 of @ERd)]	_	_	-	_	-	_	4	1
	BNOT #xx:3,@aa:8	В						4				(#xx:3 of @aa:8)← [¬ (#xx:3 of @aa:8)]	_	-	-	_	_	_	4	1
	BNOT #xx:3,@aa:16	В						6				(#xx:3 of @aa:16)← [¬ (#xx:3 of @aa:16)]	-	_	-	_	_		5	,
	BNOT #xx:3,@aa:32	В						8				(#xx:3 of @aa:32)← [¬ (#xx:3 of @aa:32)]							6	;
	BNOT Rn,Rd	В		2								(Rn8 of Rd8)← [¬ (Rn8 of Rd8)]	Ŀ	Ŀ	Ŀ		Ŀ		1	I
	BNOT Rn,@ERd	В			4							(Rn8 of @ERd)← [¬ (Rn8 of @ERd)]	Ŀ	Ŀ	Ŀ		Ŀ		4	1
	BNOT Rn,@aa:8	В						4				(Rn8 of @aa:8)← [¬ (Rn8 of @aa:8)]							4	ŀ
	BNOT Rn,@aa:16	В						6				(Rn8 of @aa:16)← [¬ (Rn8 of @aa:16)]							5	j .
	BNOT Rn,@aa:32	В						8				(Rn8 of @aa:32)← [¬ (Rn8 of @aa:32)]			-	_	_		6	; ;

			In		dre: ucti		_)		(Con	ditio	on C	Code	9	No Stat	. of tes ^{*1}
	Mnemonic	Size	#xx	Rn	@ERn	@(d,ERn)	@-ERn/@ERn+	@aa	@(d,PC)	@ @ aa	I	Operation	ı	н	N	z	v	С	Normal	Advanced
BTST	BTST #xx:3,Rd	В		2								¬ (#xx:3 of Rd8)→Z	-	_	_	‡	_	_		1
	BTST #xx:3,@ERd	В			4							¬ (#xx:3 of @ERd)→Z	_	_	_	‡	_	_	,	3
	BTST #xx:3,@aa:8	В						4				¬ (#xx:3 of @aa:8)→Z	-	_	_	‡	_	_		3
	BTST #xx:3,@aa:16	В						6				¬ (#xx:3 of @aa:16)→Z	-	_	_	‡	_	_		4
	BTST #xx:3,@aa:32	В						8				¬ (#xx:3 of @aa:32)→Z	-	_	_	‡	_	_		5
	BTST Rn,Rd	В		2								¬ (Rn8 of Rd8)→Z	-	_	_	‡	_	_		1
	BTST Rn,@ERd	В			4							¬ (Rn8 of @ERd)→Z	-	_	_	‡	_	_	:	3
	BTST Rn,@aa:8	В						4				¬ (Rn8 of @aa:8)→Z	-	_	_	‡	_	_	:	3
	BTST Rn,@aa:16	В						6				¬ (Rn8 of @aa:16)→Z	-	_	_	‡	_	_		4
	BTST Rn,@aa:32	В						8				¬ (Rn8 of @aa:32)→Z	-	_	_	‡	_	_		5
BLD	BLD #xx:3,Rd	В		2								(#xx:3 of Rd8)→C	-	_	_	_	_	‡		1
	BLD #xx:3,@ERd	В			4							(#xx:3 of @ERd)→C	-	_	_	_	_	‡		3
	BLD #xx:3,@aa:8	В						4				(#xx:3 of @aa:8)→C	_	_	_	_	_	‡	:	3
	BLD #xx:3,@aa:16	В						6				(#xx:3 of @aa:16)→C	-	_	_	_	_	‡		4
	BLD #xx:3,@aa:32	В						8				(#xx:3 of @aa:32)→C	_	_	_	_	_	\$		5
BILD	BILD #xx:3,Rd	В		2								¬ (#xx:3 of Rd8)→C	-	_	_	_	_	‡		1
	BILD #xx:3,@ERd	В			4							¬ (#xx:3 of @ERd)→C	-	_	_	_	_	‡		3
	BILD #xx:3,@aa:8	В						4				¬ (#xx:3 of @aa:8)→C	_	_	_	_	_	‡	;	3
	BILD #xx:3,@aa:16	В						6				¬ (#xx:3 of @aa:16)→C	-	_	_	_	_	‡		4
	BILD #xx:3,@aa:32	В						8				¬ (#xx:3 of @aa:32)→C	-	_	_	_	_	‡		5
BST	BST #xx:3,Rd	В		2								C→(#xx:3 of Rd8)	-	_	_	_	_	_		1
	BST #xx:3,@ERd	В			4							C→(#xx:3 of @ERd)	-	_	_	_	_	_		4
	BST #xx:3,@aa:8	В						4				C→(#xx:3 of @aa:8)	-	_	_	_	_	_		4
	BST #xx:3,@aa:16	В						6				C→(#xx:3 of @aa:16)	-	_	_	_	_	_		5
	BST #xx:3,@aa:32	В						8				C→(#xx:3 of @aa:32)	_	_	_	_	_	_	-	6
BIST	BIST #xx:3,Rd	В		2								¬ C→(#xx:3 of Rd8)	_	_	_	_	_	_		1
	BIST #xx:3,@ERd	В			4							¬ C→(#xx:3 of @ERd)	_	_	_	_	_	_		4
	BIST #xx:3,@aa:8	В						4				¬ C→(#xx:3 of @aa:8)	_	_	_	_	_	_		4
	BIST #xx:3,@aa:16	В						6				¬ C→(#xx:3 of @aa:16)	_	_	_	_	_	_		5
	BIST #xx:3,@aa:32	В						8				¬ C→(#xx:3 of @aa:32)	_	_	_	_	_	_	-	6





			lr				ng N Len)		(Con	ditio	on C	ode	e	No.	. of tes*1
	Mnemonic	Size	xx#	R	@ERn	@(d,ERn)	@-ERn/@ERn+	@aa	@(d,PC)	@ @ aa	ı	Operation	ı	н	N	z	v	С	Normal	Advanced
BAND	BAND #xx:3,Rd	В		2								C∧ (#xx:3 of Rd8)→C	<u> </u>	_	_	_	_	1		1
	BAND #xx:3,@ERd	В			4							C∧ (#xx:3 of @ERd)→C	_	_	_	_	_	‡	;	3
	BAND #xx:3,@aa:8	В						4				C∧ (#xx:3 of @aa:8)→C	_	_	_	_	_	‡	;	3
	BAND #xx:3,@aa:16	В						6				C∧ (#xx:3 of @aa:16)→C	<u> </u>	_	_	_	_	1	4	4
	BAND #xx:3,@aa:32	В						8				C∧ (#xx:3 of @aa:32)→C	<u> </u>	_	_	_	_	1		5
BIAND	BIAND #xx:3,Rd	В		2								C∧ [¬ (#xx:3 of Rd8)]→C	_	_	_	_	_	‡		1
	BIAND #xx:3,@ERd	В			4							C∧ [¬ (#xx:3 of @ERd)]→C	<u> </u>	_	_	_	_	1	:	3
	BIAND #xx:3,@aa:8	В						4				C∧ [¬ (#xx:3 of @aa:8)]→C	<u> </u>	_	_	_	_	‡	:	3
	BIAND #xx:3,@aa:16	В						6				C∧ [¬ (#xx:3 of @aa:16)]→C	<u> </u>	_	_	_	_	‡	4	4
	BIAND #xx:3,@aa:32	В						8				C∧ [¬ (#xx:3 of @aa:32)]→C	<u> </u>	_	_	_	_	‡		5
BOR	BOR #xx:3,Rd	В		2								C∨ (#xx:3 of Rd8)→C	<u> </u>	_	_	_	_	‡		1
	BOR #xx:3,@ERd	В			4							C∨ (#xx:3 of @ERd)→C	<u> </u>	_	_	_	_	‡	:	3
	BOR #xx:3,@aa:8	В						4				C∨ (#xx:3 of @aa:8)→C	_	_	_	_	_	1	;	3
	BOR #xx:3,@aa:16	В						6				C∨ (#xx:3 of @aa:16)→C	_	_	_	_	_	1	4	4
	BOR #xx:3,@aa:32	В						8				C∨ (#xx:3 of @aa:32)→C	_	_	_	_	_	1		5
BIOR	BIOR #xx:3,Rd	В		2								C∨ [¬ (#xx:3 of Rd8)]→C	_	_	_	_	_	1		1
	BIOR #xx:3,@ERd	В			4							C∨ [¬ (#xx:3 of @ERd)]→C	_	_	_	_	_	‡	;	3
	BIOR #xx:3,@aa:8	В						4				C∨ [¬ (#xx:3 of @aa:8)]→C	_	_	_	_	_	‡	;	3
	BIOR #xx:3,@aa:16	В						6				C∨ [¬ (#xx:3 of @aa:16)]→C	_	_	_	_	_	‡	4	4
	BIOR #xx:3,@aa:32	В						8				C∨ [¬ (#xx:3 of @aa:32)]→C	_	_	_	_	_	‡		5
BXOR	BXOR #xx:3,Rd	В		2								C⊕ (#xx:3 of Rd8)→C	_	_	_	_	_	‡		1
	BXOR #xx:3,@ERd	В			4							C⊕ (#xx:3 of @ERd)→C	_	_	_	_	_	‡	;	3
	BXOR #xx:3,@aa:8	В						4				C⊕ (#xx:3 of @aa:8)→C	_	_	_	_	_	‡	;	3
	BXOR #xx:3,@aa:16	В						6				C⊕ (#xx:3 of @aa:16)→C	_	_	_	_	_	‡	4	4
	BXOR #xx:3,@aa:32	В						8				C⊕ (#xx:3 of @aa:32)→C	-	_	_	_	_	1		5
BIXOR	BIXOR #xx:3,Rd	В		2								C⊕ [¬ (#xx:3 of Rd8)]→C	_	_	_	_	_	‡		1
	BIXOR #xx:3,@ERd	В			4							C⊕ [¬ (#xx:3 of @ERd)]→C	_	_	_	_	_	‡	:	3
	BIXOR #xx:3,@aa:8	В						4				C⊕ [¬ (#xx:3 of @aa:8)]→C	-	_	_	_	_	‡	;	3
	BIXOR #xx:3,@aa:16	В						6				C⊕ [¬ (#xx:3 of @aa:16)]→C	-	_	_	_	_	‡	4	4
	BIXOR #xx:3,@aa:32	В						8				C⊕ [¬ (#xx:3 of @aa:32)]→C	_	_	_	_	_	‡		5

6. Branch Instructions

			In		dre:					nd rtes)	,			(Con	ditio	on C	od	е	No. State	
	Mnemonic	Size	xx#	Rn	@ERn	@ (d,ERn)	@-ERn/@ERn+	@ aa	@ (d,PC)	@ @aa	1	Operation	Branch Condition	ı	н	N	z	v	С	Normal	Advanced
Всс	BRA d:8(BT d:8)	_							2			if condition is true then	Always	_	_	_	_	_	-	2	
	BRA d:16(BT d:16)	_							4			PC←PC+d else next;		_	_	_	_	_	-	3	
	BRN d:8(BF d:8)								2			, , , , , , , , , , , , , , , , , , , ,	Never	_	_	_	_	_	_	2	
	BRN d:16(BF d:16)	T-							4					_	_	_	_	_	-	3	
	BHI d:8	T-							2				CvZ=0	_	_	_	_	_	_	2	
	BHI d:16								4					_	_	_	_	_	_	3	
	BLS d:8	_							2				C∨Z=1	_	_	_	_	_	<u> </u>	2	
	BLS d:16	T-							4					_	_	_	_	_	-	3	
	BCC d:8(BHS d:8)	T-							2				C=0	_	_	_	_	_	_	2	
	BCC d:16(BHS d:16)	T-							4					_	_	_	_	_	_	3	
	BCS d:8(BLO d:8)	1-							2				C=1	_	_	_	_	_	_	2	
	BCS d:16(BLO d:16)	1-							4					_	_	_	_	_	_	3	
	BNE d:8	T-							2				Z=0	_	_	_	_	_	-	2	
	BNE d:16	T-							4					_	_	_	_	_	-	3	
	BEQ d:8	T-							2				Z=1	_	_	_	_	_	-	2	
	BEQ d:16	1-							4					_	_	_	_	_	_	3	
	BVC d:8	1-							2				V=0	_	_	_	_	_	_	2	
	BVC d:16	_							4					_	_	_	_	_	_	3	
	BVS d:8	1-							2				V=1	_	_	_	_	_	_	2	
	BVS d:16	1-							4					_	_	_	_	_	_	3	
	BPL d:8	1-							2				N=0	_	_	_	_	_	_	2	
	BPL d:16	_							4					_	_	_	_	_	-	3	
	BMI d:8	T-							2				N=1	_	_	_	_	_	-	2	
	BMI d:16	T-							4					_	_	_	_	_	-	3	
	BGE d:8	_							2				N⊕V=0	_	_	<u> </u>	_	_	_	2	
	BGE d:16	1-							4					_	_	<u> </u>	_	_	_	3	
	BLT d:8	1-							2				N⊕V=1	_	_	<u> </u>	_	_	_	2	
	BLT d:16	1-							4					_	_	_	_	_	_	3	
	BGT d:8	1-							2			1	Z√(N⊕V)=0	_	_	<u> </u>	_	_	<u> </u>	2	
	BGT d:16	1-							4			1		_	_	<u> </u>	_	_	<u> </u>	3	
	BLE d:8	1-							2				Z∨(N⊕V)=1	_	_	<u> </u>	_	_	_	2	
	BLE d:16	1_							4			1		_	_	_	_	_	-	3	



			Ir				ng N Len				ı		(Con	diti	on (Cod	е	No. Stat	of tes*1
	Mnemonic	Size	*xx	Rn	@ERn	@(d,ERn)	@-ERn/@ERn+	@aa	@(d,PC)	@ @ aa	ı	Operation	ı	н	N	z	v	O	Normal	Advanced
JMP	JMP @ERn	1-			2							PC←ERn	-	-	_	_	_	_	2	2
	JMP @aa:24	1-						4				PC←aa:24	-	_	_	_	_	_	3	3
	JMP @@aa:8	-								2		PC←@aa:8	-	_	_	_	_	_	4	5
BSR	BSR d:8	-							2			PC→@-SP,PC←PC+d:8	-	_	_	_	_	_	3	4
	BSR d:16	1-							4			PC→@-SP,PC←PC+d:16	_	_	_	_	_	_	4	5
JSR	JSR @ERn	_			2							PC→@-SP,PC←ERn	-	_	_	_	_	_	3	4
	JSR @aa:24	-						4				PC→@-SP,PC←aa:24	_	_	_	_	_	_	4	5
	JSR @@aa:8	-								2		PC→@-SP,PC←@aa:8	_	_	_	_	_	_	4	6
RTS	RTS	_									2	PC←@SP+	_	_	_	_	_	_	4	5

7. System Control Instructions

			Ir			ssir)		(Con	diti	on (Cod	е	No. Stat	of es*1
	Mnemonic	Size	xx#	Rn	@ERn	@(d,ERn)	@-ERn/@ERn+	@aa	@(d,PC)	@ @ aa	I	Operation	ı	н	N	z	v	С	Normal	Advanced
TRAPA	TRAPA #xx:2	-										PC→@-SP,CCR→@-SP, EXR→@-SP, <vector>→PC</vector>	1	_	-	_	_	_	7 [9]	8 [9]
RTE	RTE	-										EXR←@SP+,CCR←@SP+, PC←@SP+	\$	\$	\$	\$	‡	\$	5 [9]
SLEEP	SLEEP	-										Transition to power-down state	-	_	_	_	_	_	2	2
LDC	LDC #xx:8,CCR	В	2									#xx:8→CCR	1	‡	\$	\$	‡	‡	1	
	LDC #xx:8,EXR	В	4									#xx:8→EXR	-	_	_	_	_	_	2	?
	LDC Rs,CCR	В		2								Rs8→CCR	1	‡	\$	1	‡	1	1	
	LDC Rs,EXR	В		2								Rs8→EXR	-	_	_	_	_	_	1	
	LDC @ERs,CCR	W			4							@ERs-CCR	‡	‡	\$	\$	‡	‡	3	}
	LDC @ERs,EXR	W			4							@ERs-EXR	-	_	_	_	_	_	3	}
	LDC @(d:16,ERs),CCR	W				6						@(d:16,ERs)→CCR	‡	‡	\$	\$	‡	‡	4	ļ
	LDC @(d:16,ERs),EXR	W				6						@(d:16,ERs)→EXR	_	_	_	_	_	_	4	ļ
	LDC @(d:32,ERs),CCR	W				10						@(d:32,ERs)→CCR	\$	‡	\$	\$	‡	‡	6	;
	LDC @(d:32,ERs),EXR	W				10						@(d:32,ERs)→EXR	-	_	-	_	_	_	6	;
	LDC @ERs+,CCR	W					4					@ERs→CCR,ERs32+2→ERs32	1	‡	\$	\$	‡	‡	4	ļ
	LDC @ERs+,EXR	W					4					@ERs→EXR,ERs32+2→ERs32	-	_	-	_	_	_	4	ļ
	LDC @aa:16,CCR	W						6				@aa:16→CCR	\$	‡	\$	\$	‡	‡	4	1
	LDC @aa:16,EXR	W						6				@aa:16→EXR	-		-	_		_	4	1
	LDC @aa:32,CCR	W						8				@aa:32→CCR	\$	\$	\$	\$	‡	‡	5	;
	LDC @aa:32,EXR	W						8				@aa:32→EXR	[-	_	_	_	_	=	5	;



			Ir	Ad nstr		ssir on l)		(Con	diti	on (od	е		of tes*1
	Mnemonic	Size	xx#	Rn	@ERn	@(d,ERn)	@-ERn/@ERn+	@ aa	@(d,PC)	@ @aa	I	Operation	ı	Н	N	z	v	С	Normal	Advanced
STC	STC CCR,Rd	В		2								CCR→Rd8	-	_	_	_	_	_		1
	STC EXR,Rd	В		2								EXR→Rd8	-	_	_	_	_	_	,	1
	STC CCR,@ERd	w			4							CCR→@ERd	-	_	_	_	_	_	,	3
	STC EXR,@ERd	w			4							EXR→@ERd	-	_	_	_	_	_	,	3
	STC CCR,@(d:16,ERd)	w				6						CCR→@(d:16,ERd)	-	_	_	_	_	_	4	4
	STC EXR,@(d:16,ERd)	w				6						EXR→@(d:16,ERd)	-	_	_	_	_	_	4	4
	STC CCR,@(d:32,ERd)	w				10						CCR→@(d:32,ERd)	-	_	_	_	_	_	(6
	STC EXR,@(d:32,ERd)	w				10						EXR→@(d:32,ERd)	-	_	_	_	_	_	(6
	STC CCR,@-ERd	w					4					ERd32-2→ERd32,CCR→@ERd	-	_	_	_	_	_	4	4
	STC EXR,@-ERd	w					4					ERd32-2→ERd32,EXR→@ERd	-	_	_	_	_	_	4	4
	STC CCR,@aa:16	w						6				CCR→@aa:16	-	_	_	_	_	_	4	4
	STC EXR,@aa:16	w						6				EXR→@aa:16	-	_	_	_	_	_	4	4
	STC CCR,@aa:32	w						8				CCR→@aa:32	-	_	_	_	_	_		5
	STC EXR,@aa:32	w						8				EXR→@aa:32	-	_	_	_	_	_		5
ANDC	ANDC #xx:8,CCR	В	2									CCR∧#xx:8→CCR	1	‡	‡	‡	1	1	,	1
	ANDC #xx:8,EXR	В	4									EXR∧#xx:8→EXR	-	_	_	_	_	_	2	2
ORC	ORC #xx:8,CCR	В	2									CCR∨#xx:8→CCR	1	‡	1	‡	1	1		1
	ORC #xx:8,EXR	В	4									EXR∨#xx:8→EXR	-	_	_	_	_	_	2	2
XORC	XORC #xx:8,CCR	В	2									CCR⊕#xx:8→CCR	1	‡	1	‡	‡	1		1
	XORC #xx:8,EXR	В	4									EXR⊕#xx:8→EXR	-	_	_	_	_	_	2	2
NOP	NOP	-									2	PC←PC+2	-	_	_	_	_	_		1

8. Block Transfer Instructions

			In	Ad nstru	dre: ucti)		ď	Con	ditio	on C	Cod	е	No. Stat	of es*1
	Mnemonic	Size	xx#	Rn	@ERn	@ (d,ERn)	@-ERn/@ERn+	@ aa	@ (q,PC)	@ @aa	1	Operation	ı	н	N	z	v	С	Normal	Advanced
EEPMOV	EEPMOV.B	_									4	if R4L≠0 Repeat @ER5→@ER6 ER5+1→ER5 ER6+1→ER6 R4L-1→R4L Until R4L=0 else next;		_					4+2	n*2
	EEPMOV.W	_									4	if R4≠0 Repeat @ER5→@ER6 ER5+1→ER5 ER6+1→ER6 R4-1→R4 Until R4=0 else next;	_		_			_	4+2	n*2

- Notes: 1. The number of states is the number of states required for execution when the instruction and its operands are located in on-chip memory.
 - 2. n is the initial value set in R4L or R4.
 - 3. Only register ER0, ER1, ER4, or ER5 should be used when using the TAS instruction.
 - 4. Only registers ER0 to ER6 should be used when using the STM/LDM instruction.
 - [1] 7 states when the number of saved/restored registers is 2, 9 states when 3, and 11 states when 4.
 - [2] Cannot be used with this LSI.
 - [3] Set to 1 when there is a carry from or borrow to bit 11; otherwise cleared to 0.
 - [4] Set to 1 when there is a carry from or borrow to bit 27; otherwise cleared to 0.
 - [5] If the result is zero, the previous value of the flag is retained; otherwise the flag is cleared to 0.
 - [6] Set to 1 if the divisor is negative; otherwise cleared to 0.
 - [7] Set to 1 if the divisor is zero; otherwise cleared to 0.
 - [8] Set to 1 if the quotient is negative; otherwise cleared to 0.
 - [9] When EXR is valid, the number of states is increased by 1.



A.2 Instruction Codes

Table A.2 Instruction Codes

Instruc-	Monomork								Instruc	Instruction Format				
tion		Size	1st Byte	yte	2nd Byte	3yte	3rd Byte	4th Byte	5th Byte	6th Byte	7th Byte	8th Byte	9th Byte	10th Byte
ADD	ADD.B #xx:8,Rd	В	ω	Þ	IMM	Σ								
	ADD.B Rs,Rd	В	0	8	rs	rd								
	ADD.W #xx:16,Rd	>	7	6	-	ы	2	IMM						
	ADD.W Rs,Rd	8	0	6	ß	rd								
	ADD.L #xx:32,ERd	_	7	∢	-	0 erd		IMM	Σ					
	ADD.L ERS,ERd	_	0	∢	1 ers 0 erd	0 erd								
ADDS	ADDS #1,ERd	_	0	В	0	0 erd								
	ADDS #2,ERd	_	0	В	ω	0 erd								
	ADDS #4,ERd	_	0	В	6	0 erd								
ADDX	ADDX #xx:8,Rd	В	6	ъ	IMM	Σ								
	ADDX Rs,Rd	В	0	ш	ន	ē								
AND	AND.B #xx:8,Rd	В	ш	ъ	IMM	Σ								
	AND.B Rs,Rd	В	-	9	LS.	ъ								
	AND.W #xx:16,Rd	Ν	7	6	9	rd	N	IMM						
	AND.W Rs,Rd	>	9	9	ß	ы								
	AND.L #xx:32,ERd	_	7	٧	9	0 erd		IMM	Σ					
	AND.L ERS,ERd	_	0	-	ш	0	9 9	0 ers 0 erd						
ANDC	ANDC #xx:8,CCR	В	0	9	IMM	Σ								
	ANDC #xx:8,EXR	В	0	1	4	1	9 0	IMM						
BAND	BAND #xx:3,Rd	В	7	9	ОІММ	p								
	BAND #xx:3,@ERd	В	7	O	0 erd	0	9 /	0 MMI:0						
	BAND #xx:3,@aa:8	В	7	ш	aps	S	9 /	0 MMI 0						
	BAND #xx:3,@aa:16	В	9	∢	-	0	a	abs	2 6	0 IMM 0				
	BAND #xx:3,@aa:32	В	9	۷	3	0		abs	S		7 6	O IMM O		
Bcc	BRA d:8 (BT d:8)	I	4	0	disp	ά								
	BRA d:16 (BT d:16)	I	2	80	0	0	ਰ	disp						
	BRN d:8 (BF d:8)	I	4	-	disp	ď								
	BRN d:16 (BF d:16)	Ι	2	80	-	0	ā	disp						

Instruc-	Monoin							Instruction Format	n Format				
tion		Size	1st Byte	3yte	2nd Byte	3rd Byte	4th Byte	5th Byte	6th Byte	7th Byte	8th Byte	9th Byte	10th Byte
Bcc	BHI d:8	Ι	4	2	dsib								
	BHI d:16	I	2	æ	2 0	dsip	d						
	BLS d:8	Ι	4	3	dsib								
	BLS d:16	I	2	8	3 0	disp	d						
	BCC d:8 (BHS d:8)	I	4	4	dsib								
	BCC d:16 (BHS d:16)	I	5	8	4 0	dsip	d						
	BCS d:8 (BLO d:8)	Ι	4	2	dsib								
	BCS d:16 (BLO d:16)	I	2	8	5 0	dsip	d						
	BNE d:8	Ι	4	9	dsib								
	BNE d:16	Ι	2	8	0 9	disp	d						
	BEQ d:8	I	4	7	dsib								
	BEQ d:16	I	2	8	0 2	dsip	d						
	BVC d:8	Ι	4	8	dsib								
	BVC d:16	I	2	80	0 8	dsib	ď						
	BVS d:8	I	4	6	dsib								
	BVS d:16	I	2	8	0 6	disp	d						
	BPL d:8	I	4	А	dsib								
	BPL d:16	I	2	8	A 0	disp	ф						
	BMI d:8	Ι	4	В	dsib								
	BMI d:16	I	2	8	B 0	disp	р						
	BGE d:8	I	4	С	dsib								
	BGE d:16	Ι	2	8	0 O	disp	ф						
	BLT d:8	I	4	۵	dsib								
	BLT d:16	I	2	œ	0	disp	d						
	BGT d:8	I	4	ш	dsib								
	BGT d:16	I	2	8	Е 0	dsip	р						
	BLE d:8	I	4	Ь	dsib								
	BLE d:16	Ι	5	8	F 0	disp	d						

Instruc-	Magazic	į								드	structic	Instruction Format				
tion		Size		1st Byte	2nd Byte	yte	3rd Byte		4th Byte	5th	5th Byte	6th Byte	7th Byte	8th Byte	9th Byte	10th Byte
BCLR	BCLR #xx:3,Rd	В	7	2	ОПММ	Þ										
	BCLR #xx:3, @ ERd	В	2	۵	0 erd	0	7	2 0 11	O IMM O							
	BCLR #xx:3, @aa:8	В	2	н	abs	,	2 2	2 0.11	0 MMI:0							
	BCLR #xx:3, @aa:16	В	9	A	-	8		abs		7	5	0 IMM 0				
	BCLR #xx:3, @aa:32	В	9	٨	က	80				abs			7 2	0 MMI 0		
	BCLR Rn,Rd	В	9	2	E	p										
	BCLR Rn, @ERd	В	7	۵	0 erd	0	9	2	0							
	BCLR Rn, @aa:8	В	7	ш	aps	,,	9	2 r	0							
	BCLR Rn, @aa:16	Ф	9	4	·····	∞		abs		9		0				
	BCLR Rn, @aa:32	В	9	۷	က	80				aps			6	0		
BIAND	BIAND #xx:3,Rd	Ф	7	9	1 IMM	5										
	BIAND #xx:3, @ERd	Ф	7	ပ	0 erd	0	2	9	1 IMM 0							
	BIAND #xx:3, @aa:8	В	2	Ш	abs	,) /	9	0 MMI 1							
	BIAND #xx:3, @aa:16	В	9	A	Ψ-	0		abs		7	9	1 IMM 0				
	BIAND #xx:3, @aa:32	В	9	A	က	0				abs			9 /	1 IMM 0		
BILD	BILD #xx:3,Rd	В	2	7	1 IMM	rd										
	BILD #xx:3,@ERd	В	2	၁	0 erd	0	7 7	7 1	1 IMM 0							
	BILD #xx:3,@aa:8	В	2	Е	abs	·^	7 7	7 1	1 IMM 0							
	BILD #xx:3,@aa:16	В	9	Α	-	0		abs		7	7	1 IMM 0				
	BILD #xx:3,@aa:32	В	9	A	3	0				abs			2 2	1 IMM 0		
BIOR	BIOR #xx:3,Rd	В	2	4	1 IMM	Þ										
	BIOR #xx:3,@ERd	В	7	C	0 erd	0	, ,	4 1:11	1 IMM 0							
	BIOR #xx:3,@aa:8	В	7	Ш	abs	.	, ,	4	1 IMM 0							
	BIOR #xx:3,@aa:16	В	9	Α	٠٠٠٠٠	0		abs		7	4	1 IMM 0				
	BIOR #xx:3,@aa:32	В	9	Α	3	0				abs			7 4	1 IMM 0		

Instruc-	Mnemonic	ا ا							Instruction	Instruction Format				
tion		Size	1st Byte	yte	2nd Byte	3rd Byte	3yte	4th Byte	5th Byte	6th Byte	7th Byte	8th Byte	9th Byte	10th Byte
BIST	BIST #xx:3,Rd	В	9	7	1 IMM rd									
	BIST #xx:3, @ERd	В	7	۵	0 erd 0	9	7	1 IMM 0						
	BIST #xx:3, @aa:8	В	7	н	abs	9	7 1	1 IMM 0						
	BIST #xx:3, @aa:16	В	9	٧	1 8		abs		2 9	1 IMM 0				
	BIST #xx:3, @aa:32	В	9	4	3			at	abs		2 9	1 IMM 0		
BIXOR	BIXOR #xx:3,Rd	В	7	5	1 IMM rd									
	BIXOR #xx:3,@ERd	В	7	ပ	0 erd 0	7	5	1 IMM 0						
	BIXOR #xx:3,@aa:8	В	7	ш	abs	7	5	1 IMM 0						
	BIXOR #xx:3,@aa:16	В	9	А	1 0		abs	\$	2 2	1 IMM 0				
	BIXOR #xx:3,@aa:32	В	9	A	3 0			at	abs		7 5	1 IMM 0		
BLD	BLD #xx:3,Rd	В	7	7	0 IMM rd									
	BLD #xx:3,@ERd	В	7	ပ	0 erd 0	7	7 (0 MMI 0						
	BLD #xx:3,@aa:8	В	7	ш	abs	7	7 (0 MMI 0						
	BLD #xx:3,@aa:16	В	9	٧	1 0		abs		7 7	0 MMI 0				
	BLD #xx:3,@aa:32	В	9	Α	3 0			at	abs		7 7	0 MMI 0		
BNOT	BNOT #xx:3,Rd	В	7	1	0 IMM rd									
	BNOT #xx:3,@ERd	В	7	D	0 erd 0	7	1	0 MMI 0						
	BNOT #xx:3,@aa:8	В	7	ш	aps	7	-	O IMMI O						
	BNOT #xx:3,@aa:16	В	9	A	1 8		abs		7 1	0 IMM 0				
	BNOT #xx:3,@aa:32	В	9	٧	3			at	abs		7 1	0 IMM 0		
	BNOT Rn,Rd	В	9	1	m rd									
	BNOT Rn,@ERd	В	7	٥	0 erd 0	9	-	n 0						
	BNOT Rn,@aa:8	В	7	ш	abs	9	-	n 0						
	BNOT Rn,@aa:16	В	9	۷	- 8		aps		9	0				
	BNOT Rn,@aa:32	В	9	4	3			aţ	abs		9	n 0		



Instruc-	e i como M										Instr	ruction	Instruction Format				
tion		Size	1st	1st Byte	2nd Byte	3yte	3rd Byte	yte	4th Byte	te	5th Byte	yte	6th Byte	7th Byte	8th Byte	9th Byte	10th Byte
BOR	BOR #xx:3,Rd	В	7	4	0 IMM	Б											
	BOR #xx:3,@ERd	В	7	С	0 erd	0	7	4 (ммі о	0							
	BOR #xx:3, @aa:8	В	7	Е	abs	s	7	4	оімм	0							
	BOR #xx:3,@aa:16	В	9	٧	-	0		abs	<u></u>		7	4	0 MMI 0				
	BOR #xx:3, @aa:32	В	9	4	က	0				aps				7 4	0 IMM 0		
BSET	BSET #xx:3,Rd	В	7	0	0 IMM	rd											
	BSET #xx:3,@ERd	В	7	D	0 erd	0	7	0	о імм	0							
	BSET #xx:3,@aa:8	В	7	F	abs	s	7	0	оімм	0							
	BSET #xx:3,@aa:16	В	9	Α	1	8		abs	"		7	0 0	0 MMI 0				
	BSET #xx:3,@aa:32	В	9	A	3	8				abs				7 0	0 MMI 0		
	BSET Rn,Rd	В	9	0	٤	p											
	BSET Rn, @ERd	Ф	7	۵	0 erd	0	9	0	 E	0							
	BSET Rn, @aa:8	В	7	ш	abs	s	9	0	 E	0							
	BSET Rn,@aa:16	В	9	А	1	8		abs	(x)		9	0	0 m				
	BSET Rn, @aa:32	В	9	А	3	8				abs				0 9	rn 0		
BSR	BSR d:8	Ι	2	5	disp	Q.											
	BSR d:16	Ι	2	ပ	0	0		disp	0								
BST	BST #xx:3,Rd	В	9	7	0 IMM	rd											
	BST #xx:3,@ERd	В	7	۵	0 erd	0	9	7	оімм	0							
	BST #xx:3,@aa:8	В	7	ч	abs	s	9	7 (ОІММ	0							
	BST #xx:3,@aa:16	В	9	Α	1	8		abs			9	7 0	O IMM O				
	BST #xx:3,@aa:32	В	9	Α	3	8				abs				6 7	0 IMM 0		
BTST	BTST #xx:3,Rd	В	7	3	0 IMM	Þ											
	BTST #xx:3,@ERd	Ф	7	ပ	0 erd	0	7	3	O IMM	0							
	BTST #xx:3,@aa:8	Ф	7	ш	abs	s	7	3	ОІММ	0							
	BTST #xx:3,@aa:16	Ф	9	4	~	0		abs	,		7	3	0 IMM 0				
	BTST #xx:3,@aa:32	Ф	9	٧	က	0				aps				7 3	0 MMI 0		
	BTST Rn,Rd	В	9	ъ	٤	Þ											
	BTST Rn,@ERd	В	7	ပ	0 erd	0	9	3	 E	0							

Instruc-	Momorio	į								Instructi	Instruction Format				
tion		Size	1st	1st Byte	2nd Byte		3rd Byte		4th Byte	5th Byte	6th Byte	7th Byte	8th Byte	9th Byte	10th Byte
BTST	BTST Rn, @aa:8	В	7	Е	abs		9	LL L	0						
	BTST Rn,@aa:16	В	9	A	1	0		abs		6 3	0 W				
	BTST Rn, @aa:32	В	9	A	3	0			abs	Sı		9	rn 0		
BXOR	BXOR #xx:3,Rd	В	7	2	O IMM	p									
	BXOR #xx:3,@ERd	В	7	С	0 erd (0	7 5	O IMM	1M 0						
	BXOR #xx:3,@aa:8	В	7	Е	abs		7 5	MMI 0	1M 0						
	BXOR #xx:3,@aa:16	В	9	A	-	0		abs		7 5	0 IMM 0				
	BXOR #xx:3,@aa:32	В	9	∢	e e	0			abs	Š		7 5	0 MMI 0		
CLRMAC	CLRMAC CLRMAC	I	Cann	ot be us	Cannot be used with this LSI	nis LSI									
CMP	CMP.B #xx:8,Rd	В	⋖	ъ	IMM										
	CMP.B Rs,Rd	В	1	С	rs r	rd									
	CMP.W #xx:16,Rd	W	7	6	2	rd		MMI							
	CMP.W Rs,Rd	8	<u>_</u>	۵	rs L	p									
	CMP.L #xx:32,ERd	_	7	4	2	0 erd			IMM	Σ					
	CMP.L ERS,ERd	L	1	Ь	1 ers 0	0 erd									
DAA	DAA Rd	В	0	ч	0	rd									
DAS	DAS Rd	В	1	F	0	rd									
DEC	DEC.B Rd	В	-	۷	0	Ð									
	DEC.W #1,Rd	≥	-	В		Đ.									
	DEC.W #2,Rd	8	1	В	٥	rd									
	DEC.L #1,ERd	_	-	В	2 0	0 erd									
	DEC.L #2,ERd	٦	_	В	F 0 erd	erd									
DIVXS	DIVXS.B Rs,Rd	В	0	1	0	0	5 1	SI IS	rd rd						
	DIVXS.W Rs,ERd	W	0	1	0	0	5 3	SI IS	0 erd						
DIVXU	DIVXU.B Rs,Rd	В	2	1	rs.	p									
	DIVXU.W Rs,ERd	8	2	3	rs 0	0 erd									
EEPMOV	EEPMOV EEPMOV.B	I	7	В		O	2	80	ш.						
	EEPMOV.W	Ι	7	В	, 	4	5 9	8	ш.						



Instruc-											Instruction Format	n Forn	nat				
tion		Size	1st E	1st Byte	2nd	2nd Byte	3rd Byte	vte	4th Byte		5th Byte	6th Byte	3yte	7th Byte	8th Byte	9th Byte	10th Byte
EXTS	EXTS.W Rd	8	_	7	۵	p											
	EXTS.L ERd	Г	1	7	ч	0 erd											
EXTU	EXTU.W Rd	Μ	1	7	2	rd											
	EXTU.L ERd	_	-	7	7	0 erd											
INC	INC.B Rd	В	0	A	0	Þ											
	INC.W #1,Rd	8	0	В	2	Ð											
	INC.W #2,Rd	Ν	0	В	۵	rd											
	INC.L #1,ERd	Γ	0	В	7	0 erd											
	INC.L #2,ERd	٦	0	В	F	0 erd											
JMP	JMP @ERn	Ι	2	6	0 ern	0											
	JMP @aa:24	Ι	2	∢			aps	,									
	JMP @@aa:8	Ι	2	В	al	abs											
JSR	JSR @ERn	I	2	۵	0 ern	0											
	JSR @aa:24	I	2	ш			abs	,									
	JSR @@aa:8	I	5	Ь	al	abs											
ГРС	LDC #xx:8,CCR	В	0	7	IN	IMM											
	LDC #xx:8,EXR	В	0	1	4	1	0	7	IMM								
	LDC Rs,CCR	В	0	3	0	S											
	LDC Rs,EXR	В	0	3	-	ত											
	LDC @ERs,CCR	>	0	-	4	0	9	6	ers 0								
	LDC @ERs,EXR	≥	0	-	4	-	9	0	0 ers 0								
	LDC @(d:16,ERs),CCR	≥	0	-	4	0	9	ь	0 ers 0		ē	disp					
	LDC @(d:16,ERs),EXR	≥	0	-	4	-	9	ь	0 ers 0		ਰ	disp					
	LDC @(d:32,ERs),CCR	≥	0	-	4	0	7	8	0 ers 0	9	_	7	0		ġ	dsib	
	LDC @(d:32,ERs),EXR	≥	0	-	4	-	7	8	0 ers 0	9	_	2	0		ġ	disp	
	LDC @ERs+,CCR	≥	0	-	4	0	9	٥	0 ers 0								
	LDC @ERs+,EXR	≥	0	-	4	-	9	٥	0 ers 0								
	LDC @aa:16,CCR	≥	0	-	4	0	9	Ф	0 0		ğ	abs					
	LDC @aa:16,EXR	≥	0	-	4	-	9	В	0	\dashv	a	abs					

Instruc-	o: company										Instruction Format	n Format				
tion		Size	1st Byte	3yte	2nd Byte	Byte	3rd Byte	3yte	4th	4th Byte	5th Byte	6th Byte	7th Byte	8th Byte	9th Byte	10th Byte
ГРС	LDC @aa:32,CCR	>	0	-	4	0	9	В	2	0		abs	Si			
	LDC @aa:32,EXR	>	0	-	4	-	9	В	2	0		abs	Si			
LDM*3	LDM.L @SP+, (ERn-ERn+1)	٦	0	1	1	0	9	D	7	0 em+1						
	LDM.L @SP+, (ERn-ERn+2)	٦	0	-	2	0	9	۵	7	0 em+2						
	LDM.L @SP+, (ERn-ERn+3)	_	0	-	3	0	9	D	7	0 ern+3						
LDMAC	LDMAC ERS,MACH	٦	Cann	ot be u	Cannot be used with this LSI	h this L	SI.									
	LDMAC ERS,MACL	٦														
MAC	MAC @ERn+, @ERm+	1														
MOV	MOV.B #xx:8,Rd	В	ъ	Þ	IMM	Σ										
	MOV.B Rs,Rd	В	0	ပ	LS	rd										
	MOV.B @ERs,Rd	В	9	8	0 ers	rd										
	MOV.B @ (d:16,ERs),Rd	В	9	Е	0 ers	rd		disp	dξ							
	MOV.B @(d:32,ERs),Rd	В	7	8	0 ers	0	9	4	2	5		dsip	ds			
	MOV.B @ERs+,Rd	В	9	C	0 ers	rd										
	MOV.B @aa:8,Rd	В	2	Б	abs	Sc										
	MOV.B @aa:16,Rd	В	9	⋖	0	Þ		abs	S							
	MOV.B @aa:32,Rd	В	9	Α	2	rd				abs	s					
	MOV.B Rs, @ERd	В	9	8	1 erd	IS										
	MOV.B Rs, @ (d:16,ERd)	В	9	ш	1 erd	গ		disp	ď							
	MOV.B Rs, @ (d:32,ERd)	В	7	8	0 erd	0	9	Α	Α	rs		disp	dε			
	MOV.B Rs, @-ERd	В	9	O	1 erd	ত										
	MOV.B Rs, @aa:8	В	က	S	aps	SC										
	MOV.B Rs, @aa:16	В	9	Α	8	rs		abs	S							
	MOV.B Rs, @aa:32	В	9	⋖	⋖	হ				aps	s					
	MOV.W #xx:16,Rd	≥	7	6	0	Б		IMM	Σ							
	MOV.W Rs,Rd	8	0	D	rs	rd										
	MOV.W @ERs,Rd	8	9	6	0 ers	Б										
	MOV.W @(d:16,ERs),Rd	>	9	ш	0 ers	ъ		disp	ď							
	MOV.W @(d:32,ERs),Rd	≥	7	8	0 ers	0	9	В	2	2		disp	ds			



Instriic.										lns	structio	Instruction Format	يو				
tion	Minemonic	Size	1st Byte	3yte	2nd Byte	3yte	3rd Byte	fe	4th Byte	5th	5th Byte	6th Byte		7th Byte	8th Byte	9th Byte	10th Byte
MOV	MOV.W @ERs+,Rd	8	9	٥	0 ers	rd											
	MOV.W @aa:16,Rd	≥	9	В	0	ē		abs									
	MOV.W @aa:32,Rd	≥	9	В	2	5				abs							
	MOV.W Rs, @ERd	8	9	6	1 erd	ย											
	MOV.W Rs, @(d:16,ERd)	8	9	ш	1 erd	গ		disp									
	MOV.W Rs, @(d:32,ERd)	×	7	8	0 erd	0	9	В	A rs				dsib				
	MOV.W Rs, @-ERd	N	9	O	1 erd	ญ											
	MOV.W Rs,@aa:16	Ν	9	В	8	S		abs									
	MOV.W Rs, @aa:32	8	9	В	∢	S				abs							
	MOV.L #xx:32,Rd	٦	7	A	0	0 erd			-	IMM							
	MOV.L ERS,ERd	٦	0	ш	1 ers	0 erd											
	MOV.L @ERS,ERd	_	0	1	0	0	9	0 6	0 ers 0 erd								
	MOV.L @ (d:16,ERs),ERd	٦	0	1	0	0	9	Ь 0	0 ers 0 erd		dsib	Q.					
	MOV.L @ (d:32,ERs),ERd	٦	0	-	0	0	7	8	0 ers 0	9	В	2 0	erd		ij	dsip	
	MOV.L @ERs+,ERd	٦	0	1	0	0	9	D 0	ers 0 erd								
	MOV.L @aa:16 ,ERd	٦	0	-	0	0	9	В	0 0 erd	_	abs	Ş					
	MOV.L @aa:32 ,ERd	٦	0	1	0	0	9	В	2 0 erd				abs				
	MOV.L ERs, @ERd	Γ	0	1	0	0	9	9 1	1 erd 0 ers								
	MOV.L ERs, @ (d:16,ERd)	٦	0	-	0	0	9	Т_	1 erd 0 ers		disp	g.					
	MOV.L ERs, @(d:32, ERd)*1	_	0	-	0	0	7	8	0 erd 0	9	Ф	⋖	ers		ij	disp	
	MOV.L ERs, @-ERd	_	0	-	0	0	9	7	1 erd 0 ers	**							
	MOV.L ERs, @aa:16	٦	0	-	0	0	9	В	8 0 ers		aps	S					
	MOV.L ERs, @aa:32	٦	0	1	0	0	9	В	A 0 ers				abs				
MOVFPE	MOVFPE MOVFPE @aa:16,Rd	В	Canno	t be us	ed with	Cannot be used with this LSI											
MOVTPE	MOVTPE MOVTPE Rs, @aa:16	В															
MULXS	MULXS.B Rs,Rd	В	0	-	ပ	0	2	0	rs Ld								
	MULXS.W Rs,ERd	≥	0	-	ပ	0	2	2	rs 0 erd								
MULXU	MULXU.B Rs,Rd	В	2	0	బ	5											
	MULXU.W Rs,ERd	≥	2	2	S	0 erd											

Instruc-	Magazic	į							Instruction Format	n Format				
tion		Size	1st Byte	yte	2nd Byte		3rd Byte	4th Byte	5th Byte	6th Byte	7th Byte	8th Byte	9th Byte	10th Byte
NEG	NEG.B Rd	В	τ-	7	9 8									
	NEG.W Rd	≥	τ-	7	p. 6									
	NEG.L ERd	_	τ-	7	B 0 erd	Ld Ld								
NOP	NOP	ı	0	0	0									
NOT	NOT.B Rd	В	τ-	7	0									
	NOT.W Rd	>	τ-	7	1 rd									
	NOT.L ERd	_	τ-	7	3 0 erd	D.								
OR	OR.B #xx:8,Rd	В	ပ	ē	MM									
	OR.B Rs,Rd	В	τ-	4	rs									
	OR.W #xx:16,Rd	>	7	6	4 rd	_	IMM	M						
	OR.W Rs,Rd	≥	9	4	rs									
	OR.L #xx:32,ERd	_	7	A	4 0 erd			MMI	W					
	OR.L ERS,ERd	_	0	1	F 0		6 4	0 ers 0 erd						
ORC	ORC #xx:8,CCR	В	0	4	IMM									
	ORC #xx:8,EXR	В	0	1	4 1)	0 4	IMM						
POP	POP.W Rn	>	9	۵	7 rn									
	POP.L ERn	_	0	1	0 0		6 D	7 0 ern						
PUSH	PUSH.W Rn	≥	9	۵	 Б									
	PUSH.L ERn	_	0	1	0 0		6 D	F 0 ern						
ROTL	ROTL.B Rd	В	τ-	2	8 rd									
	ROTL.B #2, Rd	В	~	7	C D									
	ROTL.W Rd	≥	~	7	9 rd									
	ROTL.W #2, Rd	≥	τ-	2	D									
	ROTL.L ERd	_	1	2	B 0 erd	Ld Ld								
	ROTL.L #2, ERd	_	~	2	F 0 erd	ē								



Instruc-	Moment	ä							Instruction Format	n Format				
tion		Size	1st E	Byte	2nd Byte	3yte	3rd Byte	4th Byte	5th Byte	6th Byte	7th Byte	8th Byte	9th Byte	10th Byte
ROTR	ROTR.B Rd	В	1	3	8	rd								
	ROTR.B #2, Rd	В	1	3	၁	rd								
	ROTR.W Rd	>	1	3	6	p								
	ROTR.W #2, Rd	Ν	1	3	Ω	rd								
	ROTR.L ERd	_	-	က	М	0 erd								
	ROTR.L #2, ERd	_	-	က	ш	0 erd								
ROTXL	ROTXL.B Rd	В	1	2	0	ъ								
	ROTXL.B #2, Rd	ш	-	2	4	5								
	ROTXL.W Rd	≥	<u></u>	2	τ-	5								
	ROTXL.W #2, Rd	Ν	1	2	2	rd								
	ROTXL.L ERd	_	-	2	ю	0 erd								
	ROTXL.L #2, ERd	_	1	2	7	0 erd								
ROTXR	ROTXR.B Rd	В	-	3	0	Þ								
	ROTXR.B #2, Rd	В	-	3	4	Þ								
	ROTXR.W Rd	>	-	3	-	p								
	ROTXR.W #2, Rd	Ν	1	3	2	rd								
	ROTXR.L ERd	_	-	3	ю	0 erd								
	ROTXR.L #2, ERd	٦	1	3	7	0 erd								
RTE	RTE	Ι	2	9	7	0								
RTS	RTS	Ι	2	4	7	0								
SHAL	SHAL.B Rd	В	1	0	8	Þ								
	SHAL.B #2, Rd	В	1	0	ပ	Þ								
	SHAL.W Rd	>	-	0	6	Б								
	SHAL.W #2, Rd	>	1	0	Δ	Þ								
	SHAL.L ERd	_	-	0	В	0 erd								
	SHAL.L #2, ERd	_	-	0	ш	0 erd								

Instruc-	Monomora									-	Instruction Format	n Form	at				
tion		Size	1st Byte	3yte	2nd Byte	Byte	3rd Byte		4th Byte		5th Byte	6th Byte	yte	7th Byte	8th Byte	9th Byte	10th Byte
SHAR	SHAR.B Rd	В	1	1	8	rd											
	SHAR.B #2, Rd	Ф	-	-	ပ	5											
	SHAR.W Rd	Λ	1	1	6	rd											
	SHAR.W #2, Rd	8	-	-	۵	Þ											
	SHAR.L ERd	٦	-	1	В	0 erd											
	SHAR.L #2, ERd	٦	1	1	ь	0 erd											
SHLL	SHLL.B Rd	В	-	0	0	p											
	SHLL.B #2, Rd	Ф	-	0	4	5											
	SHLL.W Rd	>	-	0	-	ē											
	SHLL.W #2, Rd	>	-	0	2	Þ											
	SHLL.L ERd	_	-	0	3	0 erd											
	SHLL.L #2, ERd	٦	1	0	7	0 erd											
SHLR	SHLR.B Rd	В	-	-	0	Þ											
	SHLR.B #2, Rd	В	-	-	4	5											
	SHLR.W Rd	Μ	1	1	1	rd											
	SHLR.W #2, Rd	>	-	-	2	Þ											
	SHLR.L ERd	_	-	-	3	0 erd											
	SHLR.L #2, ERd	٦	1	1	7	0 erd											
SLEEP	SLEEP	Ι	0	1	8	0											
STC	STC.B CCR,Rd	В	0	2	0	rd											
	STC.B EXR,Rd	В	0	2	-	Б											
	STC.W CCR,@ERd	>	0	-	4	0	9	9	erd 0								
	STC.W EXR,@ERd	Ν	0	1	4	1	9	9 1	erd 0								
	STC.W CCR,@(d:16,ERd)	>	0	1	4	0	9	Р 1	erd 0		di	disp					
	STC.W EXR,@(d:16,ERd)	>	0	-	4	-	9	Т 1	1 erd 0		disp	ds					
	STC.W CCR,@(d:32,ERd)	>	0	-	4	0	7	8	erd 0	9	В	∢	0		Ġ	disp	
	STC.W EXR,@(d:32,ERd)	>	0	1	4	-	7	8 0	erd 0	9	В	A	0		di	disp	
	STC.W CCR,@-ERd	>	0	1	4	0	9	1	1 erd 0								
	STC.W EXR,@-ERd	>	0	1	4	-	9	٦	1 erd 0								



Instruc-	Magazia										Instruction	Instruction Format				
tion		Size	1st Byte	3yte	2nd Byte	Byte	3rd Byte	yte	4th Byte	yte	5th Byte	6th Byte	7th Byte	8th Byte	9th Byte	10th Byte
STC	STC.W CCR,@aa:16	>	0	-	4	0	9	В	8	0	ש	abs				
	STC.W EXR,@aa:16	≥	0	-	4	-	9	Ф	ω	0	a	abs				
	STC.W CCR,@aa:32	8	0	-	4	0	9	В	∢	0		abs	S			
	STC.W EXR,@aa:32	>	0	-	4	-	9	В	∢	0		abs	Si			
STM*3	STM.L (ERn-ERn+1), @-SP	_	0	-	-	0	9	۵	ш	0 ern						
	STM.L (ERn-ERn+2), @-SP	_	0	-	2	0	9	٥	ъ.	0 ern						
	STM.L (ERn-ERn+3), @-SP	٦	0	1	3	0	9	۵	ш	0 ern						
STMAC	STMAC MACH,ERd	_	Canno	ot be us	sed with	Cannot be used with this LSI	SI.									
	STMAC MACL, ERd	_														
SUB	SUB.B Rs,Rd	В	-	8	SJ	5										
	SUB.W #xx:16,Rd	8	7	6	3	Б		IMM								
	SUB.W Rs,Rd	8	-	6	SJ	5										
	SUB.L #xx:32,ERd	٦	7	А	ε	0 erd				IMM	1					
	SUB.L ERS,ERd	_	-	A	1 ers	0 erd										
SUBS	SUBS #1,ERd	٦	-	В	0	0 erd										
	SUBS #2,ERd	٦	1	В	8	0 erd										
	SUBS #4,ERd	_	-	В	6	0 erd										
SUBX	SUBX #xx:8,Rd	В	В	гd	IMM	Σ										
	SUBX Rs,Rd	В	1	ш	LS	5										
TAS	TAS @ERd*2	В	0	-	Ш	0	7	В	0 erd	ပ						
TRAPA	TRAPA #x:2	I	2	7	00 IMM	0										
XOR	XOR.B #xx:8,Rd	В	D	rd	MMI	Σ										
	XOR.B Rs,Rd	В	-	2	rs	5										
	XOR.W #xx:16,Rd	8	7	6	2	Б		IMM	V							
	XOR.W Rs,Rd	≷	9	2	ន	5										
	XOR.L #xx:32,ERd	٦	7	A	2	0 erd				IMM	_					
	XOR.L ERS,ERd	_	0	-	ш	0	9	2	0 ers 0 erd	0 erd						

Instruc-								Instruction Format	n Format				
tion	Mnemonic	Size	1st byte	oyte	2nd byte	3rd byte	4th byte	5th byte	6th byte	7th byte	8th byte	9th byte	10th byte
XORC	XORC #xx:8,CCR	В	0	5	IMM								
	XORC #xx:8,EXR	В	0	1	1	0 5	IMM						
Legend: IMM: abs: disp: rs, rd, rn: ers, erd, 6	ern, erm:	data ddres ent (8 sld (4 sld (3 sld ((2, 3, 8 s (8, 1 , 16, o bits, ir bits, in	3, 16, c 6, 24, r 32 bi ndicatir ndicatir xively.	or 32 bits) or 32 bits) its) ng an 8-bit ng an addre	Immediate data (2, 3, 8, 16, or 32 bits) Absolute address (8, 16, 24, or 32 bits) Displacement (8, 16, or 32 bits) Register field (4 bits, indicating an 8-bit or 16-bit register. rs, rd, and rn correspond to operand formats Rs, Rd, and Rn, respectively.) Register field (3 bits, indicating an address register or 32-bit register. ers, erd, ern, and erm correspond to operand formats ERs, ERd, ERn, and ERm, respectively.)	ster. rs, rd, ɛ rr 32-bit regi:	and rn corre ster. ers, en	spond to op d, ern, and t	verand form: erm corresp	ats Rs, Rd, ond to oper	and Rn, resp and formats	oectively.) ERs, ERd,
Notes: 1. 2. 3.	. Bit 7 of the 4th byte of the MOV.L ERs, @ (d:32, ERd) instruction can be either 0 or 1. Only register ER0, ER1, ER4, or ER5 should be used when using the TAS instruction. Only registers ER0 to ER6 should be used when using the STM/LDM instruction.	of the ER1, E to ER	MOV ER4, o 6 shou	LER: TER5 Ild be	s, @ (d:32, should be used when	ERd) instruc used when u using the ST	tion can be sing the TA\$	either 0 or 1 S instructior ruction.	_: <i>c</i> :				
The corre	The correspondence between register fields and general registers is shown in the following table.	registe	er field	ls and	general reç	jisters is sho	wn in the fol	lowing table	ø				
Addr 32-B	Address Registers 32-Bit Registers		7	6-Bit F	16-Bit Register		8-Bit F	8-Bit Register					
Register Field	General Register		Register Field	.e.	General Register		Register Field	General Register					
000	ERO		0000		RO		0000	ROH					
•	ER1	_			잗 •		• 0001	R1H •					
•	•		•		•			•					
•	•		•		•		•	•					
111	ER7	J	0111		R7		0111	R7H					
			1000		Щ 1		1000	ROL P.1					
			3 •				<u>.</u>	•					
			•		•		•	•					
			•		•		•	•					
		`	1111		E7		1111	R7L					

A.3 Operation Code Map

Table A.3 shows the operation code map.

Table A.3 Operation Code Map (1)

Operan		Juc	1416	·P (<u> </u>													
	ш	Table A.3 (2)	Table A.3 (2)			BLE												
	ш	ADDX	SUBX			BGT	JSR		1.3 (3)									
	۵	>	۵			BLT		MOV	Table A.3 (3)									
	O	MOV	CMP			BGE	BSR											
1 is 0.	Ф	Table A.3 (2)	Table A.3 (2)			BMI			EEPMOV									
nt bit of BH nt bit of BH	⋖	Table A.3 (2)	Table A.3 (2)			BPL	JMP	Table A.3 (2)	Table A.3 (2)									
 Instruction when most significant bit of BH is 0. Instruction when most significant bit of BH is 1. 	თ	۵	a			BVS		>	Table A.3 (2)									
when mo	®	ADD	SUB	١	n.	BVC	Table A.3 (2)	MOV	MOV	Q	×	۵	×	~	<u>~</u>	۵	>	
Instructior	7	LDC	Table A.3 (2)	3	MOV.B	BEQ	TRAPA	BST	BLD BILD	ADD	ADDX	CMP	SUBX	OR	XOR	AND	MOV	
\	9	ANDC	AND			BNE	RTE	AND	3AND BIAND									
	2	XORC	XOR			BCS	BSR	XOR	3XOR BIXOR									
yte BL	4	ORC	OR			BCC	RTS	OR	BOR BIOR									
2nd byte BH BL	ო	DC LDMAC	Table A.3 (2)			BLS	DIVXU	i i	<u> </u>									
	2	STC *	Table A.3 (2)			BH	MULXU	0	3 7									
1st byte AH AL	-	Table A.3 (2)	Table A.3 (2)			BRN	DIVXU	i d	D N N									
Instruction code:	0	NOP	Table A.3 (2)			BRA	MULXU	l c	E S S E S									
Instructik	4/ H	0	-	2	ю	4	2	9	7	8	6	4	В	O	۵	ш	ш	

Table A.3 **Operation Code Map (2)**

_																
ш	Table A.3 (3)		INC		SHAL	SHAR	ROTL	ROTR	EXTS		DEC		BLE			
ш	TAS												BGT			
۵	Table A.3 (3)		NC						EXTS		DEC		BLT			
O	Table A.3 (3)	ADD		MOV	SHAL	SHAR	ROTL	ROTR		SUB		CMP	BGE	MOVTPE*		
В		AC		M					NEG	ร		S	BMI			
∢	CLRMAC*												BPL	MOV		
6			ADDS		SHAL	SHAR	ROTL	ROTR	NEG		SUBS		BVS			
8	SLEEP		AD			HS HS	RO	RO	쀨		ns		BVC	MOV		
7			INC		SHLL	SHLR	ROTXL	ROTXR	EXTU		DEC		BEQ			
9	MAC												BNE		AND	AND
2			INC						EXTU		DEC		BCS		XOR	XOR
4	LDC				SHLL	SHLR	ROTXL	ROTXR					BCC	MOVFPE	OR	OR
က	STM								NOT				BLS	Table A.3 (4)	SUB	SUB
7	$\left[\left[\right] \right]$												BHI	MOV	CMP	CMP
1	LDM				SHLL	SHLR	ROTXL	ROTXR	NOT				BRN	Table A.3 (4)	ADD	ADD
0	MOV	INC	ADDS	DAA	RS	꿄	RO.	RO.	ž	DEC	SUBS	DAS	BRA	MOV	MOV	MOV
AH AL	01	0A	0B	0F	10	11	12	13	17	1A	18	1F	58	6A	79	7.A

Note: * Cannot be used with this LSI.

1st byte Instruction code:

2nd byte 뮵

ВН

 F

ΑH

Table A.3 Operation Code Map (3)

au	,,,,	-00	10 10	тар	(3)										
Instruction when most significant bit of DH is 0.	Instruction when most significant bit of DH is 1.		ш												
ificant bit	ificant bit		ш												
most sign	most sign		٥												
tion when	tion when		O												
— Instruc	— Instruc		В												
K	<u> </u>		٨												
Į,			6												
			8												
			7					BLD	BST BIST			BLD	BST BIST		
4th byte	DL		9			AND		BAND BIAND				3AND BIAND			
4th	DH		5			XOR		3XOR BIXOR				BXOR BIXOR			
3rd byte	H CL		4			OR		BOR				BOR BIOR			
	BL CH		ဗ		DIVXS		BTST	втѕт			BTST	BTST			
2nd byte	ВН		2	MULXS					BCLR	BCLR			BCLR	BCLR	ion field
1st byte	AL		-		DIVXS				BNOT	BNOT			BNOT	BNOT	conficers
1st	АН		0	MULXS					BSET	BSET			BSET	BSET	o register
Instruction code:			CL AH AL BH BL CH	01C05	01D05	01F06	7Cr06*1	7Cr07*1	7Dr06*1	7Dr07*1	7Eaa6*2	7Eaa7*2	7Faa6*2	7Faa7*2	Notes: 1 r is the register specification field
_															_

Notes: 1. r is the register specification field.
2. aa is the absolute address specification.

Table A.3 Operation Code Map (4)

`	pe		Coa	C 1V1	ар	(•)	
		Instruction when most significant bit of FH is 0. Instruction when most significant bit of FH is 1.	ь				
		ificant bit o	Е				
		most sign most sign	O				
		tion when tion when	၁				
		— Instruct — Instruct	В				
			A				
6th byte	귙		6				
ett	Æ		8				
5th byte	립				\ <u>Q</u>)TS	
5th	픕		7		BLD ALD	BST	
byte	Ы		9		BAND BIAN		
4th byte	H		5		R BXOR BIOR BIXOR		
3rd byte	ر ا		4		BOR BXOR BIXC		
35	CH			Ę	<u> M</u>		
oyte	В			ŀ	<u>-</u>		
2nd byte	BH		2				, L
1st byte	AH AL		1			TONG	
1st	AH		0			POET	
Instruction code:			EL AHALBHBLCHCLDHDLEH	6A10aaaa6*	6A10aaaa7*	6A18aaaa6*	6A18aaaa7*

		o L	ш				
		Indicates case where MSB of HH is 0. Indicates case where MSB of HH is 1.	ш				
0	_	iere MSB iere MSB	۵				
8th byte	HH	s case wh s case wh	ပ				
7th byte	GL	Indicates Indicates	В				
7th b	НЭ		⋖				
6th byte	FL		6				
eth	FH		8				
oyte	П					\ <u>+</u>	
5th byte	H		7		AND BLD BILD	BST BIST	
4th byte	סר		9		BAND BIAND		
4th	ВΗ		2		R BXOR BAND BLD BIOR BIXOR BIAND E		
3rd byte	CL		4		BOR BOR BIOR		
31	СН				<u> </u>		
oyte	BL		.,	-	<u> </u>		
2nd byte	ВН		2			FON	פכר
byte	AL		_			FOING	
1st byte	АН		0			FIG	D20
Instruction code:			AHALBHBL FHFLGH	6A30aaaaaaaa6*	6A30aaaaaaaa7*	6A38aaaaaaaa6*	6A38aaaaaaa7*

Note: * aa is the absolute address specification.

A.4 Number of States Required for Execution

The tables in this section can be used to calculate the number of states required for instruction execution by the H8S/2000 CPU. Table A.5 shows the number of instruction fetch, data read/write, and other cycles occurring in each instruction, and table A.4 shows the number of states required per cycle according to the bus size. The number of states required for execution of an instruction can be calculated from these two tables as follows:

Number of states =
$$I \times S_i + J \times S_j + K \times S_k + L \times S_i + M \times S_M + N \times S_N$$

Examples of Calculation of Number of States Required for Execution

Examples: Advanced mode, stack located in external address space, on-chip supporting modules accessed in two states with 8-bit bus width, external devices accessed in three states with one wait state and 16-bit bus width.

1. BSET #0,@FFFFC7:8

$$I = L = 2$$
 and $J = K = M = N = 0$

From table A.4.

$$S_{1} = 4 \text{ and } S_{1} = 2$$

Number of states =
$$2 \times 4 + 2 \times 2 = 12$$

2. JSR @@30

From table A.5,

$$I = J = K = 2$$
 and $L = M = N = 0$

From table A.4,

$$S_{I} = S_{I} = S_{K} = 4$$

Number of states =
$$2 \times 4 + 2 \times 4 + 2 \times 4 = 24$$

Table A.4 Number of States per Cycle

Access Conditions

				n-Chip		Externa	al Device	
				porting lodule	8-Bi	it Bus	16-B	it Bus
Cycle		On-Chip Memory	8-Bit Bus	16-Bit Bus	2-State Access	3-State Access	2-State Access	3-State Access
Instruction fetch	Sı	1	4	2	4	6 + 2m	2	3 + m
Branch address fetch	S	-						
Stack operation	S _K	_						
Byte data access	S _L	_	2		2	3 + m	-	
Word data access	$S_{\scriptscriptstyle M}$	-	4		4	6 + 2m	=	
Internal operation	S _N	1	1	1	1	1	1	1

Legend:

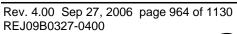
m: Number of wait states inserted into external device access



Table A.5 Number of Cycles per Instruction

Instruction	Mnemonic	Instruction Fetch I	Branch Address Read J	Stack Operation K	Byte Data Access L	Word Data Access M	Internal Operation N
ADD	ADD.B #xx:8,Rd	1					
	ADD.B Rs,Rd	1					
	ADD.W #xx:16,Rd	2					
	ADD.W Rs,Rd	1					
	ADD.L #xx:32,ERd	3					
	ADD.L ERs,ERd	1					
ADDS	ADDS #1/2/4,ERd	1					
ADDX	ADDX #xx:8,Rd	1					
	ADDX Rs,Rd	1					
AND	AND.B #xx:8,Rd	1					
	AND.B Rs,Rd	1					
	AND.W #xx:16,Rd	2					
	AND.W Rs,Rd	1					
	AND.L #xx:32,ERd	3					
	AND.L ERs,ERd	2					
ANDC	ANDC #xx:8,CCR	1					
	ANDC #xx:8,EXR	2					
BAND	BAND #xx:3,Rd	1					
	BAND #xx:3,@ERd	2			1		
	BAND #xx:3,@aa:8	2			1		
	BAND #xx:3,@aa:16	3			1		
	BAND #xx:3,@aa:32	4			1		
Bcc	BRA d:8 (BT d:8)	2					
	BRN d:8 (BF d:8)	2					
	BHI d:8	2					
	BLS d:8	2					
	BCC d:8 (BHS d:8)	2					
	BCS d:8 (BLO d:8)	2					
	BNE d:8	2					
	BEQ d:8	2					
	BVC d:8	2					
	BVS d:8	2					
	BPL d:8	2					

Instruction	Mnemonic	Instruction Fetch I	Branch Address Read J	Stack Operation K	Byte Data Access L	Word Data Access M	Internal Operation N
Bcc	BMI d:8	2					
	BGE d:8	2					
	BLT d:8	2					
	BGT d:8	2					
	BLE d:8	2					
	BRA d:16 (BT d:16)	2					1
	BRN d:16 (BF d:16)	2					1
	BHI d:16	2					1
	BLS d:16	2					1
	BCC d:16 (BHS d:16)	2					1
	BCS d:16 (BLO d:16)	2					1
	BNE d:16	2					1
	BEQ d:16	2					1
	BVC d:16	2					1
	BVS d:16	2					1
	BPL d:16	2					1
	BMI d:16	2					1
	BGE d:16	2					1
	BLT d:16	2					1
	BGT d:16	2					1
	BLE d:16	2					1
BCLR	BCLR #xx:3,Rd	1					
	BCLR #xx:3,@ERd	2			2		
	BCLR #xx:3,@aa:8	2			2		
	BCLR #xx:3,@aa:16	3			2		
	BCLR #xx:3,@aa:32	4			2		
	BCLR Rn,Rd	1					
	BCLR Rn,@ERd	2			2		
	BCLR Rn,@aa:8	2			2		
	BCLR Rn,@aa:16	3			2		
	BCLR Rn,@aa:32	4			2		
BIAND	BIAND #xx:3,Rd	1					
	BIAND #xx:3,@ERd	2			1		
	BIAND #xx:3,@aa:8	2			1		
	BIAND #xx:3,@aa:16	3			1		
	BIAND #xx:3,@aa:32	4			1		



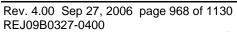


Instruction	Mnemonic	Instruction Fetch I	Branch Address Read J	Stack Operation K	Byte Data Access L	Word Data Access M	Internal Operation N
BILD	BILD #xx:3,Rd	1					
	BILD #xx:3,@ERd	2			1		
	BILD #xx:3,@aa:8	2			1		
	BILD #xx:3,@aa:16	3			1		
	BILD #xx:3,@aa:32	4			1		
BIOR	BIOR #xx:8,Rd	1					
	BIOR #xx:8,@ERd	2			1		
	BIOR #xx:8,@aa:8	2			1		
	BIOR #xx:8,@aa:16	3			1		
	BIOR #xx:8,@aa:32	4			1		
BIST	BIST #xx:3,Rd	1					
	BIST #xx:3,@ERd	2			2		
	BIST #xx:3,@aa:8	2			2		
	BIST #xx:3,@aa:16	3			2		
	BIST #xx:3,@aa:32	4			2		
BIXOR	BIXOR #xx:3,Rd	1					
	BIXOR #xx:3,@ERd	2			1		
	BIXOR #xx:3,@aa:8	2			1		
	BIXOR #xx:3,@aa:16	3			1		
	BIXOR #xx:3,@aa:32	4			1		
BLD	BLD #xx:3,Rd	1					
	BLD #xx:3,@ERd	2			1		
	BLD #xx:3,@aa:8	2			1		
	BLD #xx:3,@aa:16	3			1		
	BLD #xx:3,@aa:32	4			1		
BNOT	BNOT #xx:3,Rd	1					
	BNOT #xx:3,@ERd	2			2		
	BNOT #xx:3,@aa:8	2			2		
	BNOT #xx:3,@aa:16	3			2		
	BNOT #xx:3,@aa:32	4			2		
	BNOT Rn,Rd	1					
	BNOT Rn,@ERd	2			2		
	BNOT Rn,@aa:8	2			2		
	BNOT Rn,@aa:16	3			2		
	BNOT Rn,@aa:32	4			2		

Instruction	Mnemonic		Instruction Fetch I	Branch Address Read J	Stack Operation K	Byte Data Access L	Word Data Access M	Internal Operation N
BOR	BOR #xx:3,Rd	d	1					
	BOR #xx:3,@	ERd	2			1		
	BOR #xx:3,@	aa:8	2			1		
	BOR #xx:3,@	aa:16	3			1		
	BOR #xx:3,@	aa:32	4			1		
BSET	BSET #xx:3,R	ld.	1					
	BSET #xx:3,@	ERd	2			2		
	BSET #xx:3,@	aa:8	2			2		
	BSET #xx:3,@	@aa:16	3			2		
	BSET #xx:3,@	aa:32	4			2		
	BSET Rn,Rd		1					
	BSET Rn,@E	Rd	2			2		
	BSET Rn,@a	a:8	2			2		
	BSET Rn,@a	a:16	3			2		
	BSET Rn,@a	a:32	4			2		
BSR	BSR d:8	Normal	2		1			
		Advanced	2		2			
	BSR d:16	Normal	2		1			1
		Advanced	2		2			1
BST	BST #xx:3,Rd		1					
	BST #xx:3,@I	ERd	2			2		
	BST #xx:3,@a	aa:8	2			2		
	BST #xx:3,@a	aa:16	3			2		
	BST #xx:3,@a	aa:32	4			2		
BTST	BTST #xx:3,R	d	1					
	BTST #xx:3,@	ERd	2			1		
	BTST #xx:3,@	aa:8	2			1		
	BTST #xx:3,@	aa:16	3			1		
	BTST #xx:3,@	aa:32	4			1		
	BTST Rn,Rd		1					
	BTST Rn,@E	Rd	2			1		
	BTST Rn,@aa	a:8	2			1		
	BTST Rn,@aa	a:16	3			1		
	BTST Rn,@aa	a:32	4			1		

Instruction	Mnemonic		Instruction Fetch I	Branch Address Read J	Stack Operation K	Byte Data Access L	Word Data Access M	Internal Operation N
BXOR	BXOR #xx:3,Rd		1					
	BXOR #xx:3,@E	Rd	2			1		
	BXOR #xx:3,@a	ia:8	2			1		
	BXOR #xx:3,@a	a:16	3			1		
	BXOR #xx:3,@a	ia:32	4			1		
CLRMAC	CLRMAC		Cannot be us	sed with this I	_SI.			
CMP	CMP.B #xx:8,Rd	I	1					
	CMP.B Rs,Rd		1					
	CMP.W #xx:16,F	Rd	2					
	CMP.W Rs,Rd		1					
	CMP.L #xx:32,E	Rd	3					
	CMP.L ERs,ERo	i	1					
DAA	DAA Rd		1					
DAS	DAS Rd		1					
DEC	DEC.B Rd		1					
	DEC.W #1/2,Rd		1					
	DEC.L #1/2,ERd	İ	1					
DIVXS	DIVXS.B Rs,Rd		2					11
	DIVXS.W Rs,ER	ld	2					19
DIVXU	DIVXU.B Rs,Rd		1					11
	DIVXU.W Rs,ER	Rd	1					19
EEPMOV	EEPMOV.B		2			2n+2*2		
	EEPMOV.W		2			2n+2*2		
EXTS	EXTS.W Rd		1					
	EXTS.L ERd		1					
EXTU	EXTU.W Rd		1					
	EXTU.L ERd		1					
INC	INC.B Rd		1					
	INC.W #1/2,Rd		1					
	INC.L #1/2,ERd		1					
JMP	JMP @ERn		2					
	JMP @aa:24		2					1
	JMP @@aa:8	Normal	2	1				1
		Advanced	2	2				1

Instruction	Mnemonic		Instruction Fetch I	Branch Address Read J	Stack Operation K	Byte Data Access L	Word Data Access M	Internal Operation N
JSR	JSR @ERn	Normal	2		1			
		Advanced	2		2			
	JSR @aa:24	Normal	2		1			1
		Advanced	2		2			1
	JSR @@aa:8	Normal	2	1	1			
		Advanced	2	2	2			
LDC	LDC #xx:8,CCR		1					
	LDC #xx:8,EXR		2					
	LDC Rs,CCR		1					
	LDC Rs,EXR		1					
	LDC @ERs,CCI	R	2				1	
	LDC @ERs,EXF	₹	2				1	
	LDC @(d:16,ER	Rs),CCR	3				1	
	LDC @(d:16,ER	Rs),EXR	3				1	
	LDC @(d:32,ER	Rs),CCR	5				1	
	LDC @(d:32,ER	Rs),EXR	5				1	
	LDC @ERs+,C0	CR	2				1	1
	LDC @ERs+,E>	(R	2				1	1
	LDC @aa:16,C0	CR	3				1	
	LDC @aa:16,EX	(R	3				1	
	LDC @aa:32,C0	CR	4				1	
	LDC @aa:32,E	KR	4				1	
LDM*4	LDM.L @SP+, (E	Rn-ERn+1)	2		4			1
	LDM.L @SP+, (E	Rn-ERn+2)	2		6			1
	LDM.L @SP+, (E	Rn-ERn+3)	2		8			1
LDMAC	LDMAC ERs, M	ACH	Cannot be us	ed with this I	LSI.			
	LDMAC ERs, M	ACL						
MAC	MAC @ERn+, @	@ERm+	_					
MOV	MOV.B #xx:8,Ro	d	1					
	MOV.B Rs,Rd		1					
	MOV.B @ERs,F	₹d	1			1		
	MOV.B @(d:16,	ERs),Rd	2			1		
	MOV.B @(d:32,	ERs),Rd	4			1		
	MOV.B @ERs+	,Rd	1			1		1
	MOV.B @aa:8,F	₹d	1			1		
	MOV.B @aa:16	,Rd	2			1		



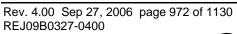


Instruction	Mnemonic	Instruction Fetch I	Branch Address Read J	Stack Operation K	Byte Data Access L	Word Data Access M	Internal Operation N
MOV	MOV.B @aa:32,Rd	3			1		
	MOV.B Rs,@ERd	1			1		
	MOV.B Rs,@(d:16,ERd)	2			1		
	MOV.B Rs,@(d:32,ERd)	4			1		
	MOV.B Rs,@-ERd	1			1		1
	MOV.B Rs,@aa:8	1			1		
	MOV.B Rs,@aa:16	2			1		
	MOV.B Rs,@aa:32	3			1		
	MOV.W #xx:16,Rd	2					
	MOV.W Rs,Rd	1					
	MOV.W @ERs,Rd	1				1	
	MOV.W @(d:16,ERs),Rd	2				1	
	MOV.W @(d:32,ERs),Rd	4				1	
	MOV.W @ERs+,Rd	1				1	1
	MOV.W @aa:16,Rd	2				1	
	MOV.W @aa:32,Rd	3				1	
	MOV.W Rs,@ERd	1				1	
	MOV.W Rs,@(d:16,ERd)	2				1	
	MOV.W Rs,@(d:32,ERd)	4				1	
	MOV.W Rs,@-ERd	1				1	1
	MOV.W Rs,@aa:16	2				1	
	MOV.W Rs,@aa:32	3				1	
	MOV.L #xx:32,ERd	3					
	MOV.L ERs,ERd	1					
	MOV.L @ERs,ERd	2				2	
	MOV.L @(d:16,ERs),ERd	3				2	
	MOV.L @(d:32,ERs),ERd	5				2	
	MOV.L @ERs+,ERd	2				2	1
	MOV.L @aa:16,ERd	3				2	
	MOV.L @aa:32,ERd	4				2	
	MOV.L ERs,@ERd	2				2	
	MOV.L ERs,@(d:16,ERd)	3				2	
	MOV.L ERs,@(d:32,ERd)	5				2	
	MOV.L ERs,@-ERd	2				2	1
	MOV.L ERs,@aa:16	3				2	
	MOV.L ERs,@aa:32	4				2	

Instruction	Mnemonic	Instruction Fetch I	Branch Address Read J	Stack Operation K	Byte Data Access L	Word Data Access M	Internal Operation N
MOVFPE	MOVFPE @:aa:16,Rd	Cannot be us	ed with this I	_SI.			
MOVTPE	MOVTPE Rs,@:aa:16						
MULXS	MULXS.B Rs,Rd	2					11
	MULXS.W Rs,ERd	2					19
MULXU	MULXU.B Rs,Rd	1					11
	MULXU.W Rs,ERd	1					19
NEG	NEG.B Rd	1					
	NEG.W Rd	1					
	NEG.L ERd	1					
NOP	NOP	1					
NOT	NOT.B Rd	1					
	NOT.W Rd	1					
	NOT.L ERd	1					
OR	OR.B #xx:8,Rd	1					
	OR.B Rs,Rd	1					
	OR.W #xx:16,Rd	2					
	OR.W Rs,Rd	1					
	OR.L #xx:32,ERd	3					
	OR.L ERs,ERd	2					
ORC	ORC #xx:8,CCR	1					
	ORC #xx:8,EXR	2					
POP	POP.W Rn	1				1	1
	POP.L ERn	2				2	1
PUSH	PUSH.W Rn	1				1	1
	PUSH.L ERn	2				2	1
ROTL	ROTL.B Rd	1					
	ROTL.B #2,Rd	1					
	ROTL.W Rd	1					
	ROTL.W #2,Rd	1					
	ROTL.L ERd	1					
	ROTL.L #2,ERd	1					

Internal Operation N
1
1
1
_

Instruction	Mnemonic	Instruction Fetch	Branch Address Read J	Stack Operation K	Byte Data Access L	Word Data Access M	Internal Operation N
SHLL	SHLL.B Rd	1					
	SHLL.B #2,Rd	1					
	SHLL.W Rd	1					
	SHLL.W #2,Rd	1					
	SHLL.L ERd	1					
	SHLL.L #2,ERd	1					
SHLR	SHLR.B Rd	1					
	SHLR.B #2,Rd	1					
	SHLR.W Rd	1					
	SHLR.W #2,Rd	1					
	SHLR.L ERd	1					
	SHLR.L #2,ERd	1					
SLEEP	SLEEP	1					1
STC	STC.B CCR,Rd	1					
	STC.B EXR,Rd	1					
	STC.W CCR,@ERd	2				1	
	STC.W EXR,@ERd	2				1	
	STC.W CCR,@(d:16,ERd)	3				1	
	STC.W EXR,@(d:16,ERd)	3				1	
	STC.W CCR,@(d:32,ERd)	5				1	
	STC.W EXR,@(d:32,ERd)	5				1	
	STC.W CCR,@-ERd	2				1	1
	STC.W EXR,@-ERd	2				1	1
	STC.W CCR,@aa:16	3				1	
	STC.W EXR,@aa:16	3				1	
	STC.W CCR,@aa:32	4				1	
	STC.W EXR,@aa:32	4				1	
STM*4	STM.L (ERn-ERn+1),@-SP	2		4			1
	STM.L (ERn-ERn+2),@-SP	2		6			1
	STM.L (ERn-ERn+3),@-SP	2		8			1
SUB	SUB.B Rs,Rd	1					
	SUB.W #xx:16,Rd	2					
	SUB.W Rs,Rd	1					
	SUB.L #xx:32,ERd	3					
	SUB.L ERs,ERd	1					





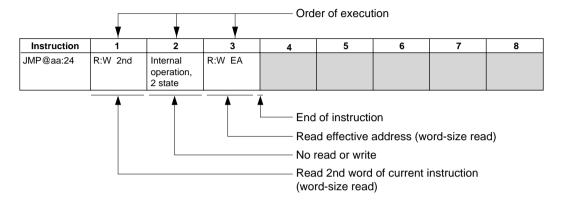
Instruction	Mnemonic		Instruction Fetch I	Branch Address Read J	Stack Operation K	Byte Data Access L	Word Data Access M	Internal Operation N
SUBS	SUBS #1/2/4,	ERd	1					
SUBX	SUBX #xx:8,F	Rd	1					
	SUBX Rs,Rd		1					
TAS	TAS @ERd*3		2			2		
TRAPA	TRAPA #x:2	Normal	2	1	2/3*1			2
		Advanced	2	2	2/3*1			2
XOR	XOR.B #xx:8,	Rd	1					
	XOR.B Rs,Rd		1					
	XOR.W #xx:1	6,Rd	2					
	XOR.W Rs,Rd	d	1					
	XOR.L #xx:32	,ERd	3					
	XOR.L ERs,E	Rd	2					
XORC	XORC #xx:8,0	CCR	1					
	XORC #xx:8,E	EXR	2					

- Notes: 1. 2 when EXR is invalid, 3 when valid.
 - 2. When n bytes of data are transferred.
 - 3. Only register ER0, ER1, ER4, or ER5 should be used when using the TAS instruction.
 - 4. Only registers ER0 to ER6 should be used when using the STM/LDM instruction.

A.5 Bus States during Instruction Execution

Table A.6 indicates the types of cycles that occur during instruction execution by the CPU. See table A.4 for the number of states per cycle.

How to Read the Table:



Legend:

R:B	Byte-size read	
R:W	Word-size read	
W:B	Byte-size write	
W:W	Word-size write	
:M	Transfer of the bus is not performed immediately after this cycle	
2nd	Address of 2nd word (3rd and 4th bytes)	
3rd	Address of 3rd word (5th and 6th bytes)	
4th	Address of 4th word (7th and 8th bytes)	
5th	Address of 5th word (9th and 10th bytes)	
NEXT	Start address of instruction following executing instruction	
EA	Effective address	
VEC	Vector address	

Figure A.1 shows timing waveforms for the address bus and the \overline{RD} , \overline{HWR} , and \overline{LWR} signals during execution of the above instruction with an 8-bit bus, using three-state access with no wait states.

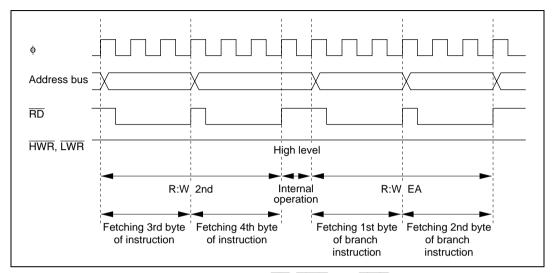


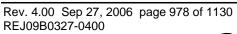
Figure A.1 Address Bus, RD, HWR, and LWR Timing (8-Bit Bus, Three-State Access, No Wait States)

Table A.6 Instruction Execution Cycle

Instruction	1	2	3	4	5	6	7	8	9
ADD.B #xx:8,Rd	R:W NEXT								
ADD.B Rs,Rd	R:W NEXT								
ADD.W #xx:16,Rd	R:W 2nd	R:W NEXT							
ADD.W Rs,Rd	R:W NEXT								
ADD.L #xx:32,ERd	R:W 2nd	R:W 3rd	R:W NEXT						
ADD.L ERs,ERd	R:W NEXT								
ADDS #1/2/4,ERd	R:W NEXT								
ADDX #xx:8,Rd	R:W NEXT								
ADDX Rs,Rd	R:W NEXT								
AND.B #xx:8,Rd	R:W NEXT								
AND.B Rs,Rd	R:W NEXT								
AND.W #xx:16,Rd	R:W 2nd	R:W NEXT							
AND.W Rs,Rd	R:W NEXT								
AND.L #xx:32,ERd	R:W 2nd	R:W 3rd	R:W NEXT						
AND.L ERs,ERd	R:W 2nd	R:W NEXT							
ANDC #xx:8,CCR	R:W NEXT								
ANDC #xx:8,EXR	R:W 2nd	R:W NEXT							
BAND #xx:3,Rd	R:W NEXT								
BAND #xx:3,@ERd	R:W 2nd	R:B EA	R:W:M NEXT						
BAND #xx:3,@aa:8	R:W 2nd	R:B EA	R:W:M NEXT						
BAND #xx:3, @aa:16	R:W 2nd	R:W 3rd	R:B EA	R:W:M NEXT					
BAND #xx:3, @aa:32	R:W 2nd	R:W 3rd	R:W 4th	R:B EA	R:W:M NEXT				
BRA d:8 (BT d:8)	R:W NEXT	R:W EA							
BRN d:8 (BF d:8)	R:W NEXT	R:W EA							
BHI d:8	R:W NEXT	R:W EA							
BLS d:8	R:W NEXT	R:W EA							
BCC d:8 (BHS d:8)	R:W NEXT	R:W EA							
BCS d:8 (BLO d:8)	R:W NEXT	R:W EA							
BNE d:8	R:W NEXT	R:W EA							
BEQ d:8	R:W NEXT	R:W EA							
BVC d:8	R:W NEXT	R:W EA							
BVS d:8	R:W NEXT	R:W EA							
BPL d:8	R:W NEXT	R:W EA							

Instruction	1	2	3	4	5	6	7	8	9
BMI d:8	R:W NEXT	R:W EA							
BGE d:8	R:W NEXT	R:W EA							
BLT d:8	R:W NEXT	R:W EA							
BGT d:8	R:W NEXT	R:W EA							
BLE d:8	R:W NEXT	R:W EA							
BRA d:16 (BT d:16)	R:W 2nd	Internal operation, 1 state	R:W EA						
BRN d:16 (BF d:16)	R:W 2nd	Internal operation, 1 state	R:W EA						
BHI d:16	R:W 2nd	Internal operation, 1 state	R:W EA						
BLS d:16	R:W 2nd	Internal operation, 1 state	R:W EA						
BCC d:16 (BHS d:16)	R:W 2nd	Internal operation, 1 state	R:W EA						
BCS d:16 (BLO d:16)	R:W 2nd	Internal operation, 1 state	R:W EA						
BNE d:16	R:W 2nd	Internal operation, 1 state	R:W EA						
BEQ d:16	R:W 2nd	Internal operation, 1 state	R:W EA						
BVC d:16	R:W 2nd	Internal operation, 1 state	R:W EA						
BVS d:16	R:W 2nd	Internal operation, 1 state	R:W EA						
BPL d:16	R:W 2nd	Internal operation, 1 state	R:W EA						
BMI d:16	R:W 2nd	Internal operation, 1 state	R:W EA						
BGE d:16	R:W 2nd	Internal operation, 1 state	R:W EA						

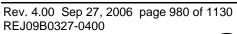
Instruction	1	2	3	4	5	6	7	8	9
BLT d:16	R:W 2nd	Internal operation, 1 state	R:W EA						
BGT d:16	R:W 2nd	Internal operation, 1 state	R:W EA						
BLE d:16	R:W 2nd	Internal operation, 1 state	R:W EA						
BCLR #xx:3,Rd	R:W NEXT								
BCLR #xx:3,@ERd	R:W 2nd	R:B:M EA	R:W:M NEXT	W:B EA					
BCLR #xx:3,@aa:8	R:W 2nd	R:B:M EA	R:W:M NEXT	W:B EA					
BCLR#xx:3,@aa:16	R:W 2nd	R:W 3rd	R:B:M EA	R:W:M NEXT	W:B EA				
BCLR#xx:3,@aa:32	R:W 2nd	R:W 3rd	R:W 4th	R:B:M EA	R:W:M NEXT	W:B EA			
BCLR Rn,Rd	R:W NEXT								
BCLR Rn,@ERd	R:W 2nd	R:B:M EA	R:W:M NEXT	W:B EA					
BCLR Rn,@aa:8	R:W 2nd	R:B:M EA	R:W:M NEXT	W:B EA					
BCLR Rn,@aa:16	R:W 2nd	R:W 3rd	R:B:M EA	R:W:M NEXT	W:B EA				
BCLR Rn,@aa:32	R:W 2nd	R:W 3rd	R:W 4th	R:B:M EA	R:W:M NEXT	W:B EA			
BIAND #xx:3,Rd	R:W NEXT								
BIAND #xx:3, @ERd	R:W 2nd	R:B EA	R:W:M NEXT						
BIAND #xx:3, @aa:8	R:W 2nd	R:B EA	R:W:M NEXT						
BIAND #xx:3, @aa:16	R:W 2nd	R:W 3rd	R:B EA	R:W:M NEXT					
BIAND #xx:3, @aa:32	R:W 2nd	R:W 3rd	R:W 4th	R:B EA	R:W:M NEXT				
BILD #xx:3,Rd	R:W NEXT								
BILD #xx:3,@ERd	R:W 2nd	R:B EA	R:W:M NEXT						
BILD #xx:3,@aa:8	R:W 2nd	R:B EA	R:W:M NEXT						
BILD #xx:3,@aa:16	R:W 2nd	R:W 3rd	R:B: EA	R:W:M NEXT					





Instruction	1	2	3	4	5	6	7	8	9
BILD #xx:3,@aa:32	R:W 2nd	R:W 3rd	R:W 4th	R:B EA	R:W:M NEXT				
BIOR #xx:3,Rd	R:W NEXT								
BIOR #xx:3,@ERd	R:W 2nd	R:B EA	R:W:M NEXT						
BIOR #xx:3,@aa:8	R:W 2nd	R:B EA	R:W:M NEXT						
BIOR #xx:3,@aa:16	R:W 2nd	R:W 3rd	R:B EA	R:W:M NEXT					
BIOR #xx:3,@aa:32	R:W 2nd	R:W 3rd	R:W 4th	R:B EA	R:W:M NEXT				
BIST #xx:3,Rd	R:W NEXT								
BIST #xx:3,@ERd	R:W 2nd	R:B:M EA	R:W:M NEXT	W:B EA					
BIST #xx:3,@aa:8	R:W 2nd	R:B:M EA	R:W:M NEXT	W:B EA					
BIST #xx:3,@aa:16	R:W 2nd	R:W 3rd	R:B:M EA	R:W:M NEXT	W:B EA				
BIST #xx:3,@aa:32	R:W 2nd	R:W 3rd	R:W 4th	R:B:M EA	R:W:M NEXT	W:B EA			
BIXOR #xx:3,Rd	R:W NEXT								
BIXOR #xx:3, @ERd	R:W 2nd	R:B EA	R:W:M NEXT						
BIXOR #xx:3, @aa:8	R:W 2nd	R:B EA	R:W:M NEXT						
BIXOR #xx:3, @aa:16	R:W 2nd	R:W 3rd	R:B EA	R:W:M NEXT					
BIXOR #xx:3, @aa:32	R:W 2nd	R:W 3rd	R:W 4th	R:B EA	R:W:M NEXT				
BLD #xx:3,Rd	R:W NEXT								
BLD #xx:3,@ERd	R:W 2nd	R:B EA	R:W:M NEXT						
BLD #xx:3,@aa:8	R:W 2nd	R:B EA	R:W:M NEXT						
BLD #xx:3,@aa:16	R:W 2nd	R:W 3rd	R:B EA	R:W:M NEXT					
BLD #xx:3,@aa:32	R:W 2nd	R:W 3rd	R:W 4th	R:B EA	R:W:M NEXT				
BNOT #xx:3,Rd	R:W NEXT								
BNOT #xx:3,@ERd	R:W 2nd	R:B:M EA	R:W:M NEXT	W:B EA					

Instruction	1	2	3	4	5	6	7	8	9
BNOT #xx:3,@aa:8	R:W 2nd	R:B:M EA	R:W:M NEXT	W:B EA					
BNOT #xx:3, @aa:16	R:W 2nd	R:W 3rd	R:B:M EA	R:W:M NEXT	W:B EA				
BNOT #xx:3, @aa:32	R:W 2nd	R:W 3rd	R:W 4th	R:B:M EA	R:W:M NEXT	W:B EA			
BNOT Rn,Rd	R:W NEXT								
BNOT Rn,@ERd	R:W 2nd	R:B:M EA	R:W:M NEXT	W:B EA					
BNOT Rn,@aa:8	R:W 2nd	R:B:M EA	R:W:M NEXT	W:B EA					
BNOT Rn,@aa:16	R:W 2nd	R:W 3rd	R:B:M EA	R:W:M NEXT	W:B EA				
BNOT Rn,@aa:32	R:W 2nd	R:W 3rd	R:W 4th	R:B:M EA	R:W:M NEXT	W:B EA			
BOR #xx:3,Rd	R:W NEXT								
BOR #xx:3,@ERd	R:W 2nd	R:B EA	R:W:M NEXT						
BOR #xx:3,@aa:8	R:W 2nd	R:B EA	R:W:M NEXT						
BOR #xx:3,@aa:16	R:W 2nd	R:W 3rd	R:B EA	R:W:M NEXT					
BOR #xx:3,@aa:32	R:W 2nd	R:W 3rd	R:W 4th	R:B EA	R:W NEXT				
BSET #xx:3,Rd	R:W NEXT								
BSET #xx:3,@ERd	R:W 2nd	R:B:M EA	R:W:M NEXT	W:B EA					
BSET #xx:3,@aa:8	R:W 2nd	R:B:M EA	R:W:M NEXT	W:B EA					
BSET #xx:3, @aa:16	R:W 2nd	R:W 3rd	R:B:M EA	R:W:M NEXT	W:B EA				
BSET #xx:3, @aa:32	R:W 2nd	R:W 3rd	R:W 4th	R:B:M EA	R:W:M NEXT	W:B EA			
BSET Rn,Rd	R:W NEXT								
BSET Rn,@ERd	R:W 2nd	R:B:M EA	R:W:M NEXT	W:B EA					
BSET Rn,@aa:8	R:W 2nd	R:B:M EA	R:W:M NEXT	W:B EA					
BSET Rn,@aa:16	R:W 2nd	R:W 3rd	R:B:M EA	R:W:M NEXT	W:B EA				
BSET Rn,@aa:32	R:W 2nd	R:W 3rd	R:W 4th	R:B:M EA	R:W:M NEXT	W:B EA			





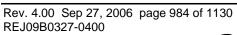
Instr	uction	1	2	3	4	5	6	7	8	9
BSR d:8	Advanced	R:W NEXT	R:W EA	W:W:M Stack (H)	W:W Stack (L)					
BSR d:16	Advanced	R:W 2nd	Internal operation, 1 state	R:W EA	W:W:M Stack (H)	W:W Stack (L)				
BST #xx:	3,Rd	R:W NEXT								
BST #xx:	3,@ERd	R:W 2nd	R:B:M EA	R:W:M NEXT	W:B EA					
BST #xx:	3,@aa:8	R:W 2nd	R:B:M EA	R:W:M NEXT	W:B EA					
BST #xx:	3, @aa:16	R:W 2nd	R:W 3rd	R:B:M EA	R:W:M NEXT	W:B EA				
BST #xx:	3,@aa:32	R:W 2nd	R:W 3rd	R:W 4th	R:B:M EA	R:W:M NEXT	W:B EA			
BTST #xx	k:3,Rd	R:W NEXT								
BTST #xx	c:3,@ERd	R:W 2nd	R:B EA	R:W:M NEXT						
BTST #xx	c:3,@aa:8	R:W 2nd	R:B EA	R:W:M NEXT						
BTST #xx @aa:16	c:3,	R:W 2nd	R:W 3rd	R:B EA	R:W:M NEXT					
BTST #xx @aa:32	c:3,	R:W 2nd	R:W 3rd	R:W 4th	R:B EA	R:W:M NEXT				
BTST Rn	,Rd	R:W NEXT								
BTST Rn	,@ERd	R:W 2nd	R:B EA	R:W:M NEXT						
BTST Rn	,@aa:8	R:W 2nd	R:B EA	R:W:M NEXT						
BTST Rn	,@aa:16	R:W 2nd	R:W 3rd	R:B EA	R:W:M NEXT					
BTST Rn	,@aa:32	R:W 2nd	R:W 3rd	R:W 4th	R:B EA	R:W:M NEXT				
BXOR #x	x:3,Rd	R:W NEXT								
BXOR #x	x:3,@ERd	R:W 2nd	R:B EA	R:W:M NEXT						
BXOR #x	x:3,@aa:8	R:W 2nd	R:B EA	R:W:M NEXT						
BXOR #x @aa:16	x:3,	R:W 2nd	R:W 3rd	R:B EA	R:W:M NEXT					
BXOR #x @aa:32	x:3,	R:W 2nd	R:W 3rd	R:W 4th	R:B EA	R:W:M NEXT				
CLRMAC	:	Cannot be	used in this	LSI	·	·				

Instr	uction	1	2	3	4	5	6	7	8	9
CMP.B #	xx:8,Rd	R:W NEXT								
CMP.B R	s,Rd	R:W NEXT								
CMP.W #	xx:16,Rd	R:W 2nd	R:W NEXT							
CMP.W F	Rs,Rd	R:W NEXT								
CMP.L #	xx:32,ERd	R:W 2nd	R:W 3rd	R:W NEXT						
CMP.L E	Rs,ERd	R:W NEXT								
DAA Rd		R:W NEXT								
DAS Rd		R:W NEXT								
DEC.B R	d	R:W NEXT								
DEC.W#	1/2,Rd	R:W NEXT								
DEC.L #1	I/2,ERd	R:W NEXT								
DIVXS.B	Rs,Rd	R:W 2nd	R:W NEXT	Internal ope	eration, 11 s	tates				
DIVXS.W	Rs,ERd	R:W 2nd	R:W NEXT	Internal ope	eration, 19 s	tates	1			
DIVXU.B	Rs,Rd	R:W NEXT	Internal ope	eration, 11 s	ates					
DIVXU.W	/ Rs,ERd	R:W NEXT	Internal ope	eration, 19 s	ates	1	•			
EEPMOV	′.B	R:W 2nd	R:B EAs*1	R:B EAd*1	R:B EAs*2	W:B EAd*2	R:W NEXT			
EEPMOV	′.W	R:W 2nd	R:B EAs*1	R:B EAd*1	R:B EAs*2	W:B EAd*2	R:W NEXT			
EXTS.W	Rd	R:W NEXT			← Repeated	d n times *_2 \rightarrow				
EXTS.L E	Rd	R:W NEXT								
EXTU.W	Rd	R:W NEXT								
EXTU.L E	ERd	R:W NEXT								
INC.B Ro		R:W NEXT								
INC.W #1	I/2,Rd	R:W NEXT								
INC.L #1	/2,ERd	R:W NEXT								
JMP @E	Rn	R:W NEXT	R:W EA							
JMP @aa	a:24	R:W 2nd	Internal operation, 1 state	R:W EA						
JMP @@aa:8		R:W NEXT	R:W:M aa:8	R:W aa:8	Internal operation, 1 state	R:W EA				
JSR @ERn	Advanced	R:W NEXT	R:W EA	W:W:M Stack (H)	W:W Stack (L)					
JSR @aa:24	Advanced	R:W 2nd	Internal operation, 1 state	R:W EA	W:W:M Stack (H)	W:W Stack (L)				
JSR @@aa:8	Advanced	R:W NEXT	R:W:M aa:8	R:W aa:8	W:W:M Stack (H)	W:W Stack (L)	R:W EA			



Instruction	1	2	3	4	5	6	7	8	9
LDC #xx:8,CCR	R:W NEXT								
LDC #xx:8,EXR	R:W 2nd	R:W NEXT							
LDC Rs,CCR	R:W NEXT								
LDC Rs,EXR	R:W NEXT								
LDC @ERs,CCR	R:W 2nd	R:W NEXT	R:W EA						
LDC @ERs,EXR	R:W 2nd	R:W NEXT	R:W EA						
LDC@(d:16,ERs), CCR	R:W 2nd	R:W 3rd	R:W NEXT	R:W EA					
LDC@(d:16,ERs), EXR	R:W 2nd	R:W 3rd	R:W NEXT	R:W EA					
LDC@(d:32,ERs), CCR	R:W 2nd	R:W 3rd	R:W 4th	R:W 5th	R:W NEXT	R:W EA			
LDC@(d:32,ERs), EXR	R:W 2nd	R:W 3rd	R:W 4th	R:W 5th	R:W NEXT	R:W EA			
LDC @ERs+,CCR	R:W 2nd	R:W NEXT	Internal operation, 1 state	R:W EA					
LDC @ERs+,EXR	R:W 2nd	R:W NEXT	Internal operation, 1 state	R:W EA					
LDC @aa:16,CCR	R:W 2nd	R:W 3rd	R:W NEXT	R:W EA					
LDC @aa:16,EXR	R:W 2nd	R:W 3rd	R:W NEXT	R:W EA					
LDC @aa:32,CCR	R:W 2nd	R:W 3rd	R:W 4th	R:W NEXT	R:W EA				
LDC @aa:32,EXR	R:W 2nd	R:W 3rd	R:W 4th	R:W NEXT	R:W EA				
LDM.L @SP+, (ERn-ERn+1)*9	R:W 2nd	R:W:M NEXT	Internal operation, 1 state	R:W:M Stack (H)*3	R:W Stack (L)*3				
LDM.L @SP+, (ERn-ERn+2)*9	R:W 2nd	R:W:M NEXT	Internal operation, 1 state	R:W:M Stack (H)*3	R:W Stack (L)*3				
LDM.L @SP+, (ERn-ERn+3)*9	R:W 2nd	R:W:M NEXT	Internal operation, 1 state	R:W:M Stack (H)*3	R:W Stack (L)*3				
LDMAC ERs,MACH	Cannot be	used in this	LSI						
LDMAC ERs,MACL									
MAC @ERn+, @ERm+									
MOV.B #xx:8,Rd	R:W NEXT								
MOV.B Rs,Rd	R:W NEXT								
MOV.B @ERs,Rd	R:W NEXT	R:B EA							
MOV.B @(d:16,ERs),Rd	R:W 2nd	R:W NEXT	R:B EA						

Instruction	1	2	3	4	5	6	7	8	9
MOV.B @(d:32,ERs),Rd	R:W 2nd	R:W 3rd	R:W 4th	R:W NEXT	R:B EA				
MOV.B @ERs+,Rd	R:W NEXT	Internal operation, 1 state	R:B EA						
MOV.B @aa:8,Rd	R:W NEXT	R:B EA							
MOV.B @aa:16,Rd	R:W 2nd	R:W NEXT	R:B EA						
MOV.B @aa:32,Rd	R:W 2nd	R:W 3rd	R:W NEXT	R:B EA					
MOV.B Rs,@ERd	R:W NEXT	W:B EA							
MOV.B Rs, @(d:16,ERd)	R:W 2nd	R:W NEXT	W:B EA						
MOV.B Rs, @(d:32,ERd)	R:W 2nd	R:W 3rd	R:W 4th	R:W NEXT	W:B EA				
MOV.B Rs,@-ERd	R:W NEXT	Internal operation, 1 state	W:B EA						
MOV.B Rs,@aa:8	R:W NEXT	W:B EA							
MOV.B Rs,@aa:16	R:W 2nd	R:W NEXT	W:B EA						
MOV.B Rs,@aa:32	R:W 2nd	R:W 3rd	R:W NEXT	W:B EA					
MOV.W #xx:16,Rd	R:W 2nd	R:W NEXT							
MOV.W Rs,Rd	R:W NEXT								
MOV.W @ERs,Rd	R:W NEXT	R:W EA							
MOV.W @(d:16,ERs),Rd	R:W 2nd	R:W NEXT	R:W EA						
MOV.W @(d:32,ERs),Rd	R:W 2nd	R:W 3rd	R:W 4th	R:W NEXT	R:W EA				
MOV.W @ERs+,Rd	R:W NEXT	Internal operation, 1 state	R:W EA						
MOV.W @aa:16,Rd	R:W 2nd	R:W NEXT	R:W EA						
MOV.W @aa:32,Rd	R:W 2nd	R:W 3rd	R:W NEXT	R:B EA					
MOV.W Rs,@ERd	R:W NEXT	W:W EA							
MOV.W Rs, @(d:16,ERd)	R:W 2nd	R:W NEXT	W:W EA						
MOV.W Rs, @(d:32,ERd)	R:W 2nd	R:W 3rd	R:W 4th	R:W NEXT	W:W EA				
MOV.W Rs,@-ERd	R:W NEXT	Internal operation, 1 state	W:W EA						
MOV.W Rs,@aa:16	R:W 2nd	R:W NEXT	W:W EA						
MOV.W Rs,@aa:32	R:W 2nd	R:W 3rd	R:W NEXT	W:W EA					





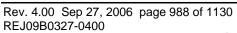
Instruction	1	2	3	4	5	6	7	8	9
MOV.L #xx:32,ERd	R:W 2nd	R:W 3rd	R:W NEXT						
MOV.L ERs,ERd	R:W NEXT								
MOV.L @ERs,ERd	R:W 2nd	R:W:M NEXT	R:W:M EA	R:W EA+2					
MOV.L @(d:16,ERs),ERd	R:W 2nd	R:W:M 3rd	R:W NEXT	R:W:M EA	R:W EA+2				
MOV.L @(d:32,ERs),ERd	R:W 2nd	R:W:M 3rd	R:W:M 4th	R:W 5th	R:W NEXT	R:W:M EA	R:W EA+2		
MOV.L @ERs+, ERd	R:W 2nd	R:W:M NEXT	Internal operation, 1 state	R:W:M EA	R:W EA+2				
MOV.L @aa:16, ERd	R:W 2nd	R:W:M 3rd	R:W NEXT	R:W:M EA	R:W EA+2				
MOV.L @aa:32, ERd	R:W 2nd	R:W:M 3rd	R:W 4th	R:W NEXT	R:W:M EA	R:W EA+2			
MOV.L ERs,@ERd	R:W 2nd	R:W:M NEXT	W:W:M EA	W:W EA+2					
MOV.L ERs, @(d:16,ERd)	R:W 2nd	R:W:M 3rd	R:W NEXT	W:W:M EA	W:W EA+2				
MOV.L ERs, @(d:32,ERd)	R:W 2nd	R:W:M 3rd	R:W:M 4th	R:W 5th	R:W NEXT	W:W:M EA	W:W EA+2		
MOV.L ERs,@-ERd	R:W 2nd	R:W:M NEXT	Internal operation, 1 state	W:W:M EA	W:W EA+2				
MOV.L ERs, @aa:16	R:W 2nd	R:W:M 3rd	R:W NEXT	W:W:M EA	W:W EA+2				
MOV.L ERs, @aa:32	R:W 2nd	R:W:M 3rd	R:W 4th	R:W NEXT	W:W:M EA	W:W EA+2			
MOVFPE @aa:16,Rd	Cannot be	used in this	LSI	I.	I.	I.			
MOVTPE Rs,@aa:16									
MULXS.B Rs,Rd	R:W 2nd	R:W NEXT	Internal ope	eration, 11 s	tates				
MULXS.W Rs,ERd	R:W 2nd	R:W NEXT	Internal ope	eration, 19 s	tates				
MULXU.B Rs,Rd	R:W NEXT	Internal ope	eration, 11 st	tates					
MULXU.W Rs,ERd	R:W NEXT	Internal ope	eration, 19 st	tates					
NEG.B Rd	R:W NEXT								
NEG.W Rd	R:W NEXT								
NEG.L ERd	R:W NEXT								
NOP	R:W NEXT								
NOT.B Rd	R:W NEXT								
NOT.W Rd	R:W NEXT								

Instruction	1	2	3	4	5	6	7	8	9
NOT.L ERd	R:W NEXT								
OR.B #xx:8,Rd	R:W NEXT								
OR.B Rs,Rd	R:W NEXT								
OR.W #xx:16,Rd	R:W 2nd	R:W NEXT							
OR.W Rs,Rd	R:W NEXT								
OR.L #xx:32,ERd	R:W 2nd	R:W 3rd	R:W NEXT						
OR.L ERs,ERd	R:W 2nd	R:W NEXT							
ORC #xx:8,CCR	R:W NEXT								
ORC #xx:8,EXR	R:W 2nd	R:W NEXT							
POP.W Rn	R:W NEXT	Internal operation, 1 state	R:W EA						
POP.L ERn	R:W 2nd	R:W:M NEXT	Internal operation, 1 state	R:W:M EA	R:W EA+2				
PUSH.W Rn	R:W NEXT	Internal operation, 1 state	W:W EA						
PUSH.L ERn	R:W 2nd	R:W:M NEXT	Internal operation, 1 state	W:W:M EA	W:W EA+2				
ROTL.B Rd	R:W NEXT								
ROTL.B #2,Rd	R:W NEXT								
ROTL.W Rd	R:W NEXT								
ROTL.W #2,Rd	R:W NEXT								
ROTL.L ERd	R:W NEXT								
ROTL.L #2,ERd	R:W NEXT								
ROTR.B Rd	R:W NEXT								
ROTR.B #2,Rd	R:W NEXT								
ROTR.W Rd	R:W NEXT								
ROTR.W #2,Rd	R:W NEXT								
ROTR.L ERd	R:W NEXT								
ROTR.L #2,ERd	R:W NEXT								
ROTXL.B Rd	R:W NEXT								
ROTXL.B #2,Rd	R:W NEXT								
ROTXL.W Rd	R:W NEXT								
ROTXL.W #2,Rd	R:W NEXT								
ROTXL.L ERd	R:W NEXT								



Instr	ruction	1	2	3	4	5	6	7	8	9
ROTXL.L	_#2,ERd	R:W NEXT								
ROTXR.	B Rd	R:W NEXT								
ROTXR.I	B #2,Rd	R:W NEXT								
ROTXR.	W Rd	R:W NEXT								
ROTXR.	W #2,Rd	R:W NEXT								
ROTXR.	L ERd	R:W NEXT								
ROTXR.	L #2,ERd	R:W NEXT								
RTE		R:W NEXT	R:W Stack (EXR)	R:W Stack (H)	R:W Stack (L)	Internal operation, 1 state	R:W*4			
RTS	Advanced	R:W NEXT	R:W:M Stack (H)	R:W Stack (L)	Internal operation, 1 state	R:W*4				
SHAL.B	Rd	R:W NEXT								
SHAL.B	#2,Rd	R:W NEXT								
SHAL.W	Rd	R:W NEXT								
SHAL.W	#2,Rd	R:W NEXT								
SHAL.L I	ERd	R:W NEXT								
SHAL.L #	#2,ERd	R:W NEXT								
SHAR.B	Rd	R:W NEXT								
SHAR.B	#2,Rd	R:W NEXT								
SHAR.W	' Rd	R:W NEXT								
SHAR.W	#2,Rd	R:W NEXT								
SHAR.L	ERd	R:W NEXT								
SHAR.L	#2,ERd	R:W NEXT								
SHLL.B I	Rd	R:W NEXT								
SHLL.B #	#2,Rd	R:W NEXT								
SHLL.W	Rd	R:W NEXT								
SHLL.W	#2,Rd	R:W NEXT								
SHLL.L E	ERd	R:W NEXT								
SHLL.L#	#2,ERd	R:W NEXT								
SHLR.B	Rd	R:W NEXT								
SHLR.B	#2,Rd	R:W NEXT								
SHLR.W	Rd	R:W NEXT								
SHLR.W	#2,Rd	R:W NEXT								
SHLR.L I	ERd	R:W NEXT								
SHLR.L	#2,ERd	R:W NEXT								

SLEEP R					5	6	7	8	9
SLEEP	R:W NEXT	Internal operation :M							
STC CCR,Rd R	R:W NEXT								
STC EXR,Rd R	R:W NEXT								
STC CCR,@ERd R	R:W 2nd	R:W NEXT	W:W EA						
STC EXR,@ERd R	R:W 2nd	R:W NEXT	W:W EA						
STC CCR, @(d:16,ERd)	R:W 2nd	R:W 3rd	R:W NEXT	W:W EA					
STC EXR, @(d:16,ERd)	R:W 2nd	R:W 3rd	R:W NEXT	W:W EA					
STC CCR, @(d:32,ERd)	R:W 2nd	R:W 3rd	R:W 4th	R:W 5th	R:W NEXT	W:W EA			
STC EXR, @(d:32,ERd)	R:W 2nd	R:W 3rd	R:W 4th	R:W 5th	R:W NEXT	W:W EA			
STC CCR,@-ERd R	R:W 2nd	R:W NEXT	Internal operation, 1 state	W:W EA					
STC EXR,@-ERd R	R:W 2nd	R:W NEXT	Internal operation, 1 state	W:W EA					
STC CCR,@aa:16 R	R:W 2nd	R:W 3rd	R:W NEXT	W:W EA					
STC EXR,@aa:16 R	R:W 2nd	R:W 3rd	R:W NEXT	W:W EA					
STC CCR,@aa:32 R	R:W 2nd	R:W 3rd	R:W 4th	R:W NEXT	W:W EA				
STC EXR,@aa:32 R	R:W 2nd	R:W 3rd	R:W 4th	R:W NEXT	W:W EA				
STM.L (ERn-ERn+1), @-SP*9	R:W 2nd	R:W:M NEXT	Internal operation, 1 state	W:W:M Stack (H)	W:W Stack (L)				
STM.L (ERn-ERn+2), @-SP*9	R:W 2nd	R:W:M NEXT	Internal operation, 1 state	W:W:M Stack (H)	W:W Stack (L)				
STM.L (ERn-ERn+3), @-SP*9	R:W 2nd	R:W:M NEXT	Internal operation, 1 state	W:W:M Stack (H)	W:W Stack (L)				
STMAC MACH,ERd C	Cannot be u	used in this	LSI						
STMAC MACL,ERd									
SUB.B Rs,Rd R	R:W NEXT								
SUB.W #xx:16,Rd R	R:W 2nd	R:W NEXT							
SUB.W Rs,Rd R	R:W NEXT								
SUB.L #xx:32,ERd R	R:W 2nd	R:W 3rd	R:W NEXT						
SUB.L ERs,ERd R	R:W NEXT								





Instr	uction	1	2	3	4	5	6	7	8	9
SUBS #1	/2/4,ERd	R:W NEXT								
SUBX #x	x:8,Rd	R:W NEXT								
SUBX Rs	,Rd	R:W NEXT								
TAS @E	ERd*8	R:W 2nd	R:W NEXT	R:B:M EA	W:B EA					
TRAPA #x:2	Advanced	R:W NEXT	Internal operation, 1 state	W:W Stack (L)	W:W Stack (H)	W:W Stack (EXR)	R:W:M VEC	R:W VEC+2	Internal operation, 1 state	R:W*7
XOR.B #	xx8,Rd	R:W NEXT								
XOR.B R	s,Rd	R:W NEXT								
XOR.W#	xx:16,Rd	R:W 2nd	R:W NEXT							
XOR.W F	Rs,Rd	R:W NEXT								
XOR.L #x	x:32,ERd	R:W 2nd	R:W 3rd	R:W NEXT						
XOR.L E	Rs,ERd	R:W 2nd	R:W NEXT							
XORC #x	x:8,CCR	R:W NEXT								
XORC #x	x:8,EXR	R:W 2nd	R:W NEXT							
Reset excep- tion handling	Advanced	R:W:M VEC	R:W VEC+2	Internal operation, 1 state	R:W*5					
Interrupt excep- tion handling	Advanced	R:W*6	Internal operation, 1 state	W:W Stack (L)	W:W Stack (H)	W:W Stack (EXR)	R:W:M VEC	R:W VEC+2	Internal operation, 1 state	R:W*7

Notes: 1. EAs is the contents of ER5. EAd is the contents of ER6.

- 2. EAs is the contents of ER5. EAd is the contents of ER6. Both registers are incremented by 1 after execution of the instruction. n is the initial value of R4L or R4. If n = 0, these bus cycles are not executed.
- Repeated two times to save or restore two registers, three times for three registers, or four times for four registers.
- 4. Start address after return.
- 5. Start address of the program.
- Prefetch address, equal to two plus the PC value pushed onto the stack. In recovery from sleep mode or software standby mode the read operation is replaced by an internal operation.
- 7. Start address of the interrupt-handling routine.
- 8. Only register ER0, ER1, ER4, or ER5 should be used when using the TAS instruction.
- 9. Only registers ER0 to ER6 should be used when using the STM/LDM instruction.

Appendix B Internal I/O Registers

B.1 Addresses

Address	Register Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Module Name	Bus Width
H'EC00	MRA	SM1	SM0	DM1	DM0	MD1	MD0	DTS	Sz	DTC	16/32*
to H'EFFF	SAR										
	MRB	CHNE	DISEL	_	_	_	_	_	_		
	DAR										
	CRA										
	CRB										
H'FE80	HICR2	_	_	_	_	_	IBFIE4	IBFIE3	_	HIF	8
H'FE84	IDR3	IDR7	IDR6	IDR5	IDR4	IDR3	IDR2	IDR1	IDR0	<u> </u>	
H'FE85	ODR3	ODR7	ODR6	ODR5	ODR4	ODR3	ODR2	ODR1	ODR0	_	
H'FE86	STR3	DBU	DBU	DBU	DBU	C/D	DBU	IBF	OBF		
H'FE8C	IDR4	IDR7	IDR6	IDR5	IDR4	IDR3	IDR2	IDR1	IDR0		
H'FE8D	ODR4	ODR7	ODR6	ODR5	ODR4	ODR3	ODR2	ODR1	ODR0	_	
H'FE8E	STR4	DBU	DBU	DBU	DBU	C/D	DBU	IBF	OBF	_	
H'FED8	KBCRH0	KBIOE	KCLKI	KDI	KBFSEL	KBIE	KBF	PER	KBS	Keyboard	8
H'FED9	KBCRL0	KBE	KCLKO	KDO	_	RXCR3	RXCR2	RXCR1	RXCR0	buffer controller	
H'FEDA	KBBR0	KB7	KB6	KB5	KB4	KB3	KB2	KB1	KB0		
H'FEDC	KBCRH1	KBIOE	KCLKI	KDI	KBFSEL	KBIE	KBF	PER	KBS		
H'FEDD	KBCRL1	KBE	KCLKO	KDO	_	RXCR3	RXCR2	RXCR1	RXCR0		
H'FEDE	KBBR1	KB7	KB6	KB5	KB4	KB3	KB2	KB1	KB0		
H'FEE0	KBCRH2	KBIOE	KCLKI	KDI	KBFSEL	KBIE	KBF	PER	KBS		
H'FEE1	KBCRL2	KBE	KCLKO	KDO	_	RXCR3	RXCR2	RXCR1	RXCR0		
H'FEE2	KBBR2	KB7	KB6	KB5	KB4	KB3	KB2	KB1	KB0		
H'FEE4	KBCOMP	IrE	IrCKS2	IrCKS1	IrCKS0	KBADE	KBCH2	KBCH1	KBCH0	IrDA/ expansion A/D	8
H'FEE6	DDCSWR	SWE	SW	IE	IF	CLR3	CLR2	CLR1	CLR0	IIC0	8

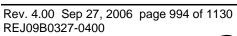
Address	Register Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Module Name	Bus Width
H'FEE8	ICRA	ICR7	ICR6	ICR5	ICR4	ICR3	ICR2	ICR1	ICR0	Interrupt	8
H'FEE9	ICRB	ICR7	ICR6	ICR5	ICR4	ICR3	ICR2	ICR1	ICR0	controller	
H'FEEA	ICRC	ICR7	ICR6	ICR5	ICR4	ICR3	ICR2	ICR1	ICR0		
H'FEEB	ISR	IRQ7F	IRQ6F	IRQ5F	IRQ4F	IRQ3F	IRQ2F	IRQ1F	IRQ0F		
H'FEEC	ISCRH	IRQ7SCB	IRQ7SCA	IRQ6SCB	IRQ6SCA	IRQ5SCB	IRQ5SCA	IRQ4SCB	IRQ4SCA		
H'FEED	ISCRL	IRQ3SCB	IRQ3SCA	IRQ2SCB	IRQ2SCA	IRQ1SCB	IRQ1SCA	IRQ0SCB	IRQ0SCA		
H'FEEE	DTCERA	DTCEA7	DTCEA6	DTCEA5	DTCEA4	DTCEA3	DTCEA2	DTCEA1	DTCEA0	DTC	8
H'FEEF	DTCERB	DTCEB7	DTCEB6	DTCEB5	DTCEB4	DTCEB3	DTCEB2	DTCEB1	DTCEB0		
H'FEF0	DTCERC	DTCEC7	DTCEC6	DTCEC5	DTCEC4	DTCEC3	DTCEC2	DTCEC1	DTCEC0		
H'FEF1	DTCERD	DTCED7	DTCED6	DTCED5	DTCED4	DTCED3	DTCED2	DTCED1	DTCED0		
H'FEF2	DTCERE	DTCEE7	DTCEE6	DTCEE5	DTCEE4	DTCEE3	DTCEE2	DTCEE1	DTCEE0		
H'FEF3	DTVECR	SWDTE	DTVEC6	DTVEC5	DTVEC4	DTVEC3	DTVEC2	DTVEC1	DTVEC0		
H'FEF4	ABRKCR	CMF	_	_	_	_	_	_	BIE	Interrupt	8
H'FEF5	BARA	A23	A22	A21	A20	A19	A18	A17	A16	controller	
H'FEF6	BARB	A15	A14	A13	A12	A11	A10	A9	A8		
H'FEF7	BARC	A7	A6	A5	A4	A3	A2	A1	_		
H'FF80	FLMCR1	FWE	SWE	_	_	EV	PV	E	Р	FLASH	8
H'FF81	FLMCR2	FLER	_	_	_	_	_	ESU	PSU		
H'FF82	PCSR	_	_	_	_	_	PWCKB	PWCKA	_	PWM	8
	EBR1	_	_	_	_	_	_	EB9/—	EB8/—	FLASH	8
H'FF83	SYSCR2	KWUL1	KWUL0	P6PUE	_	SDE	CS4E	CS3E	HI12E	HIF	8
	EBR2	EB7	EB6	EB5	EB4	EB3	EB2	EB1	EB0	FLASH	8
H'FF84	SBYCR	SSBY	STS2	STS1	STS0	_	SCK2	SCK1	SCK0	SYSTEM	8
H'FF85	LPWRCR	DTON	LSON	NESEL	EXCLE	_	_	_	_		
H'FF86	MSTPCRH	MSTP15	MSTP14	MSTP13	MSTP12	MSTP11	MSTP10	MSTP9	MSTP8		
H'FF87	MSTPCRL	MSTP7	MSTP6	MSTP5	MSTP4	MSTP3	MSTP2	MSTP1	MSTP0		
H'FF88	SMR1	C/A	CHR	PE	O/Ē	STOP	MP	CKS1	CKS0	SCI1	8
	ICCR1	ICE	IEIC	MST	TRS	ACKE	BBSY	IRIC	SCP	IIC1	- '
H'FF89	BRR1									SCI1	8
	ICSR1	ESTP	STOP	IRTR	AASX	AL	AAS	ADZ	ACKB	IIC1	
H'FF8A	SCR1	TIE	RIE	TE	RE	MPIE	TEIE	CKE1	CKE0	SCI1	8
H'FF8B	TDR1			<u></u>		<u></u>					
H'FF8C	SSR1	TDRE	RDRF	ORER	FER	PER	TEND	MPB	MPBT		
H'FF8D	RDR1										

Address	Register Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Module Name	Bus Width
H'FF8E	SCMR1	_	_	_	_	SDIR	SINV	_	SMIF	SCI1	8
	ICDR1	ICDR7	ICDR6	ICDR5	ICDR4	ICDR3	ICDR2	ICDR1	ICDR0	IIC1	8
	SARX1	SVAX6	SVAX5	SVAX4	SVAX3	SVAX2	SVAX1	SVAX0	FSX	:	
H'FF8F	ICMR1	MLS	WAIT	CKS2	CKS1	CKS0	BC2	BC1	BC0		
	SAR1	SVA6	SVA5	SVA4	SVA3	SVA2	SVA1	SVA0	FS	:	
H'FF90	TIER	ICIAE	ICIBE	ICICE	ICIDE	OCIAE	OCIBE	OVIE	_	FRT	16
H'FF91	TCSR	ICFA	ICFB	ICFC	ICFD	OCFA	OCFB	OVF	CCLRA		
H'FF92	FRCH									•	
H'FF93	FRCL										
H'FF94	OCRAH										
	OCRBH									•	
H'FF95	OCRAL										
	OCRBL									•	
H'FF96	TCR	IEDGA	IEDGB	IEDGC	IEDGD	BUFEA	BUFEB	CKS1	CKS0		
H'FF97	TOCR	ICRDMS	OCRAMS	ICRS	OCRS	OEA	OEB	OLVLA	OLVLB	•	
H'FF98	ICRAH										
	OCRARH										
H'FF99	ICRAL										
	OCRARL										
H'FF9A	ICRBH									•	
	OCRAFH									•	
H'FF9B	ICRBL									•	
	OCRAFL									•	
H'FF9C	ICRCH									•	
	OCRDMH	0	0	0	0	0	0	0	0	•	
H'FF9D	ICRCL									•	
	OCRDML									•	
H'FF9E	ICRDH										
H'FF9F	ICRDL										
H'FFA0	SMR2	C/Ā	CHR	PE	O/Ē	STOP	MP	CKS1	CKS0	SCI2	8
	DADRAH	DA13	DA12	DA11	DA10	DA9	DA8	DA7	DA6	PWMX	_
	DACR	TEST	PWME	_	_	OEB	OEA	os	CKS	•	
H'FFA1	BRR2	-			-	-		-		SCI2	8
	DADRAL	DA5	DA4	DA3	DA2	DA1	DA0	CFS	_	PWMX	•



Address	Register Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Module Name	Bus Width
H'FFA2	SCR2	TIE	RIE	TE	RE	MPIE	TEIE	CKE1	CKE0	SCI2	8
H'FFA3	TDR2										
H'FFA4	SSR2	TDRE	RDRF	ORER	FER	PER	TEND	MPB	MPBT		
H'FFA5	RDR2										
H'FFA6	SCMR2	_	_	_	_	SDIR	SINV	_	SMIF		
	DADRBH	DA13	DA12	DA11	DA10	DA9	DA8	DA7	DA6	PWMX	8
	DACNTH										
H'FFA7	DADRBL	DA5	DA4	DA3	DA2	DA1	DA0	CFS	REGS		
	DACNTL							_	REGS		
H'FFA8	TCSR0	OVF	WT/ĪT	TME	RSTS	RST/NMI	CKS2	CKS1	CKS0	WDT0	16
	TCNT0 (write)										
H'FFA9	TCNT0 (read)										
H'FFAA	PAODR	PA7ODR	PA6ODR	PA5ODR	PA4ODR	PA3ODR	PA2ODR	PA10DR	PA0ODR	Ports	8
H'FFAB	PAPIN (read)	PA7PIN	PA6PIN	PA5PIN	PA4PIN	PA3PIN	PA2PIN	PA1PIN	PA0PIN		
	PADDR (write)	PA7DDR	PA6DDR	PA5DDR	PA4DDR	PA3DDR	PA2DDR	PA1DDR	PA0DDR		
H'FFAC	P1PCR	P17PCR	P16PCR	P15PCR	P14PCR	P13PCR	P12PCR	P11PCR	P10PCR		
H'FFAD	P2PCR	P27PCR	P26PCR	P25PCR	P24PCR	P23PCR	P22PCR	P21PCR	P20PCR		
H'FFAE	P3PCR	P37PCR	P36PCR	P35PCR	P34PCR	P33PCR	P32PCR	P31PCR	P30PCR		
H'FFB0	P1DDR	P17DDR	P16DDR	P15DDR	P14DDR	P13DDR	P12DDR	P11DDR	P10DDR		
H'FFB1	P2DDR	P27DDR	P26DDR	P25DDR	P24DDR	P23DDR	P22DDR	P21DDR	P20DDR		
H'FFB2	P1DR	P17DR	P16DR	P15DR	P14DR	P13DR	P12DR	P11DR	P10DR		
H'FFB3	P2DR	P27DR	P26DR	P25DR	P24DR	P23DR	P22DR	P21DR	P20DR		
H'FFB4	P3DDR	P37DDR	P36DDR	P35DDR	P34DDR	P33DDR	P32DDR	P31DDR	P30DDR		
H'FFB5	P4DDR	P47DDR	P46DDR	P45DDR	P44DDR	P43DDR	P42DDR	P41DDR	P40DDR		
H'FFB6	P3DR	P37DR	P36DR	P35DR	P34DR	P33DR	P32DR	P31DR	P30DR		
H'FFB7	P4DR	P47DR	P46DR	P45DR	P44DR	P43DR	P42DR	P41DR	P40DR		
H'FFB8	P5DDR	_	_	_	_	_	P52DDR	P51DDR	P50DDR		
H'FFB9	P6DDR	P67DDR	P66DDR	P65DDR	P64DDR	P63DDR	P62DDR	P61DDR	P60DDR		
H'FFBA	P5DR	_	_	_	_	_	P52DR	P51DR	P50DR		
H'FFBB	P6DR	P67DR	P66DR	P65DR	P64DR	P63DR	P62DR	P61DR	P60DR		
H'FFBC	PBODR	PB7ODR	PB6ODR	PB5ODR	PB4ODR	PB3ODR	PB2ODR	PB1ODR	PB0ODR		
H'FFBD	PBPIN (read)	PB7PIN	PB6PIN	PB5PIN	PB4PIN	PB3PIN	PB2PIN	PB1PIN	PB0PIN		
	P8DDR (write)	_	P86DDR	P85DDR	P84DDR	P83DDR	P82DDR	P81DDR	P80DDR		

Address	Register Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Module Name	Bus Width
H'FFBE	P7PIN (read)	P77PIN	P76PIN	P75PIN	P74PIN	P73PIN	P72PIN	P71PIN	P70PIN	Ports	8
	PBDDR (write)	PB7DDR	PB6DDR	PB5DDR	PB4DDR	PB3DDR	PB2DDR	PB1DDR	PB0DDR		
H'FFBF	P8DR	_	P86DR	P85DR	P84DR	P83DR	P82DR	P81DR	P80DR		
H'FFC0	P9DDR	P97DDR	P96DDR	P95DDR	P94DDR	P93DDR	P92DDR	P91DDR	P90DDR		
H'FFC1	P9DR	P97DR	P96DR	P95DR	P94DR	P93DR	P92DR	P91DR	P90DR		
H'FFC2	IER	IRQ7E	IRQ6E	IRQ5E	IRQ4E	IRQ3E	IRQ2E	IRQ1E	IRQ0E	Interrupt controller	8
H'FFC3	STCR	IICS	IICX1	IICX0	IICE	FLSHE	_	ICKS1	ICKS0	System	8
H'FFC4	SYSCR	CS2E	IOSE	INTM1	INTM0	XRST	NMIEG	HIE	RAME		
H'FFC5	MDCR	EXPE	_	_	_	_	_	MDS1	MDS0		
H'FFC6	BCR	ICIS1	ICIS0	BRSTRM	BRSTS1	BRSTS0	_	IOS1	IOS0	Bus	8
H'FFC7	WSCR	RAMS	RAM0	ABW	AST	WMS1	WMS0	WC1	WC0	controller	
H'FFC8	TCR0	CMIEB	CMIEA	OVIE	CCLR1	CCLR0	CKS2	CKS1	CKS0	TMR0,	16
H'FFC9	TCR1	CMIEB	CMIEA	OVIE	CCLR1	CCLR0	CKS2	CKS1	CKS0	TMR1	
H'FFCA	TCSR0	CMFB	CMFA	OVF	ADTE	OS3	OS2	OS1	OS0		
H'FFCB	TCSR1	CMFB	CMFA	OVF	_	OS3	OS2	OS1	OS0		
H'FFCC	TCORA0										
H'FFCD	TCORA1										
H'FFCE	TCORB0								_		
H'FFCF	TCORB1										
H'FFD0	TCNT0										
H'FFD1	TCNT1								_		
H'FFD2	PWOERB	OE15	OE14	OE13	OE12	OE11	OE10	OE9	OE8	PWM	8
H'FFD3	PWOERA	OE7	OE6	OE5	OE4	OE3	OE2	OE1	OE0		
H'FFD4	PWDPRB	OS15	OS14	OS13	OS12	OS11	OS10	OS9	OS8		
H'FFD5	PWDPRA	OS7	OS6	OS5	OS4	OS3	OS2	OS1	OS0		
H'FFD6	PWSL	PWCKE	PWCKS	_	_	RS3	RS2	RS1	RS0		
H'FFD7	PWDR0 to PWDR15										
H'FFD8	SMR0	C/Ā	CHR	PE	O/E	STOP	MP	CKS1	CKS0	SCI0	8
	ICCR0	ICE	IEIC	MST	TRS	ACKE	BBSY	IRIC	SCP	IIC0	_
H'FFD9	BRR0									SCI0	_
	ICSR0	ESTP	STOP	IRTR	AASX	AL	AAS	ADZ	ACKB	IIC0	





Address	Register Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Module Name	Bus Width
H'FFDA	SCR0	TIE	RIE	TE	RE	MPIE	TEIE	CKE1	CKE0	SCI0	8
H'FFDB	TDR0										
H'FFDC	SSR0	TDRE	RDRF	ORER	FER	PER	TEND	MPB	MPBT		
H'FFDD	RDR0										
H'FFDE	SCMR0	_	_	_	_	SDIR	SINV	_	SMIF		
	ICDR0	ICDR7	ICDR6	ICDR5	ICDR4	ICDR3	ICDR2	ICDR1	ICDR0	IIC0	_
	SARX0	SVAX6	SVAX5	SVAX4	SVAX3	SVAX2	SVAX1	SVAX0	FSX		
H'FFDF	ICMR0	MLS	WAIT	CKS2	CKS1	CKS0	BC2	BC1	BC0		
	SAR0	SVA6	SVA5	SVA4	SVA3	SVA2	SVA1	SVA0	FS		
H'FFE0	ADDRAH	AD9	AD8	AD7	AD6	AD5	AD4	AD3	AD2	A/D	8
H'FFE1	ADDRAL	AD1	AD0	_	_	_	_	_	_		
H'FFE2	ADDRBH	AD9	AD8	AD7	AD6	AD5	AD4	AD3	AD2		
H'FFE3	ADDRBL	AD1	AD0	_	_	_	_	_	_		
H'FFE4	ADDRCH	AD9	AD8	AD7	AD6	AD5	AD4	AD3	AD2		
H'FFE5	ADDRCL	AD1	AD0	_	_	_	_	_	_		
H'FFE6	ADDRDH	AD9	AD8	AD7	AD6	AD5	AD4	AD3	AD2		
H'FFE7	ADDRDL	AD1	AD0	_	_	_	_	_	_		
H'FFE8	ADCSR	ADF	ADIE	ADST	SCAN	CKS	CH2	CH1	CH0		
H'FFE9	ADCR	TRGS1	TRGS0	_	_	_	_	_	_		
H'FFEA	TCSR1	OVF	WT/IT	TME	PSS	RST/NMI	CKS2	CKS1	CKS0	WDT1	16
	TCNT1 (write)										
H'FFEB	TCNT1 (read)										
H'FFF0	HICR	_	_	_	_	_	IBFIE2	IBFIE1	FGA20E	HIF	8
	TCRX	CMIEB	CMIEA	OVIE	CCLR1	CCLR0	CKS2	CKS1	CKS0	TMRX	_
	TCRY	CMIEB	CMIEA	OVIE	CCLR1	CCLR0	CKS2	CKS1	CKS0	TMRY	
H'FFF1	KMIMR	KMIMR7	KMIMR6	KMIMR5	KMIMR4	KMIMR3	KMIMR2	KMIMR1	KMIMR0	Interrupt controller	8
	TCSRX	CMFB	CMFA	OVF	ICF	OS3	OS2	OS1	OS0	TMRX	
	TCSRY	CMFB	CMFA	OVF	ICIE	OS3	OS2	OS1	OS0	TMRY	_
H'FFF2	KMPCR	KM7PCR	KM6PCR	KM5PCR	KM4PCR	KM3PCR	KM2PCR	KM1PCR	KM0PCR	Ports	_
	TICRR									TMRX	_
	TCORAY									TMRY	_
H'FFF3	KMIMRA	KMIMR15	KMIMR14	KMIMR13	KMIMR12	KMIMR11	KMIMR10	KMIMR9	KMIMR8	Interrupt controller	8
	TICRF									TMRX	_
	TCORBY									TMRY	=

	Register									Module	Bus
Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Name	Width
H'FFF4	IDR1	IDR7	IDR6	IDR5	IDR4	IDR3	IDR2	IDR1	IDR0	HIF	8
	TCNTX									TMRX	-
	TCNTY									TMRY	-
H'FFF5	ODR1	ODR7	ODR6	ODR5	ODR4	ODR3	ODR2	ODR1	ODR0	HIF	-
	TCORC									TMRX	-
	TISR	_	_	_	_	_	_	_	IS	TMRY	-
H'FFF6	STR1	DBU	DBU	DBU	DBU	C/D	DBU	IBF	OBF	HIF	-
	TCORAX									TMRX	=
H'FFF7	TCORBX										
H'FFF8	DADR0									D/A	-
H'FFF9	DADR1										
H'FFFA	DACR	DAOE1	DAOE0	DAE	_	_	_	_	_		
H'FFFC	IDR2	IDR7	IDR6	IDR5	IDR4	IDR3	IDR2	IDR1	IDR0	HIF	-
	TCONRI	SIMOD1	SIMOD0	SCONE	ICST	HFINV	VFINV	HIINV	VIINV	Timer connection	-
H'FFFD	ODR2	ODR7	ODR6	ODR5	ODR4	ODR3	ODR2	ODR1	ODR0	HIF	-
	TCONRO	HOE	VOE	CLOE	CBOE	HOINV	VOINV	CLOINV	CBOINV	Timer connection	-
H'FFFE	STR2	DBU	DBU	DBU	DBU	C/D	DBU	IBF	OBF	HIF	=
	TCONRS	TMRX/Y	ISGENE	HOMOD1	HOMOD0	VOMOD1	VOMOD0	CLMOD1	CLMOD0	Timer	=
H'FFFF	SEDGR	VEDG	HEDG	CEDG	HFEDG	VFEDG	PREQF	IHI	IVI	connection	

B.2 Register Selection Conditions

Lower Address	Register Name	H8S/2148 Group Register Selection Conditions	H8S/2147N Register Selection Conditions	H8S/2144 Group Register Selection Conditions	Module Name
H'EC00	MRA	RAME = 1 in SYSCR	_	_	DTC
to H'EFFF	SAR				
ПЕГГГ	MRB				
	DAR				
	CRA				
	CRB				
H'FE80	HICR2	MSTP2 = 0	MSTP2 = 0	_	HIF
H'FE84	IDR3				
H'FE85	ODR3				
H'FE86	STR3				
H'FE8C	IDR4				
H'FE8D	ODR4				
H'FE8E	STR4				
H'FED8	KBCRH0	MSTP2 = 0	MSTP2 = 0	_	Keyboard
H'FED9	KBCRL0				buffer
H'FEDA	KBBR0				controller
H'FEDC	KBCRH1				
H'FEDD	KBCRL1				
H'FEDE	KBBR1				
H'FEE0	KBCRH2				
H'FEE1	KBCRL2				
H'FEE2	KBBR2				
H'FEE4	KBCOMP	No conditions	No conditions	No conditions	IrDA/ expansion A/D
H'FEE6	DDCSWR	MSTP4 = 0	MSTP4 = 0	_	IIC0
H'FEE8	ICRA	No conditions	No conditions	No conditions	Interrupt
H'FEE9	ICRB				controller
H'FEEA	ICRC				
H'FEEB	ISR	1			
H'FEEC	ISCRH	1			
H'FEED	ISCRL	1			
H'FEEE	DTCERA	No conditions	_	_	DTC
H'FEEF	DTCERB	1			
H'FEF0	DTCERC	1			
H'FEF1	DTCERD				
H'FEF2	DTCERE	1			

Lower Address	Register Name	H8S/2148 Grou Selection Co		H8S/2147N Regis Conditi		H8S/2144 Group Register Selection Conditions	Module Name
H'FEF3	DTVECR	No conditions		_		_	DTC
H'FEF4	ABRKCR	No conditions		No conditions		No conditions	Interrupt
H'FEF5	BARA						controller
H'FEF6	BARB						
H'FEF7	BARC						
H'FF80	FLMCR1	FLSHE = 1 in STCR		FLSHE = 1 in STCR	l	FLSHE = 1 in STCR	Flash
H'FF81	FLMCR2						memory
H'FF82	PCSR	FLSHE = 0 in STCR		FLSHE = 0 in STCR	l	_	PWM
	EBR1	FLSHE = 1 in STCR		FLSHE = 1 in STCR	1	FLSHE = 1 in STCR	Flash memory
H'FF83	SYSCR2	FLSHE = 0 in STCR		FLSHE = 0 in STCR	l	_	HIF
	EBR2	FLSHE = 1 in STCR		FLSHE = 1 in STCR	!	FLSHE = 1 in STCR	Flash memory
H'FF84	SBYCR	FLSHE = 0 in STCR		FLSHE = 0 in STCR	l	FLSHE = 0 in STCR	System
H'FF85	LPWRCR						
H'FF86	MSTPCRH						
H'FF87	MSTPCRL						
H'FF88	SMR1	MSTP6 = 0, IICE = 0) in STCR	MSTP6 = 0, IICE = 0	0 in STCR	MSTP6 = 0, IICE = 0 in STCR	SCI1
	ICCR1	MSTP3 = 0, IICE = 1	I in STCR	MSTP3 = 0, IICE =	1 in STCR	_	IIC1
H'FF89	BRR1	MSTP6 = 0, IICE = 0) in STCR	MSTP6 = 0, IICE = 0	0 in STCR	MSTP6 = 0, IICE = 0 in STCR	SCI1
	ICSR1	MSTP3 = 0, IICE = 1	I in STCR	MSTP3 = 0, IICE =	1 in STCR	_	IIC1
H'FF8A	SCR1	MSTP6 = 0		MSTP6 = 0		MSTP6 = 0	SCI1
H'FF8B	TDR1						
H'FF8C	SSR1						
H'FF8D	RDR1						
H'FF8E	SCMR1	MSTP6 = 0, IICE = 0) in STCR	MSTP6 = 0, IICE = 0	0 in STCR	MSTP6 = 0, IICE = 0 in STCR	
	ICDR1	MSTP3 = 0, IICE = 1 in STCR	ICE = 1 in ICCR1	MSTP3 = 0, IICE = 1 in STCR	ICE = 1 in ICCR1		IIC1
	SARX1		ICE = 0 in ICCR1		ICE = 0 in ICCR1		
H'FF8F	ICMR1	1	ICE = 1 in ICCR1		ICE = 1 in ICCR1		
	SAR1		ICE = 0 in ICCR1		ICE = 0 in ICCR1	1	
H'FF90	TIER	MSTP13 = 0		MSTP13 = 0		MSTP13 = 0	FRT
H'FF91	TCSR	1					
H'FF92	FRCH	1					
H'FF93	FRCL	1					



Lower Address	Register Name	H8S/2148 Grou Selection Co		H8S/2147N Regist		H8S/2144 Grou Selection Co		Module Name
H'FF94	OCRAH	MSTP13 = 0	OCRS = 0 in	MSTP13 = 0	OCRS = 0 in	MSTP13 = 0	OCRS = 0 in	FRT
	OCRBH		OCRS = 1 in		OCRS = 1 in		OCRS = 1 in	
H'FF95	OCRAL		OCRS = 0 in		OCRS = 0 in		OCRS = 0 in	
	OCRBL		OCRS = 1 in		OCRS = 1 in		OCRS = 1 in	-
H'FF96	TCR					-		
H'FF97	TOCR							
H'FF98	ICRAH		ICRS = 0 in TOCR		ICRS = 0 in TOCR		ICRS = 0 in TOCR	
	OCRARH		ICRS = 1 in TOCR		ICRS = 1 in TOCR		ICRS = 1 in TOCR	
H'FF99	ICRAL	_	ICRS = 0 in TOCR		ICRS = 0 in TOCR		ICRS = 0 in TOCR	
	OCRARL	_	ICRS = 1 in TOCR		ICRS = 1 in TOCR		ICRS = 1 in TOCR	
H'FF9A	ICRBH		ICRS = 0 in TOCR		ICRS = 0 in TOCR		ICRS = 0 in TOCR	
	OCRAFH		ICRS = 1 in TOCR		ICRS = 1 in TOCR		ICRS = 1 in TOCR	
H'FF9B	ICRBL		ICRS = 0 in TOCR		ICRS = 0 in TOCR		ICRS = 0 in TOCR	
	OCRAFL		ICRS = 1 in TOCR		ICRS = 1 in TOCR		ICRS = 1 in TOCR	
H'FF9C	ICRCH		ICRS = 0 in TOCR		ICRS = 0 in TOCR		ICRS = 0 in TOCR	
	OCRDMH		ICRS = 1 in TOCR		ICRS = 1 in TOCR		ICRS = 1 in TOCR	
H'FF9D	ICRCL		ICRS = 0 in TOCR		ICRS = 0 in TOCR		ICRS = 0 in TOCR	
	OCRDML		ICRS = 1 in TOCR		ICRS = 1 in TOCR		ICRS = 1 in TOCR	
H'FF9E	ICRDH							
H'FF9F	ICRDL							
H'FFA0	SMR2	MSTP5 = 0, IICE = 0) in STCR	MSTP5 = 0, IICE = 0	in STCR	MSTP5 = 0, IICE = 0	in STCR	SCI2
	DADRAH	MSTP11 = 0, IICE = 1 in STCR	REGS = 0 in DACNT/ DADRB	MSTP11 = 0, IICE = 1 in STCR	REGS = 0 in DACNT/ DADRB	MSTP11 = 0, IICE = 1 in STCR	REGS = 0 in DACNT/ DADRB	PWMX
	DACR		REGS = 1 in DACNT/ DADRB		REGS = 1 in DACNT/ DADRB		REGS = 1 in DACNT/ DADRB	

Lower Address	Register Name	H8S/2148 Group Selection Cor		H8S/2147N Regis Conditi		H8S/2144 Grou Selection Co		Module Name
H'FFA1	BRR2	MSTP5 = 0, IICE = 0	in STCR	MSTP5 = 0, IICE =	0 in STCR	MSTP5 = 0, IICE =	0 in STCR	SCI2
	DADRAL	MSTP11 = 0, IICE = 1 in STCR	REGS = 0 in DACNT/ DADRB	MSTP11 = 0, IICE = 1 in STCR	REGS = 0 in DACNT/ DADRB	MSTP11 = 0, IICE = 1 in STCR	REGS = 0 in DACNT/ DADRB	PWMX
H'FFA2	SCR2	MSTP5 = 0	•	MSTP5 = 0		MSTP5 = 0	·	SCI2
H'FFA3	TDR2							
H'FFA4	SSR2							
H'FFA5	RDR2							
H'FFA6	SCMR2	MSTP5 = 0, IICE = 0	in STCR	MSTP5 = 0, IICE =	0 in STCR	MSTP5 = 0, IICE =	0 in STCR	
	DADRBH	MSTP11 = 0, IICE = 1 in STCR	REGS = 0 in DACNT/ DADRB	MSTP11 = 0, IICE = 1 in STCR	REGS = 0 in DACNT/ DADRB	MSTP11 = 0, IICE = 1 in STCR	REGS = 0 in DACNT/ DADRB	PWMX
	DACNTH		REGS = 1 in DACNT/ DADRB		REGS = 1 in DACNT/ DADRB		REGS = 1 in DACNT/ DADRB	
H'FFA7	DADRBL		REGS = 0 in DACNT/ DADRB		REGS = 0 in DACNT/ DADRB		REGS = 0 in DACNT/ DADRB	PWMX
	DACNTL		REGS = 1 in DACNT/ DADRB		REGS = 1 in DACNT/ DADRB		REGS = 1 in DACNT/ DADRB	
H'FFA8	TCSR0	No conditions		No conditions	•	No conditions		WDT0
	TCNT0 (write)							
H'FFA9	TCNT0 (read)							
H'FFAA	PAODR0	No conditions		No conditions		No conditions		Ports
H'FFAB	PAPIN (read)							
	PADDR (write)							
H'FFAC	P1PCR							
H'FFAD	P2PCR							
H'FFAE	P3PCR							
H'FFB0	P1DDR							
H'FFB1	P2DDR]						
H'FFB2	P1DR							
H'FFB3	P2DR							
H'FFB4	P3DDR							
H'FFB5	P4DDR							
H'FFB6	P3DR							
H'FFB7	P4DR							



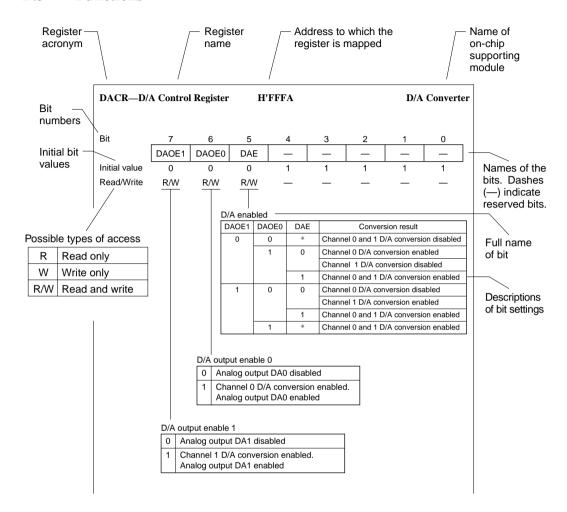
Lower Address	Register Name	H8S/2148 Group Register Selection Conditions	H8S/2147N Register Selection Conditions	H8S/2144 Group Register Selection Conditions	Module Name
H'FFB8	P5DDR	No conditions	No conditions	No conditions	Ports
H'FFB9	P6DDR	-			
H'FFBA	P5DR	-			
H'FFBB	P6DR				
H'FFBC	PBODR	-			
H'FFBD	P8DDR (write)				
	PBPIN (read)				
H'FFBE	P7PIN (read)				
	PBDDR (write)				
H'FFBF	P8DR				
H'FFC0	P9DDR				
H'FFC1	P9DR				
H'FFC2	IER	No conditions	No conditions	No conditions	Interrupt controller
H'FFC3	STCR	No conditions	No conditions	No conditions	System
H'FFC4	SYSCR				
H'FFC5	MDCR				
H'FFC6	BCR				Bus
H'FFC7	WSCR				controller
H'FFC8	TCR0	MSTP12 = 0	MSTP12 = 0	MSTP12 = 0	TMR0,
H'FFC9	TCR1				TMR1
H'FFCA	TCSR0				
H'FFCB	TCSR1				
H'FFCC	TCORA0				
H'FFCD	TCORA1				
H'FFCE	TCORB0				
H'FFCF	TCORB1				
H'FFD0	TCNT0				
H'FFD1	TCNT1				
H'FFD2	PWOERB	No conditions	No conditions	_	PWM
H'FFD3	PWOERA				
H'FFD4	PWDPRB				
H'FFD5	PWDPRA				

Lower Address	Register Name	H8S/2148 Grou Selection Co		H8S/2147N Regis		H8S/2144 Group Register Selection Conditions	Module Name
H'FFD6	PWSL	MSTP11 = 0		MSTP11 = 0		_	PWM
H'FFD7	PWDR0 to PWDR15	-					
H'FFD8	SMR0	MSTP7 = 0, IICE = 0) in STCR	MSTP7 = 0, IICE =	0 in STCR	MSTP7 = 0, IICE = 0 in STCR	SCI0
	ICCR0	MSTP4 = 0, IICE = 1	1 in STCR	MSTP4 = 0, IICE =	1 in STCR	_	IIC0
H'FFD9	BRR0	MSTP7 = 0, IICE = 0) in STCR	MSTP7 = 0, IICE =	0 in STCR	MSTP7 = 0, IICE = 0 in STCR	SCI0
	ICSR0	MSTP4 = 0, IICE = 1	1 in STCR	MSTP4 = 0, IICE =	1 in STCR	_	IIC0
H'FFDA	SCR0	MSTP7 = 0		MSTP7 = 0		MSTP7 = 0	SCI0
H'FFDB	TDR0						
H'FFDC	SSR0						
H'FFDD	RDR0						
H'FFDE	SCMR0	MSTP7 = 0, IICE = 0) in STCR	MSTP7 = 0, IICE = 0 in STCR		MSTP7 = 0, IICE = 0 in STCR	
	ICDR0	MSTP4 = 0, IICE = 1 in STCR	ICE = 1 in ICCR0	MSTP4 = 0, IICE = 1 in STCR	ICE = 1 in ICCR0	_	IIC0
	SARX0		ICE = 0 in ICCR0		ICE = 0 in ICCR0		
H'FFDF	ICMR0		ICE = 1 in ICCR0		ICE = 1 in ICCR0	-	
	SAR0	-	ICE = 0 in		ICE = 0 in	-	
H'FFE0	ADDRAH	MSTP9 = 0	1	MSTP9 = 0		MSTP9 = 0	A/D
H'FFE1	ADDRAL	-					
H'FFE2	ADDRBH	-					
H'FFE3	ADDRBL						
H'FFE4	ADDRCH	-					
H'FFE5	ADDRCL						
H'FFE6	ADDRDH						
H'FFE7	ADDRDL						
H'FFE8	ADCSR						
H'FFE9	ADCR	1					
H'FFEA	TCSR1	No conditions		No conditions		No conditions	WDT1
	TCNT1 (write)						
H'FFEB	TCNT1 (read)						
H'FFF0	HICR	MSTP2 = 0, HIE = 1	in SYSCR	MSTP2 = 0, HIE = 1	I in SYSCR	_	HIF
	TCRX	MSTP8 = 0, HIE = 0 in SYSCR	TMRX/Y = 0 in TCONRS	_		_	TMRX
	TCRY		TMRX/Y = 1 in TCONRS	MSTP8 = 0, HIE = 0) in SYSCR	MSTP8 = 0, HIE = 0 in SYSCR	TMRY



Lower Address	Register Name	H8S/2148 Grou Selection Co		H8S/2147N Register Selection Conditions	H8S/2144 Group Register Selection Conditions	Module Name
H'FFF1	KMIMR	MSTP2 = 0, HIE = 1	in SYSCR	MSTP2 = 0, HIE = 1 in SYSCR	MSTP2 = 0, HIE = 1 in SYSCR	Interrupt controller
	TCSRX	MSTP8 = 0, HIE = 0 in SYSCR	TMRX/Y = 0 in TCONRS	_	_	TMRX
	TCSRY		TMRX/Y = 1 in TCONRS	MSTP8 = 0, HIE = 0 in SYSCR	MSTP8 = 0, HIE = 0 in SYSCR	TMRY
H'FFF2	KMPCR	MSTP2 = 0, HIE = 1	in SYSCR	MSTP2 = 0, HIE = 1 in SYSCR	MSTP2 = 0, HIE = 1 in SYSCR	Ports
	TICRR	MSTP8 = 0, HIE = 0 in SYSCR	TMRX/Y = 0 in TCONRS	_	_	TMRX
TCORAY		TMRX/Y = 1 in TCONRS	MSTP8 = 0, HIE = 0 in SYSCR	MSTP8 = 0, HIE = 0 in SYSCR	TMRY	
H'FFF3	KMIMRA	MSTP2 = 0, HIE = 1	in SYSCR	MSTP2 = 0, HIE = 1 in SYSCR	MSTP2 = 0, HIE = 1 in SYSCR	Interrupt controller
	TICRF	MSTP8 = 0, TMRX/Y = 0 in TCONRS		_	_	TMRX
	TCORBY	TMRX/Y = 1 in TCONRS		MSTP8 = 0, HIE = 0 in SYSCR	MSTP8 = 0, HIE = 0 in SYSCR	TMRY
H'FFF4	IDR1	MSTP2 = 0, HIE = 1 in SYSCR		MSTP2 = 0, HIE = 1 in SYSCR	_	HIF
	TCNTX	MSTP8 = 0, HIE = 0 in SYSCR	TMRX/Y = 0 in TCONRS	_	-	TMRX
	TCNTY		TMRX/Y = 1 in TCONRS	MSTP8 = 0, HIE = 0 in SYSCR	MSTP8 = 0, HIE = 0 in SYSCR	TMRY
H'FFF5	FFF5 ODR1 MSTP	MSTP2 = 0, HIE = 1	in SYSCR	MSTP2 = 0, HIE = 1 in SYSCR	_	HIF
	TCORC	MSTP8 = 0, HIE = 0 in SYSCR	TMRX/Y = 0 in TCONRS	_	-	TMRX
	TISR		TMRX/Y = 1 in TCONRS	MSTP8 = 0, HIE = 0 in SYSCR	MSTP8 = 0, HIE = 0 in SYSCR	TMRY
H'FFF6	STR1	MSTP2 = 0, HIE = 1	in SYSCR	MSTP2 = 0, HIE = 1 in SYSCR	_	HIF
	TCORAX	MSTP8 = 0,	TMRX/Y = 0	_		TMRX
H'FFF7	TCORBX	HIE = 0 in SYSCR	in TCONRS			
H'FFF8	DADR0	MSTP10 = 0		MSTP10 = 0	MSTP10 = 0	D/A
H'FFF9	DADR1]				
H'FFFA	DACR					
H'FFFC	IDR2	MSTP2 = 0, HIE = 1	in SYSCR	MSTP2 = 0, HIE = 1 in SYSCR	_	HIF
	TCONRI	MSTP8 = 0, HIE = 0	in SYSCR	_		Timer connection
H'FFFD	ODR2	MSTP2 = 0, HIE = 1	in SYSCR	MSTP2 = 0, HIE = 1 in SYSCR		HIF
	TCONRO	MSTP8 = 0, HIE = 0	in SYSCR	_		Timer connection
H'FFFE	STR2	MSTP2 = 0, HIE = 1	in SYSCR	MSTP2 = 0, HIE = 1 in SYSCR		HIF
	TCONRS	MSTP8 = 0, HIE = 0	in SYSCR	_		Timer
H'FFFF	SEDGR	1				connection

B.3 Functions



MRA—DTC Mode Register A H'EC00-H'EFFF DTC Bit 7 6 5 4 3 2 1 0 SM1 SM0 MD0 DTS DM1 DM0 MD1 Sz Initial value Undefined Undefined Undefined Undefined Undefined Undefined Undefined Read/Write DTC data transfer size Byte-size transfer Word-size transfer DTC transfer mode select Destination side is repeat area or block area Source side is repeat area or block area DTC mode 0 Normal mode Repeat mode Block transfer mode 1 0 1 Destination address mode DAR is fixed DAR is incremented after a transfer (by 1 when Sz = 0; by 2 when Sz = 1) DAR is decremented after a transfer (by 1 when Sz = 0; by 2 when Sz = 1) Source address mode SAR is fixed 1 SAR is incremented after a transfer (by 1 when Sz = 0; by 2 when Sz = 1)

SAR is decremented after a transfer (by 1 when Sz = 0; by 2 when Sz = 1)

MRB—DTC M	Iode Regi	ster B			H'EC00-H'EFFF					
Bit	7	6	5	4	3	2	1	0		
	CHNE	DISEL	_	_	_	_	_	_		
Initial value	Undefined	Undefined	Undefined	Undefined	Undefined	Undefined	Undefined	Undefined		
Read/Write		0 A di	nterrupt se fter a data isabled un fter a data nabled	transfer e less the tra	ansfer cou	nter is 0		_		
	DTC chain transfer enable									
	0 Er									
	1 D									

SAR—DTC Se	ource	Add	ress F	Regist	er	H'EC00-H'EFFF					DTC		
Bit	23	22	21	20	19		4	3	2	1	0		
Initial value	Unde- fined	Unde- fined	Unde- fined	Unde- fined	Unde- fined		Unde- fined	Unde- fined	Unde- fined	Unde- fined	Unde- fined		
Read/Write		_	_	_	_		_	_	_	_			

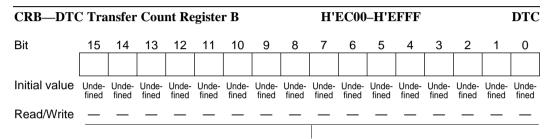
Specifies DTC transfer data source address

DAR—DTC D	estina	ation	Addı	ress I	Regist	er H'EC00–H	H'EC00-H'EFFF				
Bit	23	22	21	20	19		4	3	2	1	0
Initial value	Unde- fined	Unde- fined	Unde- fined	Unde- fined	Unde- fined		Unde- fined	Unde- fined	Unde- fined	Unde- fined	Unde- fined
Read/Write		_	_	_	_		_	_	_	_	

Specifies DTC transfer data destination address

CRA—DTC Transfer Count Register A								H'EC00-H'EFFF					DTC			
Bit	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Initial value	Unde- fined	Unde- fined	Unde- fined	Unde- fined	Unde- fined	Unde- fined	Unde- fined	Unde- fined	Unde- fined	Unde- fined	Unde- fined	Unde- fined	Unde- fined	Unde- fined	Unde- fined	Unde- fined
Read/Write	_	_	_	_	_	_	_	_	_	_	_	_	_	_	_	_
	← CRAH —							-	—			CR	RAL			
									I							

Specifies the number of DTC data transfers



Specifies the number of DTC block data transfers

HICR2—Host	Interface	Control F	Register 2		H'FE80			I	HIF
Bit	7	6	5	4	3	2	1	0	
	_	_	_	_	_	IBFIE4	IBFIE3	_	
Initial value	1	1	1	1	1	0	0	0	
Slave R/W	_	_	_	_	_	R/W	R/W	_	
Host R/W	_	_	_	_	_			_	

Input data register full interrupt enable

IBFIE4	IBFIE3	Description
_	0	Input data register (IDR3) receive complete interrupt is disabled
_	1	Input data register (IDR3) receive complete interrupt is enabled
0	_	Input data register (IDR4) receive complete interrupt is disabled
1	_	Input data register (IDR4) receive complete interrupt is enabled

IDR3—Input D IDR4—Input D	_				HIF HIF				
Bit	7	6	5	4	3	2	1	0	_
	IDR7	IDR6	IDR5	IDR4	IDR3	IDR2	IDR1	IDR0	
Initial value	_	_	_	_	_	_	_	_	-
Slave R/W	R	R	R	R	R	R	R	R	
Host R/W	W	W	W	W	W	W	W	W	

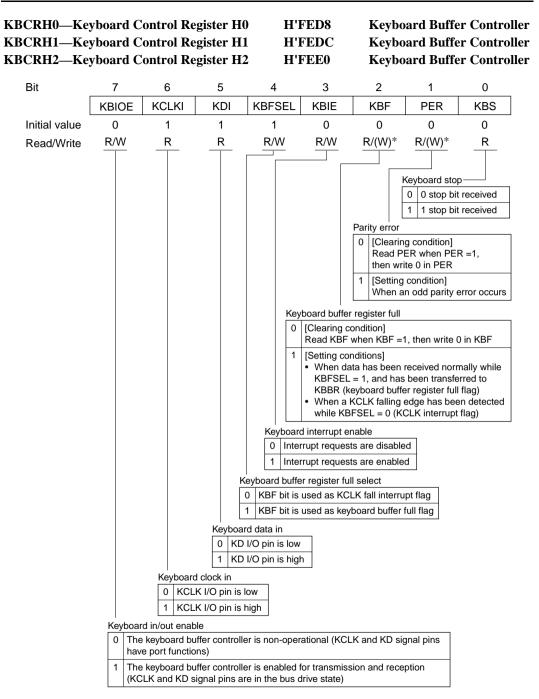
Stores host data bus contents at rise of \overline{IOW} when \overline{CS} is low

ODR3—Outpu ODR4—Outpu		_		H'FE85 H'FE8D				
Bit	7	6	5	4	3	2	1	0
	ODR7	ODR6	ODR5	ODR4	ODR3	ODR2	ODR1	ODR0
Initial value	_	_	_	_	_	_	_	
Slave R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Host R/W	R	R	R	R	R	R	R	R

ODR contents are output to the host data bus when HA0 is low, $\overline{\text{CS}}$ is low, and $\overline{\text{IOR}}$ is low

STR3—Status	_			H'FE86 H'FE8E						
Bit	7	6	5	4		3	2	1	0	
	DBU	DBU	DBU	DBU		C/D	DBU	IBF	OBF	
Initial value	0	0	0	0		0	0	0	0	
Slave R/W	R/W	R/W	R/W	R/W		R	R/W	R	R/(W)	
Host R/W	R	R	R	R		R	R	R	R	
		User-def	ined bits				Whe read write 1 [Sett Whe	offer full aring cond on the hos s ODR or es 0 in the ing condi on the slaves to ODR	t process the slave OBF bit tion]	e
					Inpu	ut buffer	full			
					0		ng conditio		reads ID	R
					1		g condition the host pr		vrites to I	DR
			Con	nmand/da	ıta					
0 Contents of input data register (IDR) are data									data	

Contents of input data register (IDR) are a command



Note: * Only 0 can be written, to clear the flag.

KBCRL0—F KBCRL1—F KBCRL2—F	Keyboard	Control R	egister l	L1	H'FI H'FI H'FI	E D	D	Keyboa	rd Buffer	Controller Controller Controller
Bit	7	6	5	4			3	2	1	0
	KBE	KCLKO	KDO	_	-	RX	CR3	RXCR2	RXCR1	RXCR0
Initial value	0	1	1	1	•		0	0	0	0
Read/Write	R/W	R/W	R/W	_	-		R	R	R	R
				Receive o	counter	Т	RXCR1	RXCR0	Daneitie de	
				0	Receive da	a contents				
				0	0		0	1	Star	t bit
						-	1	0	KE	
								1	KE	31
					1		0	0	KB2	
								1	KE	33
							1	0	KE	34
								1	KE	
				1	0		0	0	KE	36
								1	KE	
							1	0	Parit	y bit
					1			1	-	-
				pard data	out					
								ta I/O pin i		
			1 K	keyboard i	outter c	cont	roller da	ta I/O pin i	s nign	
		 	ما مامماد می							
			d clock ou	fer control	ler cloc	-k 1/	O nin is	low		
				fer control						
							о р	9		
	 Kevboa	rd enable								
		ading of recei	ve data in	to KBBR	is disal	blec	t			
		ading of recei					_			

KBBR0—Keyboard Data Buffer Register 0 KBBR1—Keyboard Data Buffer Register 1 KBBR2—Keyboard Data Buffer Register 2					H'FEDA H'FEDE H'FEE2		Keyboard Buffer Controller Keyboard Buffer Controller Keyboard Buffer Controller		
Bit	7	6	5	4	3	2	1	0	
	KB7	KB6	KB5	KB4	KB3	KB2	KB1	KB0	
Initial value	0	0	0	0	0	0	0	0	
Read/Write	R	R	R	R	R	R	R	R	
				Stores re	ceive data				

Stores receive data

KBCOMP—Kevboard Comparator Control Register H'FEE4 IrDA/Expansion A/D Bit 7 6 5 4 3 2 1 0 IrE IrCKS2 IrCKS1 IrCKS0 **KBADE** KBCH2 KBCH1 KBCH0 0 0 0 0 0 Initial value 0 0 0 Read/Write R/W R/W R/W R/W R/W R/W R/W R/W Keyboard comparator control Bit 3 Bit 2 Bit 1 Bit 0 A/D converter A/D converter channel 7 input channel 6 input **KBADE** KBCH0 KBCH2 KBCH1 AN6 AN7 1 0 0 0 CIN₀ CIN8 1 CIN1 CIN9 1 0 CIN₂ CIN₁₀ 1 CIN3 CIN11 1 0 0 CIN4 CIN12 1 CIN₅ CIN13 1 0 CIN14 CIN₆ 1 CIN7 CIN15 IrDA Clock select 2 to 0 0 $B \times 3/16$ (3/16 of the bit rate) 1 φ/2 1 0 φ/4 1 φ/8 1 0 0 φ/16 1 φ/32 1 0 φ/64 1 φ/128

IrDA enable

	The TxD2/IrTxD and RxD2/IrRxD pins function as TxD2 and RxD2
1	The TxD2/IrTxD and RxD2/IrRxD pins function as IrTxD and IrRxD

IIC0 DDCSWR—DDC Switch Register H'FEE6 Bit 7 6 5 4 3 2 1 0 SWF SW ΙE IF CLR3 CLR2 CLR1 CLR₀ Initial value 0 0 0 0 1 1 1 1 R/W W*2 W*2 Read/Write R/W R/W R/(W)*1 W*2 W*2 IIC clear bits Bit 3 | Bit 2 | Bit 1 Bit 0 Description CLR3 CLR2 CLR1 CLR0 0 Setting prohibited 1 0 0 Setting prohibited 1 IIC0 internal latch cleared 1 IIC1 internal latch cleared 0 1 IIC0 and IIC1 internal latches cleared 1 Invalid setting DDC mode switch interrupt flag No interrupt is requested when automatic format switching is executed [Clearing condition] When 0 is written in IF after reading IF = 1 An interrupt is requested when automatic format switching is executed [Setting condition] When a falling edge is detected on the SCL pin when SWE = 1 DDC mode switch interrupt enable bit Interrupt when automatic format switching is executed is disabled Interrupt when automatic format switching is executed is enabled DDC mode switch IIC channel 0 is used with the I2C bus format [Clearing conditions] · When 0 is written by software • When a falling edge is detected on the SCL pin when SWE = 1 IIC channel 0 is used in formatless mode [Setting condition] When 1 is written in SW after reading SW = 0 DDC mode switch enable Automatic switching of IIC channel 0 from formatless mode to I²C bus format is disabled Automatic switching of IIC channel 0 from formatless mode to I²C bus format is enabled

Notes: 1. Only 0 can be written, to clear the flag.

2. Always read as 1.



ICRA—Interro ICRB—Interro ICRC—Interro	ıpt Contr	ol Registe	er B	H'l	FEE8 FEE9 FEEA		Interrup	t Controller t Controller t Controller
Bit	7	6	5	4	3	2	1	0
	ICR7	ICR6	ICR5	ICR4	ICR3	ICR2	ICR1	ICR0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
	Interrupt control level O Corresponding interrupt source is control level 0 (non-priority) 1 Corresponding interrupt source is control level 1 (priority)							

Correspondence between Interrupt Sources and ICR Settings

Register				В	ts			
Register	7	6	5	4	3	2	1	0
ICRA	IRQ0	IRQ1	IRQ2 IRQ3	IRQ4 IRQ5	IRQ6 IRQ7	DTC	Watchdog timer 0	Watchdog timer 1
ICRB	A/D converter	Free- running timer	_	_	8-bit timer channel 0		8-bit timer channels X, Y	HIF, keyboard buffer controller
ICRC	SCI channel 0	SCI channel 1	SCI channel 2	IIC channel 0 (option)	IIC channel 1 (option)	_	_	_

ISR—IRQ Sta		Η'.	FEEB	Interrupt Controller					
Bit	7	6	5	4	3	2	1	0	
	IRQ7F	IRQ6F	IRQ5F	IRQ4F	IRQ3F	IRQ2F	IRQ1F	IRQ0F	
Initial value	0	0	0	0	0	0	0	0	
Read/Write	R/(W)*	R/(W)*	R/(W)*	R/(W)*	R/(W)*	R/(W)*	R/(W)*	R/(W)*	

IRQ7 to IRQ0 flags

- 0 [Clearing conditions]
 - Cleared by reading IRQnF when set to 1, then writing 0 in IRQnF
 - When interrupt exception handling is executed while low-level detection is set (IRQnSCB = IRQnSCA = 0) and IRQn input is high*
 - When IRQn interrupt exception handling is executed while falling, rising, or both-edge detection is set (IRQnSCB = 1 or IRQnSCA = 1)*
- 1 [Setting conditions]
 - When IRQn input goes low when low-level detection is set (IRQnSCB = IRQnSCA = 0)
 - When a falling edge occurs in IRQn input while falling edge detection is set (IRQnSCB = 0, IRQnSCA = 1)
 - When a rising edge occurs in IRQn input while rising edge detection is set (IRQnSCB = 1, IRQnSCA = 0)
 - When a falling or rising edge occurs in IRQn input while both-edge detection is set (IRQnSCB = IRQnSCA = 1)

Notes: n = 7 to 0

- * When a product, in which a DTC is incorporated, is used in the following settings, the corresponding flag bit is not automatically cleared even when exception handling, which is a clear condition, is executed and the bit is held at 1.
 - (1) When DTCEA3 is set to 1 (ADI is set to an interrupt source), IRQ4F flag is not automatically cleared.
 - (2) When DTCEA2 is set to 1 (ICIA is set to an interrupt source), IRQ5F flag is not automatically cleared.
 - (3) When DTCEA1 is set to 1 (ICIB is set to an interrupt source), IRQ6F flag is not automatically cleared.
 - (4) When DTCEA0 is set to 1 (OCIA is set to an interrupt souce), IRQ7F flag is not automatically cleared.

When activation interrupt sources of DTC and IRQ interrupts are used with the above combinations, clear the interrupt flag by software in the interrupt handling routine of the corresponding IRQ.

Note: * Only 0 can be written, to clear the flag.

Interrupt Controller

ISCRL—IRQ	Sense Co	ntrol Regi	ster L	ter L H'FEED			Interrupt Controller	
ISCRH								
Bit	15	14	13	12	11	10	9	8
	IRQ7SCB	IRQ7SCA	IRQ6SCB	IRQ6SCA	IRQ5SCB	IRQ5SCA	IRQ4SCB	IRQ4SCA
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
			IRQ7 to	IRQ4 sens	se control	A and B		
ISCRL								
Bit	7	6	5	4	3	2	1	0
	IRQ3SCB	IRQ3SCA	IRQ2SCB	IRQ2SCA	IRQ1SCB	IRQ1SCA	IRQ0SCB	IRQ0SCA
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

ISCRH—IRQ Sense Control Register H

IRQ3 to IRQ0 sense control A and B

H'FEEC

1	bits 7 to 0 bits 7 to 0	Description
IRQ7SCB to IRQ0SCB	IRQ7SCA to IRQ0SCA	Description
0	0	Interrupt request generated at IRQ7 to IRQ0 input at low level
	1	Interrupt request generated at falling edge of IRQ7 to IRQ0 input
1	0	Interrupt request generated at rising edge of IRQ7 to IRQ0 input
	1	Interrupt request generated at both falling and rising edges of IRQ7 to IRQ0 input

DTCER—DTC	C Enable	Register		H'1	D	ГС			
Bit	7	6	5	4	3	2	1	0	_
	DTCE7	DTCE6	DTCE5	DTCE4	DTCE3	DTCE2	DTCE1	DTCE0	
Initial value	0	0	0	0	0	0	0	0	
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	
	0 D	octivation e TC activation of When data When the TC activation of Holding co When the E ave not er	tion by interpretations by interpretations; a transfer specified tion by interpretation]	ends with number o	the DISEI f transfers nabled	end			

DTVECR—D	ΓC Vector	r Register		H'l	FEF3			DTC
Bit	7	6	5	4	3	2	1	0
	SWDTE	DTVEC6	DTVEC5	DTVEC4	DTVEC3	DTVEC2	DTVEC1	DTVEC0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/(W)*	R/W	R/W	R/W	R/W	R/W	R/W	R/W
	DTC software activation enable 0 DTC software activation is disabled [Clearing condition] When the DISEL bit is 0 and the specified number of transfers have not ended 1 DTC software activation is enabled [Holding conditions] • When data transfer ends with the DISEL bit set to 1							
	•	When the During so	•			end		

Note: * A value of 1 can always be written to the SWDTE bit, but 0 can only be written after 1 is read.

ABRKCR—Ad	ldress Bro	eak Contr	ol Regist	er	H'FEF4		Interrup	t Controller
Bit	7	6	5	4	3	2	1	0
	CMF	_	_	_	_	_	_	BIE
Initial value	0	0	0	0	0	0	0	0
Read/Write	<u>R</u>	_	_	_		Addres	— upt enable ss break di	sabled
	0 [Cl WI 1 [Se	etting cond	ndition] ss break in dition]	•	cception ha	andling is	executed	

BARA—Break Address Register A BARB—Break Address Register B BARC—Break Address Register C					H'FEF5 H'FEF6 H'FEF7		Interrupt Controller Interrupt Controller Interrupt Controller		
Bit	7	6	5	4	3	2	1	0	
BARA	A23	A22	A21	A20	A19	A18	A17	A16	
Initial value	0	0	0	0	0	0	0	0	
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	

Specifies address (bits 23 to 16) at which address break is to be generated

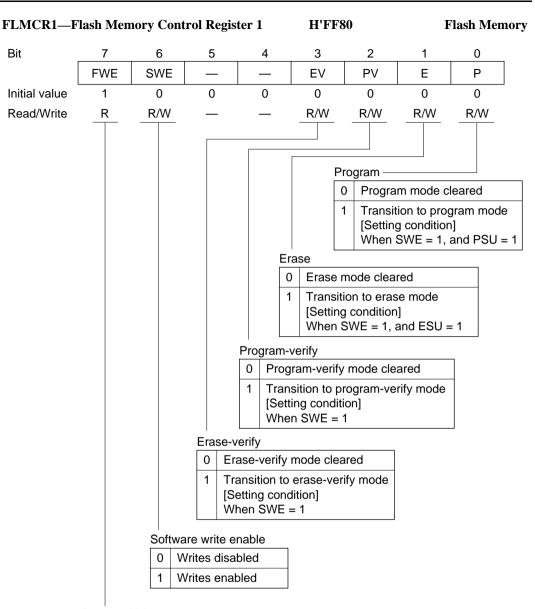
Bit	7	6	5	4	3	2	1	0
BARB	A15	A14	A13	A12	A11	A10	A9	A8
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W							
-								

Specifies address (bits 15 to 8) at which address break is to be generated

Bit	7	6	5	4	3	2	1	0
BARC	A7	A6	A5	A4	A3	A2	A1	_
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	_						
•								

Specifies address (bits 7 to 1) at which address break is to be generated





Reserved bit

FLMCR2—F	lash M	Iemory Cont	rol Regis	ter 2	H'FF	`81	F	lash Memory
Bit	7	6	5	4	3	2	1	0
	FLE	R —	_	_	_	_	ESU	PSU
Initial value	0	0	0	0	0	0	0	0
Read/Write	<u>R</u>	_	_	_		0 P	m setup rogram se rogram se Setting cor/hen SWE	ndition]
					0 1		ondition]	d
	Flas	sh memory eri	or					
	0	Flash memo Flash memo [Clearing col Reset or har	ry is opera ry prograr ndition]	n/erase pro	otection	(error proted	ction) is di	sabled
	1	An error has Flash memo [Setting cond See section	ry prograr dition]	n/erase pro	otection	(error proted		

PCSR—Periph	eral Cloc	k Select F	Register		H'FF82			PW	M
Bit	7	6	5	4	3	2	1	0	_
	_	_	_	_	_	PWCKB	PWCKA	_	
Initial value	0	0	0	0	0	0	0	0	-
Read/Write	_	_	_	_	_	R/W	R/W	_	

PWM clock select -

PWSL		PC	SR	
Bit 7	Bit 6	Bit 2	Bit 1	Description
PWCKE	PWCKS	PWCKB	PWCKA	
0	_	_	_	Clock input is disabled
1	0	_	_	φ (system clock) is selected
	1	0	0	φ/2 is selected
			1	φ/4 is selected
		1	0	φ/8 is selected
			1	φ/16 is selected

SYSCR2—Sys	stem Cont	rol Regis	ter 2		H'FF8	33		HIF
Bit	7	6	5	4	3	2	1	0
	KWUL1	KWUL0	P6PUE	_	SDE	CS4E	CS3E	HI12E
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	_	R/W	R/W	R/W	R/W
					CS4 (0)	0 Hos are 1 Hos are 2S3 enable 0 Host in function 1 Host in	disabled at interface enabled terface pin as disable terface pin as enable ace pin chaisabled ace pin chaisabled	e functions e functions n channel 3 d n channel 3 d
				Shutdown				
								n disabled
				1 Host	пцегтасе	pin shutdo	wn iunctio	n enabled
		Pc 0	input pu	rd current ull-up func	specificat	ion is selected f		
	⊦ Ke	y wakeup	level 1 an	d 0				
	0				selected as	s port 6 inp	ut level	
		1 Inpu	t level 1 is	selected	as port 6	input level		
	1					input level		
		1 Inpu	t level 3 is	selected	as port 6	input level		

EBR1—Eras EBR2—Eras		U			H'FF8 H'FF8			Flash Memo Flash Memo	•
Bit	7	6	5	4	3	2	1	0	
EBR1	_	_	_	_	_	_	EB9/—*2	EB8/—*2	
Initial value	0	0	0	0	0	0	0	0	
Read/Write	_	_	_	_	_	_	R/W*1*2	R/W*1*2	
Bit	7	6	5	4	3	2	1	0	
EBR2	EB7	EB6	EB5	EB4	EB3	EB2	EB1	EB0	
Initial value	0	0	0	0	0	0	0	0	
Read/Write	R/W*1	R/W	R/W	R/W	R/W	R/W	R/W	R/W	

Notes:

- 1. In normal mode, these bits cannot be modified and are always read as 0.
- 2. Bits EB8 and EB9 are not present in the 64-kbyte versions; they must not be set to 1.

Erase Blocks

Block	(Size)	Addresses	
128-kbyte versions	64-kbyte versions	Addresses	
EB0 (1 kbyte)	EB0 (1 kbyte)	H'(00)0000 to H'(00)03FF	
EB1 (1 kbyte)	EB1 (1 kbyte)	H'(00)0400 to H'(00)07FF	
EB2 (1 kbyte)	EB2 (1 kbyte)	H'(00)0800 to H'(00)0BFF	
EB3 (1 kbyte)	EB3 (1 kbyte)	H'(00)0C00 to H'(00)0FFF	
EB4 (28 kbytes)	EB4 (28 kbytes)	H'(00)1000 to H'(00)7FFF	
EB5 (16 kbytes)	EB5 (16 kbytes)	H'(00)8000 to H'(00)BFFF	
EB6 (8 kbytes)	EB6 (8 kbytes)	H'(00)C000 to H'(00)DFFF	
EB7 (8 kbytes)	EB7 (8 kbytes)	H'00E000 to H'00FFFF	
EB8 (32 kbytes)	_	H'010000 to H'017FFF	
EB9 (32 kbytes)	_	H'018000 to H'01FFFF	

SBYCR—Standby Control Register H'FF84 System 5 2 Bit 7 6 4 3 1 0 SSBY STS2 STS1 SCK2 SCK1 SCK0 STS0 0 0 0 0 0 0 0 Initial value 0 Read/Write R/W R/W R/W R/W R/W R/W R/W

System	clock	select	2	to	0	-
--------	-------	--------	---	----	---	---

0	0	0	Bus master is in high-speed mode
		1	Medium-speed clock = φ/2
	1	0	Medium-speed clock = φ/4
		1	Medium-speed clock = φ/8
1	0	0	Medium-speed clock = φ/16
		1	Medium-speed clock = φ/32
	1	_	_

Standby timer select 2 to 0

,			
0	0	0	Standby time = 8192 states
		1	Standby time = 16384 states
	1	0	Standby time = 32768 states
		1	Standby time = 65536 states
1	0	0	Standby time = 131072 states
		1	Standby time = 262144 states
	1	0	Reserved
		1	Standby time = 16 states*

Note: * This setting must not be used in the flash memory version.

Software standby

0	Transition to sleep mode after execution of SLEEP instruction in high-speed mode or medium-speed mode Transition to subsleep mode on execution of SLEEP instruction in subactive mode
1	Transition to software standby mode, subactive mode, or watch mode after execution of SLEEP instruction in high-speed mode or medium-speed mode Transition to watch mode or high-speed mode after execution of SLEEP instruction in subactive mode

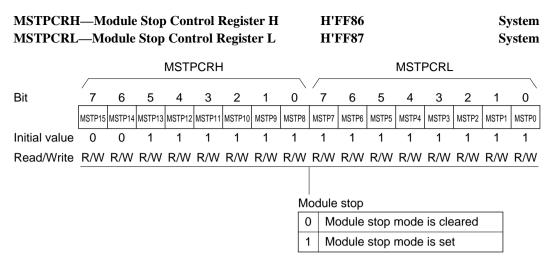
LPWRCR—I	Low-Powe	er Contro	l Register		H'FF8	5		System
Bit	7	6	5	4	3	2	1	0
	DTON	LSON	NESEL	EXCLE	_	_	_	_
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	R/W	_	_	_	_
			0 S	0 S	ubclock in sampling	put from E put from E frequency by 32	XCL pin i	s disabled s enabled
	Low-	speed on	flag					
	1	medium standby • When a is made • After wa • When a transitio • When a is made	-speed mode, or SLEEP in to watch reach mode SLEEP in is made SLEEP in to subslee	mode, or d is cleared struction is to watch r	sition is m de* s executed irectly to h , a transition s executed node or su s executed r watch m	ade to sle d in subact high-speed on is made d in high-s hactive m d in subact ode	ep mode, tive mode, I mode e to high-s peed mod ode* tive mode,	software , a transition speed mode e a , a transition

Note: * When a transition is made to watch mode or subactive mode, high-speed mode must be set.

Direct-transfer on flag

- When a SLEEP instruction is executed in high-speed mode or medium-speed mode, a transition is made to sleep mode, software standby mode, or watch mode*
 - When a SLEEP instruction is executed in subactive mode, a transition is made to subsleep mode or watch mode
- When a SLEEP instruction is executed in high-speed mode or medium-speed mode, a transition is made directly to subactive mode*, or a transition is made to sleep mode or software standby mode
 - When a SLEEP instruction is executed in subactive mode, a transition is made directly to high-speed mode, or a transition is made to subsleep mode

Note: * When a transition is made to watch mode or subactive mode, high-speed mode must be set.



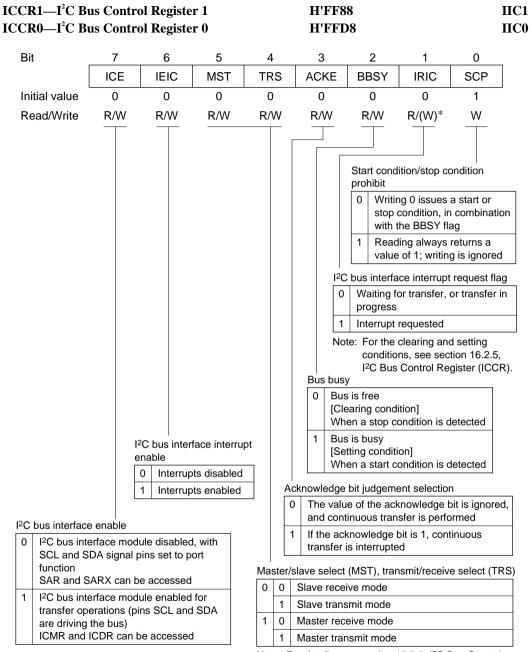
The correspondence between MSTPCR bits and on-chip supporting modules is shown below.

·							
Register	Bit	Module					
MSTPCRH	MSTP15	_					
	MSTP14*	Data transfer controller (DTC)					
	MSTP13	16-bit free-running timer (FRT)					
	MSTP12	8-bit timers (TMR0, TMR1)					
	MSTP11	8-bit PWM timer (PWM), 14-bit PWM timer (PWMX)					
	MSTP10	D/A converter					
	MSTP9	A/D converter					
	MSTP8	8-bit timers (TMRX, TMRY), timer connection					
MSTPCRL	MSTP7	Serial communication interface 0 (SCI0)					
	MSTP6	Serial communication interface 1 (SCI1)					
	MSTP5	Serial communication interface 2 (SCI2)					
	MSTP4*	I ² C bus interface (IIC) channel 0 (option)					
	MSTP3*	I ² C bus interface (IIC) channel 1 (option)					
	MSTP2*	Host interface (HIF), keyboard buffer controller (PS2)					
	MSTP1*	_					
	MSTP0*	_					

Notes: Do not set bit 15 to 1. Bits 1 and 0 can be read and written but do not affect operation.

* Must be set to 1 in the H8S/2144 Group.

Mode Ro	egister 2			H'FF	A 0			SCI1 SCI2 SCI0
7	6	5	4	3	2	1	0	
C/A	CHR	PE	O/E	STOP	MP	CKS1	CKS0	
0	0	0	0	0	0	0	0	_
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	
0 As	0 8-1 7-Note:	0 P 1 P teter length bit data bit data* When 7 of TDR MSB-firs mode us mode	Parity O E 1 C enable Parity bit ac Parity bit ac rity bit data is is not tran	top bit lenge of the length of	Multiproce Multiproce gth bit bits I checking I checking the MSB	0 0 1 1 1 mode essor funderssor form	φ clock φ/4 clock φ/16 cloc φ/64 clock	ck ck
	Commit O As	C/A CHR 0 0 R/W R/W Charac 0 8- 1 7- Note:	Mode Register 2 Mode Register 0 7 6 5 C/Ā CHR PE 0 0 0 0 R/W R/W R/W Parity 0 F 1 F Character length 0 8-bit data 1 7-bit data* Note: * When 7 of TDR MSB-fir Communication mode 0 Asynchronous mode	Node Register 2 Mode Register 0	Mode Register 2 Mode Register 0 T 6 5 4 3 C/A CHR PE O/E STOP O 0 0 0 0 0 R/W R/W R/W R/W R/W Stop bit leng O 1 stop 1 2 stop Parity mode O Even parity 1 Odd parity Parity enable O Parity bit addition and 1 Parity bit addition and 2 Note: * When 7-bit data is selected, of TDR is not transmitted. A MSB-first selection is not av	Mode Register 2 Mode Register 0 T 6 5 4 3 2 C/Ā CHR PE O/Ē STOP MP O 0 0 0 0 0 0 0 R/W R/W R/W R/W R/W R/W O 1 stop bit 1 2 stop bits Parity mode O Even parity 1 Odd parity Parity bit addition and checking 1 Parity bit addition and checking Character length O 8-bit data 1 7-bit data* Note: * When 7-bit data is selected, the MSB of TDR is not transmitted. Also, LSB-fi MSB-first selection is not available. Communication mode O Asynchronous mode	Mode Register 2	Mode Register 2

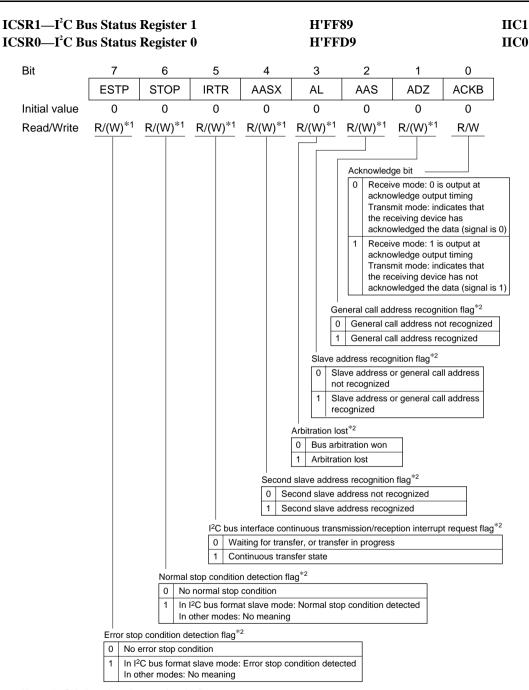


Note: For details, see section 16.2.5, I²C Bus Control Register (ICCR).

Note: $\,^*\,$ Only 0 can be written, to clear the flag.

BRR1—Bit Rate Register 1 BRR2—Bit Rate Register 2 BRR0—Bit Rate Register 0						H'FF89 H'FFA1 H'FFD9				
Bit	7	6	5	4	3	2	1	0	7	
Initial value	1	1	1	1	1	1	1	1	•	
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W		
			0-1-11-						=	

Sets the serial transmit/receive bit rate



Notes: 1. Only 0 can be written, to clear the flag.

2. For the clearing and setting conditions, see section 16.2.6, I²C Bus Status Register (ICSR).

SCR1—Serial (SCR2—Serial (SCR0—Serial (Control R	Register 2				H'FF8 H'FFA H'FFI	12			SCI1 SCI2 SCI0
Bit	7	6	5	4		3	2	1	0	
	TIE	RIE	TE	RE	N	//PIE	TEIE	CKE1	CKE0	
Initial value	0	0	0	0		0	0	0	0	
Read/Write	R/W	R/W	R/W	R/W		R/W	R/W	R/W	R/W	
			Clock enable 1 and 0							
				0	0	1 -	ronous	Internal clock/S		
						mode Synchr	onous	functions as I/C Internal clock/S	•	
						mode		functions as se	rial clock ou	itput
					1	Asynch mode	ronous	Internal clock/s functions as clo		
						Synchr mode	onous	Internal clock/S functions as se		itput
				1	0	Asynch mode	ronous	External clock/functions as clo		
						Synchr	onous	External clock/functions as se		out
					1	Asynch mode	ronous	External clock/functions as clo	•	
						Synchr mode	onous	External clock/functions as se	•	out
				Transm	it end	d interrup	t enable			
								EI) request disab		
				1 Tr	ansn	nit-ena ini	terrupt (11	EI) request enab	olea	
				Aultiprocesso Multipro				ed (normal rece	ntion mode)	
				[Clearin	g cor	ditions]			puon mode)	
Transmit interrupt	enable		When the MPIE bit is cleared to 0 When data with MPB = 1 is received							
	ta-empty inte	rrupt					pts enable			
(TXI) reques	ta-empty inte	rrupt		I			, ,	s, receive-error i RF, FER, and O		′
(TXI) reques				SSR are		bled unti	l data with	the multiproces	ssor bit set to	0
			Rec	eive enable						
Receive interrupt	enable —— a-full interrup	t (RXI)	0	Reception	disab	led				
	receive-error	` '	1	Reception	enab	ed				
1 Receive-dat	a-full interrup	` '		it enable ansmission o	lisab	led				
(ERI) reques		шопарс	1 Tr	ansmission e	enabl	ed				

7

1

R/W

6

1

R/W

5

Bit

Initial value

Read/Write

RDR1—Receiv RDR2—Receiv RDR0—Receiv	e Data R	egister 2		S	CI1 CI2 CI0				
Bit	7	6	5	4	3	2	1	0	_
Initial value	0	0	0	0	0	0	0	0	_
Read/Write	R	R	R	R	R	R	R	R	
Stores serial receive data									
TDR1—Trans		H'FF8B							
TDR2—Trans		H'FFA3							
TDR0—Trans	mit Data	Register 0			H'FFD	В		S	CI0

4

1 1 1 1 1 1 1 R/W R/W R/W R/W

2

1

0

3

Stores serial transmit data

SSR1—Serial Status Register 1 H'FF8C SCI1 SSR2—Serial Status Register 2 H'FFA4 SCI2 SSR0—Serial Status Register 0 **H'FFDC** SCI0 Bit 7 6 5 4 3 2 1 0 TDRF **RDRF ORER FFR PFR** TEND MPB **MPBT** Initial value 1 0 0 0 0 1 0 0 Read/Write R/(W)* R/(W)* R/(W)* R/(W)* R/(W)* R R R/W Multiprocessor bit transfer Data with a 0 multi-processor bit is transmitted Data with a 1 multi-processor bit is transmitted Multiprocessor bit 0 [Clearing condition] When data with a 0 multiprocessor bit is received [Setting condition] When data with a 1 multiprocessor bit is received Transmit end [Clearing conditions] . When 0 is written in TDRE after reading TDRE = 1 . When the DTC is activated by a TXI interrupt and writes data to TDR [Setting conditions] When the TE bit in SCR is 0 . When TDRE = 1 at transmission of the last bit of a 1-byte serial transmit character Parity error [Clearing condition] When 0 is written in PER after reading PER = 1 [Setting condition] When, in reception, the number of 1 bits in the receive data plus the parity bit does not match the parity setting (even or odd) specified by the O/E bit in SMR Framing error [Clearing condition] When 0 is written in FER after reading FER = 1 [Setting condition] When the SCI checks the stop bit at the end of the receive data when reception ends, and the stop bit is 0 Overrun error 0 [Clearing condition] When 0 is written in ORER after reading ORER = 1 [Setting condition] When the next serial reception is completed while RDRF = 1 Receive data register full [Clearing conditions] When 0 is written in RDRF after reading RDRF = 1 When the DTC is activated by an RXI interrupt and reads data from RDR [Setting condition] When serial reception ends normally and receive data is transferred from RSR to RDR Transmit data register empty [Clearing conditions] . When 0 is written in TDRE after reading TDRE = 1 . When the DTC is activated by a TXI interrupt and writes data to TDR [Setting conditions] . When the TE bit in SCR is 0 Note: * Only 0 can be written, to clear the flag. · When data is transferred from TDR to TSR and data can be written in TDR

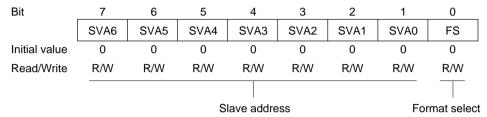
SCMR1—Ser SCMR2—Ser SCMR0—Ser	ial Interfa	ace Mo	de Register	2	H'FF8 H'FFA H'FFI	S	CI1 CI2 CI0		
Bit	7	6	5	4	3	2	1	0	
	_	_		_	SDIR	SINV	_	SMIF	
Initial value	1	1	1	1	0	0	1	0	
Read/Write	_	_	_	_	R/W	R/W	_	R/W	
			Data 0	Receive of	data is stor	ransmitted red in RDI	nterface m Norm Settir without m wifore being	munication node select al SCI mod ng prohibited nodification transmitted	le d
				TRECEIVE	uata is stoi	ed III INDI	X III IIIVEIR	50 101111	
		Data	a transfer dir	ection			_		
		0	TDR contents are transmitted LSB-first Receive data is stored in RDR LSB-first						
		1	TDR conte Receive da						

CDR1—I ² C Bus Data Register 1 H'FF8E CDR0—I ² C Bus Data Register 0 H'FFDE														
Bit	7	6	5	4	3	2	1	0						
	ICDR7	ICDR6	ICDR5	ICDR4	ICDR3	ICDR2	ICDR1	ICDR0						
Initial value	_	_	_	_	_	_	_	_	,					
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W						
ICDRR														
Bit	7	6	5	4	3	2	1	0						
	ICDRR7	ICDRR6	ICDRR5	ICDRR4	ICDRR3	ICDRR2	ICDRR1	ICDRR0						
Initial value	_	_	_	_	_	_	_	_						
Read/Write	R	R	R	R	R	R	R	R						
ICDRS														
Bit	7	6	5	4	3	2	1	0						
	ICDRS7	ICDRS6	ICDRS5	ICDRS4	ICDRS3	ICDRS2	ICDRS1	ICDRS0						
Initial value	_	_	_	_	_	_	_	_	,					
Read/Write	_	_	_	_	_	_	_	_						
ICDRT														
Bit	7	6	5	4	3	2	1	0	_					
	ICDRT7	ICDRT6	ICDRT5	ICDRT4	ICDRT3	ICDRT2	ICDRT1	ICDRT0						
Initial value	_	_	_	_	_		_	_	•					
Read/Write	W	W	W	W	W	W	W	W						
TDRE, RDR	(internal	flags)												
Bit							_	_						
							TDRE	RDRF						
Initial value							0	0	1					
Read/Write														

Note: For details, see section 16.2.1, I^2C Bus Data Register (ICDR).

SARX1—Second Slave Address Register 1	H'FF8E	IIC1
SAR1—Slave Address Register 1	H'FF8F	IIC1
SARX0—Second Slave Address Register 0	H'FFDE	IIC0
SAR0—Slave Address Register 0	H'FFDF	IIC0
SAR0—Slave Address Register 0	H'FFDF	IICO

SAR



SARX

Bit	7	6	5	4	3	2	1	0
	SVAX6	SVAX5	SVAX4	SVAX3	SVAX2	SVAX1	SVAX0	FSX
Initial value	0	0	0	0	0	0	0	1
Read/Write	R/W	R/W						

Second slave address

Format select

I Office Ser			
DDCSWR Bit 6	SAR Bit 0	SARX Bit 0	Operating Mode
SW	FS	FSX	
0	0	0	I ² C bus format • SAR and SARX slave addresses recognized
		1	I ² C bus format • SAR slave address recognized • SARX slave address ignored
	1	0	I ² C bus format • SAR slave address ignored • SARX slave address recognized
		1	Synchronous serial format • SAR and SARX slave addresses ignored
1	0	0	Formatless mode (start/stop conditions not
		1	detected)
	1	0	Acknowledge bit used
		1	Formatless mode* (start/stop conditions not detected) • No acknowledge bit

Note: * Do not set this mode when automatic switching to the I²C bus format is performed by means of the DDCSWR setting.



ICMR1—I²C Bus Mode Register 1 H'FF8F IIC1 ICMR0—I²C Bus Mode Register 0 IIC0 **H'FFDF** Bit 7 2 5 4 3 1 0 6 MLS WAIT CKS2 CKS₁ CKS0 BC₂ BC1 BC₀ Initial value 0 0 0 0 0 0 0 0 R/W R/W R/W R/W R/W R/W Read/Write R/W R/W Bit counter -Synchronous I²C bus BC2 BC0 BC₁ serial format format 0 0 0 8 9 1 1 2 2 1 0 3 1 3 4 1 0 0 4 5 1 5 6 1 0 6 7 7 1 8 Serial clock select IICX CKS2 CKS0 CKS1 Clock 0 0 0 0 φ/28 1 φ/40 1 0 φ/48 1 φ/64 1 0 0 φ/80 1 φ/100 1 0 φ/112 1 φ/128 1 0 0 0 φ/56 1 φ/80 1 0 φ/96 1 φ/128 1 0 0 φ/160 1 φ/200 1 0 φ/224 φ/256 Wait insertion bit Data and acknowledge bits transferred consecutively

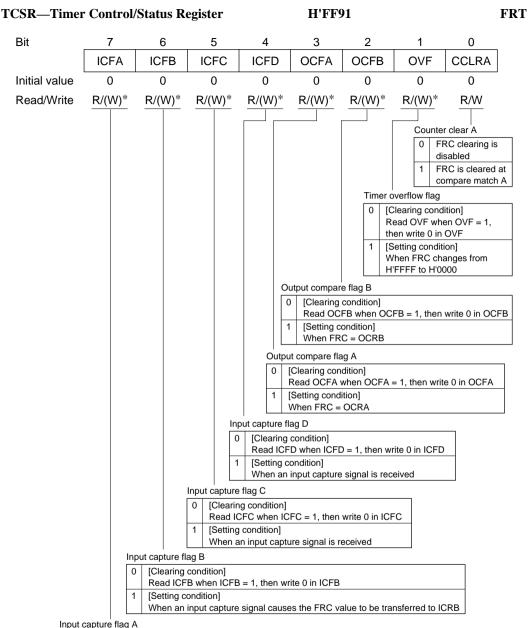
MSB-first/LSB-first select*

0	MSB-first
1	LSB-first

Note: * Do not set this bit to 1 when the I2C bus format is used.

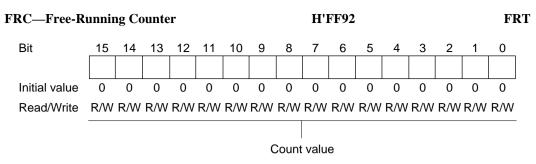
Wait inserted between data and acknowledge bits

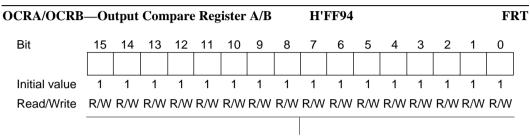
TIER—Time	r Interruj	pt Enable	Register		FRT			
Bit	7	6	5	4	3	2	1	0
	ICIAE	ICIBE	ICICE	ICIDE	OCIAE	OCIBE	OVIE	_
Initial value	0	0	0	0	0	0	0	1
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	_
Read/Write	R/W		Inp 0 1 Input c 0 In 1 In	out capture	Output com 0 Output (OCIA 1 Output (OCIA interrupt I pture inter pture inter errupt C er e interrupt e interrupt	Timer over 0 Time request 1 Time request 1 Output compare interest to compare is disabled to compare is enable of the request 1 output request 2 enable enab	erflow interer overflow uest (FOV) er overflo	s disabled aterrupt s enabled able request A
				re interrup		3 (ICIB) is	disabled	
		1 In	put captu	re interrup	t request E	B (ICIB) is	enabled	
	Input c	apture inte	errupt A er	nable				
	0 In	put captur	e interrup	t request A	(ICIA) is	disabled		
	1 In	put captur	e interrup	t request A	(ICIA) is	enabled		



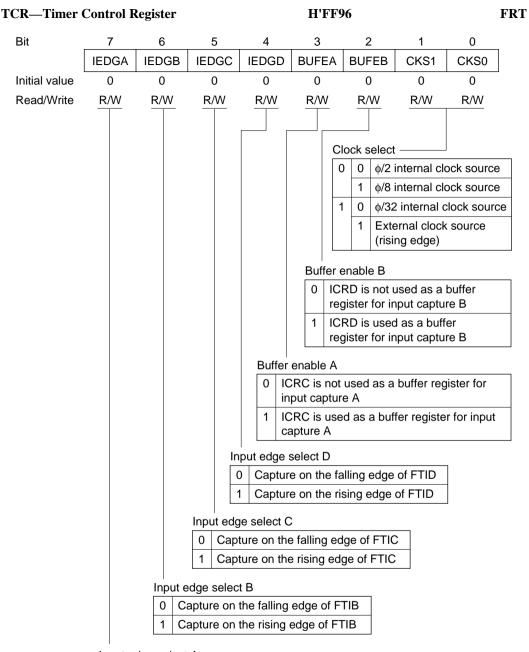
[Clearing condition] Read ICFA when ICFA = 1, then write 0 in ICFA [Setting condition] When an input capture signal causes the FRC value to be transferred to ICRA

Note: * Only 0 can be written in bits 7 to 1, to clear the flags.





Constantly compared with FRC value; OCF is set when OCR = FRC



Input edge select A

Capture on the falling edge of FTIA
 Capture on the rising edge of FTIA

TOCR—Time	er Output	Compare	Control	Register	H'FF	97			FRT	
Bit	7	6	5	4	3	2	1	0		
	ICRDMS	OCRAMS	ICRS	OCRS	OEA	OEB	OLVLA	OLVLB		
Initial value	0	0	0	0	0	0	0	0	,	
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W		
Read/Write	R/W	R/W	R/W		Output 0 0 1 0 utput enate 0 Output	Output lev 0 0 out matc 1 1 out matc enable B utput computput comp	put level E 0 output match B 1 output match B rel A put at com h A put at com h A pare B out pare B out	at compared at com	e- ed	
			(Output con	npare regi	ster select	t			
					A register					
					3 register	selected				
				oture regis						
						registers s OCRDM re		locted		
		ieciea								
	Output compare A mode select O OCRA set to normal operating mode									
						OCRAR a	and OCRA	.F		
				20.44.19 111						
		ture D mod		ating mode	<u> </u>					
	-									
	1 ICRD set to operating mode using OCRDM									

OCRAR—Out								FRT FRT								
Bit	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Initial value Read/Write	1 R/W	1 R/W	1 R/W	1 R/W	•	•	•	•	1 R/W	•	1 R/W	1 R/W	1 R/W	1 R/W	1 R/W	1 R/W

Used for OCRA operation when OCRAMS = 1 in TOCR (For details, see section 11.2.4, Output Compare Registers AR and AF (OCRAR, OCRAF).)

OCRDM—Output Compare Register DM											H'FF9C					
Bit	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Initial value	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
Read/Write	R	R	R	R	R	R	R	R	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

Used for ICRD operation when ICRDMS = 1 in TOCR (For details, see section 11.2.5, Output Compare Register DM (OCRDM).)

ICRA—Input ICRB—Input ICRC—Input ICRD—Input		H'FF98 H'FF9A H'FF9C H'FF9E						FRT FRT FRT FRT								
Bit	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Initial value	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
Read/Write	R	R	R	R	R	R	R	R	R	R	R	R	R	R	R	R

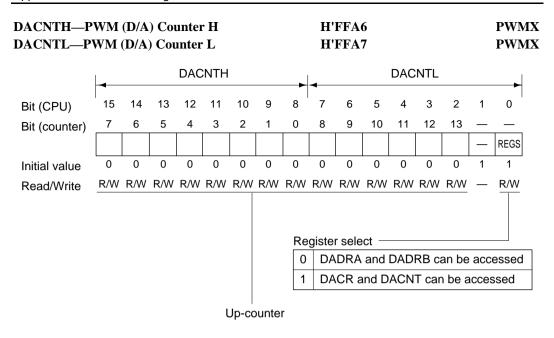
Stores FRC value when input capture signal is input (ICRC and ICRD can be used for buffer operation. For details, see section 11.2.3, Input Capture Registers A to D (ICRA to ICRD).)

DADRAH— DADRAL—I DADRBH—I DADRBL—I	PWM PWM	(D /A	A) Da A) Da	ıta R ıta R	egist egist	er Al er Bl	r AL H'FFA1 r BH H'FFA6						PW: PW: PW:	MX MX			
	—			DAE	DRH			DADRL							•		
Bit (CPU)	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0	
Bit (data)	13	12	11	10	9	8	7	6	5	4	3	2	1	0	_		
DADRA	DA13	DA12	DA11	DA10	DA9	DA8	DA7	DA6	DA5	DA4	DA3	DA2	DA1	DA0	CFS	-	
Initial value	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	_	
DADRB	DA13	DA12	DA11	DA10	DA9	DA8	DA7	DA6	DA5	DA4	DA3	DA2	DA1	DA0	CFS	REGS	
Initial value	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	
									0 0		A and	d DAI	DRB	can b		cesse	d
									•	cy se							
										e = re to H'F			Γ) × 6	4 DA	DR r	ange	
										e = re to H'F		ion (T	Γ) × 2	56 D	ADR	range	÷
					D/	1		اء ماء:	-4-								

D/A conversion data

DACR—PW	M (D/A)	Control Ro	egister		PV	VMX					
Bit	7	6	5	4		3		2	1	0	
	TEST	PWME	_	_		OEB	3	OEA	os	CKS	
Initial value	0	0	1	1		0		0	0	0	
Read/Write	R/W	R/W	_	_		R/W	_	R/W	R/W	R/W	
Read/Wille	N/VV	N/W		Output	ena	Output 60 PW (PV 11 PW ble B	lock : O O Sy 11 O Sy Dut se Dire Inve	select — perates /stem clo perates /stem clo elect ct PWM rted PW e A b/A) chai output p	at resolut ock cycle at resolut ock cycle output 'M output in) disable nnel A ou in) enable	tion (T) = time (t _{cyc}) tion (T) = time (t _{cyc}) time (t _{cyc})	
) disable			
								nel B out) enable			
		PWM e	nable								
		0 DA	ACNT ope	erates a	s a	14-bit ι	up-co	unter			
		1 DA	CNT halt	ts at H'	0003	3					
	Test r	mode									
	0 F	PWM (D/A) i	n user sta	ate: nor	mal	operat	ion				

0	PWM (D/A) in user state: normal operation
1	PWM (D/A) in test state: correct conversion results unobtainable



TCSR0—Timer	Contro	ol/Status	Register (0	H'FF	Ά8				WDT0
Bit	7	6	5	4	3		2	1	0	
	OVF	WT/ĪT	TME	RSTS	RST/NMI	Cł	(S2	CKS1	CKS0]
Initial value	0	0	0	0	0		0	0	0	_
Read/Write	R/(W)*	* R/W	R/W	R/W	R/W	R	/W	R/W	R/W	
					Clé		oloot 2	to 0		
						KS2	elect 2 CKS1	CKS0	Clock	
						0	0	0	φ/2	
						Ü		1	φ/64	
							1	0	φ/128	
								1	φ/512	
						1	0	0	φ/2048	
								1	φ/8192	
							1	0	ф/32768	
								1	φ/131072	
					Reset or	NIMI				
							rrupt ro	quested	\neg	
								-		
					1 Inter	nai r	eset re	quested		
				Reserve	ed bit					
			Timer er	nable						
			0 TC	NT is initia	lized to H'	00 ar	nd halte	ed	1	
			1 TC	NT counts						
		Time a m							J	
			node selec			DII a				
			terval timer quest (WO					ai timer	interrupt	
			atchdog tin		Generates	s a re	eset or	NMI inte	rrupt wher	ו
·	 Overflow	r flag								
Ţ		earing cond	itionel						7	
		rite 0 in the								
		ead TCSR		= 1*, the	n write 0 in	OVE	=			
		tting condit	•						1	
		en TCNT o	•	-						
		nen internal er mode, O								
		Mhan OV								

Note: * When OVF is polled and the interval timer interrupt is disabled, OVF = 1 must be read at least twice.

Note: * Only 0 can be written, to clear the flag.

TCNT0—Time				H'FFA	WE WE)Т0)Т1			
Bit	7	6	5	4	3	2	1	0	_
Initial value	0	0	0	0	0	0	0	0	J
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	
-									-
				Up-co	ounter				

PAODR—Por	t A Outpu	ıt Data Ro	egister		H'FFA	4		Port	A
Bit	7	6	5	4	3	2	1	0	
	PA7ODR	PA6ODR	PA5ODR	PA4ODR	PA3ODR	PA2ODR	PA10DR	PA0ODR	
Initial value	0	0	0	0	0	0	0	0	
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	

Output data for port A pins

PAPIN—Port	A Input I	ata Regis	ster		H'FFAI	3 (R)		Port	A
Bit	7	6	5	4	3	2	1	0	
	PA7PIN	PA6PIN	PA5PIN	PA4PIN	PA3PIN	PA2PIN	PA1PIN	PA0PIN	
Initial value	*	*	*	*	*	*	*	*	
Read/Write	R	R	R	R	R	R	R	R	
				Port A p	in states				

Note: $\,^*\,$ Determined by state of pins PA7 to PA0.

PADDR—Por	t A Data I	Direction 1	Register		H'FFAB (W)				t A
Bit	7	6	5	4	3	2	1	0	
	PA7DDR	PA6DDR	PA5DDR	PA4DDR	PA3DDR	PA2DDR	PA1DDR	PA0DDR	
Initial value	0	0	0	0	0	0	0	0	
Read/Write	W	W	W	W	W	W	W	W	

Specification of input or output for port A pins



REJ09B0327-0400



					Α	ppendix B	Internal	I/O Regist	ers
P1PCR—Port	1 MOS P	ull-Up Co	ontrol Reg	ister	H'FFA	\mathbb{C}		Por	t 1
Bit	7	6	5	4	3	2	1	0	
	P17PCR	P16PCR	P15PCR	P14PCR	P13PCR	P12PCR	P11PCR	P10PCR	
Initial value	0	0	0	0	0	0	0	0	1
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	_
		C	ontrol of p	ort 1 built-	in MOS in	put pull-up	os		
P2PCR—Port	2 MOS P	ull-Up Co	ntrol Reg	ister	H'FFAI)		Por	<u>t 2</u>
Bit	7	6	5	4	3	2	1	0	
	P27PCR	P26PCR	P25PCR	P24PCR	P23PCR	P22PCR	P21PCR	P20PCR	
Initial value	0	0	0	0	0	0	0	0	1
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	_
		С	ontrol of p	ort 2 built-	in MOS in	put pull-up	os		
P3PCR—Port	3 MOS P	ull-Up Co	ntrol Reg	gister	H'FFAI	E		Por	t 3
Bit	7	6	5	4	3	2	1	0	_
	P37PCR	P36PCR	P35PCR	P34PCR	P33PCR	P32PCR	P31PCR	P30PCR	
Initial value	0	0	0	0	0	0	0	0	•
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	_
		C	ontrol of p	ort 3 built-	in MOS in	put pull-up	os		
P1DDR—Port	1 Data D	irection R	Register		H'FFB()		Por	t 1
Bit	7	6	5	4	3	2	1	0	_
	P17DDR	P16DDR	P15DDR	P14DDR	P13DDR	P12DDR	P11DDR	P10DDR	

Specification of input or output for port 1 pins

W

W

0

W

W

0

W

0

W

W

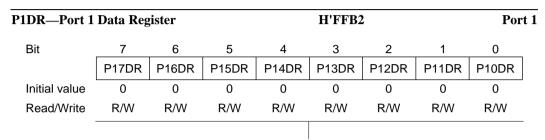
W

Initial value

Read/Write

P2DDR—Port	2 Data Di	rection R	legister		H'FFB1	-		Port	t 2
Bit	7	6	5	4	3	2	1	0	
	P27DDR	P26DDR	P25DDR	P24DDR	P23DDR	P22DDR	P21DDR	P20DDR	
Initial value	0	0	0	0	0	0	0	0	
Read/Write	W	W	W	W	W	W	W	W	

Specification of input or output for port 2 pins



Output data for port 1 pins

P2DR—Port 2	Data Reg	gister			Port 2			
Bit	7	6	5	4	3	2	1	0
	P27DR	P26DR	P25DR	P24DR	P23DR	P22DR	P21DR	P20DR
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

Output data for port 2 pins

P3DDR—Port	3 Data Di	irection R	legister		H'FFB4	Port 3		
Bit	7	6	5	4	3	2	1	0
	P37DDR	P36DDR	P35DDR	P34DDR	P33DDR	P32DDR	P31DDR	P30DDR
Initial value	0	0	0	0	0	0	0	0
Read/Write	W	W	W	W	W	W	W	W

Specification of input or output for port 3 pins



P4DDR—Port	4 Data Di	rection R	egister		H'FFB5	Port 4			
Bit	7	6	5	4	3	2	1	0	_
	P47DDR	P46DDR	P45DDR	P44DDR	P43DDR	P42DDR	P41DDR	P40DDR	
Initial value	0	0	0	0	0	0	0	0	
Read/Write	W	W	W	W	W	W	W	W	
		Sr	pecification	of input o	or output fo	or port 4 pi	ne		

Specification of input or output for port 4 pins

P3DR—Port 3	Data Reg	gister			Port 3			
Bit	7	6	5	4	3	2	1	0
	P37DR	P36DR	P35DR	P34DR	P33DR	P32DR	P31DR	P30DR
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

Output data for port 3 pins

P4DR—Port 4	Data Reg	gister		H'FFB7					
Bit	7	6	5	4	3	2	1	0	
	P47DR	P46DR	P45DR	P44DR	P43DR	P42DR	P41DR	P40DR	
Initial value	0	0	0	0	0	0	0	0	
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	
									-

Output data for port 4 pins

P5DDR—Port	5 Data D	irection R	legister		H'FFB	8		Port	t 5
Bit	7	6	5	4	3	2	1	0	
	_	_	_	_	_	P52DDR	P51DDR	P50DDR	
Initial value	1	1	1	1	1	0	0	0	
Read/Write		_	_		_	W	W	W	

Specification of input or output for port 5 pins

P6DDR—Port	6 Data Di	irection R	egister		H'FFB9	1		Por	t 6
Bit	7	6	5	4	3	2	1	0	
	P67DDR	P66DDR	P65DDR	P64DDR	P63DDR	P62DDR	P61DDR	P60DDR	
Initial value	0	0	0	0	0	0	0	0	
Read/Write	W	W	W	W	W	W	W	W	
		Sr	ecification	of input o	or output fo	or port 6 pi	ns		

P5DR—Port 5	Data Reg	gister			H'FFB	A		Por
Bit	7	6	5	4	3	2	1	0
	_	_	_	_	_	P52DR	P51DR	P50DR
Initial value	1	1	1	1	1	0	0	0
Read/Write	_	_	_	_	_	R/W	R/W	R/W

Output data for port 5 pins

P6DR—Port 6	Data Reg	gister		H'FFBB						
Bit	7	6	5	4	3	2	1	0		
	P67DR	P66DR	P65DR	P64DR	P63DR	P62DR	P61DR	P60DR		
Initial value	0	0	0	0	0	0	0	0	,	
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	_	

Output data for port 6 pins

PBODR—Por	t B Outpu	t Data Re	egister		H'FFB(C		Port B	
Bit	7	6	5	4	3	2	1	0	
	PB7ODR	PB6ODR	PB5ODR	PB4ODR	PB3ODR	PB2ODR	PB10DR	PB0ODR	
Initial value	0	0	0	0	0	0	0	0	1
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	-

Output data for port B pins

P8DDR—Port	8 Data D	irection R	legister		H'FFBI	O (W)		Port 8		
Bit	7	6	5	4	3	2	1	0		
	_	P86DDR	P85DDR	P84DDR	P83DDR	P82DDR	P81DDR	P80DDR	Ì	
Initial value	1	0	0	0	0	0	0	0		
Read/Write	_	W	W	W	W	W	W	W		

Specification of input or output for port 8 pins

PBPIN—Port	B Input D	ata Regis	ter		H'FFBI	O (R)		Por	t B
Bit	7	6	5	4	3	2	1	0	_
	PB7PIN	PB6PIN	PB5PIN	PB4PIN	PB3PIN	PB2PIN	PB1PIN	PB0PIN	
Initial value	*	*	*	*	*	*	*	*	,
Read/Write	R	R	R	R	R	R	R	R	
				Port B n	in states				

Port B pin states

Note: * Determined by state of pins PB7 to PB0.

PBDDR—Port	t B Data D	irection 1	Register		H'FFBI	E (W)		Port I		
Bit	7	6	5	4	3	2	1	0		
	PB7DDR	PB6DDR	PB5DDR	PB4DDR	PB3DDR	PB2DDR	PB1DDR	PB0DDR		
Initial value	0	0	0	0	0	0	0	0		
Read/Write	W	W	W	W	W	W	W	W		

Specification of input or output for port B pins

P7PIN—Port	7 Input Da	ata Regist	er			Port 7			
Bit	7	6	5	4	3	2	1	0	
	P77PIN	P76PIN	P75PIN	P74PIN	P73PIN	P72PIN	P71PIN	P70PIN	
Initial value	*	*	*	*	*	*	*	*	
Read/Write	R	R	R	R	R	R	R	R	
				Port 7 p	in states				

Note: * Determined by state of pins P77 to P70.

P8DR—Port 8	Data Reg	ister			H'FFBI	7		Por	t 8
Bit	7	6	5	4	3	2	1	0	_
	_	P86DR	P85DR	P84DR	P83DR	P82DR	P81DR	P80DR	
Initial value	1	0	0	0	0	0	0	0	•
Read/Write	_	R/W	R/W	R/W	R/W	R/W	R/W	R/W	
				Output o	data for po	rt 8 pins			

P9DDR—Port	9 Data Di	rection R	egister		H'FFC0		Port 9	
Bit	7	6	5	4	3	2	1	0
	P97DDR	P96DDR	P95DDR	P94DDR	P93DDR	P92DDR	P91DDR	P90DDR
Mode 1			1					
Initial value	0	1	0	0	0	0	0	0
Read/Write	W	W	W	W	W	W	W	W
Modes 2 and 3	3							
Initial value	0	0	0	0	0	0	0	0
Read/Write	W	W	W	W	W	W	W	W

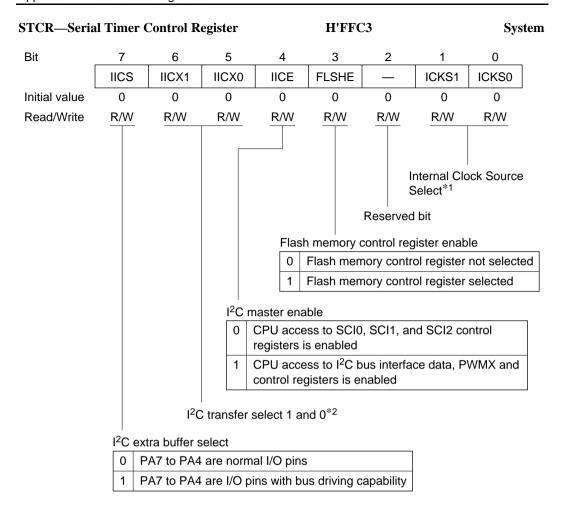
Specification of input or output for port 9 pins

P9DR—Port 9			Port 9						
Bit	7	6	5	4	3	2	1	0	_
	P97DR	P96DR	P95DR	P94DR	P93DR	P92DR	P91DR	P90DR	
Initial value	0	*	0	0	0	0	0	0	
Read/Write	R/W	R	R/W	R/W	R/W	R/W	R/W	R/W	
									-

Output data for port 9 pins

Note: $\ast\,$ Determined by state of pin P96.

IER—IRQ Ena	able Regis	ster			H'FFC2			Interrupt Controller		
Bit	7	6	5	4	3	2	1	0		
	IRQ7E	IRQ6E	IRQ5E	IRQ4E	IRQ3E	IRQ2E	IRQ1E	IRQ0E		
Initial value	0	0	0	0	0	0	0	0		
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W		
			IRC	7 to IRQ0	enable					
			0	IRQn int	errupt disa	bled				
			1	IRQn int	errupt ena	bled				
					(n =	7 to 0)				



- Notes: 1. Used for 8-bit timer input clock selection. For details, see section 12.2.4, Timer Control Register (TCR).
 - 2. Used for I²C bus interface transfer clock selection. For details, see section 16.2.4, I²C Bus Mode Register (ICMR).

H'FFC4 SYSCR—System Control Register System Bit 7 6 5 4 3 2 1 0 CS2E IOSE INTM1 **XRST NMIEG** HIF INTM₀ RAME Initial value 0 0 0 0 1 0 0 1 Read/Write R/W R/W R R/W R R/W R/W R/W RAM Enable On-chip RAM is disabled On-chip RAM is enabled Host interface enable Addresses H'(FF)FFF0 to H'(FF)FFF7 and H'(FF)FFFC to H'(FF)FFFF are used for access to 8-bit timer (channel X and Y) data registers and control registers, and timer connection control registers Addresses H'(FF)FFF0 to H'(FF)FFF7 and H'(FF)FFFC to H'(FF)FFFF are used for access to host interface data NMI edge select registers and control registers, and Falling edge keyboard controller and MOS input Rising edge pull-up control registers External reset Reset generated by watchdog timer overflow Reset generated by an external reset Interrupt control mode select INTM1 INTM0 Description 0 0 Interrupt control mode 0 Interrupt control mode 1 1 IOS enable The AS/IOS pin functions as the address strobe pin (Low output when accessing an external area) The AS/IOS pin functions as the I/O strobe pin (Low output when accessing a specified address from H'(FF)F000 to H'(FF)FE4F)*

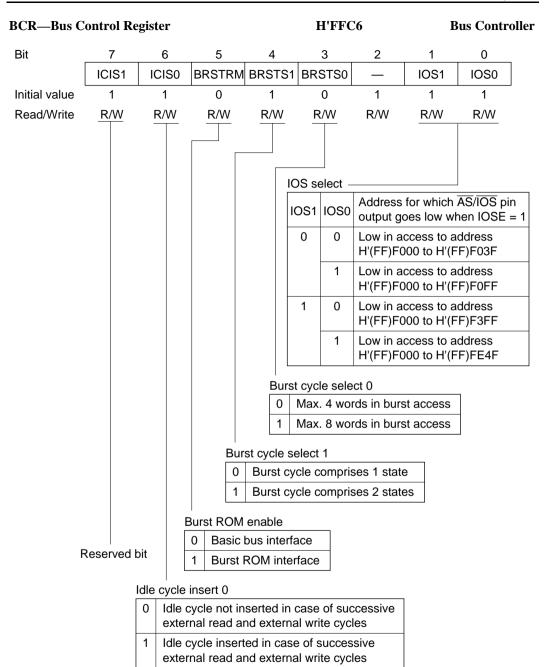
Note: * In the H8S/2148 F-ZTAT A-mask version and H8S/2147 F-ZTAT A-mask version, the address range is from H'(FF)F000 to H'(FF)F7FF.

CS2 enable

COZ CHAD	iC .	
SYSCR	HICR	
Bit 7	Bit 0	Description
CS2E	FGA20E	•
0	0	CS2 pin function halted
	1	(CS2 fixed high internally)
1	0	CS2 pin function selected for P81/CS2 pin
	1	CS2 pin function selected for P90/ECS2 pin

MDCR—Mode	e Control	Register			H'FFC5	5		System
Bit	7	6	5	4	3	2	1	0
	EXPE	_	_	_	_	_	MDS1	MDS0
Initial value	*	0	0	0	0	0	*	*
Read/Write	R/W*	_	_	_	_	_	R	R
	0 Si	led mode ongle-chip repanded m	mode sele				Mode sele	ect 1 and 0

Note: * Determined by the MD1 and MD0 pins.



Appendix B	Internal I/0	O Register	S							
WSCR—W	ait State C	Control Re	egister]	H'FF	C7]	Bus Cont	roller
Bit	7	6	5	4	;	3	2	1	0	
	RAMS	RAM0	ABW	AST	WN	/IS1	WMS	0 WC1	WC0	
Initial value	0	0	1	1		0	0	1	1	J
Read/Write	R/W	R/W	R/W	R/W	R	/W	R/W	R/W	R/W	
						W	ait coun	t 1 and 0 —		
							0 0	1	am wait e inserted	
							1	1	•	
							1 0	_	•	
	Reserv	ed bits					1	1	•	
					Wait n	node	select 1	and 0		
					0	0	Progr	am wait mod	de	
						1	Wait	disabled mo	de	
					1	0	Pin w	ait mode		
						1	Pin a	uto-wait mod	de	
			<u>A</u>	ccess sta	ite con	trol				
				acces Waits	s spac	e sertic	on in ext	s designated		е
				acces Waits	s spac	ce Isertic	on in ext	s designated		е

Bus width control

External memory space designated as 16-bit access space
 External memory space designated as 8-bit access space

TCR0—Time TCR1—Time TCRX—Tim TCRY—Tim	er Control er Control			TM TM	MR0 MR1 IRX IRY						
Bit	7	6	5	4	ļ	3	3	2	1	0	
	CMIEB	CMIEA	OVIE	CCL	.R1	CCI	_R0	CKS2	CKS1	CKS0	
Initial value	0	0	0	C	l)	()	0	0	0	J
Read/Write	R/W	R/W	R/W	R/	ν./	R/	W	R/W	R/W	R/W	
				Clock se	lect 2 to	0 0 — Bit 1	Bit 0				
				Channel	CKS2		CKS0		Descrip	tion	
	Counter clea	r 1 and 0 ——		0	0	0	0	Clock input dis	abled		\neg
	0 0 Clea	ar is disabled					1*1	Internal clock:	counting at fal	ling edge of \$\phi/8	3
		ared on compa	re					Internal clock:	counting at fal	ling edge of ø/2	?
	mate					1	0*1	Internal clock:	counting at fal	ling edge of ø/6	64
	1 0 Clea	ared on compar	re					Internal clock:	counting at fal	ling edge of $\phi/3$	32
		ared on rising e	dao				1*1	Internal clock:	counting at fal	ling edge of \phi/1	024
		kternal reset in						Internal clock:	counting at fal	ling edge of $\phi/2$	256
					1	0	0	Counting at TO	CNT1 overflow	signal*2	
				1	0	0	0	Clock input dis	abled		
	er overflow interre	•					1*1	Internal clock:	counting at fal	ling edge of \$/8	3
	OVF interrupt re	. , ,						Internal clock:	counting at fal	ling edge of \$\phi/2	<u> </u>
1	OVF interrupt re	quest (OVI) is	enabled			1	0*1			ling edge of \$/6	
										ling edge of \$\phi/1	
Compare ma	atch interrupt ena	abla A					1*1			ling edge of \$\phi/1	
	nterrupt request		hled							ling edge of \$\phi/2	2048
	nterrupt request				1	0	0	Count at TCN	•	atch A*2	
1 000 74 1	nterrupt request	(Olvin t) io criac	nou _	X	0	0	0	Clock input dis			
							1	Internal clock:			
Compare Match Inter	•					1	0			ling edge of \$\phi/2	
0 CMFB interrupt							1			ling edge of \$/4	
1 CMFB interrupt	request (CMIB) i	s enabled			1	0	0	Clock input dis			
				Y	0	0	0	Clock input dis		lina adaa af ±//	.—
						1	0			ling edge of \$/4	
						1	1			ling edge of \$\phi/2	
					1	0	0	Clock input dis		ling edge of φ/2	.040
				All	1	0	1	External clock		sina edae	-
				All	'	1	0	External clock			-+
						'	1			oth rising and fa	alling
							'	edges	. counting at be	sa noning and it	9
				Notes: 1.	Select	ed by I	CKS1 a	nd ICKS0 in ST	CR. For details	s, see section 1	2.2.4,

- Notes: 1. Selected by ICKS1 and ICKS0 in STCR. For details, see section 12.2.4. Timer Control Register (TCR).
 - If the clock input of channel 0 is the TCNT1 overflow signal and that of channel 1 is the TCNT0 compare match signal, no incrementing clock is generated. Do not use this setting.

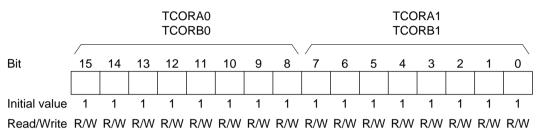
TCSR0—Time	r Contro	l/Status I	Register 0	H'FFCA						TMR0
TCSR0										
Bit	7	6	5	4		3	2	1	0	
	CMFB	CMFA	OVF	ADTE	=	OS3	OS2	OS1	OS0	
Initial value	0	0	0	0		0	0	0	0	_
Read/Write	R/(W)*	R/(W)*	R/(W)*	R/W		R/W	R/W	R/W	R/W	
						Output se	elect 1 and () ———		
						0 (No cha	nge at com	pare matc	h A
							0 outpu	t at compar	re match A	\
						1 (1 outpu	t at compar	re match A	١
								inverted at A (toggle ou	•	
				Output	ا دمام	ct 3 and		i (toggie oc	πραι)	
				0	0		z ange at con	npare matc	h B	
					1	+	ut at compa	•		
				1	0		ut at compa			
					1		inverted at	•		
						match	B (toggle or	utput)		
				trigger e						
			0	A/D co are dis			equests by	compare m	atch A	
							equests by	compare m	atch A	
				are ena			- 4			
			Timer ov	verflow fla	ag					
				earing co						
							, then write	0 in OVF		
				etting con nen TCN			om H'FF to I	H'00		
		Compar	e match flag							
			earing cond	-						
		• F	Read CMFA	when Cl						
			Vhen the D		vated	by a CN	/IIA interrup	t		
			etting conditi nen TCNT =							
C	│ ompare ma	tch flag B								
_		g conditions								
			n CMFB = 1				•			
-		condition]	activated by	y a CIVIIB	mer	τυρι				
	1	CNT = TCC	RB							

Note: * Only 0 can be written in bits 7 to 5, to clear the flags.

TCSR1—Time	r Co	Control/Status Register 1						H'FF	СВ			TMR1
TCSR1												
Bit		7	6		5	4		3	2	1	0	
	CN	/IFB	CMF	Ά	OVF	-	-	OS3	OS2	OS1	OS0)
Initial value	(0	0	•	0	1	•	0	0	0	0	
Read/Write	R/(W)*	R/(W)*	R/(W)*	_	-	R/W	R/W	R/W	R/W	1
				_								_
							Outp	ut seled	ct 1 and 0 -			
							0	0	No change	e at compa	are mat	ch A
								1	0 output a	t compare	match	Α
							1	0	1 output a	•		
								1	Output inv match A (t			
						outpu	t sele	ct 3 and	d 2			
						0	0	No ch	nange at co	mpare ma	tch B	
							1	0 out	out at comp	are match	В	
						1	0	1 out	out at comp	are match	В	
							1		ut inverted an B (toggle o	•	Э	
					Timer ov	orflow	, flag					
					$\overline{}$	earing		ition1				
									= 1, then w	rite 0 in O	VF	
					•	tting c						
					Wh	en TC	NT o	verflows	s from H'FF	to H'00		
			Com	pare	match fl	ag A						
					aring cor							
									, then write		4	
			1	[Set	ting cond	lition]						
				vvhe	en TCNT	= 100	JRA					
	Con	npare	are match flag B									
	0		Clearing conditions] Read CMFB when CMFB = 1, then write 0 in CMFB									
					when CM C is activ							
	1	[Sett	ting cor	nditio			.,					

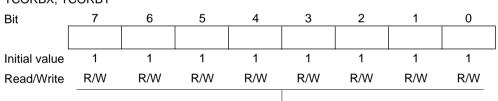
Note: * Only 0 can be written in bits 7 to 5, to clear the flags.

TCORA0—Time Constant Register A0	H'FFCC	TMR0
TCORA1—Time Constant Register A1	H'FFCD	TMR1
TCORB0—Time Constant Register B0	H'FFCE	TMR0
TCORB1—Time Constant Register B1	H'FFCF	TMR1
TCORAY—Time Constant Register AY	H'FFF2	TMRY
TCORBY—Time Constant Register BY	H'FFF3	TMRY
TCORC—Time Constant Register C	H'FFF5	TMRX
TCORAX—Time Constant Register AX	H'FFF6	TMRX
TCORBX—Time Constant Register BX	H'FFF7	TMRX



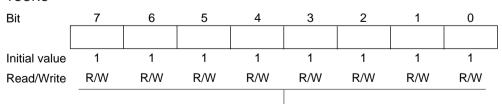
Compare match flag (CMF) is set when TCOR and TCNT values match

TCORAX, TCORAY TCORBX, TCORBY



Compare match flag (CMF) is set when TCOR and TCNT values match

TCORC



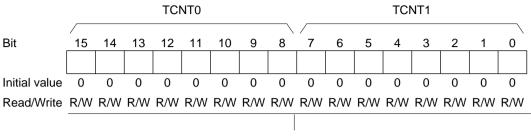
Compare match C signal is generated when sum of TCORC and TICR contents match TCNT value

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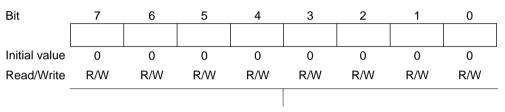


TCNT0—Timer Counter 0	H'FFD0	TMR0
TCNT1—Timer Counter 1	H'FFD1	TMR1
TCNTX—Timer Counter X	H'FFF4	TMRX
TCNTY—Timer Counter Y	H'FFF4	TMRY



Up-counter

TCNTX, TCNTY



Up-counter

PWOERA—PV PWOERB—PV	_		0		H'FFD3 H'FFD2			PW PW	
Bit	7	6	5	4	3	2	1	0	_
PWOERA	OE7	OE6	OE5	OE4	OE3	OE2	OE1	OE0	
Initial value	0	0	0	0	0	0	0	0	_
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	
Bit	7	6	5	4	3	2	1	0	_
PWOERB	OE15	OE14	OE13	OE12	OE11	OE10	OE9	OE8	
Initial value	0	0	0	0	0	0	0	0	-
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	

Switching between PWM output and port output

DDR	OE	Description
0	0	Port input
	1	Port input
1	0	Port output or PWM 256/256 output
	1	PWM output (0 to 255/256 output)

PWDPRA—PV PWDPRB—PV		H'FFD5 H'FFD4						
Bit	7	6	5	4	3	2	1	0
PWDPRA	OS7	OS6	OS5	OS4	OS3	OS2	OS1	OS0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Bit	7	6	5	4	3	2	1	0
PWDPRB	OS15	OS14	OS13	OS12	OS11	OS10	OS9	OS8
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

PWM output polarity control

0	PWM direct output (PWDR value corresponds to high width of output)
1	PWM inverted output (PWDR value corresponds to low width of output)

PWSL—PWM Register Select					H'I	FD	6				P	WM
Bit	7	6	5	4	;	3		2		1	0	
	PWCKE	PWCKS	_	_	R	S3		RS2		RS1	RS0	
Initial value	0	0	1	0	()		0		0	0	
Read/Write	R/W	R/W	_	_	_R/	W		R/W		R/W	R/W	
					Reg	ister	Sel	ect =				
					0	0	0		PW	DR0 sele	ected	İ
								1	PW	DR1 sele	ected	ı
							1	0	PW	DR2 sele	ected	1
								1	PW	DR3 sele	ected	1
						1	0	0	PW	DR4 sele	ected	1
								1	PW	DR5 sele	ected	1
							1	0	PW	DR6 sele	ected	1
								1	PW	DR7 sele	ected	1
					1	0	0	0	PW	DR8 sele	ected	1
								1	PW	DR9 sele	ected	1
							1	0	PW	DR10 se	lected	ı
								1	PW	DR11 se	lected	1
						1	0	0	PW	DR12 se	lected	ı
								1	PW	DR13 se	lected	ı
							1	0	PW	DR14 se	lected	ı
								1	PW	DR15 se	lected	Ì

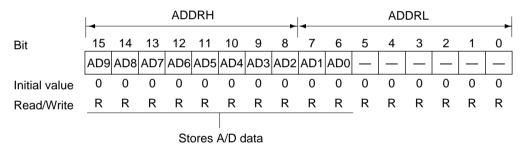
PWM clock enable, PWM clock select

PV	PWSL		SR	
Bit 7	Bit 6	Bit 2	Bit 1	Description
PWCKE	PWCKS	PWCKB	PWCKA	
0	_	_		Clock input disabled
1	0	_		φ (system clock) selected
	1	0	0	φ/2 selected
			1	φ/4 selected
		1	0	φ/8 selected
			1	φ/16 selected

PWDR0 to PW	DR15—F	PWM Dat	a Registe	rs	H'FFD7	7		PV	VM
Bit	7	6	5	4	3	2	1	0	_
Initial value	0	0	0	0	0	0	0	0	_
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	

Specifies duty factor of basic output pulse and number of additional pulses

ADDRAH—A/D Data Register AH	H'FFE0	A/D Converter
ADDRAL—A/D Data Register AL	H'FFE1	A/D Converter
ADDRBH—A/D Data Register BH	H'FFE2	A/D Converter
ADDRBL—A/D Data Register BL	H'FFE3	A/D Converter
ADDRCH—A/D Data Register CH	H'FFE4	A/D Converter
ADDRCL—A/D Data Register CL	H'FFE5	A/D Converter
ADDRDH—A/D Data Register DH	H'FFE6	A/D Converter
ADDRDL—A/D Data Register DL	H'FFE7	A/D Converter



Correspondence between analog input channels and ADDR registers

Analog Inp	A/D Data Register	
Group 0	Group 1	A/D Data Register
AN0	AN4	ADDRA
AN1	AN5	ADDRB
AN2	AN6 or CIN0 to CIN7	ADDRC
AN3	AN7 or CIN8 to CIN15	ADDRD

ADCSR—A/D Control/Status Register					I	H'FF	E8		A/D Converter			
Bit	7	6	5	4		3	2	1	0			
	ADF	ADIE	ADST	SCAN	С	KS	CH2	CH1	CH0			
Initial value	0	0	0	0	•	0	0	0	0			
Read/Write	R/(W)*	R/W	R/W	R/W	R	/W	R/W	R/W	R/W			
				Channel								
				Group selection	Cha	nnel ction		Description				
				CH2	CH1	CH0	Single mode		Scan mode			
				0	0	0	AN0	AN0				
						1	AN1	ANO, AN	N1			
					1	0	AN2	ANO, AN				
						1	AN3		N1, AN2, AN3			
				1	0	0	AN4 AN5	AN4	15			
				-	1	0	AN6 or CIN0 to	AN4, AN	15, AN6 or CIN0 to 7			
					•	1	AN7 or CIN8 to	15 AN4, AN	N5, AN6 or CIN0 to 7, CIN8 to 15			
								'				
				ock select Conversi	on time	- 266	states (max.)	\neg				
							states (max.)	_				
				Conversi	OII tillie	, = 104	states (max.)					
			Scan mo	ode								
			0 Sin	gle mode								
			1 Sca	an mode]							
		A/D	start									
		0	A/D convers	sion stoppe	d							
		1					arted. Cleared	to 0 automa	atically			
							channel ends ted. Conversio	n continuo				
							els until ADST					
							on to standby n					
			stop mode	е								
			upt enable		(ADI) =		dia alsia d					
			conversion er									
	A /D I fl		conversion er	ia interrupt	(ADI) II	equest	enabled					
	A/D end fla	ag ring condition	nel .									
	• Wh	en 0 is writter	n in the to AD s activated by									
		ng conditions			1.7							
	• Sin	gle mode: Wh	hen A/D conv en A/D conve			pecifie	d channels					
]				

Note: * Only 0 can be written, to clear the flag.

ADCR—A/D Control Register					H'FFE9	A/D Converter			
Bit	7	6	5	4	3	2	1	0	
	TRGS1	TRGS0	_	_	_	_	_	_	
Initial value	0	0	1	1	1	1	1	1	
Read/Write	R/W	R/W	_	_	_	_	_		

Timer trigger select

0	0	Start of A/D conversion by external trigger is disabled
	1	Start of A/D conversion by external trigger is disabled
1	0	Start of A/D conversion by external trigger (8-bit timer) is enabled
	1	Start of A/D conversion by external trigger pin is enabled

TCSR1—Time	r Contro	l/Status	Register 1	L	H'FF	EA				WDT1
Bit	7	6	5	4	3	2		1		0
	OVF	WT/ĪT	TME	PSS	RST/NMI	CKS2	Cł	(S1	Cł	KS0
Initial value	0	0	0	0	0	0		0		0
Read/Write	R/(W)*1	R/W	R/W	R/W	R/W	R/W	R	/W	R	/W
						Clock	select 2		CKS0	Clock
							0			Clock
						0	"	0	0	φ/2
									1	φ/64
								1	0	φ/128
							-	_	1	φ/512
							1	0	0	φ/2048
								_	0	φ/8192
								1	1	φ/32768 φ/131072
						1	0	0	0	φ/131072 φSUB/2
						'	"	"	1	φSUB/4
								1	0	φSUB/8
								'	1	φSUB/16
							1	0	0	φSUB/32
							'	"	1	φSUB/64
								1	0	φSUB/128
									1	φSUB/256
					Reset or NM	11	1			,
						terrupt reques	ted			
					1 Interna	al reset reques	ted			
				 Prescaler	select*2					
					IT counts on a ø-	based prescal	ler (PSN	Л) scale	ed clock	:
					IT counts on a øs					
			 Timer enat	ole						
					o H'00 and halte	d				
				Γ counts						
		Timer mo	ode select							
		0 Inte	rval timer mode:			(WOVI)				
			t to CPU when T							
			tchdog timer mod CPU when TCNT		MI interrupt reque	est sent				
	Overflow fla									
		ng conditions]								
	Write	0 in the TME	bit OVF = 1*, then w	vrite 0 in OVF						
		g condition]	vs (changes from	H'FF to H'00\						
	(When	internal reset	request generation cleared automatic	on is selected i						
		-	llad and the inten	, ,	,					

Note: * When OVF is polled and the interval timer interrupt is disabled, OVF = 1 must be read at least twice.

Notes: 1. Only 0 can be written, to clear the flag.

2. For operation control when a transition is made to power-down mode, see section 25.2.3, Timer Control/Status Register (TCSR).

HICR—Host In	nterface C	ontrol Re	gister		H'FFF0			HIF
Bit	7	6	5	4	3	2	1	0
	_	_	_	_	_	IBFIE2	IBFIE1	FGA20E
Initial value	1	1	1	1	1	0	0	0
Slave R/W	_	_	_	_	_	R/W	R/W	R/W
Host R/W					0 Input recei disat 1 Input	0 Fast disa 1 Fast enal register fur data register secomple oled data register complete oled	gate A20 oled all interrup ster (IDR1 te interrup	function function t enable 1 ot is

Input data register full interrupt enable 2

	1
0	Input data register (IDR2) receive complete interrupt is disabled
1	Input data register (IDR2) receive complete
	interrupt is enabled

RX—Time	er Contro	I/Status F	Kegister 2	Y	H'FI	F1		TM
CSRX								
Bit	7	6	5	4	3	2	1	0
	CMFB	CMFA	OVF	ICF	OS3	OS2	OS1	OS0
nitial value	0	0	0	0	0	0	0	0
Read/Write	R/(W)*	R/(W)*	R/(W)*	R/(W)*	R/W	R/W	R/W	R/W
					0			
					0 Output s	elect 1 and No cha		pare match A
					" -		ut at compar	
					1		ut at compai	
							inverted at	
						match	A (toggle οι	ıtput)
				Outp	ut select 3	3 and 2		
				0	0 N	lo change a	t compare m	natch B
					1 0	output at c	ompare mate	ch B
				1	0 1	output at c	ompare mate	ch B
						Output invert natch B (tog	ed at compa	are
						iateri b (tog	gie output)	
				put capture	condition	.1		
						-	eading ICF	= 1
			1	[
							y a falling ed signal after t	
						RI has been		
			Timer o	verflow flag				
				earing cond				
				nen 0 is writt		- after readi	ng OVF = 1	
				etting conditi nen TCNT o		rom H'FF to	H'00	
		 Compar	e match fla					
		` _	earing cond					
		1 1		ritten in CM		•		
				TC is activa	ted by a (MIA interru	pt	
			etting condition =					
Compare	match flag E							
	ring condition							
	en 0 is writt en the DTC			•	1			
	ng condition		, . O	oupt				
	n TCNT = T	-						

Note: * Only 0 can be written in bits 7 to 4, to clear the flags.

TCSRY—Tim	er C	Contro	ster Y	Y H'FFF1					7	ГMRY				
TCSRY														
Bit		7	(3		5	4		3		2	1	0	
	CI	ИFВ	CN	1FA	0	VF	ICIE	:	OS3		OS2	OS1	OS0	
Initial value		0	())	0		0		0	0	0	
Read/Write	R/((W)*	R/(W)*	R/(W)*	R/W	/	R/W		R/W	R/W	R/W	
	-													
									Outpu	ıt sele	ct 1 and	0 ——		
									0	0	No cha	ange at cor	npare matc	h A
										1	0 outp	ut at compa	are match A	4
									1	0	1 outp	ut at compa	are match A	A
										1		inverted a		
								, Outpu	ıt selec	t 3 ar	id 2			
								0	0	No c	hange a	compare	match B	
									1	0 ou	tput at co	mpare ma	tch B	
								1	0	1 ou	tput at co	mpare ma	tch B	
									1			ed at comp	oare	
							lanut a					gic output)		
							Input o					IX) is disal	oled	7
											` `	IX) is enab		-
					Tim	er ove	erflow fla		pr		,	., 1, 10 01142		
					0		aring co	<u> </u>	onl					
										F afte	r reading	OVF = 1		
					1		ing con			, ,				
						Whe	n ICN	ove	rflows 1	from I	H'FF to F	1'00		
						h flag								
			0		_	condit	-	MEA	A after i	readir	ıg CMFA	_ 1		
				l .							interrupt	I		
			1		•	onditio	-							
				Whe	en TC	NT = 7	CORA							
	Cor	npare i	match	flag I	3						_			
	0	[Clea												
							3 after red d by a C							
	1	[Setti					, u c			r.				
			TCN		•	3								

Note: * Only 0 can be written in bits 7 to 5, to clear the flags.

KMIMR—Key KMIMRA—K			-	U		H'FFF1 H'FFF3	_	ot Controller ot Controller
KMIMR								
Bit	7	6	5	4	3	2	1	0
	KMIMR7	KMIMR6	KMIMR5	KMIMR4	KMIMR3	KMIMR2	KMIMR1	KMIMR0
Initial value	1	0	1	1	1	1	1	1
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
			0 K		nput interi	nask rupt reques rupt reques		
KMIMRA								
Bit	7	6	5	4	3	2	1	0
	KMIMR15	KMIMR14	KMIMR13	KMIMR12	KMIMR11	KMIMR10	KMIMR9	KMIMR8
Initial value	1	1	1	1	1	1	1	1
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
				ard matrix ey-sense i		nask rupt reques	sts enable	d
			1 Ke	ey-sense i	nput interr	upt reques	sts disable	d
TICRR—Inpu	-	_			•	H'FFF2 H'FFF3		TMRX TMRX
1	•	6						

TICKK—Inpu	-	U				H'FFF3		TM	
Bit	7	6	5	4	3	2	1	0	_
Initial value	0	0	0	0	0	0	0	0	-
Read/Write	R	R	R	R	R	R	R	R	

Stores TCNT value at fall of external reset input

KMPCR—Por	t 6 MOS	Pull-Up C	Control Re	egister	H'FFF2	}		Por	t 6
Bit	7	6	5	4	3	2	1	0	
	KM7PCR	KM6PCR	KM5PCR	KM4PCR	KM3PCR	KM2PCR	KM1PCR	KM0PCR	
Initial value	0	0	0	0	0	0	0	0	
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	
		С	ontrol of p	ort 6 built-	in MOS in	put pull-up	os		

Note: KMPCR has the same address as TICRR/TCORAY of TMRX/TMRY.

When selecting KMPCR, set the HIE bit to 1 in SYSCR.

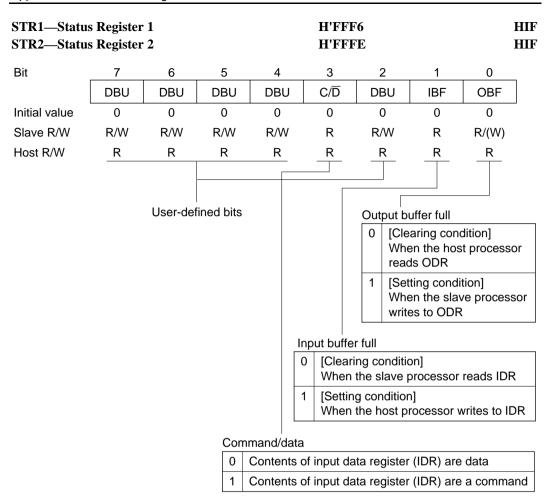
IDR1—Input IIIDR2—Input I			HII HII						
Bit	7	6	5	4	3	2	1	0	_
	IDR7	IDR6	IDR5	IDR4	IDR3	IDR2	IDR1	IDR0	
Initial value		_	_	_	_	_	_	_	_
Slave R/W	R	R	R	R	R	R	R	R	
Host R/W	W	W	W	W	W	W	W	W	

Stores host data bus contents at rise of \overline{IOW} when \overline{CS} is low

ODR1—Outpu ODR2—Outpu		H'FFF5 H'FFFD							
Bit	7	6	5	4	3	2	1	0	
	ODR7	ODR6	ODR5	ODR4	ODR3	ODR2	ODR1	ODR0	
Initial value	_	_	_	_	_	_	_	_	
Slave R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	
Host R/W	R	R	R	R	R	R	R	R	

ODR contents are output to the host data bus when HA0 is low, $\overline{\text{CS}}$ is low, and $\overline{\text{IOR}}$ is low

TISR—Time	r Input Se	elect Regi	ster		TM	1RY			
Bit	7	6	5	4	3	2	1	0	_
	_	_	_	_	_	_	_	IS	
Initial value	1	1	1	1	1	1	1	0	•
Read/Write	_	_	_	_	_	_	_	R/W	
		li	nput selec	t					
			Exte			H8S/2148 (ut is disabl		144 Group	Э,
			1 VSY	NCI/TMIY	(TMCIY/	TMRIY) is	selected		



DADR0—D/A DADR1—D/A		H'FFF8 H'FFF9	D/A Converter D/A Converter					
Bit	7	6	5	4	3	2	1	0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

Stores data for D/A conversion



							Appendix	B Interna	II I/O Regis	sters				
DACR—D/A	Control 1	Register	•			H'FFI	FA	D	/A Conve	rter				
Bit	7	6	5	4	1	3	2	1	0					
	DAOE1	DAOE	0 DAI	E -	_	_	_	_	_					
Initial value	0	0	0	'	1	1		1	1	,				
Read/Write	R/W	R/W	R/V	<u>v</u> –	_	_	_	_	_					
			D/A enal	oled										
		DAOE1 DAOE0 DAE Conv						version result						
	0 0					Chan	Channel 0 and 1 D/A conversion disabled							
				1	0									
					1	Chan	nel 0 and 1	D/A conver	sion enable	d				
			1	0	0		Channel 0 D/A conversion disabled Channel 1 D/A conversion enabled							
					1	Chan	nel 0 and 1	D/A conver	sion enable	d				
				1	*	Chan	nel 0 and 1	D/A conver	sion enable	d				
			Legend:	*: Don't ca	are									
		D/A ou	/A output enable 0											
		0 A	nalog ou	tput DA0	disa	bled		D/A conversion disabled onversion enabled D/A conversion enabled onversion disabled D/A conversion disabled						
		D/A conversion is enabled on channel 0. Analog output DA0 is enabled												
	D/A out	put enak	ole 1											

0	Analog output DA1 disabled
1	D/A conversion is enabled on channel 1. Analog output DA1 is enabled
	Analog output DA1 is enabled

TCONRI—Ti	mer Conr	nection R	egister I		H'FFFC			Т	Timer Connection		
Bit	7	6	5	4		3	2	1	0)	
	SIMOD1	SIMOD0	SCONE	ICST	F	HFINV	VFINV	HIIN\	/ VIII	٧V	
Initial value	0	0	0	0	- I	0	0	0	0		
Read/Write	R/W	R/W	R/W	R/W		R/W	R/W	R/W	R/\	W	
				nchronizati version VSYNCI	pin state						
			is used directly as the VSYNCI input 1 The VSYNCI pin st is inverted before u								
					Inni	ut cynchro	nization ci	gnal inversion	CI input		
					0	The HS	YNCI and	CSYNCI pin	states are	I	
					1	· ·		YNCI and CS CSYNCI pin			
						1		HSYNCI and		I	
			Input synchronization signal inversion								
								tly as the VF			
			1 The VFBACKI pin state is inverted before use as the VFBACKI input								
		Input synchronization signal inversion									
					•			as the HFBA			
			1 The	HFBACKI	pin st	ate is inve	rted befor	e use as the	HFBACKI	input	
			put capture s								
		a	[Clearing	condition]				is detected of			
	When a rising edge followed by a falling edge is detected on TMRIX 1 The TICRR and TICRF input capture functions are operating (Waiting for detection of a rising edge followed by a falling edge on T [Setting condition]							ng			
			-1	When 1 is written in ICST after reading ICST = 0							
		Synchron	ization signal Mode		on ena TIA	FTIB	FTIC	FTID	TMCI1	TMRI1	
		0	Normal		TIA	FTIB	FTIC	FTID	TMCI1	TMRI1	
			connectio		put	input	input	input	input	input	
		1	Synchroni signal con tion mode	nec- si	/I gnal	TMO1 signal	VFBAC input	KI IHI signal	IHI signal	IVI inverse signal	
	Input syncl	hronization m	ode select 1	and 0						_	
	SIMOD1	Мо	de		IHI sigr			IVI signal			
	0	0	No signal			BACKI in			VFBACKI input		
						CSYNCI input PDC input HSYNCI input PDC input			\dashv		
	1	0 Composite mode				o rivoi inpi	uι	PDC input			

HSYNCI input

Separate mode

VSYNCI input

TCONRO—T	imer Co	nnection I	Register (H'FFI	F D	Tim	Timer Connection				
Bit	7	6	5	4	3	2	1	0			
	HOE	VOE	CLOE	CBOE	HOINV	VOINV	CLOINV	CBOINV			
Initial value	0	0	0	0	0	0	0	0			
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W			
Read/Write	R/W	Output 0 Ti	Output of O The Inc. enable ne P64/FTIC	Output sync O The IH The IH enable Re P27/A15/Pt mode 1 (expa fine P27/A15/ modes 2 and The P27/A15/	Output O Ti th Ti Us O signal is us O signal is in W15/CBLANA anded mode v PW15/CBLAN 3 (expanded PW15/CBLAN	Output sy O The or C the or C use synchroniza he IVO signa he IVO si	Output synch signal inverse Used d CBLAN The CE used d CBLAN The CE inverte the CB nchronization CLO signal (inverted Last Signal) is inverted to the CLAM It is used directly to the coutput signal inverted to the CBLAN It is inverted to the coutput signal inverted to the CBLAN Some signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to the coutput signal inverted to t	hronization ision BLANK signal is irectly as the like output BLANK signal is defore use as LANK output a signal inversion CL1, CL2, CL3, used directly as put CL1, CL2, CL3, inverted before PO output //ersion ctly as Defore like CO output HSYNCO output //A15/PW15 pin d): 5 pin enabled): BLANK pin			
		Output enable 0 The P61/FTOA/KIN1/CIN1/VSYNCO pin functions as the P61/FTOA/KIN1/CIN1 pin									
		he P61/FTO/						ı hiii			
Output	t enable										
0 T	he P44/TMC	01/HIRQ1/HS	YNCO pin fu	unctions as th	e P44/TMO1/	HIRQ1 pin					

The P44/TMO1/HIRQ1/HSYNCO pin functions as the HSYNCO pin

TCONRS—Timer Connection Register S					I	H'FF	FE	Timer Connection					
7	6	5		4		3	2	1	0				
TMRX/Y	ISGENE	HON	/OD1 F	HOMOD0	VON	/IOD1	VOMOD0	CLMOD1	CLMOD0				
0	0		0	0		0	0	0	0				
R/W	R/W	R	/W	R/W	R	/W	R/W	R/W	R/W				
		_											
									scription gnal is selected				
							1	The CL2 signal is selected					
						1	0		gnal is selected				
							1						
					1	0	0	The CL4 sig	gnal is selected				
							1						
						1							
							1						
			Vertical	synchroniz	ation o	ation output mode select 1 and 0							
					1 VON	IOD0	DD0 Description						
			0	0		0		•					
						1	The IVI signal (without fall modification, with IHI synchronization) is selected						
				1		0	The IVI signal (with fall modification, without IHI synchronization) is selected						
						1	The IVI signa	l (with fall mo	odification and				
			1	0		0	-						
						1	· ·						
				1		0							
						1							
	Horiz	ontal	synchron	ization out	put mo	ode sel	ect 1 and 0						
	ISG	ENE	HOMOD [,]	1 HOMOD	Description								
		0	0	0	The IHI signal (without 2fH modification) is selected								
						`	• •) is selected				
			1		The	CL1 s	ignal is select	ied					
	<u> </u>	1	0										
		'	U		I ne ing signal is selected								
		ŀ	1	0	1								
				1	7								
Inte	ernal synchro	nizatio	on signal	select	•								
	7 TMRX/Y 0 R/W	7 6 TMRX/Y ISGENE 0 0 R/W R/W Horiz ISG	TMRX/Y ISGENE HOM O O R/W R/W R Horizontal ISGENE O Internal synchronization	7 6 5 TMRX/Y ISGENE HOMOD1 H 0 0 0 R/W R/W R/W Vertical SISGENE 0	7 6 5 4 TMRX/Y ISGENE HOMOD1 HOMOD0 0 0 0 0 R/W R/W R/W R/W Vertical synchroniz ISGENE VOMOD 0 0 1 1 Horizontal synchronization out ISGENE HOMOD1 HOMOD0 0 0 1 1 1 0 1 0 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 1 1 0 1 0	7 6 5 4 TMRX/Y ISGENE HOMOD1 HOMOD0 VOM 0 0 0 0 0 R/W R/W R/W R/W R/W R SGENE O	TMRX/Y ISGENE HOMOD1 HOMOD0 VOMOD1	TMRX/Y ISGENE HOMOD1 HOMOD0 VOMOD1 VOMOD0	TMRXXY ISGENE HOMOD1 HOMOD0 VOMOD1 VOMOD0 CLMOD1				

TMRX/TMRY access select

- 0 The TMRX registers are accessed at addresses H'FFF0 to H'FFF5
- 1 The TMRY registers are accessed at addresses H'FFF0 to H'FFF5

Timer Connection SEDGR—Edge Sense Register H'FFFF 7 5 2 Bit 6 4 3 0 **VEDG HEDG CEDG HFEDG** VFEDG **PREQF** IHI IVI __*2 __*2 Initial value 0 0 0 0 0 0 R/(W)*1 R/(W)*1 R/(W)*1 R/(W)*1 R/(W)*1 R/(W)*1 Read/Write R R IVI signal level The IVI signal is low The IVI signal is high IHI signal level The IHI signal is low The IHI signal is high Pre-equalization flag [Clearing condition] When 0 is written in PREQF after reading PREQF = 1 [Setting condition] When an IHI signal 2fH modification condition is detected VFBACKI edge [Clearing condition] When 0 is written in VFEDG after reading VFEDG = 1 [Setting condition] When a rising edge is detected on the VFBACKI pin HFBACKI edge [Clearing condition] When 0 is written in HFEDG after reading HFEDG = 1 [Setting condition] When a rising edge is detected on the HFBACKI pin CSYNCI edge [Clearing condition] When 0 is written in CEDG after reading CEDG = 1 1 [Setting condition] When a rising edge is detected on the CSYNCI pin HSYNCI edge [Clearing condition] When 0 is written in HEDG after reading HEDG = 1 [Setting condition] When a rising edge is detected on the HSYNCI pin VSYNCI edge [Clearing condition] When 0 is written in VEDG after reading VEDG = 1 [Setting condition] When a rising edge is detected on the VSYNCI pin

Notes: 1. Only 0 can be written, to clear the flags.

2. The initial value is undefined since it depends on the pin states.

Appendix C I/O Port Block Diagrams

C.1 Port 1 Block Diagram

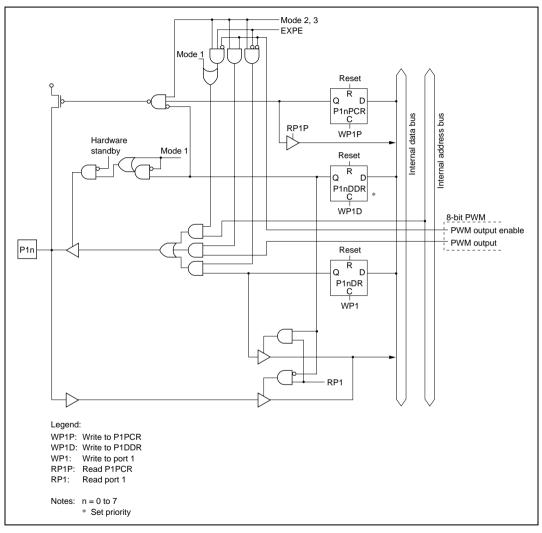


Figure C.1 Port 1 Block Diagram

C.2 Port 2 Block Diagrams

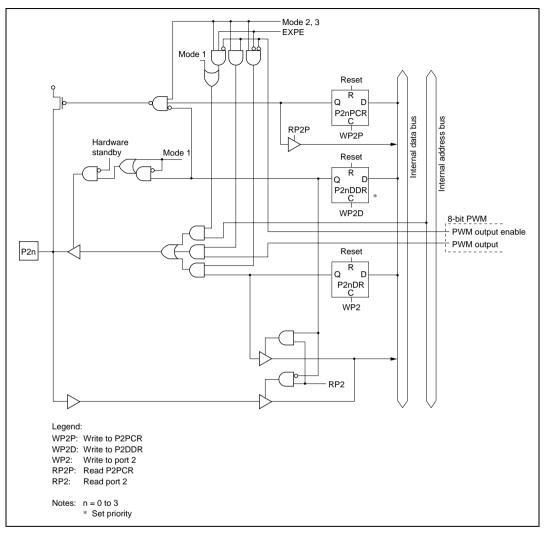


Figure C.2 Port 2 Block Diagram (Pins P20 to P23)

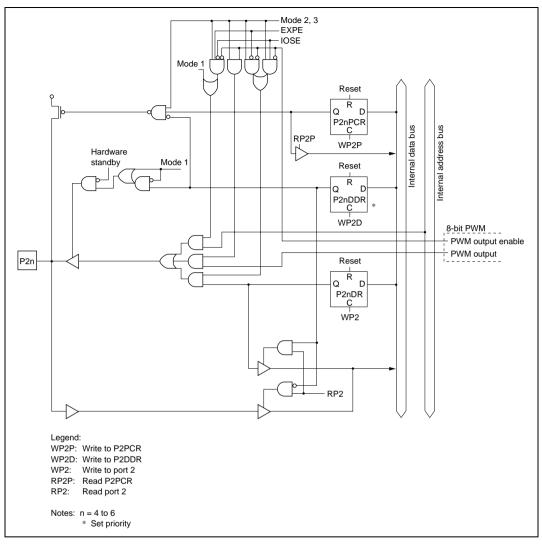


Figure C.3 Port 2 Block Diagram (Pins P24 to P26)

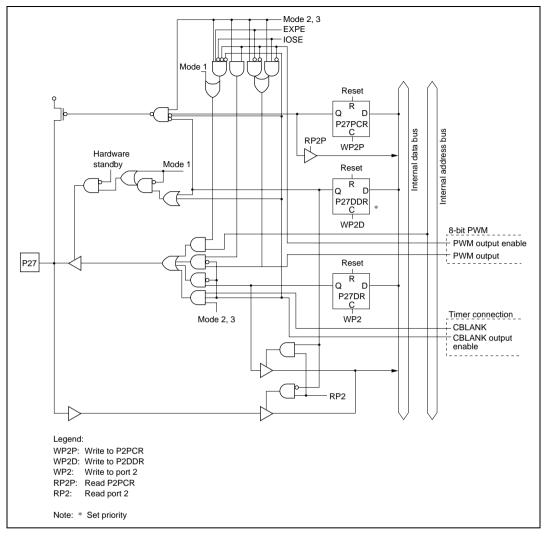


Figure C.4 Port 2 Block Diagram (Pin P27)

C.3 Port 3 Block Diagram

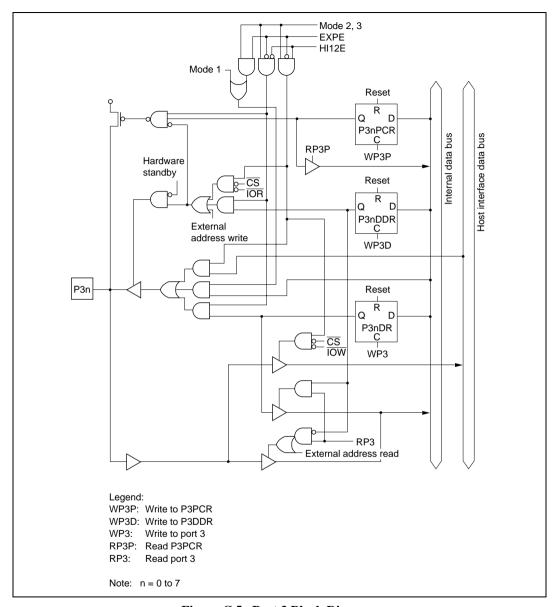


Figure C.5 Port 3 Block Diagram

C.4 Port 4 Block Diagrams

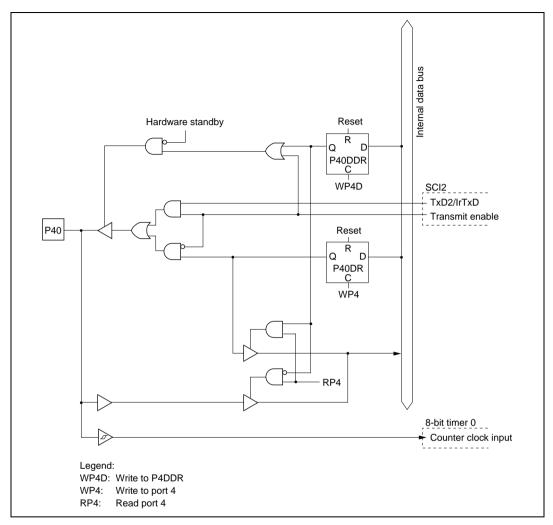


Figure C.6 Port 4 Block Diagram (Pin P40)

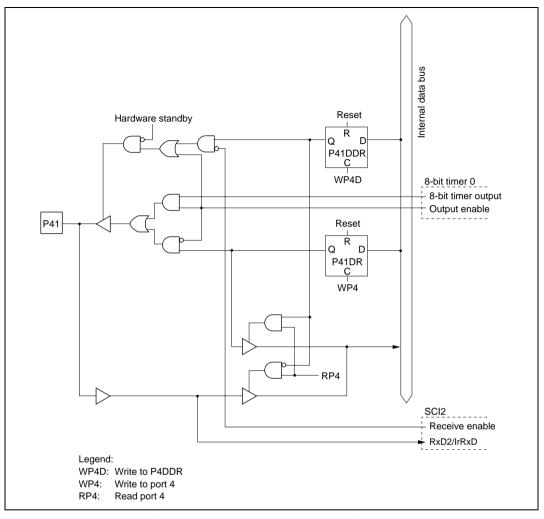


Figure C.7 Port 4 Block Diagram (Pin P41)

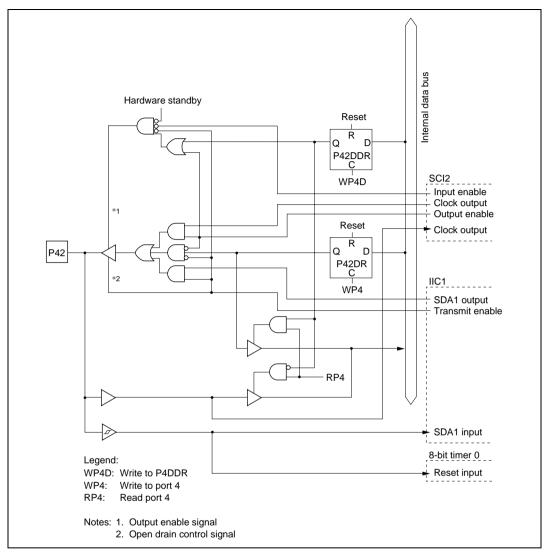


Figure C.8 Port 4 Block Diagram (Pin P42)

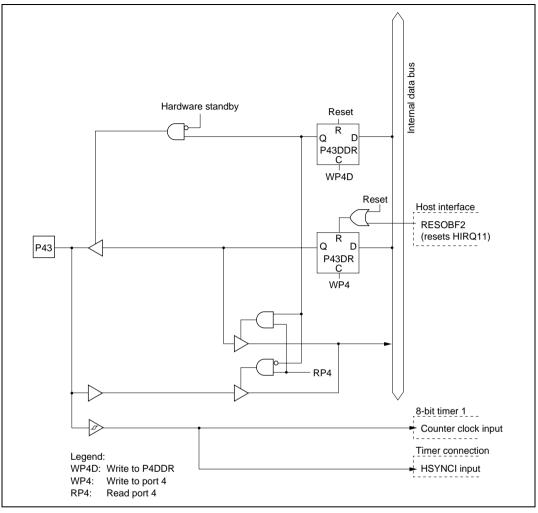


Figure C.9 Port 4 Block Diagram (Pin P43)

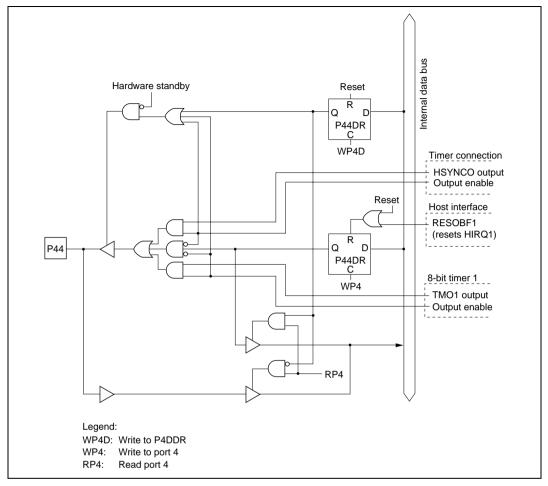


Figure C.10 Port 4 Block Diagram (Pin P44)

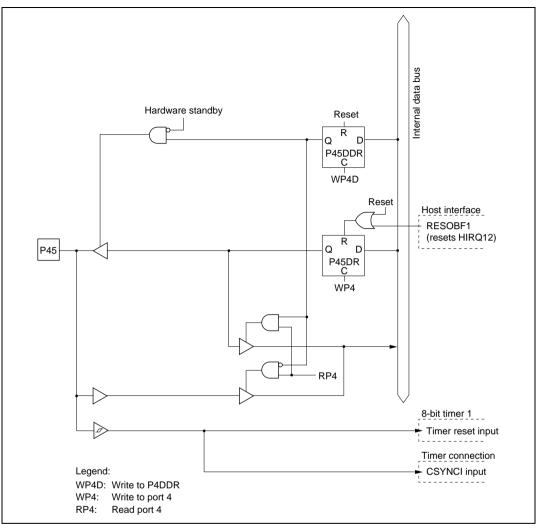


Figure C.11 Port 4 Block Diagram (Pin P45)

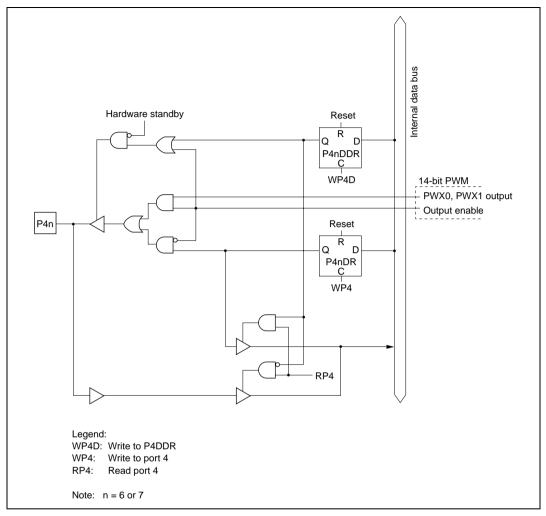


Figure C.12 Port 4 Block Diagram (Pins P46, P47)

C.5 Port 5 Block Diagrams

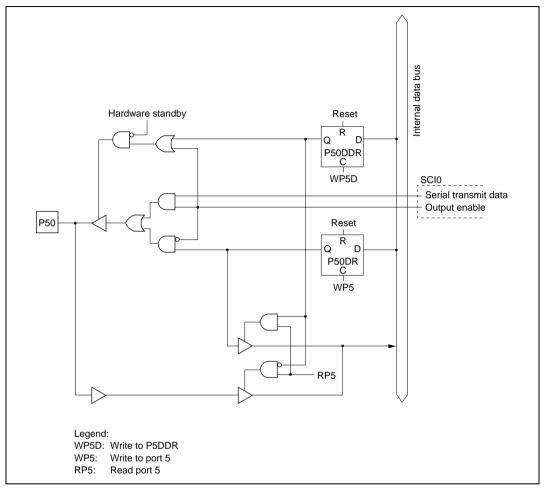


Figure C.13 Port 5 Block Diagram (Pin P50)

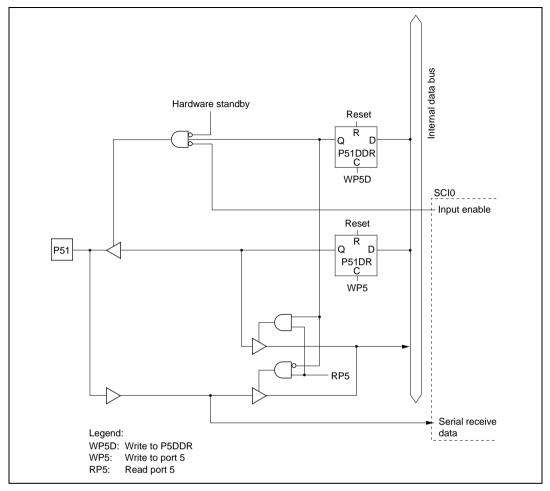


Figure C.14 Port 5 Block Diagram (Pin P51)

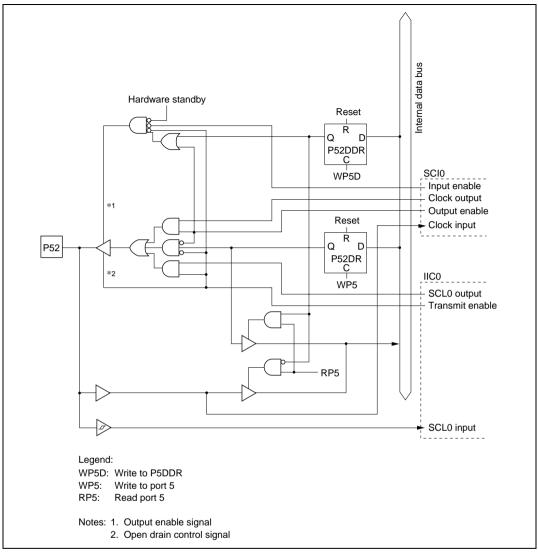


Figure C.15 Port 5 Block Diagram (Pin P52)

C.6 Port 6 Block Diagrams

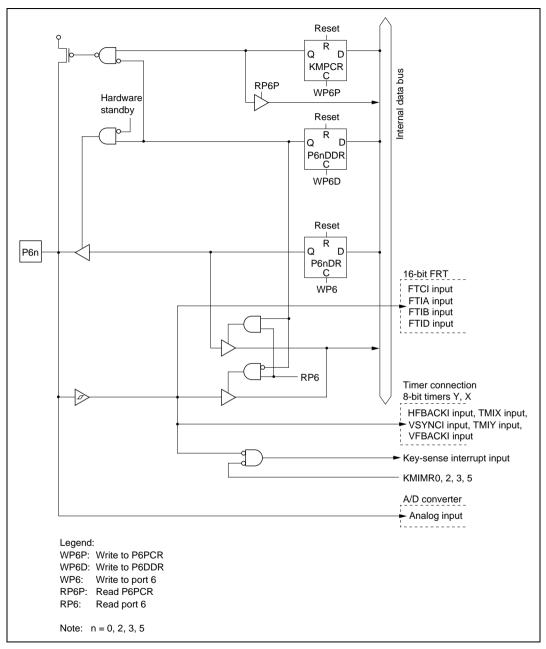


Figure C.16 Port 6 Block Diagram (Pins P60, P62, P63, P65)

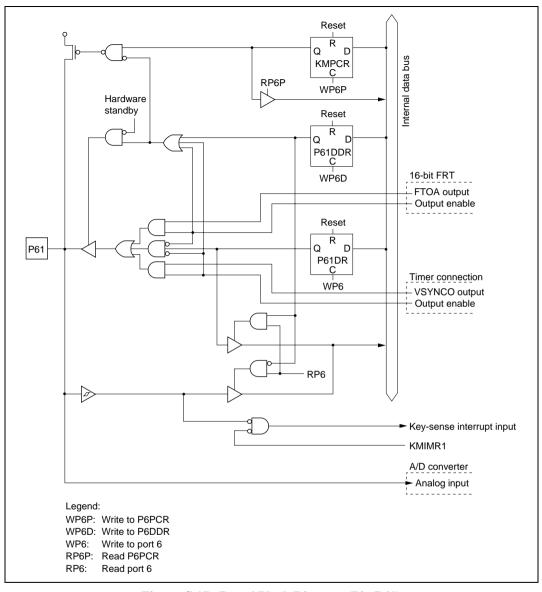


Figure C.17 Port 6 Block Diagram (Pin P61)

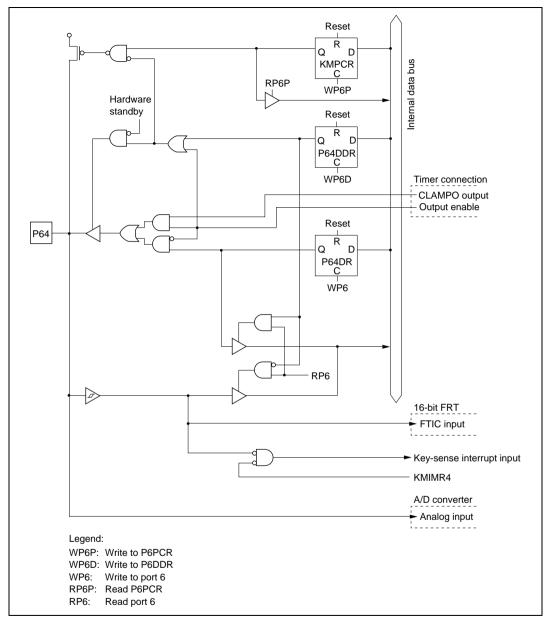


Figure C.18 Port 6 Block Diagram (Pin P64)

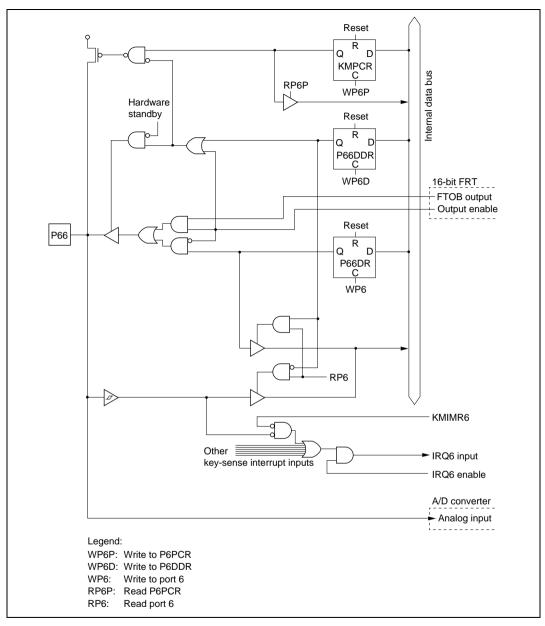


Figure C.19 Port 6 Block Diagram (Pin P66)

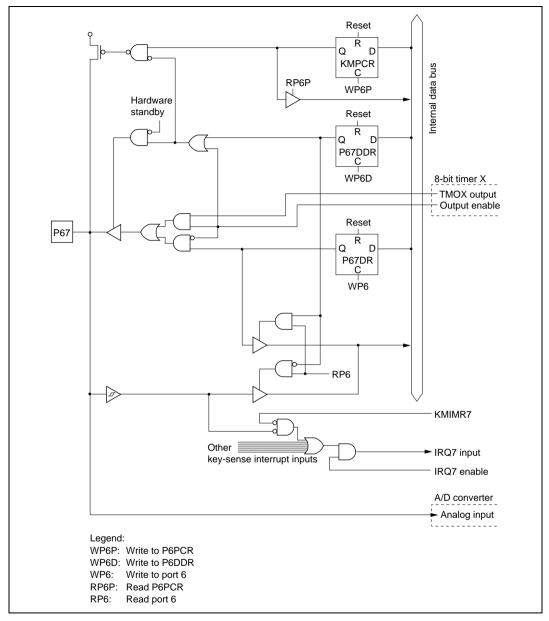


Figure C.20 Port 6 Block Diagram (Pin P67)

C.7 Port 7 Block Diagrams

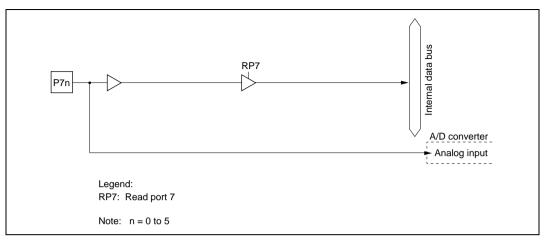


Figure C.21 Port 7 Block Diagram (Pins P70 to P75)

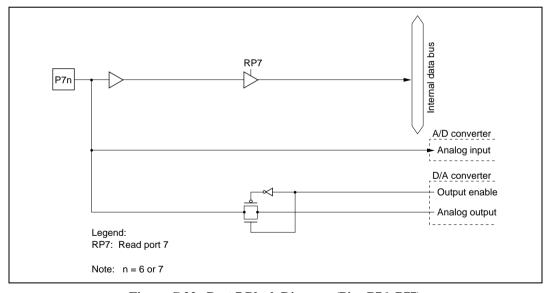


Figure C.22 Port 7 Block Diagram (Pins P76, P77)

C.8 Port 8 Block Diagrams

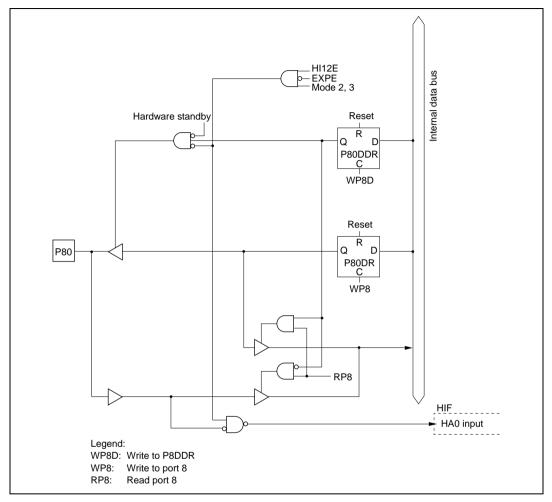


Figure C.23 Port 8 Block Diagram (Pin P80)

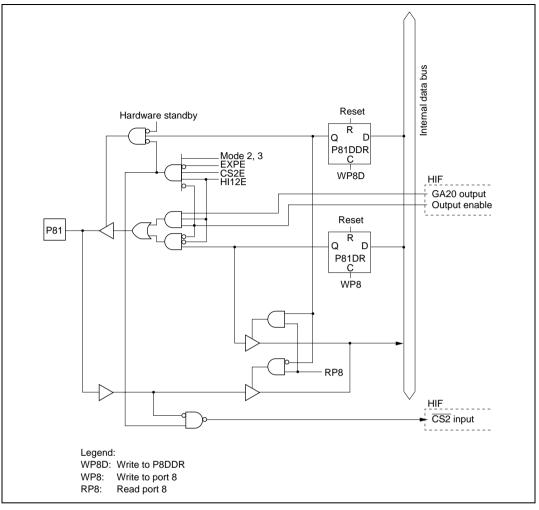


Figure C.24 Port 8 Block Diagram (Pin P81)

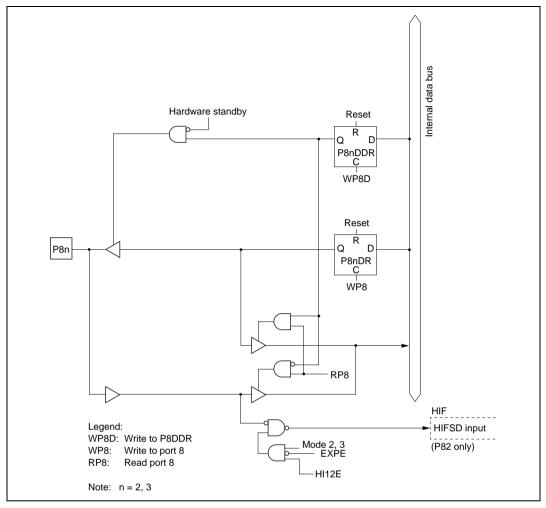


Figure C.25 Port 8 Block Diagram (Pins P82, P83)

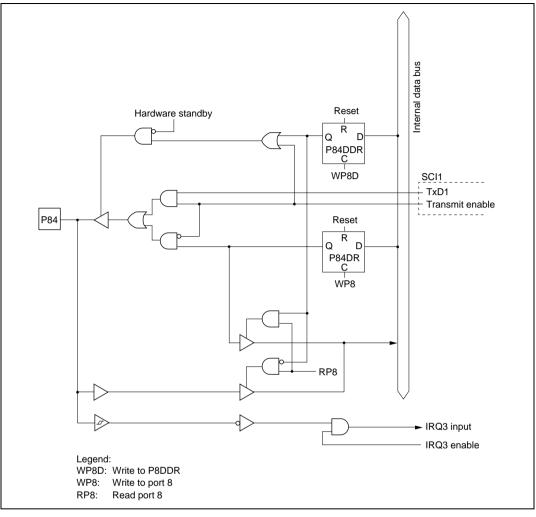


Figure C.26 Port 8 Block Diagram (Pin P84)

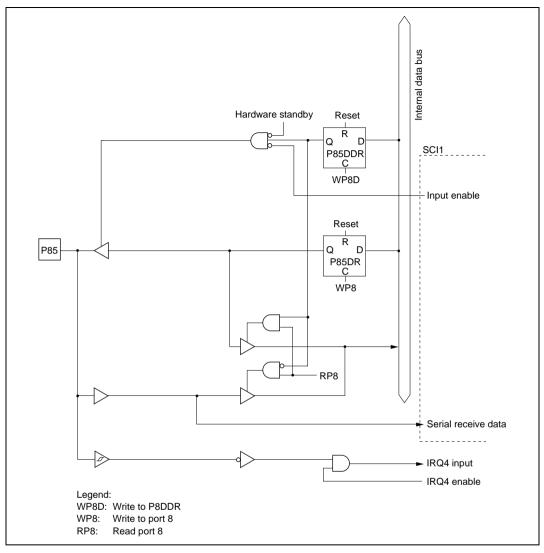


Figure C.27 Port 8 Block Diagram (Pin P85)

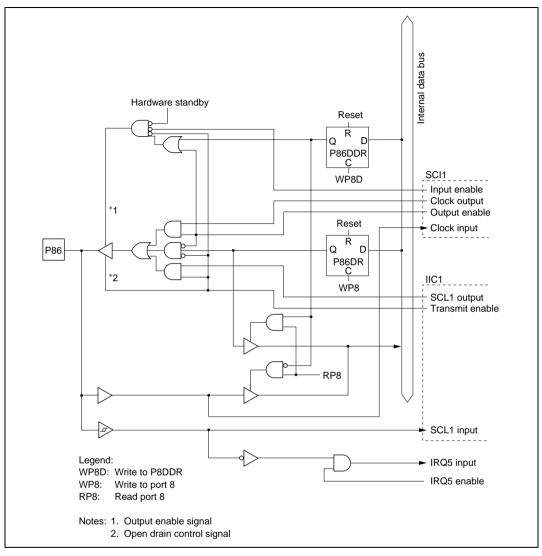


Figure C.28 Port 8 Block Diagram (Pin P86)

C.9 Port 9 Block Diagrams

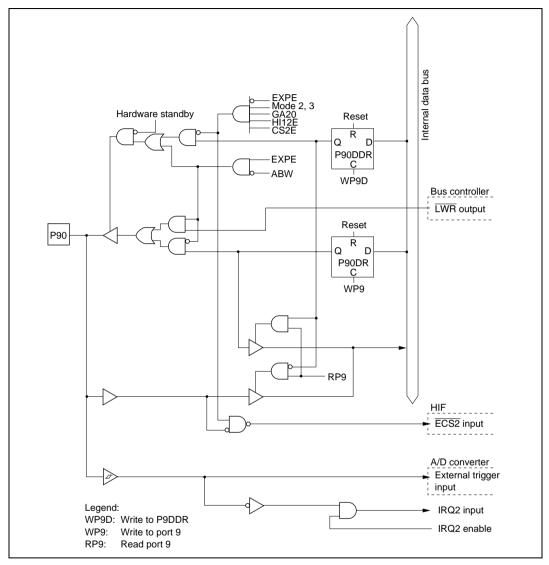


Figure C.29 Port 9 Block Diagram (Pin P90)

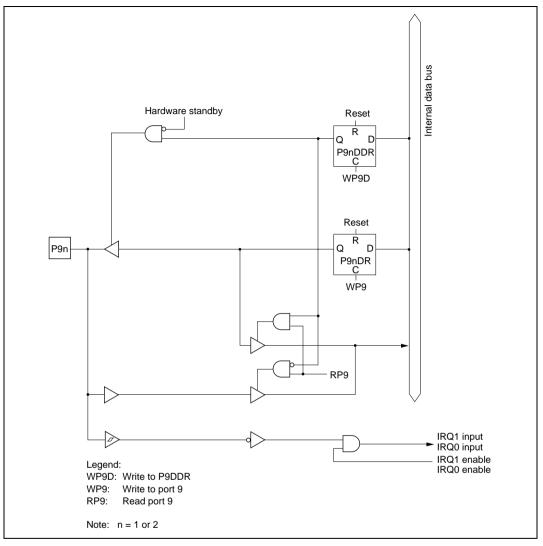


Figure C.30 Port 9 Block Diagram (Pins P91, P92)

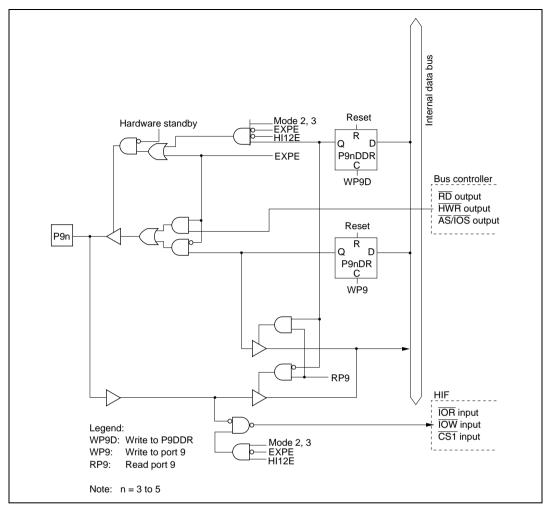


Figure C.31 Port 9 Block Diagram (Pins P93 to P95)

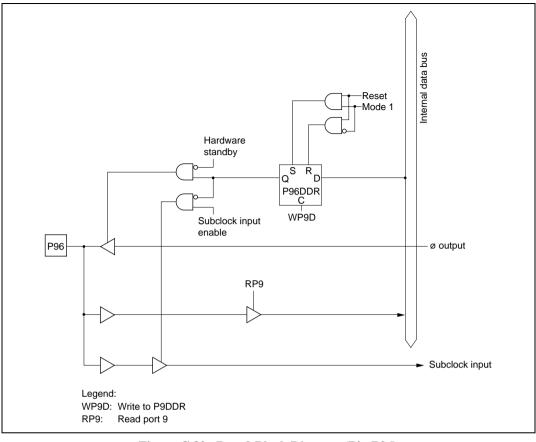


Figure C.32 Port 9 Block Diagram (Pin P96)

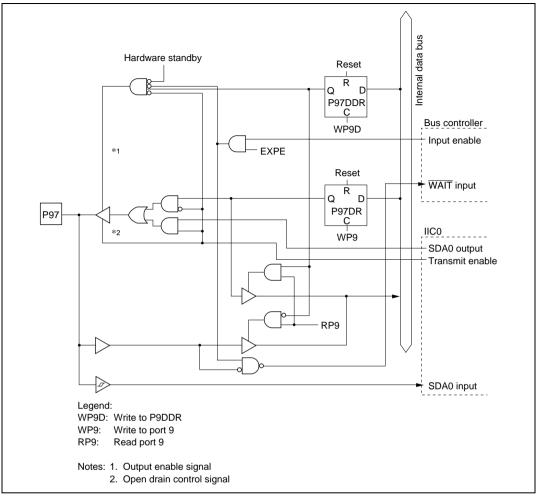


Figure C.33 Port 9 Block Diagram (Pin P97)

C.10 Port A Block Diagrams

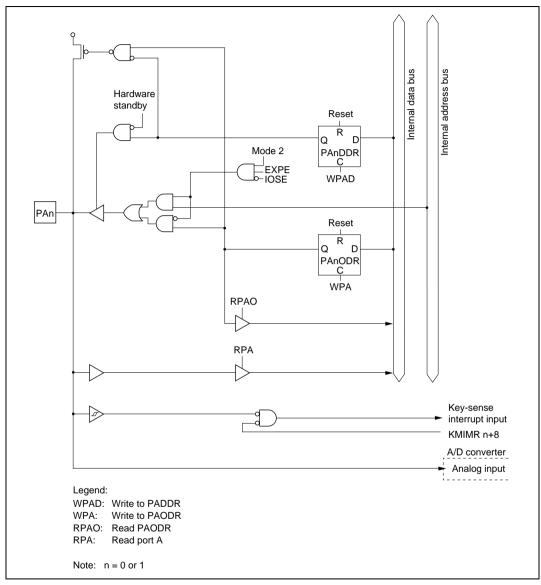


Figure C.34 Port A Block Diagram (Pins PA0, PA1)

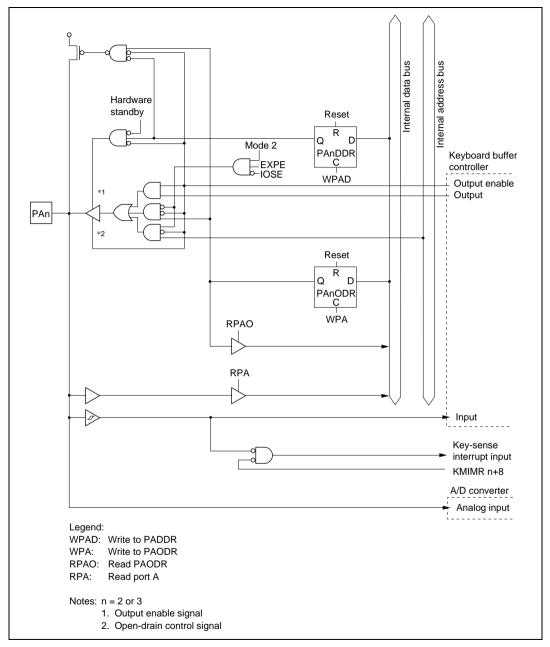


Figure C.35 Port A Block Diagram (Pins PA2, PA3)

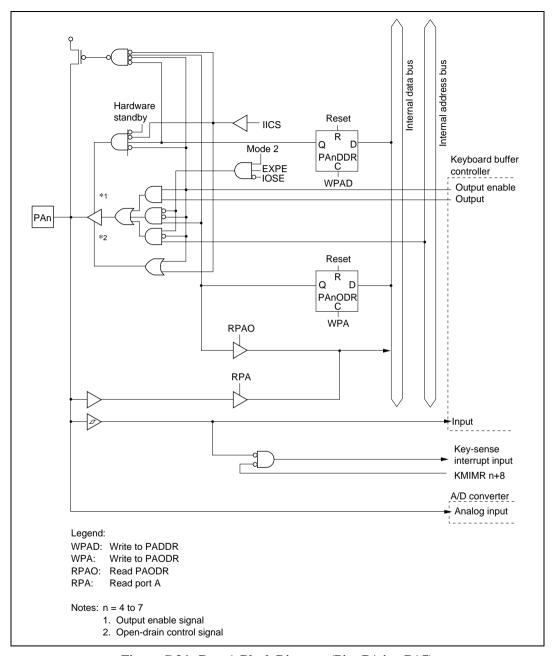


Figure C.36 Port A Block Diagram (Pins PA4 to PA7)

C.11 Port B Block Diagram

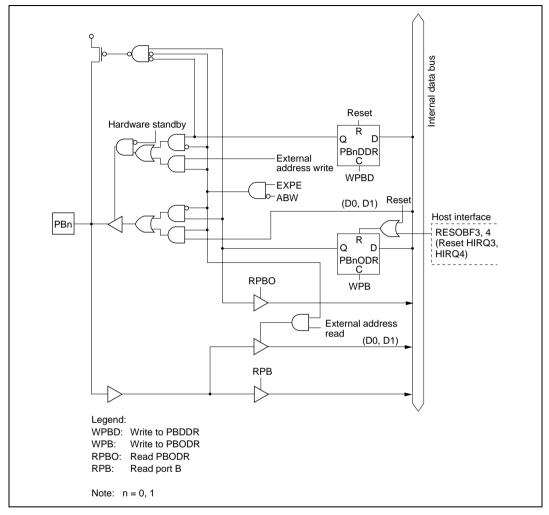


Figure C.37 Port B Block Diagram (Pins PB0 and PB1)

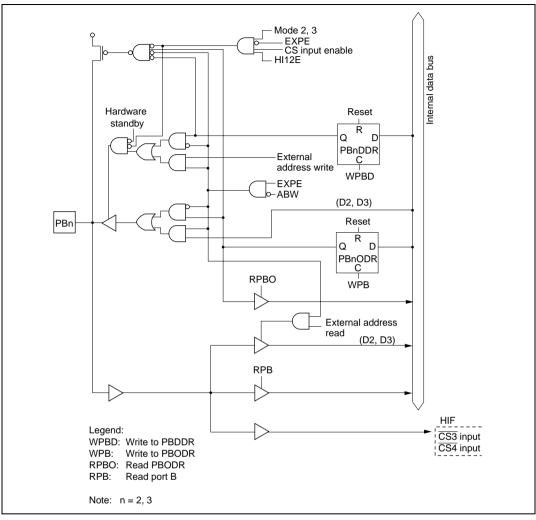


Figure C.38 Port B Block Diagram (Pins PB2 and PB3)

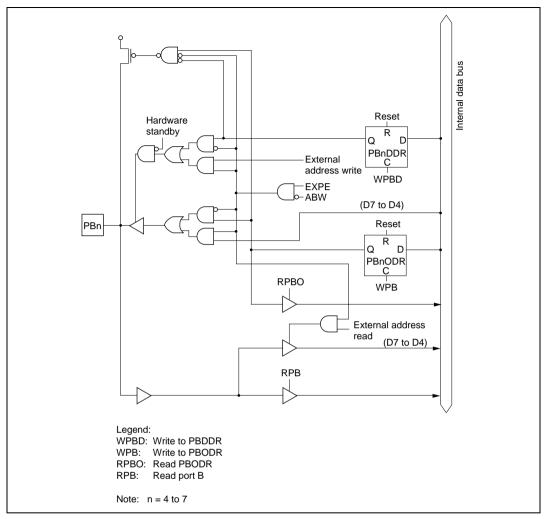


Figure C.39 Port B Block Diagram (Pins PB4 to PB7)

Appendix D Pin States

D.1 Port States in Each Processing State

Table D.1 I/O Port States in Each Processing State

Port Name Pin Name	MCU Operating Mode	Reset	Hardware Standby Mode	Software Standby Mode	Watch Mode	Sleep Mode	Sub- sleep Mode	Subactive Mode	Program Execution State
Port 1	1	L	T	keep*	keep*	keep*	keep*	A7 to A0	A7 to A0
A7 to A0	2, 3 (EXPE = 1)	Т	_					Address output/ input port	Address output/ input port
	2, 3 (EXPE = 0)	_						I/O port	I/O port
Port 2	1	L	Т	keep*	keep*	keep*	keep*	A15 to A8	A15 to A8
A15 to A8	2, 3 (EXPE = 1)	Т	_					Address output/ input port	Address output/ input port
	2, 3 (EXPE = 0)	_						I/O port	I/O port
Port 3	1	Т	T	Т	Т	Т	T	D15 to D8	D15 to D8
D15 to D8	2, 3 (EXPE = 1)	_							
	2, 3 (EXPE = 0)	_		keep	keep	keep	keep	I/O port	I/O port
Port 4	1	Т	Т	keep	keep	keep	keep	I/O port	I/O port
	2, 3 (EXPE = 1)	_							
	2, 3 (EXPE = 0)	_							
Port 5	1	Т	Т	keep	keep	keep	keep	I/O port	I/O port
	2, 3 (EXPE = 1)	_							
	2, 3 (EXPE = 0)	_							
Port 6	1	Т	Т	keep	keep	keep	keep	I/O port	I/O port
	2, 3 (EXPE = 1)	_							
	2, 3 (EXPE = 0)	_							
Port 7	1	Т	Т	Т	Т	Т	Т	Input port	Input port
	2, 3 (EXPE = 1)	_							
	2, 3 (EXPE = 0)								
Port 8	1	Т	Т	keep	keep	keep	keep	I/O port	I/O port
	2, 3 (EXPE = 1)	_							
	2, 3 (EXPE = 0)	_							
Port 97 WAIT	1	Т	Т	T/keep	T/keep	T/keep	T/keep	WAIT/	WAIT/
	2, 3 (EXPE = 1)	_						I/O port	I/O port
	2, 3 (EXPE = 0)	_		keep	keep	keep	keep	I/O port	I/O port

Port Name Pin Name	MCU Operating Mode	Reset	Hardware Standby Mode	Software Standby Mode	Watch Mode	Sleep Mode	Sub- sleep Mode	Subactive Mode	Program Execution State
Port 96	1	Clock output		[DDR = 1] H	EXCL input	[DDR = 1] clock output	EXCL input	EXCL input	Clock output/ EXCL input/ input port
	2, 3 (EXPE = 1)	T -		[DDR = 0] T		[DDR = 0]			
	2, 3 (EXPE = 0)			•		T			
Port 95 to 93	1	Н	T	Н	Н	Н	Н	AS, HWR,	AS, HWR,
AS, HWR,	2, 3 (EXPE = 1)	Т	=					RD	RD
	2, 3 (EXPE = 0)	_		keep	keep	keep	keep	I/O port	I/O port
Port 92 to 91	1	Т	Т	keep	keep	keep	keep	I/O port	I/O port
	2, 3 (EXPE = 1)	_							
	2, 3 (EXPE = 0)	_							
Port 90	1	Т	Т	H/keep	H/keep	H/keep	H/keep	LWR/	LWR/
LWR	2, 3 (EXPE = 1)	_						I/O port	I/O port
	2, 3 (EXPE = 0)	_		keep	keep	keep	keep	I/O port	I/O port
Port A	1	Т	Т	keep*	keep*	keep*	keep*	I/O port	I/O port
A23 to A16	2, 3 (EXPE = 1)	_						A23 to A16/ I/O port	A23 to A16/ I/O port
	2, 3 (EXPE = 0)	_						I/O port	I/O port
Port B	1	Т	Т	T/keep	T/keep	T/keep	T/keep	D7 to D0/	D7 to D0/
D7 to D0	2, 3 (EXPE = 1)	_						I/O port	I/O port
	2, 3 (EXPE = 0)	_		keep	keep	keep	keep	I/O port	I/O port

Legend:

H: High L: Low

T: High-impedance state

keep: Input ports are in the high-impedance state (when DDR = 0 and PCR = 1, MOS input pullups remain on).

Output ports maintain their previous state.

Depending on the pins, the on-chip supporting modules may be initialized and the I/O port function determined by DDR and DR used.

DDR: Data direction register

Note: * In the case of address output, the last address accessed is retained.

Appendix E Timing of Transition to and Recovery from Hardware Standby Mode

E.1 Timing of Transition to Hardware Standby Mode

(1) To retain RAM contents with the RAME bit set to 1 in SYSCR, drive the RES signal low 10 system clock cycles before the STBY signal goes low, as shown in figure E.1. RES must remain low until STBY signal goes low (minimum delay from STBY low to RES high: 0 ns).

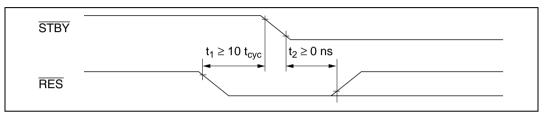


Figure E.1 Timing of Transition to Hardware Standby Mode

(2) To retain RAM contents with the RAME bit cleared to 0 in SYSCR, or when RAM contents do not need to be retained, RES does not have to be driven low as in (1).

E.2 Timing of Recovery from Hardware Standby Mode

Drive the RES signal low at least 100 ns before STBY goes high to execute a reset.

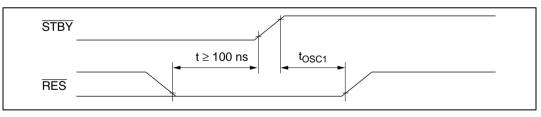


Figure E.2 Timing of Recovery from Hardware Standby Mode

Appendix F Product Code Lineup

Table F.1 H8S/2148 Group and H8S/2144 Group Product Code Lineup

Product Ty	pe			Product Code	Mark Code	Package (Package Code)	Notes
H8S/2148 Group	H8S/2148	Mask-ROM version	Standard product (5-V version, 4-V version,	HD6432148S	HD6432148S(V)(***)FA	100-pin QFP (FP-100B)	
			3-V version)		HD6432148S(V)(***)TE	100-pin TQFP (TFP-100B)	=
			Version with on-chip I ² C bus interface	HD6432148SW	HD6432148S(V)W(***)FA	100-pin QFP (FP-100B)	_
			(5-V version, 4-V version, 3-V version)		HD6432148S(V)W(***)TE	100-pin TQFP (TFP-100B)	
		F-ZTAT version	Standard product (5-V version/4-V version)	HD64F2148	HD64F2148FA20	100-pin QFP (FP-100B)	
					HD64F2148TE20	100-pin TQFP (TFP-100B)	_
			Low-voltage version (3-V version)	HD64F2148V	HD64F2148VFA10	100-pin QFP (FP-100B)	_
					HD64F2148VTE10	100-pin TQFP (TFP-100B)	
	H8S/2147	Mask-ROM version	Standard product (5-V version, 4-V version, 3-V version)	HD6432147S	HD6432147S(V)(***)FA	100-pin QFP (FP-100B)	_
					HD6432147S(V)(***)TE	100-pin TQFP (TFP-100B)	
			Version with on-chip I ² C bus interface (5-V version, 4-V version, 3-V version)	HD6432147SW	HD6432147S(V)W(***)FA	100-pin QFP (FP-100B)	_
					HD6432147S(V)W(***)TE	100-pin TQFP (TFP-100B)	_
H8S/2148 Group A-mask version	H8S/2148A	S/2148A F-ZTAT version A-mask version	Standard product (5-V version/4-V version)	HD64F2148A	HD64F2148AFA20	100-pin QFP (FP-100B)	_
					HD64F2148ATE20	100-pin TQFP (TFP-100B)	_
			Low-voltage version (3-V version)	HD64F2148AV	HD64F2148AVFA10	100-pin QFP (FP-100B)	
					HD64F2148AVTE10	100-pin TQFP (TFP-100B)	_
	H8S/2147A	2147A F-ZTAT version A-mask version	Standard product (5-V version/4-V version)	HD64F2147A	HD64F2147AFA20	100-pin QFP (FP-100B)	=
					HD64F2147ATE20	100-pin TQFP (TFP-100B)	=
			Low-voltage version (3-V version)	HD64F2147AV	HD64F2147AVFA10	100-pin QFP (FP-100B)	
					HD64F2147AVTE10	100-pin TQFP (TFP-100B)	_

Product Ty	oe			Product Code	Mark Code	Package (Package Code)	Notes
H8S/2147N	H8S/2147N	F-ZTAT version	Standard product (5-V version)	HD64F2147N	HD64F2147NFA20	100-pin QFP (FP-100B)	
					HD64F2147NTE20	100-pin TQFP (TFP-100B)	_
			Low-voltage version (3-V version)	HD64F2147NV	HD64F2147NVFA10	100-pin QFP (FP-100B)	_
					HD64F2147NVTE10	100-pin TQFP (TFP-100B)	_
H8S/2144 Group	H8\$/2144	Mask-ROM version	Standard product (5-V version, 4-V version, 3-V version)	HD6432144S	HD6432144S(V)(***)FA	100-pin QFP (FP-100B)	
					HD6432144S(V)(***)TE	100-pin TQFP (TFP-100B)	_
		F-ZTAT version	Standard product (5-V version/4-V version)	HD64F2144	HD64F2144FA20	100-pin QFP (FP-100B)	
					HD64F2144TE20	100-pin TQFP (TFP-100B)	_
			Low-voltage version (3-V version)	HD64F2144V	HD64F2144VFA10	100-pin QFP (FP-100B)	
					HD64F2144VTE10	100-pin TQFP (TFP-100B)	_
	H8S/2143	Mask-ROM version	Standard product (5-V version, 4-V version, 3-V version)	HD6432143S	HD6432143S(V)(***)FA	100-pin QFP (FP-100B)	
					HD6432143S(V)(***)TE	100-pin TQFP (TFP-100B)	_
	H8S/2142	Mask-ROM version	Standard product (5-V version, 4-V version, 3-V version)	HD6432142	HD6432142(***)FA	100-pin QFP (FP-100B)	
					HD6432142(***)TE	100-pin TQFP (TFP-100B)	
		F-ZTAT version	Standard product (5-V version/4-V version)	HD64F2142R	HD64F2142RFA20	100-pin QFP (FP-100B)	
					HD64F2142RTE20	100-pin TQFP (TFP-100B)	
			Low-voltage version (3-V version)	HD64F2142RV	HD64F2142RVFA10	100-pin QFP (FP-100B)	
					HD64F2142RVTE10	100-pin TQFP (TFP-100B)	_
H8S/2144 Group A-mask version	H8S/2144A	4A F-ZTAT version A-mask version	Standard product (5-V version/4-V version)	HD64F2144A	HD64F2144AFA20	100-pin QFP (FP-100B)	
					HD64F2144ATE20	100-pin TQFP (TFP-100B)	
			Low-voltage version (3-V version)	HD64F2144AV	HD64F2144AVFA10	100-pin QFP (FP-100B)	
					HD64F2144AVTE10	100-pin TQFP (TFP-100B)	

Note: (***) is the ROM code.

The F-ZTAT version of the H8S/2148 has an on-chip I²C bus interface as standard.

The F-ZTAT 5 V/4 V version supports the operating ranges of the 5 V version and the 4 V version.

The operating range of the F-ZTAT low-voltage version will be decided later.

The above table includes products under development. Information on the status of individual products can be obtained from Renesas sales offices.



Appendix G Package Dimensions

The package dimension that is shown in the Renesas Semiconductor Package Data Book has priority.

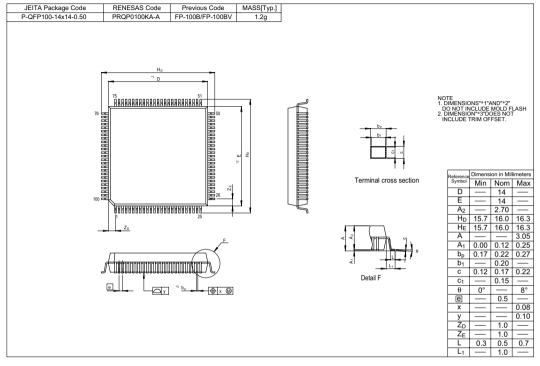


Figure G.1 Package Dimensions (FP-100B)

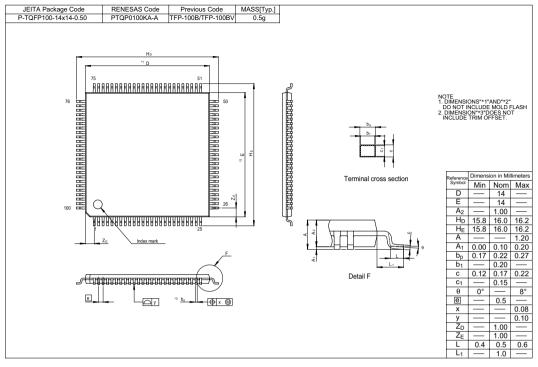


Figure G.2 Package Dimensions (TFP-100B)

Renesas 16-Bit Single-Chip Microcomputer Hardware Manual H8S/2148 Group, H8S/2144 Group, H8S/2148FZTAT™, H8S/2147N F-ZTAT™, H8S/2144F-ZTAT™, H8S/2142F-ZTAT™

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H8S/2148 Group, H8S/2144 Group, H8S/2148F-ZTAT™, H8S/2147N F-ZTAT™, H8S/2144F-ZTAT™, H8S/2142F-ZTAT™ Hardware Manual



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